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Understanding Embedded - CPLDs (Complex Programmable Logic Devices)

Embedded - CPLDs, or Complex Programmable Logic Devices, are highly versatile digital logic devices used in electronic systems. These programmable components are designed to perform complex logical operations and can be customized for specific applications. Unlike fixed-function ICs, CPLDs offer the flexibility to reprogram their configuration, making them an ideal choice for various embedded systems. They consist of a set of logic gates and programmable interconnects, allowing designers to implement complex logic circuits without needing custom hardware.

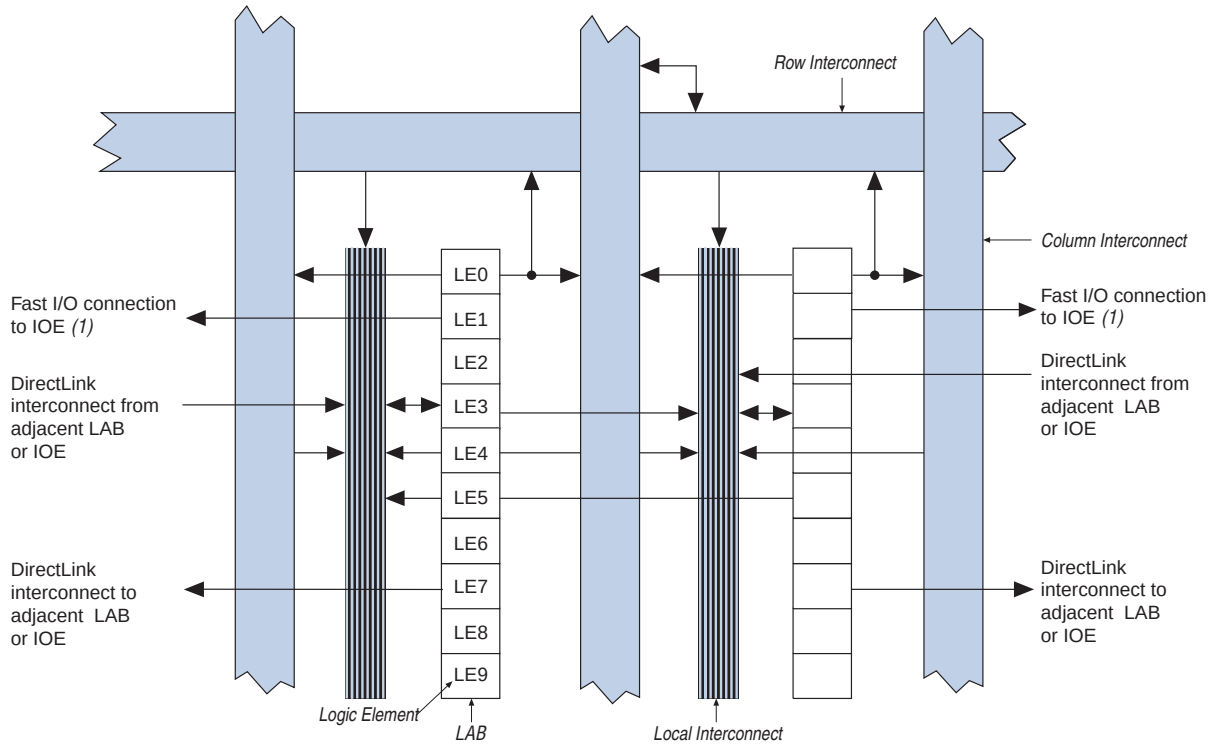
Applications of Embedded - CPLDs

Details	
Product Status	Active
Programmable Type	In System Programmable
Delay Time tpd(1) Max	5.4 ns
Voltage Supply - Internal	1.71V ~ 1.89V
Number of Logic Elements/Blocks	570
Number of Macrocells	440
Number of Gates	-
Number of I/O	116
Operating Temperature	0°C ~ 85°C (TJ)
Mounting Type	Surface Mount
Package / Case	144-LQFP
Supplier Device Package	144-TQFP (20x20)
Purchase URL	https://www.e-xfl.com/pro/item?MUrl=&PartUrl=epm570gt144c4n

Logic Array Blocks

Each LAB consists of 10 LEs, LE carry chains, LAB control signals, a local interconnect, a look-up table (LUT) chain, and register chain connection lines. There are 26 possible unique inputs into an LAB, with an additional 10 local feedback input lines fed by LE outputs in the same LAB. The local interconnect transfers signals between LEs in the same LAB. LUT chain connections transfer the output of one LE's LUT to the adjacent LE for fast sequential LUT connections within the same LAB. Register chain connections transfer the output of one LE's register to the adjacent LE's register within an LAB. The Quartus® II software places associated logic within an LAB or adjacent LABs, allowing the use of local, LUT chain, and register chain connections for performance and area efficiency. Figure 2–3 shows the MAX II LAB.

Figure 2–3. MAX II LAB Structure



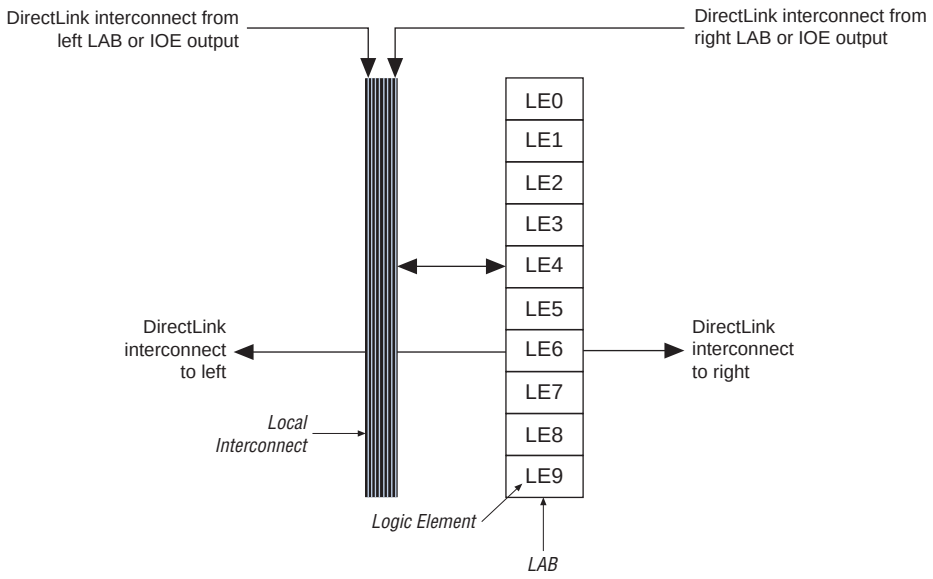
Note to Figure 2–3:

(1) Only from LABs adjacent to IOEs.

LAB Interconnects

The LAB local interconnect can drive LEs within the same LAB. The LAB local interconnect is driven by column and row interconnects and LE outputs within the same LAB. Neighboring LABs, from the left and right, can also drive an LAB's local interconnect through the DirectLink connection. The DirectLink connection feature minimizes the use of row and column interconnects, providing higher performance and flexibility. Each LE can drive 30 other LEs through fast local and DirectLink interconnects. Figure 2–4 shows the DirectLink connection.

Figure 2–4. DirectLink Connection



LAB Control Signals

Each LAB contains dedicated logic for driving control signals to its LEs. The control signals include two clocks, two clock enables, two asynchronous clears, a synchronous clear, an asynchronous preset/load, a synchronous load, and add/subtract control signals, providing a maximum of 10 control signals at a time. Although synchronous load and clear signals are generally used when implementing counters, they can also be used with other functions.

Each LAB can use two clocks and two clock enable signals. Each LAB's clock and clock enable signals are linked. For example, any LE in a particular LAB using the `labclk1` signal also uses `labclkena1`. If the LAB uses both the rising and falling edges of a clock, it also uses both LAB-wide clock signals. Deasserting the clock enable signal turns off the LAB-wide clock.

Each LAB can use two asynchronous clear signals and an asynchronous load/preset signal. By default, the Quartus II software uses a `NOT` gate push-back technique to achieve preset. If you disable the `NOT` gate push-back option or assign a given register to power-up high using the Quartus II software, the preset is then achieved using the asynchronous load signal with asynchronous load data input tied high.

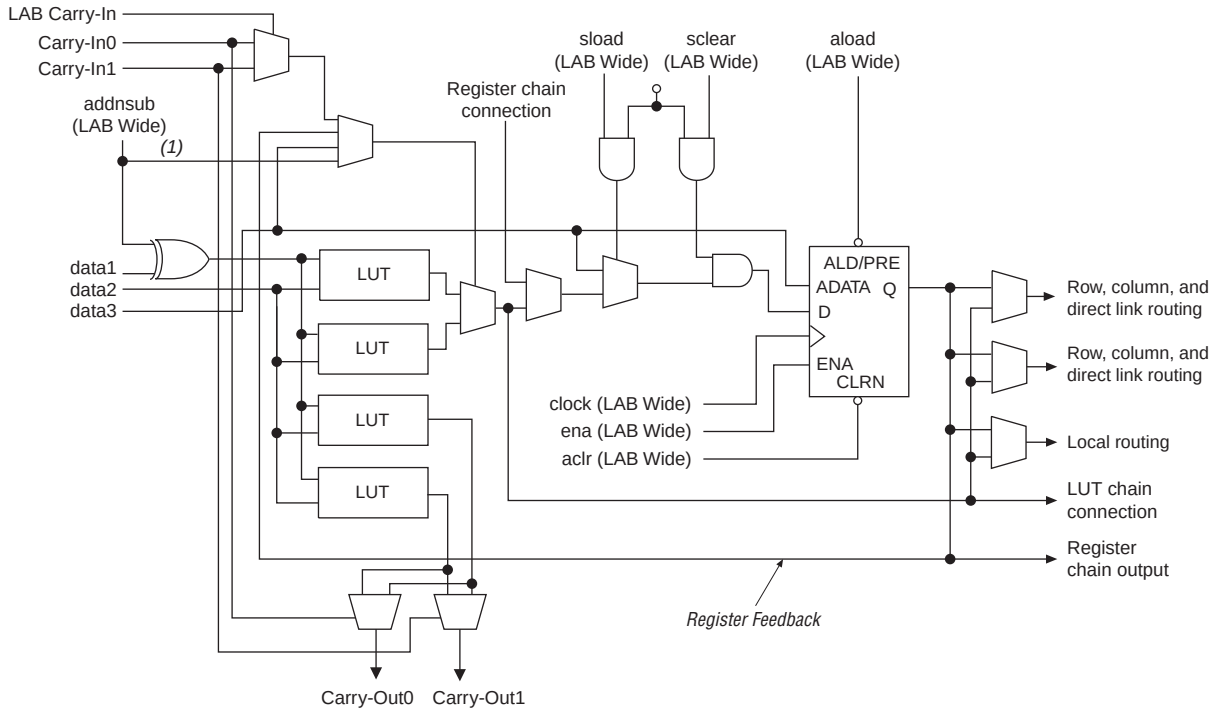
With the LAB-wide addnsub control signal, a single LE can implement a one-bit adder and subtractor. This saves LE resources and improves performance for logic functions such as correlators and signed multipliers that alternate between addition and subtraction depending on data.

The LAB column clocks [3..0], driven by the global clock network, and LAB local interconnect generate the LAB-wide control signals. The MultiTrack interconnect structure drives the LAB local interconnect for non-global control signal generation. The MultiTrack interconnect's inherent low skew allows clock and control signal distribution in addition to data. Figure 2–5 shows the LAB control signal generation circuit.

The other two LUTs use the data1 and data2 signals to generate two possible carry-out signals: one for a carry of 1 and the other for a carry of 0. The `carry-in0` signal acts as the carry-select for the `carry-out0` output and `carry-in1` acts as the carry-select for the `carry-out1` output. LEs in arithmetic mode can drive out registered and unregistered versions of the LUT output.

The dynamic arithmetic mode also offers clock enable, counter enable, synchronous up/down control, synchronous clear, synchronous load, and dynamic adder/subtractor options. The LAB local interconnect data inputs generate the counter enable and synchronous up/down control signals. The synchronous clear and synchronous load options are LAB-wide signals that affect all registers in the LAB. The Quartus II software automatically places any registers that are not used by the counter into other LABs. The addnsub LAB-wide signal controls whether the LE acts as an adder or subtractor.

Figure 2–8. LE in Dynamic Arithmetic Mode



Note to Figure 2–8:

(1) The addnsub signal is tied to the carry input for the first LE of a carry chain only.

Carry-Select Chain

The carry-select chain provides a very fast carry-select function between LEs in dynamic arithmetic mode. The carry-select chain uses the redundant carry calculation to increase the speed of carry functions. The LE is configured to calculate outputs for a possible carry-in of 0 and carry-in of 1 in parallel. The `carry-in0` and `carry-in1` signals from a lower-order bit feed forward into the higher-order bit via the parallel carry chain and feed into both the LUT and the next portion of the carry chain. Carry-select chains can begin in any LE within an LAB.

The UFM block communicates with the logic array similar to LAB-to-LAB interfaces. The UFM block connects to row and column interconnects and has local interconnect regions driven by row and column interconnects. This block also has DirectLink interconnects for fast connections to and from a neighboring LAB. For more information about the UFM interface to the logic array, see "User Flash Memory Block" on page 2–18.

Table 2–2 shows the MAX II device routing scheme.

Table 2–2. MAX II Device Routing Scheme

Source	Destination										
	LUT Chain	**Register Chain**	**Local *(1)***	**DirectLink *(1)***	**R4 *(1)***	**C4 *(1)***	**LE**	**UFM Block**	**Column IOE**	**Row IOE**	**Fast I/O *(1)***
LUT Chain	—	—	—	—	—	—	✓	—	—	—	—
Register Chain	—	—	—	—	—	—	✓	—	—	—	—
Local Interconnect	—	—	—	—	—	—	✓	✓	✓	✓	—
DirectLink Interconnect	—	—	✓	—	—	—	—	—	—	—	—
R4 Interconnect	—	—	✓	—	✓	✓	—	—	—	—	—
C4 Interconnect	—	—	✓	—	✓	✓	—	—	—	—	—
LE	✓	✓	✓	✓	✓	✓	—	—	✓	✓	✓
UFM Block	—	—	✓	✓	✓	✓	—	—	—	—	—
Column IOE	—	—	—	—	—	✓	—	—	—	—	—
Row IOE	—	—	—	✓	✓	✓	—	—	—	—	—

Note to Table 2–2:

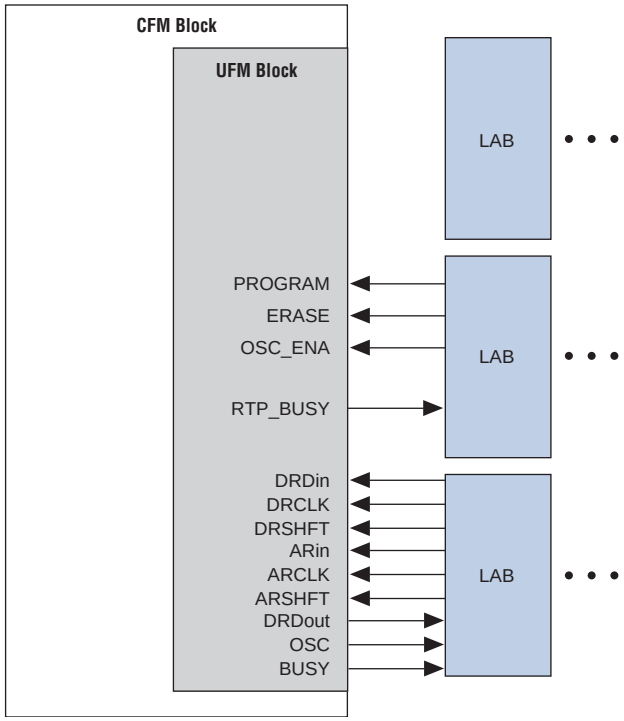
(1) These categories are interconnects.

Global Signals

Each MAX II device has four dual-purpose dedicated clock pins (`GCLK[3..0]`, two pins on the left side and two pins on the right side) that drive the global clock network for clocking, as shown in Figure 2–13. These four pins can also be used as general-purpose I/O if they are not used to drive the global clock network.

The four global clock lines in the global clock network drive throughout the entire device. The global clock network can provide clocks for all resources within the device including LEs, LAB local interconnect, IOEs, and the UFM block. The global clock lines can also be used for global control signals, such as clock enables, synchronous or asynchronous clears, presets, output enables, or protocol control signals such as `TRDY` and `IRDY` for PCI. Internal logic can drive the global clock network for internally-generated global clocks and control signals. Figure 2–13 shows the various sources that drive the global clock network.

Figure 2–16. EPM240 UFM Block LAB Row Interface *(Note 1)*

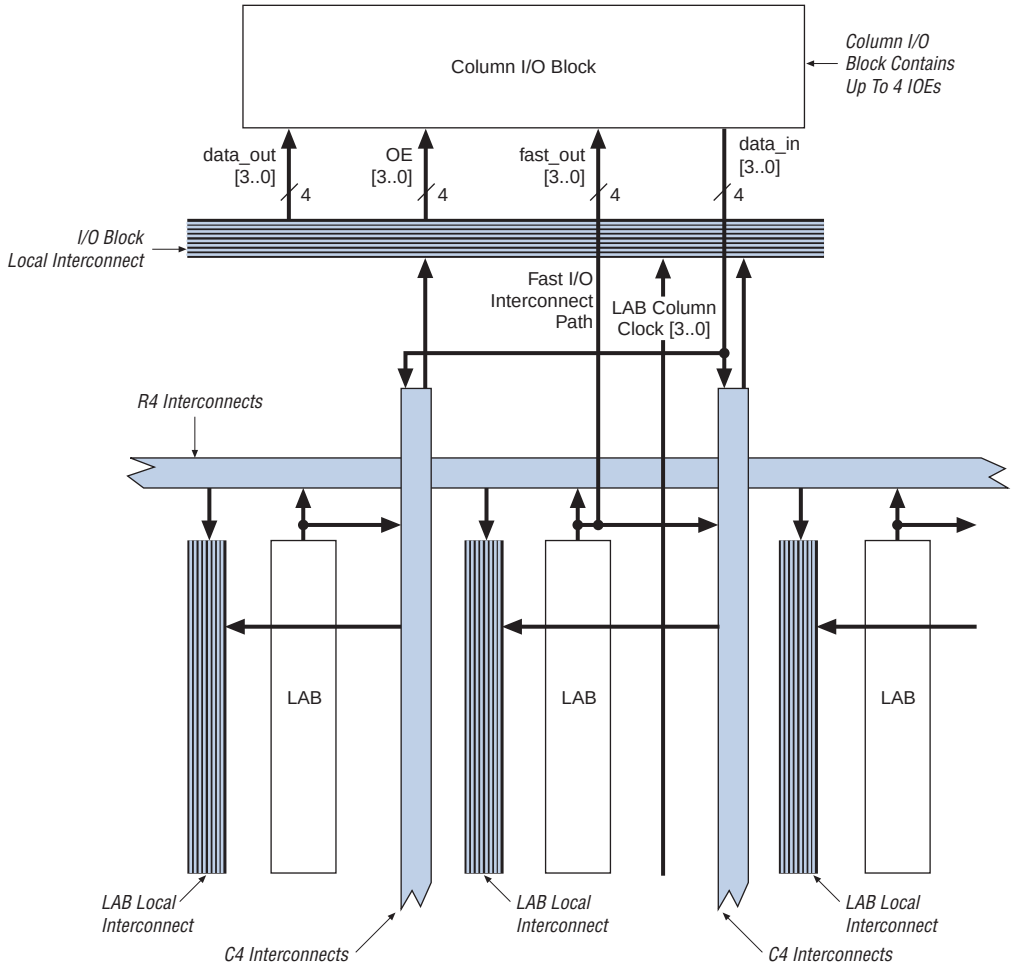


Note to Figure 2–16:

(1) The UFM block inputs and outputs can drive to/from all types of interconnects, not only DirectLink interconnects from adjacent row LABs.

Figure 2–21 shows how a column I/O block connects to the logic array.

Figure 2–21. Column I/O Block Connection to the Interconnect *(Note 1)*



Note to Figure 2–21:

(1) Each of the four IOEs in the column I/O block can have one `data_out` or `fast_out` output, one `OE` output, and one `data_in` input.

I/O Standards and Banks

MAX II device IOEs support the following I/O standards:

- 3.3-V LVTTL/LVCMOS
- 2.5-V LVTTL/LVCMOS
- 1.8-V LVTTL/LVCMOS
- 1.5-V LVCMOS
- 3.3-V PCI

Bus Hold

Each MAX II device I/O pin provides an optional bus-hold feature. The bus-hold circuitry can hold the signal on an I/O pin at its last-driven state. Since the bus-hold feature holds the last-driven state of the pin until the next input signal is present, an external pull-up or pull-down resistor is not necessary to hold a signal level when the bus is tri-stated.

The bus-hold circuitry also pulls undriven pins away from the input threshold voltage where noise can cause unintended high-frequency switching. The designer can select this feature individually for each I/O pin. The bus-hold output will drive no higher than V_{CCIO} to prevent overdriving signals. If the bus-hold feature is enabled, the device cannot use the programmable pull-up option.

The bus-hold circuitry uses a resistor to pull the signal level to the last driven state. The *DC and Switching Characteristics* chapter in the *MAX II Device Handbook* gives the specific sustaining current for each V_{CCIO} voltage level driven through this resistor and overdrive current used to identify the next-driven input level.

The bus-hold circuitry is only active after the device has fully initialized. The bus-hold circuit captures the value on the pin present at the moment user mode is entered.

Programmable Pull-Up Resistor

Each MAX II device I/O pin provides an optional programmable pull-up resistor during user mode. If the designer enables this feature for an I/O pin, the pull-up resistor holds the output to the V_{CCIO} level of the output pin's bank.

The programmable pull-up resistor feature should not be used at the same time as the bus-hold feature on a given I/O pin.

Programmable Input Delay

The MAX II IOE includes a programmable input delay that is activated to ensure zero hold times. A path where a pin directly drives a register, with minimal routing between the two, may require the delay to ensure zero hold time. However, a path where a pin drives a register through long routing or through combinational logic may not require the delay to achieve a zero hold time. The Quartus II software uses this delay to ensure zero hold times when needed.

MultiVolt I/O Interface

The MAX II architecture supports the MultiVolt I/O interface feature, which allows MAX II devices in all packages to interface with systems of different supply voltages. The devices have one set of VCC pins for internal operation (V_{CCINT}), and up to four sets for input buffers and I/O output driver buffers (V_{CCIO}), depending on the number of I/O banks available in the devices where each set of VCC pins powers one I/O bank. The EPM240 and EPM570 devices have two I/O banks respectively while the EPM1270 and EPM2210 devices have four I/O banks respectively.

Table 3–4 shows the programming times for MAX II devices using in-circuit testers to execute the algorithm vectors in hardware. Software-based programming tools used with download cables are slightly slower because of data processing and transfer limitations.

Table 3–4. MAX II Device Family Programming Times

Description	EPM240 EPM240G EPM240Z	EPM570 EPM570G EPM570Z	EPM1270 EPM1270G	EPM2210 EPM2210G	Unit
Erase + Program (1 MHz)	1.72	2.16	2.90	3.92	sec
Erase + Program (10 MHz)	1.65	1.99	2.58	3.40	sec
Verify (1 MHz)	0.09	0.17	0.30	0.49	sec
Verify (10 MHz)	0.01	0.02	0.03	0.05	sec
Complete Program Cycle (1 MHz)	1.81	2.33	3.20	4.41	sec
Complete Program Cycle (10 MHz)	1.66	2.01	2.61	3.45	sec

UFM Programming

The Quartus II software, with the use of POF, Jam, or JBC files, supports programming of the user flash memory (UFM) block independent of the logic array design pattern stored in the CFM block. This allows updating or reading UFM contents through ISP without altering the current logic array design, or vice versa. By default, these programming files and methods will program the entire flash memory contents, which includes the CFM block and UFM contents. The stand-alone embedded Jam STAPL player and Jam Byte-Code Player provides action commands for programming or reading the entire flash memory (UFM and CFM together) or each independently.

For more information, refer to the *Using Jam STAPL for ISP via an Embedded Processor* chapter in the *MAX II Device Handbook*.

In-System Programming Clamp

By default, the IEEE 1532 instruction used for entering ISP automatically tri-states all I/O pins with weak pull-up resistors for the duration of the ISP sequence. However, some systems may require certain pins on MAX II devices to maintain a specific DC logic level during an in-field update. For these systems, an optional in-system programming clamp instruction exists in MAX II circuitry to control I/O behavior during the ISP sequence. The in-system programming clamp instruction enables the device to sample and sustain the value on an output pin (an input pin would remain tri-stated if sampled) or to explicitly set a logic high, logic low, or tri-state value on any pin. Setting these options is controlled on an individual pin basis using the Quartus II software.

For more information, refer to the *Real-Time ISP and ISP Clamp for MAX II Devices* chapter in the *MAX II Device Handbook*.

Programming/Erasure Specifications

Table 5–3 shows the MAX II device family programming/erasure specifications.

Table 5–3. MAX II Device Programming/Erasure Specifications

Parameter	Minimum	Typical	Maximum	Unit
Erase and reprogram cycles	—	—	100 *(1)*	Cycles

Note to Table 5–3:

(1) This specification applies to the UFM and configuration flash memory (CFM) blocks.

DC Electrical Characteristics

Table 5–4 shows the MAX II device family DC electrical characteristics.

Table 5–4. MAX II Device DC Electrical Characteristics *(Note 1)* (Part 1 of 2)

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Unit
I_I	Input pin leakage current	$V_I = V_{CCIO}$ max to 0 V *(2)*	–10	—	10	µA
I_{OZ}	Tri-stated I/O pin leakage current	$V_O = V_{CCIO}$ max to 0 V *(2)*	–10	—	10	µA
$I_{CCSTANDBY}$	V_{CCINT} supply current (standby) *(3)*	MAX II devices	—	12	—	mA
		MAX IIG devices	—	2	—	mA
		EPM240Z (Commercial grade) *(4)*	—	25	90	µA
		EPM240Z (Industrial grade) *(5)*	—	25	139	µA
		EPM570Z (Commercial grade) *(4)*	—	27	96	µA
		EPM570Z (Industrial grade) *(5)*	—	27	152	µA
$V_{SCHMITT}$ *(6)*	Hysteresis for Schmitt trigger input *(7)*	V_{CCIO} = 3.3 V	—	400	—	mV
		V_{CCIO} = 2.5 V	—	190	—	mV
$I_{CCPOWERUP}$	V_{CCINT} supply current during power-up *(8)*	MAX II devices	—	55	—	mA
		MAX IIG and MAX IIZ devices	—	40	—	mA
R_{PULLUP}	Value of I/O pin pull-up resistor during user mode and in-system programming	V_{CCIO} = 3.3 V *(9)*	5	—	25	kΩ
		V_{CCIO} = 2.5 V *(9)*	10	—	40	kΩ
		V_{CCIO} = 1.8 V *(9)*	25	—	60	kΩ
		V_{CCIO} = 1.5 V *(9)*	45	—	95	kΩ

Table 5–4. MAX II Device DC Electrical Characteristics *(Note 1)* (Part 2 of 2)

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Unit
I_{PULLUP}	I/O pin pull-up resistor current when I/O is unprogrammed	—	—	—	300	µA
C_{IO}	Input capacitance for user I/O pin	—	—	—	8	pF
C_{GCLK}	Input capacitance for dual-purpose GCLK/user I/O pin	—	—	—	8	pF

Notes to Table 5–4:

(1) Typical values are for T_A = 25°C, V_{CCINT} = 3.3 or 2.5 V, and V_{CCIO} = 1.5 V, 1.8 V, 2.5 V, or 3.3 V.

(2) This value is specified for normal device operation. The value may vary during power-up. This applies for all V_{CCIO} settings (3.3, 2.5, 1.8, and 1.5 V).

(3) V_I = ground, no load, no toggling inputs.

(4) Commercial temperature ranges from 0°C to 85°C with maximum current at 85°C.

(5) Industrial temperature ranges from –40°C to 100°C with maximum current at 100°C.

(6) This value applies to commercial and industrial range devices. For extended temperature range devices, the $V_{SCHMITT}$ typical value is 300 mV for V_{CCIO} = 3.3 V and 120 mV for V_{CCIO} = 2.5 V.

(7) The TCK input is susceptible to high pulse glitches when the input signal fall time is greater than 200 ns for all I/O standards.

(8) This is a peak current value with a maximum duration of t_{CONFIG} time.

(9) Pin pull-up resistance values will lower if an external source drives the pin higher than V_{CCIO}.

Table 5–6. 3.3-V LVCMOS Specifications (Part 2 of 2)

Symbol	Parameter	Conditions	Minimum	Maximum	Unit
V_{OH}	High-level output voltage	V_{CCIO} = 3.0, IOH = –0.1 mA *(1)*	V_{CCIO} – 0.2	—	V
V_{OL}	Low-level output voltage	V_{CCIO} = 3.0, IOL = 0.1 mA *(1)*	—	0.2	V

Table 5–7. 2.5-V I/O Specifications

Symbol	Parameter	Conditions	Minimum	Maximum	Unit
V_{CCIO}	I/O supply voltage	—	2.375	2.625	V
V_{IH}	High-level input voltage	—	1.7	4.0	V
V_{IL}	Low-level input voltage	—	–0.5	0.7	V
V_{OH}	High-level output voltage	IOH = –0.1 mA *(1)*	2.1	—	V
		IOH = –1 mA *(1)*	2.0	—	V
		IOH = –2 mA *(1)*	1.7	—	V
V_{OL}	Low-level output voltage	IOL = 0.1 mA *(1)*	—	0.2	V
		IOL = 1 mA *(1)*	—	0.4	V
		IOL = 2 mA *(1)*	—	0.7	V

Table 5–8. 1.8-V I/O Specifications

Symbol	Parameter	Conditions	Minimum	Maximum	Unit
V_{CCIO}	I/O supply voltage	—	1.71	1.89	V
V_{IH}	High-level input voltage	—	0.65 × V_{CCIO}	2.25 *(2)*	V
V_{IL}	Low-level input voltage	—	–0.3	0.35 × V_{CCIO}	V
V_{OH}	High-level output voltage	IOH = –2 mA *(1)*	V_{CCIO} – 0.45	—	V
V_{OL}	Low-level output voltage	IOL = 2 mA *(1)*	—	0.45	V

Table 5–9. 1.5-V I/O Specifications

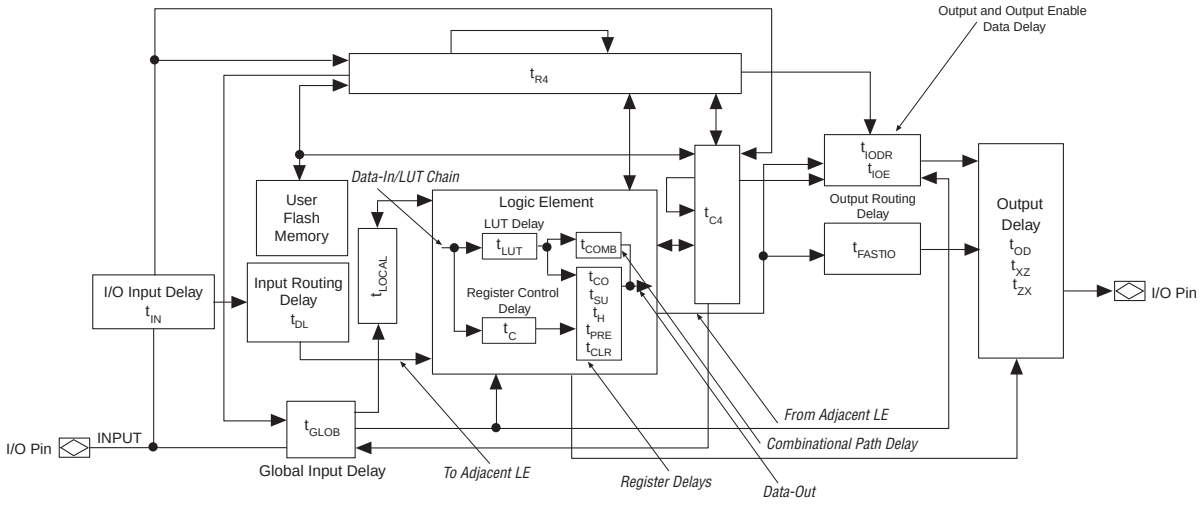
Symbol	Parameter	Conditions	Minimum	Maximum	Unit
V_{CCIO}	I/O supply voltage	—	1.425	1.575	V
V_{IH}	High-level input voltage	—	0.65 × V_{CCIO}	V_{CCIO} + 0.3 *(2)*	V
V_{IL}	Low-level input voltage	—	–0.3	0.35 × V_{CCIO}	V
V_{OH}	High-level output voltage	IOH = –2 mA *(1)*	0.75 × V_{CCIO}	—	V
V_{OL}	Low-level output voltage	IOL = 2 mA *(1)*	—	0.25 × V_{CCIO}	V

Notes to Table 5–5 through Table 5–9:

(1) This specification is supported across all the programmable drive strength settings available for this I/O standard, as shown in the *MAX II Architecture* chapter (*I/O Structure section*) in the *MAX II Device Handbook*.

(2) This maximum V_{IH} reflects the JEDEC specification. The MAX II input buffer can tolerate a V_{IH} maximum of 4.0, as specified by the V_I parameter in Table 5–2.

Figure 5–2. MAX II Device Timing Model



The timing characteristics of any signal path can be derived from the timing model and parameters of a particular device. External timing parameters, which represent pin-to-pin timing delays, can be calculated as the sum of internal parameters.

Refer to the *Understanding Timing in MAX II Devices* chapter in the *MAX II Device Handbook* for more information.

This section describes and specifies the performance, internal, external, and UFM timing specifications. All specifications are representative of the worst-case supply voltage and junction temperature conditions.

Preliminary and Final Timing

Timing models can have either preliminary or final status. The Quartus II software issues an informational message during the design compilation if the timing models are preliminary. Table 5–13 shows the status of the MAX II device timing models.

Preliminary status means the timing model is subject to change. Initially, timing numbers are created using simulation results, process data, and other known parameters. These tests are used to make the preliminary numbers as close to the actual timing parameters as possible.

Final timing numbers are based on actual device operation and testing. These numbers reflect the actual performance of the device under the worst-case voltage and junction temperature conditions.

Table 5–13. MAX II Device Timing Model Status (Part 1 of 2)

Device	Preliminary	Final
EPM240	—	✓
EPM240Z *(1)*	—	✓
EPM570	—	✓
EPM570Z *(1)*	—	✓

Table 5–20. t_{XZ} IOE Microparameter Adders for Slow Slew Rate

Standard		MAX II / MAX IIG						MAX IIZ						Unit
		–3 Speed Grade		–4 Speed Grade		–5 Speed Grade		–6 Speed Grade		–7 Speed Grade		–8 Speed Grade		
		Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	
3.3-V LVTTL	16 mA	—	206	—	–20	—	–247	—	1,433	—	1,446	—	1,454	ps
	8 mA	—	891	—	665	—	438	—	1,332	—	1,345	—	1,348	ps
3.3-V LVCMOS	8 mA	—	206	—	–20	—	–247	—	1,433	—	1,446	—	1,454	ps
	4 mA	—	891	—	665	—	438	—	1,332	—	1,345	—	1,348	ps
2.5-V LVTTL / LVCMOS	14 mA	—	222	—	–4	—	–231	—	213	—	208	—	213	ps
	7 mA	—	943	—	717	—	490	—	166	—	161	—	166	ps
3.3-V PCI	20 mA	—	161	—	210	—	258	—	1,332	—	1,345	—	1,348	ps

The default slew rate setting for MAX II devices in the Quartus II design software is "fast".

Table 5–21. UFM Block Internal Timing Microparameters (Part 1 of 3)

Symbol	Parameter	MAX II / MAX IIG						MAX IIZ						Unit
		–3 Speed Grade		–4 Speed Grade		–5 Speed Grade		–6 Speed Grade		–7 Speed Grade		–8 Speed Grade		
		Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	
t_{ACLK}	Address register clock period	100	—	100	—	100	—	100	—	100	—	100	—	ns
t_{ASU}	Address register shift signal setup to address register clock	20	—	20	—	20	—	20	—	20	—	20	—	ns
t_{AH}	Address register shift signal hold to address register clock	20	—	20	—	20	—	20	—	20	—	20	—	ns
t_{ADS}	Address register data in setup to address register clock	20	—	20	—	20	—	20	—	20	—	20	—	ns
t_{ADH}	Address register data in hold from address register clock	20	—	20	—	20	—	20	—	20	—	20	—	ns
t_{DCLK}	Data register clock period	100	—	100	—	100	—	100	—	100	—	100	—	ns
t_{DSS}	Data register shift signal setup to data register clock	60	—	60	—	60	—	60	—	60	—	60	—	ns
t_{DSH}	Data register shift signal hold from data register clock	20	—	20	—	20	—	20	—	20	—	20	—	ns

Figure 5–4. UFM Program Waveforms

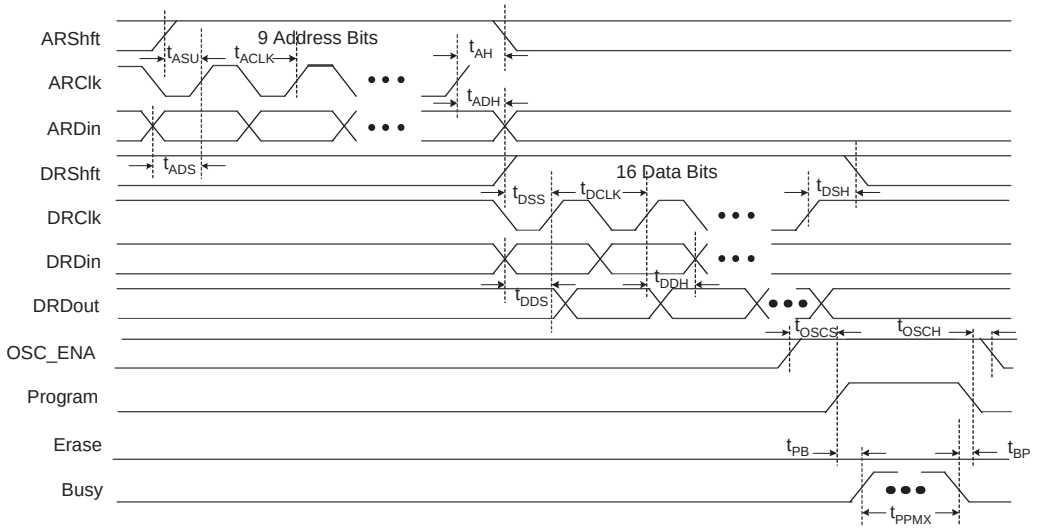


Figure 5–5. UFM Erase Waveform

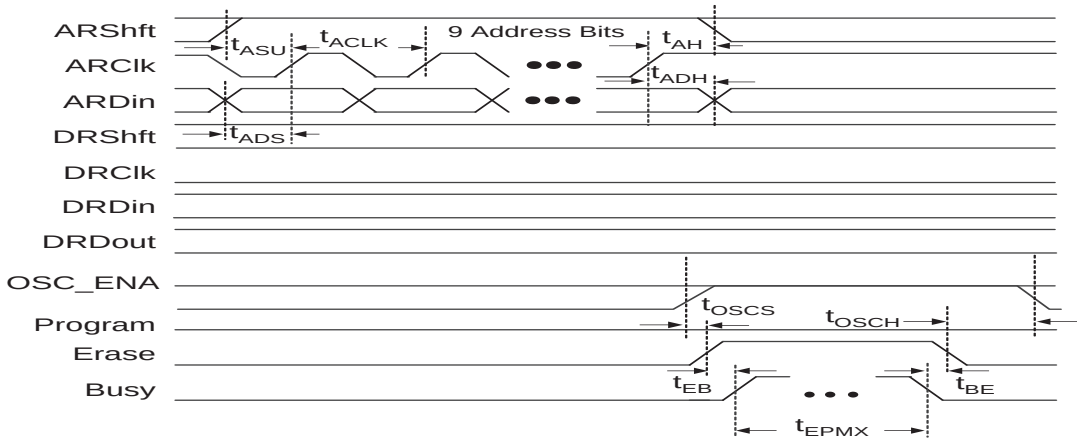


Table 5–22. Routing Delay Internal Timing Microparameters

Routing	MAX II / MAX IIG						MAX IIZ						Unit
	–3 Speed Grade		–4 Speed Grade		–5 Speed Grade		–6 Speed Grade		–7 Speed Grade		–8 Speed Grade		
	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	
t_{C4}	—	429	—	556	—	687	—	*(1)*	—	*(1)*	—	*(1)*	ps
t_{R4}	—	326	—	423	—	521	—	*(1)*	—	*(1)*	—	*(1)*	ps
t_{LOCAL}	—	330	—	429	—	529	—	*(1)*	—	*(1)*	—	*(1)*	ps

Note to Table 5–22:

(1) The numbers will only be available in a later revision.

Table 5–24. EPM570 Global Clock External I/O Timing Parameters (Part 2 of 2)

Symbol	Parameter	Condition	MAX II / MAX IIG						MAX IIZ						Unit
			–3 Speed Grade		–4 Speed Grade		–5 Speed Grade		–6 Speed Grade		–7 Speed Grade		–8 Speed Grade		
			Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	
f_{CNT}	Maximum global clock frequency for 16-bit counter	—	—	304.0 *(1)*	—	247.5	—	201.1	—	184.1	—	123.5	—	118.3	MHz

Note to Table 5–24:

(1) The maximum frequency is limited by the I/O standard on the clock input pin. The 16-bit counter critical delay performs faster than this global clock input pin maximum frequency.

Table 5–25 shows the external I/O timing parameters for EPM1270 devices.

Table 5–25. EPM1270 Global Clock External I/O Timing Parameters

Symbol	Parameter	Condition	MAX II / MAX IIG						Unit
			–3 Speed Grade		–4 Speed Grade		–5 Speed Grade		
			Min	Max	Min	Max	Min	Max	
t_{PD1}	Worst case pin-to-pin delay through 1 look-up table (LUT)	10 pF	—	6.2	—	8.1	—	10.0	ns
t_{PD2}	Best case pin-to-pin delay through 1 LUT	10 pF	—	3.7	—	4.8	—	5.9	ns
t_{SU}	Global clock setup time	—	1.2	—	1.5	—	1.9	—	ns
t_{H}	Global clock hold time	—	0	—	0	—	0	—	ns
t_{CO}	Global clock to output delay	10 pF	2.0	4.6	2.0	5.9	2.0	7.3	ns
t_{CH}	Global clock high time	—	166	—	216	—	266	—	ps
t_{CL}	Global clock low time	—	166	—	216	—	266	—	ps
t_{CNT}	Minimum global clock period for 16-bit counter	—	3.3	—	4.0	—	5.0	—	ns
f_{CNT}	Maximum global clock frequency for 16-bit counter	—	—	304.0 *(1)*	—	247.5	—	201.1	MHz

Note to Table 5–25:

(1) The maximum frequency is limited by the I/O standard on the clock input pin. The 16-bit counter critical delay performs faster than this global clock input pin maximum frequency.

Table 5–33. MAX II Maximum Output Clock Rate for I/O

I/O Standard		MAX II / MAX IIG			MAX IIZ		
		–3 Speed Grade	–4 Speed Grade	–5 Speed Grade	–6 Speed Grade	–7 Speed Grade	–8 Speed Grade
3.3-V LVTTL	304	304	304	304	304	304	MHz
3.3-V LVCMOS	304	304	304	304	304	304	MHz
2.5-V LVTTL	220	220	220	220	220	220	MHz
2.5-V LVCMOS	220	220	220	220	220	220	MHz
1.8-V LVTTL	200	200	200	200	200	200	MHz
1.8-V LVCMOS	200	200	200	200	200	200	MHz
1.5-V LVCMOS	150	150	150	150	150	150	MHz
3.3-V PCI	304	304	304	304	304	304	MHz

JTAG Timing Specifications

Figure 5–6 shows the timing waveforms for the JTAG signals.

Figure 5–6. MAX II JTAG Timing Waveforms

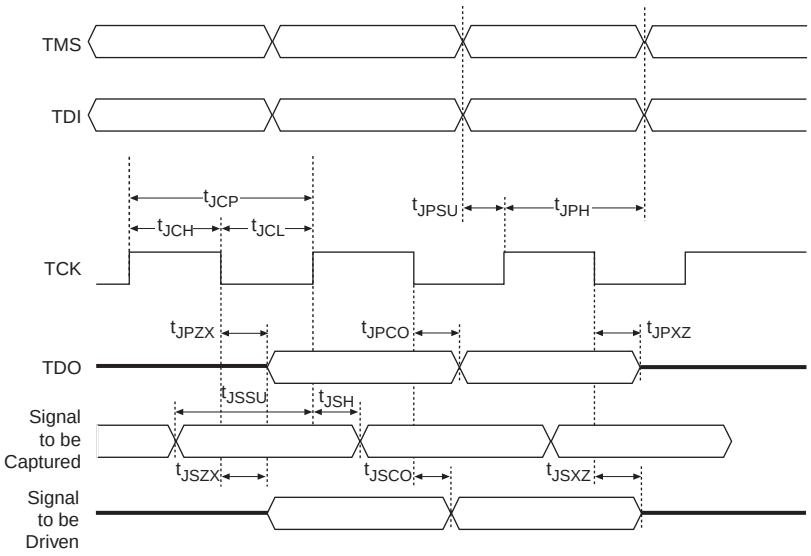


Table 5–34 shows the JTAG Timing parameters and values for MAX II devices.

Table 5–34. MAX II JTAG Timing Parameters (Part 1 of 2)

Symbol	Parameter	Min	Max	Unit
t_{JCP} *(1)*	TCK clock period for V_{CCIO1} = 3.3 V	55.5	—	ns
	TCK clock period for V_{CCIO1} = 2.5 V	62.5	—	ns
	TCK clock period for V_{CCIO1} = 1.8 V	100	—	ns
	TCK clock period for V_{CCIO1} = 1.5 V	143	—	ns
t_{JCH}	TCK clock high time	20	—	ns
t_{JCL}	TCK clock low time	20	—	ns

Document Revision History

Table 5–35 shows the revision history for this chapter.

Table 5–35. Document Revision History (Part 1 of 2)

Date and Revision	Changes Made	Summary of Changes
August 2009, version 2.5	■ Added Table 5–28, Table 5–29, and Table 5–30. ■ Updated Table 5–2, Table 5–4, Table 5–14, Table 5–15, Table 5–16, Table 5–17, Table 5–18, Table 5–19, Table 5–20, Table 5–21, Table 5–22, Table 5–23, Table 5–24, Table 5–27, Table 5–31, Table 5–32, and Table 5–33.	Added information for speed grade –8
November 2008, version 2.4	■ Updated Table 5–2. ■ Updated "Internal Timing Parameters" section.	—
October 2008, version 2.3	■ Updated New Document Format. ■ Updated Figure 5–1.	—
July 2008, version 2.2	■ Updated Table 5–14 , Table 5–23 , and Table 5–24.	—
March 2008, version 2.1	■ Added (Note 5) to Table 5–4.	—
December 2007, version 2.0	■ Updated (Note 3) and (4) to Table 5–1. ■ Updated Table 5–2 and added (Note 5). ■ Updated ICCSTANDBY and ICCPOWERUP information and added IPULLUP information in Table 5–4. ■ Added (Note 1) to Table 5–10. ■ Updated Figure 5–2. ■ Added (Note 1) to Table 5–13. ■ Updated Table 5–13 through Table 5–24, and Table 5–27 through Table 5–30. ■ Added tCOMB information to Table 5–15. ■ Updated Figure 5–6. ■ Added "Referenced Documents" section.	Updated document with MAX IIZ information.
December 2006, version 1.8	■ Added note to Table 5–1. ■ Added document revision history.	—
July 2006, version 1.7	■ Minor content and table updates.	—
February 2006, version 1.6	■ Updated "External Timing I/O Delay Adders" section. ■ Updated Table 5–29. ■ Updated Table 5–30.	—
November 2005, version 1.5	■ Updated Tables 5-2, 5-4, and 5-12.	—
August 2005, version 1.4	■ Updated Figure 5-1. ■ Updated Tables 5-13, 5-16, and 5-26. ■ Removed Note 1 from Table 5-12.	—

Table 5–35. Document Revision History (Part 2 of 2)

Date and Revision	Changes Made	Summary of Changes
June 2005, version 1.3	■ Updated the R_{PULLUP} parameter in Table 5-4. ■ Added Note 2 to Tables 5-8 and 5-9. ■ Updated Table 5-13. ■ Added "Output Drive Characteristics" section. ■ Added I^2C mode and Notes 5 and 6 to Table 5-14. ■ Updated timing values to Tables 5-14 through 5-33.	—
December 2004, version 1.2	■ Updated timing Tables 5-2, 5-4, 5-12, and Tables 15-14 through 5-34. ■ Table 5-31 is new.	—
June 2004, version 1.1	■ Updated timing Tables 5-15 through 5-32.	—