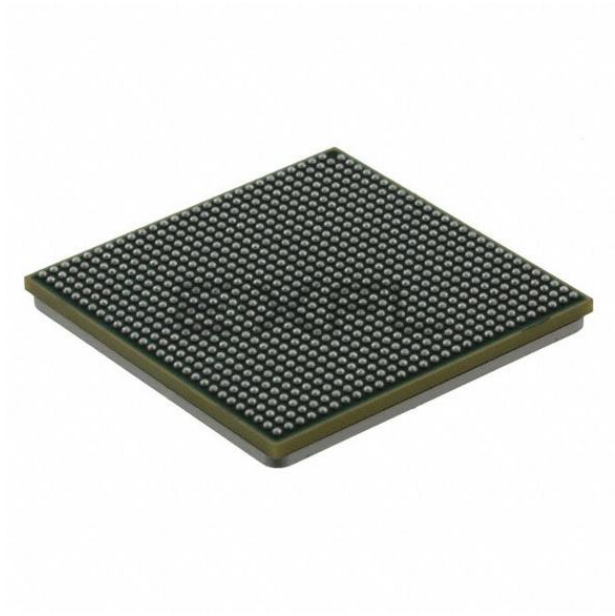

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Understanding Embedded - DSP (Digital Signal Processors)

Embedded - DSP (Digital Signal Processors) are specialized microprocessors designed to perform complex mathematical computations on digital signals in real-time. Unlike general-purpose processors, DSPs are optimized for high-speed numeric processing tasks, making them ideal for applications that require efficient and precise manipulation of digital data. These processors are fundamental in converting and processing signals in various forms, including audio, video, and communication signals, ensuring that data is accurately interpreted and utilized in embedded systems.

Applications of Embedded - DSP (Digital Signal Processors)

Details	
Product Status	Obsolete
Type	SC3850 Dual Core
Interface	Ethernet, I²C, PCI, RGMII, Serial RapidIO, SGMII, SPI, UART/USART
Clock Rate	1GHz
Non-Volatile Memory	ROM (96kB)
On-Chip RAM	576kB
Voltage - I/O	2.50V
Voltage - Core	1.00V
Operating Temperature	0°C ~ 105°C (TJ)
Mounting Type	Surface Mount
Package / Case	783-BBGA, FCBGA
Supplier Device Package	783-FCPBGA (29x29)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/msc8252svt1000b

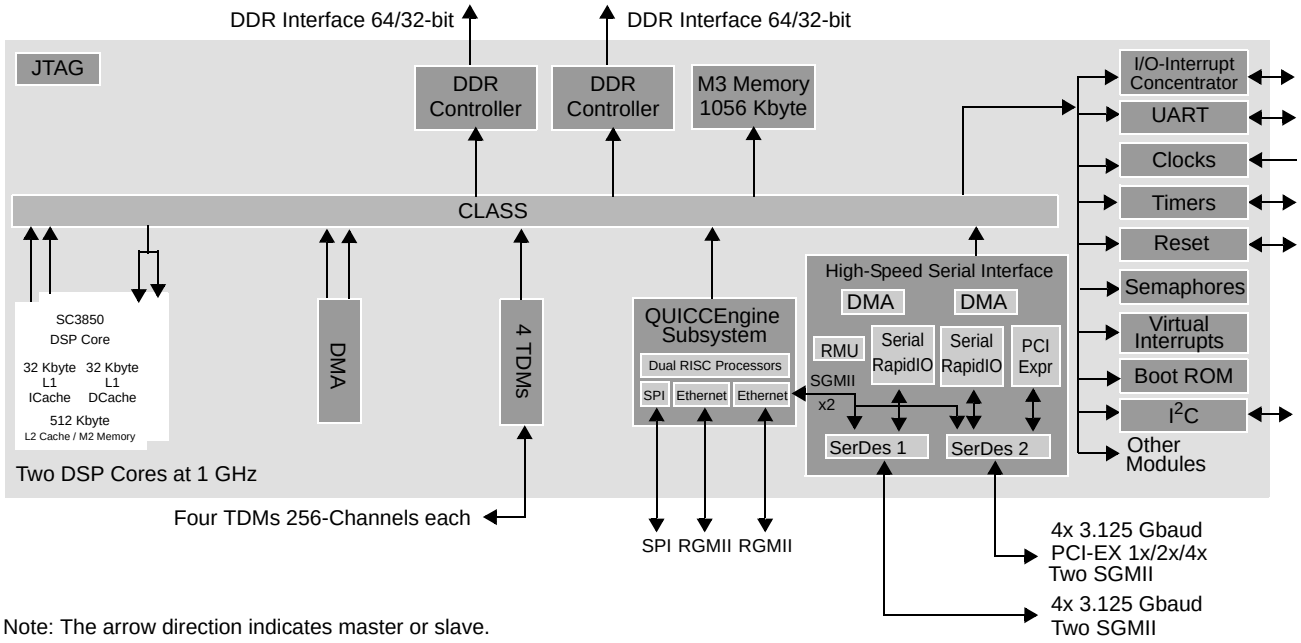


Figure 1. MSC8252 Block Diagram

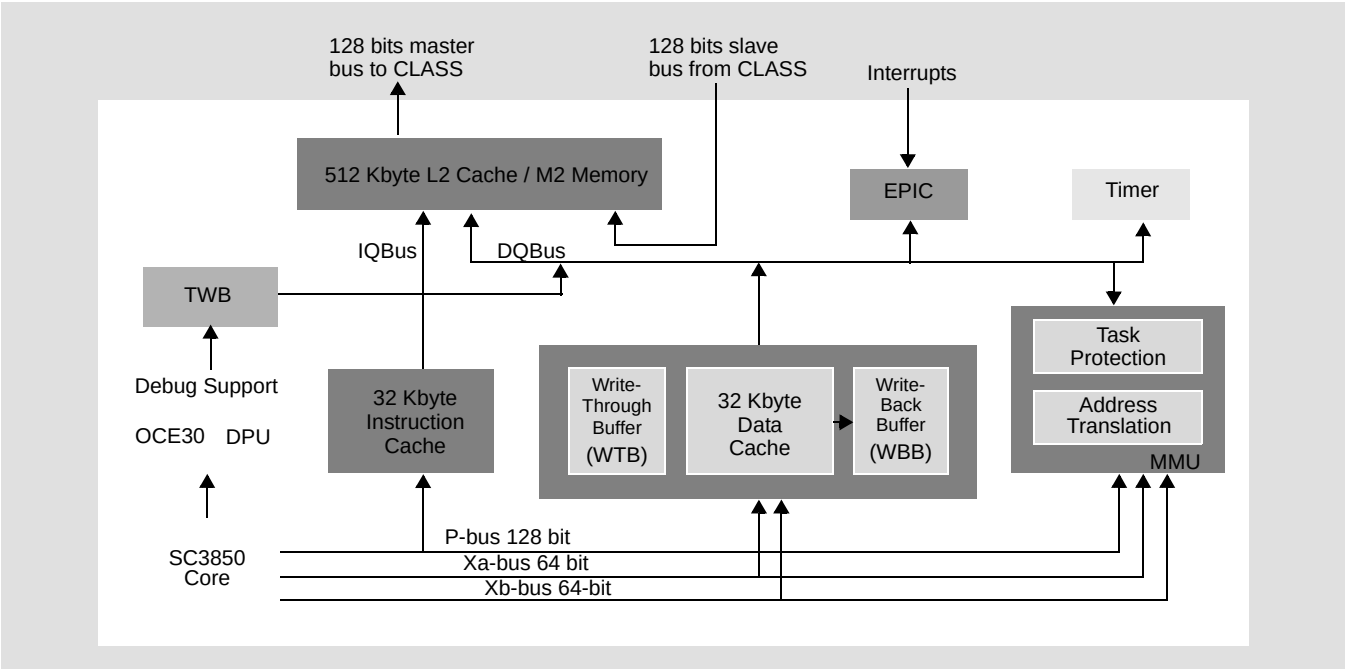


Figure 2. StarCore SC3850 DSP Subsystem Block Diagram

1 Pin Assignment

This section includes diagrams of the MSC8252 package ball grid array layouts and tables showing how the pinouts are allocated for the package.

1.1 FC-PBGA Ball Layout Diagram

The top view of the FC-PBGA package is shown in Figure 3 with the ball location index numbers.

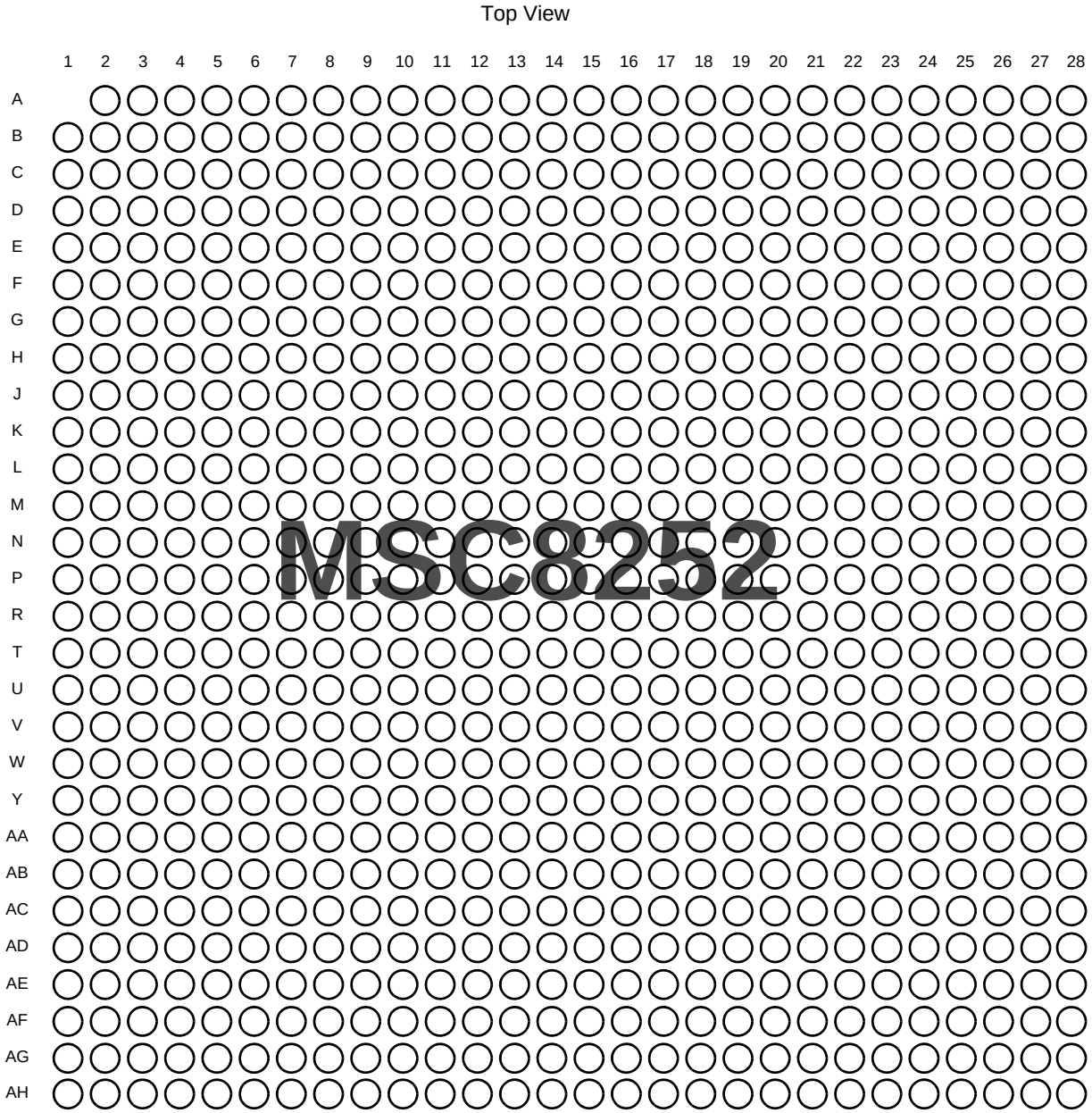


Figure 3. MSC8252 FC-PBGA Package, Top View

Table 1. Signal List by Ball Number (continued)

Ball Number	Signal Name[1,2]	Pin Type[10]	Power Rail Name
N23	SR2_TXD2/PE_TXD2/SG1_TX[4]	O	SXPVDD2
N24	$\overline{\text{SR2\_TXD2}}$/PE_TXD2/$\overline{\text{SG1\_TX}}$[4]	O	SXPVDD2
N25	SXCVDD2	Power	N/A
N26	SXCVSS2	Ground	N/A
N27	$\overline{\text{SR2\_RXD2}}$/$\overline{\text{PE\_RXD2}}$/$\overline{\text{SG1\_RX}}$[4]	I	SXCVDD2
N28	SR2_RXD2/PE_RXD2/SG1_RX[4]	I	SXCVDD2
P1	CLKIN	I	QVDD
P2	EE0	I	QVDD
P3	QVDD	Power	N/A
P4	VSS	Ground	N/A
P5	STOP_BS	I	QVDD
P6	QVDD	Power	N/A
P7	VSS	Ground	N/A
P8	PLL0_AVDD[9]	Power	VDD
P9	PLL2_AVDD[9]	Power	VDD
P10	VSS	Ground	N/A
P11	VDD	Power	N/A
P12	VSS	Ground	N/A
P13	VDD	Power	N/A
P14	VSS	Ground	N/A
P15	VSS	Ground	N/A
P16	VSS	Ground	N/A
P17	VSS	Ground	N/A
P18	VSS	Ground	N/A
P19	VDD	Power	N/A
P20	Reserved	NC	—
P21	Reserved	NC	—
P22	Reserved	NC	—
P23	SXPVDD2	Power	N/A
P24	SXPVSS2	Ground	N/A
P25	SR2_PLL_AGND[9]	Ground	SXCVSS2
P26	SR2_PLL_AVDD[9]	Power	SXCVDD2
P27	SXCVSS2	Ground	N/A
P28	SXCVDD2	Power	N/A
R1	VSS	Ground	N/A
R2	$\overline{\text{NMI}}$	I	QVDD
R3	$\overline{\text{NMI\_OUT}}$[6]	O	QVDD
R4	$\overline{\text{HRESET}}$[6,7]	I/O	QVDD
R5	$\overline{\text{INT\_OUT}}$[6]	O	QVDD
R6	EE1	O	QVDD
R7	VSS	Ground	N/A
R8	PLL1_AVDD[9]	Power	VDD
R9	VSS	Ground	N/A
R10	VDD	Power	N/A
R11	VSS	Non-user	N/A
R12	VDD	Power	N/A

Table 1. Signal List by Ball Number (continued)

Ball Number	Signal Name[1,2]	Pin Type[10]	Power Rail Name
V21	RCW_LSEL_3/RC20	I/O	NVDD
V22	RCW_LSEL_2/RC19	I/O	NVDD
V23	SXPVDD2	Power	N/A
V24	SXPVSS2	Ground	N/A
V25	RCW_LSEL_1/RC18	I/O	NVDD
V26	RC21	I	NVDD
V27	SXCVDD2	Power	N/A
V28	SXCVSS2	Ground	N/A
W1	VSS	Ground	N/A
W2	GVDD1	Power	N/A
W3	M1DM1	O	GVDD1
W4	VSS	Ground	N/A
W5	GVDD1	Power	N/A
W6	M1DQ0	I/O	GVDD1
W7	VSS	Ground	N/A
W8	GVDD1	Power	N/A
W9	M1DQ5	I/O	GVDD1
W10	VDD	Power	N/A
W11	VSS	Ground	N/A
W12	VDD	Power	N/A
W13	VSS	Ground	N/A
W14	VDD	Power	N/A
W15	VSS	Ground	N/A
W16	VDD	Power	N/A
W17	VSS	Ground	N/A
W18	VDD	Power	N/A
W19	VSS	Ground	N/A
W20	VSS	Ground	N/A
W21	RCW_LSEL0/RC17	I/O	NVDD
W22	GPIO19/SPI_MISO[5,8]	I/O	NVDD
W23	VSS	Ground	N/A
W24	NVDD	Power	N/A
W25	GPIO11/IRQ11/RC11[5,8]	I/O	NVDD
W26	GPIO3/DRQ1/IRQ3/RC3[5,8]	I/O	NVDD
W27	GPIO7/IRQ7/RC7[5,8]	I/O	NVDD
W28	GPIO2/IRQ2/RC2[5,8]	I/O	NVDD
Y1	M1DQS1	I/O	GVDD1
Y2	M1DQS1	I/O	GVDD1
Y3	M1DQ10	I/O	GVDD1
Y4	M1DQ11	I/O	GVDD1
Y5	M1DQ14	I/O	GVDD1
Y6	M1DQ23	I/O	GVDD1
Y7	M1ODT0	O	GVDD1
Y8	M1A12	O	GVDD1
Y9	M1A14	O	GVDD1
Y10	VSS	Ground	N/A

Using this waveform, the definitions are listed in Table 10. To simplify the illustration, the definitions assume that the SerDes transmitter and receiver operate in a fully symmetrical differential signalling environment.

Table 10. Differential Signal Definitions

Term	Definition
Single-Ended Swing	The transmitter output signals and the receiver input signals SR[1–2]_TX, $\overline{\text{SR[1–2]\_TX}}$, SR[1–2]_RX and $\overline{\text{SR[1–2]\_RX}}$ each have a peak-to-peak swing of A – B volts. This is also referred to as each signal wire's single-ended swing.
Differential Output Voltage, V_{OD} (or **Differential Output Swing**):	The differential output voltage (or swing) of the transmitter, V_{OD}, is defined as the difference of the two complimentary output voltages: $V_{SR[1–2]\_TX} - V_{\overline{SR[1–2]\_TX}}$. The V_{OD} value can be either positive or negative.
Differential Input Voltage, V_{ID} (or **Differential Input Swing**)	The differential input voltage (or swing) of the receiver, V_{ID}, is defined as the difference of the two complimentary input voltages: $V_{SR[1–2]\_RX} - V_{\overline{SR[1–2]\_RX}}$. The V_{ID} value can be either positive or negative.
Differential Peak Voltage, V_{DIFFp}	The peak value of the differential transmitter output signal or the differential receiver input signal is defined as the differential peak voltage, $V_{DIFFp} = \lvert A - B \rvert$ volts.
Differential Peak-to-Peak, $V_{DIFFp-p}$	Since the differential output signal of the transmitter and the differential input signal of the receiver each range from A – B to –(A – B) volts, the peak-to-peak value of the differential transmitter output signal or the differential receiver input signal is defined as differential peak-to-peak voltage, $V_{DIFFp-p} = 2 \times V_{DIFFp} = 2 \times \lvert (A - B) \rvert$ volts, which is twice the differential swing in amplitude, or twice of the differential peak. For example, the output differential peak-peak voltage can also be calculated as $V_{TX-DIFFp-p} = 2 \times \lvert V_{OD} \rvert$.
Differential Waveform	The differential waveform is constructed by subtracting the inverting signal ($\overline{\text{SR[1–2]\_TX}}$, for example) from the non-inverting signal (SR[1–2]_TX, for example) within a differential pair. There is only one signal trace curve in a differential waveform. The voltage represented in the differential waveform is not referenced to ground. Refer to Figure 16 as an example for differential waveform.
Common Mode Voltage, V_{cm}	The common mode voltage is equal to half of the sum of the voltages between each conductor of a balanced interchange circuit and ground. In this example, for SerDes output, $V_{cm\_out} = (V_{SR[1–2]\_TX} + V_{\overline{SR[1–2]\_TX}}) \div 2 = (A + B) \div 2$, which is the arithmetic mean of the two complimentary output voltages within a differential pair. In a system, the common mode voltage may often differ from one component's output to the other's input. It may be different between the receiver input and driver output circuits within the same component. It is also referred to as the DC offset on some occasions.

To illustrate these definitions using real values, consider the example of a current mode logic (CML) transmitter that has a common mode voltage of 2.25 V and outputs, TD and $\overline{\text{TD}}$. If these outputs have a swing from 2.0 V to 2.5 V, the peak-to-peak voltage swing of each signal (TD or $\overline{\text{TD}}$) is 500 mV p-p, which is referred to as the single-ended swing for each signal. Because the differential signaling environment is fully symmetrical in this example, the transmitter output differential swing (V_{OD}) has the same amplitude as each signal single-ended swing. The differential output signal ranges between 500 mV and –500 mV. In other words, V_{OD} is 500 mV in one phase and –500 mV in the other phase. The peak differential voltage (V_{DIFFp}) is 500 mV. The peak-to-peak differential voltage ($V_{DIFFp-p}$) is 1000 mV p-p.

Note: Specifications are valid at the recommended operating conditions listed in Table 3.

Table 11. PCI Express (2.5 Gbps) Differential Transmitter (Tx) Output DC Specifications

Parameter	Symbol	Min	Typical	Max	Units	Notes
Differential peak-to-peak output voltage	$V_{TX-DIFFp-p}$	800	1000	1200	mV	1
De-emphasized differential output voltage (ratio)	$V_{TX-DE-RATIO}$	3.0	3.5	4.0	dB	2
DC differential Tx impedance	$Z_{TX-DIFF-DC}$	80	100	120	Ω	3
Transmitter DC impedance	Z_{TX-DC}	40	50	60	Ω	4

Notes:
1. $V_{TX-DIFFp-p} = 2 \times |V_{TX-D+} - V_{TX-D-}|$ Measured at the package pins with a test load of 50 Ω to GND on each pin.
2. Ratio of the $V_{TX-DIFFp-p}$ of the second and following bits after a transition divided by the $V_{TX-DIFFp-p}$ of the first bit after a transition. Measured at the package pins with a test load of 50 Ω to GND on each pin.
3. Tx DC differential mode low impedance
4. Required Tx D+ as well as D– DC Impedance during all states

Table 12. PCI Express (2.5 Gbps) Differential Receiver (Rx) Input DC Specifications

Parameter	Symbol	Min	Typical	Max	Units	Notes
Differential input peak-to-peak voltage	$V_{RX-DIFFp-p}$	120	1000	1200	mV	1
DC differential Input Impedance	$Z_{RX-DIFF-DC}$	80	100	120	Ω	2
DC input impedance	Z_{RX-DC}	40	50	60	Ω	3
Powered down DC input impedance	$Z_{RX-HIGH-IMP-DC}$	50	—	—	KΩ	4
Electrical idle detect threshold	$V_{RX-IDLE-DET-DIFFp-p}$	65	—	175	mV	5

Notes:
1. $V_{RX-DIFFp-p} = 2 \times |V_{RX-D+} - V_{RX-D-}|$ Measured at the package pins with a test load of 50 Ω to GND on each pin.
2. Rx DC differential mode impedance. Impedance during all LTSSM states. When transitioning from a fundamental reset to detect (the initial state of the LTSSM), there is a 5 ms transition time before the receiver termination values must be met on all unconfigured lanes of a port.
3. Required Rx D+ as well as D– DC Impedance (50 ±20% tolerance). Measured at the package pins with a test load of 50 Ω to GND on each pin. Impedance during all LTSSM states. When transitioning from a fundamental reset to detect (the initial state of the LTSSM), there is a 5 ms transition time before the receiver termination values must be met on all unconfigured lanes of a port.
4. Required Rx D+ as well as D– DC Impedance when the receiver terminations do not have power. The Rx DC common mode impedance that exists when no power is present or fundamental reset is asserted. This helps ensure that the receiver detect circuit does not falsely assume a receiver is powered on when it is not. This term must be measured at 300 mV above the Rx ground.
5. $V_{RX-IDLE-DET-DIFFp-p} = 2 \times |V_{RX-D+} - V_{RX-D-}|$. Measured at the package pins of the receiver

2.5.3.3 DC-Level Requirements for Serial RapidIO Configurations

This sections provided various DC-level requirements for Serial RapidIO Configurations.

Note: Specifications are valid at the recommended operating conditions listed in Table 3.

Table 13. Serial RapidIO Transmitter DC Specifications

Parameter	Symbol	Min	Typical	Max	Units	Notes
Output voltage	V_O	–0.40	—	2.30	V	1
Long run differential output voltage	V_{DIFFPP}	800	—	1600	mVp-p	—
Short run differential output voltage	V_{DIFFPP}	500	—	1000	mVp-p	—

Note: Voltage relative to COMMON of either signal comprising a differential pair.

Table 14. Serial RapidIO Receiver DC Specifications

Parameter	Symbol	Min	Typical	Max	Units	Notes
Differential input voltage	V_{IN}	200	—	1600	mVp-p	1

Notes: **1.** Measured at receiver.

2.5.3.4 DC-Level Requirements for SGMII Configurations

Note: Specifications are valid at the recommended operating conditions listed in Table 3

Table 15 describes the SGMII SerDes transmitter AC-coupled DC electrical characteristics. Transmitter DC characteristics are measured at the transmitter outputs (SR[1–2]_TX[n] and $\overline{\text{SR[1–2]\_TX[n]}}$) as shown in Figure 10.

Table 15. SGMII DC Transmitter Electrical Characteristics

Parameter	Symbol	Min	Typ	Max	Unit	Notes
Output high voltage	V_{OH}	—	—	$XV_{DD\_SRDS\text{-}Typ}/2 + \lvert V_{OD}\rvert_{\text{-}max}/2$	mV	1
Output low voltage	V_{OL}	$XV_{DD\_SRDS\text{-}Typ}/2 - \lvert V_{OD}\rvert_{\text{-}max}/2$	—	—	mV	1
Output differential voltage ($XV_{DD\text{-}Typ}$ at 1.0 V)	$\lvert V_{OD}\rvert$	323	500	725	mV	2,3,4
		296	459	665		2,3,5
		269	417	604		2,3,6
		243	376	545		2,3,7
		215	333	483		2,3,8
		189	292	424		2,3,9
		162	250	362		2,3,10
Output impedance (single-ended)	R_O	40	50	60	Ω	—

Notes:
1. This does not align to DC-coupled SGMII. $XV_{DD\_SRDS2\text{-}Typ}$ = 1.1 V.
2. The $\lvert V_{OD}\rvert$ value shown in the table assumes full multitude by setting srd_smit_lvl as 000 and the following transmit equalization setting in the XMITEQ**AB** (for lanes A and B) or XMITEQ**EF** (for lanes E and F) bit field of Control Register:
• The MSB (bit 0) of the above bit field is set to zero (selecting the full $V_{DD\text{-}DIFF\text{-}p\text{-}p}$ amplitude which is power up default);
• The LSB (bit [1–3]) of the above bit field is set based on the equalization settings listed in notes 4 through 10.
3. The $\lvert V_{OD}\rvert$ value shown in the Typ column is based on the condition of $XV_{DD\_SRDS2\text{-}Typ}$ = 1.0 V, no common mode offset variation (V_{OS} =500mV), SerDes transmitter is terminated with 100-Ω differential load between
4. Equalization setting: 1.0x: 0000.
5. Equalization setting: 1.09x: 1000.
6. Equalization setting: 1.2x: 0100.
7. Equalization setting: 1.33x: 1100.
8. Equalization setting: 1.5x: 0010.
9. Equalization setting: 1.71x: 1010.
10. Equalization setting: 2.0x: 0110.
11. $\lvert V_{OD}\rvert = \lvert V_{SR[1–2]\_TXn} - V_{\overline{SR[1–2]\_TXn}}\rvert$. $\lvert V_{OD}\rvert$ is also referred to as output differential peak voltage. $V_{TX\text{-}DIFFp\text{-}p} = 2*\lvert V_{OD}\rvert$.

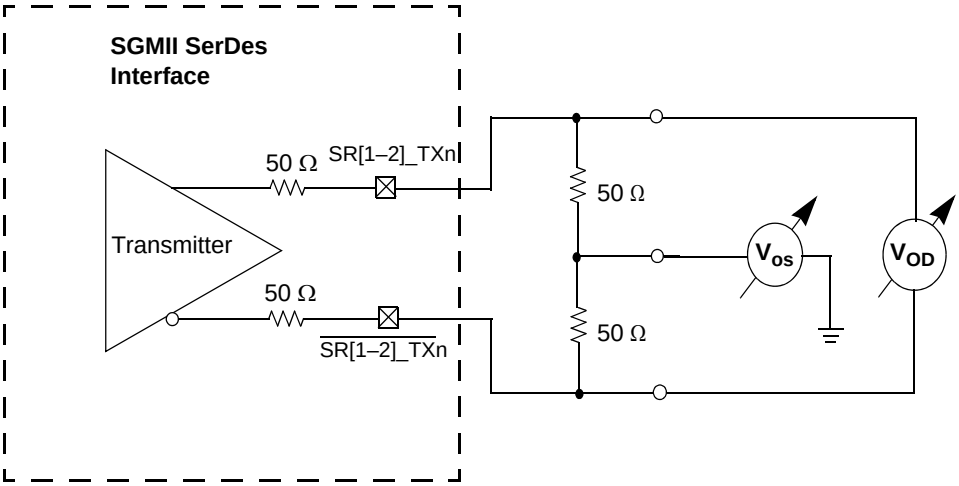


Figure 10. SGMII Transmitter DC Measurement Circuit

Table 16 describes the SGMII SerDes receiver AC-coupled DC electrical characteristics.

Table 16. SGMII DC Receiver Electrical Characteristics[5]

Parameter		Symbol	Min	Typ	Max	Unit	Notes
DC Input voltage range		—		N/A		—	1
Input differential voltage	SRDSnCR4[EICE{12:10}] = 0b001 for SGMII1 SRDSnCR4[EICF{4:2}] = 0b001 for SGMII2	$V_{RX\_DIFFp\text{-}p}$	100	—	1200	mV	2, 4
	SRDSnCR4[EICE{12:10}] = 0b100 for SGMII1 SRDSnCR4[EICF{4:2}] = 0b100 for SGMII2		175	—			
Loss of signal threshold	SRDSnCR4[EICE{12:10}] = 0b001 for SGMII1 SRDSnCR4[EICF{4:2}] = 0b001 for SGMII2	VLOS	30	—	100	mV	3, 4
	SRDSnCR4[EICE{12:10}] = 0b100 for SGMII1 SRDSnCR4[EICF{4:2}] = 0b100 for SGMII2		65	—	175		
Receiver differential input impedance		Z_{RX_DIFF}	80	—	120	W	—

Notes:
1. Input must be externally AC-coupled.
2. $V_{RX\_DIFFp\text{-}p}$ is also referred to as peak-to-peak input differential voltage.
3. The concept of this parameter is equivalent to the Electrical Idle Detect Threshold parameter in the PCI Express interface. Refer to the PCI Express Differential Receiver (RX) Input Specifications section of the *PCI Express Specification* document. for details.
4. The values for SGMII1 and SGMII2 are selected in the SRDS control registers.
5. The supply voltage is 1.0 V.

Figure 13 shows the DDR SDRAM output timing diagram.

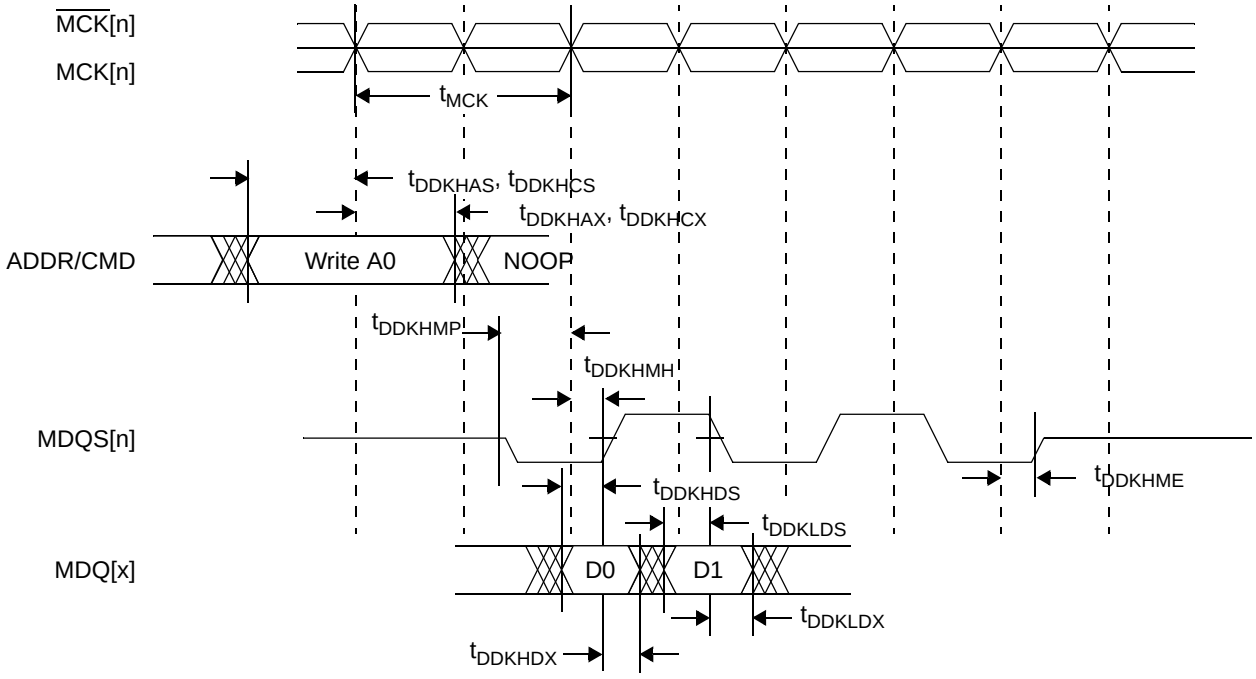


Figure 13. DDR SDRAM Output Timing

Figure 14 provides the AC test load for the DDR2 and DDR3 controller bus.

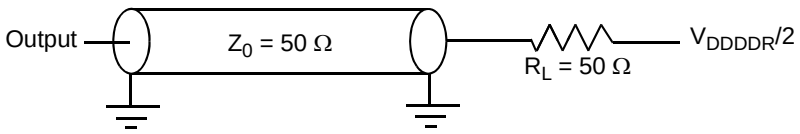


Figure 14. DDR2 and DDR3 Controller Bus AC Test Load

2.6.1.3 DDR2 and DDR3 SDRAM Differential Timing Specifications

This section describes the DC and AC differential timing specifications for the DDR2 and DDR3 SDRAM controller interface. Figure 15 shows the differential timing specification.

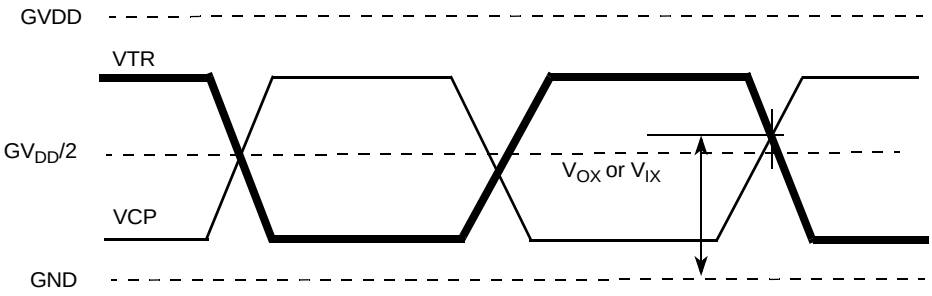


Figure 15. DDR2 and DDR3 SDRAM Differential Timing Specifications

Note: VTR specifies the true input signal (such as MCK or MDQS) and VCP is the complementary input signal (such as $\overline{\text{MCK}}$ or $\overline{\text{MDQS}}$).

2.6.2.2 PCI Express AC Physical Layer Specifications

The AC requirements for PCI Express implementations have separate requirements for the Tx and Rx lines. The MSC8252 supports a 2.5 Gbps PCI Express interface defined by the *PCI Express Base Specification, Revision 1.0a*. The transmitter specifications are defined in Table 25 and the receiver specifications are defined in Table 26. The parameters are specified at the component pins. the AC timing specifications do not include REF_CLK jitter.

Note: Specifications are valid at the recommended operating conditions listed in Table 3.

Table 25. PCI Express (2.5 Gbps) Differential Transmitter (Tx) Output AC Specifications

Parameter	Symbol	Min	Typical	Max	Units	Notes
Unit interval	UI	399.88	400.00	400.12	ps	1
Minimum Tx eye width	T_{TX-EYE}	0.70	—	—	UI	2, 3
Maximum time between the jitter median and maximum deviation from the median.	$T_{TX-EYE-MEDIAN-to-MAX-JITTER}$	—	—	0.15	UI	3, 4
AC coupling capacitor	C_{TX}	75	—	200	nF	5

Notes:
1. Each UI is 400 ps ± 300 ppm. UI does not account for spread spectrum clock dictated variations. No test load is necessarily associated with this value.
2. The maximum transmitter jitter can be derived as $T_{TX-MAX-JITTER} = 1 – T_{TX-EYE} = 0.3$ UI.
3. Specified at the measurement point into a timing and voltage compliance test load as shown in Figure 8 and measured over any 250 consecutive Tx UIs. A $T_{TX-EYE} = 0.70$ UI provides for a total sum of deterministic and random jitter budget of $T_{TX-JITTER-MAX} = 0.30$ UI for the transmitter collected over any 250 consecutive Tx UIs. The $T_{TX-EYE-MEDIAN-to-MAX-JITTER}$ median is less than half of the total Tx jitter budget collected over any 250 consecutive Tx UIs. It should be noted that the median is not the same as the mean. The jitter median describes the point in time where the number of jitter points on either side is approximately equal as opposed to the averaged time value. Jitter is defined as the measurement variation of the crossing points ($V_{TX-DIFFp-p} = 0$ V) in relation to a recovered Tx UI. A recovered Tx UI is calculated over 3500 consecutive unit intervals of sample data.
4. Jitter is measured using all edges of the 250 consecutive UI in the center of the 3500 UI used for calculating the Tx UI.
5. All transmitters shall be AC-coupled. The AC coupling is required either within the media or within the transmitting component itself. The SerDes transmitter does not have built-in Tx capacitance. An external AC coupling capacitor is required.

Table 26. PCI Express (2.5 Gbps) Differential Receiver (Rx) Input AC Specifications

Parameter	Symbol	Min	Typical	Max	Units	Notes
Unit Interval	UI	399.88	400.00	400.12	ps	1
Minimum receiver eye width	T_{RX-EYE}	0.4	—	—	UI	2, 3, 4
Maximum time between the jitter median and maximum deviation from the median.	$T_{RX-EYE-MEDIAN-to-MAX-JITTER}$	—	—	0.3	UI	3, 4, 5

Notes:
1. Each UI is 400 ps ± 300 ppm. UI does not account for spread spectrum clock dictated variations. No test load is necessarily associated with this value.
2. The maximum interconnect media and transmitter jitter that can be tolerated by the receiver can be derived as $T_{RX-MAX-JITTER} = 1 – T_{RX-EYE} = 0.6$ UI.
3. Specified at the measurement point and measured over any 250 consecutive UIs. The test load in Figure 8 should be used as the Rx device when taking measurements. If the clocks to the Rx and Tx are not derived from the same reference clock, the Tx UI recovered from 3500 consecutive UI must be used as a reference for the eye diagram.
4. A $T_{RX-EYE} = 0.40$ UI provides for a total sum of 0.60 UI deterministic and random jitter budget for the transmitter and interconnect collected any 250 consecutive UIs. The $T_{RX-EYE-MEDIAN-to-MAX-JITTER}$ specification ensures a jitter distribution in which the median and the maximum deviation from the median is less than half of the total. UI jitter budget collected over any 250 consecutive Tx UIs. It should be noted that the median is not the same as the mean. The jitter median describes the point in time where the number of jitter points on either side is approximately equal as opposed to the averaged time value. If the clocks to the Rx and Tx are not derived from the same reference clock, the Tx UI recovered from 3500 consecutive UI must be used as the reference for the eye diagram.
5. Jitter is defined as the measurement variation of the crossing points ($V_{RX-DIFFp-p} = 0$ V) in relation to a recovered Tx UI. A recovered Tx UI is calculated over 3500 consecutive unit intervals of sample data. Jitter is measured using all edges of the 250 consecutive UI in the center of the 3500 UI used for calculating the Tx UI. It is recommended that the recovered Tx UI is calculated using all edges in the 3500 consecutive UI interval with a fit algorithm using a minimization merit function. Least squares and median deviation fits have worked well with experimental and simulated data.

2.6.5 Ethernet Timing

This section describes the AC electrical characteristics for the Ethernet interface.

There are programmable delay units (PDU) that should be programmed differently for each interface to meet timing. There is a general configuration register 4 (GCR4) used to configure the timing. For additional information, see the *MSC8252 Reference Manual*.

2.6.5.1 Management Interface Timing

Table 33 lists the timer input Ethernet controller management interface timing specifications shown in Figure 24.

Table 33. Ethernet Controller Management Interface Timing

Characteristics	Symbol	Min	Max	Unit
GE_MDC frequency	f_{MDC}	—	2.5	MHz
GE_MDC period	t_{MDC}	400	—	ns
GE_MDC clock pulse width high	t_{MDC_H}	160	—	ns
GE_MDC clock pulse width low	t_{MDC_L}	160	—	ns
GE_MDC to GE_MDIO delay[2]	t_{MDKHDX}	10	70	ns
GE_MDIO to GE_MDC rising edge setup time	t_{MDDVKH}	20	—	ns
GE_MDC rising edge to GE_MDIO hold time	t_{MDDXKH}	0	—	ns

Notes:
1. Program the GE_MDC frequency (f_{MDC}) to a maximum value of 2.5 MHz (400 ns period for t_{MDC}). The value depends on the source clock and configuration of MIIMCFG[MCS] and UPSMR[MDCP]. For example, for a source clock of 400 MHz to achieve f_{MDC} = 2.5 MHz, program MIIMCFG[MCS] = 0x4 and UPSMR[MDCP] = 0. See the *MSC8252 Reference Manual* for configuration details.
2. The value depends on the source clock. For example, for a source clock of 267 MHz, the delay is 70 ns. For a source clock of 333 MHz, the delay is 58 ns.

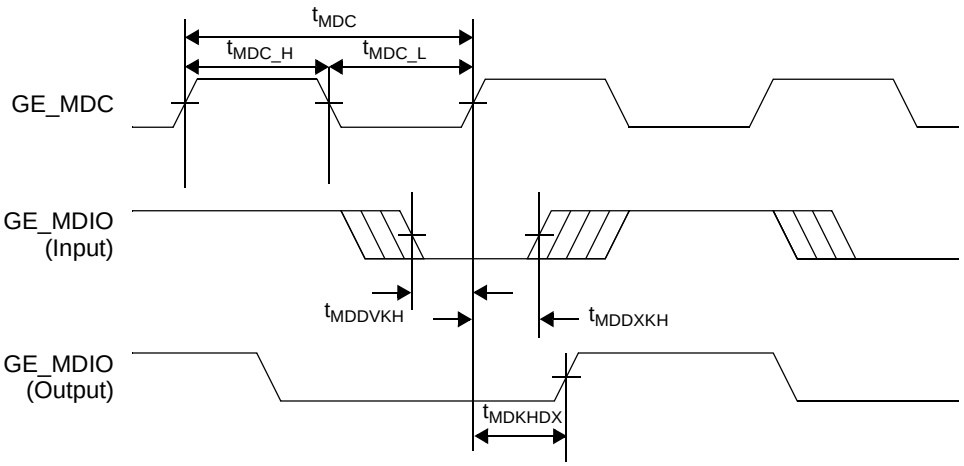


Figure 24. MII Management Interface Timing

2.6.5.2 RGMII AC Timing Specifications

Table 34 presents the RGMII AC timing specifications for applications requiring an on-board delayed clock.

Table 34. RGMII at 1 Gbps[2] with On-Board Delay[3] AC Timing Specifications

Parameter/Condition	Symbol	Min	Typ	Max	Unit
Data to clock output skew (at transmitter)[4]	t_{SKEWT}	–-0.5	—	0.5	ns
Data to clock input skew (at receiver) [4]	t_{SKEWR}	1	—	2.6	ns

Notes:
1. At recommended operating conditions with V_{DDIO} of 2.5 V ± 5%.
2. RGMII at 100 Mbps support is guaranteed by design.
3. Program GCR4 as 0x00000000.
4. This implies that PC board design requires clocks to be routed such that an additional trace delay of greater than 1.5 ns and less than 2.0 ns is added to the associated clock signal.

Table 35 presents the RGMII AC timing specification for applications required non-delayed clock on board.

Table 35. RGMII at 1 Gbps[2] with No On-Board Delay[3] AC Timing Specifications

Parameter/Condition	Symbol	Min	Typ	Max	Unit
Data to clock output skew (at transmitter)[4]	t_{SKEWT}	–2.6	—	–1.0	ns
Data to clock input skew (at receiver)[4]	t_{SKEWR}	–0.5	—	0.5	ns

Notes:
1. At recommended operating conditions with V_{DDIO} of 2.5 V ± 5%.
2. RGMII at 100 Mbps support is guaranteed by design.
3. GCR4 should be programmed as 0x000CC330.
4. This implies that PC board design requires clocks to be routed with no additional trace delay

Figure 25 shows the RGMII AC timing and multiplexing diagrams.

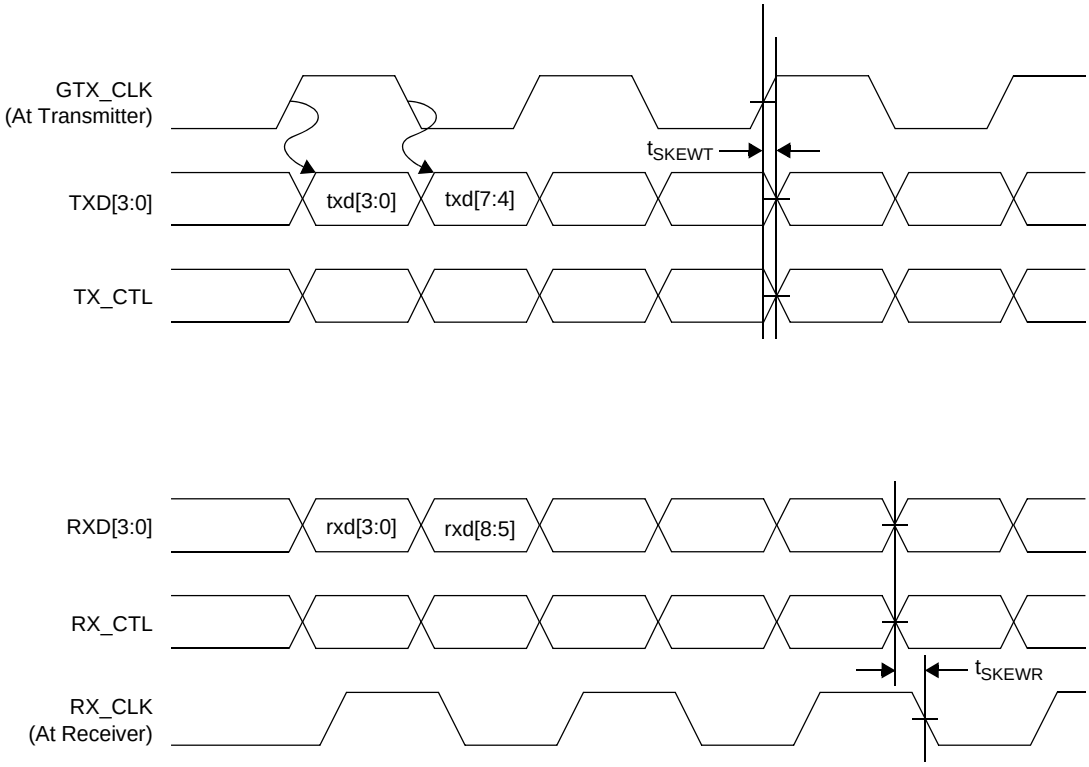
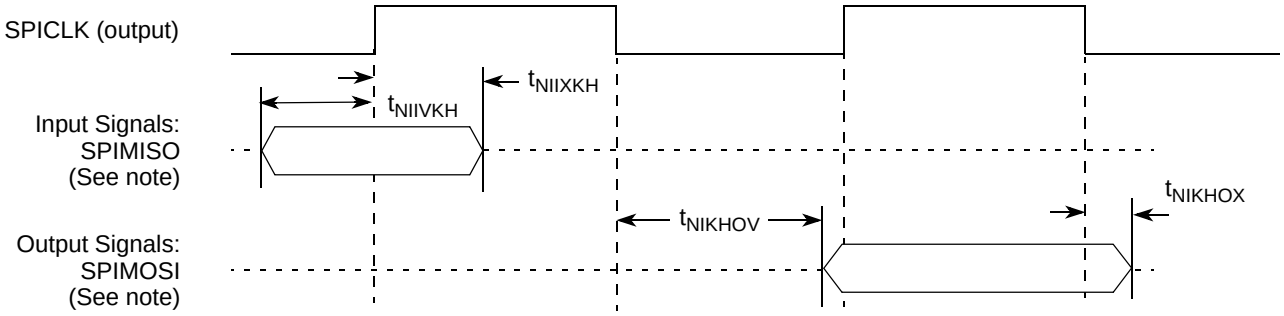


Figure 25. RGMII AC Timing and Multiplexing



Note: measured with SPMODE[CI] = 0, SPMODE[CP] = 0

Figure 28. SPI AC Timing in Master Mode (Internal Clock)

2.6.7 Asynchronous Signal Timing

Table 35 lists the asynchronous signal timing specifications.

Table 37. Signal Timing

Characteristics	Symbol	Type	Min
Input	t_{IN}	Asynchronous	One CLKIN cycle
Output	t_{OUT}	Asynchronous	Application dependent
Note: Input value relevant for EE0, $\overline{\text{IRQ}}$[15–0], and $\overline{\text{NMI}}$ only.			

The following interfaces use the specified asynchronous signals:

- *GPIO*. Signals GPIO[31–0], when used as GPIO signals, that is, when the alternate multiplexed special functions are not selected.

Note: When used as a general purpose input (GPI), the input signal should be driven until it is acknowledged by the MSC8252 device, that is, when the expected input value is read from the GPIO data register.

- *EE port*. Signals EE0, EE1.
- *Boot function*. Signal STOP_BS.
- *I^2C interface*. Signals I2C_SCL and I2C_SDA.
- *Interrupt inputs*. Signals $\overline{\text{IRQ}}$[15–0] and $\overline{\text{NMI}}$.
- *Interrupt outputs*. Signals $\overline{\text{INT\_OUT}}$ and $\overline{\text{NMI\_OUT}}$ (minimum pulse width is 32 ns).

2.6.8 JTAG Signals

Table 38 lists the JTAG timing specifications shown in Figure 29 through Figure 32.

Table 38. JTAG Timing

Characteristics	Symbol	All frequencies		Unit
		Min	Max	
TCK cycle time	t_{TCKX}	36.0	—	ns
TCK clock high phase measured at $V_M = V_{DDIO}/2$	t_{TCKH}	15.0	—	ns
Boundary scan input data setup time	t_{BSVKH}	0.0	—	ns
Boundary scan input data hold time	t_{BSXKH}	15.0	—	ns
TCK fall to output data valid	t_{TCKHOV}	—	20.0	ns
TCK fall to output high impedance	t_{TCKHOZ}	—	24.0	ns
TMS, TDI data setup time	t_{TDIVKH}	0.0	—	ns
TMS, TDI data hold time	t_{TDIXKH}	5.0	—	ns
TCK fall to TDO data valid	t_{TDOHOV}	—	10.0	ns
TCK fall to TDO high impedance	t_{TDOHOZ}	—	12.0	ns
$\overline{\text{TRST}}$ assert time	t_{TRST}	100.0	—	ns
Note: All timings apply to OnCE module data transfers as well as any other transfers via the JTAG port.				

Figure 29 shows the test clock input timing diagram

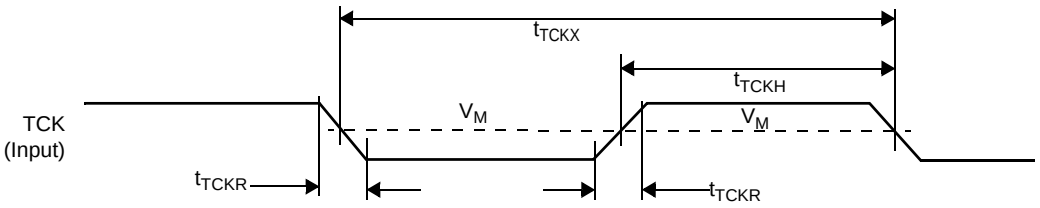


Figure 29. Test Clock Input Timing

2. After the above rails rise to 90% of their nominal voltage, the following I/O power rails may rise in any sequence (see Figure 34): QVDD, NVDD, GVDD1, and GVDD2.

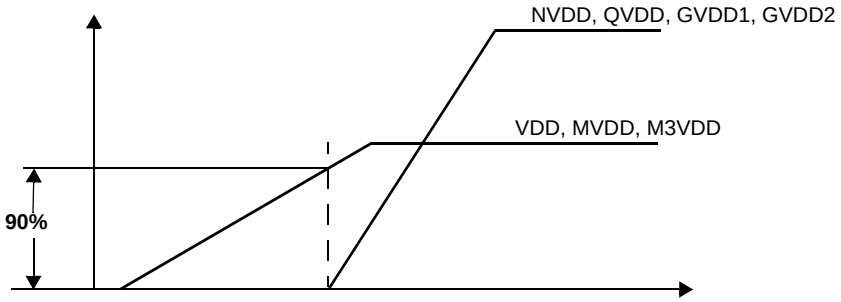


Figure 34. Supply Ramp-Up Sequence

Notes:
1. If the M3 memory is not used, M3VDD can be tied to GND.
2. If the HSSI port1 is not used, SXCVDD1and SXPVDD1 must be connected to the designated power supplies.
3. If the HSSI port2 is not used, SXCVDD2 and SXPVDD2 must be connected to the designated power supplies.
4. If the DDR port 1 interface is not used, it is recommended that GVDD1 be left unconnected.
5. If the DDR port 2 interface is not used, it is recommended that GVDD2 be left unconnected.

3.1.4 Reset Guidelines

When a debugger is not used, implement the connection scheme shown in Figure 35.

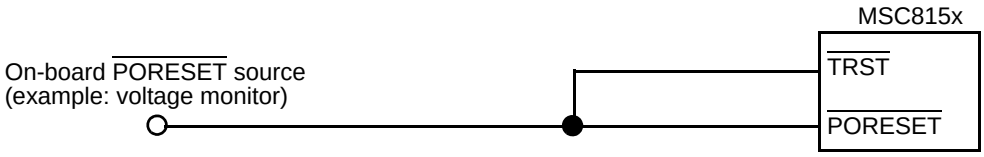


Figure 35. Reset Connection in Functional Application

When a debugger is used, implement the connection scheme shown in Figure 36.

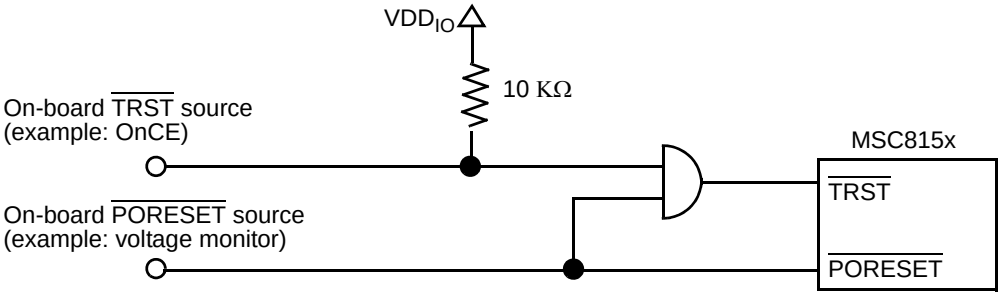


Figure 36. Reset Connection in Debugger Application

3.3 Clock and Timing Signal Board Layout Considerations

When laying out the system board, use the following guidelines:

- Keep clock and timing signal paths as short as possible and route with 50 Ω impedance.
- Use a serial termination resistor placed close to the clock buffer to minimize signal reflection. Use the following equation to compute the resistor value:

  Rterm = Rim – Rbuf

  where Rim = trace characteristic impedance

  Rbuf = clock buffer internal impedance.

3.4 SGMII AC-Coupled Serial Link Connection Example

Figure 39 shows an example of a 4-wire AC-coupled serial link connection. For additional layout suggestions, see *AN3556 MSC815x High Speed Serial Interface Hardware Design Considerations*, available on the Freescale website or from your local sales office or representative.

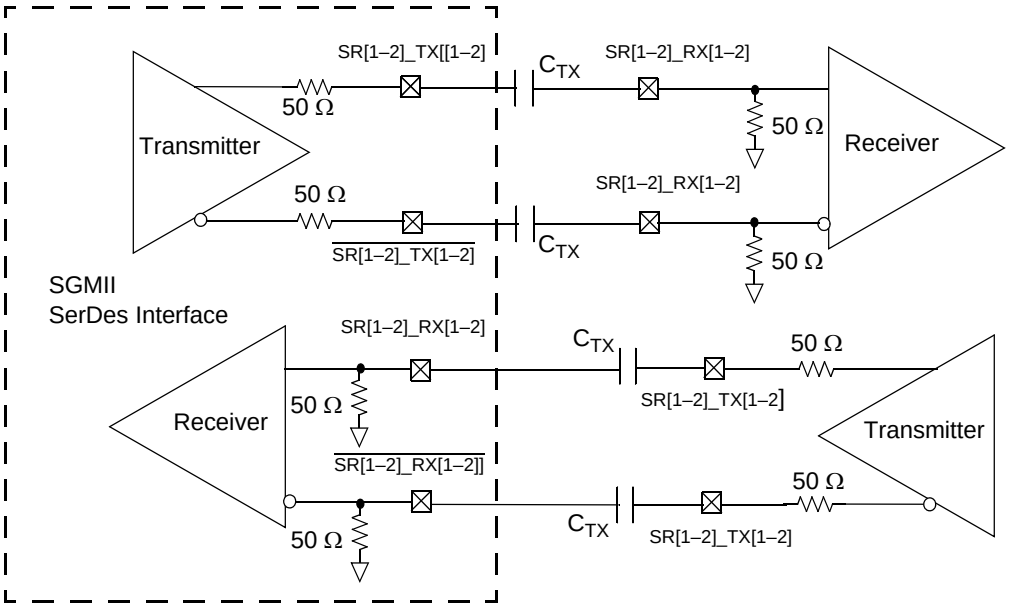


Figure 39. 4-Wire AC-Coupled SGMII Serial Link Connection Example

3.5 Connectivity Guidelines

Note: Although the package actually uses a ball grid array, the more conventional term pin is used to denote signal connections in this discussion.

First, select the pin multiplexing mode to allocate the required I/O signals. Then use the guidelines presented in the following subsections for board design and connections. The following conventions are used in describing the connectivity requirements:

1. GND indicates using a 10 kΩ pull-down resistor (recommended) or a direct connection to the ground plane. Direct connections to the ground plane may yield DC current up to 50 mA through the I/O supply that adds to overall power consumption.
2. V_{DD} indicates using a 10 kΩ pull-up resistor (recommended) or a direct connection to the appropriate power supply. Direct connections to the supply may yield DC current up to 50 mA through the I/O supply that adds to overall power consumption.
3. Mandatory use of a pull-up or pull-down resistor is clearly indicated as “pull-up/pull-down.” For buses, each pin on the bus should have its own resistor.
4. NC indicates “not connected” and means do not connect anything to the pin.
5. The phrase “in use” indicates a typical pin connection for the required function.

Note: Please see recommendations #1 and #2 as mandatory pull-down or pull-up connection for unused pins in case of subset interface connection.

3.5.1.2 DDR Interface Is Used With 32-Bit DDR Memory Only

Table 41 lists unused pin connection when using 32-bit DDR memory. The 32 most significant data lines are not used.

Table 41. Connectivity of DDR Related Pins When Using 32-bit DDR Memory Only

Signal Name	Pin Connection
MDQ[31–0]	in use
MDQ[63–32]	NC
MDQS[3–0]	in use
MDQS[7–4]	NC
$\overline{\text{MDQS}}$[3–0]	in use
$\overline{\text{MDQS}}$[7–4]	NC
MA[15–0]	in use
MCK[2–0]	in use
$\overline{\text{MCK}}$[2–0]	in use
$\overline{\text{MCS}}$[1–0]	in use
MDM[3–0]	in use
MDM[7–4]	NC
MBA[2–0]	in use
$\overline{\text{MCAS}}$	in use
MCKE[1–0]	in use
MODT[1–0]	in use
MMDIC[1–0]	in use
$\overline{\text{MRAS}}$	in use
$\overline{\text{MWE}}$	in use
MVREF	in use
GVDD1/GVDD2	in use

Notes:
1. For the signals listed in this table, the initial M stands for M1 or M2 depending on which DDR controller is not used.
2. For MSC8252 Revision 1 silicon, these pins were connected to GND (or VDD). For newer revisions of the MSC8252, connecting these pins to GND increases device power consumption.

3.5.1.3 ECC Unused Pin Connections

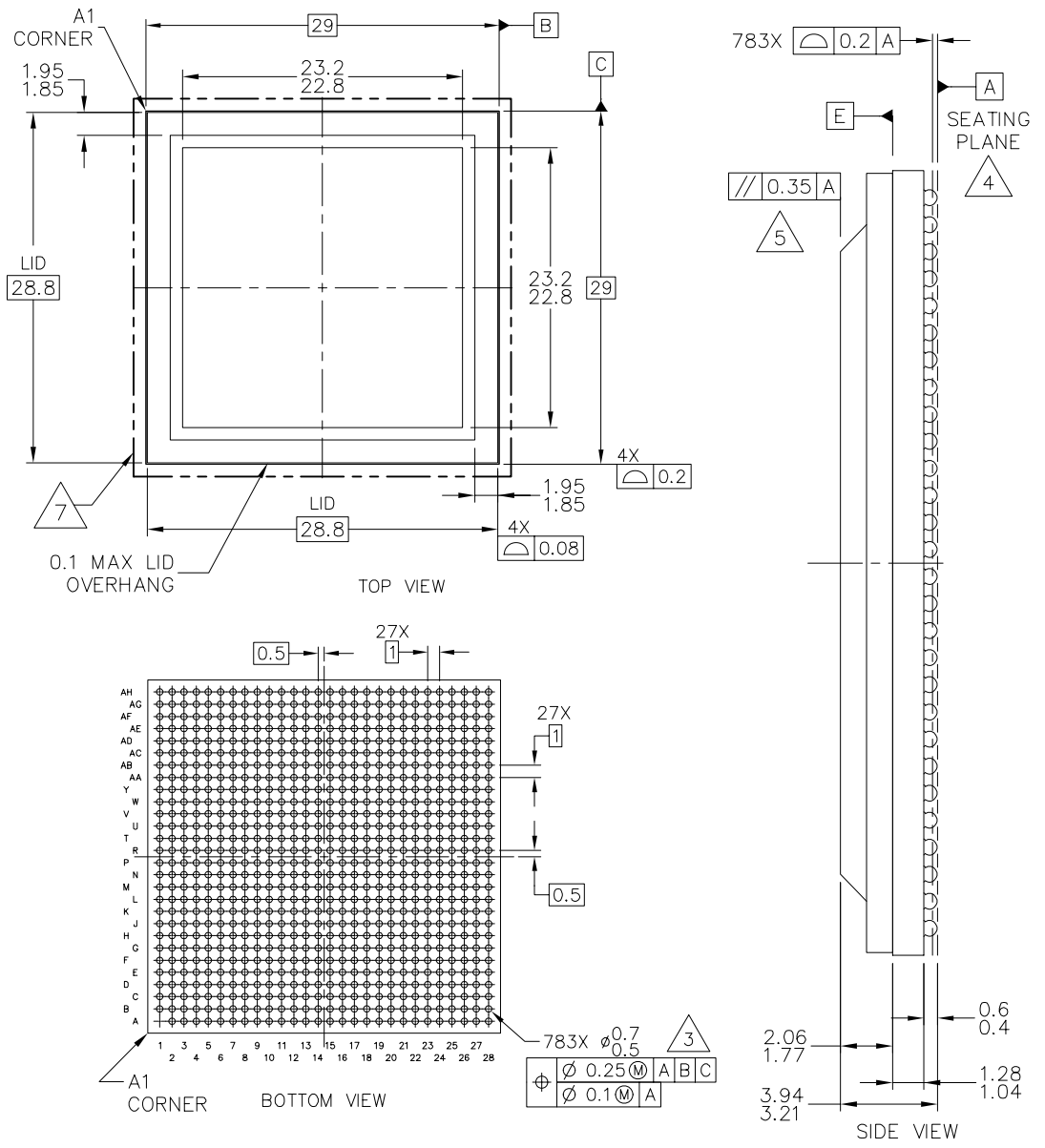
When the error code correction mechanism is not used in any 32- or 64-bit DDR configuration, refer to Table 42 to determine the correct pin connections.

Table 42. Connectivity of Unused ECC Mechanism Pins

Signal Name	Pin connection
MECC[7–0]	NC
MDM8	NC
MDQS8	NC
$\overline{\text{MDQS8}}$	NC

Notes:
1. For the signals listed in this table, the initial M stands for M1 or M2 depending on which DDR controller is not used.
2. For MSC8252 Revision 1 silicon, these pins were connected to GND (or VDD). For newer revisions of the MSC8252, connecting these pins to GND increases device power consumption.

5 Package Information



NOTES:

1. ALL DIMENSIONS IN MILLIMETERS.
2. DIMENSIONING AND TOLERANCING PER ASME Y14.5M-1994.
3. MAXIMUM SOLDER BALL DIAMETER MEASURE PARALLEL TO DATUM A.
4. DATUM A, THE SEATING PLANE, IS DETERMINED BY THE SPHERICAL CROWNS OF THE SOLDER BALLS.
5. PARALLELISM MEASUREMENT SHALL EXCLUDE ANY EFFECT OF MARK ON TOP SURFACE OF PACKAGE.
6. ALL DIMENSIONS ARE SYMMETRIC ACROSS THE PACKAGE CENTER LINES, UNLESS DIMENSIONED OTHERWISE.
7. 29.2MM MAXIMUM PACKAGE ASSEMBLY (LID + LAMINATE) X AND Y.

Figure 40. MSC8252 Mechanical Information, 783-ball FC-PBGA Package

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