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Understanding Embedded - CPLDs (Complex Programmable Logic Devices)

Embedded - CPLDs, or Complex Programmable Logic Devices, are highly versatile digital logic devices used in electronic systems. These programmable components are designed to perform complex logical operations and can be customized for specific applications. Unlike fixed-function ICs, CPLDs offer the flexibility to reprogram their configuration, making them an ideal choice for various embedded systems. They consist of a set of logic gates and programmable interconnects, allowing designers to implement complex logic circuits without needing custom hardware.

Applications of Embedded - CPLDs

Details	
Product Status	Active
Programmable Type	In System Programmable
Delay Time tpd(1) Max	5.4 ns
Voltage Supply - Internal	2.5V, 3.3V
Number of Logic Elements/Blocks	570
Number of Macrocells	440
Number of Gates	-
Number of I/O	116
Operating Temperature	0°C ~ 85°C (TJ)
Mounting Type	Surface Mount
Package / Case	144-LQFP
Supplier Device Package	144-TQFP (20x20)
Purchase URL	https://www.e-xfl.com/product-detail/intel/epm570t144c5

Table 1–1 shows the MAX II family features.

Table 1–1. MAX II Family Features

Feature	EPM240 EPM240G	EPM570 EPM570G	EPM1270 EPM1270G	EPM2210 EPM2210G	EPM240Z	EPM570Z
LEs	240	570	1,270	2,210	240	570
Typical Equivalent Macrocells	192	440	980	1,700	192	440
Equivalent Macrocell Range	128 to 240	240 to 570	570 to 1,270	1,270 to 2,210	128 to 240	240 to 570
UFM Size (bits)	8,192	8,192	8,192	8,192	8,192	8,192
Maximum User I/O pins	80	160	212	272	80	160
t_{PD1} (ns) *(1)*	4.7	5.4	6.2	7.0	7.5	9.0
f_{CNT} (MHz) *(2)*	304	304	304	304	152	152
t_{SU} (ns)	1.7	1.2	1.2	1.2	2.3	2.2
t_{CO} (ns)	4.3	4.5	4.6	4.6	6.5	6.7

Notes to Table 1–1:

(1) t_{PD1} represents a pin-to-pin delay for the worst case I/O placement with a full diagonal path across the device and combinational logic implemented in a single LUT and LAB that is adjacent to the output pin.

(2) The maximum frequency is limited by the I/O standard on the clock input pin. The 16-bit counter critical delay will run faster than this number.

For more information about equivalent macrocells, refer to the *MAX II Logic Element to Macrocell Conversion Methodology* white paper.

MAX II and MAX IIG devices are available in three speed grades: –3, –4, and –5, with –3 being the fastest. Similarly, MAX IIZ devices are available in three speed grades: –6, –7, and –8, with –6 being the fastest. These speed grades represent the overall relative performance, not any specific timing parameter. For propagation delay timing numbers within each speed grade and density, refer to the *DC and Switching Characteristics* chapter in the *MAX II Device Handbook*.

Table 1–2 shows MAX II device speed-grade offerings.

Table 1–2. MAX II Speed Grades

Device	Speed Grade					
	–3	–4	–5	–6	–7	–8
EPM240 EPM240G	✓	✓	✓	—	—	—
EPM570 EPM570G	✓	✓	✓	—	—	—
EPM1270 EPM1270G	✓	✓	✓	—	—	—
EPM2210 EPM2210G	✓	✓	✓	—	—	—
EPM240Z	—	—	—	✓	✓	✓
EPM570Z	—	—	—	✓	✓	✓

MAX II devices have an internal linear voltage regulator which supports external supply voltages of 3.3 V or 2.5 V, regulating the supply down to the internal operating voltage of 1.8 V. MAX IIG and MAX IIZ devices only accept 1.8 V as the external supply voltage. MAX IIZ devices are pin-compatible with MAX IIG devices in the 100-pin Micro FineLine BGA and 256-pin Micro FineLine BGA packages. Except for external supply voltage requirements, MAX II and MAX II G devices have identical pin-outs and timing specifications. Table 1–5 shows the external supply voltages supported by the MAX II family.

Table 1–5. MAX II External Supply Voltages

Devices	EPM240 EPM570 EPM1270 EPM2210	EPM240G EPM570G EPM1270G EPM2210G EPM240Z EPM570Z *(1)*
MultiVolt core external supply voltage (V_{CCINT}) *(2)*	3.3 V, 2.5 V	1.8 V
MultiVolt I/O interface voltage levels (V_{CCIO})	1.5 V, 1.8 V, 2.5 V, 3.3 V	1.5 V, 1.8 V, 2.5 V, 3.3 V

Notes to Table 1–5:

(1) MAX IIG and MAX IIZ devices only accept 1.8 V on their `VCCINT` pins. The 1.8-V V_{CCINT} external supply powers the device core directly.

(2) MAX II devices operate internally at 1.8 V.

Referenced Documents

This chapter references the following documents:

- *DC and Switching Characteristics* chapter in the *MAX II Device Handbook*
- *MAX II Logic Element to Macrocell Conversion Methodology* white paper

Document Revision History

Table 1–6 shows the revision history for this chapter.

Table 1–6. Document Revision History

Date and Revision	Changes Made	Summary of Changes
August 2009, version 1.9	■ Updated Table 1–2.	Added information for speed grade –8
October 2008, version 1.8	■ Updated "Introduction" section. ■ Updated new Document Format.	—
December 2007, version1.7	■ Updated Table 1–1 through Table 1–5. ■ Added "Referenced Documents" section.	Updated document with MAX IIZ information.
December 2006, version 1.6	■ Added document revision history.	—
August 2006, version 1.5	■ Minor update to features list.	—
July 2006, version 1.4	■ Minor updates to tables.	—

2. MAX II Architecture

MII51002-2.2

Introduction

This chapter describes the architecture of the MAX II device and contains the following sections:

Functional Description

MAX® II devices contain a two-dimensional row- and column-based architecture to implement custom logic. Row and column interconnects provide signal interconnects between the logic array blocks (LABs).

The logic array consists of LABs, with 10 logic elements (LEs) in each LAB. An LE is a small unit of logic providing efficient implementation of user logic functions. LABs are grouped into rows and columns across the device. The MultiTrack interconnect provides fast granular timing delays between LABs. The fast routing between LEs provides minimum timing delay for added levels of logic versus globally routed interconnect structures.

The MAX II device I/O pins are fed by I/O elements (IOE) located at the ends of LAB rows and columns around the periphery of the device. Each IOE contains a bidirectional I/O buffer with several advanced features. I/O pins support Schmitt trigger inputs and various single-ended standards, such as 66-MHz, 32-bit PCI, and LVTTL.

MAX II devices provide a global clock network. The global clock network consists of four global clock lines that drive throughout the entire device, providing clocks for all resources within the device. The global clock lines can also be used for control signals such as clear, preset, or output enable.

Table 2–1. MAX II Device Resources

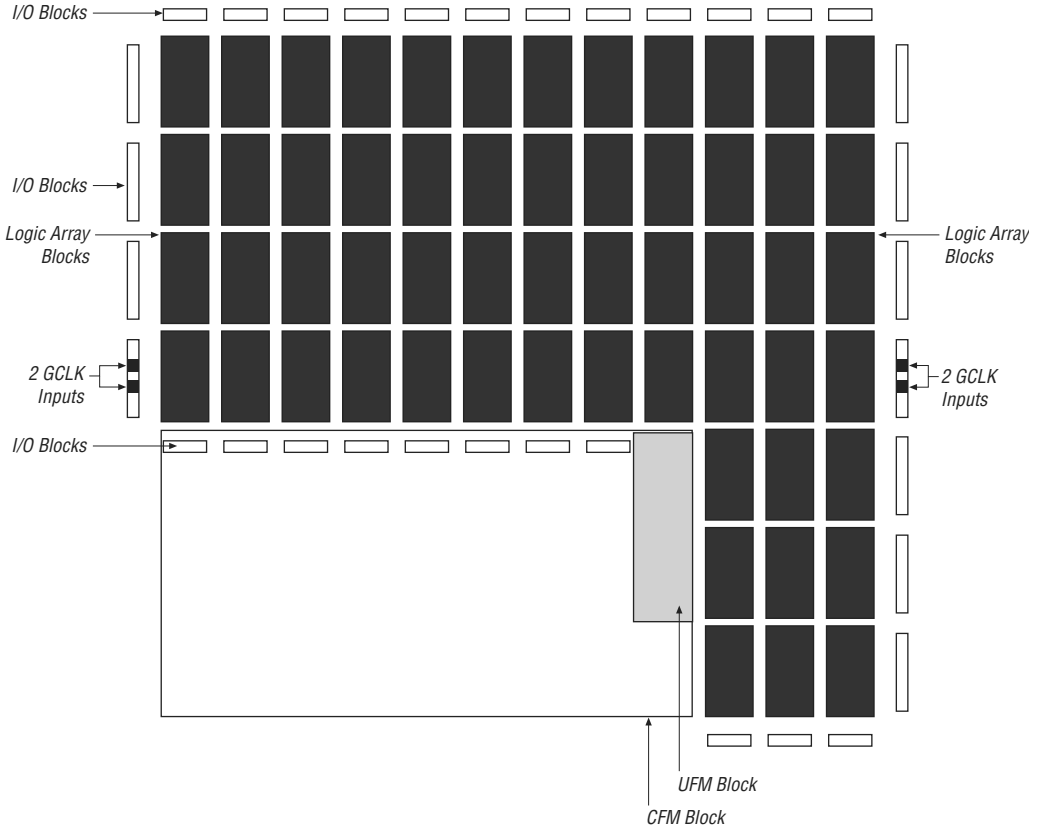
Devices	UFM Blocks	LAB Columns	LAB Rows		Total LABs
			Long LAB Rows	Short LAB Rows (Width) *(1)*	
EPM240	1	6	4	—	24
EPM570	1	12	4	3 (3)	57
EPM1270	1	16	7	3 (5)	127
EPM2210	1	20	10	3 (7)	221

Note to Table 2–1:

(1) The width is the number of LAB columns in length.

Figure 2–2 shows a floorplan of a MAX II device.

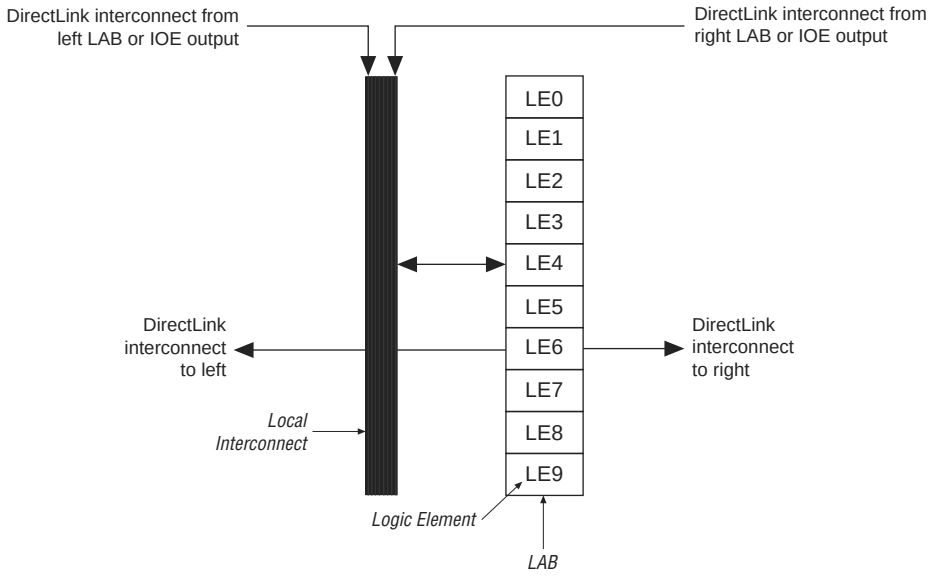
Figure 2–2. MAX II Device Floorplan *(Note 1)*



Note to Figure 2–2:

(1) The device shown is an EPM570 device. EPM1270 and EPM2210 devices have a similar floorplan with more LABs. For EPM240 devices, the CFM and UFM blocks are located on the left side of the device.

Figure 2–4. DirectLink Connection



LAB Control Signals

Each LAB contains dedicated logic for driving control signals to its LEs. The control signals include two clocks, two clock enables, two asynchronous clears, a synchronous clear, an asynchronous preset/load, a synchronous load, and add/subtract control signals, providing a maximum of 10 control signals at a time. Although synchronous load and clear signals are generally used when implementing counters, they can also be used with other functions.

Each LAB can use two clocks and two clock enable signals. Each LAB's clock and clock enable signals are linked. For example, any LE in a particular LAB using the `labclk1` signal also uses `labclkena1`. If the LAB uses both the rising and falling edges of a clock, it also uses both LAB-wide clock signals. Deasserting the clock enable signal turns off the LAB-wide clock.

Each LAB can use two asynchronous clear signals and an asynchronous load/preset signal. By default, the Quartus II software uses a `NOT` gate push-back technique to achieve preset. If you disable the `NOT` gate push-back option or assign a given register to power-up high using the Quartus II software, the preset is then achieved using the asynchronous load signal with asynchronous load data input tied high.

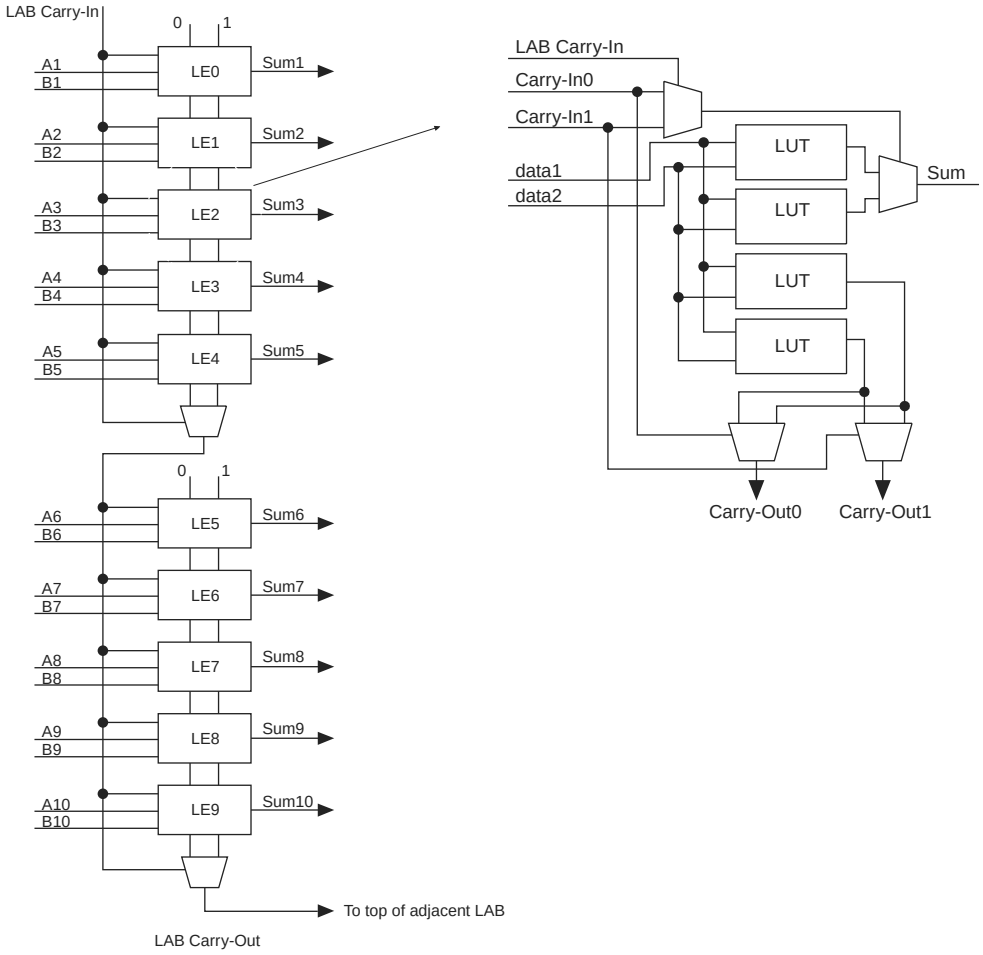
With the LAB-wide addnsub control signal, a single LE can implement a one-bit adder and subtractor. This saves LE resources and improves performance for logic functions such as correlators and signed multipliers that alternate between addition and subtraction depending on data.

The LAB column clocks [3..0], driven by the global clock network, and LAB local interconnect generate the LAB-wide control signals. The MultiTrack interconnect structure drives the LAB local interconnect for non-global control signal generation. The MultiTrack interconnect's inherent low skew allows clock and control signal distribution in addition to data. Figure 2–5 shows the LAB control signal generation circuit.

The speed advantage of the carry-select chain is in the parallel precomputation of carry chains. Since the LAB carry-in selects the precomputed carry chain, not every LE is in the critical path. Only the propagation delays between LAB carry-in generation (LE 5 and LE 10) are now part of the critical path. This feature allows the MAX II architecture to implement high-speed counters, adders, multipliers, parity functions, and comparators of arbitrary width.

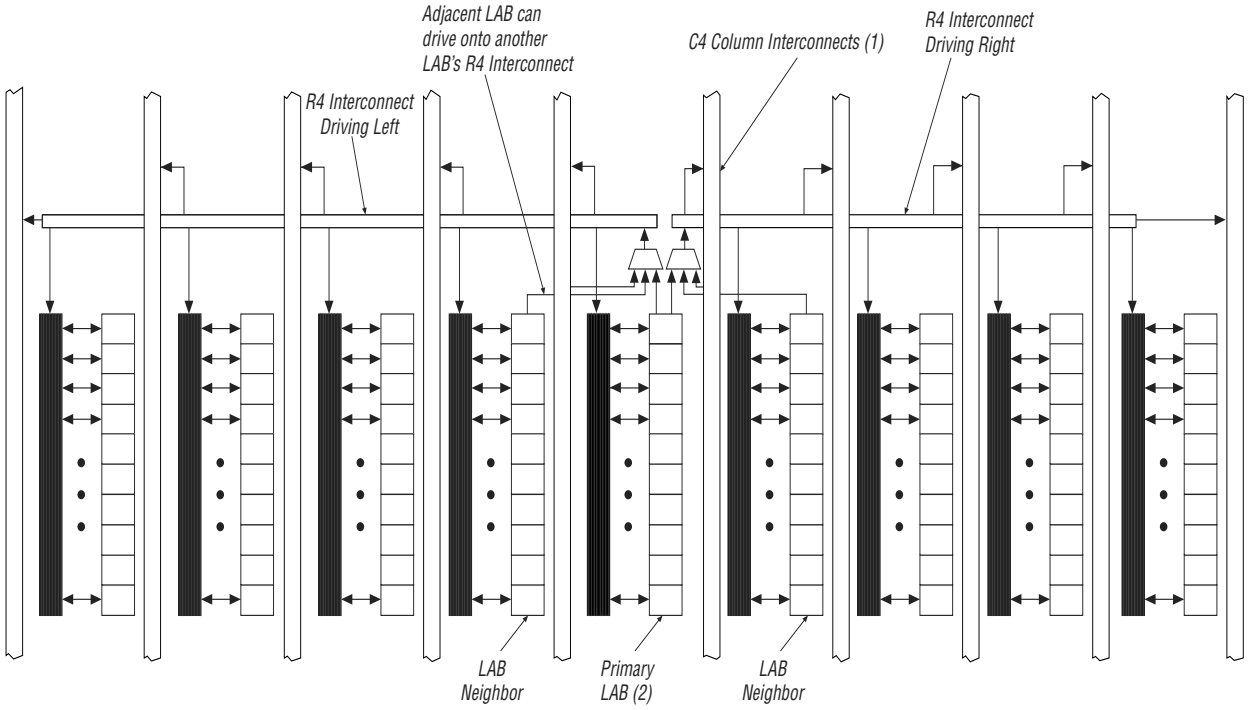
Figure 2–9 shows the carry-select circuitry in an LAB for a 10-bit full adder. One portion of the LUT generates the sum of two bits using the input signals and the appropriate carry-in bit; the sum is routed to the output of the LE. The register can be bypassed for simple adders or used for accumulator functions. Another portion of the LUT generates carry-out bits. An LAB-wide carry-in bit selects which chain is used for the addition of given inputs. The carry-in signal for each chain, `carry-in0` or `carry-in1`, selects the carry-out to carry forward to the carry-in signal of the next-higher-order bit. The final carry-out signal is routed to an LE, where it is fed to local, row, or column interconnects.

Figure 2–9. Carry-Select Chain



The R4 interconnects span four LABs and are used for fast row connections in a four-LAB region. Every LAB has its own set of R4 interconnects to drive either left or right. Figure 2–10 shows R4 interconnect connections from an LAB. R4 interconnects can drive and be driven by row IOEs. For LAB interfacing, a primary LAB or horizontal LAB neighbor can drive a given R4 interconnect. For R4 interconnects that drive to the right, the primary LAB and right neighbor can drive on to the interconnect. For R4 interconnects that drive to the left, the primary LAB and its left neighbor can drive on to the interconnect. R4 interconnects can drive other R4 interconnects to extend the range of LABs they can drive. R4 interconnects can also drive C4 interconnects for connections from one row to another.

Figure 2–10. R4 Interconnect Connections



Notes to Figure 2–10:

(1) C4 interconnects can drive R4 interconnects.

(2) This pattern is repeated for every LAB in the LAB row.

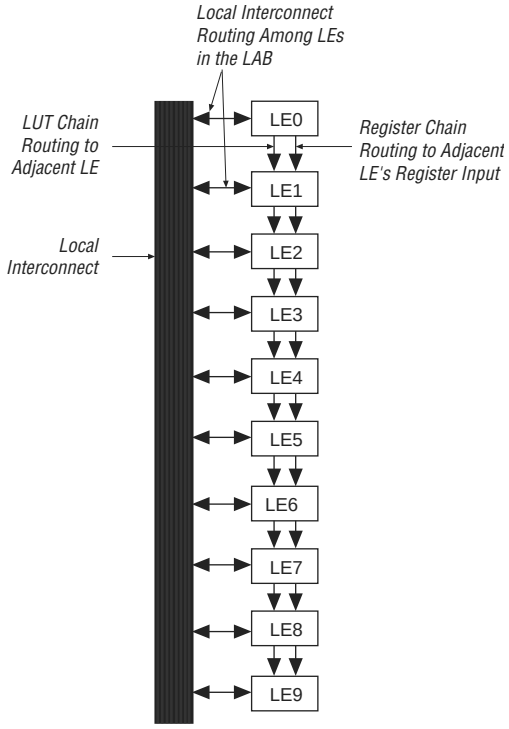
The column interconnect operates similarly to the row interconnect. Each column of LABs is served by a dedicated column interconnect, which vertically routes signals to and from LABs and row and column IOEs. These column resources include:

- LUT chain interconnects within an LAB
- Register chain interconnects within an LAB
- C4 interconnects traversing a distance of four LABs in an up and down direction

MAX II devices include an enhanced interconnect structure within LABs for routing LE output to LE input connections faster using LUT chain connections and register chain connections. The LUT chain connection allows the combinational output of an LE to directly drive the fast input of the LE right below it, bypassing the local interconnect. These resources can be used as a high-speed connection for wide fan-in

functions from LE 1 to LE 10 in the same LAB. The register chain connection allows the register output of one LE to connect directly to the register input of the next LE in the LAB for fast shift registers. The Quartus II Compiler automatically takes advantage of these resources to improve utilization and performance. Figure 2–11 shows the LUT chain and register chain interconnects.

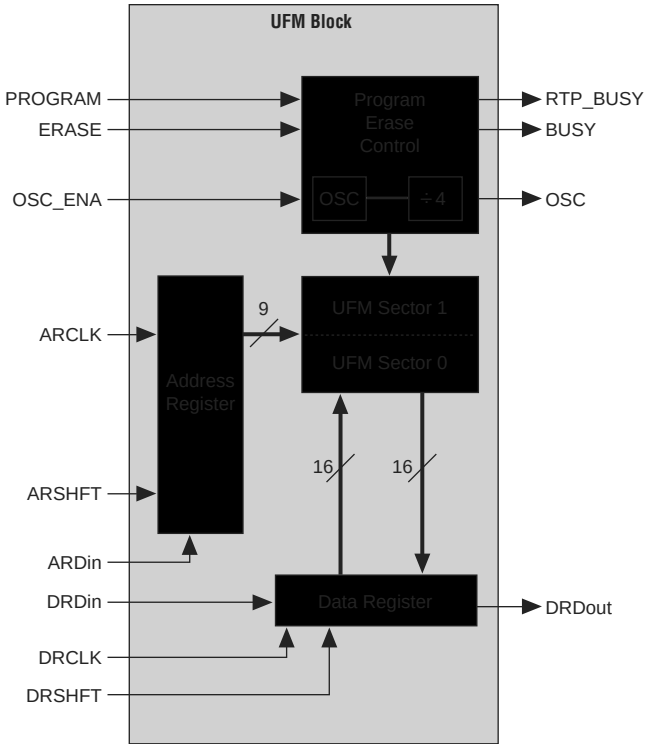
Figure 2–11. LUT Chain and Register Chain Interconnects



The C4 interconnects span four LABs up or down from a source LAB. Every LAB has its own set of C4 interconnects to drive either up or down. Figure 2–12 shows the C4 interconnect connections from an LAB in a column. The C4 interconnects can drive and be driven by column and row IOEs. For LAB interconnection, a primary LAB or its vertical LAB neighbor can drive a given C4 interconnect. C4 interconnects can drive each other to extend their range as well as drive row interconnects for column-to-column connections.

- Auto-increment addressing
- Serial interface to logic array with programmable interface

Figure 2–15. UFM Block and Interface Signals



UFM Storage

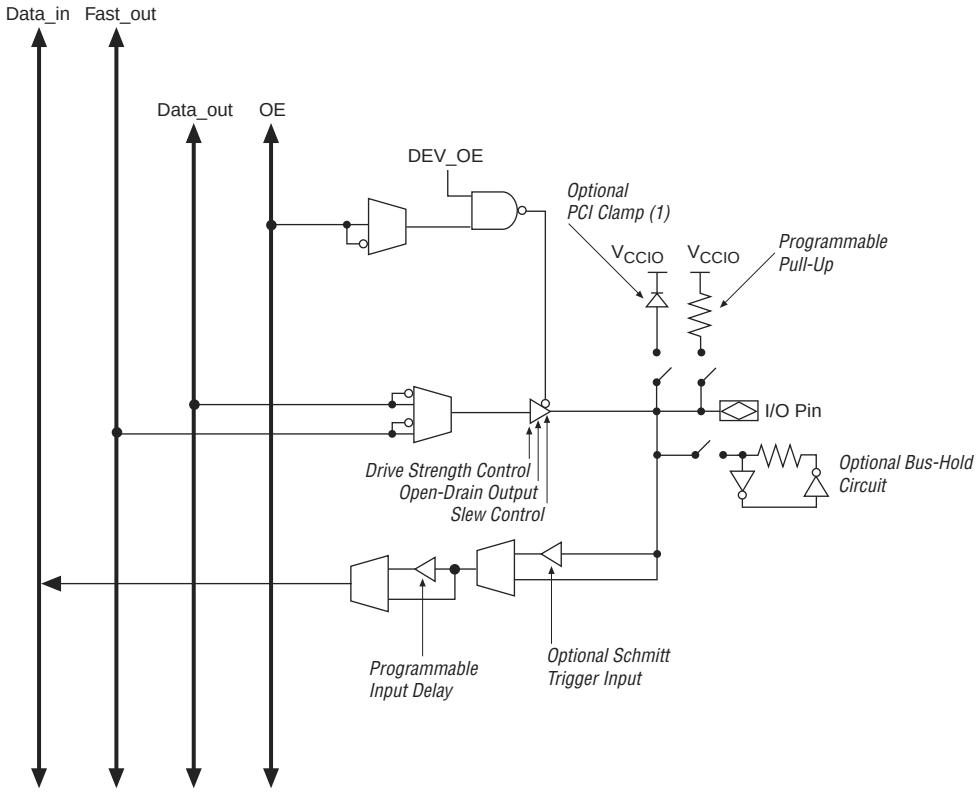
Each device stores up to 8,192 bits of data in the UFM block. Table 2–3 shows the data size, sector, and address sizes for the UFM block.

Table 2–3. UFM Array Size

Device	Total Bits	Sectors	Address Bits	Data Width
EPM240 EPM570 EPM1270 EPM2210	8,192	2 (4,096 bits/sector)	9	16

There are 512 locations with 9-bit addressing ranging from `000h` to `1FFh`. Sector 0 address space is `000h` to `0FFh` and Sector 1 address space is from `100h` to `1FFh`. The data width is up to 16 bits of data. The Quartus II software automatically creates logic to accommodate smaller read or program data widths. Erasure of the UFM involves individual sector erasing (that is, one erase of sector 0 and one erase of sector 1 is required to erase the entire UFM block). Since sector erase is required before a program or write, having two sectors enables a sector size of data to be left untouched while the other sector is erased and programmed with new data.

Figure 2–19. MAX II IOE Structure



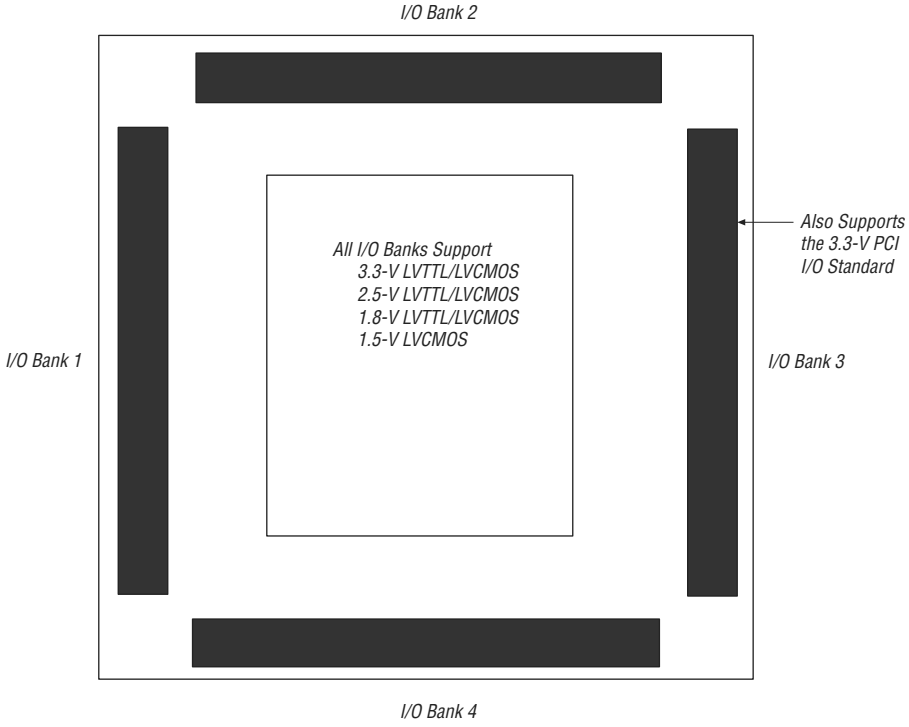
Note to Figure 2–19:

(1) Available in EPM1270 and EPM2210 devices only.

I/O Blocks

The IOEs are located in I/O blocks around the periphery of the MAX II device. There are up to seven IOEs per row I/O block (5 maximum in the EPM240 device) and up to four IOEs per column I/O block. Each column or row I/O block interfaces with its adjacent LAB and MultiTrack interconnect to distribute signals throughout the device. The row I/O blocks drive row, column, or DirectLink interconnects. The column I/O blocks drive column interconnects.

Figure 2–23. MAX II I/O Banks for EPM1270 and EPM2210 *(Note 1), (2)*



Notes to Figure 2–23:

(1) Figure 2–23 is a top view of the silicon die.

(2) Figure 2–23 is a graphical representation only. Refer to the pin list and the Quartus II software for exact pin locations.

Each I/O bank has dedicated V_{CCIO} pins that determine the voltage standard support in that bank. A single device can support 1.5-V, 1.8-V, 2.5-V, and 3.3-V interfaces; each individual bank can support a different standard. Each I/O bank can support multiple standards with the same V_{CCIO} for input and output pins. For example, when V_{CCIO} is 3.3 V, Bank 3 can support LVTTL, LVCMOS, and 3.3-V PCI. V_{CCIO} powers both the input and output buffers in MAX II devices.

The JTAG pins for MAX II devices are dedicated pins that cannot be used as regular I/O pins. The pins `TMS`, `TDI`, `TDO`, and `TCK` support all the I/O standards shown in Table 2–4 on page 2–27 except for PCI. These pins reside in Bank 1 for all MAX II devices and their I/O standard support is controlled by the V_{CCIO} setting for Bank 1.

PCI Compliance

The MAX II EPM1270 and EPM2210 devices are compliant with PCI applications as well as all 3.3-V electrical specifications in the *PCI Local Bus Specification Revision 2.2*. These devices are also large enough to support PCI intellectual property (IP) cores. Table 2–5 shows the MAX II device speed grades that meet the PCI timing specifications.

Table 2–6. Programmable Drive Strength *(Note 1)*

I/O Standard	IOH/IOL Current Strength Setting (mA)
3.3-V LVTTL	16
	8
3.3-V LVCMOS	8
	4
2.5-V LVTTL/LVCMOS	14
	7
1.8-V LVTTL/LVCMOS	6
	3
1.5-V LVCMOS	4
	2

Note to Table 2–6:

(1) The I_{OH} current strength numbers shown are for a condition of a $V_{OUT} = V_{OH}$ minimum, where the V_{OH} minimum is specified by the I/O standard. The I_{OL} current strength numbers shown are for a condition of a $V_{OUT} = V_{OL}$ maximum, where the V_{OL} maximum is specified by the I/O standard. For 2.5-V LVTTL/LVCMOS, the I_{OH} condition is V_{OUT} = 1.7 V and the I_{OL} condition is V_{OUT} = 0.7 V.

Slew-Rate Control

The output buffer for each MAX II device I/O pin has a programmable output slew-rate control that can be configured for low noise or high-speed performance. A faster slew rate provides high-speed transitions for high-performance systems. However, these fast transitions may introduce noise transients into the system. A slow slew rate reduces system noise, but adds a nominal output delay to rising and falling edges. The lower the voltage standard (for example, 1.8-V LVTTL) the larger the output delay when slow slew is enabled. Each I/O pin has an individual slew-rate control, allowing the designer to specify the slew rate on a pin-by-pin basis. The slew-rate control affects both the rising and falling edges.

Open-Drain Output

MAX II devices provide an optional open-drain (equivalent to open-collector) output for each I/O pin. This open-drain output enables the device to provide system-level control signals (for example, interrupt and write enable signals) that can be asserted by any of several devices. This output can also provide an additional wired-OR plane.

Programmable Ground Pins

Each unused I/O pin on MAX II devices can be used as an additional ground pin. This programmable ground feature does not require the use of the associated LEs in the device. In the Quartus II software, unused pins can be set as programmable GND on a global default basis or they can be individually assigned. Unused pins also have the option of being set as tri-stated input pins.

Table 3–3. 32-Bit MAX II Device IDCODE (Part 2 of 2)

Device	Binary IDCODE (32 Bits) *(1)*				HEX IDCODE
	Version (4 Bits)	Part Number	Manufacturer Identity (11 Bits)	LSB (1 Bit) *(2)*	
EPM240Z	0000	0010 0000 1010 0101	000 0110 1110	1	0x020A50DD
EPM570Z	0000	0010 0000 1010 0110	000 0110 1110	1	0x020A60DD

Notes to Table 3–2:

(1) The most significant bit (MSB) is on the left.

(2) The IDCODE's least significant bit (LSB) is always 1.

For JTAG AC characteristics, refer to the *DC and Switching Characteristics* chapter in the *MAX II Device Handbook*.

For more information about JTAG BST, refer to the *IEEE 1149.1 (JTAG) Boundary-Scan Testing for MAX II Devices* chapter in the *MAX II Device Handbook*.

JTAG Block

The MAX II JTAG block feature allows you to access the JTAG TAP and state signals when either the USER0 or USER1 instruction is issued to the JTAG TAP. The USER0 and USER1 instructions bring the JTAG boundary-scan chain (TDI) through the user logic instead of the MAX II device's boundary-scan cells. Each USER instruction allows for one unique user-defined JTAG chain into the logic array.

Parallel Flash Loader

The JTAG block ability to interface JTAG to non-JTAG devices is ideal for general-purpose flash memory devices (such as Intel- or Fujitsu-based devices) that require programming during in-circuit test. The flash memory devices can be used for FPGA configuration or be part of system memory. In many cases, the MAX II device is already connected to these devices as the configuration control logic between the FPGA and the flash device. Unlike ISP-capable CPLD devices, bulk flash devices do not have JTAG TAP pins or connections. For small flash devices, it is common to use the serial JTAG scan chain of a connected device to program the non-JTAG flash device. This is slow and inefficient in most cases and impractical for large parallel flash devices. Using the MAX II device's JTAG block as a parallel flash loader, with the Quartus II software, to program and verify flash contents provides a fast and cost-effective means of in-circuit programming during test. Figure 3–1 shows MAX II being used as a parallel flash loader.

Real-Time ISP

For systems that require more than DC logic level control of I/O pins, the real-time ISP feature allows you to update the CFM block with a new design image while the current design continues to operate in the SRAM logic array and I/O pins. A new programming file is updated into the MAX II device without halting the original design's operation, saving down-time costs for remote or field upgrades. The updated CFM block configures the new design into the SRAM upon the next power cycle. It is also possible to execute an immediate configuration of the SRAM without a power cycle by using a specific sequence of ISP commands. The configuration of SRAM without a power cycle takes a specific amount of time (t_{CONFIG}). During this time, the I/O pins are tri-stated and weakly pulled-up to V_{CCIO}.

Design Security

All MAX II devices contain a programmable security bit that controls access to the data programmed into the CFM block. When this bit is programmed, design programming information, stored in the CFM block, cannot be copied or retrieved. This feature provides a high level of design security because programmed data within flash memory cells is invisible. The security bit that controls this function, as well as all other programmed data, is reset only when the device is erased. The SRAM is also invisible and cannot be accessed regardless of the security bit setting. The UFM block data is not protected by the security bit and is accessible through JTAG or logic array connections.

Programming with External Hardware

MAX II devices can be programmed by downloading the information via in-circuit testers, embedded processors, the Altera® ByteblasterMV™, MasterBlaster™, ByteBlaster™ II, and USB-Blaster cables.

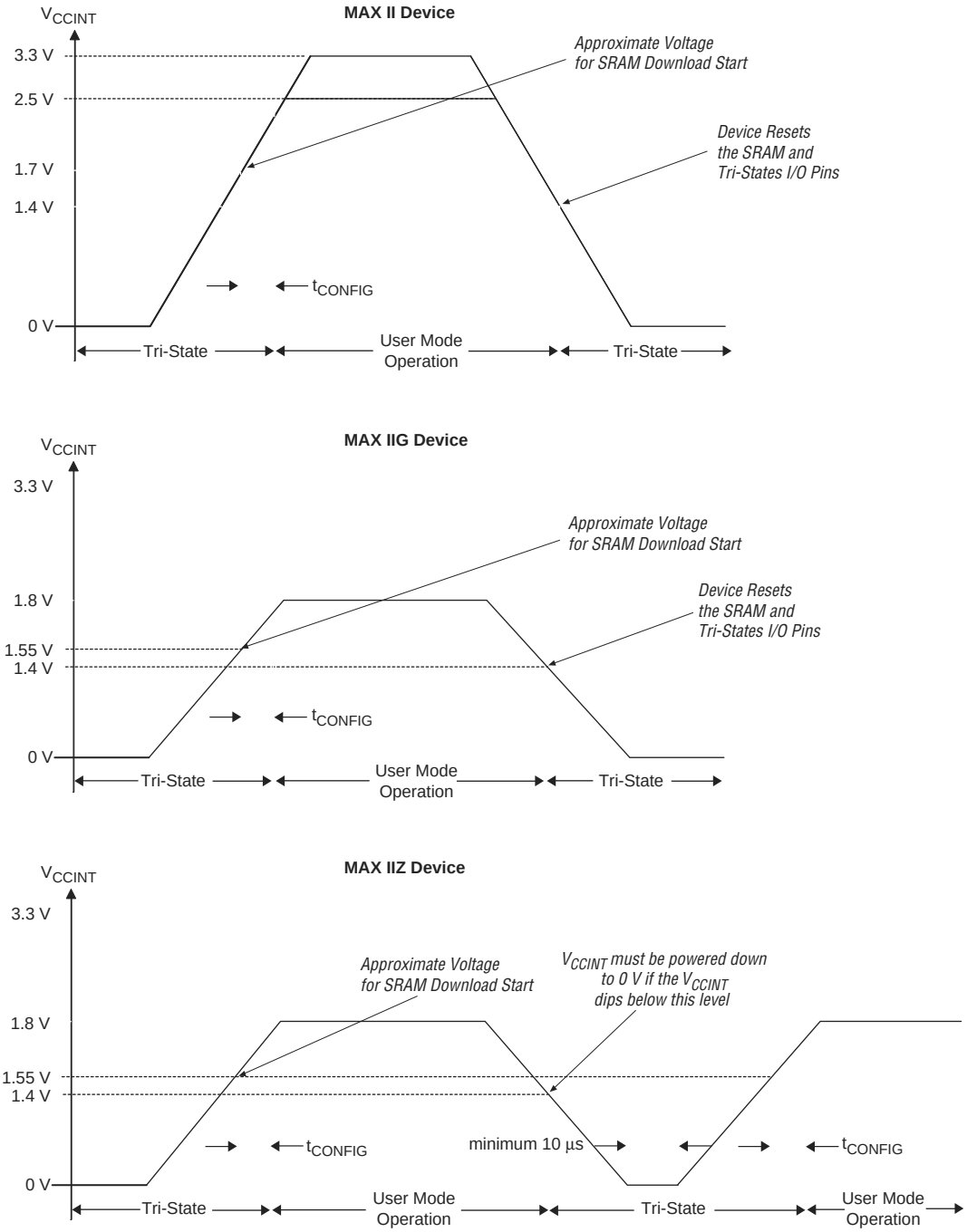
BP Microsystems, System General, and other programming hardware manufacturers provide programming support for Altera devices. Check their websites for device support information.

Referenced Documents

This chapter references the following documents:

- *DC and Switching Characteristics* chapter in the *MAX II Device Handbook*
- *IEEE 1149.1 (JTAG) Boundary-Scan Testing for MAX II Devices* chapter in the *MAX II Device Handbook*
- *Real-Time ISP and ISP Clamp for MAX II Devices* chapter in the *MAX II Device Handbook*
- *Using Jam STAPL for ISP via an Embedded Processor* chapter in the *MAX II Device Handbook*

Figure 4–5. Power-Up Characteristics for MAX II, MAX IIG, and MAX IIZ Devices *(Note 1), (2)*



Notes to Figure 4–5:

(1) Time scale is relative.

(2) Figure 4–5 assumes all V_{CCIO} banks power up simultaneously with the V_{CCINT} profile shown. If not, t_{CONFIG} stretches out until all V_{CCIO} banks are powered.

☞ After SRAM configuration, all registers in the device are cleared and released into user function before I/O tri-states are released. To release clears after tri-states are released, use the DEV_CLRn pin option. To hold the tri-states beyond the power-up configuration time, use the DEV_OE pin option.

External Timing Parameters

External timing parameters are specified by device density and speed grade. All external I/O timing parameters shown are for the 3.3-V LVTTL I/O standard with the maximum drive strength and fast slew rate. For external I/O timing using standards other than LVTTL or for different drive strengths, use the I/O standard input and output delay adders in Table 5–27 through Table 5–31.

For more information about each external timing parameters symbol, refer to the *Understanding Timing in MAX II Devices* chapter in the *MAX II Device Handbook*.

Table 5–23 shows the external I/O timing parameters for EPM240 devices.

Table 5–23. EPM240 Global Clock External I/O Timing Parameters (Part 1 of 2)

			MAX II / MAX IIG						MAX IIZ						
			–3 Speed Grade		–4 Speed Grade		–5 Speed Grade		–6 Speed Grade		–7 Speed Grade		–8 Speed Grade		
Symbol	Parameter	Condition	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Unit
t_{PD1}	Worst case pin-to-pin delay through 1 look-up table (LUT)	10 pF	—	4.7	—	6.1	—	7.5	—	7.9	—	12.0	—	14.0	ns
t_{PD2}	Best case pin-to-pin delay through 1 LUT	10 pF	—	3.7	—	4.8	—	5.9	—	5.8	—	7.8	—	8.5	ns
t_{SU}	Global clock setup time	—	1.7	—	2.2	—	2.7	—	2.4	—	4.1	—	4.6	—	ns
t_{H}	Global clock hold time	—	0	—	0	—	0	—	0	—	0	—	0	—	ns
t_{CO}	Global clock to output delay	10 pF	2.0	4.3	2.0	5.6	2.0	6.9	2.0	6.6	2.0	8.1	2.0	8.6	ns
t_{CH}	Global clock high time	—	166	—	216	—	266	—	253	—	335	—	339	—	ps
t_{CL}	Global clock low time	—	166	—	216	—	266	—	253	—	335	—	339	—	ps
t_{CNT}	Minimum global clock period for 16-bit counter	—	3.3	—	4.0	—	5.0	—	5.4	—	8.1	—	8.4	—	ns

Table 5–26 shows the external I/O timing parameters for EPM2210 devices.

Table 5–26. EPM2210 Global Clock External I/O Timing Parameters

Symbol	Parameter	Condition	MAX II / MAX IIG						Unit
			–3 Speed Grade		–4 Speed Grade		–5 Speed Grade		
			Min	Max	Min	Max	Min	Max	
t_{PD1}	Worst case pin-to-pin delay through 1 look-up table (LUT)	10 pF	—	7.0	—	9.1	—	11.2	ns
t_{PD2}	Best case pin-to-pin delay through 1 LUT	10 pF	—	3.7	—	4.8	—	5.9	ns
t_{SU}	Global clock setup time	—	1.2	—	1.5	—	1.9	—	ns
t_{H}	Global clock hold time	—	0	—	0	—	0	—	ns
t_{CO}	Global clock to output delay	10 pF	2.0	4.6	2.0	6.0	2.0	7.4	ns
t_{CH}	Global clock high time	—	166	—	216	—	266	—	ps
t_{CL}	Global clock low time	—	166	—	216	—	266	—	ps
t_{CNT}	Minimum global clock period for 16-bit counter	—	3.3	—	4.0	—	5.0	—	ns
f_{CNT}	Maximum global clock frequency for 16-bit counter	—	—	304.0 *(1)*	—	247.5	—	201.1	MHz

Note to Table 5–26:

(1) The maximum frequency is limited by the I/O standard on the clock input pin. The 16-bit counter critical delay performs faster than this global clock input pin maximum frequency.

External Timing I/O Delay Adders

The I/O delay timing parameters for I/O standard input and output adders, and input delays are specified by speed grade independent of device density.

Table 5–27 through Table 5–31 show the adder delays associated with I/O pins for all packages. The delay numbers for –3, –4, and –5 speed grades shown in Table 5–27 through Table 5–33 are based on an EPM1270 device target, while –6, –7, and –8 speed grade values are based on an EPM570Z device target. If an I/O standard other than 3.3-V LVTTL is selected, add the input delay adder to the external t_{SU} timing parameters shown in Table 5–23 through Table 5–26. If an I/O standard other than 3.3-V LVTTL with 16 mA drive strength and fast slew rate is selected, add the output delay adder to the external t_{CO} and t_{PD} shown in Table 5–23 through Table 5–26.

Table 5–27. External Timing Input Delay Adders (Part 1 of 2)

I/O Standard		MAX II / MAX IIG						MAX IIZ						Unit
		–3 Speed Grade		–4 Speed Grade		–5 Speed Grade		–6 Speed Grade		–7 Speed Grade		–8 Speed Grade		
		Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	
3.3-V LVTTL	Without Schmitt Trigger	—	0	—	0	—	0	—	0	—	0	—	0	ps
	With Schmitt Trigger	—	334	—	434	—	535	—	387	—	434	—	442	ps

Table 5–29. External Timing Output Delay and t_{OD} Adders for Fast Slew Rate

I/O Standard		MAX II / MAX IIG						MAX IIZ						Unit
		–3 Speed Grade		–4 Speed Grade		–5 Speed Grade		–6 Speed Grade		–7 Speed Grade		–8 Speed Grade		
		Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	
3.3-V LVTTL	16 mA	—	0	—	0	—	0	—	0	—	0	—	0	ps
	8 mA	—	65	—	84	—	104	—	–6	—	–2	—	–3	ps
3.3-V LVCMOS	8 mA	—	0	—	0	—	0	—	0	—	0	—	0	ps
	4 mA	—	65	—	84	—	104	—	–6	—	–2	—	–3	ps
2.5-V LVTTL / LVCMOS	14 mA	—	122	—	158	—	195	—	–63	—	–71	—	–88	ps
	7 mA	—	193	—	251	—	309	—	10	—	–1	—	1	ps
1.8-V LVTTL / LVCMOS	6 mA	—	568	—	738	—	909	—	128	—	118	—	118	ps
	3 mA	—	654	—	850	—	1,046	—	352	—	327	—	332	ps
1.5-V LVCMOS	4 mA	—	1,059	—	1,376	—	1,694	—	421	—	400	—	400	ps
	2 mA	—	1,167	—	1,517	—	1,867	—	757	—	743	—	743	ps
3.3-V PCI	20 mA	—	3	—	4	—	5	—	–6	—	–2	—	–3	ps

Table 5–30. External Timing Output Delay and t_{OD} Adders for Slow Slew Rate

I/O Standard		MAX II / MAX IIG						MAX IIZ						Unit
		–3 Speed Grade		–4 Speed Grade		–5 Speed Grade		–6 Speed Grade		–7 Speed Grade		–8 Speed Grade		
		Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	
3.3-V LVTTL	16 mA	—	7,064	—	6,745	—	6,426	—	5,966	—	5,992	—	6,118	ps
	8 mA	—	7,946	—	7,627	—	7,308	—	6,541	—	6,570	—	6,720	ps
3.3-V LVCMOS	8 mA	—	7,064	—	6,745	—	6,426	—	5,966	—	5,992	—	6,118	ps
	4 mA	—	7,946	—	7,627	—	7,308	—	6,541	—	6,570	—	6,720	ps
2.5-V LVTTL / LVCMOS	14 mA	—	10,434	—	10,115	—	9,796	—	9,141	—	9,154	—	9,297	ps
	7 mA	—	11,548	—	11,229	—	10,910	—	9,861	—	9,874	—	10,037	ps
1.8-V LVTTL / LVCMOS	6 mA	—	22,927	—	22,608	—	22,289	—	21,811	—	21,854	—	21,857	ps
	3 mA	—	24,731	—	24,412	—	24,093	—	23,081	—	23,034	—	23,107	ps
1.5-V LVCMOS	4 mA	—	38,723	—	38,404	—	38,085	—	39,121	—	39,124	—	39,124	ps
	2 mA	—	41,330	—	41,011	—	40,692	—	40,631	—	40,634	—	40,634	ps
3.3-V PCI	20 mA	—	261	—	339	—	418	—	6,644	—	6,627	—	6,914	ps

Table 5–31. MAX II IOE Programmable Delays

Parameter	MAX II / MAX IIG						MAX IIZ						Unit
	–3 Speed Grade		–4 Speed Grade		–5 Speed Grade		–6 Speed Grade		–7 Speed Grade		–8 Speed Grade		
	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	Min	Max	
Input Delay from Pin to Internal Cells = 1	—	1,225	—	1,592	—	1,960	—	1,858	—	2,171	—	2,214	ps
Input Delay from Pin to Internal Cells = 0	—	89	—	115	—	142	—	569	—	609	—	616	ps

Maximum Input and Output Clock Rates

Table 5–32 and Table 5–33 show the maximum input and output clock rates for standard I/O pins in MAX II devices.

Table 5–32. MAX II Maximum Input Clock Rate for I/O

I/O Standard		MAX II / MAX IIG			MAX IIZ			Unit
		–3 Speed Grade	–4 Speed Grade	–5 Speed Grade	–6 Speed Grade	–7 Speed Grade	–8 Speed Grade	
3.3-V LVTTL	Without Schmitt Trigger	304	304	304	304	304	304	MHz
	With Schmitt Trigger	250	250	250	250	250	250	MHz
3.3-V LVCMOS	Without Schmitt Trigger	304	304	304	304	304	304	MHz
	With Schmitt Trigger	250	250	250	250	250	250	MHz
2.5-V LVTTL	Without Schmitt Trigger	220	220	220	220	220	220	MHz
	With Schmitt Trigger	188	188	188	188	188	188	MHz
2.5-V LVCMOS	Without Schmitt Trigger	220	220	220	220	220	220	MHz
	With Schmitt Trigger	188	188	188	188	188	188	MHz
1.8-V LVTTL	Without Schmitt Trigger	200	200	200	200	200	200	MHz
1.8-V LVCMOS	Without Schmitt Trigger	200	200	200	200	200	200	MHz
1.5-V LVCMOS	Without Schmitt Trigger	150	150	150	150	150	150	MHz
3.3-V PCI	Without Schmitt Trigger	304	304	304	304	304	304	MHz

Document Revision History

Table 5–35 shows the revision history for this chapter.

Table 5–35. Document Revision History (Part 1 of 2)

Date and Revision	Changes Made	Summary of Changes
August 2009, version 2.5	■ Added Table 5–28, Table 5–29, and Table 5–30. ■ Updated Table 5–2, Table 5–4, Table 5–14, Table 5–15, Table 5–16, Table 5–17, Table 5–18, Table 5–19, Table 5–20, Table 5–21, Table 5–22, Table 5–23, Table 5–24, Table 5–27, Table 5–31, Table 5–32, and Table 5–33.	Added information for speed grade –8
November 2008, version 2.4	■ Updated Table 5–2. ■ Updated "Internal Timing Parameters" section.	—
October 2008, version 2.3	■ Updated New Document Format. ■ Updated Figure 5–1.	—
July 2008, version 2.2	■ Updated Table 5–14 , Table 5–23 , and Table 5–24.	—
March 2008, version 2.1	■ Added (Note 5) to Table 5–4.	—
December 2007, version 2.0	■ Updated (Note 3) and (4) to Table 5–1. ■ Updated Table 5–2 and added (Note 5). ■ Updated ICCSTANDBY and ICCPOWERUP information and added IPULLUP information in Table 5–4. ■ Added (Note 1) to Table 5–10. ■ Updated Figure 5–2. ■ Added (Note 1) to Table 5–13. ■ Updated Table 5–13 through Table 5–24, and Table 5–27 through Table 5–30. ■ Added tCOMB information to Table 5–15. ■ Updated Figure 5–6. ■ Added "Referenced Documents" section.	Updated document with MAX IIZ information.
December 2006, version 1.8	■ Added note to Table 5–1. ■ Added document revision history.	—
July 2006, version 1.7	■ Minor content and table updates.	—
February 2006, version 1.6	■ Updated "External Timing I/O Delay Adders" section. ■ Updated Table 5–29. ■ Updated Table 5–30.	—
November 2005, version 1.5	■ Updated Tables 5-2, 5-4, and 5-12.	—
August 2005, version 1.4	■ Updated Figure 5-1. ■ Updated Tables 5-13, 5-16, and 5-26. ■ Removed Note 1 from Table 5-12.	—