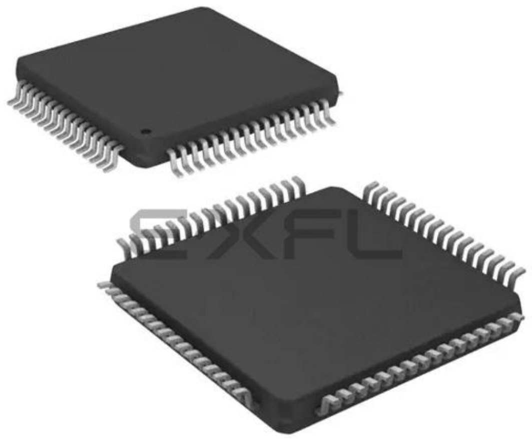

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Peripherals	Brown-out Detect/Reset, Motor Control PWM, QEI, POR, PWM, WDT
Number of I/O	52
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Program Memory Type	FLASH
EEPROM Size	1K x 8
RAM Size	2K x 8
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Data Converters	A/D 16x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
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The core does not support a multi-stage instruction pipeline. However, a single stage instruction prefetch mechanism is used, which accesses and partially decodes instructions a cycle ahead of execution, in order to maximize available execution time. Most instructions execute in a single cycle, with certain exceptions.

The core features a vectored exception processing structure for traps and interrupts, with 62 independent vectors. The exceptions consist of up to 8 traps (of which 4 are reserved) and 54 interrupts. Each interrupt is prioritized based on a user assigned priority between 1 and 7 (1 being the lowest priority and 7 being the highest) in conjunction with a predetermined ‘natural order’. Traps have fixed priorities, ranging from 8 to 15.

2.2 Programmer’s Model

The programmer’s model is shown in Figure 2-1 and consists of 16x16-bit working registers (W0 through W15), 2x40-bit accumulators (ACCA and ACCB), STATUS register (SR), Data Table Page register (TBLPAG), Program Space Visibility Page register (PSVPAG), DO and REPEAT registers (DOSTART, DOEND, DCOUNT and RCOUNT) and Program Counter (PC). The working registers can act as data, address or offset registers. All registers are memory mapped. W0 acts as the W register for file register addressing.

Some of these registers have a shadow register associated with each of them, as shown in Figure 2-1. The shadow register is used as a temporary holding register and can transfer its contents to or from its host register upon the occurrence of an event. None of the shadow registers are accessible directly. The following rules apply for transfer of registers into and out of shadows.

- `PUSH.S` and `POP.S`
  W0, W1, W2, W3, SR (DC, N, OV, Z and C bits only) are transferred.
- `DO` instruction
  DOSTART, DOEND, DCOUNT shadows are pushed on loop start, and popped on loop end.

When a byte operation is performed on a working register, only the Least Significant Byte (LSB) of the target register is affected. However, a benefit of memory mapped working registers is that both the Least and Most Significant Bytes (MSBs) can be manipulated through byte-wide data memory space accesses.

2.2.1 SOFTWARE STACK POINTER/ FRAME POINTER

The dsPIC DSC devices contain a software stack. W15 is the dedicated software Stack Pointer (SP), and will be automatically modified by exception processing and subroutine calls and returns. However, W15 can be referenced by any instruction in the same manner as all other W registers. This simplifies the reading, writing and manipulation of the Stack Pointer (e.g., creating stack frames).

Note:	In order to protect against misaligned stack accesses, W15<0> is always clear.

W15 is initialized to 0x0800 during a Reset. The user may reprogram the SP during initialization to any location within data space.

W14 has been dedicated as a Stack Frame Pointer as defined by the `LNK` and `ULNK` instructions. However, W14 can be referenced by any instruction in the same manner as all other W registers.

2.2.2 STATUS REGISTER

The dsPIC DSC core has a 16-bit STATUS register (SR), the LSB of which is referred to as the SR Low Byte (SRL) and the MSB as the SR High Byte (SRH). See Figure 2-1 for SR layout.

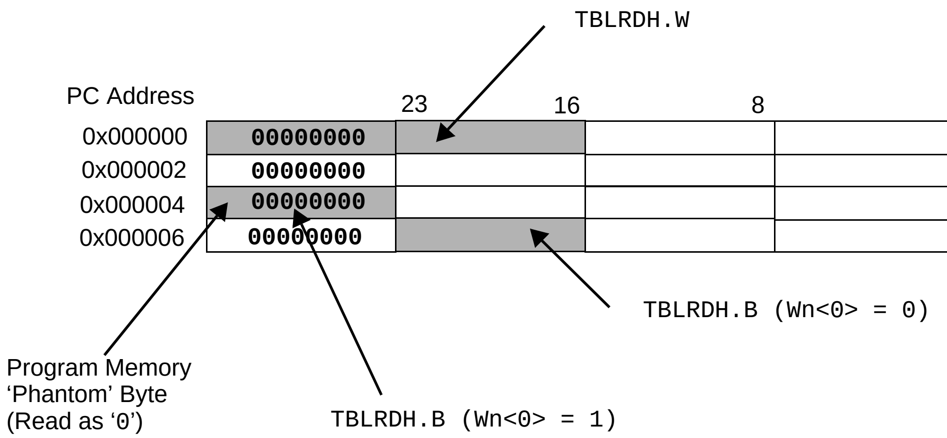
SRL contains all the MCU ALU operation Status flags (including the Z bit), as well as the CPU Interrupt Priority Level Status bits, IPL<2:0>, and the Repeat Active Status bit, RA. During exception processing, SRL is concatenated with the MSB of the PC to form a complete word value which is then stacked.

The upper byte of the SR register contains the DSP Adder/Subtracter Status bits, the DO Loop Active bit (DA) and the Digit Carry (DC) Status bit.

2.2.3 PROGRAM COUNTER

The Program Counter is 23 bits wide. Bit 0 is always clear. Therefore, the PC can address up to 4M instruction words.

FIGURE 3-4: PROGRAM DATA TABLE ACCESS (MSB)



3.1.2 DATA ACCESS FROM PROGRAM MEMORY USING PROGRAM SPACE VISIBILITY

The upper 32 Kbytes of data space may optionally be mapped into any 16K word program space page. This provides transparent access of stored constant data from X data space, without the need to use special instructions (i.e., TBLRDL/H, TBLWTL/H instructions).

Program space access through the data space occurs if the MSb of the data space EA is set and program space visibility is enabled, by setting the PSV bit in the Core Control register (CORCON). The functions of CORCON are discussed in **Section 2.4 "DSP Engine"**.

Data accesses to this area add an additional cycle to the instruction being executed, since two program memory fetches are required.

Note that the upper half of addressable data space is always part of the X data space. Therefore, when a DSP operation uses program space mapping to access this memory region, Y data space should typically contain state (variable) data for DSP operations, whereas X data space should typically contain coefficient (constant) data.

Although each data space address, 0x8000 and higher, maps directly into a corresponding program memory address (see Figure 3-5), only the lower 16 bits of the 24-bit program word are used to contain the data. The upper 8 bits should be programmed to force an illegal instruction to maintain machine robustness. Refer to the *"16-bit MCU and DSC Programmer's Reference Manual"* (DS70157) for details on instruction encoding.

Note that by incrementing the PC by 2 for each program memory word, the Least Significant 15 bits of data space addresses directly map to the Least Significant 15 bits in the corresponding program space addresses. The remaining bits are provided by the Program Space Visibility Page register, PSVPAG<7:0>, as shown in Figure 3-5.

Note:	PSV access is temporarily disabled during table reads/writes.

For instructions which use PSV that are executed outside a REPEAT loop:

- The following instructions will require one instruction cycle in addition to the specified execution time:
  - MAC class of instructions with data operand prefetch
  - MOV instructions
  - MOV.D instructions
- All other instructions will require two instruction cycles in addition to the specified execution time of the instruction.

For instructions that use PSV which are executed inside a REPEAT loop:

- The following instances will require two instruction cycles in addition to the specified execution time of the instruction:
  - Execution in the first iteration
  - Execution in the last iteration
  - Execution prior to exiting the loop due to an interrupt
  - Execution upon re-entering the loop after an interrupt is serviced
- Any other iteration of the REPEAT loop will allow the instruction, accessing data using PSV, to execute in a single cycle.

3.2.2 DATA SPACES

The X data space is used by all instructions and supports all addressing modes. There are separate read and write data buses. The X read data bus is the return data path for all instructions that view data space as combined X and Y address space. It is also the X address space data path for the dual operand read instructions (`MAC` class). The X write data bus is the only write path to data space for all instructions.

The X data space also supports Modulo Addressing for all instructions, subject to addressing mode restrictions. Bit-Reversed Addressing is only supported for writes to X data space.

The Y data space is used in concert with the X data space by the `MAC` class of instructions (`CLR`, `ED`, `EDAC`, `MAC`, `MOVSAC`, `MPY`, `MPY.N` and `MSC`) to provide two concurrent data read paths. No writes occur across the Y bus. This class of instructions dedicates two W register pointers, W10 and W11, to always address Y data space, independent of X data space, whereas W8 and W9 always address X data space. Note that during accumulator write back, the data address space is considered a combination of X and Y data spaces, so the write occurs across the X bus. Consequently, the write can be to any address in the entire data space.

The Y data space can only be used for the data prefetch operation associated with the `MAC` class of instructions. It also supports Modulo Addressing for automated circular buffers. Of course, all other instructions can access the Y data address space through the X data path, as part of the composite linear space.

The boundary between the X and Y data spaces is defined as shown in Figure 3-6 and is not user programmable. Should an EA point to data outside its own assigned address space, or to a location outside physical memory, an all-zero word/byte will be returned. For example, although Y address space is visible by all non-`MAC` instructions using any addressing mode, an attempt by a `MAC` instruction to fetch data from that space, using W8 or W9 (X space pointers), will return 0x0000.

TABLE 3-2: EFFECT OF INVALID MEMORY ACCESSES

Attempted Operation	Data Returned
EA = an unimplemented address	0x0000
W8 or W9 used to access Y data space in a `MAC` instruction	0x0000
W10 or W11 used to access X data space in a `MAC` instruction	0x0000

All effective addresses are 16 bits wide and point to bytes within the data space. Therefore, the data space address range is 64 Kbytes or 32K words.

3.2.3 DATA SPACE WIDTH

The core data width is 16 bits. All internal registers are organized as 16-bit wide words. Data space memory is organized in byte addressable, 16-bit wide blocks.

3.2.4 DATA ALIGNMENT

To help maintain backward compatibility with PIC® devices and improve data space memory usage efficiency, the dsPIC30F instruction set supports both word and byte operations. Data is aligned in data memory and registers as words, but all data space EAs resolve to bytes. Data byte reads will read the complete word, which contains the byte, using the LSb of any EA to determine which byte to select. The selected byte is placed onto the LSB of the X data path (no byte accesses are possible from the Y data path as the `MAC` class of instruction can only fetch words). That is, data memory and registers are organized as two parallel byte wide entities with shared (word) address decode, but separate write lines. Data byte writes only write to the corresponding side of the array or register which matches the byte address.

As a consequence of this byte accessibility, all effective address calculations (including those generated by the DSP operations, which are restricted to word-sized data) are internally scaled to step through word-aligned memory. For example, the core would recognize that Post-Modified Register Indirect Addressing mode, [Ws++], will result in a value of Ws + 1 for byte operations and Ws + 2 for word operations.

All word accesses must be aligned to an even address. Misaligned word data fetches are not supported, so care must be taken when mixing byte and word operations, or translating from 8-bit MCU code. Should a misaligned read or write be attempted, an address error trap will be generated. If the error occurred on a read, the instruction underway is completed, whereas if it occurred on a write, the instruction will be executed but the write will not occur. In either case, a trap will then be executed, allowing the system and/or user to examine the machine state prior to execution of the address fault.

FIGURE 3-8: DATA ALIGNMENT

	MSB (15 … 8)	LSB (7 … 0)	
`0001`	Byte 1	Byte 0	`0000`
`0003`	Byte 3	Byte 2	`0002`
`0005`	Byte 5	Byte 4	`0004`

4.1.3 MOVE AND ACCUMULATOR INSTRUCTIONS

Move instructions and the DSP Accumulator class of instructions provide a greater degree of addressing flexibility than other instructions. In addition to the addressing modes supported by most MCU instructions, Move and Accumulator instructions also support Register Indirect with Register Offset Addressing mode, also referred to as Register Indexed mode.

> **Note:** For the `MOV` instructions, the addressing mode specified in the instruction can differ for the source and destination EA. However, the 4-bit Wb (Register Offset) field is shared between both source and destination (but typically only used by one).

In summary, the following addressing modes are supported by Move and Accumulator instructions:

- Register Direct
- Register Indirect
- Register Indirect Post-Modified
- Register Indirect Pre-Modified
- Register Indirect with Register Offset (Indexed)
- Register Indirect with Literal Offset
- 8-bit Literal
- 16-bit Literal

> **Note:** Not all instructions support all the addressing modes given above. Individual instructions may support different subsets of these addressing modes.

4.1.4 `MAC` INSTRUCTIONS

The dual source operand DSP instructions (`CLR`, `ED`, `EDAC`, `MAC`, `MPY`, `MPY.N`, `MOVSAC` and `MSC`), also referred to as `MAC` instructions, utilize a simplified set of addressing modes to allow the user to effectively manipulate the data pointers through Register Indirect tables.

The two source operand prefetch registers must be a member of the set {W8, W9, W10, W11}. For data reads, W8 and W9 will always be directed to the X RAGU and W10 and W11 will always be directed to the Y AGU. The effective addresses generated (before and after modification) must, therefore, be valid addresses within X data space for W8 and W9 and Y data space for W10 and W11.

> **Note:** Register Indirect with Register Offset Addressing is only available for W9 (in X space) and W11 (in Y space).

In summary, the following addressing modes are supported by the `MAC` class of instructions:

- Register Indirect
- Register Indirect Post-Modified by 2
- Register Indirect Post-Modified by 4
- Register Indirect Post-Modified by 6
- Register Indirect with Register Offset (Indexed)

4.1.5 OTHER INSTRUCTIONS

Besides the various addressing modes outlined above, some instructions use literal constants of various sizes. For example, `BRA` (branch) instructions use 16-bit signed literals to specify the branch destination directly, whereas the `DISI` instruction uses a 14-bit unsigned literal field. In some instructions, such as `ADD Acc`, the source of an operand or result is implied by the opcode itself. Certain operations, such as `NOP`, do not have any operands.

4.2 Modulo Addressing

Modulo Addressing is a method of providing an automated means to support circular data buffers using hardware. The objective is to remove the need for software to perform data address boundary checks when executing tightly looped code, as is typical in many DSP algorithms.

Modulo Addressing can operate in either data or program space (since the data pointer mechanism is essentially the same for both). One circular buffer can be supported in each of the X (which also provides the pointers into program space) and Y data spaces. Modulo Addressing can operate on any W register pointer. However, it is not advisable to use W14 or W15 for Modulo Addressing, since these two registers are used as the Stack Frame Pointer and Stack Pointer, respectively.

In general, any particular circular buffer can only be configured to operate in one direction, as there are certain restrictions on the buffer start address (for incrementing buffers) or end address (for decrementing buffers) based upon the direction of the buffer.

The only exception to the usage restrictions is for buffers which have a power-of-2 length. As these buffers satisfy the start and end address criteria, they may operate in a Bidirectional mode, (i.e., address boundary checks will be performed on both the lower and upper address boundaries).

TABLE 4-2: BIT-REVERSED ADDRESS SEQUENCE (16-ENTRY)

Normal Address					Bit-Reversed Address				
A3	A2	A1	A0	Decimal	A3	A2	A1	A0	Decimal
0	0	0	0	0	0	0	0	0	0
0	0	0	1	1	1	0	0	0	8
0	0	1	0	2	0	1	0	0	4
0	0	1	1	3	1	1	0	0	12
0	1	0	0	4	0	0	1	0	2
0	1	0	1	5	1	0	1	0	10
0	1	1	0	6	0	1	1	0	6
0	1	1	1	7	1	1	1	0	14
1	0	0	0	8	0	0	0	1	1
1	0	0	1	9	1	0	0	1	9
1	0	1	0	10	0	1	0	1	5
1	0	1	1	11	1	1	0	1	13
1	1	0	0	12	0	0	1	1	3
1	1	0	1	13	1	0	1	1	11
1	1	1	0	14	0	1	1	1	7
1	1	1	1	15	1	1	1	1	15

TABLE 4-3: BIT-REVERSED ADDRESS MODIFIER VALUES FOR XBREV REGISTER

Buffer Size (Words)	XB<14:0> Bit-Reversed Address Modifier Value
4096	0x0800
2048	0x0400
1024	0x0200
512	0x0100
256	0x0080
128	0x0040
64	0x0020
32	0x0010
16	0x0008
8	0x0004
4	0x0002
2	0x0001

6.4 RTSP Operation

The dsPIC30F Flash program memory is organized into rows and panels. Each row consists of 32 instructions, or 96 bytes. Each panel consists of 128 rows, or 4K x 24 instructions. RTSP allows the user to erase one row (32 instructions) at a time and to program 32 instructions at one time.

Each panel of program memory contains write latches that hold 32 instructions of programming data. Prior to the actual programming operation, the write data must be loaded into the panel write latches. The data to be programmed into the panel is loaded in sequential order into the write latches; instruction 0, instruction 1, etc. The addresses loaded must always be from an even group of 32 boundary.

The basic sequence for RTSP programming is to set up a Table Pointer, then do a series of `TBLWT` instructions to load the write latches. Programming is performed by setting the special bits in the NVMCON register. 32 `TBLWTL` and 32 `TBLWTH` instructions are required to load the 32 instructions.

All of the table write operations are single-word writes (2 instruction cycles), because only the table latches are written.

After the latches are written, a programming operation needs to be initiated to program the data.

The Flash program memory is readable, writable and erasable during normal operation over the entire V_{DD} range.

6.5 RTSP Control Registers

The four SFRs used to read and write the program Flash memory are:

- NVMCON
- NVMADR
- NVMADRU
- NVMKEY

6.5.1 NVMCON REGISTER

The NVMCON register controls which blocks are to be erased, which memory type is to be programmed and start of the programming cycle.

6.5.2 NVMADR REGISTER

The NVMADR register is used to hold the lower two bytes of the Effective Address. The NVMADR register captures the EA<15:0> of the last table instruction that has been executed and selects the row to write.

6.5.3 NVMADRU REGISTER

The NVMADRU register is used to hold the upper byte of the Effective Address. The NVMADRU register captures the EA<23:16> of the last table instruction that has been executed.

6.5.4 NVMKEY REGISTER

NVMKEY is a write-only register that is used for write protection. To start a programming or an erase sequence, the user must consecutively write 0x55 and 0xAA to the NVMKEY register. Refer to **Section 6.6 "Programming Operations"** for further details.

Note:	The user can also directly write to the NVMADR and NVMADRU registers to specify a program memory address for erasing or programming.

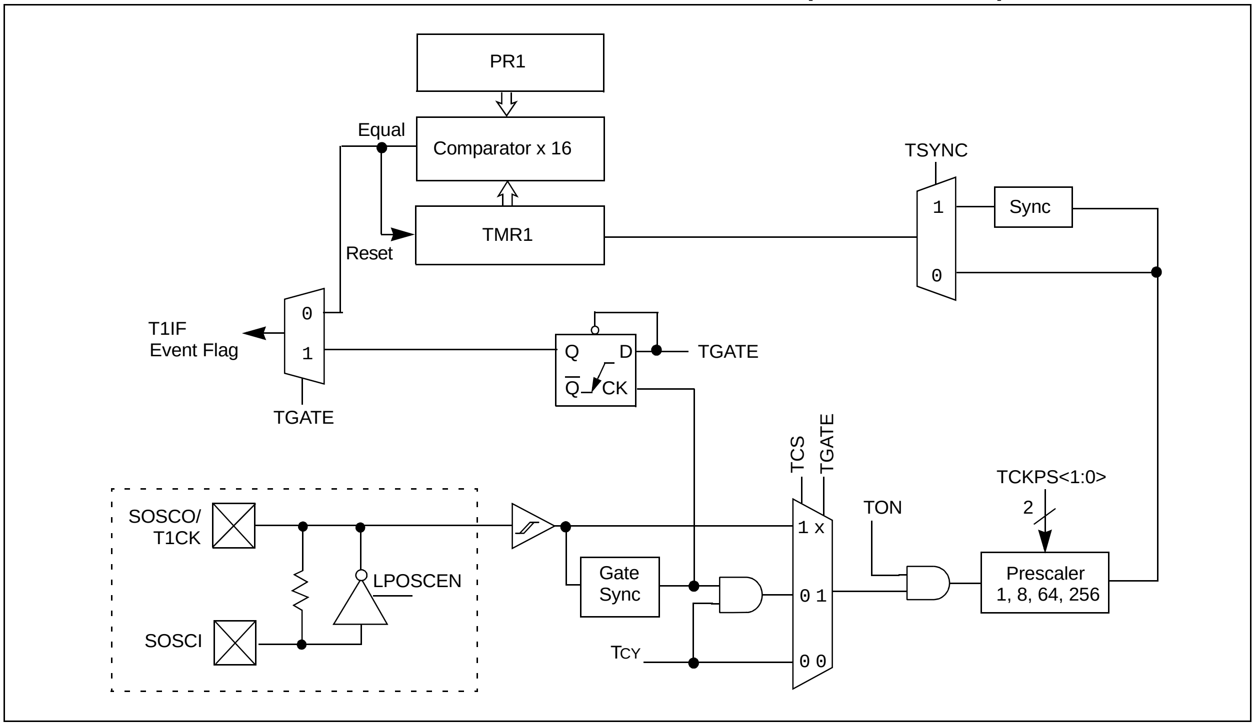
TABLE 6-1: NVM REGISTER MAP[1]

File Name	Addr.	Bit 15	Bit 14	Bit 13	Bit 12	Bit 11	Bit 10	Bit 9	Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	All Resets
NVMCON	0760	WR	WREN	WRERR	—	—	—	—	TWRI	—	PROGOP<6:0>							0000 0000 0000 0000
NVMADR	0762	NVMADR<15:0>																uuuu uuuu uuuu uuuu
NVMADRU	0764	—	—	—	—	—	—	—	—	NVMADR<23:16>								0000 0000 uuuu uuuu
NVMKEY	0766	—	—	—	—	—	—	—	—	KEY<7:0>								0000 0000 0000 0000

Legend: u = uninitialized bit; — = unimplemented bit, read as '0'

Note 1: Refer to the *"dsPIC30F Family Reference Manual"* (DS70046) for descriptions of register bit fields.

FIGURE 9-1: 16-BIT TIMER1 MODULE BLOCK DIAGRAM (TYPE A TIMER)



9.1 Timer Gate Operation

The 16-bit timer can be placed in the Gated Time Accumulation mode. This mode allows the internal TCY to increment the respective timer when the gate input signal (T1CK pin) is asserted high. Control bit TGATE (T1CON<6>) must be set to enable this mode. The timer must be enabled (TON = 1) and the timer clock source set to internal (TCS = 0).

When the CPU goes into the Idle mode, the timer will stop incrementing unless TSIDL = 0. If TSIDL = 1, the timer will resume the incrementing sequence upon termination of the CPU Idle mode.

9.2 Timer Prescaler

The input clock (FOSC/4 or external clock) to the 16-bit timer has a prescale option of 1:1, 1:8, 1:64 and 1:256, selected by control bits, TCKPS<1:0> (T1CON<5:4>). The prescaler counter is cleared when any of the following occurs:

- A write to the TMR1 register
- Clearing the TON bit (T1CON<15>)
- A device Reset, such as POR and BOR

However, if the timer is disabled (TON = 0), then the timer prescaler cannot be reset since the prescaler clock is halted.

TMR1 is not cleared when T1CON is written. It is cleared by writing to the TMR1 register.

9.3 Timer Operation During Sleep Mode

During CPU Sleep mode, the timer will operate if:

- The timer module is enabled (TON = 1) and
- The timer clock source is selected as external (TCS = 1) and
- The TSYNC bit (T1CON<2>) is asserted to a logic '0', which defines the external clock source as asynchronous

When all three conditions are true, the timer will continue to count up to the Period register and be reset to 0x0000.

When a match between the timer and the Period register occurs, an interrupt can be generated, if the respective timer interrupt enable bit is asserted.

9.4 Timer Interrupt

The 16-bit timer has the ability to generate an interrupt on period match. When the timer count matches the Period register, the T1IF bit is asserted and an interrupt will be generated, if enabled. The T1IF bit must be cleared in software. The Timer Interrupt Flag, T1IF, is located in the IFS0 Control register in the interrupt controller.

When the Gated Time Accumulation mode is enabled, an interrupt will also be generated on the falling edge of the gate signal (at the end of the accumulation cycle).

Enabling an interrupt is accomplished via the respective Timer Interrupt Enable bit, T1IE. The Timer Interrupt Enable bit is located in the IEC0 Control register in the interrupt controller.

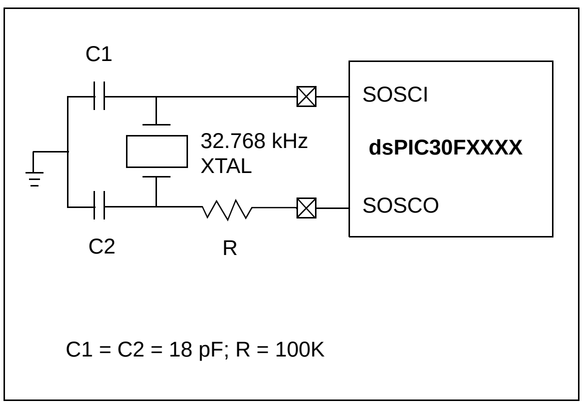
9.5 Real-Time Clock

Timer1, when operating in Real-Time Clock (RTC) mode, provides time-of-day and event time-stamping capabilities. Key operational features of the RTC are:

- Operation from 32 kHz LP oscillator
- 8-bit prescaler
- Low power
- Real-Time Clock Interrupts

These operating modes are determined by setting the appropriate bit(s) in the T1CON Control register.

FIGURE 9-2: RECOMMENDED COMPONENTS FOR TIMER1 LP OSCILLATOR RTC



9.5.1 RTC OSCILLATOR OPERATION

When the TON = `1`, TCS = `1` and TGATE = `0`, the timer increments on the rising edge of the 32 kHz LP oscillator output signal, up to the value specified in the Period register, and is then reset to ‘`0`’.

The TSYNC bit must be asserted to a logic ‘`0`’ (Asynchronous mode) for correct operation.

Enabling LPOSCEN (OSCCON<1>) will disable the normal Timer and Counter modes and enable a timer carry-out wake-up event.

When the CPU enters Sleep mode, the RTC will continue to operate, provided the 32 kHz external crystal oscillator is active and the control bits have not been changed. The TSIDL bit should be cleared to ‘`0`’ in order for RTC to continue operation in Idle mode.

9.5.2 RTC INTERRUPTS

When an interrupt event occurs, the respective interrupt flag, T1IF, is asserted and an interrupt will be generated, if enabled. The T1IF bit must be cleared in software. The respective Timer Interrupt Flag, T1IF, is located in the IFS0 Status register in the interrupt controller.

Enabling an interrupt is accomplished via the respective Timer Interrupt Enable bit, T1IE. The Timer Interrupt Enable bit is located in the IEC0 Control register in the interrupt controller.

NOTES:

13.0 OUTPUT COMPARE MODULE

Note:	This data sheet summarizes features of this group of dsPIC30F devices and is not intended to be a complete reference source. For more information on the CPU, peripherals, register descriptions and general device functionality, refer to the "*dsPIC30F Family Reference Manual*" (DS70046). For more information on the device instruction set and programming, refer to the *"16-bit MCU and DSC Programmer's Reference Manual"* (DS70157).

This section describes the output compare module and associated operational modes. The features provided by this module are useful in applications requiring operational modes such as:

- Generation of Variable Width Output Pulses
- Power Factor Correction

Figure 13-1 depicts a block diagram of the output compare module.

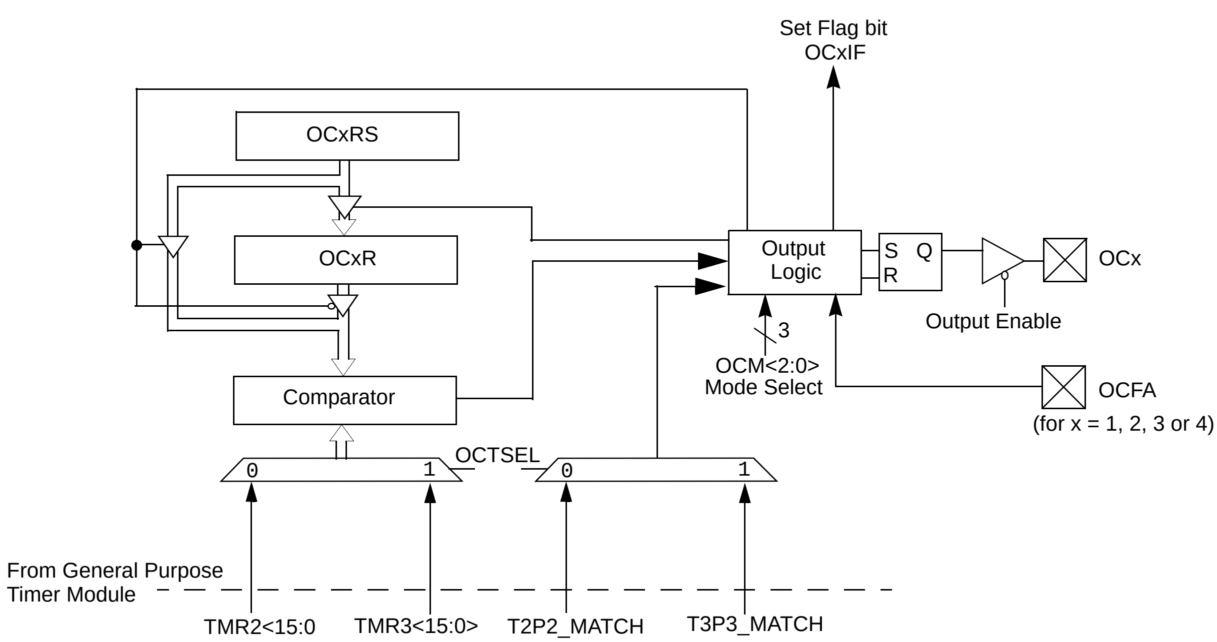
The key operational features of the output compare module include:

- Timer2 and Timer3 Selection mode
- Simple Output Compare Match mode
- Dual Output Compare Match mode
- Simple PWM mode
- Output Compare during Sleep and Idle modes
- Interrupt on Output Compare/PWM Event

These operating modes are determined by setting the appropriate bits in the 16-bit OCxCON SFR (where x = 1,2,3,...,N). The dsPIC30F5015/5016 device has 4 compare channels.

OCxRS and OCxR in the figure represent the Dual Compare registers. In the Dual Compare mode, the OCxR register is used for the first compare and OCxRS is used for the second compare.

FIGURE 13-1: OUTPUT COMPARE MODE BLOCK DIAGRAM



Note:	Where 'x' is shown, reference is made to the registers associated with the respective output compare channels 1 through N.

15.11 PWM Output and Polarity Control

There are three device Configuration bits associated with the PWM module that provide PWM output pin control:

- HPOL Configuration bit
- LPOL Configuration bit
- PWMPIN Configuration bit

These three bits in the FBORPOR Configuration register (see **Section 21.6 "Device Configuration Registers"**) work in conjunction with the four PWM Enable bits (PWMEN<4:1>) located in the PWMCON1 SFR. The Configuration bits and PWM enable bits ensure that the PWM pins are in the correct states after a device Reset occurs. The PWMPIN configuration fuse allows the PWM module outputs to be optionally enabled on a device Reset. If PWMPIN = `0`, the PWM outputs will be driven to their inactive states at Reset. If PWMPIN = `1` (default), the PWM outputs will be tri-stated. The HPOL bit specifies the polarity for the PWMxH outputs, whereas the LPOL bit specifies the polarity for the PWMxL outputs.

15.11.1 OUTPUT PIN CONTROL

The PENxH and PENxL control bits in the PWMCON1 SFR enable each high PWM output pin and each low PWM output pin, respectively. If a particular PWM output pin is not enabled, it is treated as a general purpose I/O pin.

15.12 PWM FAULT Pins

There are two Fault pins ($\overline{\text{FLTA}}$ and $\overline{\text{FLTB}}$) associated with the PWM module. When asserted, these pins can optionally drive each of the PWM I/O pins to a defined state.

15.12.1 FAULT PIN ENABLE BITS

The FLTACON and FLTBCON SFRs each have 4 control bits that determine whether a particular pair of PWM I/O pins is to be controlled by the Fault input pin. To enable a specific PWM I/O pin pair for Fault overrides, the corresponding bit should be set in the FLTACON or FLTBCON register.

If all enable bits are cleared in the FLTACON or FLTBCON registers, then the corresponding Fault input pin has no effect on the PWM module and the pin may be used as a general purpose interrupt or I/O pin.

Note:	The Fault pin logic can operate independent of the PWM logic. If all the enable bits in the FLTACON/FLTBCON registers are cleared, then the Fault pin(s) could be used as general purpose interrupt pin(s). Each Fault pin has an interrupt vector, interrupt flag bit and interrupt priority bits associated with it.

15.12.2 FAULT STATES

The FLTACON and FLTBCON Special Function Registers have 8 bits each that determine the state of each PWM I/O pin when it is overridden by a Fault input. When these bits are cleared, the PWM I/O pin is driven to the inactive state. If the bit is set, the PWM I/O pin will be driven to the active state. The active and inactive states are referenced to the polarity defined for each PWM I/O pin (HPOL and LPOL polarity control bits).

A special case exists when a PWM module I/O pair is in the Complementary mode and both pins are programmed to be active on a Fault condition. The PWMxH pin always has priority in the Complementary mode, so that both I/O pins cannot be driven active simultaneously.

15.12.3 FAULT PIN PRIORITY

If both Fault input pins have been assigned to control a particular PWM I/O pin, the Fault state programmed for the Fault A input pin will take priority over the Fault B input pin.

15.12.4 FAULT INPUT MODES

Each of the Fault input pins has two modes of operation:

- **Latched Mode:** When the Fault pin is driven low, the PWM outputs will go to the states defined in the FLTACON/FLTBCON register. The PWM outputs will remain in this state until the Fault pin is driven high and the corresponding interrupt flag has been cleared in software. When both of these actions have occurred, the PWM outputs will return to normal operation at the beginning of the next PWM cycle or half-cycle boundary. If the interrupt flag is cleared before the Fault condition ends, the PWM module will wait until the Fault pin is no longer asserted, to restore the outputs.
- **Cycle-by-Cycle Mode:** When the Fault input pin is driven low, the PWM outputs remain in the defined Fault states for as long as the Fault pin is held low. After the Fault pin is driven high, the PWM outputs return to normal operation at the beginning of the following PWM cycle or half-cycle boundary.

The operating mode for each Fault input pin is selected using the FLTAM and FLTBM control bits in the FLTACON and FLTBCON Special Function Registers.

Each of the Fault pins can be controlled manually in software.

15.13 PWM Update Lockout

For a complex PWM application, the user may need to write up to four Duty Cycle registers and the Time Base Period register, PTPER, at a given time. In some applications, it is important that all buffer registers be written before the new duty cycle and period values are loaded for use by the module.

The PWM update lockout feature is enabled by setting the UDIS control bit and clearing the IUE control bit in the PWMCON2 SFR. The UDIS bit affects all Duty Cycle Buffer registers and the PWM time base period buffer, PTPER. No duty cycle changes or period value changes will have effect while UDIS = 1.

If the IUE bit is set, any change to the Duty Cycle registers is immediately updated regardless of the bit state of the UDI. The PWM Period register update (PTPER) is not affected by the IUE control bit.

15.14 PWM Special Event Trigger

The PWM module has a special event trigger that allows A/D conversions to be synchronized to the PWM time base. The A/D sampling and conversion time may be programmed to occur at any point within the PWM period. The special event trigger allows the user to minimize the delay between the time when A/D conversion results are acquired and the time when the duty cycle value is updated.

The PWM special event trigger has an SFR named SEVTCMP, and five control bits to control its operation. The PTMR value for which a special event trigger should occur is loaded into the SEVTCMP register. When the PWM time base is in an Up/Down Counting mode, an additional control bit is required to specify the counting phase for the special event trigger. The count phase is selected using the SEVTDIR control bit in the SEVTCMP SFR. If the SEVTDIR bit is cleared, the special event trigger will occur on the upward counting cycle of the PWM time base. If the SEVTDIR bit is set, the special event trigger will occur on the downward count cycle of the PWM time base. The SEVTDIR control bit has no effect unless the PWM time base is configured for an Up/Down Counting mode.

15.14.1 SPECIAL EVENT TRIGGER POSTSCALER

The PWM special event trigger has a postscaler that allows a 1:1 to 1:16 postscale ratio. The postscaler is configured by writing the SEVOPS<3:0> control bits in the PWMCON2 SFR.

The special event output postscaler is cleared on the following events:

- Any write to the SEVTCMP register
- Any device Reset

15.15 PWM Operation During CPU Sleep Mode

The Fault A and Fault B input pins have the ability to wake the CPU from Sleep mode. The PWM module generates an interrupt if either of the Fault pins is driven low while in Sleep.

15.16 PWM Operation During CPU Idle Mode

The PTCON SFR contains a PTSIDL control bit. This bit determines if the PWM module will continue to operate or stop when the device enters Idle mode. If PTSIDL = 0, the module will continue to operate. If PTSIDL = 1, the module will stop operation as long as the CPU remains in Idle mode.

TABLE 17-2: I²C™ REGISTER MAP(1)

SFR Name	Addr.	Bit 15	Bit 14	Bit 13	Bit 12	Bit 11	Bit 10	Bit 9	Bit 8	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset State
I2CRCV	0200	—	—	—	—	—	—	—	—	Receive Register								0000 0000 0000 0000
I2CTRN	0202	—	—	—	—	—	—	—	—	Transmit Register								0000 0000 1111 1111
I2CBRG	0204	—	—	—	—	—	—	—	Baud Rate Generator									0000 0000 0000 0000
I2CCON	0206	I2CEN	—	I2CSIDL	SCLREL	IPMIEN	A10M	DISSLW	SMEN	GCEN	STREN	ACKDT	ACKEN	RCEN	PEN	RSEN	SEN	0001 0000 0000 0000
I2CSTAT	0208	ACKSTAT	TRSTAT	—	—	—	BCL	GCSTAT	ADD10	IWCOL	I2COV	D_A	P	S	R_W	RBF	TBF	0000 0000 0000 0000
I2CADD	020A	—	—	—	—	—	—	Address Register										0000 0000 0000 0000

Legend: — = unimplemented bit, read as '0'

Note 1: Refer to the *"dsPIC30F Family Reference Manual"* (DS70046) for descriptions of register bit fields.

18.2 Enabling and Setting Up UART

18.2.1 ENABLING THE UART

The UART module is enabled by setting the UARTEN bit in the UxMODE register (where x = 1). Once enabled, the UxTX and UxRX pins are configured as an output and an input respectively, overriding the TRIS and LATCH register bit settings for the corresponding I/O port pins. The UxTX pin is at logic '1' when no transmission is taking place.

18.2.2 DISABLING THE UART

The UART module is disabled by clearing the UARTEN bit in the UxMODE register. This is the default state after any Reset. If the UART is disabled, all I/O pins operate as port pins under the control of the latch and TRIS bits of the corresponding port pins.

Disabling the UART module resets the buffers to empty states. Any data characters in the buffers are lost, and the baud rate counter is reset.

All error and status flags associated with the UART module are reset when the module is disabled. The URXDA, OERR, FERR, PERR, UTXEN, UTXBRK and UTXBF bits are cleared, whereas RIDLE and TRMT are set. Other control bits, including ADDEN, URXISEL<1:0>, UTXISEL, as well as the UxMODE and UxBRG registers, are not affected.

Clearing the UARTEN bit while the UART is active will abort all pending transmissions and receptions and reset the module as defined above. Re-enabling the UART will restart the UART in the same configuration.

18.2.3 SETTING UP DATA, PARITY AND STOP BIT SELECTIONS

Control bits, PDSEL<1:0>, in the UxMODE register are used to select the data length and parity used in the transmission. The data length may either be 8 bits with even, odd or no parity, or 9 bits with no parity.

The STSEL bit determines whether one or two Stop bits will be used during data transmission.

The default (Power-on) setting of the UART is 8 bits, no parity, 1 Stop bit (typically represented as 8, N, 1).

18.3 Transmitting Data

18.3.1 TRANSMITTING IN 8-BIT DATA MODE

The following steps must be performed in order to transmit 8-bit data:

1. Set up the UART:
   First, the data length, parity and number of Stop bits must be selected. Then, the Transmit and Receive Interrupt enable and priority bits are setup in the UxMODE and UxSTA registers. Also, the appropriate baud rate value must be written to the UxBRG register.
2. Enable the UART by setting the UARTEN bit (UxMODE<15>).
3. Set the UTXEN bit (UxSTA<10>), thereby enabling a transmission.

   > **Note:** The UTXEN bit must be set after the UARTEN bit is set to enable UART transmissions.

4. Write the byte to be transmitted to the lower byte of UxTXREG. The value will be transferred to the Transmit Shift register (UxTSR) immediately and the serial bit stream will start shifting out during the next rising edge of the baud clock. Alternatively, the data byte may be written while UTXEN = 0, following which, the user may set UTXEN. This will cause the serial bit stream to begin immediately because the baud clock will start from a cleared state.
5. A Transmit interrupt will be generated depending on the value of the interrupt control bit UTXISEL (UxSTA<15>).

18.3.2 TRANSMITTING IN 9-BIT DATA MODE

The sequence of steps involved in the transmission of 9-bit data is similar to 8-bit transmission, except that a 16-bit data word (of which the upper 7 bits are always clear) must be written to the UxTXREG register.

18.3.3 TRANSMIT BUFFER (UxTXB)

The transmit buffer is 9 bits wide and 4 characters deep. Including the Transmit Shift register (UxTSR), the user effectively has a 5-deep FIFO (First-In, First-Out) buffer. The UTXBF Status bit (UxSTA<9>) indicates whether the transmit buffer is full.

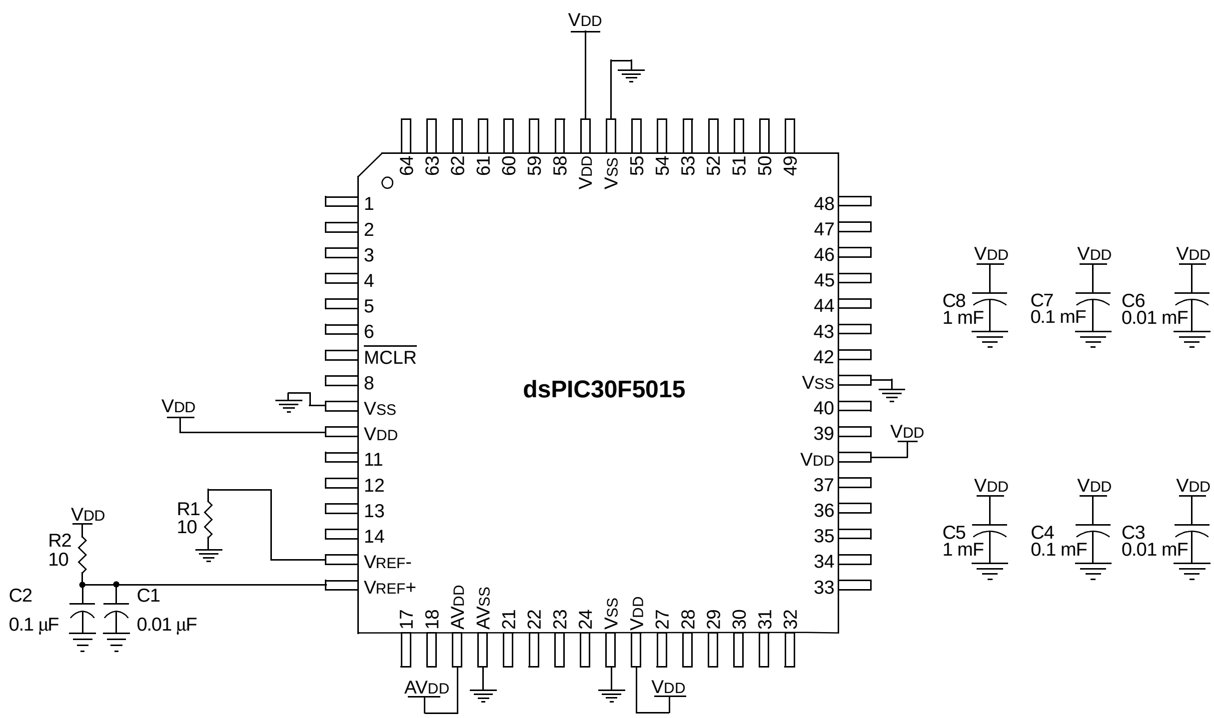
If a user attempts to write to a full buffer, the new data will not be accepted into the FIFO, and no data shift will occur within the buffer. This enables recovery from a buffer overrun condition.

The FIFO is reset during any device Reset, but is not affected when the device enters or wakes up from a Power-Saving mode.

The configuration guidelines give the required setup values for the conversion speeds above 500 ksps, since they require external VREF pins usage and there are some differences in the configuration procedure. Configuration details that are not critical to the conversion speed have been omitted.

The following figure depicts the recommended circuit for the conversion rates above 500 ksps.

FIGURE 20-2: ADC VOLTAGE REFERENCE SCHEMATIC



20.7.1 1 Msps CONFIGURATION GUIDELINE

The configuration for 1 Msps operation is dependent on whether a single input pin is to be sampled or whether multiple pins will be sampled.

20.7.1.1 Single Analog Input

For conversions at 1 Msps for a single analog input, at least two sample and hold channels must be enabled. The analog input multiplexer must be configured so that the same input pin is connected to both sample and hold channels. The ADC converts the value held on one S/H channel, while the second S/H channel acquires a new input sample.

20.7.1.2 Multiple Analog Inputs

The ADC can also be used to sample multiple analog inputs using multiple sample and hold channels. In this case, the total 1 Msps conversion rate is divided among the different input signals. For example, four inputs can be sampled at a rate of 250 ksps for each signal or two inputs could be sampled at a rate of 500 ksps for each signal. Sequential sampling must be used in this configuration to allow adequate sampling time on each input.

TABLE 24-8: DC CHARACTERISTICS: POWER-DOWN CURRENT (IPD)

DC CHARACTERISTICS			Standard Operating Conditions: 2.5V to 5.5V (unless otherwise stated) Operating temperature -40°C ≤TA ≤+85°C for Industrial -40°C ≤TA ≤+125°C for Extended			
Parameter No.	Typical(1)	Max	Units	Conditions		
Power Down Current (IPD)(2)						
DC60a	0.2	—	μA	25°C	3.3V	Base Power Down Current
DC60b	0.7	40	μA	85°C		
DC60c	12	65	μA	125°C		
DC60e	0.4	—	μA	25°C	5V	
DC60f	1.7	55	μA	85°C		
DC60g	16	90	μA	125°C		
DC61a	10	30	μA	25°C	3.3V	Watchdog Timer Current: ΔIWDT(3)
DC61b	12	30	μA	85°C		
DC61c	12	30	μA	125°C		
DC61e	20	40	μA	25°C	5V	
DC61f	22	40	μA	85°C		
DC61g	23	40	μA	125°C		
DC62a	4	10	μA	25°C	3.3V	Timer 1 w/32 kHz Crystal: ΔITI32(3)
DC62b	5	10	μA	85°C		
DC62c	4	10	μA	125°C		
DC62e	4	15	μA	25°C	5V	
DC62f	6	15	μA	85°C		
DC62g	5	15	μA	125°C		
DC63a	33	65	μA	25°C	3.3V	BOR On: ΔIBOR(3)
DC63b	38	65	μA	85°C		
DC63c	39	65	μA	125°C		
DC63e	38	70	μA	25°C	5V	
DC63f	41	70	μA	85°C		
DC63g	42	70	μA	125°C		

Note 1: Data in “Typ” column is at 5V, 25°C unless otherwise stated. Parameters are for design guidance only and are not tested.

2: These parameters are characterized but not tested in manufacturing.

3: The Δ current is the additional current consumed when the module is enabled. This current should be added to the base IPD current.

TABLE 24-41: 10-BIT HIGH-SPEED A/D CONVERSION TIMING REQUIREMENTS

AC CHARACTERISTICS			Standard Operating Conditions: 2.7V to 5.5V (unless otherwise stated) Operating temperature -40°C ≤TA ≤+85°C for Industrial -40°C ≤TA ≤+125°C for Extended				
Param No.	**Symbol**	**Characteristic**	**Min.**	**Typ(1)**	**Max.**	**Units**	**Conditions**
Clock Parameters							
AD50	TAD	A/D Clock Period	—	83.33(2)	—	ns	See Table 20-1(3)
AD51	tRC	A/D Internal RC Oscillator Period	700	900	1100	ns	—
Conversion Rate							
AD55	tCONV	Conversion Time	—	12 TAD	—	—	—
AD56	FCNV	Throughput Rate	—	1.0	—	Msps	See Table 20-1(3)
AD57	TSAMP	Sample Time	—	1 TAD	—	—	See Table 20-1(3)
Timing Parameters							
AD60	tPCS	Conversion Start from Sample Trigger(3)	—	1.0 TAD	—	—	Auto-Convert Trigger (SSRC = 111) not selected
AD61	tPSS	Sample Start from Setting Sample (SAMP) Bit	0.5 TAD	—	1.5 TAD	—	—
AD62	tCSS	Conversion Completion to Sample Start (ASAM = 1)(4)	—	0.5 TAD	—	ns	—
AD63	tDPU(4)	Time to Stabilize Analog Stage from A/D Off to A/D On(4)	—	—	20	µs	—

Note 1: These parameters are characterized but not tested in manufacturing.

2: Operating Temperature: -40°C to +85°C

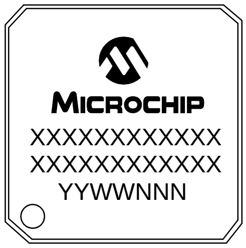
3: Because the sample caps will eventually lose charge, clock rates below 10 kHz can affect linearity performance, especially at elevated temperatures.

4: tDPU is the time required for the ADC module to stabilize when it is turned on (ADCON1<ADON> = 1). During this time the ADC result is indeterminate.

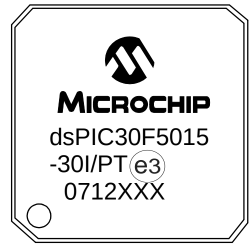
25.0 PACKAGING INFORMATION

25.1 Package Marking Information

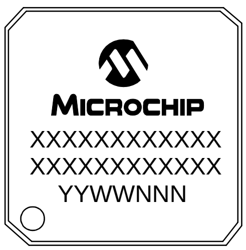
64-Lead TQFP



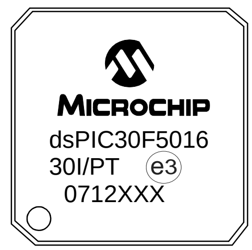
Example



80-Lead TQFP



Example



Legend:	XX...X	Customer-specific information
	Y	Year code (last digit of calendar year)
	YY	Year code (last 2 digits of calendar year)
	WW	Week code (week of January 1 is week '01')
	NNN	Alphanumeric traceability code
	e3	Pb-free JEDEC designator for Matte Tin (Sn)
	*	This package is Pb-free. The Pb-free JEDEC designator (e3) can be found on the outer packaging for this package.

Note: In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line, thus limiting the number of available characters for customer-specific information.