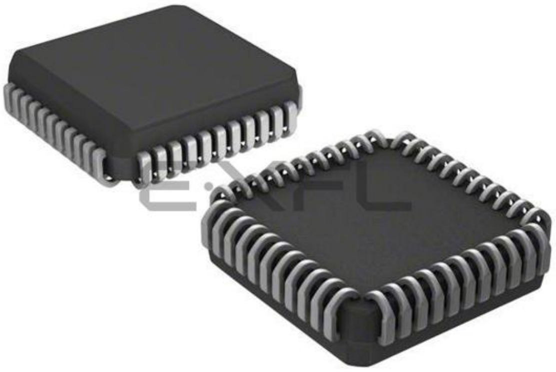

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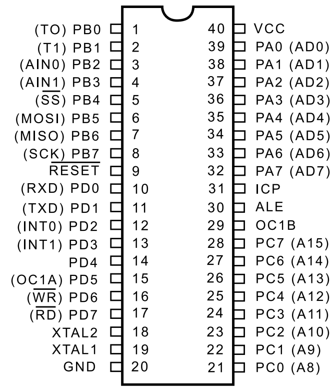
Applications of "Embedded - Microcontrollers"

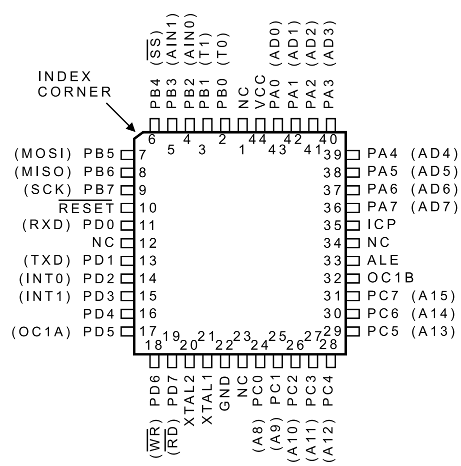
Details	
Product Status	Obsolete
Core Processor	AVR
Core Size	8-Bit
Speed	8MHz
Connectivity	SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	32
Program Memory Size	8KB (4K x 16)
Program Memory Type	FLASH
EEPROM Size	512 x 8
RAM Size	512 x 8
Voltage - Supply (Vcc/Vdd)	4V ~ 6V
Data Converters	-
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C
Mounting Type	Surface Mount
Package / Case	44-LCC (J-Lead)
Supplier Device Package	44-PLCC (16.6x16.6)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/at90s8515a-8ji

Pin Configurations

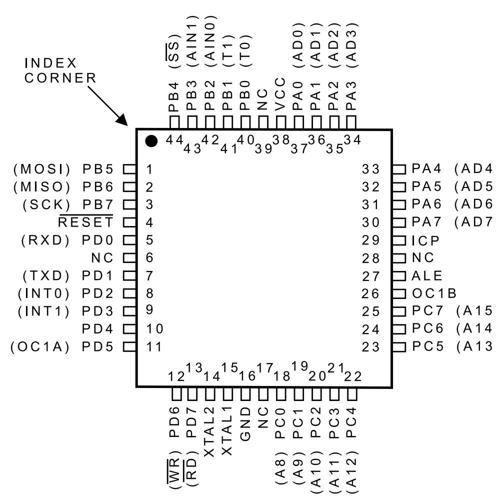
PDIP



PLCC



TQFP



Architectural Overview

The fast-access register file concept contains 32 x 8-bit general-purpose working registers with a single clock cycle access time. This means that during one single clock cycle, one ALU (Arithmetic Logic Unit) operation is executed. Two operands are output from the register file, the operation is executed and the result is stored back in the register file – in one clock cycle.

Six of the 32 registers can be used as three 16-bit indirect address register pointers for Data Space addressing, enabling efficient address calculations. One of the three address pointers is also used as the address pointer for the constant table look-up function. These added function registers are the 16-bit X-, Y-, and Z-register.

The ALU supports arithmetic and logic functions between registers or between a constant and a register. Single register operations are also executed in the ALU. Figure 4 shows the AT90S8515 AVR RISC microcontroller architecture.

In addition to the register operation, the conventional memory addressing modes can be used on the register file as well. This is enabled by the fact that the register file is assigned the 32 lowermost Data Space addresses ($00 - $1F), allowing them to be accessed as though they were ordinary memory locations.

The I/O memory space contains 64 addresses for CPU peripheral functions such as Control Registers, Timer/Counters, A/D converters and other I/O functions. The I/O memory can be accessed directly or as the Data Space locations following those of the register file, $20 - $5F.

The AVR uses a Harvard architecture concept – with separate memories and buses for program and data. The program memory is executed with a two-stage pipeline. While one instruction is being executed, the next instruction is pre-fetched from the program memory. This concept enables instructions to be executed in every clock cycle. The program memory is In-System Programmable Flash memory.

With the relative jump and call instructions, the whole 4K address space is directly accessed. Most AVR instructions have a single 16-bit word format. Every program memory address contains a 16- or 32-bit instruction.

During interrupts and subroutine calls, the return address Program Counter (PC) is stored on the stack. The stack is effectively allocated in the general data SRAM and consequently, the stack size is only limited by the total SRAM size and the usage of the SRAM. All user programs must initialize the SP in the reset routine (before subroutines or interrupts are executed). The 16-bit Stack Pointer (SP) is read/write-accessible in the I/O space.

The 512-byte data SRAM can be easily accessed through the five different addressing modes supported in the AVR architecture.

The memory spaces in the AVR architecture are all linear and regular memory maps.

The user can select the start-up time according to typical oscillator start-up. The number of WDT oscillator cycles used for each time-out is shown in Table 4. The frequency of the Watchdog Oscillator is voltage-dependent as shown in “Typical Characteristics” on page 95.

Table 4. Number of Watchdog Oscillator Cycles

FSTRT	Time-out at V_{CC} = 5V	Number of WDT Cycles
Programmed	0.28 ms	256
Unprogrammed	16.0 ms	16K

Power-on Reset

A Power-on Reset (POR) circuit ensures that the device is reset from power-on. As shown in Figure 23, an internal timer clocked from the Watchdog Timer oscillator prevents the MCU from starting until after a certain period after V_{CC} has reached the Power-on Threshold Voltage (V_{POT}), regardless of the V_{CC} rise time (see Figure 24). The FSTRT Fuse bit in the Flash can be programmed to give a shorter start-up time if a ceramic resonator or any other fast-start oscillator is used to clock the MCU.

If the built-in start-up delay is sufficient, $\overline{\text{RESET}}$ can be connected to V_{CC} directly or via an external pull-up resistor. By holding the pin low for a period after V_{CC} has been applied, the Power-on Reset period can be extended. Refer to Figure 25 for a timing example of this.

Figure 24. MCU Start-up, $\overline{\text{RESET}}$ Tied to V_{CC}.

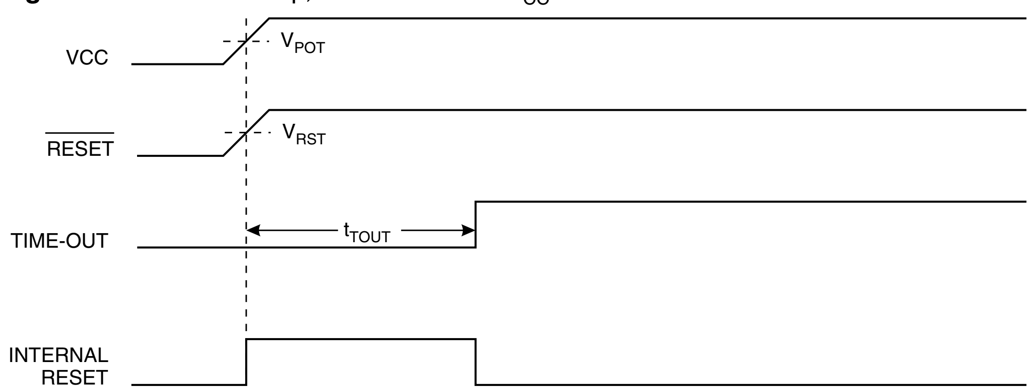
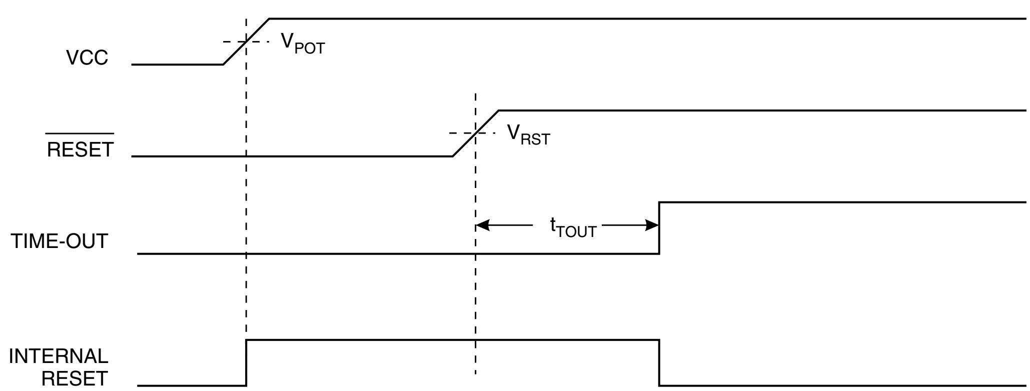


Figure 25. MCU Start-up, $\overline{\text{RESET}}$ Controlled Externally



• **Bit 6 – INTF0: External Interrupt Flag0**

When an edge on the INT0 pin triggers an interrupt request, the corresponding interrupt flag, INTF0, becomes set (one). If the I-bit in SREG and the corresponding interrupt enable bit, INT0 in GIMSK are set (one), the MCU will jump to the interrupt vector. The flag is cleared when the interrupt routine is executed. Alternatively, the flag is cleared by writing a logical “1” to it. This flag is always cleared when INT0 is configured as level interrupt.

• **Bits 5..0 – Res: Reserved Bits**

These bits are reserved bits in the AT90S8515 and always read as zero.

Timer/Counter Interrupt Mask Register – TIMSK

Bit	7	6	5	4	3	2	1	0	
$39 ($59)	TOIE1	OCIE1A	OCIE1B	–	TICIE1	–	TOIE0	–	TIMSK
Read/Write	R/W	R/W	R/W	R	R/W	R	R/W	R	
Initial Value	0	0	0	0	0	0	0	0	

• **Bit 7 – TOIE1: Timer/Counter1 Overflow Interrupt Enable**

When the TOIE1 bit is set (one) and the I-bit in the Status Register is set (one), the Timer/Counter1 Overflow interrupt is enabled. The corresponding interrupt (at vector $006) is executed if an overflow in Timer/Counter1 occurs, i.e., when the TOV1 bit is set in the Timer/Counter Interrupt Flag Register (TIFR).

• **Bit 6 – OCE1A: Timer/Counter1 Output CompareA Match Interrupt Enable**

When the OCIE1A bit is set (one) and the I-bit in the Status Register is set (one), the Timer/Counter1 CompareA Match interrupt is enabled. The corresponding interrupt (at vector $004) is executed if a CompareA match in Timer/Counter1 occurs, i.e., when the OCF1A bit is set in the Timer/Counter Interrupt Flag Register (TIFR).

• **Bit 5 – OCIE1B: Timer/Counter1 Output CompareB Match Interrupt Enable**

When the OCIE1B bit is set (one) and the I-bit in the Status Register is set (one), the Timer/Counter1 CompareB Match interrupt is enabled. The corresponding interrupt (at vector $005) is executed if a CompareB match in Timer/Counter1 occurs, i.e., when the OCF1B bit is set in the Timer/Counter Interrupt Flag Register (TIFR).

• **Bit 4 – Res: Reserved Bit**

This bit is a reserved bit in the AT90S8515 and always reads zero.

• **Bit 3 – TICIE1: Timer/Counter1 Input Capture Interrupt Enable**

When the TICIE1 bit is set (one) and the I-bit in the Status Register is set (one), the Timer/Counter1 Input Capture Event interrupt is enabled. The corresponding interrupt (at vector $003) is executed if a capture-triggering event occurs on pin 31, ICP, i.e., when the ICF1 bit is set in the Timer/Counter Interrupt Flag Register (TIFR).

• **Bit 2 – Res: Reserved Bit**

This bit is a reserved bit in the AT90S8515 and always reads zero.

• **Bit 1 – TOIE0: Timer/Counter0 Overflow Interrupt Enable**

When the TOIE0 bit is set (one) and the I-bit in the Status Register is set (one), the Timer/Counter0 Overflow interrupt is enabled. The corresponding interrupt (at vector $007) is executed if an overflow in Timer/Counter0 occurs, i.e., when the TOV0 bit is set in the Timer/Counter Interrupt Flag Register (TIFR).

• **Bit 0 – Res: Reserved Bit**

This bit is a reserved bit in the AT90S8515 and always reads zero.

The TEMP register is also used when accessing TCNT1, OCR1A and OCR1B. If the main program and interrupt routines perform access to registers using TEMP, interrupts must be disabled during access from the main program (and from interrupt routines if interrupts are allowed from within interrupt routines).

Timer/Counter1 in PWM Mode

When the PWM mode is selected, Timer/Counter1, the Output Compare Register1A (OCR1A) and the Output Compare Register1B (OCR1B) form a dual 8-, 9- or 10-bit, free-running, glitch-free and phase-correct PWM with outputs on the PD5(OC1A) and OC1B pins. Timer/Counter1 acts as an up/down counter, counting up from $0000 to TOP (see Table 11), where it turns and counts down again to zero before the cycle is repeated. When the counter value matches the contents of the 10 least significant bits of OCR1A or OCR1B, the PD5(OC1A)/OC1B pins are set or cleared according to the settings of the COM1A1/COM1A0 or COM1B1/COM1B0 bits in the Timer/Counter1 Control Register (TCCR1A). Refer to Table 12 for details.

Table 11. Timer TOP Values and PWM Frequency

PWM Resolution	Timer TOP Value	Frequency
8-bit	$00FF (255)	$f_{TCK1}/510$
9-bit	$01FF (511)	$f_{TCK1}/1022$
10-bit	$03FF(1023)	$f_{TCK1}/2046$

Table 12. Compare1 Mode Select in PWM Mode

COM1X1	COM1X0	Effect on OCX1
0	0	Not connected
0	1	Not connected
1	0	Cleared on compare match, up-counting. Set on compare match, down-counting (non-inverted PWM).
1	1	Cleared on compare match, down-counting. Set on compare match, up-counting (inverted PWM).

Note: X = A or B

Note that in the PWM mode, the 10 least significant OCR1A/OCR1B bits, when written, are transferred to a temporary location. They are latched when Timer/Counter1 reaches the value TOP. This prevents the occurrence of odd-length PWM pulses (glitches) in the event of an unsynchronized OCR1A/OCR1B write. See Figure 32 for an example.

1. In the same operation, write a logical “1” to WDTOE and WDE. A logical “1” must be written to WDE even though it is set to one before the disable operation starts.
2. Within the next four clock cycles, write a logical “0” to WDE. This disables the Watchdog.

- **Bits 2..0 – WDP2, WDP1, WDP0: Watchdog Timer Prescaler 2, 1 and 0**

The WDP2, WDP1 and WDP0 bits determine the Watchdog Timer prescaling when the Watchdog Timer is enabled. The different prescaling values and their corresponding Time-out periods are shown in Table 14.

Table 14. Watchdog Timer Prescale Select

WDP2	WDP1	WDP0	Number of WDT Oscillator Cycles	Typical Time-out at V_{CC} = 3.0V	Typical Time-out at V_{CC} = 5.0V
0	0	0	16K cycles	47.0 ms	15.0 ms
0	0	1	32K cycles	94.0 ms	30.0 ms
0	1	0	64K cycles	0.19 s	60.0 ms
0	1	1	128K cycles	0.38 s	0.12 s
1	0	0	256K cycles	0.75 s	0.24 s
1	0	1	512K cycles	1.5 s	0.49 s
1	1	0	1,024K cycles	3.0 s	0.97 s
1	1	1	2,048K cycles	6.0 s	1.9 s

Note: The frequency of the Watchdog oscillator is voltage-dependent as shown in the Electrical Characteristics section.
The WDR (Watchdog Reset) instruction should always be executed before the Watchdog Timer is enabled. This ensures that the reset period will be in accordance with the Watchdog Timer prescale settings. If the Watchdog Timer is enabled without reset, the Watchdog Timer may not start to count from zero.
To avoid unintentional MCU reset, the Watchdog Timer should be disabled or reset before changing the Watchdog Timer Prescale Select.

- **Bit 2 – EEMWE: EEPROM Master Write Enable**

The EEMWE bit determines whether setting EEWE to one causes the EEPROM to be written. When EEMWE is set (one), setting EEWE will write data to the EEPROM at the selected address. If EEMWE is zero, setting EEWE will have no effect. When EEMWE has been set (one) by software, hardware clears the bit to zero after four clock cycles. See the description of the EEWE bit for a EEPROM write procedure.

- **Bit 1 – EEWE: EEPROM Write Enable**

The EEPROM Write Enable signal (EEWE) is the write strobe to the EEPROM. When address and data are correctly set up, the EEWE bit must be set to write the value into the EEPROM. The EEMWE bit must be set when the logical "1" is written to EEWE, otherwise no EEPROM write takes place. The following procedure should be followed when writing the EEPROM (the order of steps 2 and 3 is unessential):

1. Wait until EEWE becomes zero.
2. Write new EEPROM address to EEARL and EEARH (optional).
3. Write new EEPROM data to EEDR (optional).
4. Write a logical "1" to the EEMWE bit in EECR (to be able to write a logical "1" to the EEMWE bit, the EEWE bit must be written to zero in the same cycle).
5. Within four clock cycles after setting EEMWE, write a logical "1" to EEWE.

When the write access time (typically 2.5 ms at V_{CC} = 5V or 4 ms at V_{CC} = 2.7V) has elapsed, the EEWE bit is cleared (zero) by hardware. The user software can poll this bit and wait for a zero before writing the next byte. When EEWE has been set, the CPU is halted for two cycles before the next instruction is executed.

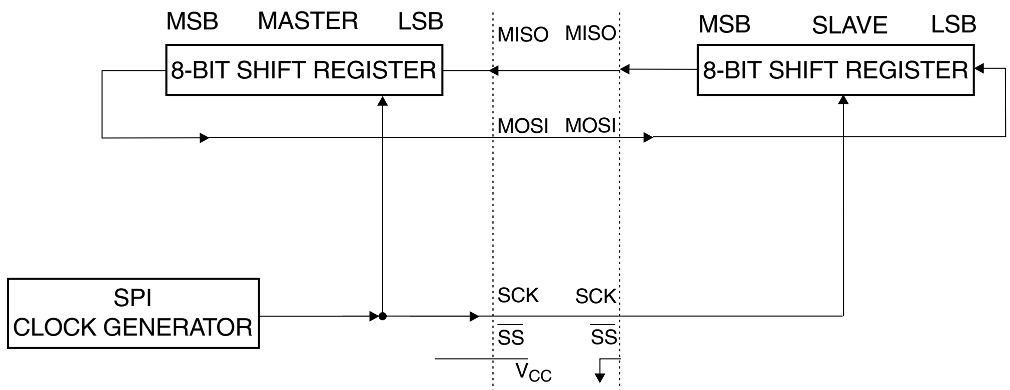
Caution: An interrupt between step 4 and step 5 will make the write cycle fail, since the EEPROM Master Write Enable will time-out. If an interrupt routine accessing the EEPROM is interrupting another EEPROM access, the EEAR or EEDR registers will be modified, causing the interrupted EEPROM access to fail. It is recommended to have the global interrupt flag cleared during the four last steps to avoid these problems.

- **Bit 0 – EERE: EEPROM Read Enable**

The EEPROM Read Enable signal EERE is the read strobe to the EEPROM. When the correct address is set up in the EEAR register, the EERE bit must be set. When the EERE bit is cleared (zero) by hardware, requested data is found in the EEDR register. The EEPROM read access takes one instruction and there is no need to poll the EERE bit. When EERE has been set, the CPU is halted for four cycles before the next instruction is executed.

The user should poll the EEWE bit before starting the read operation. If a write operation is in progress when new data or address is written to the EEPROM I/O registers, the write operation will be interrupted and the result is undefined.

Figure 35. SPI Master-slave Interconnection



The system is single-buffered in the transmit direction and double-buffered in the receive direction. This means that bytes to be transmitted cannot be written to the SPI Data Register before the entire shift cycle is completed. When receiving data, however, a received byte must be read from the SPI Data Register before the next byte has been completely shifted in. Otherwise, the first byte is lost.

When the SPI is enabled, the data direction of the MOSI, MISO, SCK and $\overline{SS}$ pins is overridden according to Table 15.

Table 15. SPI Pin Overrides

Pin	Direction, Master SPI	Direction, Slave SPI
MOSI	User Defined	Input
MISO	Input	User Defined
SCK	User Defined	Input
$\overline{SS}$	User Defined	Input

Note: See “Alternate Functions of Port B” on page 66 for a detailed description of how to define the direction of the user-defined SPI pins.

$\overline{SS}$ Pin Functionality

When the SPI is configured as a master (MSTR in SPCR is set), the user can determine the direction of the $\overline{SS}$ pin. If $\overline{SS}$ is configured as an output, the pin is a general output pin, which does not affect the SPI system. If $\overline{SS}$ is configured as an input, it must be held high to ensure master SPI operation. If the $\overline{SS}$ pin is driven low by peripheral circuitry when the SPI is configured as master with the $\overline{SS}$ pin defined as an input, the SPI system interprets this as another master selecting the SPI as a slave and starts to send data to it. To avoid bus contention, the SPI system takes the following actions:

1. The MSTR bit in SPCR is cleared and the SPI system becomes a slave. As a result of the SPI becoming a slave, the MOSI and SCK pins become inputs.
2. The SPIF flag in SPSR is set, and if the SPI interrupt is enabled and the I-bit in SREG is set, the interrupt routine will be executed.

Thus, when interrupt-driven SPI transmittal is used in Master Mode and there exists a possibility that $\overline{SS}$ is driven low, the interrupt should always check that the MSTR bit is still set. Once the MSTR bit has been cleared by a slave select, it must be set by the user to re-enable SPI Master Mode.

When the SPI is configured as a slave, the $\overline{SS}$ pin is always input. When $\overline{SS}$ is held low, the SPI is activated and MISO becomes an output if configured so by the user. All other

UART

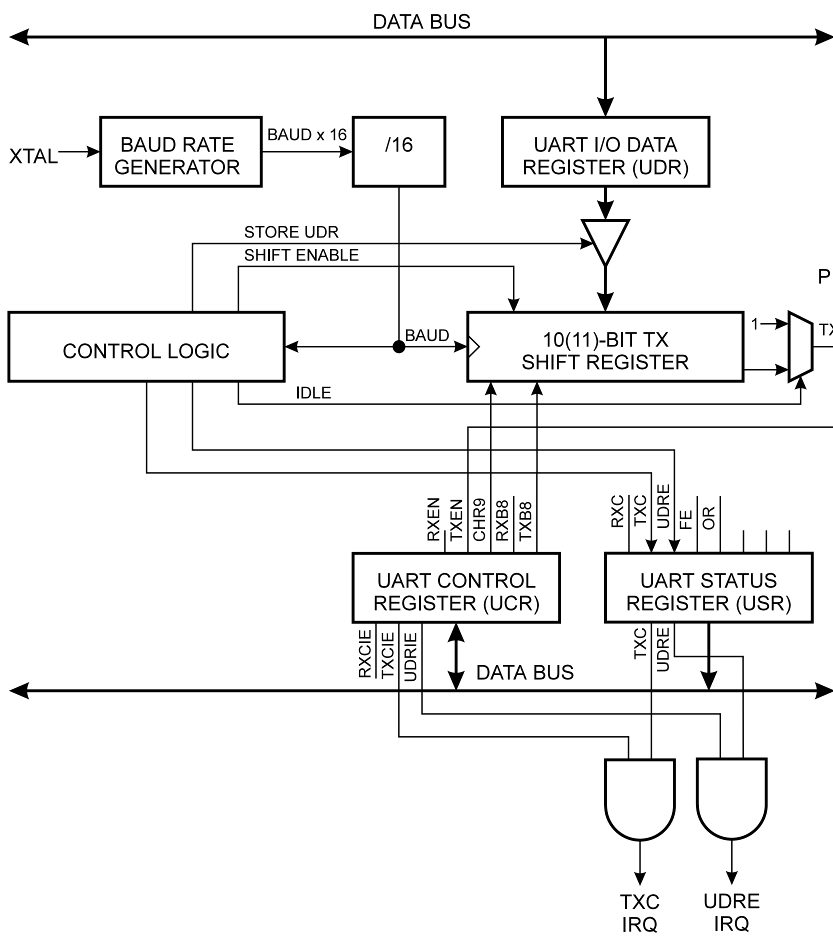
The AT90S8515 features a full duplex (separate receive and transmit registers) Universal Asynchronous Receiver and Transmitter (UART). The main features are:

- **Baud Rate Generator that can Generate a large Number of Baud Rates (bps)**
- **High Baud Rates at Low XTAL Frequencies**
- **8 or 9 Bits Data**
- **Noise Filtering**
- **Overrun Detection**
- **Framing Error Detection**
- **False Start Bit Detection**
- **Three separate Interrupts on TX Complete, TX Data Register Empty and RX Complete**

Data Transmission

A block schematic of the UART transmitter is shown in Figure 38.

Figure 38. UART Transmitter



Data transmission is initiated by writing the data to be transmitted to the UART I/O Data Register, UDR. Data is transferred from UDR to the Transmit shift register when:

- A new character has been written to UDR after the stop bit from the previous character has been shifted out. The shift register is loaded immediately.
- A new character has been written to UDR before the stop bit from the previous character has been shifted out. The shift register is loaded when the stop bit of the character currently being transmitted has been shifted out.

The FE bit is cleared when the stop bit of received data is one.

- **Bit 3 – OR: Overrun**

This bit is set if an Overrun condition is detected, i.e., when a character already present in the UDR register is not read before the next character has been shifted into the Receiver Shift register. The OR bit is buffered, which means that it will be set once the valid data still in UDRE is read.

The OR bit is cleared (zero) when data is received and transferred to UDR.

- **Bits 2..0 – Res: Reserved Bits**

These bits are reserved bits in the AT90S8515 and will always read as zero.

UART Control Register – UCR

Bit	7	6	5	4	3	2	1	0	
$0A ($2A)	RXCIE	TXCIE	UDRIE	RXEN	TXEN	CHR9	RXB8	TXB8	UCR
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R	W	
Initial Value	0	0	0	0	0	0	1	0	

- **Bit 7 – RXCIE: RX Complete Interrupt Enable**

When this bit is set (one), a setting of the RXC bit in USR will cause the Receive Complete Interrupt routine to be executed provided that global interrupts are enabled.

- **Bit 6 – TXCIE: TX Complete Interrupt Enable**

When this bit is set (one), a setting of the TXC bit in USR will cause the Transmit Complete Interrupt routine to be executed provided that global interrupts are enabled.

- **Bit 5 – UDRIE: UART Data Register Empty Interrupt Enable**

When this bit is set (one), a setting of the UDRE bit in USR will cause the UART Data Register Empty Interrupt routine to be executed provided that global interrupts are enabled.

- **Bit 4 – RXEN: Receiver Enable**

This bit enables the UART receiver when set (one). When the receiver is disabled, the RXC, OR and FE status flags cannot become set. If these flags are set, turning off RXEN does not cause them to be cleared.

- **Bit 3 – TXEN: Transmitter Enable**

This bit enables the UART transmitter when set (one). When disabling the transmitter while transmitting a character, the transmitter is not disabled before the character in the shift register plus any following character in UDR has been completely transmitted.

- **Bit 2 – CHR9: 9-bit Characters**

When this bit is set (one) transmitted and received characters are 9 bits long plus start and stop bits. The ninth bit is read and written by using the RXB8 and TXB8 bits in UCR, respectively. The ninth data bit can be used as an extra stop bit or a parity bit.

- **Bit 1 – RXB8: Receive Data Bit 8**

When CHR9 is set (one), RXB8 is the ninth data bit of the received character.

- **Bit 0 – TXB8: Transmit Data Bit 8**

When CHR9 is set (one), TXB8 is the ninth data bit in the character to be transmitted.

Table 17. UBRR Settings at Various Crystal Frequencies

Baud Rate	**1 MHz**	%Error	**1.8432 MHz**	%Error	**2 MHz**	%Error	**2.4576 MHz**	%Error
2400	UBRR= **25**	**0.2**	UBRR= **47**	**0.0**	UBRR= **51**	**0.2**	UBRR= **63**	**0.0**
4800	UBRR= **12**	**0.2**	UBRR= **23**	**0.0**	UBRR= **25**	**0.2**	UBRR= **31**	**0.0**
9600	UBRR= 6	7.5	UBRR= **11**	**0.0**	UBRR= **12**	**0.2**	UBRR= **15**	**0.0**
14400	UBRR= 3	7.8	UBRR= **7**	**0.0**	UBRR= 8	3.7	UBRR= 10	3.1
19200	UBRR= 2	7.8	UBRR= **5**	**0.0**	UBRR= 6	7.5	UBRR= **7**	**0.0**
28800	UBRR= 1	7.8	UBRR= **3**	**0.0**	UBRR= 3	7.8	UBRR= 4	6.3
38400	UBRR= 1	22.9	UBRR= **2**	**0.0**	UBRR= 2	7.8	UBRR= **3**	**0.0**
57600	UBRR= 0	7.8	UBRR= **1**	**0.0**	UBRR= 1	7.8	UBRR= 2	12.5
76800	UBRR= 0	22.9	UBRR= 1	33.3	UBRR= 1	22.9	UBRR= **1**	**0.0**
115200	UBRR= 0	84.3	UBRR= **0**	**0.0**	UBRR= 0	7.8	UBRR= 0	25.0

Baud Rate	**3.2768 MHz**	%Error	**3.6864 MHz**	%Error	**4 MHz**	%Error	**4.608 MHz**	%Error
2400	UBRR= **84**	**0.4**	UBRR= **95**	**0.0**	UBRR= **103**	**0.2**	UBRR= **119**	**0.0**
4800	UBRR= **42**	**0.8**	UBRR= **47**	**0.0**	UBRR= **51**	**0.2**	UBRR= **59**	**0.0**
9600	UBRR= **20**	**1.6**	UBRR= **23**	**0.0**	UBRR= **25**	**0.2**	UBRR= **29**	**0.0**
14400	UBRR= **13**	**1.6**	UBRR= **15**	**0.0**	UBRR= 16	2.1	UBRR= **19**	**0.0**
19200	UBRR= 10	3.1	UBRR= **11**	**0.0**	UBRR= **12**	**0.2**	UBRR= **14**	**0.0**
28800	UBRR= **6**	**1.6**	UBRR= **7**	**0.0**	UBRR= 8	3.7	UBRR= **9**	**0.0**
38400	UBRR= 4	6.3	UBRR= **5**	**0.0**	UBRR= 6	7.5	UBRR= 7	6.7
57600	UBRR= 3	12.5	UBRR= **3**	**0.0**	UBRR= 3	7.8	UBRR= **4**	**0.0**
76800	UBRR= 2	12.5	UBRR= **2**	**0.0**	UBRR= 2	7.8	UBRR= 3	6.7
115200	UBRR= 1	12.5	UBRR= **1**	**0.0**	UBRR= 1	7.8	UBRR= 2	20.0

Baud Rate	**7.3728 MHz**	%Error	**8 MHz**	%Error	**9.216 MHz**	%Error	**11.059 MHz**	%Error
2400	UBRR= **191**	**0.0**	UBRR= **207**	**0.2**	UBRR= **239**	**0.0**	UBRR= 287	-
4800	UBRR= **95**	**0.0**	UBRR= **103**	**0.2**	UBRR= **119**	**0.0**	UBRR= **143**	**0.0**
9600	UBRR= **47**	**0.0**	UBRR= **51**	**0.2**	UBRR= **59**	**0.0**	UBRR= **71**	**0.0**
14400	UBRR= **31**	**0.0**	UBRR= **34**	**0.8**	UBRR= **39**	**0.0**	UBRR= **47**	**0.0**
19200	UBRR= **23**	**0.0**	UBRR= **25**	**0.2**	UBRR= **29**	**0.0**	UBRR= **35**	**0.0**
28800	UBRR= **15**	**0.0**	UBRR= 16	2.1	UBRR= **19**	**0.0**	UBRR= **23**	**0.0**
38400	UBRR= **11**	**0.0**	UBRR= **12**	**0.2**	UBRR= **14**	**0.0**	UBRR= **17**	**0.0**
57600	UBRR= **7**	**0.0**	UBRR= 8	3.7	UBRR= **9**	**0.0**	UBRR= **11**	**0.0**
76800	UBRR= **5**	**0.0**	UBRR= 6	7.5	UBRR= 7	6.7	UBRR= **8**	**0.0**
115200	UBRR= **3**	**0.0**	UBRR= 3	7.8	UBRR= **4**	**0.0**	UBRR= **5**	**0.0**

UART BAUD Rate Register – UBRR

Bit	7	6	5	4	3	2	1	0	
$09 ($29)	**MSB**							**LSB**	**UBRR**
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	
Initial Value	0	0	0	0	0	0	0	0	

The UBRR register is an 8-bit read/write register that specifies the UART Baud Rate according to the equation on the previous page.

- **AIN0 – Port B, Bit 2**

AIN0: Analog Comparator Positive Input. When configured as an input (DDB2 is cleared [zero]) and with the internal MOS pull-up resistor switched off (PB2 is cleared [zero]), this pin also serves as the positive input of the On-chip Analog Comparator.

- **T1 – Port B, Bit 1**

T1: Timer/Counter1 counter source. See the timer description for further details

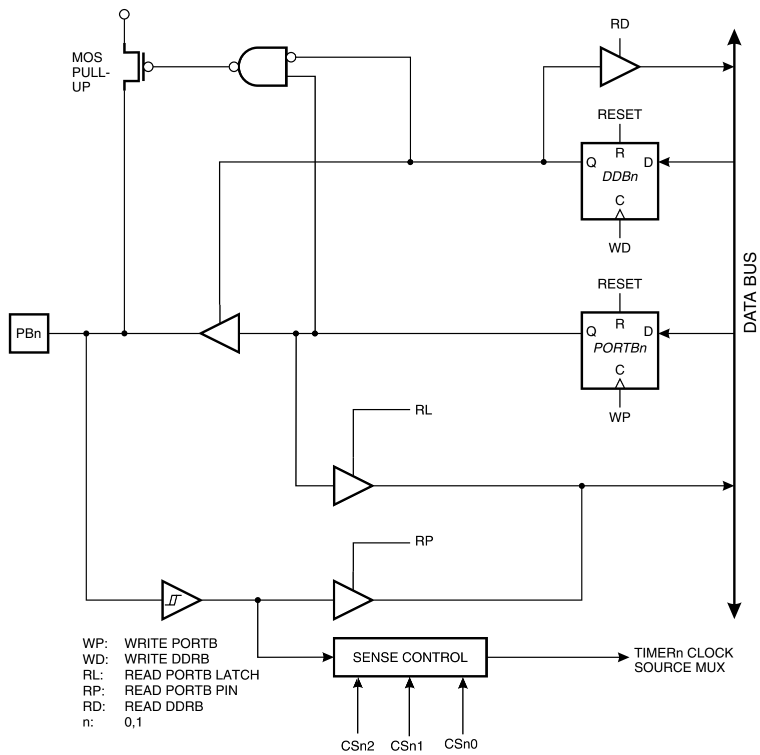
- **T0 – Port B, Bit 0**

T0: Timer/Counter0 counter source. See the timer description for further details.

Port B Schematics

Note that all port pins are synchronized. The synchronization latches are, however, not shown in the figures.

Figure 46. Port B Schematic Diagram (Pins PB0 and PB1)



- **INT1 – Port D, Bit 3**

INT1: External Interrupt source 1. The PD3 pin can serve as an external interrupt source to the MCU. See the interrupt description for further details and how to enable the source.

- **INT0 – Port D, Bit 2**

INT0: External Interrupt source 0. The PD2 pin can serve as an external interrupt source to the MCU. See the interrupt description for further details and how to enable the source.

- **TXD – Port D, Bit 1**

Transmit Data (data output pin for the UART). When the UART transmitter is enabled, this pin is configured as an output, regardless of the value of DDRD1.

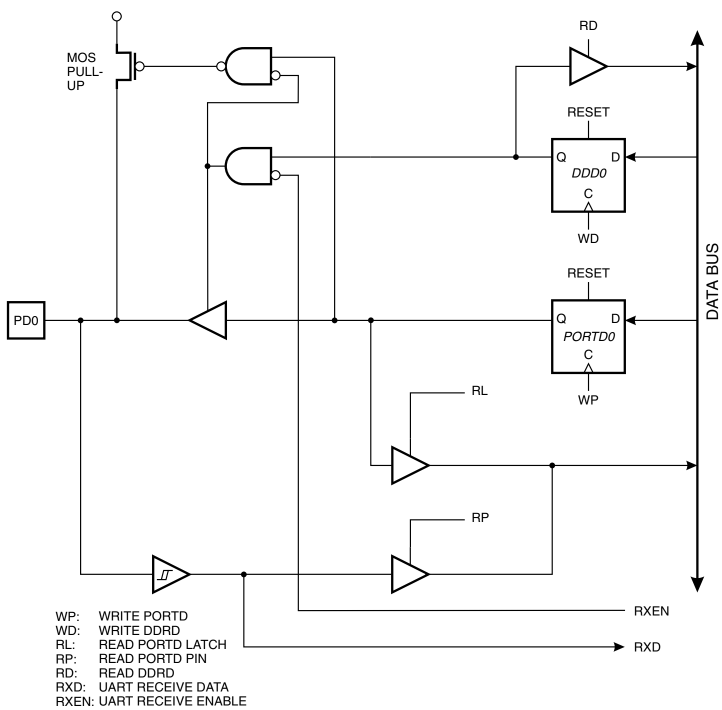
- **RXD – Port D, Bit 0**

Receive Data (data input pin for the UART). When the UART receiver is enabled, this pin is configured as an input, regardless of the value of DDRD0. When the UART forces this pin to be an input, a logical "1" in PORTD0 will turn on the internal pull-up.

Port D Schematics

Note that all port pins are synchronized. The synchronization latches are, however, not shown in the figures.

Figure 53. Port D Schematic Diagram (Pin PD0)



Bit 5 = SPIEN Fuse bit

Bit 0 = FSTRT Fuse bit

Bit 7 - 6, 4 - 1 = “1”. These bits are reserved and should be left unprogrammed (“1”).

3. Give $\overline{WR}$ a t_{WLWH_PFB}-wide negative pulse to execute the programming, t_{WLWH_PFB} is found in Table 30. Programming the Fuse bits does not generate any activity on the RDY/$\overline{BSY}$ pin.

Programming the Lock Bits

The algorithm for programming the Lock bits is as follows (refer to “Programming the Flash” on page 81 for details on command and data loading):

1. A: Load Command “0010 0000”.
2. D: Load Data Low Byte. Bit n = “0” programs the Lock bit.

   Bit 2 = Lock Bit2

   Bit 1 = Lock Bit1

   Bit 7 - 3, 0 = “1”. These bits are reserved and should be left unprogrammed (“1”).
3. E: Write Data Low Byte.

The Lock bits can only be cleared by executing Chip Erase.

Reading the Fuse and Lock Bits

The algorithm for reading the Fuse and Lock bits is as follows (refer to “Programming the Flash” on page 81 for details on Command loading):

1. A: Load Command “0000 0100”.
2. Set $\overline{OE}$ to “0”, and BS to “1”. The status of the Fuse and Lock bits can now be read at DATA (“0” means programmed).

   Bit 7 = Lock Bit1

   Bit 6 = Lock Bit2

   Bit 5 = SPIEN Fuse bit

   Bit 0 = FSTRT Fuse bit
3. Set $\overline{OE}$ to “1”.

Observe that BS needs to be set to “1”.

Reading the Signature Bytes

The algorithm for reading the signature bytes is as follows (refer to “Programming the Flash” on page 81 for details on command and address loading):

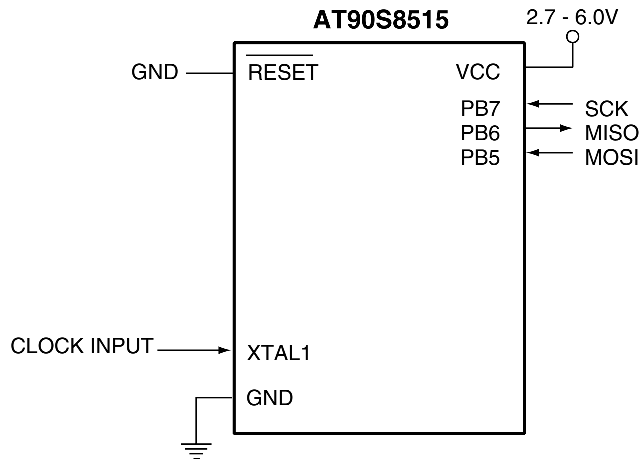
1. A: Load Command “0000 1000”.
2. C: Load Address Low Byte ($00 - $02).

   Set $\overline{OE}$ to “0”, and BS to “0”. The selected signature byte can now be read at DATA.
3. Set $\overline{OE}$ to “1”.

Serial Downloading

Both the program and data memory arrays can be programmed using the SPI bus while $\overline{\text{RESET}}$ is pulled to GND. The serial interface consists of pins SCK, MOSI (input) and MISO (output). See Figure 64. After $\overline{\text{RESET}}$ is set low, the Programming Enable instruction needs to be executed first before program/erase instructions can be executed.

Figure 64. Serial Programming and Verify



For the EEPROM, an auto-erase cycle is provided within the self-timed Write instruction and there is no need to first execute the Chip Erase instruction. The Chip Erase instruction turns the content of every memory location in both the program and EEPROM arrays into $FF.

The program and EEPROM memory arrays have separate address spaces: $0000 to $0FFF (AT90S8515) for program memory and $0000 to $01FF (AT90S8515) for EEPROM memory.

Either an external clock is supplied at pin XTAL1 or a crystal needs to be connected across pins XTAL1 and XTAL2. The minimum low and high periods for the serial clock (SCK) input are defined as follows:

Low: > 2 XTAL1 clock cycles

High: > 2 XTAL1 clock cycles

Serial Programming Algorithm

When writing serial data to the AT90S8515, data is clocked on the rising edge of SCK.

When reading data from the AT90S8515, data is clocked on the falling edge of SCK. See Figure 65, Figure 66 and Table 33 on page 89 for timing details.

To program and verify the AT90S8515 in the Serial Programming Mode, the following sequence is recommended (see 4-byte instruction formats in Table 32):

1. Power-up sequence:

   Apply power between V_{CC} and GND while $\overline{\text{RESET}}$ and SCK are set to "0". If a crystal is not connected across pins XTAL1 and XTAL2, apply a clock signal to the XTAL1 pin. In some systems, the programmer cannot guarantee that SCK is held low during power-up. In this case, $\overline{\text{RESET}}$ must be given a positive pulse of at least two XTAL1 cycles duration after SCK has been set to "0".
2. Wait for at least 20 ms and enable serial programming by sending the Programming Enable serial instruction to the MOSI (PB5) pin.
3. The serial programming instructions will not work if the communication is out of synchronization. When in sync, the second byte ($53) will echo back when issu-

Serial Programming Characteristics

Figure 66. Serial Programming Timing

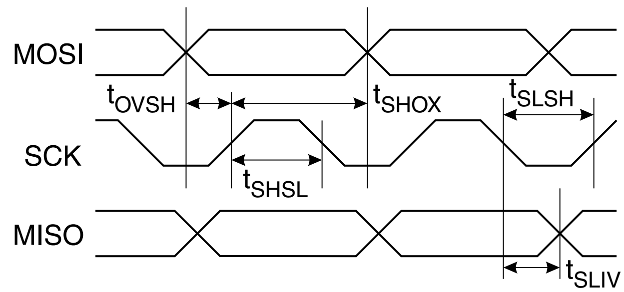


Table 33. Serial Programming Characteristics, T_A = -40°C to 85°C, V_{CC} = 2.7V - 6.0V (unless otherwise noted)

Symbol	Parameter	Min	Typ	Max	Units
$1/t_{CLCL}$	Oscillator Frequency (V_{CC} = 2.7 - 4.0V)	0		4.0	MHz
t_{CLCL}	Oscillator Period (V_{CC} = 2.7 - 4.0V)	250.0			ns
$1/t_{CLCL}$	Oscillator Frequency (V_{CC} = 4.0 - 6.0V)	0		8.0	MHz
t_{CLCL}	Oscillator Period (V_{CC} = 4.0 - 6.0V)	125.0			ns
t_{SHSL}	SCK Pulse Width High	2.0 t_{CLCL}			ns
t_{SLSH}	SCK Pulse Width Low	2.0 t_{CLCL}			ns
t_{OVSH}	MOSI Setup to SCK High	t_{CLCL}			ns
t_{SHOX}	MOSI Hold after SCK High	2.0 t_{CLCL}			ns
t_{SLIV}	SCK Low to MISO Valid	10.0	16.0	32.0	ns

Table 34. Minimum Wait Delay after the Chip Erase Instruction

Symbol	3.2V	3.6V	4.0V	5.0V
t_{WD_ERASE}	18 ms	14 ms	12 ms	8 ms

Table 35. Minimum Wait Delay after Writing a Flash or EEPROM Location

Symbol	3.2V	3.6V	4.0V	5.0V
t_{WD_PROG}	9 ms	7 ms	6 ms	4 ms

Typical Characteristics

The following charts show typical behavior. These figures are not tested during manufacturing. All current consumption measurements are performed with all I/O pins configured as inputs and with internal pull-ups enabled. ICP is pulled high externally. A sine wave generator with rail-to-rail output is used as clock source.

The power consumption in Power-down mode is independent of clock selection.

The current consumption is a function of several factors such as: operating voltage, operating frequency, loading of I/O pins, switching rate of I/O pins, code executed and ambient temperature. The dominating factors are operating voltage and frequency.

The current drawn from capacitive loaded pins may be estimated (for one pin) as $C_L \bullet V_{CC} \bullet f$ where C_L = load capacitance, V_{CC} = operating voltage and f = average switching frequency of I/O pin.

The parts are characterized at frequencies higher than test limits. Parts are not guaranteed to function properly at frequencies higher than the ordering code indicates.

The difference between current consumption in Power-down mode with Watchdog Timer enabled and Power-down mode with Watchdog Timer disabled represents the differential current drawn by the Watchdog Timer.

Figure 69. Active Supply Current vs. Frequency

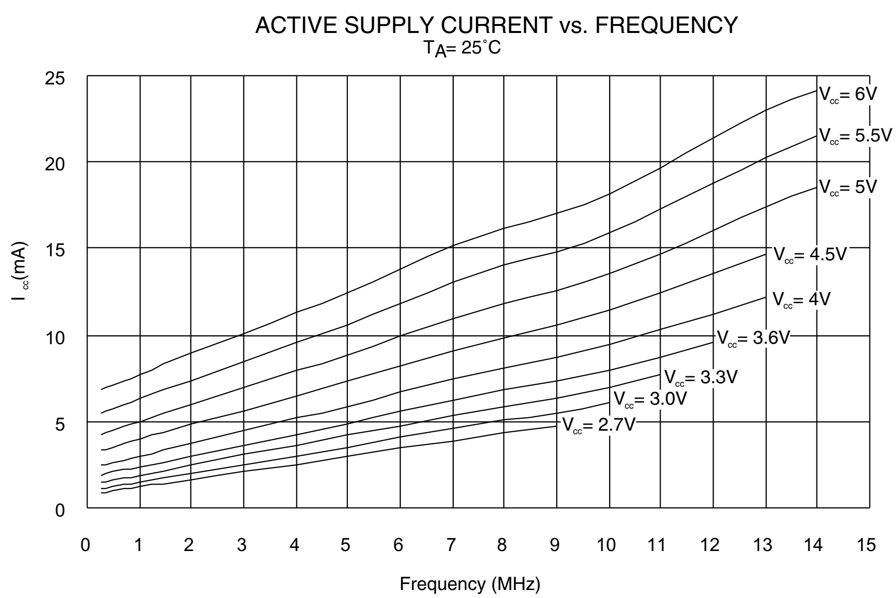


Figure 78. Analog Comparator Input Leakage Current

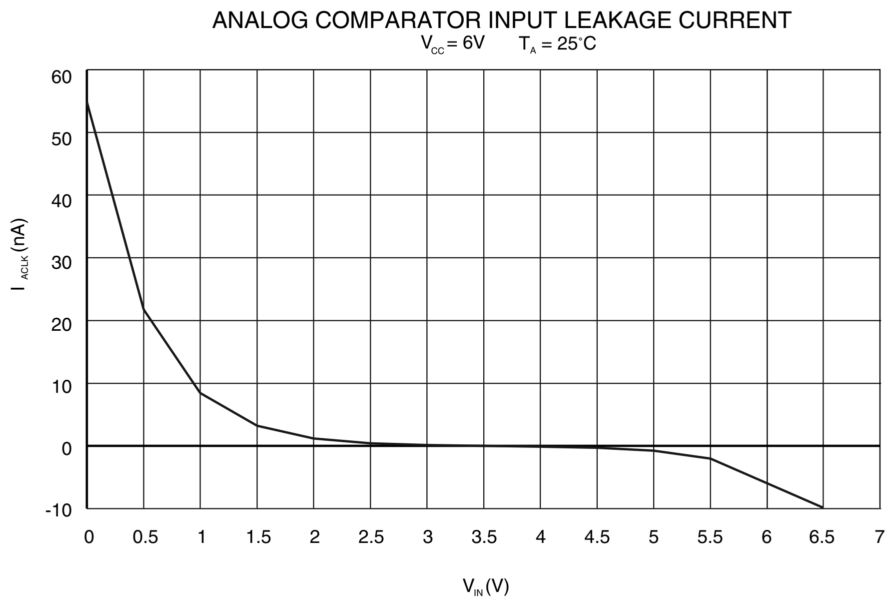


Figure 79. Watchdog Oscillator Frequency vs. V_{CC}

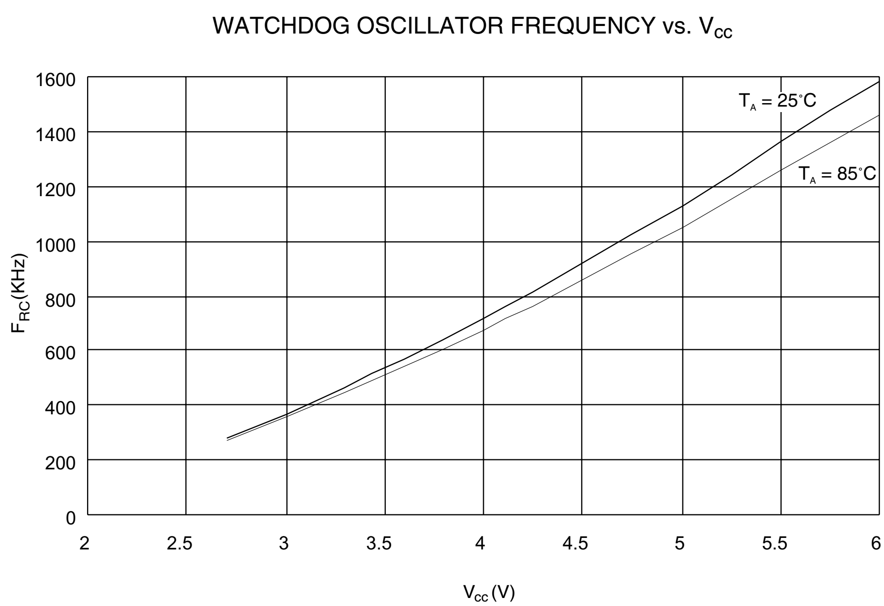


Figure 84. I/O Pin Source Current vs. Output Voltage

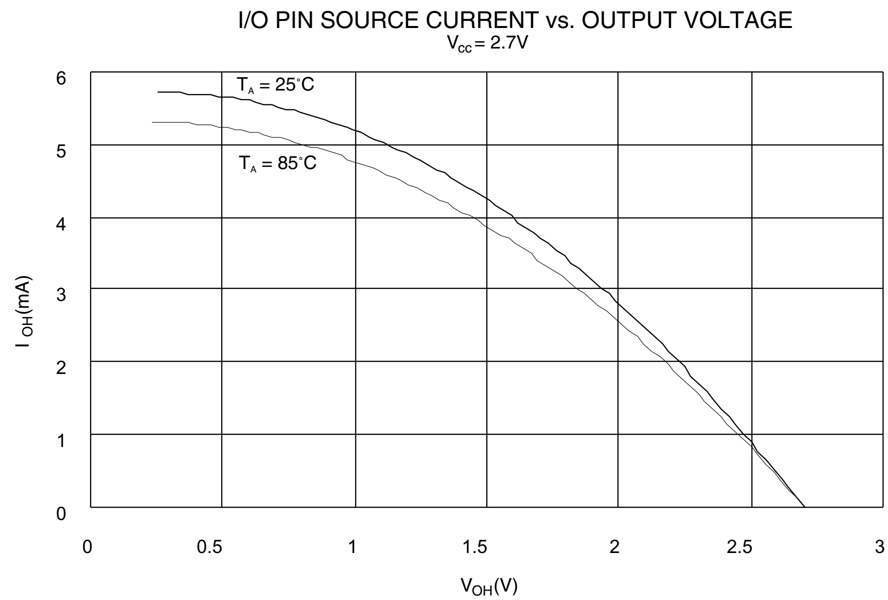
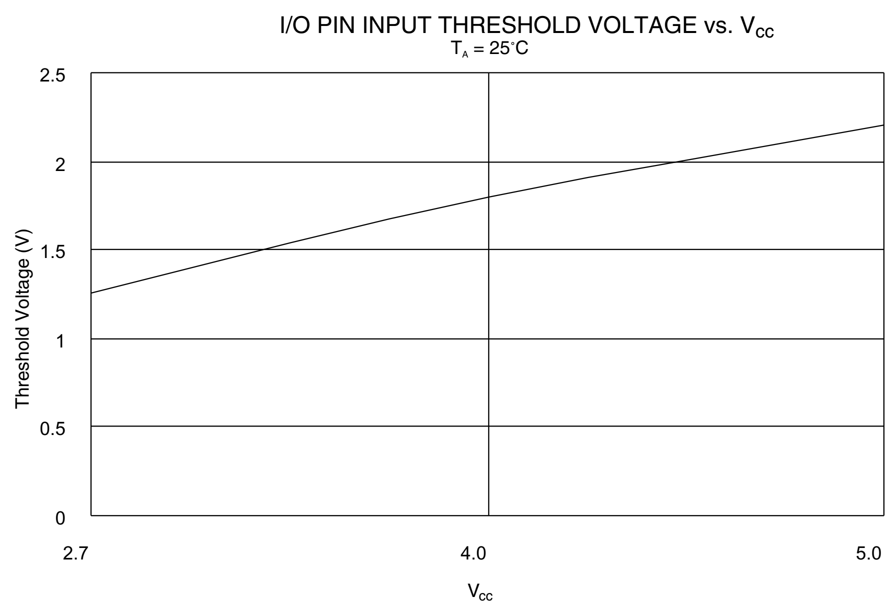
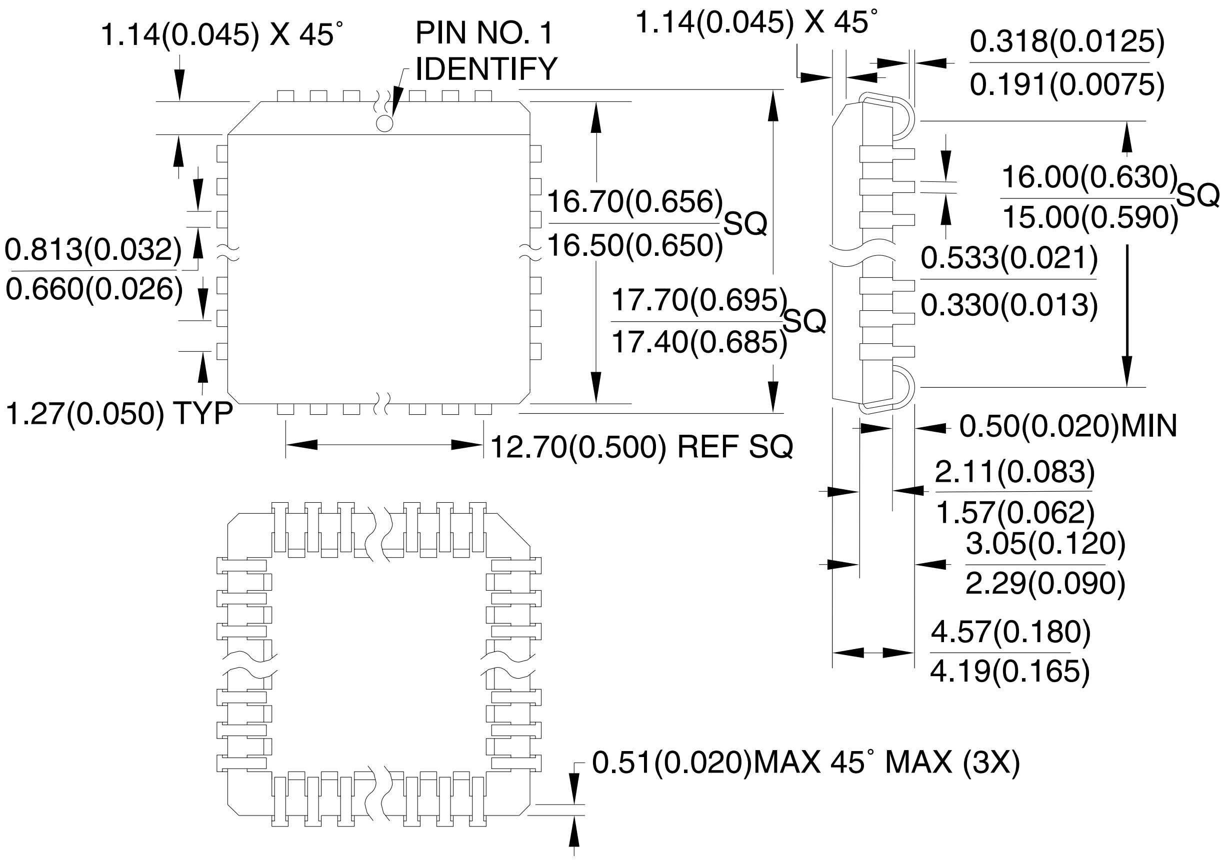


Figure 85. I/O Pin Input Threshold Voltage vs. V_{CC}



44J

44J, 44-lead, Plastic J-leaded Chip Carrier (PLCC)
Dimensions in Milimeters and (Inches)*
JEDEC STANDARD MS-018 AC



*Controlling dimensions: Inches

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