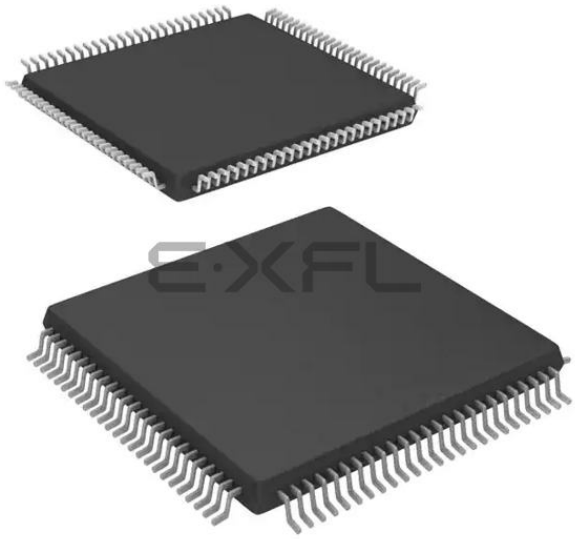

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Understanding Embedded - CPLDs (Complex Programmable Logic Devices)

Embedded - CPLDs, or Complex Programmable Logic Devices, are highly versatile digital logic devices used in electronic systems. These programmable components are designed to perform complex logical operations and can be customized for specific applications. Unlike fixed-function ICs, CPLDs offer the flexibility to reprogram their configuration, making them an ideal choice for various embedded systems. They consist of a set of logic gates and programmable interconnects, allowing designers to implement complex logic circuits without needing custom hardware.

Applications of Embedded - CPLDs

Details	
Product Status	Active
Programmable Type	In System Programmable
Delay Time tpd(1) Max	5.4 ns
Voltage Supply - Internal	1.71V ~ 1.89V
Number of Logic Elements/Blocks	570
Number of Macrocells	440
Number of Gates	-
Number of I/O	76
Operating Temperature	0°C ~ 85°C (TJ)
Mounting Type	Surface Mount
Package / Case	100-TQFP
Supplier Device Package	100-TQFP (14x14)
Purchase URL	https://www.e-xfl.com/product-detail/intel/epm570gt100c5n

1. Introduction

MII51001-1.9

Introduction

The MAX® II family of instant-on, non-volatile CPLDs is based on a 0.18-µm, 6-layer-metal-flash process, with densities from 240 to 2,210 logic elements (LEs) (128 to 2,210 equivalent macrocells) and non-volatile storage of 8 Kbits. MAX II devices offer high I/O counts, fast performance, and reliable fitting versus other CPLD architectures. Featuring MultiVolt core, a user flash memory (UFM) block, and enhanced in-system programmability (ISP), MAX II devices are designed to reduce cost and power while providing programmable solutions for applications such as bus bridging, I/O expansion, power-on reset (POR) and sequencing control, and device configuration control.

Features

The MAX II CPLD has the following features:

- Low-cost, low-power CPLD
- Instant-on, non-volatile architecture
- Standby current as low as 25 µA
- Provides fast propagation delay and clock-to-output times
- Provides four global clocks with two clocks available per logic array block (LAB)
- UFM block up to 8 Kbits for non-volatile storage
- MultiVolt core enabling external supply voltages to the device of either 3.3 V/2.5 V or 1.8 V
- MultiVolt I/O interface supporting 3.3-V, 2.5-V, 1.8-V, and 1.5-V logic levels
- Bus-friendly architecture including programmable slew rate, drive strength, bus-hold, and programmable pull-up resistors
- Schmitt triggers enabling noise tolerant inputs (programmable per pin)
- I/Os are fully compliant with the Peripheral Component Interconnect Special Interest Group (PCI SIG) PCI Local Bus Specification, Revision 2.2 for 3.3-V operation at 66 MHz
- Supports hot-socketing
- Built-in Joint Test Action Group (JTAG) boundary-scan test (BST) circuitry compliant with IEEE Std. 1149.1-1990
- ISP circuitry compliant with IEEE Std. 1532

MAX II devices have an internal linear voltage regulator which supports external supply voltages of 3.3 V or 2.5 V, regulating the supply down to the internal operating voltage of 1.8 V. MAX IIG and MAX IIZ devices only accept 1.8 V as the external supply voltage. MAX IIZ devices are pin-compatible with MAX IIG devices in the 100-pin Micro FineLine BGA and 256-pin Micro FineLine BGA packages. Except for external supply voltage requirements, MAX II and MAX II G devices have identical pin-outs and timing specifications. Table 1–5 shows the external supply voltages supported by the MAX II family.

Table 1–5. MAX II External Supply Voltages

Devices	EPM240 EPM570 EPM1270 EPM2210	EPM240G EPM570G EPM1270G EPM2210G EPM240Z EPM570Z *(1)*
MultiVolt core external supply voltage (V_{CCINT}) *(2)*	3.3 V, 2.5 V	1.8 V
MultiVolt I/O interface voltage levels (V_{CCIO})	1.5 V, 1.8 V, 2.5 V, 3.3 V	1.5 V, 1.8 V, 2.5 V, 3.3 V

Notes to Table 1–5:

(1) MAX IIG and MAX IIZ devices only accept 1.8 V on their `VCCINT` pins. The 1.8-V V_{CCINT} external supply powers the device core directly.

(2) MAX II devices operate internally at 1.8 V.

Referenced Documents

This chapter references the following documents:

- *DC and Switching Characteristics* chapter in the *MAX II Device Handbook*
- *MAX II Logic Element to Macrocell Conversion Methodology* white paper

Document Revision History

Table 1–6 shows the revision history for this chapter.

Table 1–6. Document Revision History

Date and Revision	Changes Made	Summary of Changes
August 2009, version 1.9	■ Updated Table 1–2.	Added information for speed grade –8
October 2008, version 1.8	■ Updated "Introduction" section. ■ Updated new Document Format.	—
December 2007, version1.7	■ Updated Table 1–1 through Table 1–5. ■ Added "Referenced Documents" section.	Updated document with MAX IIZ information.
December 2006, version 1.6	■ Added document revision history.	—
August 2006, version 1.5	■ Minor update to features list.	—
July 2006, version 1.4	■ Minor updates to tables.	—

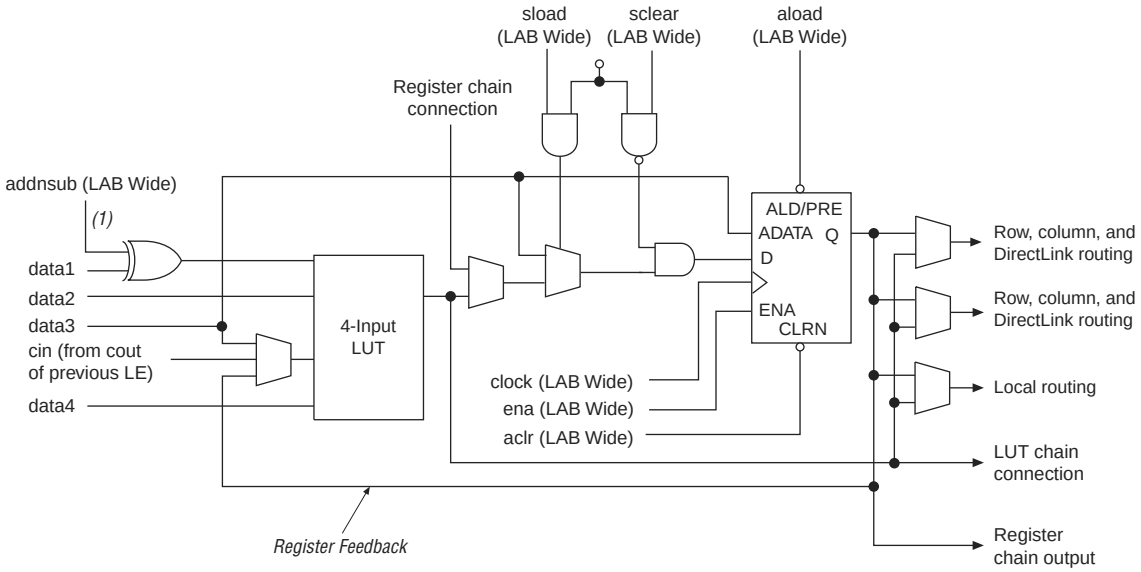
Table 1–6. Document Revision History

Date and Revision	Changes Made	Summary of Changes
June 2005, version 1.3	■ Updated timing numbers in Table 1-1.	—
December 2004, version 1.2	■ Updated timing numbers in Table 1-1.	—
June 2004, version 1.1	■ Updated timing numbers in Table 1-1.	—

Normal Mode

The normal mode is suitable for general logic applications and combinational functions. In normal mode, four data inputs from the LAB local interconnect are inputs to a four-input LUT (see Figure 2–7). The Quartus II Compiler automatically selects the carry-in or the data3 signal as one of the inputs to the LUT. Each LE can use LUT chain connections to drive its combinational output directly to the next LE in the LAB. Asynchronous load data for the register comes from the data3 input of the LE. LEs in normal mode support packed registers.

Figure 2–7. LE in Normal Mode



Note to Figure 2–7:

(1) This signal is only allowed in normal mode if the LE is at the end of an adder/subtractor chain.

Dynamic Arithmetic Mode

The dynamic arithmetic mode is ideal for implementing adders, counters, accumulators, wide parity functions, and comparators. An LE in dynamic arithmetic mode uses four 2-input LUTs configurable as a dynamic adder/subtractor. The first two 2-input LUTs compute two summations based on a possible carry-in of 1 or 0; the other two LUTs generate carry outputs for the two chains of the carry-select circuitry. As shown in Figure 2–8, the LAB carry-in signal selects either the `carry-in0` or carry-in1 chain. The selected chain's logic level in turn determines which parallel sum is generated as a combinational or registered output. For example, when implementing an adder, the sum output is the selection of two possible calculated sums:

```
data1 + data2 + carry in0
```

or

```
data1 + data2 + carry-in1
```

The speed advantage of the carry-select chain is in the parallel precomputation of carry chains. Since the LAB carry-in selects the precomputed carry chain, not every LE is in the critical path. Only the propagation delays between LAB carry-in generation (LE 5 and LE 10) are now part of the critical path. This feature allows the MAX II architecture to implement high-speed counters, adders, multipliers, parity functions, and comparators of arbitrary width.

Figure 2–9 shows the carry-select circuitry in an LAB for a 10-bit full adder. One portion of the LUT generates the sum of two bits using the input signals and the appropriate carry-in bit; the sum is routed to the output of the LE. The register can be bypassed for simple adders or used for accumulator functions. Another portion of the LUT generates carry-out bits. An LAB-wide carry-in bit selects which chain is used for the addition of given inputs. The carry-in signal for each chain, `carry-in0` or `carry-in1`, selects the carry-out to carry forward to the carry-in signal of the next-higher-order bit. The final carry-out signal is routed to an LE, where it is fed to local, row, or column interconnects.

Figure 2–9. Carry-Select Chain

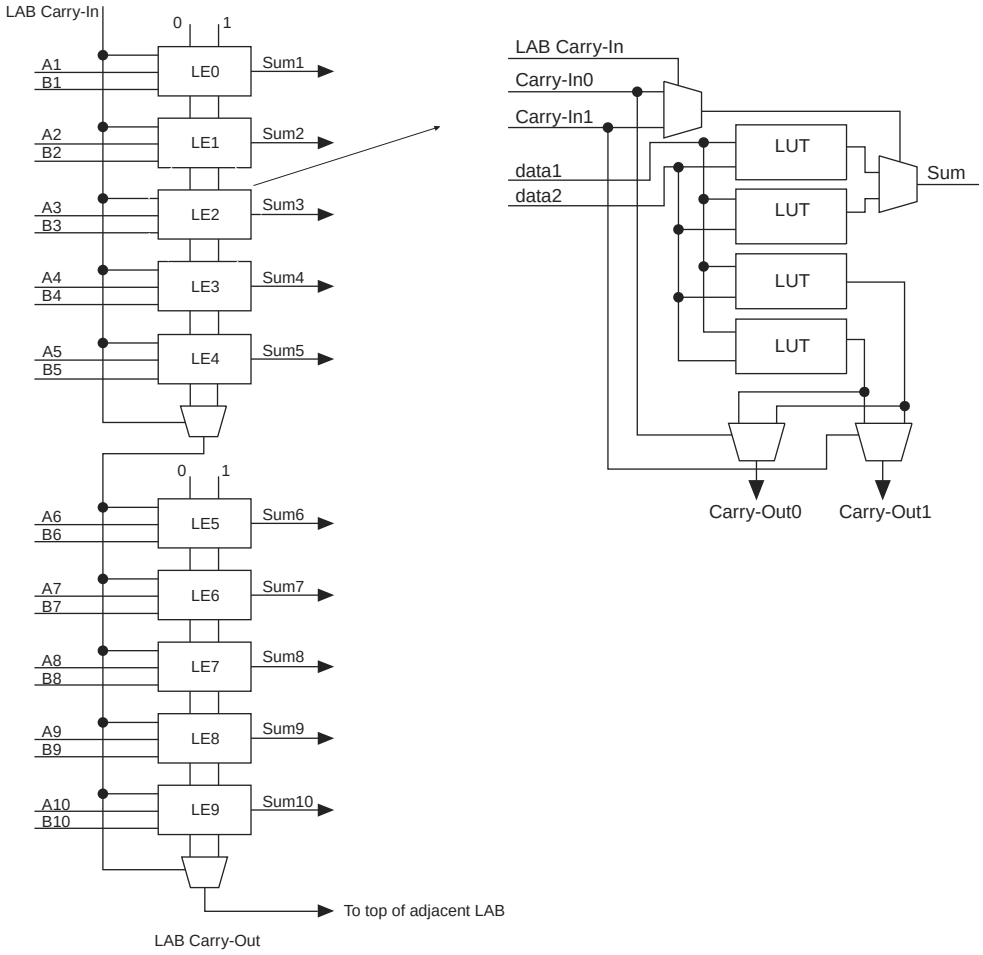
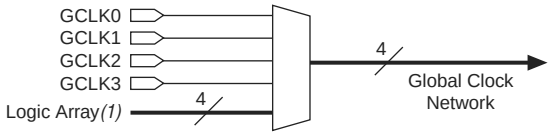


Figure 2–13. Global Clock Generation



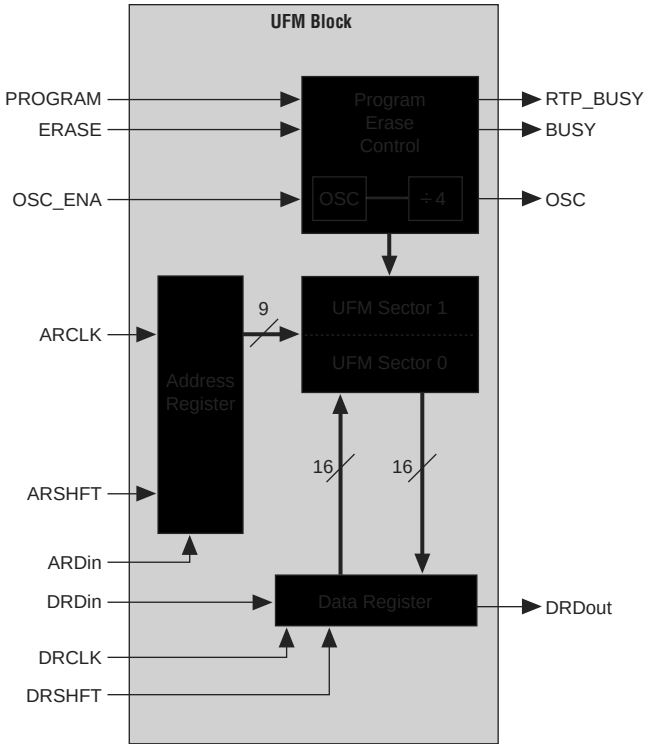
Note to Figure 2–13:

(1) Any I/O pin can use a MultiTrack interconnect to route as a logic array-generated global clock signal.

The global clock network drives to individual LAB column signals, LAB column clocks [3..0], that span an entire LAB column from the top to the bottom of the device. Unused global clocks or control signals in a LAB column are turned off at the LAB column clock buffers shown in Figure 2–14. The LAB column clocks [3..0] are multiplexed down to two LAB clock signals and one LAB clear signal. Other control signal types route from the global clock network into the LAB local interconnect. See "LAB Control Signals" on page 2–5 for more information.

- Auto-increment addressing
- Serial interface to logic array with programmable interface

Figure 2–15. UFM Block and Interface Signals



UFM Storage

Each device stores up to 8,192 bits of data in the UFM block. Table 2–3 shows the data size, sector, and address sizes for the UFM block.

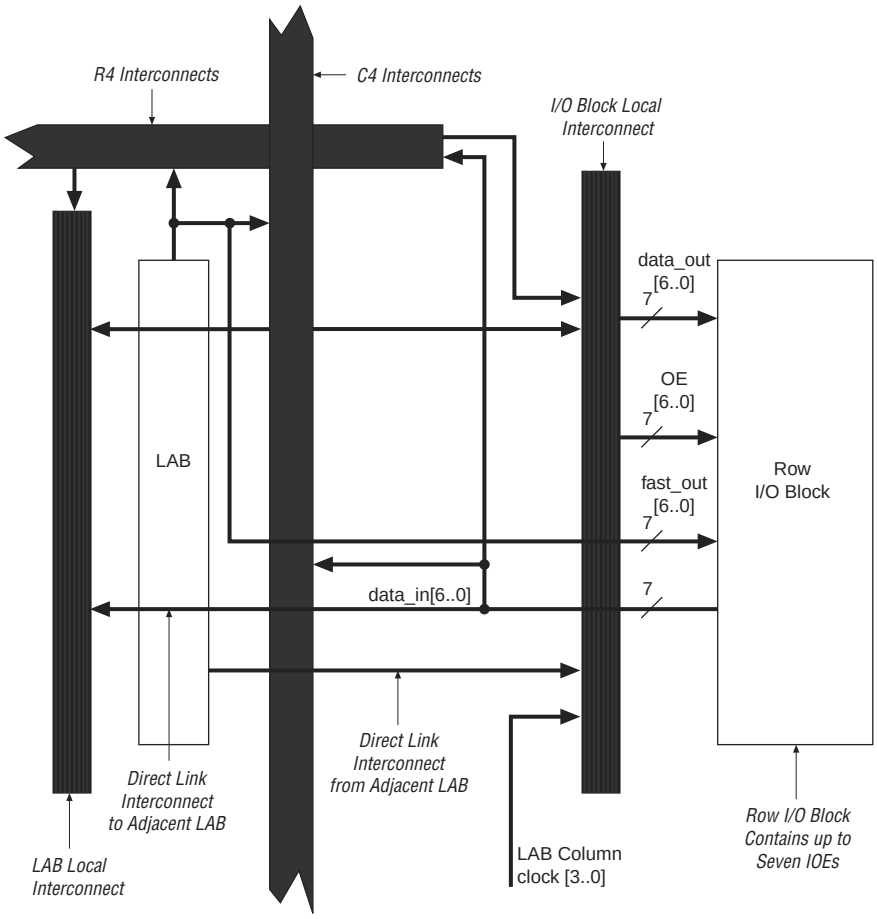
Table 2–3. UFM Array Size

Device	Total Bits	Sectors	Address Bits	Data Width
EPM240 EPM570 EPM1270 EPM2210	8,192	2 (4,096 bits/sector)	9	16

There are 512 locations with 9-bit addressing ranging from `000h` to `1FFh`. Sector 0 address space is `000h` to `0FFh` and Sector 1 address space is from `100h` to `1FFh`. The data width is up to 16 bits of data. The Quartus II software automatically creates logic to accommodate smaller read or program data widths. Erasure of the UFM involves individual sector erasing (that is, one erase of sector 0 and one erase of sector 1 is required to erase the entire UFM block). Since sector erase is required before a program or write, having two sectors enables a sector size of data to be left untouched while the other sector is erased and programmed with new data.

Figure 2–20 shows how a row I/O block connects to the logic array.

Figure 2–20. Row I/O Block Connection to the Interconnect *(Note 1)*



Note to Figure 2–20:

(1) Each of the seven IOEs in the row I/O block can have one `data_out` or `fast_out` output, one `OE` output, and one `data_in` input.

Connect VCCIO pins to either a 1.5-V, 1.8 V, 2.5-V, or 3.3-V power supply, depending on the output requirements. The output levels are compatible with systems of the same voltage as the power supply (that is, when VCCIO pins are connected to a 1.5-V power supply, the output levels are compatible with 1.5-V systems). When VCCIO pins are connected to a 3.3-V power supply, the output high is 3.3 V and is compatible with 3.3-V or 5.0-V systems. Table 2–7 summarizes MAX II MultiVolt I/O support.

Table 2–7. MAX II MultiVolt I/O Support *(Note 1)*

	Input Signal					Output Signal				
VCCIO (V)	**1.5 V**	**1.8 V**	**2.5 V**	**3.3 V**	**5.0 V**	**1.5 V**	**1.8 V**	**2.5 V**	**3.3 V**	**5.0 V**
1.5	✓	✓	✓	✓	—	✓	—	—	—	—
1.8	✓	✓	✓	✓	—	✓ *(2)*	✓	—	—	—
2.5	—	—	✓	✓	—	✓ *(3)*	✓ *(3)*	✓	—	—
3.3	—	—	✓ *(4)*	✓	✓ *(5)*	✓ *(6)*	✓ *(6)*	✓ *(6)*	✓	✓ *(7)*

Notes to Table 2–7:

(1) To drive inputs higher than V_{CCIO} but less than 4.0 V including the overshoot, disable the I/O clamp diode. However, to drive 5.0-V inputs to the device, enable the I/O clamp diode to prevent V_I from rising above 4.0 V.

(2) When V_{CCIO} = 1.8 V, a MAX II device can drive a 1.5-V device with 1.8-V tolerant inputs.

(3) When V_{CCIO} = 2.5 V, a MAX II device can drive a 1.5-V or 1.8-V device with 2.5-V tolerant inputs.

(4) When V_{CCIO} = 3.3 V and a 2.5-V input signal feeds an input pin, the VCCIO supply current will be slightly larger than expected.

(5) MAX II devices can be 5.0-V tolerant with the use of an external resistor and the internal I/O clamp diode on the EPM1270 and EPM2210 devices.

(6) When V_{CCIO} = 3.3 V, a MAX II device can drive a 1.5-V, 1.8-V, or 2.5-V device with 3.3-V tolerant inputs.

(7) When V_{CCIO} = 3.3 V, a MAX II device can drive a device with 5.0-V TTL inputs but not 5.0-V CMOS inputs. In the case of 5.0-V CMOS, open-drain setting with internal I/O clamp diode (available only on EPM1270 and EPM2210 devices) and external resistor is required.

For information about output pin source and sink current guidelines, refer to the *AN 428: MAX II CPLD Design Guidelines*.

Referenced Documents

This chapter referenced the following documents:

- *AN 428: MAX II CPLD Design Guidelines*
- *DC and Switching Characteristics* chapter in the *MAX II Device Handbook*
- *Hot Socketing and Power-On Reset in MAX II Devices* chapter in the *MAX II Device Handbook*
- *Using User Flash Memory in MAX II Devices* chapter in the *MAX II Device Handbook*

Table 3–1. MAX II JTAG Instructions (Part 2 of 2)

JTAG Instruction	Instruction Code	Description
CLAMP *(1)*	00 0000 1010	Places the 1-bit bypass register between the TDI and TDO pins, which allows the boundary scan test data to pass synchronously through selected devices to adjacent devices during normal device operation, while holding I/O pins to a state defined by the data in the boundary-scan register.
USER0	00 0000 1100	This instruction allows you to define the scan chain between TDI and TDO in the MAX II logic array. This instruction is also used for custom logic and JTAG interfaces.
USER1	00 0000 1110	This instruction allows you to define the scan chain between TDI and TDO in the MAX II logic array. This instruction is also used for custom logic and JTAG interfaces.
IEEE 1532 instructions	*(2)*	IEEE 1532 ISC instructions used when programming a MAX II device via the JTAG port.

Notes to Table 3–1:

(1) HIGHZ, CLAMP, and EXTEST instructions do not disable weak pull-up resistors or bus hold features.

(2) These instructions are shown in the 1532 BSDL files, which will be posted on the Altera® website at www.altera.com when they are available.

Unsupported JTAG instructions should not be issued to the MAX II device as this may put the device into an unknown state, requiring a power cycle to recover device operation.

The MAX II device instruction register length is 10 bits and the USERCODE register length is 32 bits. Table 3–2 and Table 3–3 show the boundary-scan register length and device IDCODE information for MAX II devices.

Table 3–2. MAX II Boundary-Scan Register Length

Device	Boundary-Scan Register Length
EPM240	240
EPM570	480
EPM1270	636
EPM2210	816

Table 3–3. 32-Bit MAX II Device IDCODE (Part 1 of 2)

Device	Binary IDCODE (32 Bits) *(1)*				HEX IDCODE
	Version (4 Bits)	Part Number	Manufacturer Identity (11 Bits)	LSB (1 Bit) *(2)*	
EPM240 EPM240G	0000	0010 0000 1010 0001	000 0110 1110	1	0x020A10DD
EPM570 EPM570G	0000	0010 0000 1010 0010	000 0110 1110	1	0x020A20DD
EPM1270 EPM1270G	0000	0010 0000 1010 0011	000 0110 1110	1	0x020A30DD
EPM2210 EPM2210G	0000	0010 0000 1010 0100	000 0110 1110	1	0x020A40DD

IEEE 1532 Support

The JTAG circuitry and ISP instruction set in MAX II devices is compliant to the IEEE 1532-2002 programming specification. This provides industry-standard hardware and software for in-system programming among multiple vendor programmable logic devices (PLDs) in a JTAG chain.

The MAX II 1532 BSDL files will be released on the Altera website when available.

Jam Standard Test and Programming Language (STAPL)

The Jam STAPL JEDEC standard, JESD71, can be used to program MAX II devices with in-circuit testers, PCs, or embedded processors. The Jam byte code is also supported for MAX II devices. These software programming protocols provide a compact embedded solution for programming MAX II devices.

For more information, refer to the *Using Jam STAPL for ISP via an Embedded Processor* chapter in the *MAX II Device Handbook*.

Programming Sequence

During in-system programming, 1532 instructions, addresses, and data are shifted into the MAX II device through the `TDI` input pin. Data is shifted out through the `TDO` output pin and compared against the expected data. Programming a pattern into the device requires the following six ISP steps. A stand-alone verification of a programmed pattern involves only stages 1, 2, 5, and 6. These steps are automatically executed by third-party programmers, the Quartus II software, or the Jam STAPL and Jam Byte-Code Players.

1. *Enter ISP*—The enter ISP stage ensures that the I/O pins transition smoothly from user mode to ISP mode.
2. *Check ID*—Before any program or verify process, the silicon ID is checked. The time required to read this silicon ID is relatively small compared to the overall programming time.
3. *Sector Erase*—Erasing the device in-system involves shifting in the instruction to erase the device and applying an erase pulse(s). The erase pulse is automatically generated internally by waiting in the run/test/idle state for the specified erase pulse time of 500 ms for the CFM block and 500 ms for each sector of the UFM block.
4. *Program*—Programming the device in-system involves shifting in the address, data, and program instruction and generating the program pulse to program the flash cells. The program pulse is automatically generated internally by waiting in the run/test/idle state for the specified program pulse time of 75 µs. This process is repeated for each address in the CFM and UFM blocks.
5. *Verify*—Verifying a MAX II device in-system involves shifting in addresses, applying the verify instruction to generate the read pulse, and shifting out the data for comparison. This process is repeated for each CFM and UFM address.
6. *Exit ISP*—An exit ISP stage ensures that the I/O pins transition smoothly from ISP mode to user mode.

Table 3–4 shows the programming times for MAX II devices using in-circuit testers to execute the algorithm vectors in hardware. Software-based programming tools used with download cables are slightly slower because of data processing and transfer limitations.

Table 3–4. MAX II Device Family Programming Times

Description	EPM240 EPM240G EPM240Z	EPM570 EPM570G EPM570Z	EPM1270 EPM1270G	EPM2210 EPM2210G	Unit
Erase + Program (1 MHz)	1.72	2.16	2.90	3.92	sec
Erase + Program (10 MHz)	1.65	1.99	2.58	3.40	sec
Verify (1 MHz)	0.09	0.17	0.30	0.49	sec
Verify (10 MHz)	0.01	0.02	0.03	0.05	sec
Complete Program Cycle (1 MHz)	1.81	2.33	3.20	4.41	sec
Complete Program Cycle (10 MHz)	1.66	2.01	2.61	3.45	sec

UFM Programming

The Quartus II software, with the use of POF, Jam, or JBC files, supports programming of the user flash memory (UFM) block independent of the logic array design pattern stored in the CFM block. This allows updating or reading UFM contents through ISP without altering the current logic array design, or vice versa. By default, these programming files and methods will program the entire flash memory contents, which includes the CFM block and UFM contents. The stand-alone embedded Jam STAPL player and Jam Byte-Code Player provides action commands for programming or reading the entire flash memory (UFM and CFM together) or each independently.

For more information, refer to the *Using Jam STAPL for ISP via an Embedded Processor* chapter in the *MAX II Device Handbook*.

In-System Programming Clamp

By default, the IEEE 1532 instruction used for entering ISP automatically tri-states all I/O pins with weak pull-up resistors for the duration of the ISP sequence. However, some systems may require certain pins on MAX II devices to maintain a specific DC logic level during an in-field update. For these systems, an optional in-system programming clamp instruction exists in MAX II circuitry to control I/O behavior during the ISP sequence. The in-system programming clamp instruction enables the device to sample and sustain the value on an output pin (an input pin would remain tri-stated if sampled) or to explicitly set a logic high, logic low, or tri-state value on any pin. Setting these options is controlled on an individual pin basis using the Quartus II software.

For more information, refer to the *Real-Time ISP and ISP Clamp for MAX II Devices* chapter in the *MAX II Device Handbook*.

4. Hot Socketing and Power-On Reset in MAX II Devices

MII51004-2.1

Introduction

MAX® II devices offer hot socketing, also known as hot plug-in or hot swap, and power sequencing support. Designers can insert or remove a MAX II board in a system during operation without undesirable effects to the system bus. The hot socketing feature removes some of the difficulties designers face when using components on printed circuit boards (PCBs) that contain a mixture of 3.3-, 2.5-, 1.8-, and 1.5-V devices.

The MAX II device hot socketing feature provides:

- Board or device insertion and removal
- Support for any power-up sequence
- Non-intrusive I/O buffers to system buses during hot insertion

This chapter contains the following sections:

MAX II Hot-Socketing Specifications

MAX II devices offer all three of the features required for the hot-socketing capability listed above without any external components or special design requirements. The following are hot-socketing specifications:

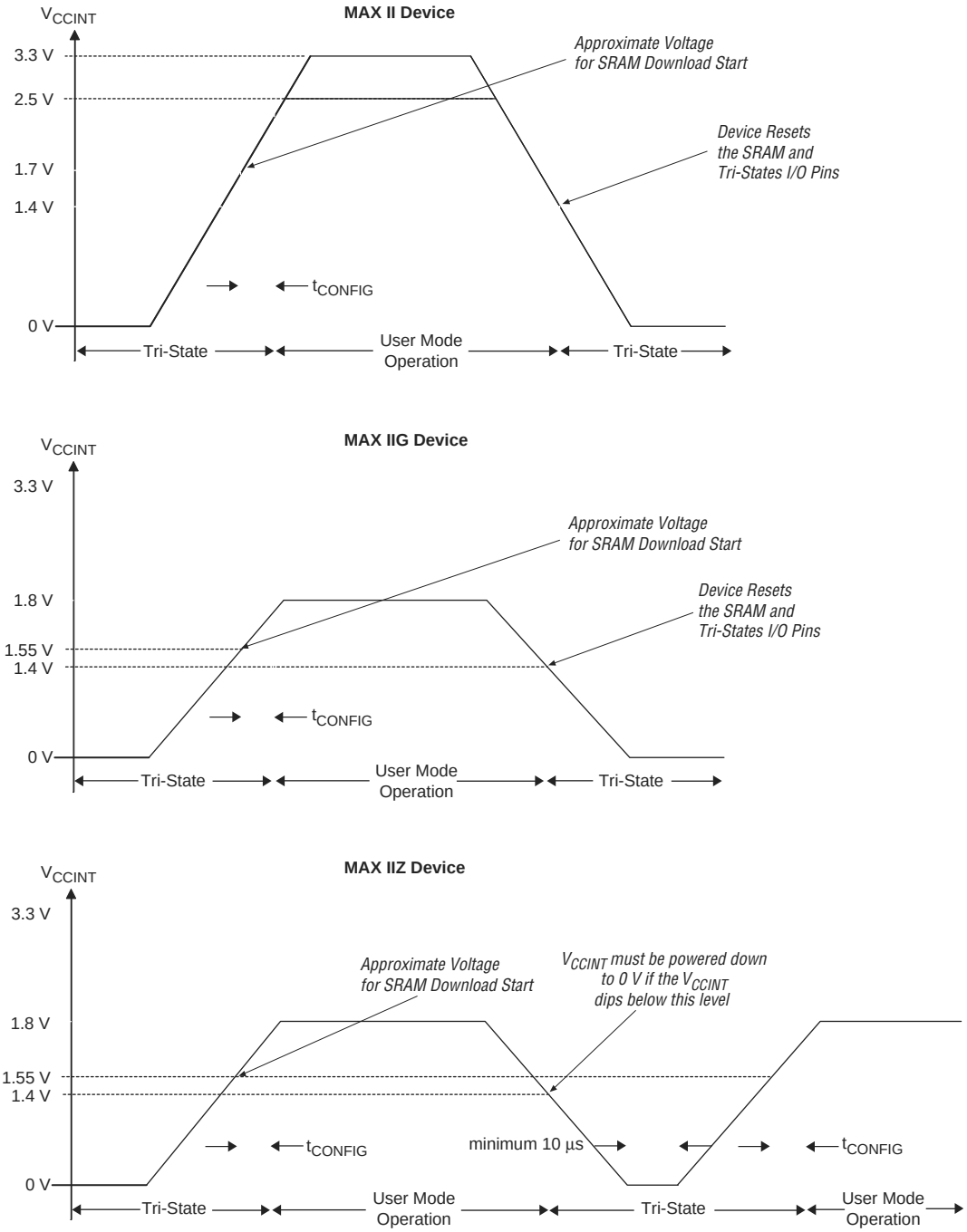
- The device can be driven before and during power-up or power-down without any damage to the device itself.
- I/O pins remain tri-stated during power-up. The device does not drive out before or during power-up, thereby affecting other buses in operation.
- Signal pins do not drive the V_{CCIO} or V_{CCINT} power supplies. External input signals to device I/O pins do not power the device V_{CCIO} or V_{CCINT} power supplies via internal paths. This is true if the V_{CCINT} and the V_{CCIO} supplies are held at GND.

☞ Altera uses `GND` as reference for the hot-socketing and I/O buffers circuitry designs. You must connect the `GND` between boards before connecting the V_{CCINT} and the V_{CCIO} power supplies to ensure device reliability and compliance to the hot-socketing specifications.

Devices Can Be Driven before Power-Up

Signals can be driven into the MAX II device I/O pins and `GCLK[3..0]` pins before or during power-up or power-down without damaging the device. MAX II devices support any power-up or power-down sequence (V_{CCIO1}, V_{CCIO2}, V_{CCIO3}, V_{CCIO4}, V_{CCINT}), simplifying the system-level design.

Figure 4–5. Power-Up Characteristics for MAX II, MAX IIG, and MAX IIZ Devices *(Note 1), (2)*



Notes to Figure 4–5:

(1) Time scale is relative.

(2) Figure 4–5 assumes all V_{CCIO} banks power up simultaneously with the V_{CCINT} profile shown. If not, t_{CONFIG} stretches out until all V_{CCIO} banks are powered.

☞ After SRAM configuration, all registers in the device are cleared and released into user function before I/O tri-states are released. To release clears after tri-states are released, use the DEV_CLRn pin option. To hold the tri-states beyond the power-up configuration time, use the DEV_OE pin option.

5. DC and Switching Characteristics

MII51005-2.5

Introduction

System designers must consider the recommended DC and switching conditions discussed in this chapter to maintain the highest possible performance and reliability of the MAX® II devices. This chapter contains the following sections:

Operating Conditions

Table 5–1 through Table 5–12 provide information about absolute maximum ratings, recommended operating conditions, DC electrical characteristics, and other specifications for MAX II devices.

Absolute Maximum Ratings

Table 5–1 shows the absolute maximum ratings for the MAX II device family.

Table 5–1. MAX II Device Absolute Maximum Ratings *(Note 1)*, *(2)*

Symbol	Parameter	Conditions	Minimum	Maximum	Unit
V_{CCINT}	Internal supply voltage *(3)*	With respect to ground	–0.5	4.6	V
V_{CCIO}	I/O supply voltage	—	–0.5	4.6	V
V_I	DC input voltage	—	–0.5	4.6	V
I_{OUT}	DC output current, per pin *(4)*	—	–25	25	mA
T_{STG}	Storage temperature	No bias	–65	150	°C
T_{AMB}	Ambient temperature	Under bias *(5)*	–65	135	°C
T_J	Junction temperature	TQFP and BGA packages under bias	—	135	°C

Notes to Table 5–1:

(1) Refer to the *Operating Requirements for Altera Devices Data Sheet*.

(2) Conditions beyond those listed in Table 5–1 may cause permanent damage to a device. Additionally, device operation at the absolute maximum ratings for extended periods of time may have adverse affects on the device.

(3) Maximum V_{CCINT} for MAX II devices is 4.6 V. For MAX IIG and MAX IIZ devices, it is 2.4 V.

(4) Refer to *AN 286: Implementing LED Drivers in MAX & MAX II Devices* for more information about the maximum source and sink current for MAX II devices.

(5) Refer to Table 5–2 for information about "under bias" conditions.

Table 5–13. MAX II Device Timing Model Status (Part 2 of 2)

Device	Preliminary	Final
EPM1270	—	✓
EPM2210	—	✓

Note to Table 5–13:

(1) The MAX IIZ device timing models are only available in the Quartus II software version 8.0 and later.

Performance

Table 5–14 shows the MAX II device performance for some common designs. All performance values were obtained with the Quartus II software compilation of megafunctions. Performance values for –3, –4, and –5 speed grades are based on an EPM1270 device target, while –6, –7, and –8 speed grades are based on an EPM570Z device target.

Table 5–14. MAX II Device Performance

Resource Used	Design Size and Function	Resources Used			Performance						Unit
					MAX II / MAX IIG			MAX IIZ			
		Mode	LEs	UFM Blocks	–3 Speed Grade	–4 Speed Grade	–5 Speed Grade	–6 Speed Grade	–7 Speed Grade	–8 Speed Grade	
LE	16-bit counter *(1)*	—	16	0	304.0	247.5	201.1	184.1	123.5	118.3	MHz
	64-bit counter *(1)*	—	64	0	201.5	154.8	125.8	83.2	83.2	80.5	MHz
	16-to-1 multiplexer	—	11	0	6.0	8.0	9.3	17.4	17.3	20.4	ns
	32-to-1 multiplexer	—	24	0	7.1	9.0	11.4	12.5	22.8	25.3	ns
	16-bit XOR function	—	5	0	5.1	6.6	8.2	9.0	15.0	16.1	ns
	16-bit decoder with single address line	—	5	0	5.2	6.6	8.2	9.2	15.0	16.1	ns
UFM	512 × 16	None	3	1	10.0	10.0	10.0	10.0	10.0	10.0	MHz
	512 × 16	SPI *(2)*	37	1	8.0	8.0	8.0	9.7	9.7	9.7	MHz
	512 × 8	Parallel *(3)*	73	1	*(4)*	*(4)*	*(4)*	*(4)*	*(4)*	*(4)*	MHz
	512 × 16	I^2C *(3)*	142	1	100 *(5)*	100 *(5)*	100 *(5)*	100 *(5)*	100 *(5)*	100 *(5)*	kHz

Notes to Table 5–14:

(1) This design is a binary loadable up counter.

(2) This design is configured for read-only operation in Extended mode. Read and write ability increases the number of LEs used.

(3) This design is configured for read-only operation. Read and write ability increases the number of LEs used.

(4) This design is asynchronous.

(5) The I^2C megafunction is verified in hardware up to 100-kHz serial clock line (SCL) rate.

Internal Timing Parameters

Internal timing parameters are specified on a speed grade basis independent of device density. Table 5–15 through Table 5–22 describe the MAX II device internal timing microparameters for logic elements (LEs), input/output elements (IOEs), UFM blocks, and MultiTrack interconnects. The timing values for –3, –4, and –5 speed grades shown in Table 5–15 through Table 5–22 are based on an EPM1270 device target, while –6, –7, and –8 speed grade values are based on an EPM570Z device target.

For more explanations and descriptions about each internal timing microparameters symbol, refer to the *Understanding Timing in MAX II Devices* chapter in the *MAX II Device Handbook*.

Table 5–15. LE Internal Timing Microparameters

		MAX II / MAX IIG						MAX IIZ						
		–3 Speed Grade		–4 Speed Grade		–5 Speed Grade		–6 Speed Grade		–7 Speed Grade		–8 Speed Grade		
Symbol	Parameter	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Unit
t_{LUT}	LE combinational LUT delay	—	571	—	742	—	914	—	1,215	—	2,247	—	2,247	ps
t_{COMB}	Combinational path delay	—	147	—	192	—	236	—	243	—	305	—	309	ps
t_{CLR}	LE register clear delay	238	—	309	—	381	—	401	—	541	—	545	—	ps
t_{PRE}	LE register preset delay	238	—	309	—	381	—	401	—	541	—	545	—	ps
t_{SU}	LE register setup time before clock	208	—	271	—	333	—	260	—	319	—	321	—	ps
t_{H}	LE register hold time after clock	0	—	0	—	0	—	0	—	0	—	0	—	ps
t_{CO}	LE register clock-to-output delay	—	235	—	305	—	376	—	380	—	489	—	494	ps
t_{CLKHL}	Minimum clock high or low time	166	—	216	—	266	—	253	—	335	—	339	—	ps
t_{C}	Register control delay	—	857	—	1,114	—	1,372	—	1,356	—	1,722	—	1,741	ps

Table 5–29. External Timing Output Delay and t_{OD} Adders for Fast Slew Rate

I/O Standard		MAX II / MAX IIG						MAX IIZ						Unit
		–3 Speed Grade		–4 Speed Grade		–5 Speed Grade		–6 Speed Grade		–7 Speed Grade		–8 Speed Grade		
		Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	
3.3-V LVTTL	16 mA	—	0	—	0	—	0	—	0	—	0	—	0	ps
	8 mA	—	65	—	84	—	104	—	–6	—	–2	—	–3	ps
3.3-V LVCMOS	8 mA	—	0	—	0	—	0	—	0	—	0	—	0	ps
	4 mA	—	65	—	84	—	104	—	–6	—	–2	—	–3	ps
2.5-V LVTTL / LVCMOS	14 mA	—	122	—	158	—	195	—	–63	—	–71	—	–88	ps
	7 mA	—	193	—	251	—	309	—	10	—	–1	—	1	ps
1.8-V LVTTL / LVCMOS	6 mA	—	568	—	738	—	909	—	128	—	118	—	118	ps
	3 mA	—	654	—	850	—	1,046	—	352	—	327	—	332	ps
1.5-V LVCMOS	4 mA	—	1,059	—	1,376	—	1,694	—	421	—	400	—	400	ps
	2 mA	—	1,167	—	1,517	—	1,867	—	757	—	743	—	743	ps
3.3-V PCI	20 mA	—	3	—	4	—	5	—	–6	—	–2	—	–3	ps

Table 5–30. External Timing Output Delay and t_{OD} Adders for Slow Slew Rate

I/O Standard		MAX II / MAX IIG						MAX IIZ						Unit
		–3 Speed Grade		–4 Speed Grade		–5 Speed Grade		–6 Speed Grade		–7 Speed Grade		–8 Speed Grade		
		Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	
3.3-V LVTTL	16 mA	—	7,064	—	6,745	—	6,426	—	5,966	—	5,992	—	6,118	ps
	8 mA	—	7,946	—	7,627	—	7,308	—	6,541	—	6,570	—	6,720	ps
3.3-V LVCMOS	8 mA	—	7,064	—	6,745	—	6,426	—	5,966	—	5,992	—	6,118	ps
	4 mA	—	7,946	—	7,627	—	7,308	—	6,541	—	6,570	—	6,720	ps
2.5-V LVTTL / LVCMOS	14 mA	—	10,434	—	10,115	—	9,796	—	9,141	—	9,154	—	9,297	ps
	7 mA	—	11,548	—	11,229	—	10,910	—	9,861	—	9,874	—	10,037	ps
1.8-V LVTTL / LVCMOS	6 mA	—	22,927	—	22,608	—	22,289	—	21,811	—	21,854	—	21,857	ps
	3 mA	—	24,731	—	24,412	—	24,093	—	23,081	—	23,034	—	23,107	ps
1.5-V LVCMOS	4 mA	—	38,723	—	38,404	—	38,085	—	39,121	—	39,124	—	39,124	ps
	2 mA	—	41,330	—	41,011	—	40,692	—	40,631	—	40,634	—	40,634	ps
3.3-V PCI	20 mA	—	261	—	339	—	418	—	6,644	—	6,627	—	6,914	ps