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Understanding [Embedded - CPLDs \(Complex Programmable Logic Devices\)](#)

Embedded - CPLDs, or Complex Programmable Logic Devices, are highly versatile digital logic devices used in electronic systems. These programmable components are designed to perform complex logical operations and can be customized for specific applications. Unlike fixed-function ICs, CPLDs offer the flexibility to reprogram their configuration, making them an ideal choice for various embedded systems. They consist of a set of logic gates and programmable interconnects, allowing designers to implement complex logic circuits without needing custom hardware.

Applications of Embedded - CPLDs

Details

Product Status	Active
Programmable Type	In System Programmable
Delay Time tpd(1) Max	5.4 ns
Voltage Supply - Internal	1.71V ~ 1.89V
Number of Logic Elements/Blocks	570
Number of Macrocells	440
Number of Gates	-
Number of I/O	76
Operating Temperature	0°C ~ 85°C (TJ)
Mounting Type	Surface Mount
Package / Case	100-TQFP
Supplier Device Package	100-TQFP (14x14)
Purchase URL	https://www.e-xfl.com/pro/item?MUrl=&PartUrl=epm570gt100c3

MAX II devices are available in space-saving FineLine BGA, Micro FineLine BGA, and thin quad flat pack (TQFP) packages (refer to [Table 1-3](#) and [Table 1-4](#)). MAX II devices support vertical migration within the same package (for example, you can migrate between the EPM570, EPM1270, and EPM2210 devices in the 256-pin FineLine BGA package). Vertical migration means that you can migrate to devices whose dedicated pins and JTAG pins are the same and power pins are subsets or supersets for a given package across device densities. The largest density in any package has the highest number of power pins; you must lay out for the largest planned density in a package to provide the necessary power pins for migration. For I/O pin migration across densities, cross reference the available I/O pins using the device pin-outs for all planned densities of a given package type to identify which I/O pins can be migrated. The Quartus® II software can automatically cross-reference and place all pins for you when given a device migration list.

Table 1-3. MAX II Packages and User I/O Pins

Device	68-Pin Micro FineLine BGA (1)	100-Pin Micro FineLine BGA (1)	100-Pin FineLine BGA	100-Pin TQFP	144-Pin TQFP	144-Pin Micro FineLine BGA (1)	256-Pin Micro FineLine BGA (1)	256-Pin FineLine BGA	324-Pin FineLine BGA
EPM240 EPM240G	—	80	80	80	—	—	—	—	—
EPM570 EPM570G	—	76	76	76	116	—	160	160	—
EPM1270 EPM1270G	—	—	—	—	116	—	212	212	—
EPM2210 EPM2210G	—	—	—	—	—	—	—	204	272
EPM240Z	54	80	—	—	—	—	—	—	—
EPM570Z	—	76	—	—	—	116	160	—	—

Note to Table 1-3:

(1) Packages available in lead-free versions only.

Table 1-4. MAX II TQFP, FineLine BGA, and Micro FineLine BGA Package Sizes

Package	68-Pin Micro FineLine BGA	100-Pin Micro FineLine BGA	100-Pin FineLine BGA	100-Pin TQFP	144-Pin TQFP	144-Pin Micro FineLine BGA	256-Pin Micro FineLine BGA	256-Pin FineLine BGA	324-Pin FineLine BGA
Pitch (mm)	0.5	0.5	1	0.5	0.5	0.5	0.5	1	1
Area (mm ²)	25	36	121	256	484	49	121	289	361
Length × width (mm × mm)	5 × 5	6 × 6	11 × 11	16 × 16	22 × 22	7 × 7	11 × 11	17 × 17	19 × 19

MAX II devices have an internal linear voltage regulator which supports external supply voltages of 3.3 V or 2.5 V, regulating the supply down to the internal operating voltage of 1.8 V. MAX IIG and MAX IIZ devices only accept 1.8 V as the external supply voltage. MAX IIZ devices are pin-compatible with MAX IIG devices in the 100-pin Micro FineLine BGA and 256-pin Micro FineLine BGA packages. Except for external supply voltage requirements, MAX II and MAX II G devices have identical pin-outs and timing specifications. Table 1–5 shows the external supply voltages supported by the MAX II family.

Table 1–5. MAX II External Supply Voltages

Devices	EPM240 EPM570 EPM1270 EPM2210	EPM240G EPM570G EPM1270G EPM2210G EPM240Z EPM570Z (1)
MultiVolt core external supply voltage (V_{CCINT}) (2)	3.3 V, 2.5 V	1.8 V
MultiVolt I/O interface voltage levels (V_{CCIO})	1.5 V, 1.8 V, 2.5 V, 3.3 V	1.5 V, 1.8 V, 2.5 V, 3.3 V

Notes to Table 1–5:

- (1) MAX IIG and MAX IIZ devices only accept 1.8 V on their V_{CCINT} pins. The 1.8-V V_{CCINT} external supply powers the device core directly.
- (2) MAX II devices operate internally at 1.8 V.

Referenced Documents

This chapter references the following documents:

- *DC and Switching Characteristics* chapter in the *MAX II Device Handbook*
- *MAX II Logic Element to Macrocell Conversion Methodology* white paper

Document Revision History

Table 1–6 shows the revision history for this chapter.

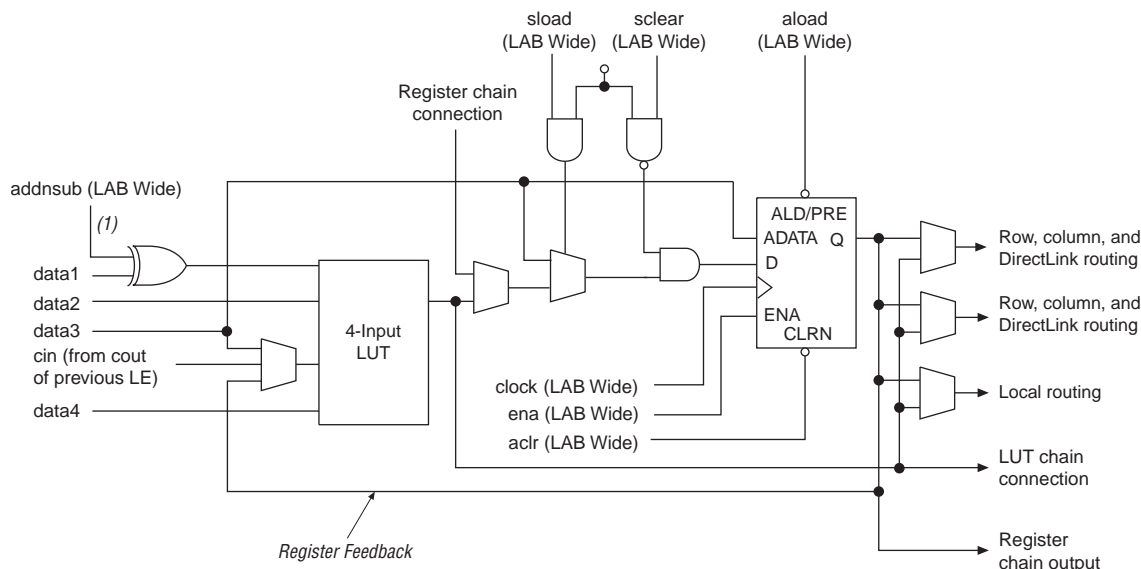
Table 1–6. Document Revision History

Date and Revision	Changes Made	Summary of Changes
August 2009, version 1.9	■ Updated Table 1–2.	Added information for speed grade –8
October 2008, version 1.8	■ Updated “Introduction” section. ■ Updated new Document Format.	—
December 2007, version 1.7	■ Updated Table 1–1 through Table 1–5. ■ Added “Referenced Documents” section.	Updated document with MAX IIZ information.
December 2006, version 1.6	■ Added document revision history.	—
August 2006, version 1.5	■ Minor update to features list.	—
July 2006, version 1.4	■ Minor updates to tables.	—

Normal Mode

The normal mode is suitable for general logic applications and combinational functions. In normal mode, four data inputs from the LAB local interconnect are inputs to a four-input LUT (see Figure 2-7). The Quartus II Compiler automatically selects the carry-in or the data3 signal as one of the inputs to the LUT. Each LE can use LUT chain connections to drive its combinational output directly to the next LE in the LAB. Asynchronous load data for the register comes from the data3 input of the LE. LEs in normal mode support packed registers.

Figure 2-7. LE in Normal Mode



Note to Figure 2-7:

(1) This signal is only allowed in normal mode if the LE is at the end of an adder/subtractor chain.

Dynamic Arithmetic Mode

The dynamic arithmetic mode is ideal for implementing adders, counters, accumulators, wide parity functions, and comparators. An LE in dynamic arithmetic mode uses four 2-input LUTs configurable as a dynamic adder/subtractor. The first two 2-input LUTs compute two summations based on a possible carry-in of 1 or 0; the other two LUTs generate carry outputs for the two chains of the carry-select circuitry. As shown in Figure 2-8, the LAB carry-in signal selects either the carry-in0 or carry-in1 chain. The selected chain's logic level in turn determines which parallel sum is generated as a combinational or registered output. For example, when implementing an adder, the sum output is the selection of two possible calculated sums:

$$\text{data1} + \text{data2} + \text{carry in0}$$

or

$$\text{data1} + \text{data2} + \text{carry-in1}$$

The Quartus II software automatically creates carry chain logic during design processing, or you can create it manually during design entry. Parameterized functions such as LPM functions automatically take advantage of carry chains for the appropriate functions. The Quartus II software creates carry chains longer than 10 LEs by linking adjacent LABs within the same row together automatically. A carry chain can extend horizontally up to one full LAB row, but does not extend between LAB rows.

Clear and Preset Logic Control

LAB-wide signals control the logic for the register's clear and preset signals. The LE directly supports an asynchronous clear and preset function. The register preset is achieved through the asynchronous load of a logic high. MAX II devices support simultaneous preset/asynchronous load and clear signals. An asynchronous clear signal takes precedence if both signals are asserted simultaneously. Each LAB supports up to two clears and one preset signal.

In addition to the clear and preset ports, MAX II devices provide a chip-wide reset pin (`DEV_CLRn`) that resets all registers in the device. An option set before compilation in the Quartus II software controls this pin. This chip-wide reset overrides all other control signals and uses its own dedicated routing resources (that is, it does not use any of the four global resources). Driving this signal low before or during power-up prevents user mode from releasing clears within the design. This allows you to control when clear is released on a device that has just been powered-up. If not set for its chip-wide reset function, the `DEV_CLRn` pin is a regular I/O pin.

By default, all registers in MAX II devices are set to power-up low. However, this power-up state can be set to high on individual registers during design entry using the Quartus II software.

MultiTrack Interconnect

In the MAX II architecture, connections between LEs, the UFM, and device I/O pins are provided by the MultiTrack interconnect structure. The MultiTrack interconnect consists of continuous, performance-optimized routing lines used for inter- and intra-design block connectivity. The Quartus II Compiler automatically places critical design paths on faster interconnects to improve design performance.

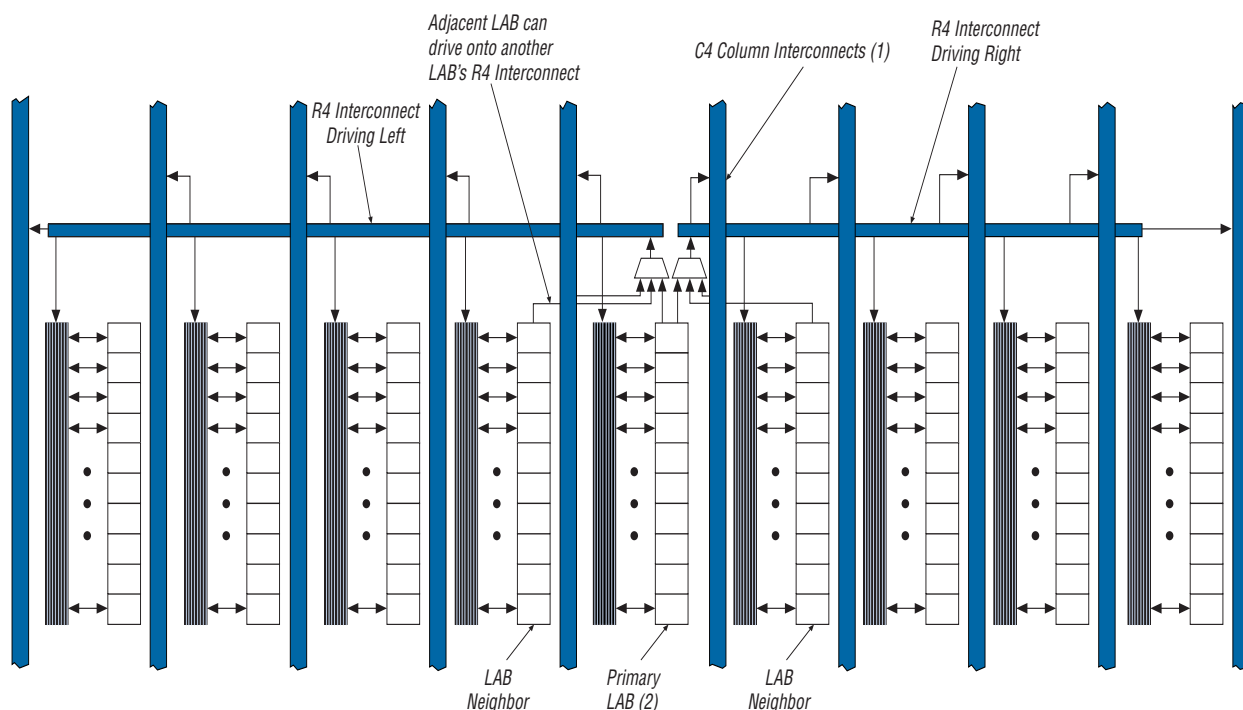
The MultiTrack interconnect consists of row and column interconnects that span fixed distances. A routing structure with fixed length resources for all devices allows predictable and short delays between logic levels instead of large delays associated with global or long routing lines. Dedicated row interconnects route signals to and from LABs within the same row. These row resources include:

- DirectLink interconnects between LABs
- R4 interconnects traversing four LABs to the right or left

The DirectLink interconnect allows an LAB to drive into the local interconnect of its left and right neighbors. The DirectLink interconnect provides fast communication between adjacent LABs and/or blocks without using row interconnect resources.

The R4 interconnects span four LABs and are used for fast row connections in a four-LAB region. Every LAB has its own set of R4 interconnects to drive either left or right. Figure 2-10 shows R4 interconnect connections from an LAB. R4 interconnects can drive and be driven by row IOEs. For LAB interfacing, a primary LAB or horizontal LAB neighbor can drive a given R4 interconnect. For R4 interconnects that drive to the right, the primary LAB and right neighbor can drive on to the interconnect. For R4 interconnects that drive to the left, the primary LAB and its left neighbor can drive on to the interconnect. R4 interconnects can drive other R4 interconnects to extend the range of LABs they can drive. R4 interconnects can also drive C4 interconnects for connections from one row to another.

Figure 2-10. R4 Interconnect Connections



Notes to Figure 2-10:

- (1) C4 interconnects can drive R4 interconnects.
- (2) This pattern is repeated for every LAB in the LAB row.

The column interconnect operates similarly to the row interconnect. Each column of LABs is served by a dedicated column interconnect, which vertically routes signals to and from LABs and row and column IOEs. These column resources include:

- LUT chain interconnects within an LAB
- Register chain interconnects within an LAB
- C4 interconnects traversing a distance of four LABs in an up and down direction

MAX II devices include an enhanced interconnect structure within LABs for routing LE output to LE input connections faster using LUT chain connections and register chain connections. The LUT chain connection allows the combinational output of an LE to directly drive the fast input of the LE right below it, bypassing the local interconnect. These resources can be used as a high-speed connection for wide fan-in

The UFM block communicates with the logic array similar to LAB-to-LAB interfaces. The UFM block connects to row and column interconnects and has local interconnect regions driven by row and column interconnects. This block also has DirectLink interconnects for fast connections to and from a neighboring LAB. For more information about the UFM interface to the logic array, see “User Flash Memory Block” on page 2-18.

Table 2-2 shows the MAX II device routing scheme.

Table 2-2. MAX II Device Routing Scheme

Source	Destination										
	LUT Chain	Register Chain	Local (1)	DirectLink (1)	R4 (1)	C4 (1)	LE	UFM Block	Column IOE	Row IOE	Fast I/O (1)
LUT Chain	—	—	—	—	—	—	✓	—	—	—	—
Register Chain	—	—	—	—	—	—	✓	—	—	—	—
Local Interconnect	—	—	—	—	—	—	✓	✓	✓	✓	—
DirectLink Interconnect	—	—	✓	—	—	—	—	—	—	—	—
R4 Interconnect	—	—	✓	—	✓	✓	—	—	—	—	—
C4 Interconnect	—	—	✓	—	✓	✓	—	—	—	—	—
LE	✓	✓	✓	✓	✓	✓	—	—	✓	✓	✓
UFM Block	—	—	✓	✓	✓	✓	—	—	—	—	—
Column IOE	—	—	—	—	—	✓	—	—	—	—	—
Row IOE	—	—	—	✓	✓	✓	—	—	—	—	—

Note to Table 2-2:

(1) These categories are interconnects.

Global Signals

Each MAX II device has four dual-purpose dedicated clock pins (GCLK[3..0], two pins on the left side and two pins on the right side) that drive the global clock network for clocking, as shown in Figure 2-13. These four pins can also be used as general-purpose I/O if they are not used to drive the global clock network.

The four global clock lines in the global clock network drive throughout the entire device. The global clock network can provide clocks for all resources within the device including LEs, LAB local interconnect, IOEs, and the UFM block. The global clock lines can also be used for global control signals, such as clock enables, synchronous or asynchronous clears, presets, output enables, or protocol control signals such as TRDY and IRDY for PCI. Internal logic can drive the global clock network for internally-generated global clocks and control signals. Figure 2-13 shows the various sources that drive the global clock network.

Document Revision History

Table 2-8 shows the revision history for this chapter.

Table 2-8. Document Revision History

Date and Revision	Changes Made	Summary of Changes
October 2008, version 2.2	<ul style="list-style-type: none"> ■ Updated Table 2-4 and Table 2-6. ■ Updated “I/O Standards and Banks” section. ■ Updated New Document Format. 	—
March 2008, version 2.1	<ul style="list-style-type: none"> ■ Updated “Schmitt Trigger” section. 	—
December 2007, version 2.0	<ul style="list-style-type: none"> ■ Updated “Clear and Preset Logic Control” section. ■ Updated “MultiVolt Core” section. ■ Updated “MultiVolt I/O Interface” section. ■ Updated Table 2-7. ■ Added “Referenced Documents” section. 	Updated document with MAX IIZ information.
December 2006, version 1.7	<ul style="list-style-type: none"> ■ Minor update in “Internal Oscillator” section. Added document revision history. 	—
August 2006, version 1.6	<ul style="list-style-type: none"> ■ Updated functional description and I/O structure sections. 	—
July 2006, version 1.5	<ul style="list-style-type: none"> ■ Minor content and table updates. 	—
February 2006, version 1.4	<ul style="list-style-type: none"> ■ Updated “LAB Control Signals” section. ■ Updated “Clear and Preset Logic Control” section. ■ Updated “Internal Oscillator” section. ■ Updated Table 2-5. 	—
August 2005, version 1.3	<ul style="list-style-type: none"> ■ Removed Note 2 from Table 2-7. 	—
December 2004, version 1.2	<ul style="list-style-type: none"> ■ Added a paragraph to page 2-15. 	—
June 2004, version 1.1	<ul style="list-style-type: none"> ■ Added CFM acronym. Corrected Figure 2-19. 	—

I/O Pins Remain Tri-Stated during Power-Up

A device that does not support hot-socketing may interrupt system operation or cause contention by driving out before or during power-up. In a hot socketing situation, the MAX II device's output buffers are turned off during system power-up. MAX II devices do not drive out until the device attains proper operating conditions and is fully configured. Refer to [“Power-On Reset Circuitry” on page 4-5](#) for information about turn-on voltages.

Signal Pins Do Not Drive the V_{CCIO} or V_{CCINT} Power Supplies

MAX II devices do not have a current path from I/O pins or $GCLK[3..0]$ pins to the V_{CCIO} or V_{CCINT} pins before or during power-up. A MAX II device may be inserted into (or removed from) a system board that was powered up without damaging or interfering with system-board operation. When hot socketing, MAX II devices may have a minimal effect on the signal integrity of the backplane.

AC and DC Specifications

You can power up or power down the V_{CCIO} and V_{CCINT} pins in any sequence. During hot socketing, the I/O pin capacitance is less than 8 pF. MAX II devices meet the following hot socketing specifications:

- The hot socketing DC specification is: $|I_{IOPIN}| < 300 \mu A$.
- The hot socketing AC specification is: $|I_{IOPIN}| < 8 \text{ mA}$ for 10 ns or less.




MAX II devices are immune to latch-up when hot socketing. If the TCK JTAG input pin is driven high during hot socketing, the current on that pin might exceed the specifications above.

I_{IOPIN} is the current at any user I/O pin on the device. The AC specification applies when the device is being powered up or powered down. This specification takes into account the pin capacitance but not board trace and external loading capacitance. Additional capacitance for trace, connector, and loading must be taken into consideration separately. The peak current duration due to power-up transients is 10 ns or less.

The DC specification applies when all V_{CC} supplies to the device are stable in the powered-up or powered-down conditions.

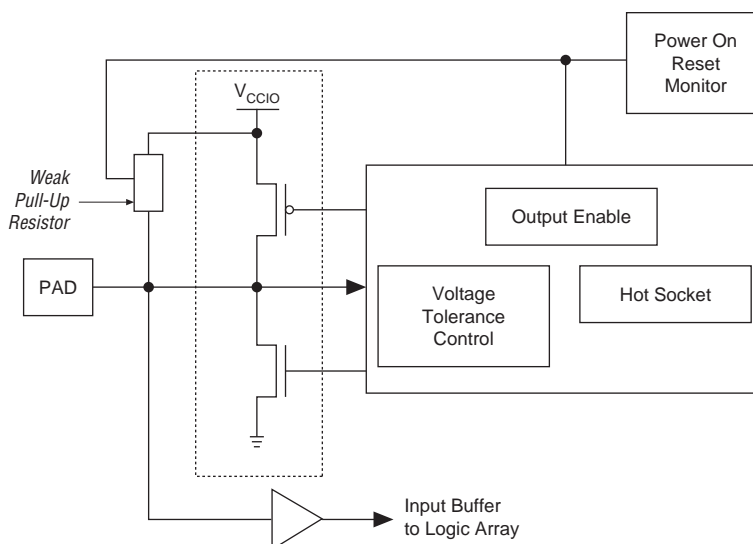
Hot Socketing Feature Implementation in MAX II Devices

The hot socketing feature turns off (tri-states) the output buffer during the power-up event (either V_{CCINT} or V_{CCIO} supplies) or power-down event. The hot-socket circuit generates an internal HOTSKT signal when either V_{CCINT} or V_{CCIO} is below the threshold voltage during power-up or power-down. The HOTSKT signal cuts off the output buffer to make sure that no DC current (except for weak pull-up leaking) leaks through the pin. When V_{CC} ramps up very slowly during power-up, V_{CC} may still be relatively low even after the power-on reset (POR) signal is released and device configuration is complete.


 Make sure that the V_{CCINT} is within the recommended operating range even though SRAM download has completed.

Each I/O and clock pin has the circuitry shown in [Figure 4-1](#).

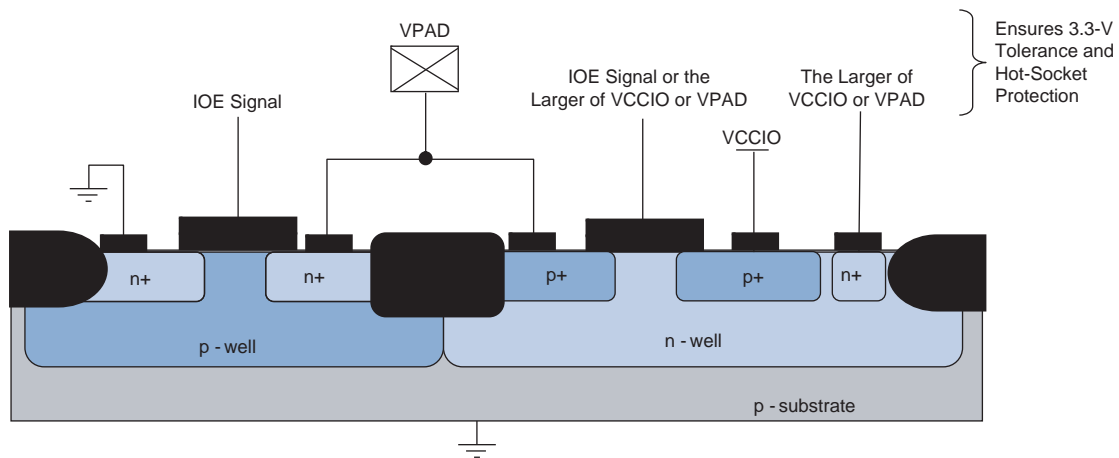
Figure 4-1. Hot Socketing Circuit Block Diagram for MAX II Devices



The POR circuit monitors V_{CCINT} and V_{CCIO} voltage levels and keeps I/O pins tri-stated until the device has completed its flash memory configuration of the SRAM logic. The weak pull-up resistor (R) from the I/O pin to V_{CCIO} is enabled during download to keep the I/O pins from floating. The 3.3-V tolerance control circuit permits the I/O pins to be driven by 3.3 V before V_{CCIO} and/or V_{CCINT} are powered, and it prevents the I/O pins from driving out when the device is not fully powered or operational. The hot socket circuit prevents I/O pins from internally powering V_{CCIO} and V_{CCINT} when driven by external signals before the device is powered.

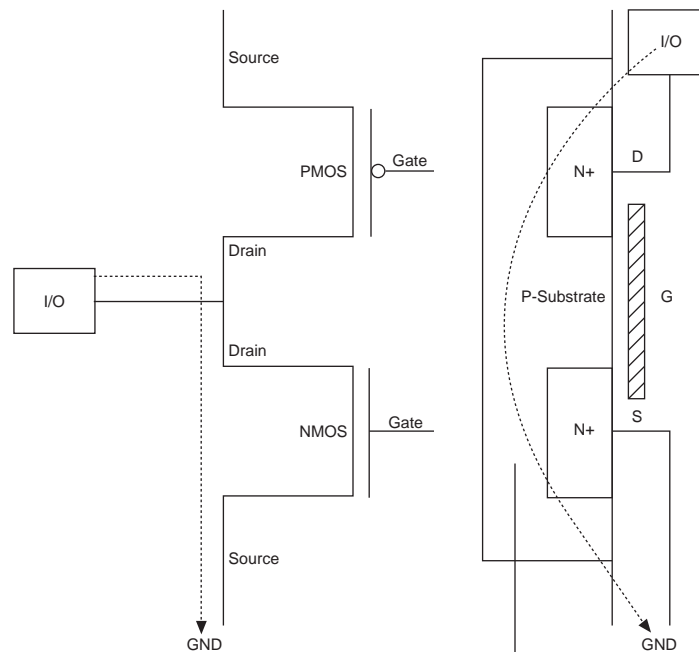
 For information about 5.0-V tolerance, refer to the [Using MAX II Devices in Multi-Voltage Systems](#) chapter in the *MAX II Device Handbook*.

[Figure 4-2](#) shows a transistor-level cross section of the MAX II device I/O buffers. This design ensures that the output buffers do not drive when V_{CCIO} is powered before V_{CCINT} or if the I/O pad voltage is higher than V_{CCIO} . This also applies for sudden voltage spikes during hot insertion. The V_{PAD} leakage current charges the 3.3-V tolerant circuit capacitance.

Figure 4-2. Transistor-Level Diagram of MAX II Device I/O Buffers

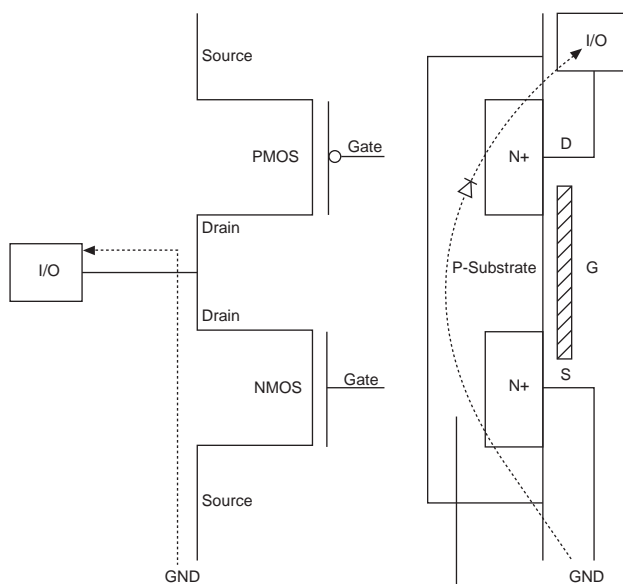
The CMOS output drivers in the I/O pins intrinsically provide electrostatic discharge (ESD) protection. There are two cases to consider for ESD voltage strikes: positive voltage zap and negative voltage zap.

A positive ESD voltage zap occurs when a positive voltage is present on an I/O pin due to an ESD charge event. This can cause the N+ (Drain)/ P-Substrate junction of the N-channel drain to break down and the N+ (Drain)/P-Substrate/N+ (Source) intrinsic bipolar transistor turn on to discharge ESD current from I/O pin to GND. The dashed line (see [Figure 4-3](#)) shows the ESD current discharge path during a positive ESD zap.

Figure 4-3. ESD Protection During Positive Voltage Zap

When the I/O pin receives a negative ESD zap at the pin that is less than -0.7 V (0.7 V is the voltage drop across a diode), the intrinsic P-Substrate/N+ drain diode is forward biased. Therefore, the discharge ESD current path is from GND to the I/O pin, as shown in Figure 4-4.

Figure 4-4. ESD Protection During Negative Voltage Zap



Power-On Reset Circuitry

MAX II devices have POR circuits to monitor V_{CCINT} and V_{CCIO} voltage levels during power-up. The POR circuit monitors these voltages, triggering download from the non-volatile configuration flash memory (CFM) block to the SRAM logic, maintaining tri-state of the I/O pins (with weak pull-up resistors enabled) before and during this process. When the MAX II device enters user mode, the POR circuit releases the I/O pins to user functionality. The POR circuit of the MAX II (except MAX IIZ) device continues to monitor the V_{CCINT} voltage level to detect a brown-out condition. The POR circuit of the MAX IIZ device does not monitor the V_{CCINT} voltage level after the device enters into user mode. More details are provided in the following sub-sections.

Recommended Operating Conditions

Table 5–2 shows the MAX II device family recommended operating conditions.

Table 5–2. MAX II Device Recommended Operating Conditions

Symbol	Parameter	Conditions	Minimum	Maximum	Unit
V_{CCINT} (1)	3.3-V supply voltage for internal logic and ISP	MAX II devices	3.00	3.60	V
	2.5-V supply voltage for internal logic and ISP	MAX II devices	2.375	2.625	V
	1.8-V supply voltage for internal logic and ISP	MAX IIG and MAX IIZ devices	1.71	1.89	V
V_{CCIO} (1)	Supply voltage for I/O buffers, 3.3-V operation	—	3.00	3.60	V
	Supply voltage for I/O buffers, 2.5-V operation	—	2.375	2.625	V
	Supply voltage for I/O buffers, 1.8-V operation	—	1.71	1.89	V
	Supply voltage for I/O buffers, 1.5-V operation	—	1.425	1.575	V
V_I	Input voltage	(2), (3), (4)	–0.5	4.0	V
V_O	Output voltage	—	0	V_{CCIO}	V
T_J	Operating junction temperature	Commercial range	0	85	°C
		Industrial range	–40	100	°C
		Extended range (5)	–40	125	°C

Notes to Table 5–2:

- (1) MAX II device in-system programming and/or user flash memory (UFM) programming via JTAG or logic array is not guaranteed outside the recommended operating conditions (for example, if brown-out occurs in the system during a potential write/program sequence to the UFM, users are recommended to read back UFM contents and verify against the intended write data).
- (2) Minimum DC input is –0.5 V. During transitions, the inputs may undershoot to –2.0 V for input currents less than 100 mA and periods shorter than 20 ns.
- (3) During transitions, the inputs may overshoot to the voltages shown in the following table based upon input duty cycle. The DC case is equivalent to 100% duty cycle. For more information about 5.0-V tolerance, refer to the *Using MAX II Devices in Multi-Voltage Systems* chapter in the *MAX II Device Handbook*.

V_{IN}	Max. Duty Cycle
4.0 V	100% (DC)
4.1	90%
4.2	50%
4.3	30%
4.4	17%
4.5	10%
- (4) All pins, including clock, I/O, and JTAG pins, may be driven before V_{CCINT} and V_{CCIO} are powered.
- (5) For the extended temperature range of 100 to 125° C, MAX II UFM programming (erase/write) is only supported via the JTAG interface. UFM programming via the logic array interface is not guaranteed in this range.

Power-Up Timing

Table 5–12 shows the power-up timing characteristics for MAX II devices.

Table 5–12. MAX II Power-Up Timing

Symbol	Parameter	Device	Min	Typ	Max	Unit
t_{CONFIG} (1)	The amount of time from when minimum V_{CCINT} is reached until the device enters user mode (2)	EPM240	—	—	200	μs
		EPM570	—	—	300	μs
		EPM1270	—	—	300	μs
		EPM2210	—	—	450	μs

Notes to Table 5–12:

- (1) Table 5–12 values apply to commercial and industrial range devices. For extended temperature range devices, the t_{CONFIG} maximum values are as follows:
- | | |
|---------|-------------------|
| Device | Maximum |
| EPM240 | 300 μs |
| EPM570 | 400 μs |
| EPM1270 | 400 μs |
| EPM2210 | 500 μs |
- (2) For more information about POR trigger voltage, refer to the *Hot Socketing and Power-On Reset in MAX II Devices* chapter in the *MAX II Device Handbook*.

Power Consumption

Designers can use the Altera® PowerPlay Early Power Estimator and PowerPlay Power Analyzer to estimate the device power.



For more information about these power analysis tools, refer to the *Understanding and Evaluating Power in MAX II Devices* chapter in the *MAX II Device Handbook* and the *PowerPlay Power Analysis* chapter in volume 3 of the *Quartus II Handbook*.

Timing Model and Specifications

MAX II devices timing can be analyzed with the Altera Quartus® II software, a variety of popular industry-standard EDA simulators and timing analyzers, or with the timing model shown in Figure 5–2.

MAX II devices have predictable internal delays that enable the designer to determine the worst-case timing of any design. The software provides timing simulation, point-to-point delay prediction, and detailed timing analysis for device-wide performance evaluation.

Table 5-21. UFM Block Internal Timing Microparameters (Part 3 of 3)

Symbol	Parameter	MAX II / MAX IIG						MAX IIZ						Unit
		-3 Speed Grade		-4 Speed Grade		-5 Speed Grade		-6 Speed Grade		-7 Speed Grade		-8 Speed Grade		
		Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	
t _{OE}	Delay from data register clock to data register output	180	—	180	—	180	—	180	—	180	—	180	—	ns
t _{RA}	Maximum read access time	—	65	—	65	—	65	—	65	—	65	—	65	ns
t _{OSCS}	Maximum delay between the OSC_ENA rising edge to the erase/program signal rising edge	250	—	250	—	250	—	250	—	250	—	250	—	ns
t _{OSCH}	Minimum delay allowed from the erase/program signal going low to OSC_ENA signal going low	250	—	250	—	250	—	250	—	250	—	250	—	ns

Figure 5-3 through Figure 5-5 show the read, program, and erase waveforms for UFM block timing parameters shown in Table 5-21.

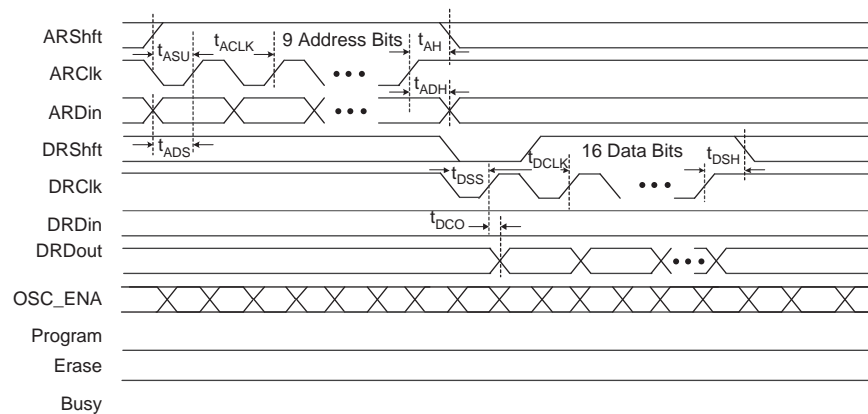
Figure 5-3. UFM Read Waveforms

Figure 5-4. UFM Program Waveforms

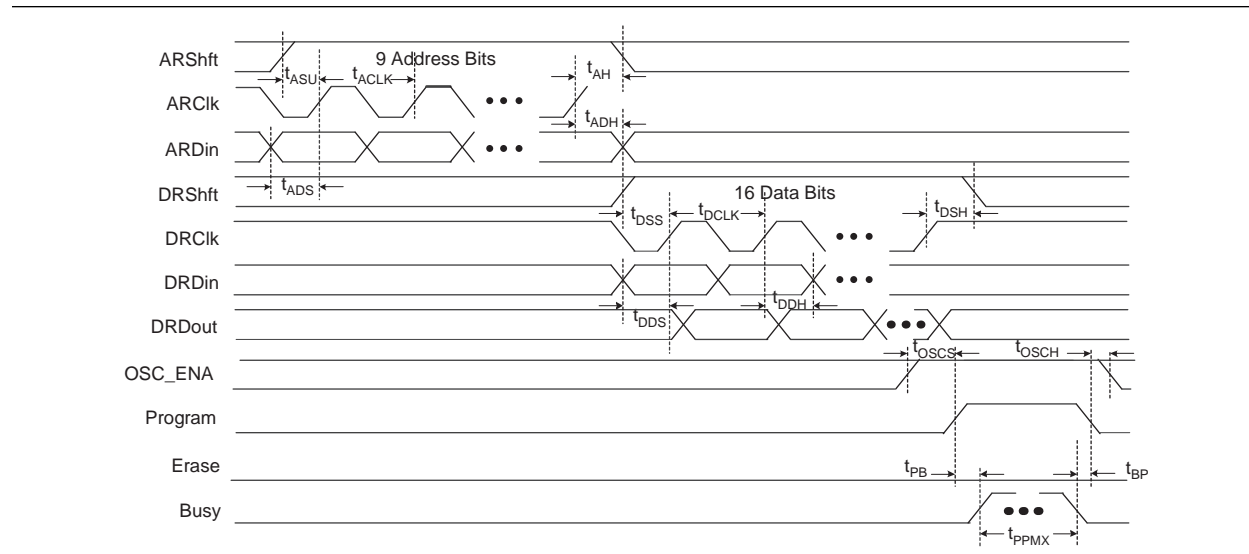


Figure 5-5. UFM Erase Waveform

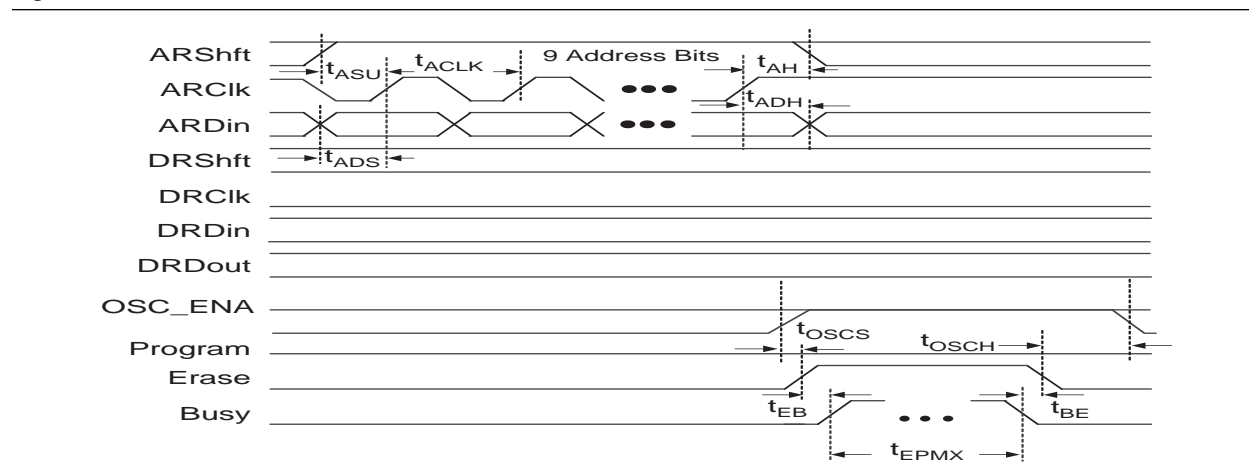


Table 5-22. Routing Delay Internal Timing Microparameters

Routing	MAX II / MAX IIG						MAX IIZ						Unit
	-3 Speed Grade		-4 Speed Grade		-5 Speed Grade		-6 Speed Grade		-7 Speed Grade		-8 Speed Grade		
	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	
t _{C4}	—	429	—	556	—	687	—	(1)	—	(1)	—	(1)	ps
t _{R4}	—	326	—	423	—	521	—	(1)	—	(1)	—	(1)	ps
t _{LOCAL}	—	330	—	429	—	529	—	(1)	—	(1)	—	(1)	ps

Note to Table 5-22:

(1) The numbers will only be available in a later revision.

Table 5-33. MAX II Maximum Output Clock Rate for I/O

I/O Standard		MAX II / MAX IIG			MAX IIZ		
		-3 Speed Grade	-4 Speed Grade	-5 Speed Grade	-6 Speed Grade	-7 Speed Grade	-8 Speed Grade
3.3-V LVTTTL	304	304	304	304	304	304	MHz
3.3-V LVCMOS	304	304	304	304	304	304	MHz
2.5-V LVTTTL	220	220	220	220	220	220	MHz
2.5-V LVCMOS	220	220	220	220	220	220	MHz
1.8-V LVTTTL	200	200	200	200	200	200	MHz
1.8-V LVCMOS	200	200	200	200	200	200	MHz
1.5-V LVCMOS	150	150	150	150	150	150	MHz
3.3-V PCI	304	304	304	304	304	304	MHz

JTAG Timing Specifications

Figure 5-6 shows the timing waveforms for the JTAG signals.

Figure 5-6. MAX II JTAG Timing Waveforms

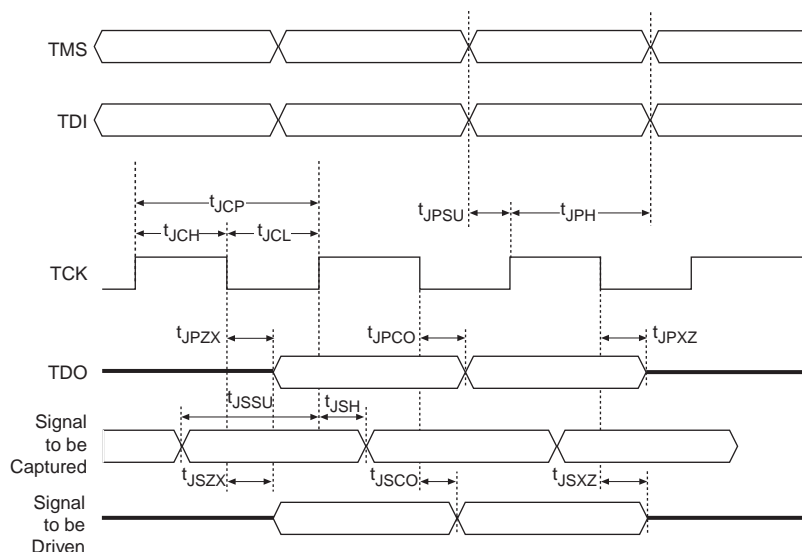


Table 5-34 shows the JTAG Timing parameters and values for MAX II devices.

Table 5-34. MAX II JTAG Timing Parameters (Part 1 of 2)

Symbol	Parameter	Min	Max	Unit
t_{JCP} (1)	TCK clock period for $V_{CCI01} = 3.3\text{ V}$	55.5	—	ns
	TCK clock period for $V_{CCI01} = 2.5\text{ V}$	62.5	—	ns
	TCK clock period for $V_{CCI01} = 1.8\text{ V}$	100	—	ns
	TCK clock period for $V_{CCI01} = 1.5\text{ V}$	143	—	ns
t_{JCH}	TCK clock high time	20	—	ns
t_{JCL}	TCK clock low time	20	—	ns

Document Revision History

Table 5-35 shows the revision history for this chapter.

Table 5-35. Document Revision History (Part 1 of 2)

Date and Revision	Changes Made	Summary of Changes
August 2009, version 2.5	<ul style="list-style-type: none"> Added Table 5-28, Table 5-29, and Table 5-30. Updated Table 5-2, Table 5-4, Table 5-14, Table 5-15, Table 5-16, Table 5-17, Table 5-18, Table 5-19, Table 5-20, Table 5-21, Table 5-22, Table 5-23, Table 5-24, Table 5-27, Table 5-31, Table 5-32, and Table 5-33. 	Added information for speed grade –8
November 2008, version 2.4	<ul style="list-style-type: none"> Updated Table 5-2. Updated “Internal Timing Parameters” section. 	—
October 2008, version 2.3	<ul style="list-style-type: none"> Updated New Document Format. Updated Figure 5-1. 	—
July 2008, version 2.2	<ul style="list-style-type: none"> Updated Table 5-14, Table 5-23, and Table 5-24. 	—
March 2008, version 2.1	<ul style="list-style-type: none"> Added (Note 5) to Table 5-4. 	—
December 2007, version 2.0	<ul style="list-style-type: none"> Updated (Note 3) and (4) to Table 5-1. Updated Table 5-2 and added (Note 5). Updated ICCSTANDBY and ICCPOWERUP information and added IPULLUP information in Table 5-4. Added (Note 1) to Table 5-10. Updated Figure 5-2. Added (Note 1) to Table 5-13. Updated Table 5-13 through Table 5-24, and Table 5-27 through Table 5-30. Added tCOMB information to Table 5-15. Updated Figure 5-6. Added “Referenced Documents” section. 	Updated document with MAX IIZ information.
December 2006, version 1.8	<ul style="list-style-type: none"> Added note to Table 5-1. Added document revision history. 	—
July 2006, version 1.7	<ul style="list-style-type: none"> Minor content and table updates. 	—
February 2006, version 1.6	<ul style="list-style-type: none"> Updated “External Timing I/O Delay Adders” section. Updated Table 5-29. Updated Table 5-30. 	—
November 2005, version 1.5	<ul style="list-style-type: none"> Updated Tables 5-2, 5-4, and 5-12. 	—
August 2005, version 1.4	<ul style="list-style-type: none"> Updated Figure 5-1. Updated Tables 5-13, 5-16, and 5-26. Removed Note 1 from Table 5-12. 	—

Table 5-35. Document Revision History (Part 2 of 2)

Date and Revision	Changes Made	Summary of Changes
June 2005, version 1.3	<ul style="list-style-type: none"> ■ Updated the R_{PULLUP} parameter in Table 5-4. ■ Added Note 2 to Tables 5-8 and 5-9. ■ Updated Table 5-13. ■ Added “Output Drive Characteristics” section. ■ Added I²C mode and Notes 5 and 6 to Table 5-14. ■ Updated timing values to Tables 5-14 through 5-33. 	—
December 2004, version 1.2	<ul style="list-style-type: none"> ■ Updated timing Tables 5-2, 5-4, 5-12, and Tables 15-14 through 5-34. ■ Table 5-31 is new. 	—
June 2004, version 1.1	<ul style="list-style-type: none"> ■ Updated timing Tables 5-15 through 5-32. 	—

Software

MAX® II devices are supported by the Altera® Quartus® II design software with new, optional MAX+PLUS® II look and feel, which provides HDL and schematic design entry, compilation and logic synthesis, full simulation and advanced timing analysis, and device programming. Refer to the Design Software Selector Guide for more details about the Quartus II software features.

The Quartus II software supports the Windows XP/2000/NT, Sun Solaris, Linux Red Hat v8.0, and HP-UX operating systems. It also supports seamless integration with industry-leading EDA tools through the NativeLink interface.

Device Pin-Outs

Printed device pin-outs for MAX II devices are available on the Altera website (www.altera.com).

Ordering Information

Figure 6–1 describes the ordering codes for MAX II devices. For more information about a specific package, refer to the *Package Information* chapter in the *MAX II Device Handbook*.

Figure 6–1. MAX II Device Packaging Ordering Information

