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**Understanding Embedded - CPLDs (Complex Programmable Logic Devices)** 

Embedded - CPLDs, or Complex Programmable Logic Devices, are highly versatile digital logic devices used in electronic systems. These programmable components are designed to perform complex logical operations and can be customized for specific applications. Unlike fixed-function ICs, CPLDs offer the flexibility to reprogram their configuration, making them an ideal choice for various embedded systems. They consist of a set of logic gates and programmable interconnects, allowing designers to implement complex logic circuits without needing custom hardware.

### **Applications of Embedded - CPLDs**

Details	
Product Status	Active
Programmable Type	In System Programmable
Delay Time tpd(1) Max	5.4 ns
Voltage Supply - Internal	1.71V ~ 1.89V
Number of Logic Elements/Blocks	570
Number of Macrocells	440
Number of Gates	-
Number of I/O	116
Operating Temperature	0°C ~ 85°C (TJ)
Mounting Type	Surface Mount
Package / Case	144-LQFP
Supplier Device Package	144-TQFP (20x20)
Purchase URL	https://www.e-xfl.com/product-detail/intel/epm570gt144c5n

Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

Table 2-1. MAX II Device Resources

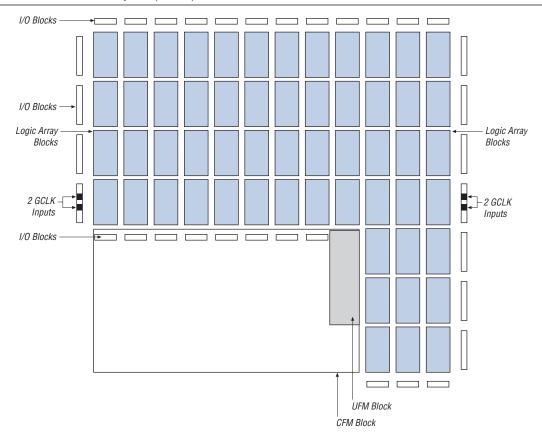
			LAB	Rows	
Devices	UFM Blocks	LAB Columns	Long LAB Rows	Total LABs	
EPM240	1	6	4	_	24
EPM570	1	12	4	3 (3)	57
EPM1270	1	16	7	3 (5)	127
EPM2210	1	20	10	3 (7)	221

### Note to Table 2-1:

(1) The width is the number of LAB columns in length.

Figure 2–2 shows a floorplan of a MAX II device.

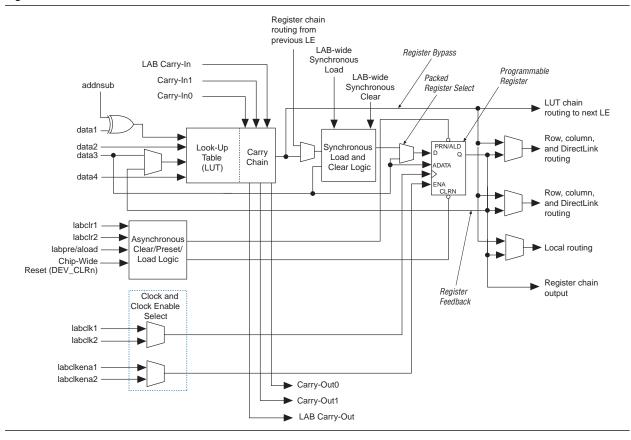
**Figure 2–2.** MAX II Device Floorplan (Note 1)



### Note to Figure 2-2:

(1) The device shown is an EPM570 device. EPM1270 and EPM2210 devices have a similar floorplan with more LABs. For EPM240 devices, the CFM and UFM blocks are located on the left side of the device.

Figure 2-6. MAX II LE



Each LE's programmable register can be configured for D, T, JK, or SR operation. Each register has data, true asynchronous load data, clock, clock enable, clear, and asynchronous load/preset inputs. Global signals, general-purpose I/O pins, or any LE can drive the register's clock and clear control signals. Either general-purpose I/O pins or LEs can drive the clock enable, preset, asynchronous load, and asynchronous data. The asynchronous load data input comes from the data3 input of the LE. For combinational functions, the LUT output bypasses the register and drives directly to the LE outputs.

Each LE has three outputs that drive the local, row, and column routing resources. The LUT or register output can drive these three outputs independently. Two LE outputs drive column or row and DirectLink routing connections and one drives local interconnect resources. This allows the LUT to drive one output while the register drives another output. This register packing feature improves device utilization because the device can use the register and the LUT for unrelated functions. Another special packing mode allows the register output to feed back into the LUT of the same LE so that the register is packed with its own fan-out LUT. This provides another mechanism for improved fitting. The LE can also drive out registered and unregistered versions of the LUT output.

## **LUT Chain and Register Chain**

In addition to the three general routing outputs, the LEs within an LAB have LUT chain and register chain outputs. LUT chain connections allow LUTs within the same LAB to cascade together for wide input functions. Register chain outputs allow registers within the same LAB to cascade together. The register chain output allows an LAB to use LUTs for a single combinational function and the registers to be used for an unrelated shift register implementation. These resources speed up connections between LABs while saving local interconnect resources. Refer to "MultiTrack Interconnect" on page 2–12 for more information about LUT chain and register chain connections.

## addnsub Signal

The LE's dynamic adder/subtractor feature saves logic resources by using one set of LEs to implement both an adder and a subtractor. This feature is controlled by the LAB-wide control signal addnsub. The addnsub signal sets the LAB to perform either A+B or A-B. The LUT computes addition; subtraction is computed by adding the two's complement of the intended subtractor. The LAB-wide signal converts to two's complement by inverting the B bits within the LAB and setting carry-in to 1, which adds one to the least significant bit (LSB). The LSB of an adder/subtractor must be placed in the first LE of the LAB, where the LAB-wide addnsub signal automatically sets the carry-in to 1. The Quartus II Compiler automatically places and uses the adder/subtractor feature when using adder/subtractor parameterized functions.

## **LE Operating Modes**

The MAX II LE can operate in one of the following modes:

- "Normal Mode"
- "Dynamic Arithmetic Mode"

Each mode uses LE resources differently. In each mode, eight available inputs to the LE, the four data inputs from the LAB local interconnect, carry-in0 and carry-in1 from the previous LE, the LAB carry-in from the previous carry-chain LAB, and the register chain connection are directed to different destinations to implement the desired logic function. LAB-wide signals provide clock, asynchronous clear, asynchronous preset/load, synchronous clear, synchronous load, and clock enable control for the register. These LAB-wide signals are available in all LE modes. The addnsub control signal is allowed in arithmetic mode.

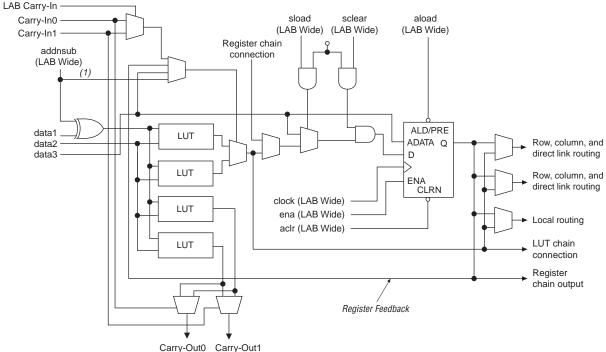
The Quartus II software, in conjunction with parameterized functions such as library of parameterized modules (LPM) functions, automatically chooses the appropriate mode for common functions such as counters, adders, subtractors, and arithmetic functions.

The other two LUTs use the data1 and data2 signals to generate two possible carry-out signals: one for a carry of 1 and the other for a carry of 0. The carry-in0 signal acts as the carry-select for the carry-out0 output and carry-in1 acts as the carry-select for the carry-out1 output. LEs in arithmetic mode can drive out registered and unregistered versions of the LUT output.

The dynamic arithmetic mode also offers clock enable, counter enable, synchronous up/down control, synchronous clear, synchronous load, and dynamic adder/subtractor options. The LAB local interconnect data inputs generate the counter enable and synchronous up/down control signals. The synchronous clear and synchronous load options are LAB-wide signals that affect all registers in the LAB. The Quartus II software automatically places any registers that are not used by the counter into other LABs. The addnsub LAB-wide signal controls whether the LE acts as an adder or subtractor.

Figure 2–8. LE in Dynamic Arithmetic Mode

LAB Carry-In



### Note to Figure 2-8:

(1) The addnsub signal is tied to the carry input for the first LE of a carry chain only.

## **Carry-Select Chain**

The carry-select chain provides a very fast carry-select function between LEs in dynamic arithmetic mode. The carry-select chain uses the redundant carry calculation to increase the speed of carry functions. The LE is configured to calculate outputs for a possible carry-in of 0 and carry-in of 1 in parallel. The carry-in0 and carry-in1 signals from a lower-order bit feed forward into the higher-order bit via the parallel carry chain and feed into both the LUT and the next portion of the carry chain. Carry-select chains can begin in any LE within an LAB.

functions from LE 1 to LE 10 in the same LAB. The register chain connection allows the register output of one LE to connect directly to the register input of the next LE in the LAB for fast shift registers. The Quartus II Compiler automatically takes advantage of these resources to improve utilization and performance. Figure 2–11 shows the LUT chain and register chain interconnects.

Local Interconnect Routing Among LEs in the LAB LE0 LUT Chain Register Chain Routing to Routing to Adjacent Adjacent LE LE's Register Input LE1 Local LE2 Interconnect LE3 LE4 LE5 LE6 LE7 LE8

Figure 2-11. LUT Chain and Register Chain Interconnects

The C4 interconnects span four LABs up or down from a source LAB. Every LAB has its own set of C4 interconnects to drive either up or down. Figure 2–12 shows the C4 interconnect connections from an LAB in a column. The C4 interconnects can drive and be driven by column and row IOEs. For LAB interconnection, a primary LAB or its vertical LAB neighbor can drive a given C4 interconnect. C4 interconnects can drive each other to extend their range as well as drive row interconnects for column-to-column connections.

The UFM block communicates with the logic array similar to LAB-to-LAB interfaces. The UFM block connects to row and column interconnects and has local interconnect regions driven by row and column interconnects. This block also has DirectLink interconnects for fast connections to and from a neighboring LAB. For more information about the UFM interface to the logic array, see "User Flash Memory Block" on page 2–18.

Table 2–2 shows the MAX II device routing scheme.

Table 2-2. MAX II Device Routing Scheme

					Des	tination					
Source	LUT Chain	Register Chain	Local	DirectLink (1)	R4 (1)	C4 (1)	LE	UFM Block	Column IOE	Row IOE	Fast I/0 (1)
LUT Chain	_	_	_	_	_	_	<b>✓</b>	_	_	_	_
Register Chain	_	_	_	_	_	_	<b>✓</b>	_	_	_	_
Local Interconnect	_	_	_	_	_	_	<b>✓</b>	<b>✓</b>	~	~	_
DirectLink Interconnect	_	_	<b>✓</b>	_	_	_	_	_	_	_	_
R4 Interconnect	_	_	<b>✓</b>	_	<b>✓</b>	<b>✓</b>	_	_	_	_	_
C4 Interconnect	_	_	<b>✓</b>	_	<b>✓</b>	<b>✓</b>	_	_	_	-	_
LE	<b>✓</b>	✓	<b>✓</b>	✓	<b>✓</b>	<b>✓</b>	_	_	<b>✓</b>	<b>✓</b>	<b>✓</b>
UFM Block	_	_	<b>✓</b>	<b>✓</b>	<b>✓</b>	<b>✓</b>	_	_	_	-	_
Column IOE	_	_	_	_	_	<b>✓</b>	_	_	_	_	_
Row IOE	_	_	_	✓	<b>✓</b>	<b>✓</b>	_	_	_	_	_

Note to Table 2-2:

(1) These categories are interconnects.

# **Global Signals**

Each MAX II device has four dual-purpose dedicated clock pins (GCLK[3..0], two pins on the left side and two pins on the right side) that drive the global clock network for clocking, as shown in Figure 2–13. These four pins can also be used as general-purpose I/O if they are not used to drive the global clock network.

The four global clock lines in the global clock network drive throughout the entire device. The global clock network can provide clocks for all resources within the device including LEs, LAB local interconnect, IOEs, and the UFM block. The global clock lines can also be used for global control signals, such as clock enables, synchronous or asynchronous clears, presets, output enables, or protocol control signals such as TRDY and IRDY for PCI. Internal logic can drive the global clock network for internally-generated global clocks and control signals. Figure 2–13 shows the various sources that drive the global clock network.

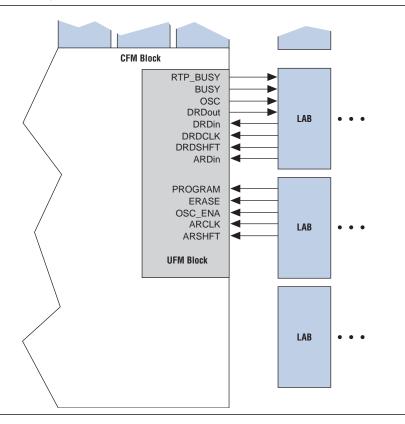


Figure 2-17. EPM570, EPM1270, and EPM2210 UFM Block LAB Row Interface

# **MultiVolt Core**

The MAX II architecture supports the MultiVolt core feature, which allows MAX II devices to support multiple  $V_{CC}$  levels on the  $V_{CCNT}$  supply. An internal linear voltage regulator provides the necessary 1.8-V internal voltage supply to the device. The voltage regulator supports 3.3-V or 2.5-V supplies on its inputs to supply the 1.8-V internal voltage to the device, as shown in Figure 2–18. The voltage regulator is not guaranteed for voltages that are between the maximum recommended 2.5-V operating voltage and the minimum recommended 3.3-V operating voltage.

The MAX IIG and MAX IIZ devices use external 1.8-V supply. The 1.8-V  $V_{CC}$  external supply powers the device core directly.

Figure 2–18. MultiVolt Core Feature in MAX II Devices

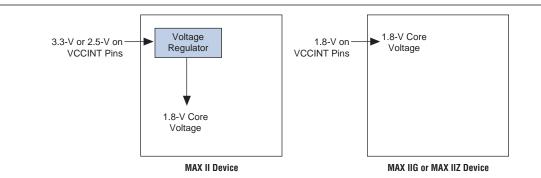
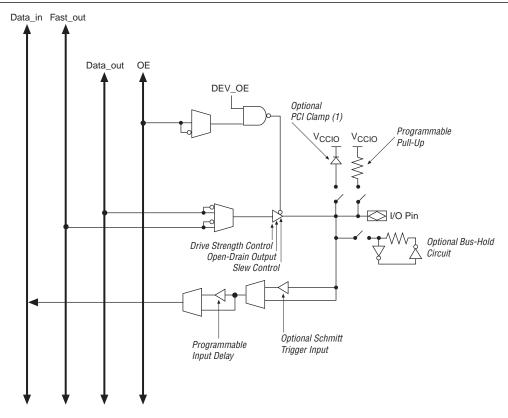


Figure 2–19. MAX II IOE Structure



Note to Figure 2-19:

(1) Available in EPM1270 and EPM2210 devices only.

## I/O Blocks

The IOEs are located in I/O blocks around the periphery of the MAX II device. There are up to seven IOEs per row I/O block (5 maximum in the EPM240 device) and up to four IOEs per column I/O block. Each column or row I/O block interfaces with its adjacent LAB and MultiTrack interconnect to distribute signals throughout the device. The row I/O blocks drive row, column, or DirectLink interconnects. The column I/O blocks drive column interconnects.

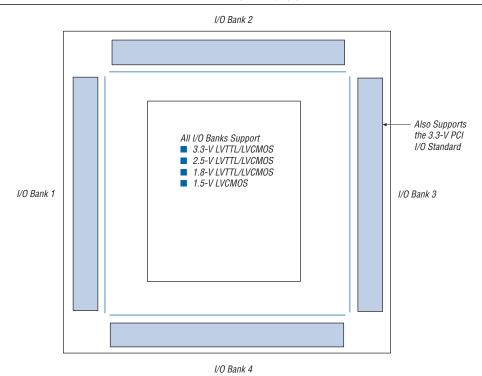


Figure 2–23. MAX II I/O Banks for EPM1270 and EPM2210 (Note 1), (2)

#### Notes to Figure 2-23:

- (1) Figure 2–23 is a top view of the silicon die.
- (2) Figure 2-23 is a graphical representation only. Refer to the pin list and the Quartus II software for exact pin locations.

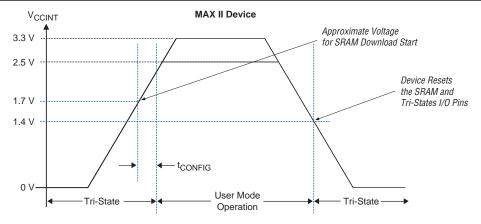
Each I/O bank has dedicated  $V_{\text{CCIO}}$  pins that determine the voltage standard support in that bank. A single device can support 1.5-V, 1.8-V, 2.5-V, and 3.3-V interfaces; each individual bank can support a different standard. Each I/O bank can support multiple standards with the same  $V_{\text{CCIO}}$  for input and output pins. For example, when  $V_{\text{CCIO}}$  is 3.3 V, Bank 3 can support LVTTL, LVCMOS, and 3.3-V PCI.  $V_{\text{CCIO}}$  powers both the input and output buffers in MAX II devices.

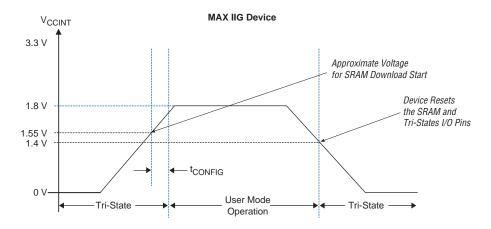
The JTAG pins for MAX II devices are dedicated pins that cannot be used as regular I/O pins. The pins TMS, TDI, TDO, and TCK support all the I/O standards shown in Table 2–4 on page 2–27 except for PCI. These pins reside in Bank 1 for all MAX II devices and their I/O standard support is controlled by the  $V_{\text{CCIO}}$  setting for Bank 1.

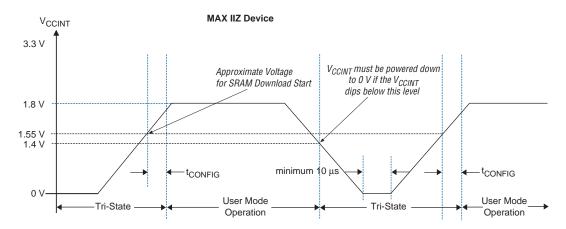
### **PCI Compliance**

The MAX II EPM1270 and EPM2210 devices are compliant with PCI applications as well as all 3.3-V electrical specifications in the *PCI Local Bus Specification Revision 2.2*. These devices are also large enough to support PCI intellectual property (IP) cores. Table 2–5 shows the MAX II device speed grades that meet the PCI timing specifications.

Figure 4–5. Power-Up Characteristics for MAX II, MAX IIG, and MAX IIZ Devices (Note 1), (2)







## Notes to Figure 4-5:

- (1) Time scale is relative.
- (2) Figure 4-5 assumes all V<sub>CCIO</sub> banks power up simultaneously with the V<sub>CCINT</sub> profile shown. If not, t<sub>CONFIG</sub> stretches out until all V<sub>CCIO</sub> banks are powered.



After SRAM configuration, all registers in the device are cleared and released into user function before I/O tri-states are released. To release clears after tri-states are released, use the DEV\_CLRn pin option. To hold the tri-states beyond the power-up configuration time, use the DEV\_OE pin option.



# 5. DC and Switching Characteristics

MII51005-2.5

## Introduction

System designers must consider the recommended DC and switching conditions discussed in this chapter to maintain the highest possible performance and reliability of the MAX® II devices. This chapter contains the following sections:

- "Operating Conditions" on page 5–1
- "Power Consumption" on page 5–8
- "Timing Model and Specifications" on page 5–8

# **Operating Conditions**

Table 5–1 through Table 5–12 provide information about absolute maximum ratings, recommended operating conditions, DC electrical characteristics, and other specifications for MAX II devices.

## **Absolute Maximum Ratings**

Table 5–1 shows the absolute maximum ratings for the MAX II device family.

**Table 5–1.** MAX II Device Absolute Maximum Ratings (Note 1), (2)

0	D	0			
Symbol	Parameter	Conditions	Minimum	Maximum	Unit
V <sub>CCINT</sub>	Internal supply voltage (3)	With respect to ground	-0.5	4.6	V
$V_{\text{CCIO}}$	I/O supply voltage	_	-0.5	4.6	V
Vı	DC input voltage	_	-0.5	4.6	V
I <sub>OUT</sub>	DC output current, per pin (4)	_	-25	25	mA
$T_{STG}$	Storage temperature	No bias	-65	150	°C
T <sub>AMB</sub>	Ambient temperature	Under bias (5)	-65	135	°C
T <sub>J</sub>	Junction temperature	TQFP and BGA packages under bias	_	135	°C

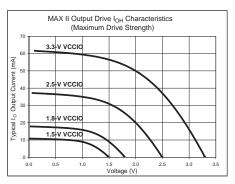
### Notes to Table 5-1:

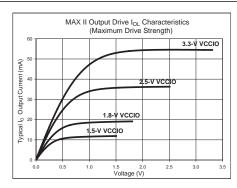
- (1) Refer to the Operating Requirements for Altera Devices Data Sheet.
- (2) Conditions beyond those listed in Table 5–1 may cause permanent damage to a device. Additionally, device operation at the absolute maximum ratings for extended periods of time may have adverse affects on the device.
- (3) Maximum V<sub>CCINT</sub> for MAX II devices is 4.6 V. For MAX IIG and MAX IIZ devices, it is 2.4 V.
- (4) Refer to AN 286: Implementing LED Drivers in MAX & MAX II Devices for more information about the maximum source and sink current for MAX II devices.
- (5) Refer to Table 5-2 for information about "under bias" conditions.

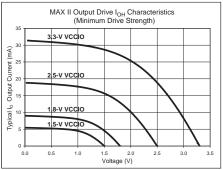
## **Output Drive Characteristics**

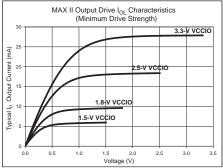
Figure 5–1 shows the typical drive strength characteristics of MAX II devices.

Figure 5-1. Output Drive Characteristics of MAX II Devices









### Note to Figure 5-1:

(1) The DC output current per pin is subject to the absolute maximum rating of Table 5-1.

# I/O Standard Specifications

Table 5–5 through Table 5–10 show the MAX II device family I/O standard specifications.

Table 5-5. 3.3-V LVTTL Specifications

Symbol	Parameter	Conditions	Minimum	Maximum	Unit
V <sub>CCIO</sub>	I/O supply voltage	_	3.0	3.6	V
V <sub>IH</sub>	High-level input voltage	_	1.7	4.0	V
V <sub>IL</sub>	Low-level input voltage	_	-0.5	0.8	V
V <sub>OH</sub>	High-level output voltage	IOH = -4 mA (1)	2.4	_	V
V <sub>oL</sub>	Low-level output voltage	IOL = 4 mA (1)	_	0.45	V

Table 5-6. 3.3-V LVCMOS Specifications (Part 1 of 2)

Symbol	Parameter	Conditions	Minimum	Maximum	Unit
V <sub>CCIO</sub>	I/O supply voltage	_	3.0	3.6	V
V <sub>IH</sub>	High-level input voltage	_	1.7	4.0	V
V <sub>IL</sub>	Low-level input voltage	_	-0.5	0.8	V

**Table 5–10.** 3.3-V PCI Specifications (*Note 1*)

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Unit
V <sub>ccio</sub>	I/O supply voltage	_	3.0	3.3	3.6	V
V <sub>IH</sub>	High-level input voltage	_	0.5 × V <sub>ccio</sub>	_	V <sub>cc10</sub> + 0.5	V
V <sub>IL</sub>	Low-level input voltage	_	-0.5	_	0.3 × V <sub>cc10</sub>	V
V <sub>OH</sub>	High-level output voltage	IOH = -500 μA	0.9 × V <sub>ccio</sub>	_	_	V
V <sub>OL</sub>	Low-level output voltage	IOL = 1.5 mA	_	_	0.1 × V <sub>CC10</sub>	V

### Note to Table 5-10:

(1) 3.3-V PCI I/O standard is only supported in Bank 3 of the EPM1270 and EPM2210 devices.

## **Bus Hold Specifications**

Table 5–11 shows the MAX II device family bus hold specifications.

**Table 5–11.** Bus Hold Specifications

			V <sub>ccio</sub> Level									
		1.	5 V	1.8 V		2.5 V		3.3 V				
Parameter	Conditions	Min	Max	Min	Max	Min	Max	Min	Max	Unit		
Low sustaining current	$V_{IN} > V_{IL}$ (maximum)	20	_	30	_	50	_	70	_	μА		
High sustaining current	$V_{IN} < V_{IH}$ (minimum)	-20	_	-30	_	-50	_	-70	_	μА		
Low overdrive current	0 V < V <sub>IN</sub> < V <sub>CCIO</sub>	_	160	_	200	_	300	_	500	μΑ		
High overdrive current	0 V < V <sub>IN</sub> < V <sub>CCIO</sub>	_	-160	_	-200	_	-300	_	-500	μА		

**Table 5–17.**  $t_{ZX}$  IOE Microparameter Adders for Fast Slew Rate (Part 2 of 2)

MAX II / MAX IIG										MA	X IIZ			
		1	-3 Speed -4 Speed Grade Grade		•	–5 Speed Grade			peed ade		peed ade	–8 Speed Grade		
Standard	Standard M		Max	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Unit
1.5-V LVCMOS	4 mA	_	1,118	_	1,454	_	1,789	_	580	_	588	_	588	ps
	2 mA	_	2,410	_	3,133	_	3,856	_	915	_	923	_	923	ps
3.3-V PCI	20 mA	_	19	_	25	_	31	_	72	_	71	_	74	ps

**Table 5–18.**  $t_{\text{ZX}}$  IOE Microparameter Adders for Slow Slew Rate

			ı	II XAN	/ MAX IIG					MA	X IIZ			
		-3 Speed -4 Speed Grade Grade				-6 Speed Grade		-7 Speed Grade		–8 Speed Grade				
Standar	d	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Unit
3.3-V LVTTL	16 mA	_	6,350	_	6,050	_	5,749	_	5,951	_	5,952	_	6,063	ps
	8 mA	_	9,383	_	9,083	_	8,782	_	6,534	_	6,533	_	6,662	ps
3.3-V LVCMOS	8 mA	_	6,350	_	6,050	_	5,749	_	5,951	_	5,952	_	6,063	ps
	4 mA		9,383	_	9,083	_	8,782	_	6,534	_	6,533	_	6,662	ps
2.5-V LVTTL /	14 mA	_	10,412	_	10,112	_	9,811	_	9,110	_	9,105	_	9,237	ps
LVCMOS	7 mA	_	13,613	_	13,313	_	13,012	_	9,830	_	9,835	_	9,977	ps
3.3-V PCI	20 mA	_	-75	_	-97	_	-120	_	6,534	_	6,533	_	6,662	ps

**Table 5–19.**  $t_{\chi\chi}$  IOE Microparameter Adders for Fast Slew Rate

			N	MAX II /	MAX II	G				MA	X IIZ			
		-3 Speed -4 Speed Grade Grade		1	–5 Speed –6 Speed Grade Grade			ı	peed ade		peed ade			
Standar	d	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Unit
3.3-V LVTTL	16 mA	_	0	_	0	_	0	_	0	_	0	_	0	ps
	8 mA	_	-56	_	-72	_	-89		-69	_	-69	_	-69	ps
3.3-V LVCMOS	8 mA	_	0	_	0	_	0		0	_	0	_	0	ps
	4 mA	_	-56	_	-72	_	-89		-69	_	-69	_	-69	ps
2.5-V LVTTL /	14 mA	_	-3	_	-4	_	<b>-</b> 5		-7	_	-11	_	-11	ps
LVCMOS	7 mA	_	<del>-4</del> 7	_	-61	_	-75		-66	_	-70	_	-70	ps
1.8-V LVTTL /	6 mA	_	119	_	155	_	191	_	45	_	34	_	37	ps
LVCMOS	3 mA	_	207	_	269	_	331		34	_	22	_	25	ps
1.5-V LVCMOS	4 mA	_	606	_	788	_	970	_	166	_	154	_	155	ps
	2 mA	_	673	_	875	_	1,077	_	190	_	177	_	179	ps
3.3-V PCI	20 mA	_	71	_	93	_	114	_	-69	_	-69	_	-69	ps

			N	IAX II /	MAX II	G				MA	X IIZ			
	-3 Speed Grade		-4 Speed -5 Spe Grade Grad		•	d –6 Speed Grade		-7 Speed Grade		–8 Speed Grade				
Standar	d	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Unit
3.3-V LVTTL	16 mA	_	206	_	-20	_	-247	_	1,433	_	1,446	_	1,454	ps
	8 mA	_	891	_	665	_	438	_	1,332	_	1,345	_	1,348	ps
3.3-V LVCMOS	8 mA	_	206	_	-20	_	-247	_	1,433	_	1,446	_	1,454	ps
	4 mA	_	891	_	665	_	438	_	1,332	_	1,345	_	1,348	ps
2.5-V LVTTL /	14 mA	_	222	_	-4	_	-231	_	213	_	208	_	213	ps
LVCMOS	7 mA	_	943	_	717	_	490	_	166	_	161	_	166	ps
3.3-V PCI	20 mA	_	161	_	210	_	258	_	1,332	_	1,345	_	1,348	ps

**Table 5–20.**  $t_{XZ}$  IOE Microparameter Adders for Slow Slew Rate

The default slew rate setting for MAX II devices in the Quartus II design software is "fast".

**Table 5–21.** UFM Block Internal Timing Microparameters (Part 1 of 3)

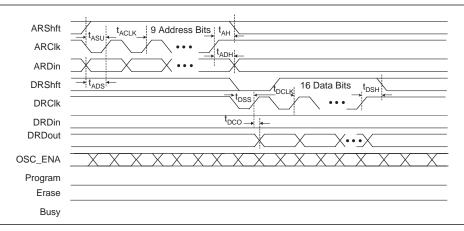
			M	IAX II /	MAX II	IG		MAX IIZ						
Symbol		–3 S <sub>l</sub> Gra			peed ade	−5 S Gra	peed ide	−6 S <sub> </sub> Gra			peed ade	–8 S Gra	peed ade	
	Parameter	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Unit
t <sub>ACLK</sub>	Address register clock period	100	_	100	_	100	_	100	_	100	_	100	_	ns
t <sub>ASU</sub>	Address register shift signal setup to address register clock	20		20	_	20	_	20	_	20	_	20	_	ns
t <sub>AH</sub>	Address register shift signal hold to address register clock	20	_	20	_	20	_	20	_	20	_	20	_	ns
t <sub>ADS</sub>	Address register data in setup to address register clock	20	_	20	_	20	_	20	_	20	_	20	_	ns
t <sub>ADH</sub>	Address register data in hold from address register clock	20	_	20	_	20	_	20	_	20	_	20	_	ns
t <sub>dclk</sub>	Data register clock period	100		100	_	100	_	100	_	100	_	100	_	ns
t <sub>DSS</sub>	Data register shift signal setup to data register clock	60	_	60	_	60	_	60	_	60	_	60	_	ns
t <sub>DSH</sub>	Data register shift signal hold from data register clock	20	_	20	_	20	_	20	_	20	_	20	_	ns

**Table 5–21.** UFM Block Internal Timing Microparameters (Part 3 of 3)

			M	AX II /	MAX II	G		MAX IIZ						
		-3 Speed Grade			peed ade	–5 Speed Grade		-6 Speed Grade		-7 Speed Grade		–8 Speed Grade		
Symbol	Parameter	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Unit
toe	Delay from data register clock to data register output	180	_	180	_	180	_	180	_	180	_	180	_	ns
t <sub>RA</sub>	Maximum read access time	_	65	_	65	_	65	_	65	_	65	_	65	ns
t <sub>oscs</sub>	Maximum delay between the OSC_ENA rising edge to the erase/program signal rising edge	250	_	250	_	250	_	250	_	250	_	250	_	ns
t <sub>osch</sub>	Minimum delay allowed from the erase/program signal going low to OSC_ENA signal going low	250	_	250	_	250	_	250	_	250	_	250	_	ns

Figure 5–3 through Figure 5–5 show the read, program, and erase waveforms for UFM block timing parameters shown in Table 5–21.

Figure 5-3. UFM Read Waveforms



**Table 5–23.** EPM240 Global Clock External I/O Timing Parameters (Part 2 of 2)

			MAX II / MAX IIG							MAX IIZ						
			–3 Speed Grade		-4 Speed Grade		–5 Speed Grade		–6 Speed Grade		-7 Speed Grade		-8 Speed Grade			
Symbol	Parameter	Condition	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Unit	
f <sub>CNT</sub>	Maximum global clock frequency for 16-bit counter	_	_	304.0	_	247.5	_	201.1	_	184.1	_	123.5	_	118.3	MHz	

#### Note to Table 5-23:

Table 5–24 shows the external I/O timing parameters for EPM570 devices.

**Table 5–24.** EPM570 Global Clock External I/O Timing Parameters (Part 1 of 2)

	Parameter			N	/AX II	/ MAX I	IG				MA	X IIZ			
				Speed ade		Speed ade	–5 Speed Grade		-6 Speed Grade		-7 Speed Grade		–8 Speed Grade		
Symbol		Condition	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Unit
t <sub>PD1</sub>	Worst case pin- to-pin delay through 1 look- up table (LUT)	10 pF	_	5.4	_	7.0	_	8.7	_	9.5	_	15.1	_	17.7	ns
t <sub>PD2</sub>	Best case pin- to-pin delay through 1 LUT	10 pF	_	3.7	_	4.8	_	5.9	_	5.7	_	7.7	_	8.5	ns
t <sub>su</sub>	Global clock setup time	_	1.2	_	1.5	_	1.9	_	2.2	_	3.9	_	4.4	_	ns
t <sub>H</sub>	Global clock hold time	_	0	_	0	_	0	_	0	_	0	_	0	_	ns
t <sub>co</sub>	Global clock to output delay	10 pF	2.0	4.5	2.0	5.8	2.0	7.1	2.0	6.7	2.0	8.2	2.0	8.7	ns
t <sub>ch</sub>	Global clock high time	_	166	_	216	_	266	_	253	_	335	_	339	_	ps
t <sub>CL</sub>	Global clock low time	_	166	_	216	_	266	_	253	_	335	_	339	_	ps
t <sub>cnt</sub>	Minimum global clock period for 16-bit counter	_	3.3	_	4.0	_	5.0	_	5.4	_	8.1	_	8.4	_	ns

<sup>(1)</sup> The maximum frequency is limited by the I/O standard on the clock input pin. The 16-bit counter critical delay performs faster than this global clock input pin maximum frequency.

Table 5–26 shows the external I/O timing parameters for EPM2210 devices.

Table 5–26. EPM2210 Global Clock External I/O Timing Parameters

			–3 Spee	ed Grade	-4 Spee	d Grade	–5 Spe		
Symbol	Parameter	Condition	Min	Max	Min	Max	Min	Max	Unit
t <sub>PD1</sub>	Worst case pin-to-pin delay through 1 look-up table (LUT)	10 pF	_	7.0	_	9.1	_	11.2	ns
t <sub>PD2</sub>	Best case pin-to-pin delay through 1 LUT	10 pF	_	3.7	_	4.8	_	5.9	ns
t <sub>su</sub>	Global clock setup time	_	1.2	_	1.5	_	1.9	_	ns
t <sub>H</sub>	Global clock hold time	_	0	_	0	_	0	_	ns
t <sub>co</sub>	Global clock to output delay	10 pF	2.0	4.6	2.0	6.0	2.0	7.4	ns
t <sub>CH</sub>	Global clock high time	_	166	_	216	_	266	_	ps
t <sub>cL</sub>	Global clock low time	_	166	_	216	_	266	_	ps
t <sub>cnt</sub>	Minimum global clock period for 16-bit counter	_	3.3	_	4.0	_	5.0	_	ns
$f_{\text{CNT}}$	Maximum global clock frequency for 16-bit counter	_	_	304.0 (1)	_	247.5	_	201.1	MHz

#### Note to Table 5-26:

# **External Timing I/O Delay Adders**

The I/O delay timing parameters for I/O standard input and output adders, and input delays are specified by speed grade independent of device density.

Table 5–27 through Table 5–31 show the adder delays associated with I/O pins for all packages. The delay numbers for –3, –4, and –5 speed grades shown in Table 5–27 through Table 5–33 are based on an EPM1270 device target, while –6, –7, and –8 speed grade values are based on an EPM570Z device target. If an I/O standard other than 3.3-V LVTTL is selected, add the input delay adder to the external  $t_{\rm SU}$  timing parameters shown in Table 5–23 through Table 5–26. If an I/O standard other than 3.3-V LVTTL with 16 mA drive strength and fast slew rate is selected, add the output delay adder to the external  $t_{\rm CO}$  and  $t_{\rm PD}$  shown in Table 5–23 through Table 5–26.

**Table 5–27.** External Timing Input Delay Adders (Part 1 of 2)

			N	MAX II ,	MAX I	IG	MAX IIZ								
		1	-3 Speed		•	-8 Speed Grade									
I/O Standard		Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Unit	
3.3-V LVTTL	Without Schmitt Trigger	_	0	_	0	_	0	_	0	_	0	_	0	ps	
	With Schmitt Trigger	_	334	_	434	_	535	_	387	_	434	_	442	ps	

<sup>(1)</sup> The maximum frequency is limited by the I/O standard on the clock input pin. The 16-bit counter critical delay performs faster than this global clock input pin maximum frequency.