Welcome to E-XFL.COM

Understanding Embedded - CPLDs (Complex Programmable Logic Devices)

Embedded - CPLDs, or Complex Programmable Logic Devices, are highly versatile digital logic devices used in electronic systems. These programmable components are designed to perform complex logical operations and can be customized for specific applications. Unlike fixed-function ICs, CPLDs offer the flexibility to reprogram their configuration, making them an ideal choice for various embedded systems. They consist of a set of logic gates and programmable interconnects, allowing designers to implement complex logic circuits without needing custom hardware.

Applications of Embedded - CPLDs

Details	
Product Status	Active
Programmable Type	In System Programmable
Delay Time tpd(1) Max	5.4 ns
Voltage Supply - Internal	2.5V, 3.3V
Number of Logic Elements/Blocks	570
Number of Macrocells	440
Number of Gates	-
Number of I/O	116
Operating Temperature	0°C ~ 85°C (TJ)
Mounting Type	Surface Mount
Package / Case	144-LQFP
Supplier Device Package	144-TQFP (20x20)
Purchase URL	https://www.e-xfl.com/product-detail/intel/epm570t144c4

1. Introduction

MII51001-1.9

Introduction

The MAX® II family of instant-on, non-volatile CPLDs is based on a 0.18-µm, 6-layer-metal-flash process, with densities from 240 to 2,210 logic elements (LEs) (128 to 2,210 equivalent macrocells) and non-volatile storage of 8 Kbits. MAX II devices offer high I/O counts, fast performance, and reliable fitting versus other CPLD architectures. Featuring MultiVolt core, a user flash memory (UFM) block, and enhanced in-system programmability (ISP), MAX II devices are designed to reduce cost and power while providing programmable solutions for applications such as bus bridging, I/O expansion, power-on reset (POR) and sequencing control, and device configuration control.

Features

The MAX II CPLD has the following features:

- Low-cost, low-power CPLD
- Instant-on, non-volatile architecture
- Standby current as low as 25 µA
- Provides fast propagation delay and clock-to-output times
- Provides four global clocks with two clocks available per logic array block (LAB)
- UFM block up to 8 Kbits for non-volatile storage
- MultiVolt core enabling external supply voltages to the device of either 3.3 V/2.5 V or 1.8 V
- MultiVolt I/O interface supporting 3.3-V, 2.5-V, 1.8-V, and 1.5-V logic levels
- Bus-friendly architecture including programmable slew rate, drive strength, bus-hold, and programmable pull-up resistors
- Schmitt triggers enabling noise tolerant inputs (programmable per pin)
- I/Os are fully compliant with the Peripheral Component Interconnect Special Interest Group (PCI SIG) PCI Local Bus Specification, Revision 2.2 for 3.3-V operation at 66 MHz
- Supports hot-socketing
- Built-in Joint Test Action Group (JTAG) boundary-scan test (BST) circuitry compliant with IEEE Std. 1149.1-1990
- ISP circuitry compliant with IEEE Std. 1532

MAX II devices have an internal linear voltage regulator which supports external supply voltages of 3.3 V or 2.5 V, regulating the supply down to the internal operating voltage of 1.8 V. MAX IIG and MAX IIZ devices only accept 1.8 V as the external supply voltage. MAX IIZ devices are pin-compatible with MAX IIG devices in the 100-pin Micro FineLine BGA and 256-pin Micro FineLine BGA packages. Except for external supply voltage requirements, MAX II and MAX II G devices have identical pin-outs and timing specifications. Table 1–5 shows the external supply voltages supported by the MAX II family.

Table 1–5. MAX II External Supply Voltages

Devices	EPM240 EPM570 EPM1270 EPM2210	EPM240G EPM570G EPM1270G EPM2210G EPM240Z EPM570Z *(1)*
MultiVolt core external supply voltage (V_{CCINT}) *(2)*	3.3 V, 2.5 V	1.8 V
MultiVolt I/O interface voltage levels (V_{CCIO})	1.5 V, 1.8 V, 2.5 V, 3.3 V	1.5 V, 1.8 V, 2.5 V, 3.3 V

Notes to Table 1–5:

(1) MAX IIG and MAX IIZ devices only accept 1.8 V on their `VCCINT` pins. The 1.8-V V_{CCINT} external supply powers the device core directly.

(2) MAX II devices operate internally at 1.8 V.

Referenced Documents

This chapter references the following documents:

- *DC and Switching Characteristics* chapter in the *MAX II Device Handbook*
- *MAX II Logic Element to Macrocell Conversion Methodology* white paper

Document Revision History

Table 1–6 shows the revision history for this chapter.

Table 1–6. Document Revision History

Date and Revision	Changes Made	Summary of Changes
August 2009, version 1.9	■ Updated Table 1–2.	Added information for speed grade –8
October 2008, version 1.8	■ Updated "Introduction" section. ■ Updated new Document Format.	—
December 2007, version1.7	■ Updated Table 1–1 through Table 1–5. ■ Added "Referenced Documents" section.	Updated document with MAX IIZ information.
December 2006, version 1.6	■ Added document revision history.	—
August 2006, version 1.5	■ Minor update to features list.	—
July 2006, version 1.4	■ Minor updates to tables.	—

2. MAX II Architecture

MII51002-2.2

Introduction

This chapter describes the architecture of the MAX II device and contains the following sections:

Functional Description

MAX® II devices contain a two-dimensional row- and column-based architecture to implement custom logic. Row and column interconnects provide signal interconnects between the logic array blocks (LABs).

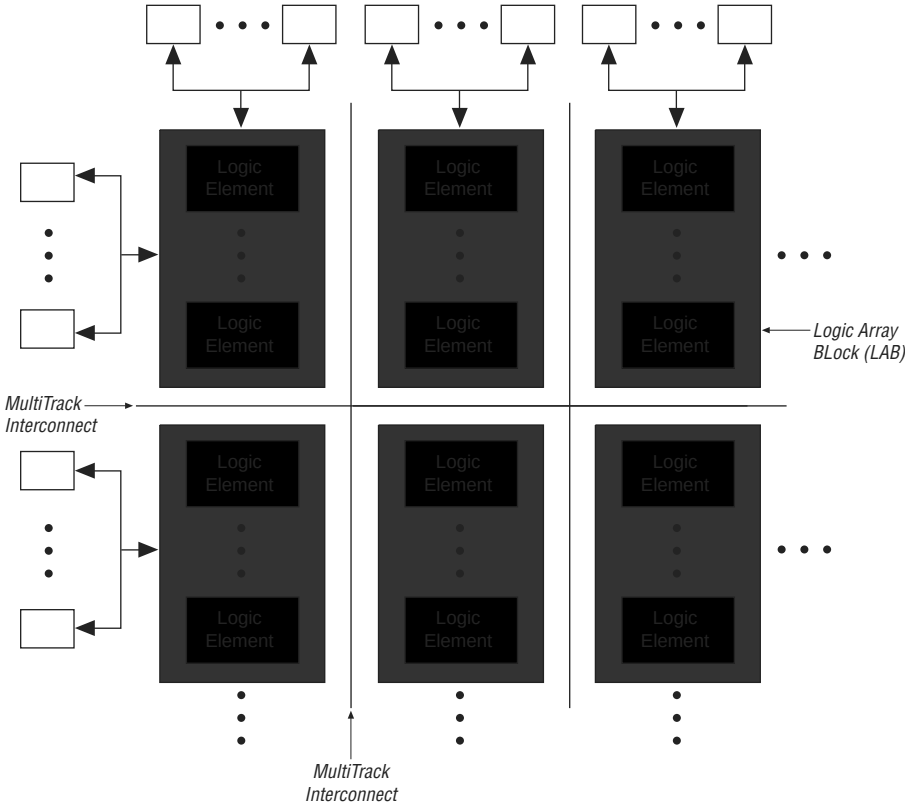
The logic array consists of LABs, with 10 logic elements (LEs) in each LAB. An LE is a small unit of logic providing efficient implementation of user logic functions. LABs are grouped into rows and columns across the device. The MultiTrack interconnect provides fast granular timing delays between LABs. The fast routing between LEs provides minimum timing delay for added levels of logic versus globally routed interconnect structures.

The MAX II device I/O pins are fed by I/O elements (IOE) located at the ends of LAB rows and columns around the periphery of the device. Each IOE contains a bidirectional I/O buffer with several advanced features. I/O pins support Schmitt trigger inputs and various single-ended standards, such as 66-MHz, 32-bit PCI, and LVTTL.

MAX II devices provide a global clock network. The global clock network consists of four global clock lines that drive throughout the entire device, providing clocks for all resources within the device. The global clock lines can also be used for control signals such as clear, preset, or output enable.

Figure 2–1 shows a functional block diagram of the MAX II device.

Figure 2–1. MAX II Device Block Diagram



Each MAX II device contains a flash memory block within its floorplan. On the EPM240 device, this block is located on the left side of the device. On the EPM570, EPM1270, and EPM2210 devices, the flash memory block is located on the bottom-left area of the device. The majority of this flash memory storage is partitioned as the dedicated configuration flash memory (CFM) block. The CFM block provides the non-volatile storage for all of the SRAM configuration information. The CFM automatically downloads and configures the logic and I/O at power-up, providing instant-on operation.

For more information about configuration upon power-up, refer to the *Hot Socketing and Power-On Reset in MAX II Devices* chapter in the *MAX II Device Handbook*.

A portion of the flash memory within the MAX II device is partitioned into a small block for user data. This user flash memory (UFM) block provides 8,192 bits of general-purpose user storage. The UFM provides programmable port connections to the logic array for reading and writing. There are three LAB rows adjacent to this block, with column numbers varying by device.

Table 2–1 shows the number of LAB rows and columns in each device, as well as the number of LAB rows and columns adjacent to the flash memory area in the EPM570, EPM1270, and EPM2210 devices. The long LAB rows are full LAB rows that extend from one side of row I/O blocks to the other. The short LAB rows are adjacent to the UFM block; their length is shown as width in LAB columns.

Table 2–1. MAX II Device Resources

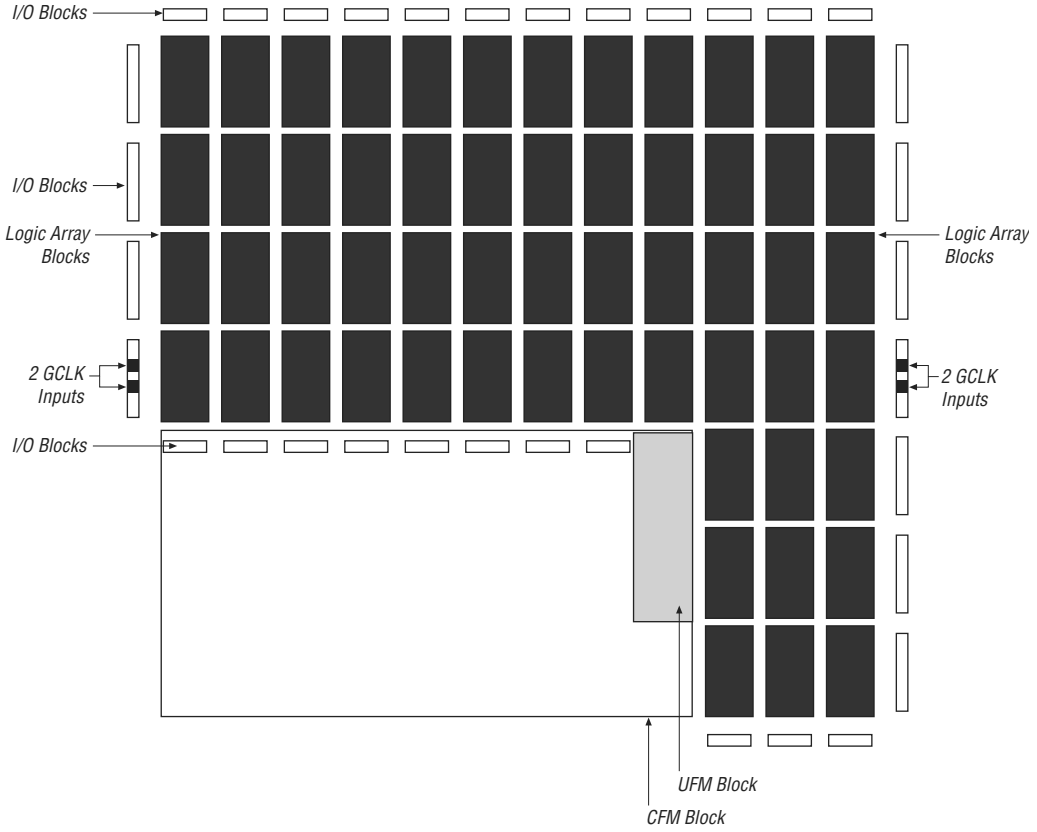
Devices	UFM Blocks	LAB Columns	LAB Rows		Total LABs
			Long LAB Rows	Short LAB Rows (Width) *(1)*	
EPM240	1	6	4	—	24
EPM570	1	12	4	3 (3)	57
EPM1270	1	16	7	3 (5)	127
EPM2210	1	20	10	3 (7)	221

Note to Table 2–1:

(1) The width is the number of LAB columns in length.

Figure 2–2 shows a floorplan of a MAX II device.

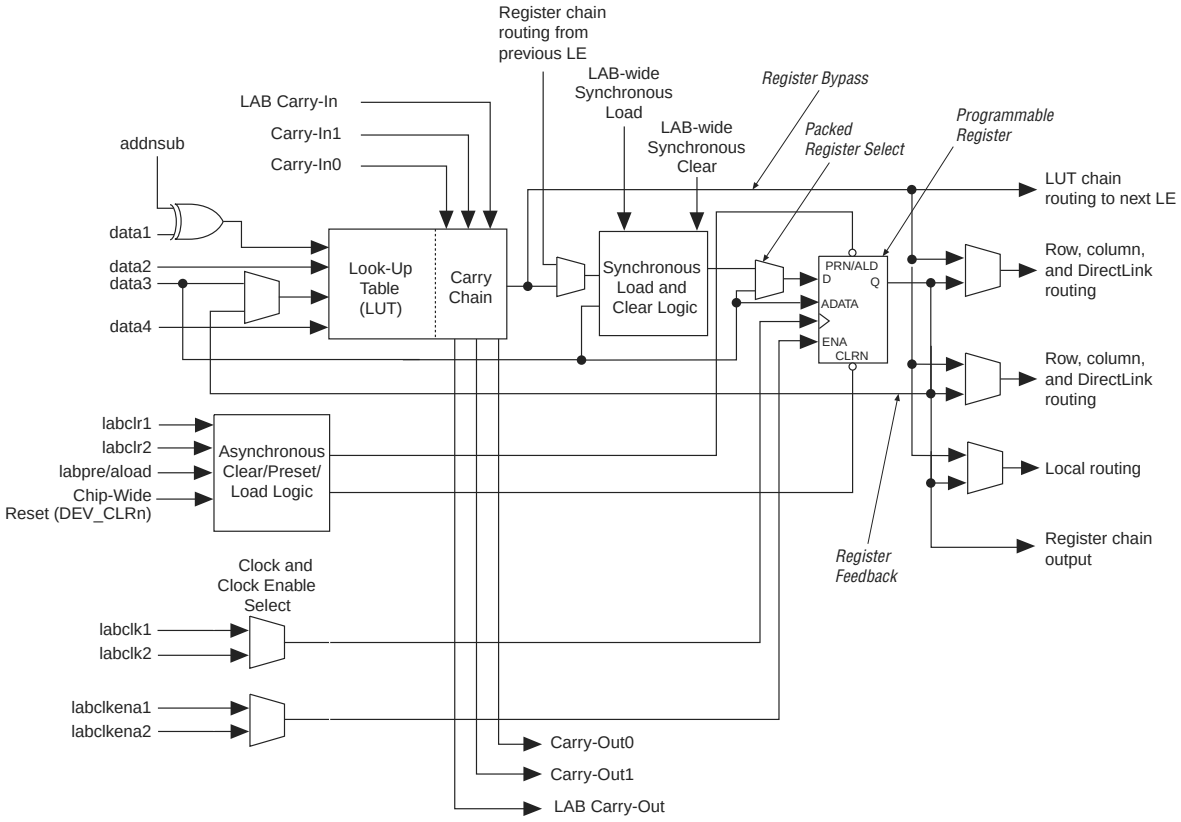
Figure 2–2. MAX II Device Floorplan *(Note 1)*



Note to Figure 2–2:

(1) The device shown is an EPM570 device. EPM1270 and EPM2210 devices have a similar floorplan with more LABs. For EPM240 devices, the CFM and UFM blocks are located on the left side of the device.

Figure 2–6. MAX II LE



Each LE's programmable register can be configured for D, T, JK, or SR operation. Each register has data, true asynchronous load data, clock, clock enable, clear, and asynchronous load/preset inputs. Global signals, general-purpose I/O pins, or any LE can drive the register's clock and clear control signals. Either general-purpose I/O pins or LEs can drive the clock enable, preset, asynchronous load, and asynchronous data. The asynchronous load data input comes from the data3 input of the LE. For combinational functions, the LUT output bypasses the register and drives directly to the LE outputs.

Each LE has three outputs that drive the local, row, and column routing resources. The LUT or register output can drive these three outputs independently. Two LE outputs drive column or row and DirectLink routing connections and one drives local interconnect resources. This allows the LUT to drive one output while the register drives another output. This register packing feature improves device utilization because the device can use the register and the LUT for unrelated functions. Another special packing mode allows the register output to feed back into the LUT of the same LE so that the register is packed with its own fan-out LUT. This provides another mechanism for improved fitting. The LE can also drive out registered and unregistered versions of the LUT output.

LUT Chain and Register Chain

In addition to the three general routing outputs, the LEs within an LAB have LUT chain and register chain outputs. LUT chain connections allow LUTs within the same LAB to cascade together for wide input functions. Register chain outputs allow registers within the same LAB to cascade together. The register chain output allows an LAB to use LUTs for a single combinational function and the registers to be used for an unrelated shift register implementation. These resources speed up connections between LABs while saving local interconnect resources. Refer to "MultiTrack Interconnect" on page 2–12 for more information about LUT chain and register chain connections.

addnsub Signal

The LE's dynamic adder/subtractor feature saves logic resources by using one set of LEs to implement both an adder and a subtractor. This feature is controlled by the LAB-wide control signal addnsub. The addnsub signal sets the LAB to perform either A + B or A – B. The LUT computes addition; subtraction is computed by adding the two's complement of the intended subtractor. The LAB-wide signal converts to two's complement by inverting the B bits within the LAB and setting carry-in to 1, which adds one to the least significant bit (LSB). The LSB of an adder/subtractor must be placed in the first LE of the LAB, where the LAB-wide addnsub signal automatically sets the carry-in to 1. The Quartus II Compiler automatically places and uses the adder/subtractor feature when using adder/subtractor parameterized functions.

LE Operating Modes

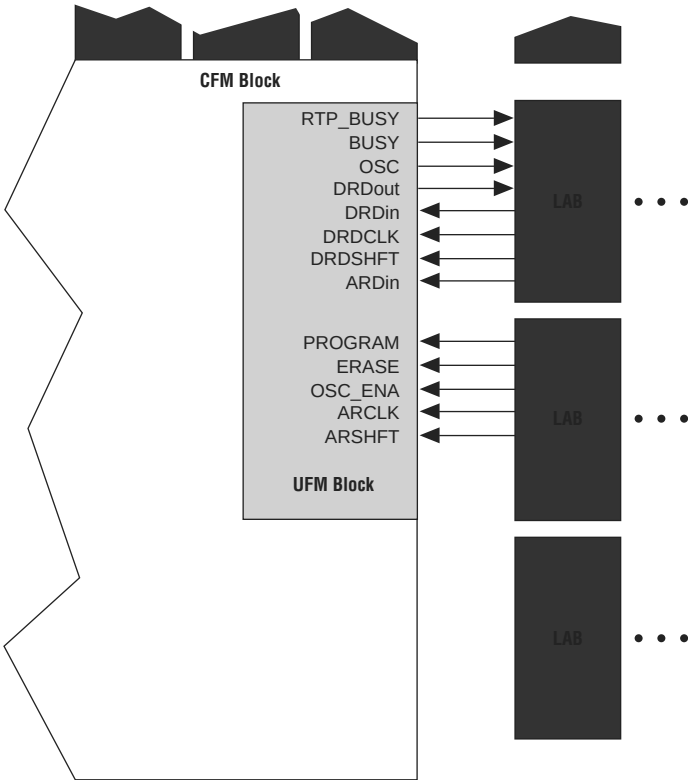
The MAX II LE can operate in one of the following modes:

- "Normal Mode"
- "Dynamic Arithmetic Mode"

Each mode uses LE resources differently. In each mode, eight available inputs to the LE, the four data inputs from the LAB local interconnect, `carry-in0` and `carry-in1` from the previous LE, the LAB carry-in from the previous carry-chain LAB, and the register chain connection are directed to different destinations to implement the desired logic function. LAB-wide signals provide clock, asynchronous clear, asynchronous preset/load, synchronous clear, synchronous load, and clock enable control for the register. These LAB-wide signals are available in all LE modes. The addnsub control signal is allowed in arithmetic mode.

The Quartus II software, in conjunction with parameterized functions such as library of parameterized modules (LPM) functions, automatically chooses the appropriate mode for common functions such as counters, adders, subtractors, and arithmetic functions.

Figure 2–17. EPM570, EPM1270, and EPM2210 UFM Block LAB Row Interface

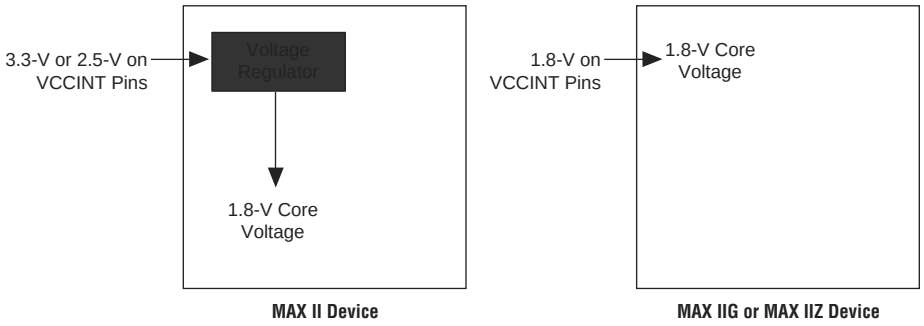


MultiVolt Core

The MAX II architecture supports the MultiVolt core feature, which allows MAX II devices to support multiple V_{CC} levels on the V_{CCINT} supply. An internal linear voltage regulator provides the necessary 1.8-V internal voltage supply to the device. The voltage regulator supports 3.3-V or 2.5-V supplies on its inputs to supply the 1.8-V internal voltage to the device, as shown in Figure 2–18. The voltage regulator is not guaranteed for voltages that are between the maximum recommended 2.5-V operating voltage and the minimum recommended 3.3-V operating voltage.

The MAX IIG and MAX IIZ devices use external 1.8-V supply. The 1.8-V V_{CC} external supply powers the device core directly.

Figure 2–18. MultiVolt Core Feature in MAX II Devices



I/O Structure

IOEs support many features, including:

- LVTTL and LVCMOS I/O standards
- 3.3-V, 32-bit, 66-MHz PCI compliance
- Joint Test Action Group (JTAG) boundary-scan test (BST) support
- Programmable drive strength control
- Weak pull-up resistors during power-up and in system programming
- Slew-rate control
- Tri-state buffers with individual output enable control
- Bus-hold circuitry
- Programmable pull-up resistors in user mode
- Unique output enable per pin
- Open-drain outputs
- Schmitt trigger inputs
- Fast I/O connection
- Programmable input delay

MAX II device IOEs contain a bidirectional I/O buffer. Figure 2–19 shows the MAX II IOE structure. Registers from adjacent LABs can drive to or be driven from the IOE's bidirectional I/O buffers. The Quartus II software automatically attempts to place registers in the adjacent LAB with fast I/O connection to achieve the fastest possible clock-to-output and registered output enable timing. For input registers, the Quartus II software automatically routes the register to guarantee zero hold time. You can set timing assignments in the Quartus II software to achieve desired I/O timing.

Fast I/O Connection

A dedicated fast I/O connection from the adjacent LAB to the IOEs within an I/O block provides faster output delays for clock-to-output and t_{PD} propagation delays. This connection exists for data output signals, not output enable signals or input signals. Figure 2–20, Figure 2–21, and Figure 2–22 illustrate the fast I/O connection.

Table 2–5. MAX II Devices and Speed Grades that Support 3.3-V PCI Electrical Specifications and Meet PCI Timing

Device	33-MHz PCI	66-MHz PCI
EPM1270	All Speed Grades	–3 Speed Grade
EPM2210	All Speed Grades	–3 Speed Grade

Schmitt Trigger

The input buffer for each MAX II device I/O pin has an optional Schmitt trigger setting for the 3.3-V and 2.5-V standards. The Schmitt trigger allows input buffers to respond to slow input edge rates with a fast output edge rate. Most importantly, Schmitt triggers provide hysteresis on the input buffer, preventing slow-rising noisy input signals from ringing or oscillating on the input signal driven into the logic array. This provides system noise tolerance on MAX II inputs, but adds a small, nominal input delay.

The JTAG input pins (TMS, TCK, and TDI) have Schmitt trigger buffers that are always enabled.

The TCK input is susceptible to high pulse glitches when the input signal fall time is greater than 200 ns for all I/O standards.

Output Enable Signals

Each MAX II IOE output buffer supports output enable signals for tri-state control. The output enable signal can originate from the GCLK[3..0] global signals or from the MultiTrack interconnect. The MultiTrack interconnect routes output enable signals and allows for a unique output enable for each output or bidirectional pin.

MAX II devices also provide a chip-wide output enable pin (DEV_OE) to control the output enable for every output pin in the design. An option set before compilation in the Quartus II software controls this pin. This chip-wide output enable uses its own routing resources and does not use any of the four global resources. If this option is turned on, all outputs on the chip operate normally when DEV_OE is asserted. When the pin is deasserted, all outputs are tri-stated. If this option is turned off, the DEV_OE pin is disabled when the device operates in user mode and is available as a user I/O pin.

Programmable Drive Strength

The output buffer for each MAX II device I/O pin has two levels of programmable drive strength control for each of the LVTTL and LVCMOS I/O standards. Programmable drive strength provides system noise reduction control for high performance I/O designs. Although a separate slew-rate control feature exists, using the lower drive strength setting provides signal slew-rate control to reduce system noise and signal overshoot without the large delay adder associated with the slew-rate control feature. Table 2–6 shows the possible settings for the I/O standards with drive strength control. The Quartus II software uses the maximum current strength as the default setting. The PCI I/O standard is always set at 20 mA with no alternate setting.

Table 3–1. MAX II JTAG Instructions (Part 2 of 2)

JTAG Instruction	Instruction Code	Description
CLAMP *(1)*	00 0000 1010	Places the 1-bit bypass register between the TDI and TDO pins, which allows the boundary scan test data to pass synchronously through selected devices to adjacent devices during normal device operation, while holding I/O pins to a state defined by the data in the boundary-scan register.
USER0	00 0000 1100	This instruction allows you to define the scan chain between TDI and TDO in the MAX II logic array. This instruction is also used for custom logic and JTAG interfaces.
USER1	00 0000 1110	This instruction allows you to define the scan chain between TDI and TDO in the MAX II logic array. This instruction is also used for custom logic and JTAG interfaces.
IEEE 1532 instructions	*(2)*	IEEE 1532 ISC instructions used when programming a MAX II device via the JTAG port.

Notes to Table 3–1:

(1) HIGHZ, CLAMP, and EXTEST instructions do not disable weak pull-up resistors or bus hold features.

(2) These instructions are shown in the 1532 BSDL files, which will be posted on the Altera® website at www.altera.com when they are available.

Unsupported JTAG instructions should not be issued to the MAX II device as this may put the device into an unknown state, requiring a power cycle to recover device operation.

The MAX II device instruction register length is 10 bits and the USERCODE register length is 32 bits. Table 3–2 and Table 3–3 show the boundary-scan register length and device IDCODE information for MAX II devices.

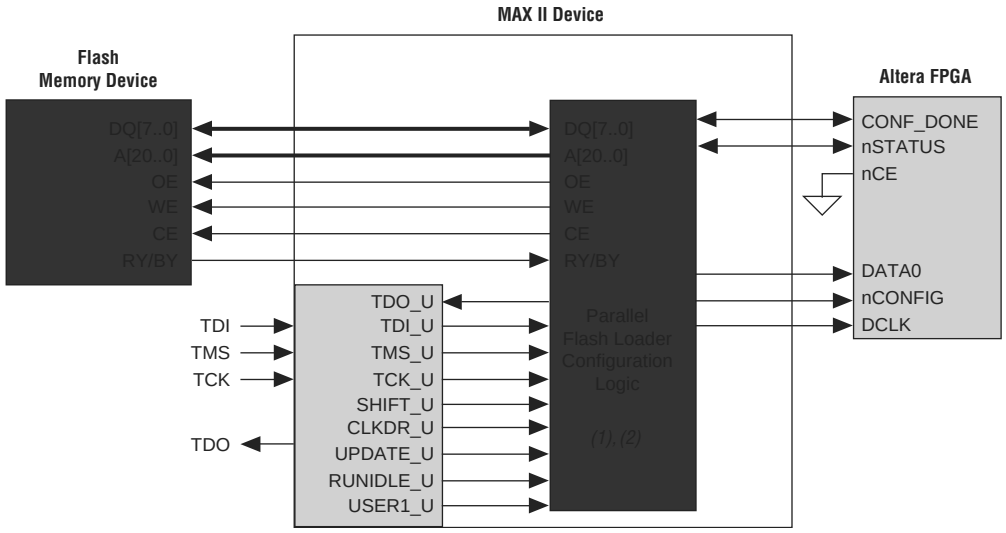
Table 3–2. MAX II Boundary-Scan Register Length

Device	Boundary-Scan Register Length
EPM240	240
EPM570	480
EPM1270	636
EPM2210	816

Table 3–3. 32-Bit MAX II Device IDCODE (Part 1 of 2)

Device	Binary IDCODE (32 Bits) *(1)*				HEX IDCODE
	Version (4 Bits)	Part Number	Manufacturer Identity (11 Bits)	LSB (1 Bit) *(2)*	
EPM240 EPM240G	0000	0010 0000 1010 0001	000 0110 1110	1	0x020A10DD
EPM570 EPM570G	0000	0010 0000 1010 0010	000 0110 1110	1	0x020A20DD
EPM1270 EPM1270G	0000	0010 0000 1010 0011	000 0110 1110	1	0x020A30DD
EPM2210 EPM2210G	0000	0010 0000 1010 0100	000 0110 1110	1	0x020A40DD

Figure 3–1. MAX II Parallel Flash Loader



Notes to Figure 3–1:

(1) This block is implemented in LEs.

(2) This function is supported in the Quartus II software.

In System Programmability

MAX II devices can be programmed in-system via the industry standard 4-pin IEEE Std. 1149.1 (JTAG) interface. In-system programmability (ISP) offers quick, efficient iterations during design development and debugging cycles. The logic, circuitry, and interconnects in the MAX II architecture are configured with flash-based SRAM configuration elements. These SRAM elements require configuration data to be loaded each time the device is powered. The process of loading the SRAM data is called configuration. The on-chip configuration flash memory (CFM) block stores the SRAM element's configuration data. The CFM block stores the design's configuration pattern in a reprogrammable flash array. During ISP, the MAX II JTAG and ISP circuitry programs the design pattern into the CFM block's non-volatile flash array.

The MAX II JTAG and ISP controller internally generate the high programming voltages required to program the CFM cells, allowing in-system programming with any of the recommended operating external voltage supplies (that is, 3.3 V/2.5 V or 1.8 V for the MAX IIG and MAX IIZ devices). ISP can be performed anytime after V_{CCINT} and all V_{CCIO} banks have been fully powered and the device has completed the configuration power-up time. By default, during in-system programming, the I/O pins are tri-stated and weakly pulled-up to V_{CCIO} to eliminate board conflicts. The in-system programming clamp and real-time ISP feature allow user control of I/O state or behavior during ISP.

For more information, refer to "In-System Programming Clamp" on page 3–6 and "Real-Time ISP" on page 3–7.

These devices also offer an `ISP_DONE` bit that provides safe operation when in-system programming is interrupted. This `ISP_DONE` bit, which is the last bit programmed, prevents all I/O pins from driving until the bit is programmed.

Table 3–4 shows the programming times for MAX II devices using in-circuit testers to execute the algorithm vectors in hardware. Software-based programming tools used with download cables are slightly slower because of data processing and transfer limitations.

Table 3–4. MAX II Device Family Programming Times

Description	EPM240 EPM240G EPM240Z	EPM570 EPM570G EPM570Z	EPM1270 EPM1270G	EPM2210 EPM2210G	Unit
Erase + Program (1 MHz)	1.72	2.16	2.90	3.92	sec
Erase + Program (10 MHz)	1.65	1.99	2.58	3.40	sec
Verify (1 MHz)	0.09	0.17	0.30	0.49	sec
Verify (10 MHz)	0.01	0.02	0.03	0.05	sec
Complete Program Cycle (1 MHz)	1.81	2.33	3.20	4.41	sec
Complete Program Cycle (10 MHz)	1.66	2.01	2.61	3.45	sec

UFM Programming

The Quartus II software, with the use of POF, Jam, or JBC files, supports programming of the user flash memory (UFM) block independent of the logic array design pattern stored in the CFM block. This allows updating or reading UFM contents through ISP without altering the current logic array design, or vice versa. By default, these programming files and methods will program the entire flash memory contents, which includes the CFM block and UFM contents. The stand-alone embedded Jam STAPL player and Jam Byte-Code Player provides action commands for programming or reading the entire flash memory (UFM and CFM together) or each independently.

For more information, refer to the *Using Jam STAPL for ISP via an Embedded Processor* chapter in the *MAX II Device Handbook*.

In-System Programming Clamp

By default, the IEEE 1532 instruction used for entering ISP automatically tri-states all I/O pins with weak pull-up resistors for the duration of the ISP sequence. However, some systems may require certain pins on MAX II devices to maintain a specific DC logic level during an in-field update. For these systems, an optional in-system programming clamp instruction exists in MAX II circuitry to control I/O behavior during the ISP sequence. The in-system programming clamp instruction enables the device to sample and sustain the value on an output pin (an input pin would remain tri-stated if sampled) or to explicitly set a logic high, logic low, or tri-state value on any pin. Setting these options is controlled on an individual pin basis using the Quartus II software.

For more information, refer to the *Real-Time ISP and ISP Clamp for MAX II Devices* chapter in the *MAX II Device Handbook*.

Document Revision History

Table 3–5 shows the revision history for this chapter.

Table 3–5. Document Revision History

Date and Revision	Changes Made	Summary of Changes
October 2008, version 1.6	■ Updated New Document Format.	—
December 2007, version 1.5	■ Added warning note after Table 3–1. ■ Updated Table 3–3 and Table 3–4. ■ Added "Referenced Documents" section.	—
December 2006, version 1.4	■ Added document revision history.	—
June 2005, version 1.3	■ Added text and Table 3-4.	—
June 2005, version 1.3	■ Updated text on pages 3-5 to 3-8.	—
June 2004, version 1.1	■ Corrected Figure 3-1. Added CFM acronym.	—

I/O Pins Remain Tri-Stated during Power-Up

A device that does not support hot-socketing may interrupt system operation or cause contention by driving out before or during power-up. In a hot socketing situation, the MAX II device's output buffers are turned off during system power-up. MAX II devices do not drive out until the device attains proper operating conditions and is fully configured. Refer to "Power-On Reset Circuitry" on page 4–5 for information about turn-on voltages.

Signal Pins Do Not Drive the V_{CCIO} or V_{CCINT} Power Supplies

MAX II devices do not have a current path from I/O pins or GCLK[3..0] pins to the V_{CCIO} or V_{CCINT} pins before or during power-up. A MAX II device may be inserted into (or removed from) a system board that was powered up without damaging or interfering with system-board operation. When hot socketing, MAX II devices may have a minimal effect on the signal integrity of the backplane.

AC and DC Specifications

You can power up or power down the V_{CCIO} and V_{CCINT} pins in any sequence. During hot socketing, the I/O pin capacitance is less than 8 pF. MAX II devices meet the following hot socketing specifications:

- The hot socketing DC specification is: $| I_{IOPIN} | < 300$ µA.
- The hot socketing AC specification is: $| I_{IOPIN} | < 8$ mA for 10 ns or less.

MAX II devices are immune to latch-up when hot socketing. If the TCK JTAG input pin is driven high during hot socketing, the current on that pin might exceed the specifications above.

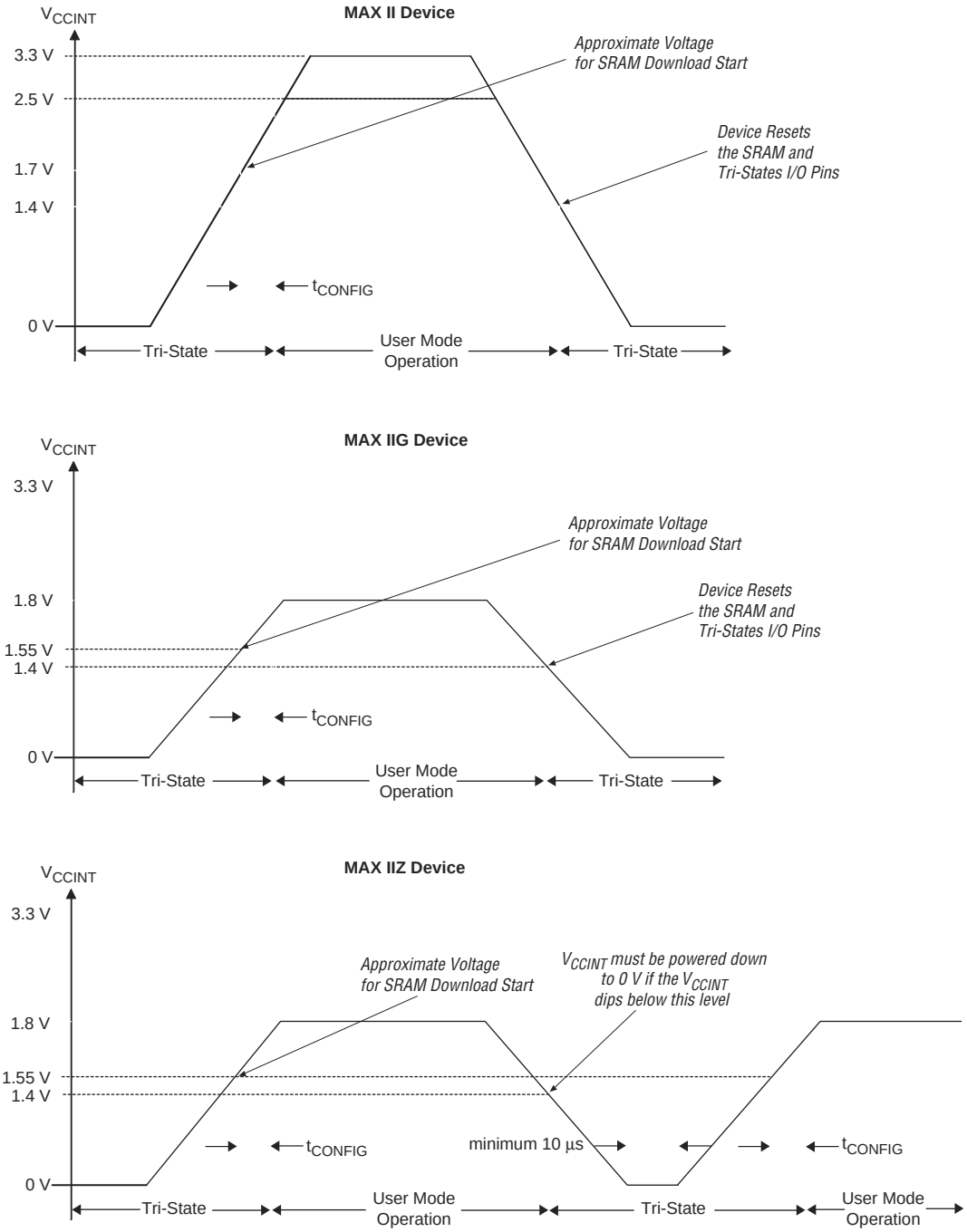
I_{IOPIN} is the current at any user I/O pin on the device. The AC specification applies when the device is being powered up or powered down. This specification takes into account the pin capacitance but not board trace and external loading capacitance. Additional capacitance for trace, connector, and loading must be taken into consideration separately. The peak current duration due to power-up transients is 10 ns or less.

The DC specification applies when all V_{CC} supplies to the device are stable in the powered-up or powered-down conditions.

Hot Socketing Feature Implementation in MAX II Devices

The hot socketing feature turns off (tri-states) the output buffer during the power-up event (either V_{CCINT} or V_{CCIO} supplies) or power-down event. The hot-socket circuit generates an internal HOTSCKT signal when either V_{CCINT} or V_{CCIO} is below the threshold voltage during power-up or power-down. The HOTSCKT signal cuts off the output buffer to make sure that no DC current (except for weak pull-up leaking) leaks through the pin. When V_{CC} ramps up very slowly during power-up, V_{CC} may still be relatively low even after the power-on reset (POR) signal is released and device configuration is complete.

Figure 4–5. Power-Up Characteristics for MAX II, MAX IIG, and MAX IIZ Devices *(Note 1), (2)*



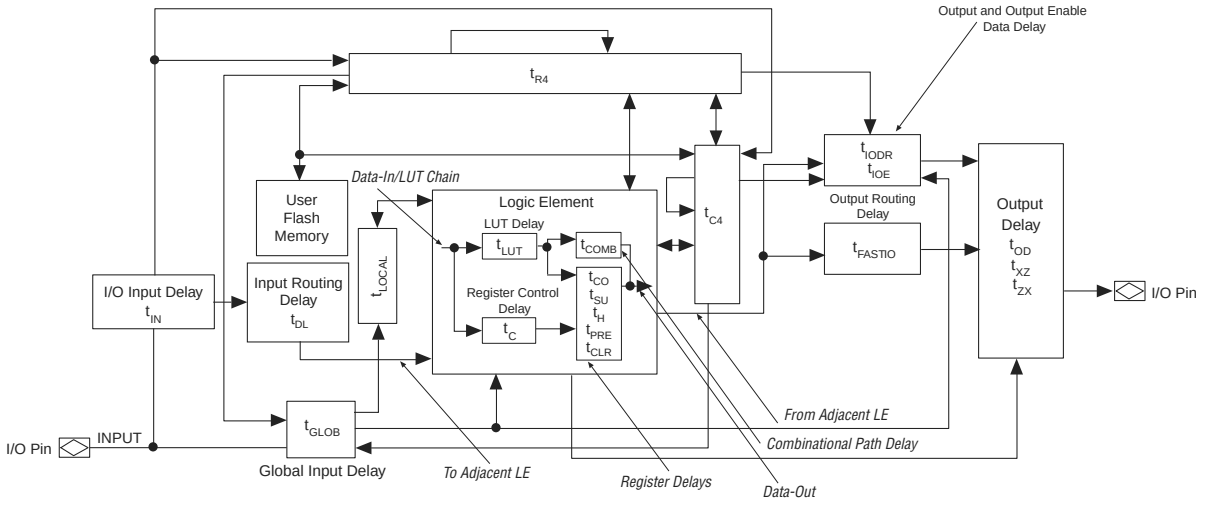
Notes to Figure 4–5:

(1) Time scale is relative.

(2) Figure 4–5 assumes all V_{CCIO} banks power up simultaneously with the V_{CCINT} profile shown. If not, t_{CONFIG} stretches out until all V_{CCIO} banks are powered.

☞ After SRAM configuration, all registers in the device are cleared and released into user function before I/O tri-states are released. To release clears after tri-states are released, use the DEV_CLRn pin option. To hold the tri-states beyond the power-up configuration time, use the DEV_OE pin option.

Figure 5–2. MAX II Device Timing Model



The timing characteristics of any signal path can be derived from the timing model and parameters of a particular device. External timing parameters, which represent pin-to-pin timing delays, can be calculated as the sum of internal parameters.

Refer to the *Understanding Timing in MAX II Devices* chapter in the *MAX II Device Handbook* for more information.

This section describes and specifies the performance, internal, external, and UFM timing specifications. All specifications are representative of the worst-case supply voltage and junction temperature conditions.

Preliminary and Final Timing

Timing models can have either preliminary or final status. The Quartus II software issues an informational message during the design compilation if the timing models are preliminary. Table 5–13 shows the status of the MAX II device timing models.

Preliminary status means the timing model is subject to change. Initially, timing numbers are created using simulation results, process data, and other known parameters. These tests are used to make the preliminary numbers as close to the actual timing parameters as possible.

Final timing numbers are based on actual device operation and testing. These numbers reflect the actual performance of the device under the worst-case voltage and junction temperature conditions.

Table 5–13. MAX II Device Timing Model Status (Part 1 of 2)

Device	Preliminary	Final
EPM240	—	✓
EPM240Z *(1)*	—	✓
EPM570	—	✓
EPM570Z *(1)*	—	✓

Internal Timing Parameters

Internal timing parameters are specified on a speed grade basis independent of device density. Table 5–15 through Table 5–22 describe the MAX II device internal timing microparameters for logic elements (LEs), input/output elements (IOEs), UFM blocks, and MultiTrack interconnects. The timing values for –3, –4, and –5 speed grades shown in Table 5–15 through Table 5–22 are based on an EPM1270 device target, while –6, –7, and –8 speed grade values are based on an EPM570Z device target.

For more explanations and descriptions about each internal timing microparameters symbol, refer to the *Understanding Timing in MAX II Devices* chapter in the *MAX II Device Handbook*.

Table 5–15. LE Internal Timing Microparameters

		MAX II / MAX IIG						MAX IIZ						
		–3 Speed Grade		–4 Speed Grade		–5 Speed Grade		–6 Speed Grade		–7 Speed Grade		–8 Speed Grade		
Symbol	Parameter	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Unit
t_{LUT}	LE combinational LUT delay	—	571	—	742	—	914	—	1,215	—	2,247	—	2,247	ps
t_{COMB}	Combinational path delay	—	147	—	192	—	236	—	243	—	305	—	309	ps
t_{CLR}	LE register clear delay	238	—	309	—	381	—	401	—	541	—	545	—	ps
t_{PRE}	LE register preset delay	238	—	309	—	381	—	401	—	541	—	545	—	ps
t_{SU}	LE register setup time before clock	208	—	271	—	333	—	260	—	319	—	321	—	ps
t_{H}	LE register hold time after clock	0	—	0	—	0	—	0	—	0	—	0	—	ps
t_{CO}	LE register clock-to-output delay	—	235	—	305	—	376	—	380	—	489	—	494	ps
t_{CLKHL}	Minimum clock high or low time	166	—	216	—	266	—	253	—	335	—	339	—	ps
t_{C}	Register control delay	—	857	—	1,114	—	1,372	—	1,356	—	1,722	—	1,741	ps

Table 5–17. t_{ZX} IOE Microparameter Adders for Fast Slew Rate (Part 2 of 2)

Standard		MAX II / MAX IIG						MAX IIZ						Unit
		–3 Speed Grade		–4 Speed Grade		–5 Speed Grade		–6 Speed Grade		–7 Speed Grade		–8 Speed Grade		
		Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	
1.5-V LVCMOS	4 mA	—	1,118	—	1,454	—	1,789	—	580	—	588	—	588	ps
	2 mA	—	2,410	—	3,133	—	3,856	—	915	—	923	—	923	ps
3.3-V PCI	20 mA	—	19	—	25	—	31	—	72	—	71	—	74	ps

Table 5–18. t_{ZX} IOE Microparameter Adders for Slow Slew Rate

Standard		MAX II / MAX IIG						MAX IIZ						Unit
		–3 Speed Grade		–4 Speed Grade		–5 Speed Grade		–6 Speed Grade		–7 Speed Grade		–8 Speed Grade		
		Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	
3.3-V LVTTL	16 mA	—	6,350	—	6,050	—	5,749	—	5,951	—	5,952	—	6,063	ps
	8 mA	—	9,383	—	9,083	—	8,782	—	6,534	—	6,533	—	6,662	ps
3.3-V LVCMOS	8 mA	—	6,350	—	6,050	—	5,749	—	5,951	—	5,952	—	6,063	ps
	4 mA	—	9,383	—	9,083	—	8,782	—	6,534	—	6,533	—	6,662	ps
2.5-V LVTTL / LVCMOS	14 mA	—	10,412	—	10,112	—	9,811	—	9,110	—	9,105	—	9,237	ps
	7 mA	—	13,613	—	13,313	—	13,012	—	9,830	—	9,835	—	9,977	ps
3.3-V PCI	20 mA	—	–75	—	–97	—	–120	—	6,534	—	6,533	—	6,662	ps

Table 5–19. t_{XZ} IOE Microparameter Adders for Fast Slew Rate

Standard		MAX II / MAX IIG						MAX IIZ						Unit
		–3 Speed Grade		–4 Speed Grade		–5 Speed Grade		–6 Speed Grade		–7 Speed Grade		–8 Speed Grade		
		Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	
3.3-V LVTTL	16 mA	—	0	—	0	—	0	—	0	—	0	—	0	ps
	8 mA	—	–56	—	–72	—	–89	—	–69	—	–69	—	–69	ps
3.3-V LVCMOS	8 mA	—	0	—	0	—	0	—	0	—	0	—	0	ps
	4 mA	—	–56	—	–72	—	–89	—	–69	—	–69	—	–69	ps
2.5-V LVTTL / LVCMOS	14 mA	—	–3	—	–4	—	–5	—	–7	—	–11	—	–11	ps
	7 mA	—	–47	—	–61	—	–75	—	–66	—	–70	—	–70	ps
1.8-V LVTTL / LVCMOS	6 mA	—	119	—	155	—	191	—	45	—	34	—	37	ps
	3 mA	—	207	—	269	—	331	—	34	—	22	—	25	ps
1.5-V LVCMOS	4 mA	—	606	—	788	—	970	—	166	—	154	—	155	ps
	2 mA	—	673	—	875	—	1,077	—	190	—	177	—	179	ps
3.3-V PCI	20 mA	—	71	—	93	—	114	—	–69	—	–69	—	–69	ps

Table 5–23. EPM240 Global Clock External I/O Timing Parameters (Part 2 of 2)

Symbol	Parameter	Condition	MAX II / MAX IIG						MAX IIZ						Unit
			–3 Speed Grade		–4 Speed Grade		–5 Speed Grade		–6 Speed Grade		–7 Speed Grade		–8 Speed Grade		
			Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	
f_{CNT}	Maximum global clock frequency for 16-bit counter	—	—	304.0 *(1)*	—	247.5	—	201.1	—	184.1	—	123.5	—	118.3	MHz

Note to Table 5–23:

(1) The maximum frequency is limited by the I/O standard on the clock input pin. The 16-bit counter critical delay performs faster than this global clock input pin maximum frequency.

Table 5–24 shows the external I/O timing parameters for EPM570 devices.

Table 5–24. EPM570 Global Clock External I/O Timing Parameters (Part 1 of 2)

Symbol	Parameter	Condition	MAX II / MAX IIG						MAX IIZ						Unit
			–3 Speed Grade		–4 Speed Grade		–5 Speed Grade		–6 Speed Grade		–7 Speed Grade		–8 Speed Grade		
			Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	
t_{PD1}	Worst case pin-to-pin delay through 1 look-up table (LUT)	10 pF	—	5.4	—	7.0	—	8.7	—	9.5	—	15.1	—	17.7	ns
t_{PD2}	Best case pin-to-pin delay through 1 LUT	10 pF	—	3.7	—	4.8	—	5.9	—	5.7	—	7.7	—	8.5	ns
t_{SU}	Global clock setup time	—	1.2	—	1.5	—	1.9	—	2.2	—	3.9	—	4.4	—	ns
t_{H}	Global clock hold time	—	0	—	0	—	0	—	0	—	0	—	0	—	ns
t_{CO}	Global clock to output delay	10 pF	2.0	4.5	2.0	5.8	2.0	7.1	2.0	6.7	2.0	8.2	2.0	8.7	ns
t_{CH}	Global clock high time	—	166	—	216	—	266	—	253	—	335	—	339	—	ps
t_{CL}	Global clock low time	—	166	—	216	—	266	—	253	—	335	—	339	—	ps
t_{CNT}	Minimum global clock period for 16-bit counter	—	3.3	—	4.0	—	5.0	—	5.4	—	8.1	—	8.4	—	ns