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### What is "[Embedded - Microcontrollers](#)"?

"[Embedded - Microcontrollers](#)" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

### Applications of "[Embedded - Microcontrollers](#)"

#### Details

Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	32MHz
Connectivity	I <sup>2</sup> C, LINbus, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	26
Program Memory Size	28KB (16K x 14)
Program Memory Type	FLASH
EEPROM Size	256 x 8
RAM Size	2K x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.6V
Data Converters	A/D 24x12b; D/A 1x5b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	28-SOIC (0.295", 7.50mm Width)
Supplier Device Package	28-SOIC
Purchase URL	<a href="https://www.e-xfl.com/product-detail/microchip-technology/pic16lf18456-i-so">https://www.e-xfl.com/product-detail/microchip-technology/pic16lf18456-i-so</a>

#### 4.7.5 CONFIG5

Name: CONFIG5  
Address: 0x800B

Configuration Word 5

Code Protection

Bit	15	14	13	12	11	10	9	8
	[Bit Fields]							
Access			U	U	U	U	U	U
Reset			1	1	1	1	1	1
Bit	7	6	5	4	3	2	1	0
	[Bit Fields]							
Access	U	U	U	U	U	U	U	R/P
Reset	1	1	1	1	1	1	1	1

**Bit 0 –  $\overline{CP}$**  Program Flash Memory Code Protection bit

Value	Description
1	Program Flash Memory code protection disabled
0	Program Flash Memory code protection enabled

#### 4.8 Register Summary - Device and Revision

Offset	Name	Bit Pos.							
0x8005	REVISION ID	7:0	MJRREV[1:0]		MNRREV[5:0]				
		13:8		1	0	MJRREV[5:2]			
0x8006	DEVICE ID	7:0	DEV[7:0]						
		13:8	1	1	DEV[11:8]				

#### 4.9 Register Definitions: Device and Revision

**4.9.2 REVISION ID**

**Name:** REVISION ID  
**Address:** 0x8005

Revision ID Register

	Bit	15	14	13	12	11	10	9	8
				1	0	MJRREV[5:2]			
Access				R	R	R	R	R	R
Reset				1	0				
	Bit	7	6	5	4	3	2	1	0
		MJRREV[1:0]				MNRREV[5:0]			
Access		R	R	R	R	R	R	R	R
Reset									

**Bit 13 – 1** Read as ‘1’  
 These bits are fixed with value ‘1’ for all devices in this family.

**Bit 12 – 0** Read as ‘0’  
 These bits are fixed with value ‘0’ for all devices in this family.

**Bits 11:6 – MJRREV[5:0]** Major Revision ID bits  
 These bits are used to identify a major revision.

**Bits 5:0 – MNRREV[5:0]** Minor Revision ID bits  
 These bits are used to identify a minor revision.

### 7.1.1.1 RETLW Instruction

The RETLW instruction can be used to provide access to tables of constants. The recommended way to create such a table is shown in the following example.

```
constants
    BRW                ;Add Index in W to
                      ;program counter to
                      ;select data
    RETLW DATA0       ;Index0 data
    RETLW DATA1       ;Index1 data
    RETLW DATA2
    RETLW DATA3
my_function
;... LOTS OF CODE...
    MOVLW             DATA_INDEX
    call constants
;... THE CONSTANT IS IN W
```

The BRW instruction makes this type of table very simple to implement.

### 7.1.1.2 Indirect Read with FSR

The program memory can be accessed as data by setting bit 7 of an FSRxH register and reading the matching INDFx register. The MOVIW instruction will place the lower eight bits of the addressed word in the W register. Writes to the program memory cannot be performed via the INDFx registers. Instructions that read the program memory via the FSR require one extra instruction cycle to complete. The following example demonstrates reading the program memory via an FSR.

The HIGH directive will set bit 7 if a label points to a location in the program memory. This applies to the assembly code shown below.

```
constants
    RETLW DATA0       ;Index0 data
    RETLW DATA1       ;Index1 data
    RETLW DATA2
    RETLW DATA3
my_function
;... LOTS OF CODE...
    MOVLW             LOW constants
    MOVWF            FSR1L
    MOVLW             HIGH constants
    MOVWF            FSR1H
    MOVIW            2[FSR1
;DATA2 IS IN W
```

## 7.2 Memory Access Partition (MAP)

User Flash is partitioned into:

- Application Block
- Boot Block, and
- Storage Area Flash (SAF) Block

The user can allocate the memory usage by setting the  $\overline{\text{BBEN}}$  bit, selecting the size of the partition defined by BBSIZE bits and enabling the Storage Area Flash by the  $\overline{\text{SAFEN}}$  bit of the Configuration Word. Refer to the following links for the different user Flash memory partitions.

### Related Links

### **7.3.3 Special Function Register**

The Special Function Registers (SFR) are registers used by the application to control the desired operation of peripheral functions in the device. The SFRs occupy the first 20 bytes of the data banks 0-59 and the first 100 bytes of the data banks 60-63, after the core registers.

The SFRs associated with the operation of the peripherals are described in the appropriate peripheral chapter of this data sheet.

### **7.3.4 General Purpose RAM**

There are up to 80 bytes of GPR in each data memory bank.

#### **7.3.4.1 Linear Access to GPR**

The general purpose RAM can be accessed in a non-banked method via the FSRs. This can simplify access to large memory structures.

#### **Related Links**

[7.6.2 Linear Data Memory](#)

### **7.3.5 Common RAM**

There are 16 bytes of common RAM accessible from all banks.

## **7.4 PCL and PCLATH**

The Program Counter (PC) is 15 bits wide. The low byte comes from the PCL register, which is a readable and writable register. The high byte (PC<14:8>) is not directly readable or writable and comes from PCLATH. On any Reset, the PC is cleared. The following figure shows the five situations for the loading of the PC.

**7.8.3 PCL**

**Name:** PCL

**Address:** 0x02 + n\*0x80 [n=0..63]

Low byte of the Program Counter

Bit	7	6	5	4	3	2	1	0
	PCL[7:0]							
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset	0	0	0	0	0	0	0	0

**Bits 7:0 – PCL[7:0]**

Provides direct read and write access to the Program Counter

**9.4.2 Fail-Safe Operation**

When the external clock fails, the FSCM overwrites the **COSC** bits to select HFINTOSC (3'b110). The frequency of HFINTOSC would be determined by the previous state of the HFFRQ bits and the **NDIV/CDIV** bits. The bit flag **OSCFIF** of the PIR1 register is set. Setting this flag will generate an interrupt if the **OSCFIE** bit of the PIE1 register is also set. The device firmware can then take steps to mitigate the problems that may arise from a failed clock. The system clock will continue to be sourced from the internal clock source until the device firmware successfully restarts the external oscillator and switches back to external operation, by writing to the **NOSC** and **NDIV** bits.

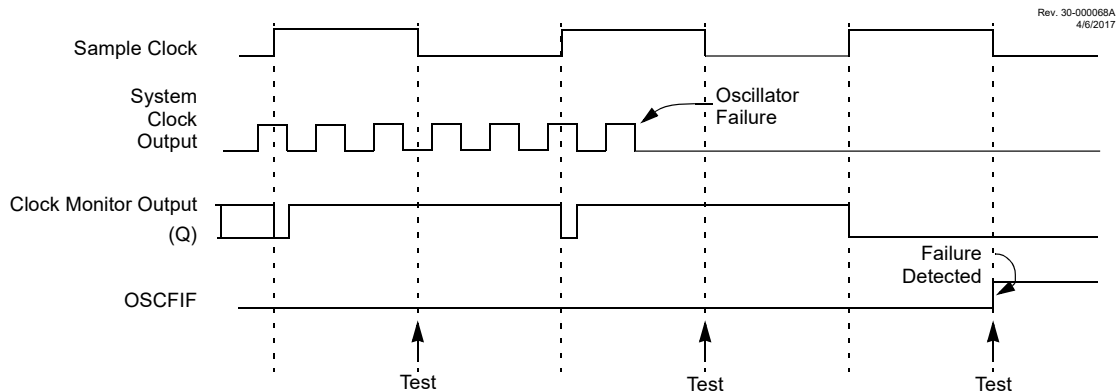
**9.4.3 Fail-Safe Condition Clearing**

The Fail-Safe condition is cleared after a Reset, executing a **SLEEP** instruction or changing the **NOSC** and **NDIV** bits. When switching to the external oscillator or PLL, the **OST** is restarted. While the **OST** is running, the device continues to operate from the **INTOSC** selected in **OSCCON1**. When the **OST** times out, the Fail-Safe condition is cleared after successfully switching to the external clock source. The **OSCFIF** bit should be cleared prior to switching to the external clock source. If the Fail-Safe condition still exists, the **OSCFIF** flag will again become set by hardware.

**9.4.4 Reset or Wake-up from Sleep**

The FSCM is designed to detect an oscillator failure after the Oscillator Start-up Timer (**OST**) has expired. The **OST** is used after waking up from Sleep and after any type of Reset. The **OST** is not used with the **EC** Clock modes so that the FSCM will be active as soon as the Reset or wake-up has completed.

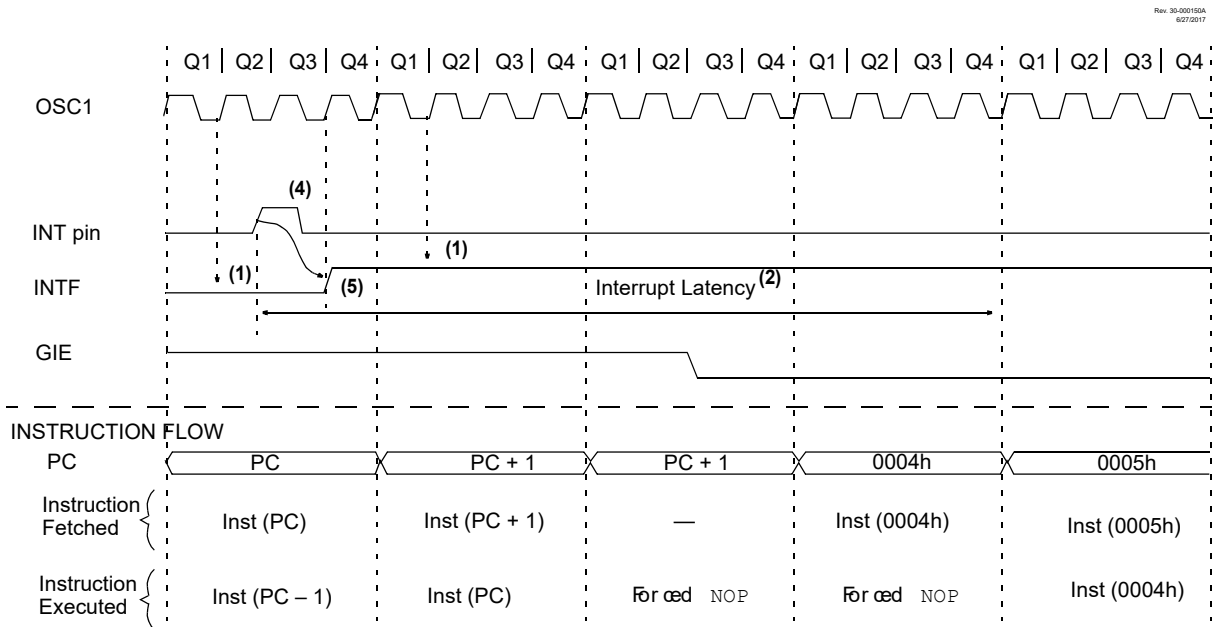
**Figure 9-10. FSCM Timing Diagram**



**Note:** The system clock is normally at a much higher frequency than the sample clock. The relative frequencies in this example have been chosen for clarity.

1. An interrupt may occur at any time during the interrupt window.
2. Since an interrupt may occur any time during the interrupt window, the actual latency can vary.

Figure 10-3. INT Pin Interrupt Timing



**Note:**

1. INTF flag is sampled here (every Q1).
2. Asynchronous interrupt latency = 3-5  $T_{CY}$ . Synchronous latency = 3-4  $T_{CY}$ , where  $T_{CY}$  = instruction cycle time. Latency is the same whether Inst (PC) is a single cycle or a 2-cycle instruction.
3. For minimum width of INT pulse, refer to AC specifications in the “Electrical Specifications” section.
4. INTF may be set any time during the Q4-Q1 cycles.

### 10.3 Interrupts During Sleep

Interrupts can be used to wake from Sleep. To wake from Sleep, the peripheral must be able to operate without the system clock. The interrupt source must have the appropriate Interrupt Enable bit(s) set prior to entering Sleep.

On waking from Sleep, if the GIE bit is also set, the processor will branch to the interrupt vector. Otherwise, the processor will continue executing instructions after the SLEEP instruction. The instruction directly after the SLEEP instruction will always be executed before branching to the ISR.

**Related Links**

[11. Power-Saving Operation Modes](#)

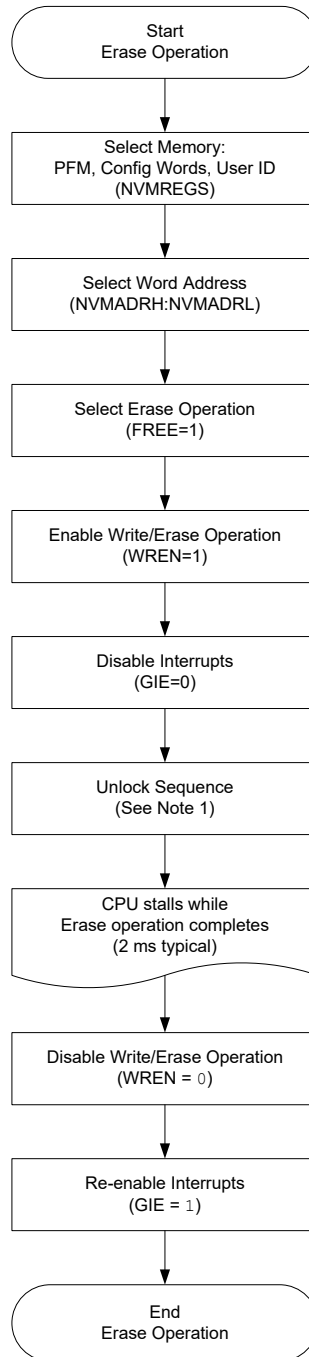
### 10.4 INT Pin

The INT pin can be used to generate an asynchronous edge-triggered interrupt. Refer to Figure 10-3. This interrupt is enabled by setting the INTE bit of the PIE0 register. The INTEDG bit of the INTCON register determines on which edge the interrupt will occur. When the INTEDG bit is set, the rising edge will cause the interrupt. When the INTEDG bit is clear, the falling edge will cause the interrupt. The INTF



Figure 13-3. NVM Erase Flowchart

Rev. 10-000048B  
03/24/2015



**Note:** See previous figure.

**Example 13-3. Erasing One Row of Program Flash Memory**

```

; This sample row erase routine assumes the following:
; 1.A valid address within the erase row is loaded in variables ADDRH:ADDRL
; 2.ADDRH and ADDRL are located in common RAM (locations 0x70 - 0x7F)
  
```

14.7.9 LATB

**Name:** LATB  
**Address:** 0x019

Output Latch Register

Bit	7	6	5	4	3	2	1	0
	LATB7	LATB6	LATB5	LATB4	LATB3	LATB2	LATB1	LATB0
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset	x	x	x	x	x	x	x	x

**Bits 0, 1, 2, 3, 4, 5, 6, 7 – LATBn** Output Latch B Value bits

Reset States: POR/BOR = xxxxxxxx

All Other Resets = uuuuuuuu

**Note:** Writes to LATB are equivalent with writes to the corresponding PORTB register. Reads from LATB register return register values, not I/O pin values.

## 16.4 Register Summary - PMD

Address	Name	Bit Pos.								
0x0796	<a href="#">PMD0</a>	7:0	SYSCMD	FVRMD				NVMMD	CLKRMD	IOCMD
0x0797	<a href="#">PMD1</a>	7:0		TMR6MD	TMR5MD	TMR4MD	TMR3MD	TMR2MD	TMR1MD	TMR0MD
0x0798	<a href="#">PMD2</a>	7:0	NCO1MD							
0x0799	<a href="#">PMD3</a>	7:0		DAC1MD	ADCMD			C2MD	C1MD	ZCDMD
0x079A	<a href="#">PMD4</a>	7:0		PWM7MD	PWM6MD	CCP5MD	CCP4MD	CCP3MD	CCP2MD	CCP1MD
0x079B	<a href="#">PMD5</a>	7:0	CWG3MD	CWG2MD	CWG1MD					
0x079C	<a href="#">PMD6</a>	7:0			UART2MD	UART1MD			MSSP2MD	MSSP1MD
0x079D	<a href="#">PMD7</a>	7:0		SMT2MD	SMT1MD	CLC4MD	CLC3MD	CLC2MD	CLC1MD	DSM1MD

## 16.5 Register Definitions: Peripheral Module Disable

$$V_{APPLIED} \left( 1 - e^{-\frac{T_C}{R_C}} \right) = V_{APPLIED} \left( 1 - \frac{1}{(2^n + 1) - 1} \right) \quad ; \text{combining [1] and [2]}$$

**Note:** Where n = number of bits of the ADC.

Solving for  $T_C$ :

$$T_C = -C_{HOLD}(R_{IC} + R_{SS} + R_S) \ln(1/8191)$$

$$T_C = -28pF(1k\Omega + 7k\Omega + 10k\Omega) \ln(0.0001221)$$

$$T_C = 4.54us$$

Therefore:

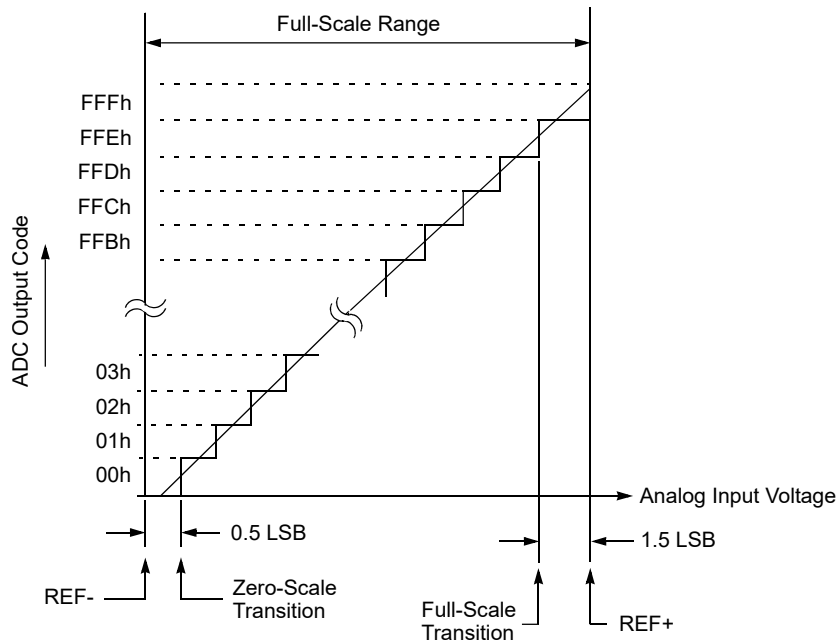
$$T_{ACQ} = 2us + 4.54us + [(50^\circ C - 25^\circ C)(0.05us/^\circ C)]$$

$$T_{ACQ} = 7.79us$$

**Note:**

1. The reference voltage ( $V_{REF}$ ) has no effect on the equation, since it cancels itself out.
2. The charge holding capacitor ( $C_{HOLD}$ ) is not discharged after each conversion.
3. The maximum recommended impedance for analog sources is 10 k $\Omega$ . This is required to meet the pin leakage specification.

**Figure 20-5. ADC Transfer Function**



**Related Links**

[42.4.4 I/O and CLKOUT Timing Specifications](#)

synchronization is only possible with the Timer1 clock source. Synchronization with the other odd numbered timers is only possible when they use the same clock source as Timer1.

**Related Links**

[26.7 Timer1 Gate](#)

**23.4.1 Comparator Output Synchronization**

The output from a comparator can be synchronized with Timer1 by setting the [SYNC](#) bit.

Once enabled, the comparator output is latched on the falling edge of the Timer1 source clock. If a prescaler is used with Timer1, the comparator output is latched after the prescaling function. To prevent a race condition, the comparator output is latched on the falling edge of the Timer1 clock source and Timer1 increments on the rising edge of its clock source. See the [Figure 23-2 Comparator Block Diagram](#) and the [Timer1 Block Diagram](#) for more information.

**Related Links**

[26. Timer1 Module with Gate Control](#)

**23.5 Comparator Interrupt**

An interrupt can be generated upon a change in the output value of the comparator for each comparator; a rising edge detector and a falling edge detector are present.

When either edge detector is triggered and its associated enable bit is set ([CxINTP](#) and/or [CxINTN](#) bits), the Corresponding Interrupt Flag bit (CxIF bit of the PIR2 register) will be set.

To enable the interrupt, the following bits must be set:

- [EN](#) and [POL](#) bits
- CxIE bit of the PIE2 register
- [INTP](#) bit (for a rising edge detection)
- [INTN](#) bit (for a falling edge detection)
- PEIE and GIE bits of the INTCON register

The associated interrupt flag bit, CxIF bit of the PIR2 register, must be cleared in software. If another edge is detected while this flag is being cleared, the flag will still be set at the end of the sequence.



**Important:** Although a comparator is disabled, an interrupt can be generated by changing the output polarity with the [CxPOL](#) bit, or by switching the comparator on or off with the [CxEN](#) bit.

**23.6 Comparator Positive Input Selection**

Configuring the [PCH](#) bits direct an internal voltage reference or an analog pin to the non-inverting input of the comparator:

PCH	Positive Input Source
111	CxV <sub>P</sub> connects to V <sub>SS</sub>
110	CxV <sub>P</sub> connects to FVR Buffer 2

## 23.8 Comparator Response Time

The comparator output is indeterminate for a period of time after the change of an input source or the selection of a new reference voltage. This period is referred to as the response time. The response time of the comparator differs from the settling time of the voltage reference. Therefore, both of these times must be considered when determining the total response time to a comparator input change. See the Comparator and Voltage Reference Specifications in Comparator Specifications and Fixed Voltage Reference (FVR) Specifications for more details.

### Related Links

[42.4.9 Comparator Specifications](#)

[42.4.11 Fixed Voltage Reference \(FVR\) Specifications](#)

## 23.9 Analog Input Connection Considerations

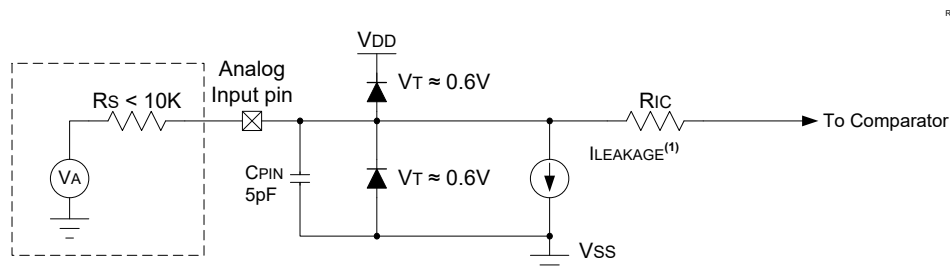
A simplified circuit for an analog input is shown in [Figure 23-3](#). Since the analog input pins share their connection with a digital input, they have reverse biased ESD protection diodes to  $V_{DD}$  and  $V_{SS}$ . The analog input, therefore, must be between  $V_{SS}$  and  $V_{DD}$ . If the input voltage deviates from this range by more than 0.6V in either direction, one of the diodes is forward biased and a latch-up may occur.

A maximum source impedance of 10 k $\Omega$  is recommended for the analog sources. Also, any external component connected to an analog input pin, such as a capacitor or a Zener diode, should have very little leakage current to minimize inaccuracies introduced.

### Note:

1. When reading a PORT register, all pins configured as analog inputs will read as a '0'. Pins configured as digital inputs will convert as an analog input, according to the input specification.
2. Analog levels on any pin defined as a digital input, may cause the input buffer to consume more current than is specified.

**Figure 23-3. Analog Input Model**



<b>Legend:</b>	CPIN	= Input Capacitance
	ILEAKAGE	= Leakage Current at the pin due to various junctions
	RIC	= Interconnect Resistance
	RS	= Source Impedance
	VA	= Analog Voltage
	VT	= Threshold Voltage

**Note:** See *Electrical Specifications* chapter.

### Related Links

[42. Electrical Specifications](#)

27.9.5 TxCLKCON

**Name:** TxCLKCON  
**Address:** 0x290,0x296,0x29C

Timer Clock Source Selection Register



**Bits 3:0 – CS[3:0]** Timer Clock Source Selection bits

Value	Description
n	See <a href="#">Clock Source Selection</a> table

# PIC16(L)F18455/56

## (MSSP) Master Synchronous Serial Port Module

---

- Master Transmit mode  
(master is transmitting data to a slave)
- Master Receive mode  
(master is receiving data from a slave)
- Slave Transmit mode  
(slave is transmitting data to a master)
- Slave Receive mode  
(slave is receiving data from the master)

To begin communication, a master device starts out in Master Transmit mode. The master device sends out a Start bit followed by the address byte of the slave it intends to communicate with. This is followed by a single Read/Write bit, which determines whether the master intends to transmit to or receive data from the slave device.

If the requested slave exists on the bus, it will respond with an Acknowledge bit, otherwise known as an  $\overline{\text{ACK}}$ . The master then continues in either Transmit mode or Receive mode and the slave continues in the complement, either in Receive mode or Transmit mode, respectively.

A Start bit is indicated by a high-to-low transition of the SDA line while the SCL line is held high. Address and data bytes are sent out, Most Significant bit (MSb) first. The Read/Write bit is sent out as a logical one when the master intends to read data from the slave, and is sent out as a logical zero when it intends to write data to the slave.

The Acknowledge bit ( $\overline{\text{ACK}}$ ) is an active-low signal, which holds the SDA line low to indicate to the transmitter that the slave device has received the transmitted data and is ready to receive more.

The transition of a data bit is always performed while the SCL line is held low. Transitions that occur while the SCL line is held high are used to indicate Start and Stop bits.

If the master intends to write to the slave, then it repeatedly sends out a byte of data, with the slave responding after each byte with an  $\overline{\text{ACK}}$  bit. In this example, the master device is in Master Transmit mode and the slave is in Slave Receive mode.

If the master intends to read from the slave, then it repeatedly receives a byte of data from the slave, and responds after each byte with an  $\overline{\text{ACK}}$  bit. In this example, the master device is in Master Receive mode and the slave is Slave Transmit mode.

On the last byte of data communicated, the master device may end the transmission by sending a Stop bit. If the master device is in Receive mode, it sends the Stop bit in place of the last  $\overline{\text{ACK}}$  bit. A Stop bit is indicated by a low-to-high transition of the SDA line while the SCL line is held high.

In some cases, the master may want to maintain control of the bus and re-initiate another transmission. If so, the master device may send another Start bit in place of the Stop bit or last  $\overline{\text{ACK}}$  bit when it is in receive mode.

The I<sup>2</sup>C bus specifies three message protocols;

- Single message where a master writes data to a slave.
- Single message where a master reads data from a slave.
- Combined message where a master initiates a minimum of two writes, or two reads, or a combination of writes and reads, to one or more slaves.

When one device is transmitting a logical one, or letting the line float, and a second device is transmitting a logical zero, or holding the line low, the first device can detect that the line is not a logical one. This detection, when used on the SCL line, is called clock stretching. Clock stretching gives slave devices a



10417	10417	0.00	767	10417	0.00	479	10425	0.08	441	10433	0.16	264
19.2k	19.18k	-0.08	416	19.23k	0.16	259	19.20k	0.00	239	19.20k	0.00	143
57.6k	57.55k	-0.08	138	57.47k	-0.22	86	57.60k	0.00	79	57.60k	0.00	47
115.2k	115.9k	0.64	68	116.3k	0.94	42	115.2k	0.00	39	115.2k	0.00	23

BAUD RATE	SYNC = 0, BRGH = 1, BRG16 = 1 or SYNC = 1, BRG16 = 1											
	Fosc = 8.000 MHz			Fosc = 4.000 MHz			Fosc = 3.6864 MHz			Fosc = 1.000 MHz		
	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)
300	300.0	0.00	6666	300.0	0.01	3332	300.0	0.00	3071	300.1	0.04	832
1200	1200	-0.02	1666	1200	0.04	832	1200	0.00	767	1202	0.16	207
2400	2401	0.04	832	2398	0.08	416	2400	0.00	383	2404	0.16	103
9600	9615	0.16	207	9615	0.16	103	9600	0.00	95	9615	0.16	25
10417	10417	0	191	10417	0.00	95	10473	0.53	87	10417	0.00	23
19.2k	19.23k	0.16	103	19.23k	0.16	51	19.20k	0.00	47	19.23k	0.16	12
57.6k	57.14k	-0.79	34	58.82k	2.12	16	57.60k	0.00	15	—	—	—
115.2k	117.6k	2.12	16	111.1k	-3.55	8	115.2k	0.00	7	—	—	—

### 36.2.1 Auto-Baud Detect

The EUSART module supports automatic detection and calibration of the baud rate.

In the Auto-Baud Detect (ABD) mode, the clock to the BRG is reversed. Rather than the BRG clocking the incoming RX signal, the RX signal is timing the BRG. The Baud Rate Generator is used to time the period of a received 55h (ASCII "U") which is the Sync character for the LIN bus. The unique feature of this character is that it has five rising edges including the Stop bit edge.

Setting the ABDEN bit of the BAUDxCON register starts the auto-baud calibration sequence. While the ABD sequence takes place, the EUSART state machine is held in Idle. On the first rising edge of the receive line, after the Start bit, the SPxBRG begins counting up using the BRG counter clock as shown in [Figure 36-6](#). The fifth rising edge will occur on the RXx pin at the end of the eighth bit period. At that time, an accumulated value totaling the proper BRG period is left in the SPxBRGH, SPxBRGL register pair, the ABDEN bit is automatically cleared and the RCxIF interrupt flag is set. The value in the RCxREG needs to be read to clear the RCxIF interrupt. RCxREG content should be discarded. When calibrating for modes that do not use the SPxBRGH register the user can verify that the SPxBRGL register did not overflow by checking for 00h in the SPxBRGH register.

The BRG auto-baud clock is determined by the BRG16 and BRGH bits as shown in [Table 36-3](#). During ABD, both the SPxBRGH and SPxBRGL registers are used as a 16-bit counter, independent of the BRG16 bit setting. While calibrating the baud rate period, the SPxBRGH and SPxBRGL registers are clocked at 1/8<sup>th</sup> the BRG base clock rate. The resulting byte measurement is the average bit time when clocked at full speed.

**Note:**

4. Write '55h' to TXxREG to load the Sync character into the transmit FIFO buffer.
5. After the Break has been sent, the SENDB bit is reset by hardware and the Sync character is then transmitted.

When the TXxREG becomes empty, as indicated by the TXxIF, the next data byte can be written to TXxREG.

### 36.2.5 Receiving a Break Character

The EUSART module can receive a Break character in two ways.

The first method to detect a Break character uses the FERR bit of the RCxSTA register and the received data as indicated by RCxREG. The Baud Rate Generator is assumed to have been initialized to the expected baud rate.

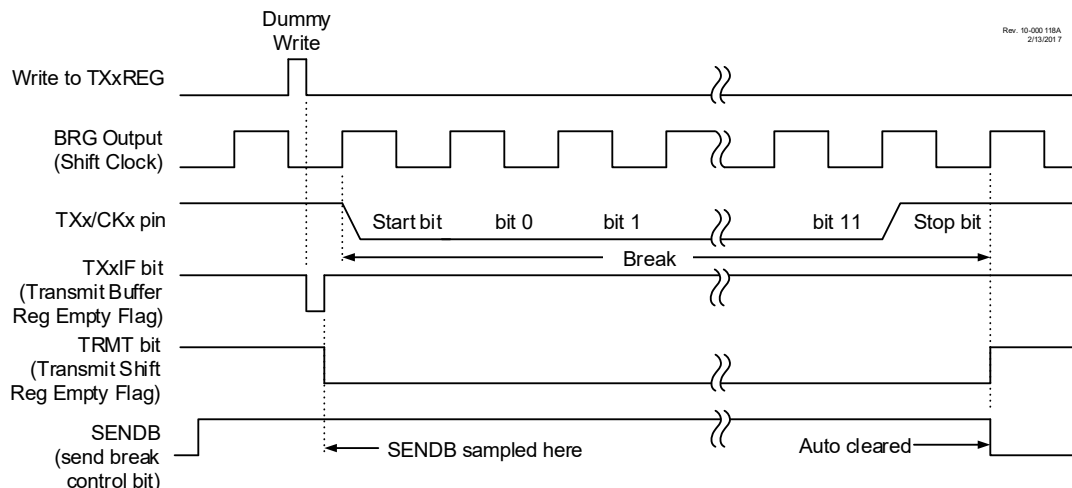
A Break character has been received when all three of the following conditions are true:

- RCxIF bit is set
- FERR bit is set
- RCxREG = 00h

The second method uses the Auto-Wake-up feature described in [36.2.3 Auto-Wake-up on Break](#). By enabling this feature, the EUSART will sample the next two transitions on RX/DT, cause an RCxIF interrupt, and receive the next data byte followed by another interrupt.

Note that following a Break character, the user will typically want to enable the Auto-Baud Detect feature. For both methods, the user can set the ABDEN bit of the BAUDxCON register before placing the EUSART in Sleep mode.

**Figure 36-9. Send Break Character Sequence**



## 36.3 EUSART Synchronous Mode

Synchronous serial communications are typically used in systems with a single master and one or more slaves. The master device contains the necessary circuitry for baud rate generation and supplies the clock for all devices in the system. Slave devices can take advantage of the master clock by eliminating the internal clock generation circuitry.

**37.3.7 SMTxTMR**

**Name:** SMTxTMR  
**Address:** 0x48C,0x50C

SMT Timer Register

Bit	23	22	21	20	19	18	17	16
TMRU[7:0]								
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset	0	0	0	0	0	0	0	0
Bit	15	14	13	12	11	10	9	8
TMRH[7:0]								
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset	0	0	0	0	0	0	0	0
Bit	7	6	5	4	3	2	1	0
TMRL[7:0]								
Access	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Reset	0	0	0	0	0	0	0	0

**Bits 23:16 – TMRU[7:0]** Upper byte of the SMT timer register

**Bits 15:8 – TMRH[7:0]** High byte of the SMT timer register

**Bits 7:0 – TMRL[7:0]** Lower byte of the SMT timer register

- Fully customizable interface: toolbars, toolbar buttons, windows, window placement, etc.
- Call graph window

Project-Based Workspaces:

- Multiple projects
- Multiple tools
- Multiple configurations
- Simultaneous debugging sessions

File History and Bug Tracking:

- Local file history feature
- Built-in support for Bugzilla issue tracker

## **41.2 MPLAB XC Compilers**

The MPLAB XC Compilers are complete ANSI C compilers for all of Microchip's 8, 16, and 32-bit MCU and DSC devices. These compilers provide powerful integration capabilities, superior code optimization and ease of use. MPLAB XC Compilers run on Windows, Linux or MAC OS X.

For easy source level debugging, the compilers provide debug information that is optimized to the MPLAB X IDE.

The free MPLAB XC Compiler editions support all devices and commands, with no time or memory restrictions, and offer sufficient code optimization for most applications.

MPLAB XC Compilers include an assembler, linker and utilities. The assembler generates relocatable object files that can then be archived or linked with other relocatable object files and archives to create an executable file. MPLAB XC Compiler uses the assembler to produce its object file. Notable features of the assembler include:

- Support for the entire device instruction set
- Support for fixed-point and floating-point data
- Command-line interface
- Rich directive set
- Flexible macro language
- MPLAB X IDE compatibility

## **41.3 MPASM Assembler**

The MPASM Assembler is a full-featured, universal macro assembler for PIC10/12/16/18 MCUs.

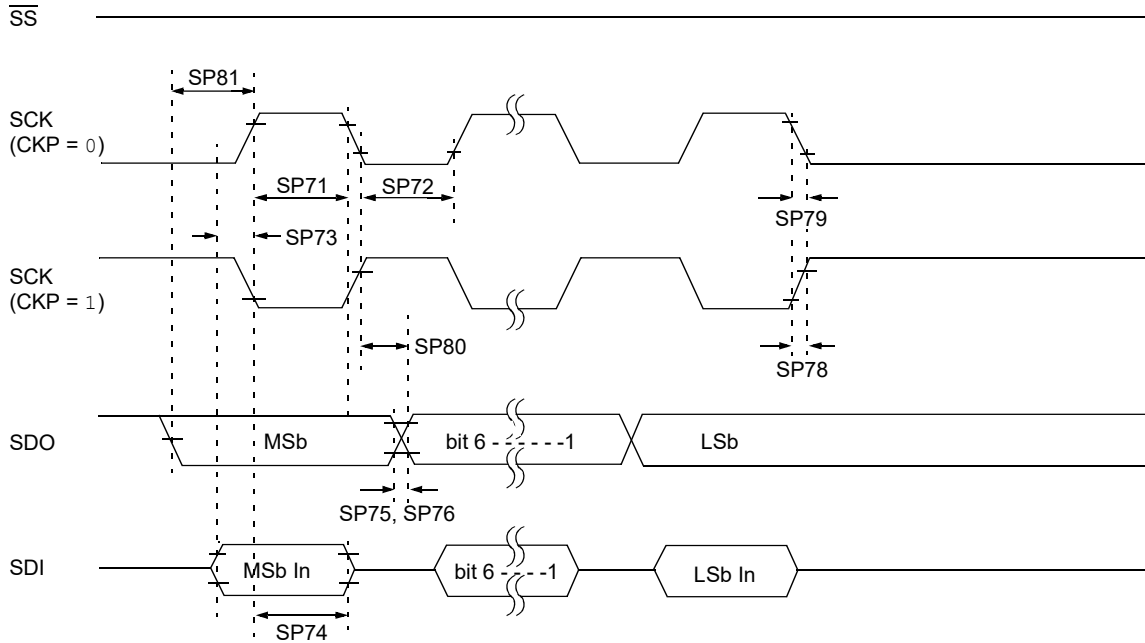
The MPASM Assembler generates relocatable object files for the MPLINK Object Linker, Intel® standard HEX files, MAP files to detail memory usage and symbol reference, absolute LST files that contain source lines and generated machine code, and COFF files for debugging.

The MPASM Assembler features include:

- Integration into MPLAB X IDE projects
- User-defined macros to streamline assembly code

Figure 42-18. SPI Master Mode Timing (CKE = 1, SMP = 1)

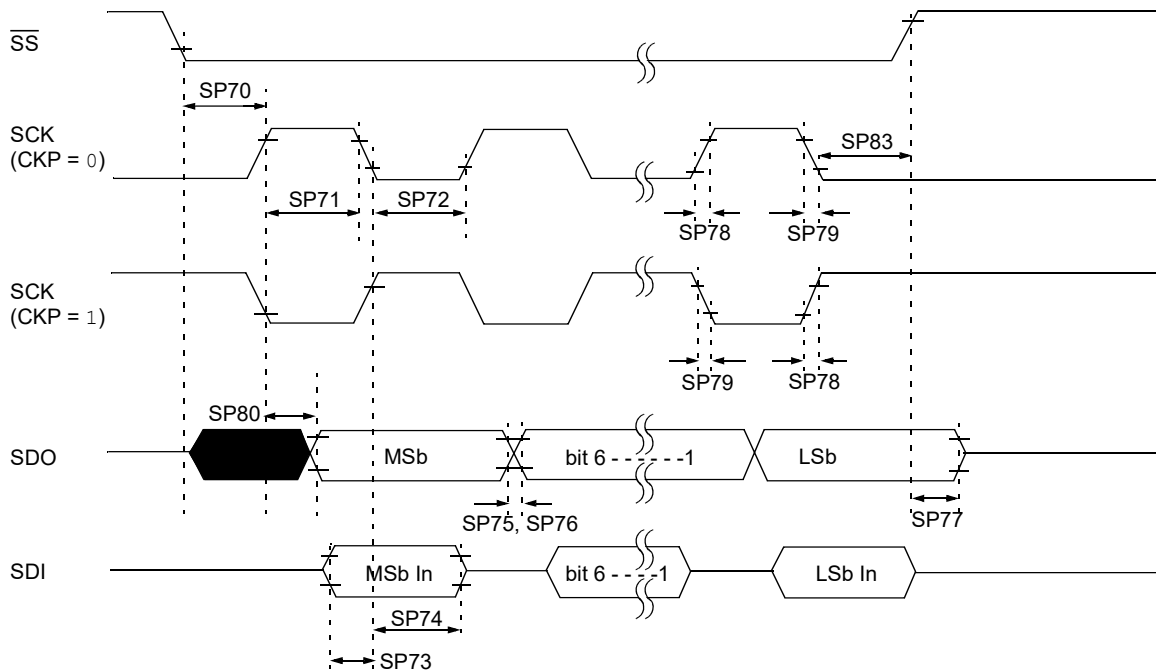
Rev. 30-00086A  
4/6/2017



**Note:** Refer to Figure 42-4 for load conditions.

Figure 42-19. SPI Slave Mode Timing (CKE = 0)

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4/6/2017



**Note:** Refer to Figure 42-4 for load conditions.