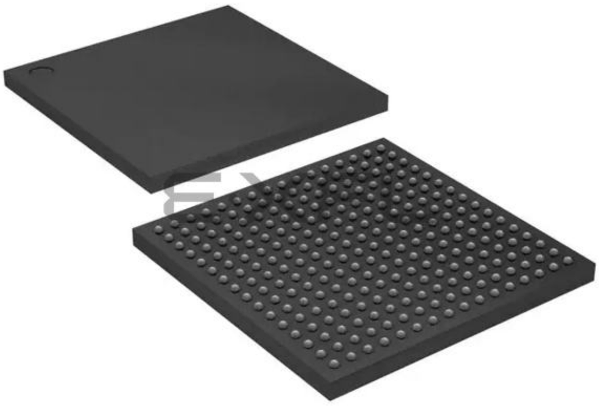


Welcome to E-XFL.COM

Understanding Embedded - CPLDs (Complex Programmable Logic Devices)

Embedded - CPLDs, or Complex Programmable Logic Devices, are highly versatile digital logic devices used in electronic systems. These programmable components are designed to perform complex logical operations and can be customized for specific applications. Unlike fixed-function ICs, CPLDs offer the flexibility to reprogram their configuration, making them an ideal choice for various embedded systems. They consist of a set of logic gates and programmable interconnects, allowing designers to implement complex logic circuits without needing custom hardware.

Applications of Embedded - CPLDs

Details	
Product Status	Active
Programmable Type	In System Programmable
Delay Time tpd(1) Max	5.4 ns
Voltage Supply - Internal	2.5V, 3.3V
Number of Logic Elements/Blocks	570
Number of Macrocells	440
Number of Gates	-
Number of I/O	160
Operating Temperature	0°C ~ 85°C (TJ)
Mounting Type	Surface Mount
Package / Case	256-BGA
Supplier Device Package	256-FBGA (17x17)
Purchase URL	https://www.e-xfl.com/product-detail/intel/epm570f256c4

1. Introduction

MII51001-1.9

Introduction

The MAX® II family of instant-on, non-volatile CPLDs is based on a 0.18-µm, 6-layer-metal-flash process, with densities from 240 to 2,210 logic elements (LEs) (128 to 2,210 equivalent macrocells) and non-volatile storage of 8 Kbits. MAX II devices offer high I/O counts, fast performance, and reliable fitting versus other CPLD architectures. Featuring MultiVolt core, a user flash memory (UFM) block, and enhanced in-system programmability (ISP), MAX II devices are designed to reduce cost and power while providing programmable solutions for applications such as bus bridging, I/O expansion, power-on reset (POR) and sequencing control, and device configuration control.

Features

The MAX II CPLD has the following features:

- Low-cost, low-power CPLD
- Instant-on, non-volatile architecture
- Standby current as low as 25 µA
- Provides fast propagation delay and clock-to-output times
- Provides four global clocks with two clocks available per logic array block (LAB)
- UFM block up to 8 Kbits for non-volatile storage
- MultiVolt core enabling external supply voltages to the device of either 3.3 V/2.5 V or 1.8 V
- MultiVolt I/O interface supporting 3.3-V, 2.5-V, 1.8-V, and 1.5-V logic levels
- Bus-friendly architecture including programmable slew rate, drive strength, bus-hold, and programmable pull-up resistors
- Schmitt triggers enabling noise tolerant inputs (programmable per pin)
- I/Os are fully compliant with the Peripheral Component Interconnect Special Interest Group (PCI SIG) PCI Local Bus Specification, Revision 2.2 for 3.3-V operation at 66 MHz
- Supports hot-socketing
- Built-in Joint Test Action Group (JTAG) boundary-scan test (BST) circuitry compliant with IEEE Std. 1149.1-1990
- ISP circuitry compliant with IEEE Std. 1532

Table 2–1. MAX II Device Resources

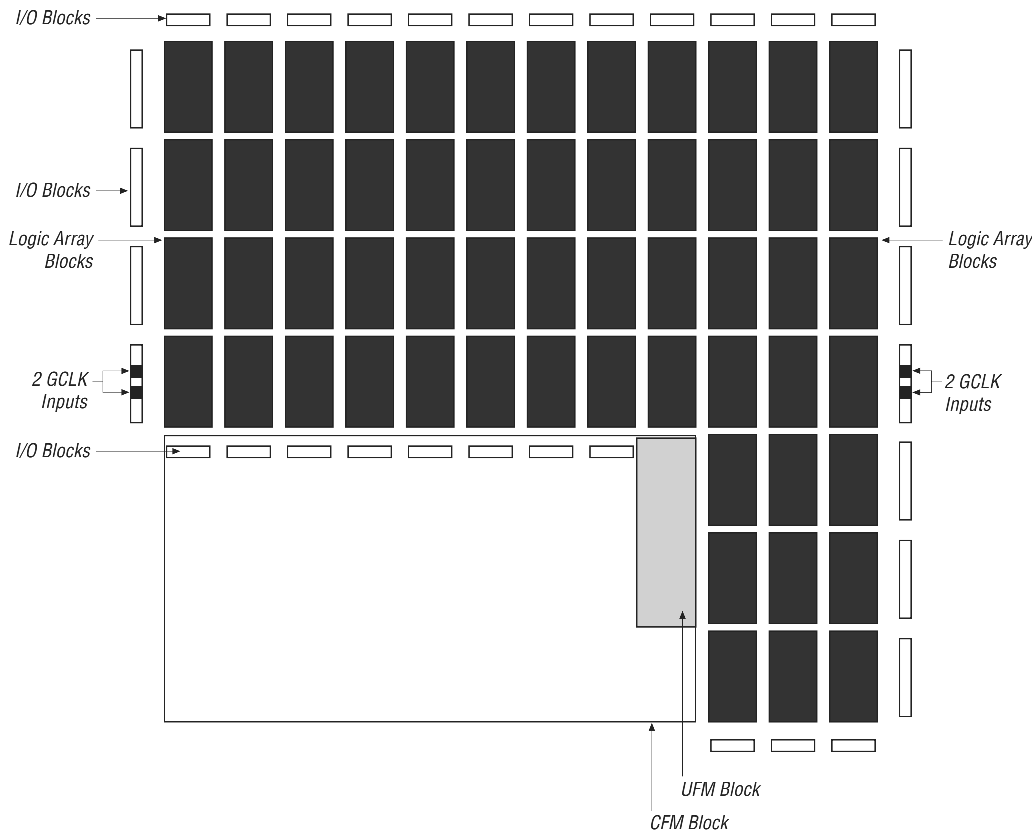
Devices	UFM Blocks	LAB Columns	LAB Rows		Total LABs
			Long LAB Rows	Short LAB Rows (Width) *(1)*	
EPM240	1	6	4	—	24
EPM570	1	12	4	3 (3)	57
EPM1270	1	16	7	3 (5)	127
EPM2210	1	20	10	3 (7)	221

Note to Table 2–1:

(1) The width is the number of LAB columns in length.

Figure 2–2 shows a floorplan of a MAX II device.

Figure 2–2. MAX II Device Floorplan *(Note 1)*



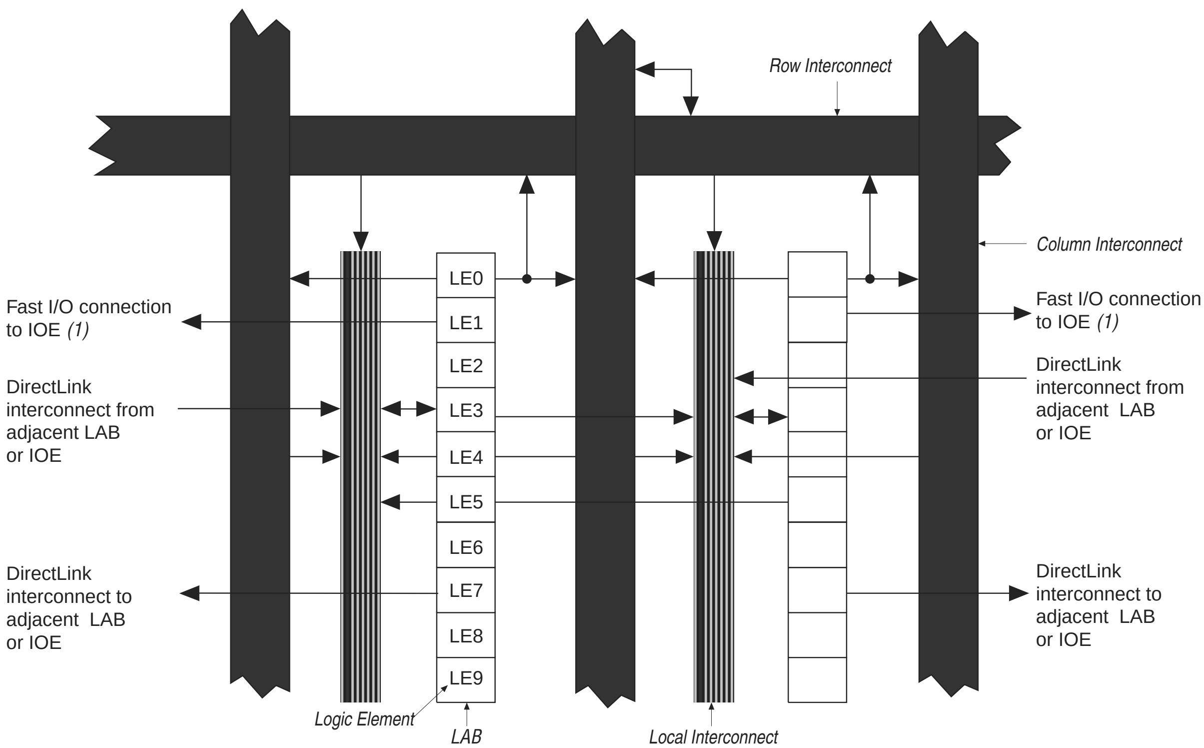
Note to Figure 2–2:

(1) The device shown is an EPM570 device. EPM1270 and EPM2210 devices have a similar floorplan with more LABs. For EPM240 devices, the CFM and UFM blocks are located on the left side of the device.

Logic Array Blocks

Each LAB consists of 10 LEs, LE carry chains, LAB control signals, a local interconnect, a look-up table (LUT) chain, and register chain connection lines. There are 26 possible unique inputs into an LAB, with an additional 10 local feedback input lines fed by LE outputs in the same LAB. The local interconnect transfers signals between LEs in the same LAB. LUT chain connections transfer the output of one LE's LUT to the adjacent LE for fast sequential LUT connections within the same LAB. Register chain connections transfer the output of one LE's register to the adjacent LE's register within an LAB. The Quartus® II software places associated logic within an LAB or adjacent LABs, allowing the use of local, LUT chain, and register chain connections for performance and area efficiency. Figure 2–3 shows the MAX II LAB.

Figure 2–3. MAX II LAB Structure



Note to Figure 2–3:

(1) Only from LABs adjacent to IOEs.

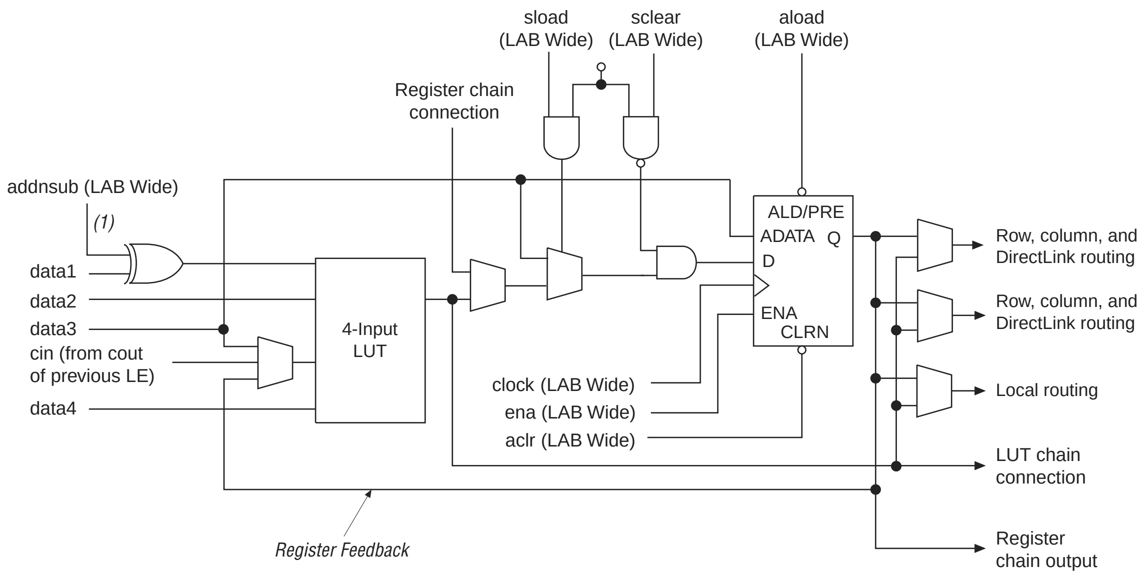
LAB Interconnects

The LAB local interconnect can drive LEs within the same LAB. The LAB local interconnect is driven by column and row interconnects and LE outputs within the same LAB. Neighboring LABs, from the left and right, can also drive an LAB's local interconnect through the DirectLink connection. The DirectLink connection feature minimizes the use of row and column interconnects, providing higher performance and flexibility. Each LE can drive 30 other LEs through fast local and DirectLink interconnects. Figure 2–4 shows the DirectLink connection.

Normal Mode

The normal mode is suitable for general logic applications and combinational functions. In normal mode, four data inputs from the LAB local interconnect are inputs to a four-input LUT (see Figure 2–7). The Quartus II Compiler automatically selects the carry-in or the data3 signal as one of the inputs to the LUT. Each LE can use LUT chain connections to drive its combinational output directly to the next LE in the LAB. Asynchronous load data for the register comes from the data3 input of the LE. LEs in normal mode support packed registers.

Figure 2–7. LE in Normal Mode



Note to Figure 2–7:

(1) This signal is only allowed in normal mode if the LE is at the end of an adder/subtractor chain.

Dynamic Arithmetic Mode

The dynamic arithmetic mode is ideal for implementing adders, counters, accumulators, wide parity functions, and comparators. An LE in dynamic arithmetic mode uses four 2-input LUTs configurable as a dynamic adder/subtractor. The first two 2-input LUTs compute two summations based on a possible carry-in of 1 or 0; the other two LUTs generate carry outputs for the two chains of the carry-select circuitry. As shown in Figure 2–8, the LAB carry-in signal selects either the `carry-in0` or carry-in1 chain. The selected chain's logic level in turn determines which parallel sum is generated as a combinational or registered output. For example, when implementing an adder, the sum output is the selection of two possible calculated sums:

`data1 + data2 + carry in0`

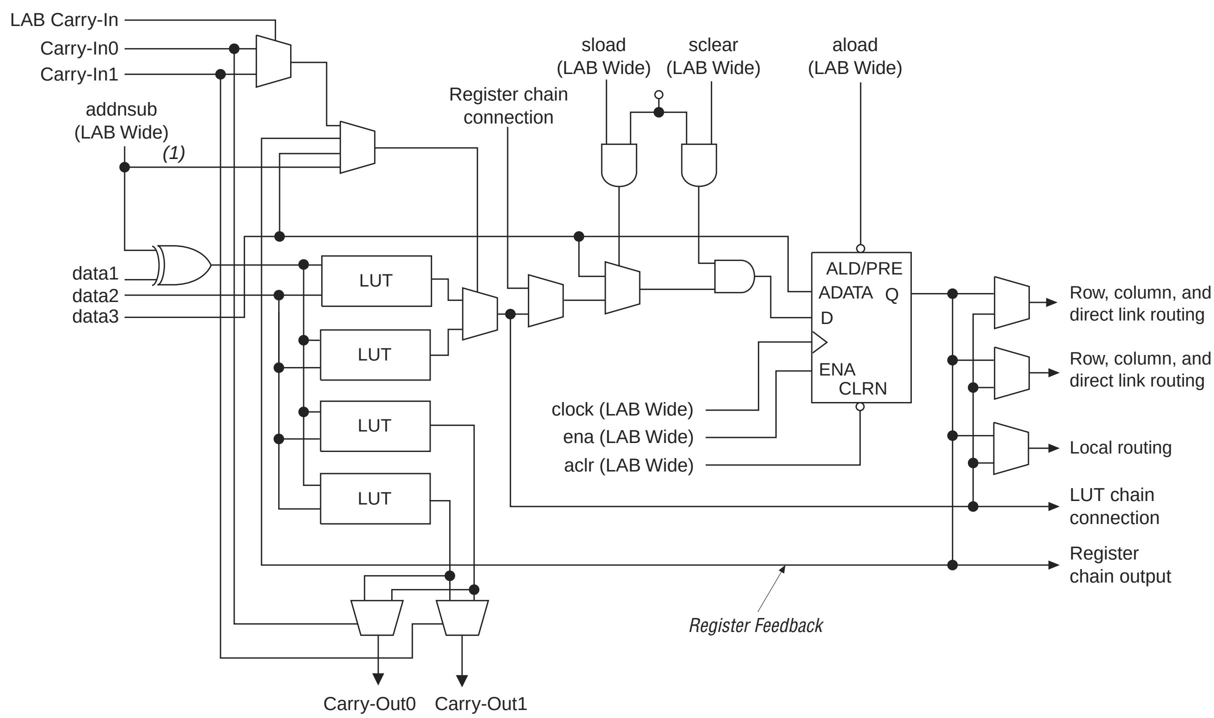
or

`data1 + data2 + carry-in1`

The other two LUTs use the data1 and data2 signals to generate two possible carry-out signals: one for a carry of 1 and the other for a carry of 0. The `carry-in0` signal acts as the carry-select for the `carry-out0` output and `carry-in1` acts as the carry-select for the `carry-out1` output. LEs in arithmetic mode can drive out registered and unregistered versions of the LUT output.

The dynamic arithmetic mode also offers clock enable, counter enable, synchronous up/down control, synchronous clear, synchronous load, and dynamic adder/subtractor options. The LAB local interconnect data inputs generate the counter enable and synchronous up/down control signals. The synchronous clear and synchronous load options are LAB-wide signals that affect all registers in the LAB. The Quartus II software automatically places any registers that are not used by the counter into other LABs. The addnsub LAB-wide signal controls whether the LE acts as an adder or subtractor.

Figure 2–8. LE in Dynamic Arithmetic Mode



Note to Figure 2–8:

(1) The `addnsub` signal is tied to the carry input for the first LE of a carry chain only.

Carry-Select Chain

The carry-select chain provides a very fast carry-select function between LEs in dynamic arithmetic mode. The carry-select chain uses the redundant carry calculation to increase the speed of carry functions. The LE is configured to calculate outputs for a possible carry-in of 0 and carry-in of 1 in parallel. The `carry-in0` and `carry-in1` signals from a lower-order bit feed forward into the higher-order bit via the parallel carry chain and feed into both the LUT and the next portion of the carry chain. Carry-select chains can begin in any LE within an LAB.

The UFM block communicates with the logic array similar to LAB-to-LAB interfaces. The UFM block connects to row and column interconnects and has local interconnect regions driven by row and column interconnects. This block also has DirectLink interconnects for fast connections to and from a neighboring LAB. For more information about the UFM interface to the logic array, see “User Flash Memory Block” on page 2–18.

Table 2–2 shows the MAX II device routing scheme.

Table 2–2. MAX II Device Routing Scheme

Source	Destination										
	LUT Chain	Register Chain	Local *(1)*	DirectLink *(1)*	R4 *(1)*	C4 *(1)*	LE	UFM Block	Column IOE	Row IOE	Fast I/O *(1)*
LUT Chain	—	—	—	—	—	—	✓	—	—	—	—
Register Chain	—	—	—	—	—	—	✓	—	—	—	—
Local Interconnect	—	—	—	—	—	—	✓	✓	✓	✓	—
DirectLink Interconnect	—	—	✓	—	—	—	—	—	—	—	—
R4 Interconnect	—	—	✓	—	✓	✓	—	—	—	—	—
C4 Interconnect	—	—	✓	—	✓	✓	—	—	—	—	—
LE	✓	✓	✓	✓	✓	✓	—	—	✓	✓	✓
UFM Block	—	—	✓	✓	✓	✓	—	—	—	—	—
Column IOE	—	—	—	—	—	✓	—	—	—	—	—
Row IOE	—	—	—	✓	✓	✓	—	—	—	—	—

Note to Table 2–2:

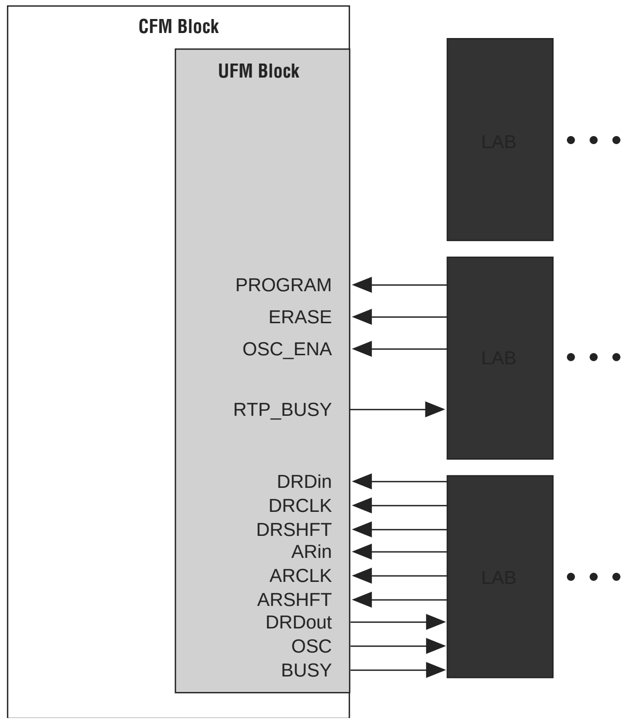
(1) These categories are interconnects.

Global Signals

Each MAX II device has four dual-purpose dedicated clock pins (`GCLK[3..0]`, two pins on the left side and two pins on the right side) that drive the global clock network for clocking, as shown in Figure 2–13. These four pins can also be used as general-purpose I/O if they are not used to drive the global clock network.

The four global clock lines in the global clock network drive throughout the entire device. The global clock network can provide clocks for all resources within the device including LEs, LAB local interconnect, IOEs, and the UFM block. The global clock lines can also be used for global control signals, such as clock enables, synchronous or asynchronous clears, presets, output enables, or protocol control signals such as `TRDY` and `IRDY` for PCI. Internal logic can drive the global clock network for internally-generated global clocks and control signals. Figure 2–13 shows the various sources that drive the global clock network.

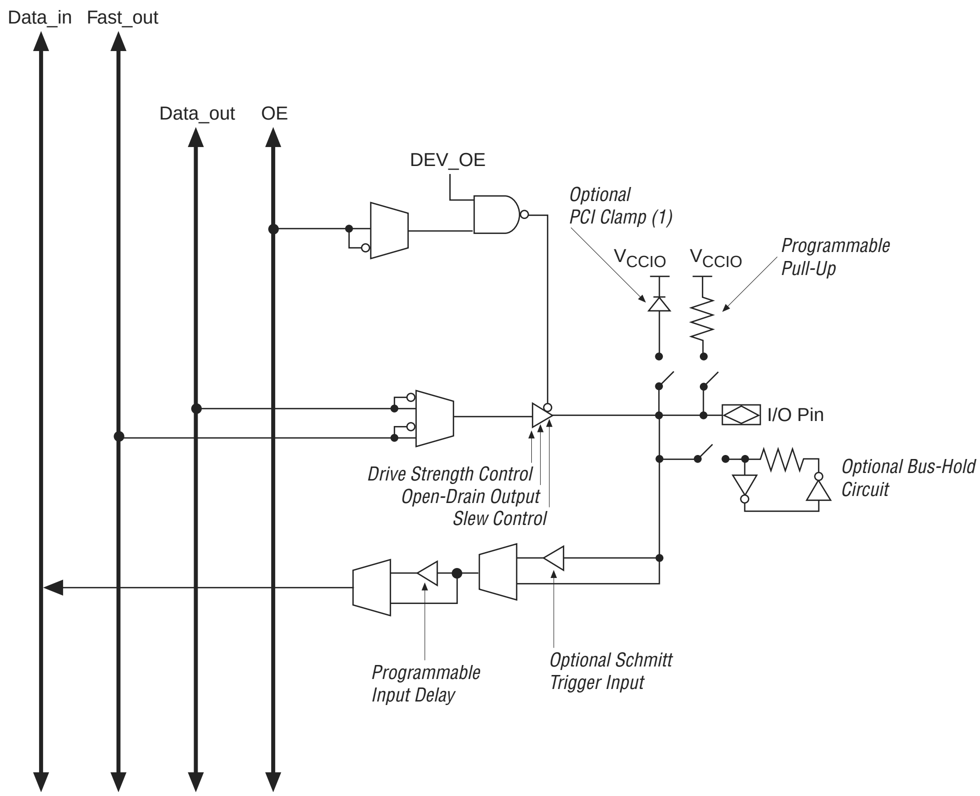
Figure 2–16. EPM240 UFM Block LAB Row Interface *(Note 1)*



Note to Figure 2–16:

(1) The UFM block inputs and outputs can drive to/from all types of interconnects, not only DirectLink interconnects from adjacent row LABs.

Figure 2–19. MAX II IOE Structure



Note to Figure 2–19:

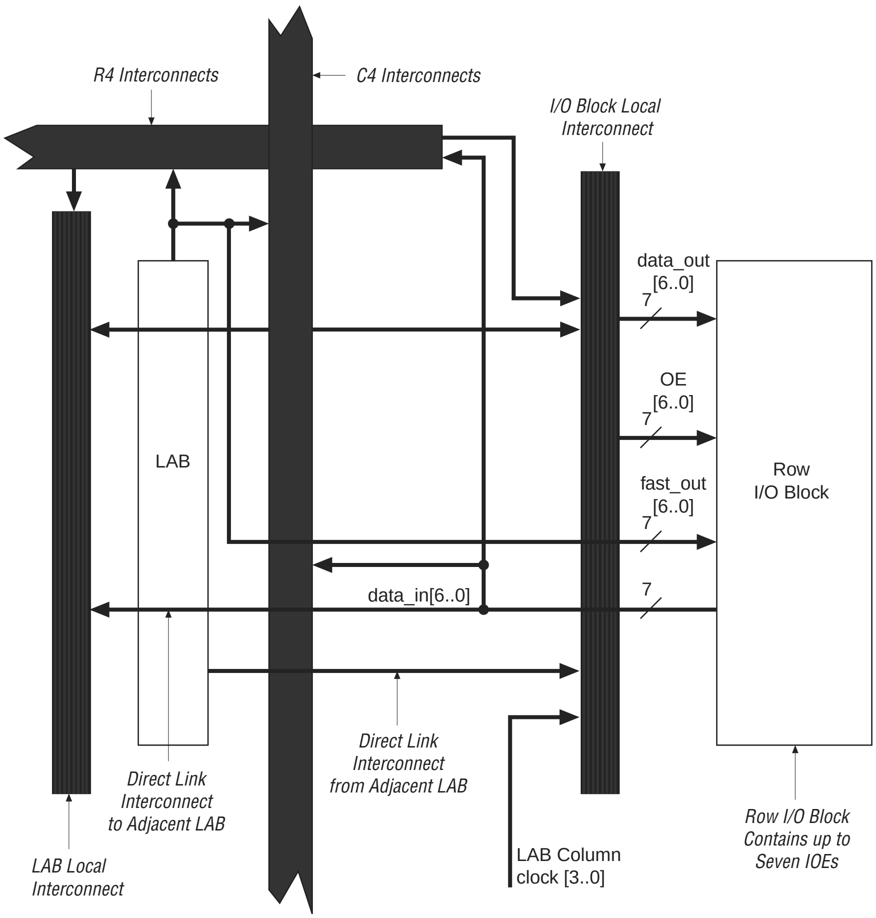
(1) Available in EPM1270 and EPM2210 devices only.

I/O Blocks

The IOEs are located in I/O blocks around the periphery of the MAX II device. There are up to seven IOEs per row I/O block (5 maximum in the EPM240 device) and up to four IOEs per column I/O block. Each column or row I/O block interfaces with its adjacent LAB and MultiTrack interconnect to distribute signals throughout the device. The row I/O blocks drive row, column, or DirectLink interconnects. The column I/O blocks drive column interconnects.

Figure 2–20 shows how a row I/O block connects to the logic array.

Figure 2–20. Row I/O Block Connection to the Interconnect *(Note 1)*



Note to Figure 2–20:

(1) Each of the seven IOEs in the row I/O block can have one `data_out` or `fast_out` output, one `OE` output, and one `data_in` input.

Table 3–1. MAX II JTAG Instructions (Part 2 of 2)

JTAG Instruction	Instruction Code	Description
`CLAMP` *(1)*	`00 0000 1010`	Places the 1-bit bypass register between the `TDI` and `TDO` pins, which allows the boundary scan test data to pass synchronously through selected devices to adjacent devices during normal device operation, while holding I/O pins to a state defined by the data in the boundary-scan register.
`USER0`	`00 0000 1100`	This instruction allows you to define the scan chain between `TDI` and `TDO` in the MAX II logic array. This instruction is also used for custom logic and JTAG interfaces.
`USER1`	`00 0000 1110`	This instruction allows you to define the scan chain between `TDI` and `TDO` in the MAX II logic array. This instruction is also used for custom logic and JTAG interfaces.
`IEEE 1532 instructions`	*(2)*	IEEE 1532 ISC instructions used when programming a MAX II device via the JTAG port.

Notes to Table 3–1:

(1) `HIGHZ`, `CLAMP`, and `EXTEST` instructions do not disable weak pull-up resistors or bus hold features.

(2) These instructions are shown in the 1532 BSDL files, which will be posted on the Altera® website at www.altera.com when they are available.

Unsupported JTAG instructions should not be issued to the MAX II device as this may put the device into an unknown state, requiring a power cycle to recover device operation.

The MAX II device instruction register length is 10 bits and the `USERCODE` register length is 32 bits. Table 3–2 and Table 3–3 show the boundary-scan register length and device `IDCODE` information for MAX II devices.

Table 3–2. MAX II Boundary-Scan Register Length

Device	Boundary-Scan Register Length
EPM240	240
EPM570	480
EPM1270	636
EPM2210	816

Table 3–3. 32-Bit MAX II Device IDCODE (Part 1 of 2)

Device	Binary IDCODE (32 Bits) *(1)*				HEX IDCODE
	Version (4 Bits)	Part Number	Manufacturer Identity (11 Bits)	LSB (1 Bit) *(2)*	
EPM240 EPM240G	`0000`	`0010 0000 1010 0001`	`000 0110 1110`	1	0x020A10DD
EPM570 EPM570G	`0000`	`0010 0000 1010 0010`	`000 0110 1110`	1	0x020A20DD
EPM1270 EPM1270G	`0000`	`0010 0000 1010 0011`	`000 0110 1110`	1	0x020A30DD
EPM2210 EPM2210G	`0000`	`0010 0000 1010 0100`	`000 0110 1110`	1	0x020A40DD

Table 3–4 shows the programming times for MAX II devices using in-circuit testers to execute the algorithm vectors in hardware. Software-based programming tools used with download cables are slightly slower because of data processing and transfer limitations.

Table 3–4. MAX II Device Family Programming Times

Description	EPM240 EPM240G EPM240Z	EPM570 EPM570G EPM570Z	EPM1270 EPM1270G	EPM2210 EPM2210G	Unit
Erase + Program (1 MHz)	1.72	2.16	2.90	3.92	sec
Erase + Program (10 MHz)	1.65	1.99	2.58	3.40	sec
Verify (1 MHz)	0.09	0.17	0.30	0.49	sec
Verify (10 MHz)	0.01	0.02	0.03	0.05	sec
Complete Program Cycle (1 MHz)	1.81	2.33	3.20	4.41	sec
Complete Program Cycle (10 MHz)	1.66	2.01	2.61	3.45	sec

UFM Programming

The Quartus II software, with the use of POF, Jam, or JBC files, supports programming of the user flash memory (UFM) block independent of the logic array design pattern stored in the CFM block. This allows updating or reading UFM contents through ISP without altering the current logic array design, or vice versa. By default, these programming files and methods will program the entire flash memory contents, which includes the CFM block and UFM contents. The stand-alone embedded Jam STAPL player and Jam Byte-Code Player provides action commands for programming or reading the entire flash memory (UFM and CFM together) or each independently.

For more information, refer to the *Using Jam STAPL for ISP via an Embedded Processor* chapter in the *MAX II Device Handbook*.

In-System Programming Clamp

By default, the IEEE 1532 instruction used for entering ISP automatically tri-states all I/O pins with weak pull-up resistors for the duration of the ISP sequence. However, some systems may require certain pins on MAX II devices to maintain a specific DC logic level during an in-field update. For these systems, an optional in-system programming clamp instruction exists in MAX II circuitry to control I/O behavior during the ISP sequence. The in-system programming clamp instruction enables the device to sample and sustain the value on an output pin (an input pin would remain tri-stated if sampled) or to explicitly set a logic high, logic low, or tri-state value on any pin. Setting these options is controlled on an individual pin basis using the Quartus II software.

For more information, refer to the *Real-Time ISP and ISP Clamp for MAX II Devices* chapter in the *MAX II Device Handbook*.

Document Revision History

Table 3–5 shows the revision history for this chapter.

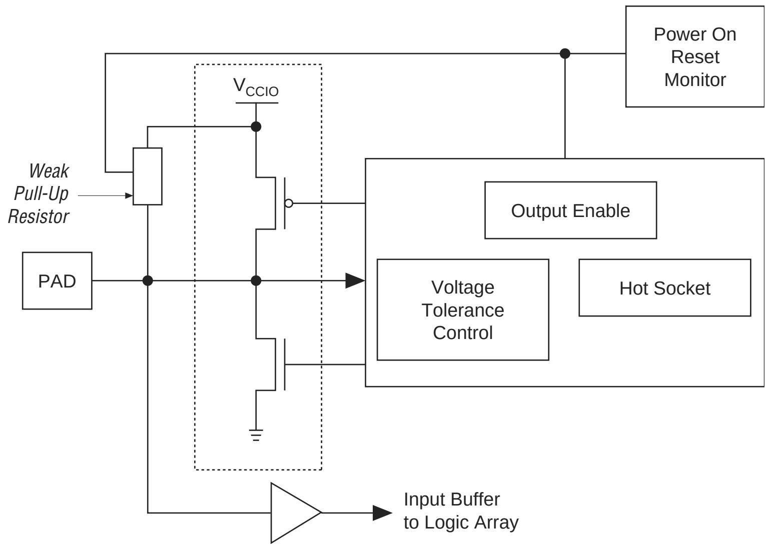
Table 3–5. Document Revision History

Date and Revision	Changes Made	Summary of Changes
October 2008, version 1.6	■ Updated New Document Format.	—
December 2007, version 1.5	■ Added warning note after Table 3–1. ■ Updated Table 3–3 and Table 3–4. ■ Added “Referenced Documents” section.	—
December 2006, version 1.4	■ Added document revision history.	—
June 2005, version 1.3	■ Added text and Table 3-4.	—
June 2005, version 1.3	■ Updated text on pages 3-5 to 3-8.	—
June 2004, version 1.1	■ Corrected Figure 3-1. Added CFM acronym.	—

Make sure that the V_{CCINT} is within the recommended operating range even though SRAM download has completed.

Each I/O and clock pin has the circuitry shown in Figure 4–1.

Figure 4–1. Hot Socketing Circuit Block Diagram for MAX II Devices

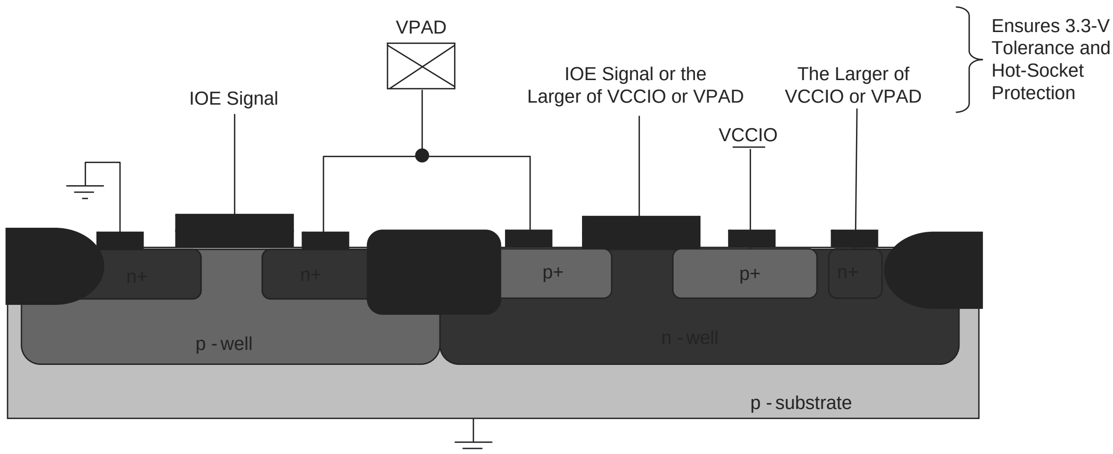


The POR circuit monitors V_{CCINT} and V_{CCIO} voltage levels and keeps I/O pins tri-stated until the device has completed its flash memory configuration of the SRAM logic. The weak pull-up resistor (R) from the I/O pin to V_{CCIO} is enabled during download to keep the I/O pins from floating. The 3.3-V tolerance control circuit permits the I/O pins to be driven by 3.3 V before V_{CCIO} and/or V_{CCINT} are powered, and it prevents the I/O pins from driving out when the device is not fully powered or operational. The hot socket circuit prevents I/O pins from internally powering V_{CCIO} and V_{CCINT} when driven by external signals before the device is powered.

For information about 5.0-V tolerance, refer to the *Using MAX II Devices in Multi-Voltage Systems* chapter in the *MAX II Device Handbook*.

Figure 4–2 shows a transistor-level cross section of the MAX II device I/O buffers. This design ensures that the output buffers do not drive when V_{CCIO} is powered before V_{CCINT} or if the I/O pad voltage is higher than V_{CCIO}. This also applies for sudden voltage spikes during hot insertion. The V_{PAD} leakage current charges the 3.3-V tolerant circuit capacitance.

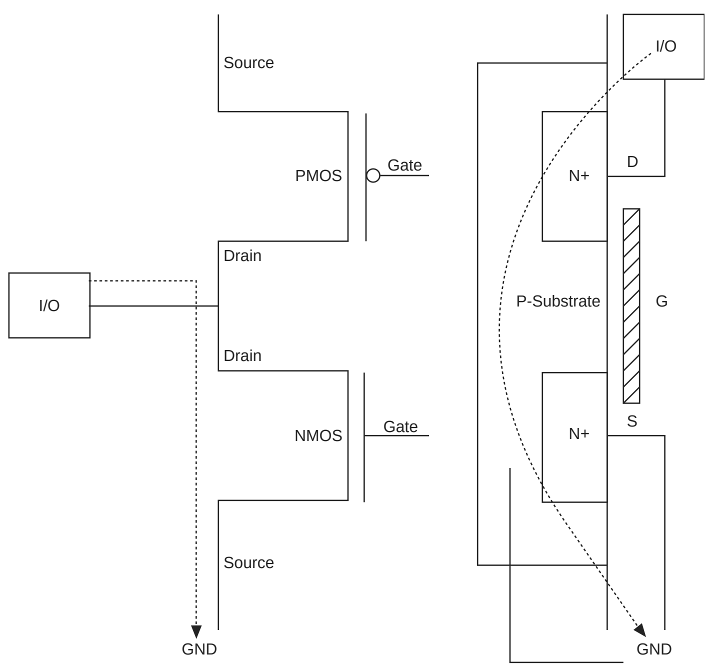
Figure 4–2. Transistor-Level Diagram of MAX II Device I/O Buffers



The CMOS output drivers in the I/O pins intrinsically provide electrostatic discharge (ESD) protection. There are two cases to consider for ESD voltage strikes: positive voltage zap and negative voltage zap.

A positive ESD voltage zap occurs when a positive voltage is present on an I/O pin due to an ESD charge event. This can cause the N+ (Drain)/ P-Substrate junction of the N-channel drain to break down and the N+ (Drain)/P-Substrate/N+ (Source) intrinsic bipolar transistor turn on to discharge ESD current from I/O pin to GND. The dashed line (see Figure 4–3) shows the ESD current discharge path during a positive ESD zap.

Figure 4–3. ESD Protection During Positive Voltage Zap



Programming/Erasure Specifications

Table 5–3 shows the MAX II device family programming/erasure specifications.

Table 5–3. MAX II Device Programming/Erasure Specifications

Parameter	Minimum	Typical	Maximum	Unit
Erase and reprogram cycles	—	—	100 *(1)*	Cycles

Note to Table 5–3:

(1) This specification applies to the UFM and configuration flash memory (CFM) blocks.

DC Electrical Characteristics

Table 5–4 shows the MAX II device family DC electrical characteristics.

Table 5–4. MAX II Device DC Electrical Characteristics *(Note 1)* (Part 1 of 2)

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Unit
I_I	Input pin leakage current	$V_I = V_{CCIO}$ max to 0 V *(2)*	–10	—	10	µA
I_{OZ}	Tri-stated I/O pin leakage current	$V_O = V_{CCIO}$ max to 0 V *(2)*	–10	—	10	µA
$I_{CCSTANDBY}$	V_{CCINT} supply current (standby) *(3)*	MAX II devices	—	12	—	mA
		MAX IIG devices	—	2	—	mA
		EPM240Z (Commercial grade) *(4)*	—	25	90	µA
		EPM240Z (Industrial grade) *(5)*	—	25	139	µA
		EPM570Z (Commercial grade) *(4)*	—	27	96	µA
		EPM570Z (Industrial grade) *(5)*	—	27	152	µA
$V_{SCHMITT}$ *(6)*	Hysteresis for Schmitt trigger input *(7)*	V_{CCIO} = 3.3 V	—	400	—	mV
		V_{CCIO} = 2.5 V	—	190	—	mV
$I_{CCPOWERUP}$	V_{CCINT} supply current during power-up *(8)*	MAX II devices	—	55	—	mA
		MAX IIG and MAX IIZ devices	—	40	—	mA
R_{PULLUP}	Value of I/O pin pull-up resistor during user mode and in-system programming	V_{CCIO} = 3.3 V *(9)*	5	—	25	kΩ
		V_{CCIO} = 2.5 V *(9)*	10	—	40	kΩ
		V_{CCIO} = 1.8 V *(9)*	25	—	60	kΩ
		V_{CCIO} = 1.5 V *(9)*	45	—	95	kΩ

Table 5–4. MAX II Device DC Electrical Characteristics *(Note 1)* (Part 2 of 2)

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Unit
I_{PULLUP}	I/O pin pull-up resistor current when I/O is unprogrammed	—	—	—	300	µA
C_{IO}	Input capacitance for user I/O pin	—	—	—	8	pF
C_{GCLK}	Input capacitance for dual-purpose GCLK/user I/O pin	—	—	—	8	pF

Notes to Table 5–4:

(1) Typical values are for T_A = 25°C, V_{CCINT} = 3.3 or 2.5 V, and V_{CCIO} = 1.5 V, 1.8 V, 2.5 V, or 3.3 V.

(2) This value is specified for normal device operation. The value may vary during power-up. This applies for all V_{CCIO} settings (3.3, 2.5, 1.8, and 1.5 V).

(3) V_I = ground, no load, no toggling inputs.

(4) Commercial temperature ranges from 0°C to 85°C with maximum current at 85°C.

(5) Industrial temperature ranges from –40°C to 100°C with maximum current at 100°C.

(6) This value applies to commercial and industrial range devices. For extended temperature range devices, the $V_{SCHMITT}$ typical value is 300 mV for V_{CCIO} = 3.3 V and 120 mV for V_{CCIO} = 2.5 V.

(7) The TCK input is susceptible to high pulse glitches when the input signal fall time is greater than 200 ns for all I/O standards.

(8) This is a peak current value with a maximum duration of t_{CONFIG} time.

(9) Pin pull-up resistance values will lower if an external source drives the pin higher than V_{CCIO}.

Power-Up Timing

Table 5–12 shows the power-up timing characteristics for MAX II devices.

Table 5–12. MAX II Power-Up Timing

Symbol	Parameter	Device	Min	Typ	Max	Unit
t_{CONFIG} *(1)*	The amount of time from when minimum V_{CCINT} is reached until the device enters user mode *(2)*	EPM240	—	—	200	µs
		EPM570	—	—	300	µs
		EPM1270	—	—	300	µs
		EPM2210	—	—	450	µs

Notes to Table 5–12:

(1) Table 5–12 values apply to commercial and industrial range devices. For extended temperature range devices, the t_{CONFIG} maximum values are as follows:

Device	Maximum
EPM240	300 µs
EPM570	400 µs
EPM1270	400 µs
EPM2210	500 µs

(2) For more information about POR trigger voltage, refer to the *Hot Socketing and Power-On Reset in MAX II Devices* chapter in the *MAX II Device Handbook*.

Power Consumption

Designers can use the Altera® PowerPlay Early Power Estimator and PowerPlay Power Analyzer to estimate the device power.

For more information about these power analysis tools, refer to the *Understanding and Evaluating Power in MAX II Devices* chapter in the *MAX II Device Handbook* and the *PowerPlay Power Analysis* chapter in volume 3 of the *Quartus II Handbook.*

Timing Model and Specifications

MAX II devices timing can be analyzed with the Altera Quartus® II software, a variety of popular industry-standard EDA simulators and timing analyzers, or with the timing model shown in Figure 5–2.

MAX II devices have predictable internal delays that enable the designer to determine the worst-case timing of any design. The software provides timing simulation, point-to-point delay prediction, and detailed timing analysis for device-wide performance evaluation.

Table 5–13. MAX II Device Timing Model Status (Part 2 of 2)

Device	Preliminary	Final
EPM1270	—	✓
EPM2210	—	✓

Note to Table 5–13:

(1) The MAX IIZ device timing models are only available in the Quartus II software version 8.0 and later.

Performance

Table 5–14 shows the MAX II device performance for some common designs. All performance values were obtained with the Quartus II software compilation of megafunctions. Performance values for –3, –4, and –5 speed grades are based on an EPM1270 device target, while –6, –7, and –8 speed grades are based on an EPM570Z device target.

Table 5–14. MAX II Device Performance

Resource Used	Design Size and Function	Resources Used			Performance						Unit
					MAX II / MAX IIG			MAX IIZ			
		Mode	LEs	UFM Blocks	–3 Speed Grade	–4 Speed Grade	–5 Speed Grade	–6 Speed Grade	–7 Speed Grade	–8 Speed Grade	
LE	16-bit counter *(1)*	—	16	0	304.0	247.5	201.1	184.1	123.5	118.3	MHz
	64-bit counter *(1)*	—	64	0	201.5	154.8	125.8	83.2	83.2	80.5	MHz
	16-to-1 multiplexer	—	11	0	6.0	8.0	9.3	17.4	17.3	20.4	ns
	32-to-1 multiplexer	—	24	0	7.1	9.0	11.4	12.5	22.8	25.3	ns
	16-bit XOR function	—	5	0	5.1	6.6	8.2	9.0	15.0	16.1	ns
	16-bit decoder with single address line	—	5	0	5.2	6.6	8.2	9.2	15.0	16.1	ns
UFM	512 × 16	None	3	1	10.0	10.0	10.0	10.0	10.0	10.0	MHz
	512 × 16	SPI *(2)*	37	1	8.0	8.0	8.0	9.7	9.7	9.7	MHz
	512 × 8	Parallel *(3)*	73	1	*(4)*	*(4)*	*(4)*	*(4)*	*(4)*	*(4)*	MHz
	512 × 16	I^2C *(3)*	142	1	100 *(5)*	100 *(5)*	100 *(5)*	100 *(5)*	100 *(5)*	100 *(5)*	kHz

Notes to Table 5–14:

(1) This design is a binary loadable up counter.
(2) This design is configured for read-only operation in Extended mode. Read and write ability increases the number of LEs used.
(3) This design is configured for read-only operation. Read and write ability increases the number of LEs used.
(4) This design is asynchronous.
(5) The I^2C megafunction is verified in hardware up to 100-kHz serial clock line (SCL) rate.

Internal Timing Parameters

Internal timing parameters are specified on a speed grade basis independent of device density. Table 5–15 through Table 5–22 describe the MAX II device internal timing microparameters for logic elements (LEs), input/output elements (IOEs), UFM blocks, and MultiTrack interconnects. The timing values for –3, –4, and –5 speed grades shown in Table 5–15 through Table 5–22 are based on an EPM1270 device target, while –6, –7, and –8 speed grade values are based on an EPM570Z device target.

For more explanations and descriptions about each internal timing microparameters symbol, refer to the *Understanding Timing in MAX II Devices* chapter in the *MAX II Device Handbook*.

Table 5–15. LE Internal Timing Microparameters

Symbol	Parameter	MAX II / MAX IIG						MAX IIZ						Unit
		–3 Speed Grade		–4 Speed Grade		–5 Speed Grade		–6 Speed Grade		–7 Speed Grade		–8 Speed Grade		
		Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	
t_{LUT}	LE combinational LUT delay	—	571	—	742	—	914	—	1,215	—	2,247	—	2,247	ps
t_{COMB}	Combinational path delay	—	147	—	192	—	236	—	243	—	305	—	309	ps
t_{CLR}	LE register clear delay	238	—	309	—	381	—	401	—	541	—	545	—	ps
t_{PRE}	LE register preset delay	238	—	309	—	381	—	401	—	541	—	545	—	ps
t_{SU}	LE register setup time before clock	208	—	271	—	333	—	260	—	319	—	321	—	ps
t_H	LE register hold time after clock	0	—	0	—	0	—	0	—	0	—	0	—	ps
t_{CO}	LE register clock-to-output delay	—	235	—	305	—	376	—	380	—	489	—	494	ps
t_{CLKHL}	Minimum clock high or low time	166	—	216	—	266	—	253	—	335	—	339	—	ps
t_C	Register control delay	—	857	—	1,114	—	1,372	—	1,356	—	1,722	—	1,741	ps

Table 5–31. MAX II IOE Programmable Delays

Parameter	MAX II / MAX IIG						MAX IIZ						Unit
	–3 Speed Grade		–4 Speed Grade		–5 Speed Grade		–6 Speed Grade		–7 Speed Grade		–8 Speed Grade		
	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	
Input Delay from Pin to Internal Cells = 1	—	1,225	—	1,592	—	1,960	—	1,858	—	2,171	—	2,214	ps
Input Delay from Pin to Internal Cells = 0	—	89	—	115	—	142	—	569	—	609	—	616	ps

Maximum Input and Output Clock Rates

Table 5–32 and Table 5–33 show the maximum input and output clock rates for standard I/O pins in MAX II devices.

Table 5–32. MAX II Maximum Input Clock Rate for I/O

I/O Standard		MAX II / MAX IIG			MAX IIZ			Unit
		–3 Speed Grade	–4 Speed Grade	–5 Speed Grade	–6 Speed Grade	–7 Speed Grade	–8 Speed Grade	
3.3-V LVTTL	Without Schmitt Trigger	304	304	304	304	304	304	MHz
	With Schmitt Trigger	250	250	250	250	250	250	MHz
3.3-V LVCMOS	Without Schmitt Trigger	304	304	304	304	304	304	MHz
	With Schmitt Trigger	250	250	250	250	250	250	MHz
2.5-V LVTTL	Without Schmitt Trigger	220	220	220	220	220	220	MHz
	With Schmitt Trigger	188	188	188	188	188	188	MHz
2.5-V LVCMOS	Without Schmitt Trigger	220	220	220	220	220	220	MHz
	With Schmitt Trigger	188	188	188	188	188	188	MHz
1.8-V LVTTL	Without Schmitt Trigger	200	200	200	200	200	200	MHz
1.8-V LVCMOS	Without Schmitt Trigger	200	200	200	200	200	200	MHz
1.5-V LVCMOS	Without Schmitt Trigger	150	150	150	150	150	150	MHz
3.3-V PCI	Without Schmitt Trigger	304	304	304	304	304	304	MHz