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Details

Product Status	Active
Core Processor	S08
Core Size	8-Bit
Speed	40MHz
Connectivity	I ² C, SCI, SPI
Peripherals	LCD, LVD, POR, PWM, WDT
Number of I/O	18
Program Memory Size	60KB (60K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	4K x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.6V
Data Converters	A/D 2x12b
Oscillator Type	External
Operating Temperature	0°C ~ 70°C (TA)
Mounting Type	Surface Mount
Package / Case	64-LQFP
Supplier Device Package	64-LQFP (10x10)
Purchase URL	https://www.e-xfl.com/pro/item?MUrl=&PartUrl=mc9s08lc60lh

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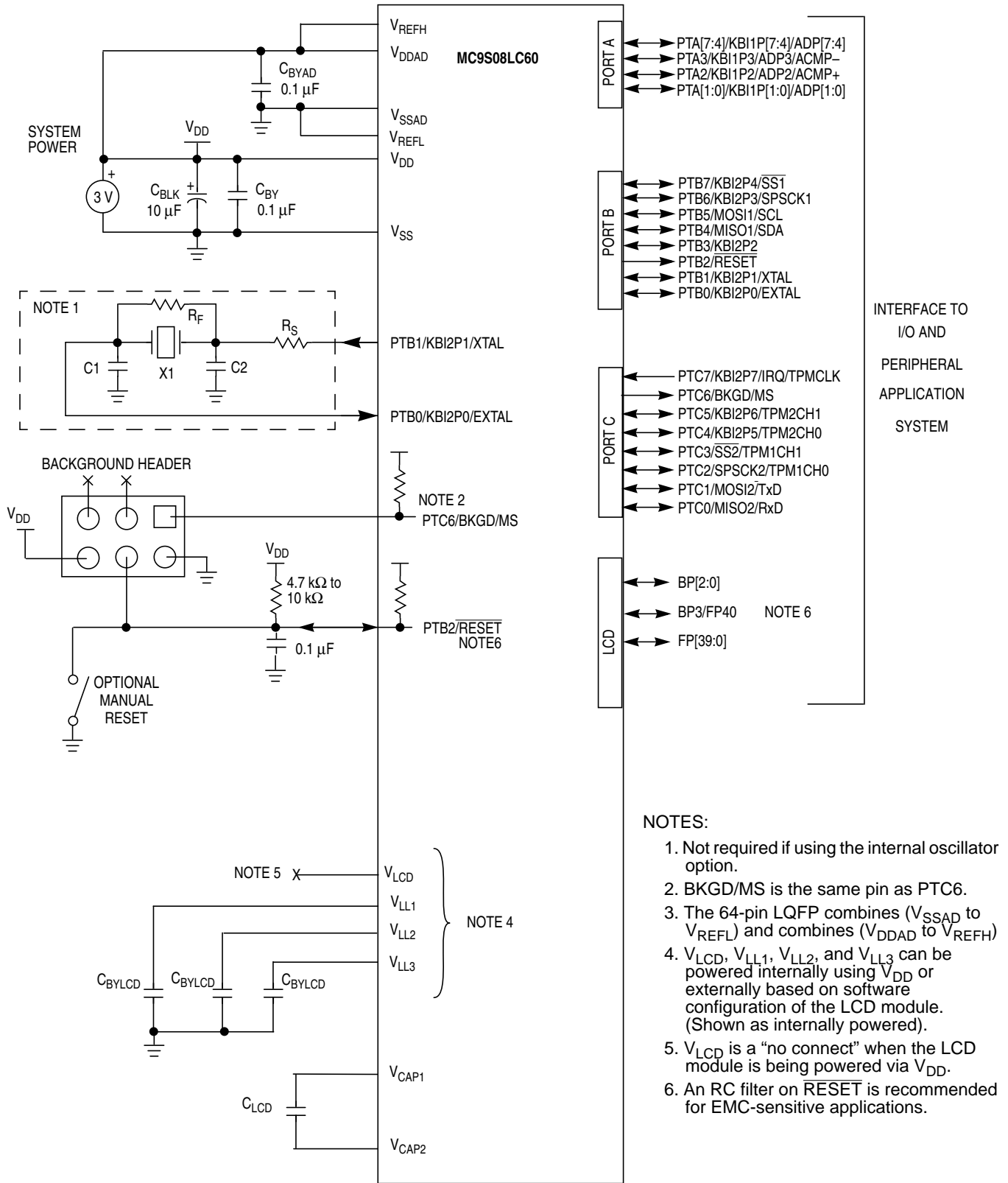


Figure 2-3. Basic System Connections

Table 3-2. Stop Mode Behavior

Peripheral	Mode		
	Stop1	Stop2	Stop3
CPU	Off	Off	Standby
RAM	Off	Standby	Standby
FLASH	Off	Off	Standby
Parallel Port Registers	Off	Off	Standby
ADC	Off	Off	Optionally On ¹
ACMP	Off	Off	Standby
ICG	Off	Off	Optionally On ²
IIC	Off	Off	Standby
LCD	Off	Off	Optionally On ³
SCI	Off	Off	Standby
SPI	Off	Off	Standby
TPM	Off	Off	Standby
Voltage Regulator	Off	Standby	Standby
I/O Pins	Hi-Z	States Held	States Held

¹ Requires the asynchronous ADC clock and LVD to be enabled, else in standby.

² OSCSPEN set in ICGC1, else in standby.

³ LCDSTP3 = 1 in the LCDCR1 register.

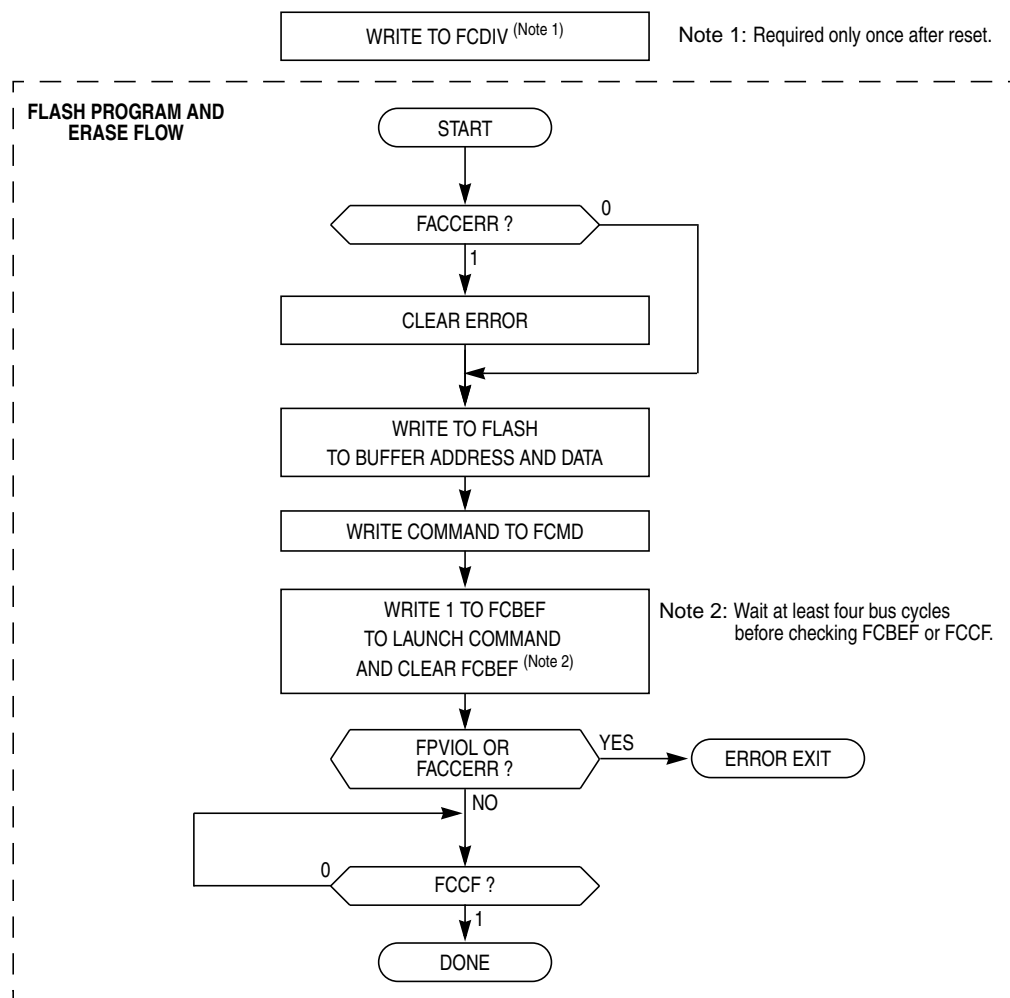


Figure 4-2. FLASH Program and Erase Flowchart

4.4.4 Burst Program Execution

The burst program command is used to program sequential bytes of data in less time than would be required using the standard program command. For the MC9S08LC60 Series, it is possible to burst across FLASH array boundaries as long as the addresses are consecutive. This is possible because the high voltage to the FLASH array does not need to be disabled between program operations. Ordinarily, when a program or erase command is issued, an internal charge pump associated with the FLASH memory must be enabled to supply high voltage to the array. Upon completion of the command, the charge pump is turned off. When a burst program command is issued, the charge pump is enabled and then remains enabled after completion of the burst program operation if these two conditions are met:

- The next burst program command has been queued before the current program operation has completed.
- The next sequential address selects a byte on the same physical row as the current byte being programmed. A row of FLASH memory consists of 64 bytes. A byte within a row is selected by addresses A5 through A0. A new row begins when addresses A5 through A0 are all zero.

Chapter 7

Keyboard Interrupt (S08KBIV2)

7.1 Introduction

This on-chip peripheral module is called a keyboard interrupt (KBI) module because originally it was designed to simplify the connection and use of row-column matrices of keyboard switches. However, these inputs are also useful as extra external interrupt inputs and as an external means of waking up the MCU from stop or wait low-power modes.

The KBI module allows up to eight pins to act as additional interrupt sources. These pins can be configured for either rising-edge sensing or falling-edge sensing. The sensing mode for all eight pins can also be modified to detect edges and levels instead of only edges.

MC9S08LC60 Series MCUs have two KBIs. When they are described individually, they are called KBI1 and/or KBI2. When referring to the module in general or both KBIs collectively, they are called KBIX.

Figure 7-1 Shows the MC9S08LC60 Series block guide with the KBIs highlighted.

Synchronous logic is used to detect edges. Prior to detecting an edge, enabled keyboard inputs must be at the deasserted logic level. A falling edge is detected when an enabled keyboard input signal is seen as a logic 1 (the deasserted level) during one bus cycle and then a logic 0 (the asserted level) during the next cycle. A rising edge is detected when the input signal is seen as a logic 0 during one bus cycle and then a logic 1 during the next cycle.

7.4.1 Edge Only Sensitivity

A valid edge on an enabled KBI pin will set KBF in KBIxSC. If KBIE in KBIxSC is set, an interrupt request will be presented to the CPU. Clearing of KBF is accomplished by writing a 1 to KBACK in KBIxSC.

7.4.2 Edge and Level Sensitivity

A valid edge or level on an enabled KBI pin will set KBF in KBIxSC. If KBIE in KBIxSC is set, an interrupt request will be presented to the CPU. Clearing of KBF is accomplished by writing a 1 to KBACK in KBIxSC provided all enabled keyboard inputs are at their deasserted levels. KBF will remain set if any enabled KBI pin is asserted while attempting to clear by writing a 1 to KBACK.

7.4.3 KBI Pullup/Pulldown Resistors

The KBI pins can be configured to use an internal pullup/pulldown resistor using the associated I/O port pullup enable register. If an internal resistor is enabled, the KBIxES register is used to select whether the resistor is a pullup (KBEDGn = 0) or a pulldown (KBEDGn = 1).

7.4.4 KBI Initialization

When a keyboard interrupt pin is first enabled it is possible to get a false keyboard interrupt flag. To prevent a false interrupt request during keyboard initialization, the user should do the following:

1. Mask keyboard interrupts by clearing KBIE in KBIxSC.
2. Enable the KBI polarity by setting the appropriate KBEDGn bits in KBIxES.
3. If using internal pullup/pulldown device, configure the associated pullup enable bits in PTxPE.
4. Enable the KBI pins by setting the appropriate KBIPEn bits in KBIxPE.
5. Write to KBACK in KBIxSC to clear any false interrupts.
6. Set KBIE in KBIxSC to enable interrupts.

8.3.5 Extended Addressing Mode (EXT)

In extended addressing mode, the full 16-bit address of the operand is located in the next two bytes of program memory after the opcode (high byte first).

8.3.6 Indexed Addressing Mode

Indexed addressing mode has seven variations including five that use the 16-bit H:X index register pair and two that use the stack pointer as the base reference.

8.3.6.1 Indexed, No Offset (IX)

This variation of indexed addressing uses the 16-bit value in the H:X index register pair as the address of the operand needed to complete the instruction.

8.3.6.2 Indexed, No Offset with Post Increment (IX+)

This variation of indexed addressing uses the 16-bit value in the H:X index register pair as the address of the operand needed to complete the instruction. The index register pair is then incremented ($H:X = H:X + 0x0001$) after the operand has been fetched. This addressing mode is only used for MOV and CBEQ instructions.

8.3.6.3 Indexed, 8-Bit Offset (IX1)

This variation of indexed addressing uses the 16-bit value in the H:X index register pair plus an unsigned 8-bit offset included in the instruction as the address of the operand needed to complete the instruction.

8.3.6.4 Indexed, 8-Bit Offset with Post Increment (IX1+)

This variation of indexed addressing uses the 16-bit value in the H:X index register pair plus an unsigned 8-bit offset included in the instruction as the address of the operand needed to complete the instruction. The index register pair is then incremented ($H:X = H:X + 0x0001$) after the operand has been fetched. This addressing mode is used only for the CBEQ instruction.

8.3.6.5 Indexed, 16-Bit Offset (IX2)

This variation of indexed addressing uses the 16-bit value in the H:X index register pair plus a 16-bit offset included in the instruction as the address of the operand needed to complete the instruction.

8.3.6.6 SP-Relative, 8-Bit Offset (SP1)

This variation of indexed addressing uses the 16-bit value in the stack pointer (SP) plus an unsigned 8-bit offset included in the instruction as the address of the operand needed to complete the instruction.

Table 8-3. Opcode Map (Sheet 2 of 2)

Bit-Manipulation	Branch	Read-Modify-Write			Control			Register/Memory																
					9E60	6							9ED0	5	9EE0	4								
					NEG	SP1							SUB	SP2	SUB	SP1								
					9E61	6							9ED1	5	9EE1	4								
					CBEQ	SP1							CMP	SP2	CMP	SP1								
													9ED2	5	9EE2	4								
													SBC	SP2	SBC	SP1								
					9E63	6							9ED3	5	9EE3	4	9EF3	6						
					COM	SP1							CPX	SP2	CPX	SP1	CPHX	SP1						
					9E64	6							9ED4	5	9EE4	4								
					LSR	SP1							AND	SP2	AND	SP1								
													9ED5	5	9EE5	4								
													BIT	SP2	BIT	SP1								
					9E66	6							9ED6	5	9EE6	4								
					ROR	SP1							LDA	SP2	LDA	SP1								
					9E67	6							9ED7	5	9EE7	4								
					ASR	SP1							STA	SP2	STA	SP1								
					9E68	6							9ED8	5	9EE8	4								
					LSL	SP1							EOR	SP2	EOR	SP1								
					9E69	6							9ED9	5	9EE9	4								
					ROL	SP1							ADC	SP2	ADC	SP1								
					9E6A	6							9EDA	5	9EEA	4								
					DEC	SP1							ORA	SP2	ORA	SP1								
					9E6B	8							9EDB	5	9EEB	4								
					DBNZ	SP1							ADD	SP2	ADD	SP1								
					9E6C	6																		
					INC	SP1																		
					9E6D	5																		
					TST	SP1																		
													9EAE	5	9EBE	6	9ECE	5	9EDE	5	9EEE	4	9EFE	5
													LDHX	IX	LDHX	IX2	LDHX	IX1	LDX	SP2	LDX	SP1	LDX	SP1
					9E6F	6																		
					CLR	SP1																		
													9EDF	5	9EEF	4	9EFF	5						
													STX	SP2	STX	SP1	STHX	SP1						

INH Inherent REL Relative SP1 Stack Pointer, 8-Bit Offset
 IMM Immediate IX Indexed, No Offset SP2 Stack Pointer, 16-Bit Offset
 DIR Direct IX1 Indexed, 8-Bit Offset IX+ Indexed, No Offset with
 EXT Extended IX2 Indexed, 16-Bit Offset Post Increment
 DD DIR to DIR IMD IMM to DIR IX1+ Indexed, 1-Byte Offset with
 IX+D IX+ to DIR DIX+ DIR to IX+ Post Increment

Note: All Sheet 2 Opcodes are Preceded by the Page 2 Prebyte (9E)

Prebyte (9E) and Opcode in Hexadecimal 9E60 6 HCS08 Cycles
 Number of Bytes 3 SP1 Instruction Mnemonic
 Addressing Mode

9.6.1.1 LCD Module Waveforms

DUTY = 1/3

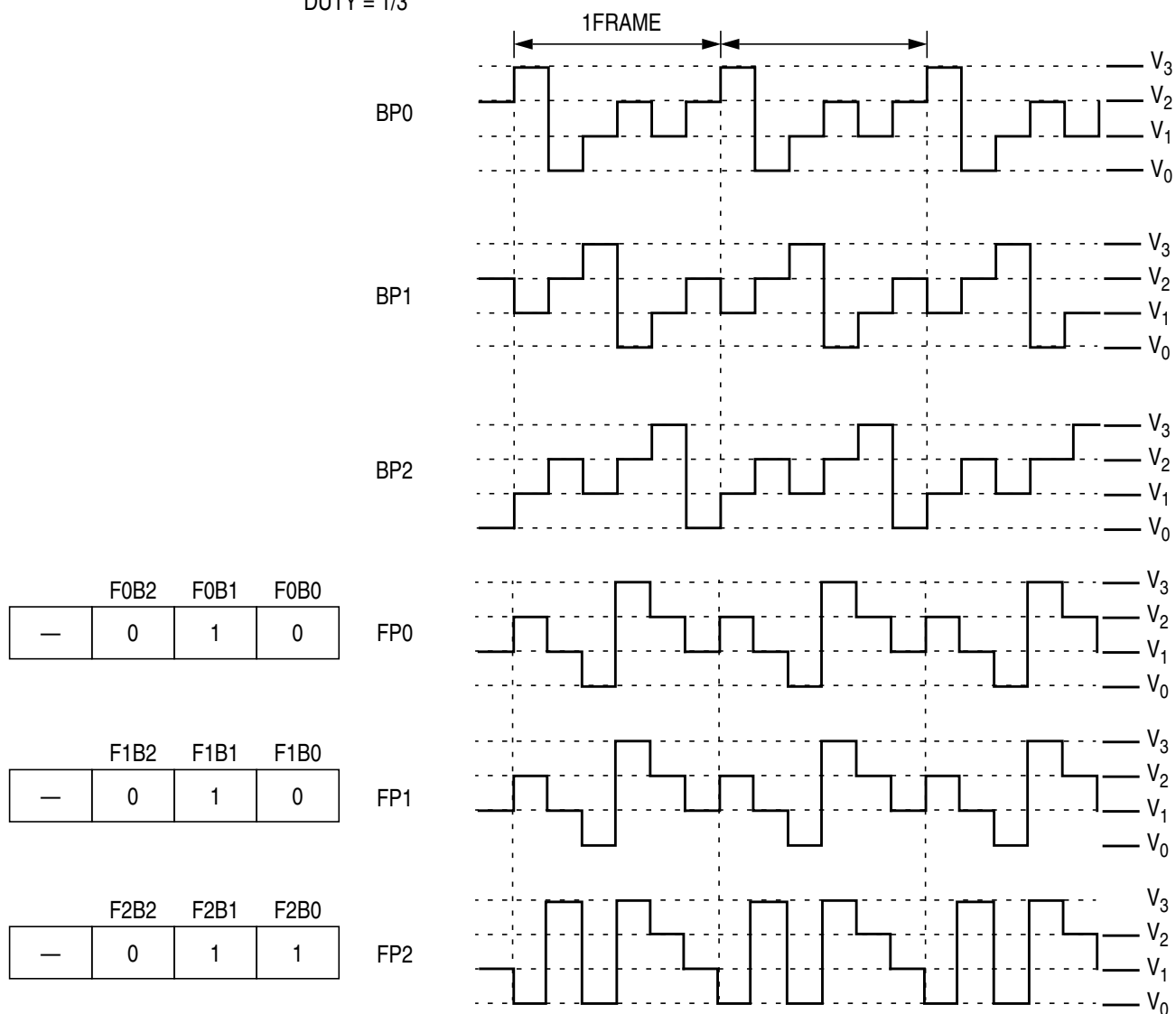


Figure 9-20. LCD Waveforms (LPWAVE = 0)

Chapter 10 Internal Clock Generator (S08ICGV4)

10.1 Introduction

The ICG module is used to generate the system clocks for the MC9S08LC60/36/20 MCU. Figure 10-1 shows the clock distribution for the MC9S08LC60/36/20 MCU. Electrical parametric data for the ICG may be found in Appendix.

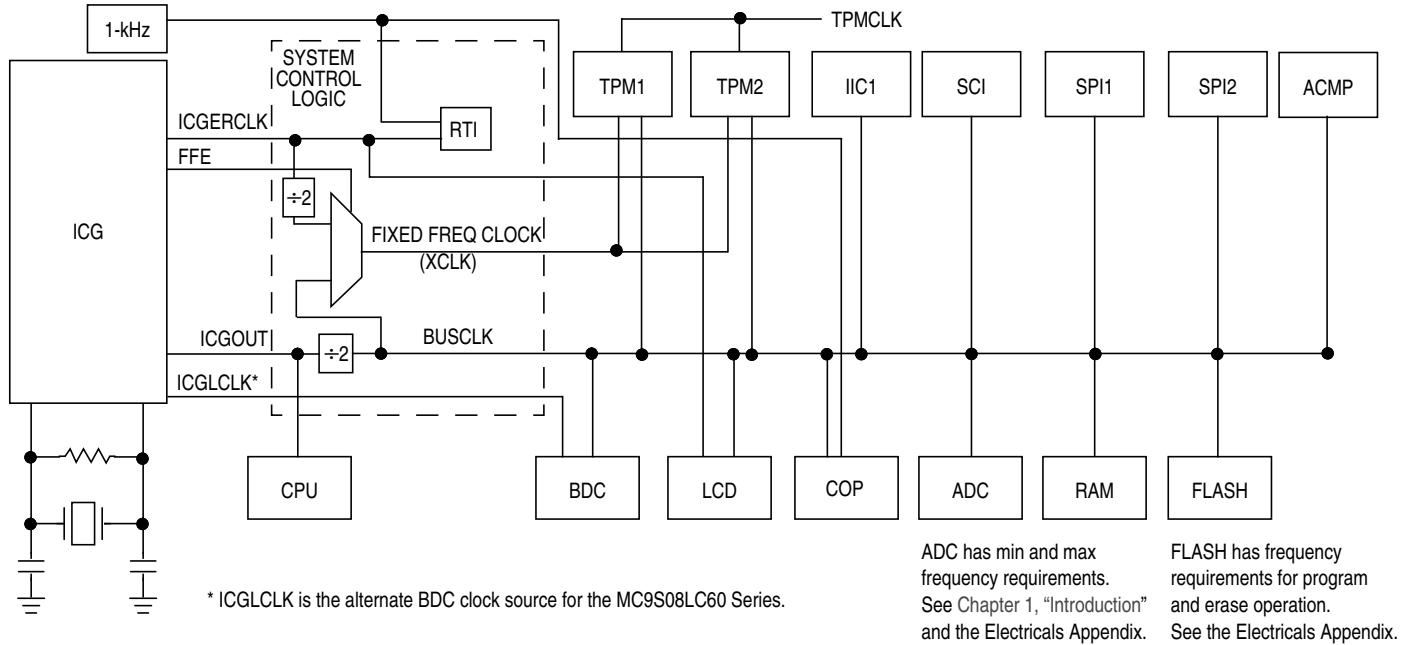


Figure 10-1. System Clock Distribution Diagram

NOTE

Freescale Semiconductor programs a factory trim value for ICGTRM into the FLASH location \$FFBE (NVICGTRM). Leaving this address for the ICGTRM value also allows debugger and programmer vendors to perform a manual trim operation and store the resultant ICGTRM value into NVICGTRM for users to access at a later time. The value in NVICGTRM is not automatically loaded and therefore must be copied into ICGTRM by user code.

Figure 10-2 shows the MC9S08LC60 Series block diagram with the ICG highlighted.

10.5.1 Off Mode (Off)

Normally when the CPU enters stop mode, the ICG will cease all clock activity and is in the off state. However there are two cases to consider when clock activity continues while the CPU is in stop mode,

10.5.1.1 BDM Active

When the BDM is enabled, the ICG continues activity as originally programmed. This allows access to memory and control registers via the BDC controller.

10.5.1.2 OSCSTEN Bit Set

When the oscillator is enabled in stop mode ($OSCSTEN = 1$), the individual clock generators are enabled but the clock feed to the rest of the MCU is turned off. This option is provided to avoid long oscillator startup times if necessary, or to run the RTI from the oscillator during stop3.

10.5.1.3 Stop/Off Mode Recovery

Upon the CPU exiting stop mode due to an interrupt, the previously set control bits are valid and the system clock feed resumes. If FEE is selected, the ICG will source the internal reference until the external clock is stable. If FBE is selected, the ICG will wait for the external clock to stabilize before enabling ICGOUT.

Upon the CPU exiting stop mode due to a reset, the previously set ICG control bits are ignored and the default reset values applied. Therefore the ICG will exit stop in SCM mode configured for an approximately 8 MHz DCO output (4 MHz bus clock) with trim value maintained. If using a crystal, 4096 clocks are detected prior to engaging ICGERCLK. This is incorporated in crystal start-up time.

10.5.2 Self-Clocked Mode (SCM)

Self-clocked mode (SCM) is the default mode of operation and is entered when any of the following conditions occur:

- After any reset.
- Exiting from off mode when $CLKS$ does not equal 10. If $CLKS = X1$, the ICG enters this state temporarily until the DCO is stable ($DCOS = 1$).
- $CLKS$ bits are written from X1 to 00.
- $CLKS = 1X$ and ICGERCLK is not detected (both $ERCS = 0$ and $LOCS = 1$).

In this state, the FLL loop is open. The DCO is on, and the output clock signal ICGOUT frequency is given by $f_{ICGDCLK} / R$. The ICGDCLK frequency can be varied from 8 MHz to 40 MHz by writing a new value into the filter registers (ICGFLTH and ICGFLTL). This is the only mode in which the filter registers can be written.

If this mode is entered due to a reset, $f_{ICGDCLK}$ will default to f_{Self_reset} which is nominally 8 MHz. If this mode is entered from FLL engaged internal, $f_{ICGDCLK}$ will maintain the previous frequency. If this mode is entered from FLL engaged external (either by programming $CLKS$ or due to a loss of external reference clock), $f_{ICGDCLK}$ will maintain the previous frequency, but ICGOUT will double if the FLL was unlocked. If this mode is entered from off mode, $f_{ICGDCLK}$ will be equal to the frequency of ICGDCLK before

10.6.4 Example #3: No External Crystal Connection, 5.4 MHz Bus Frequency

In this example, the FLL will be used (in FEI mode) to multiply the internal 243 kHz (approximate) reference clock up to 10.8 MHz to achieve 5.4 MHz bus frequency. This system will also use the trim function to fine tune the frequency based on an external reference signal.

After the MCU is released from reset, the ICG is in self-clocked mode (SCM) and supplies approximately 8 MHz on ICGOUT which corresponds to a 4 MHz bus frequency (f_{BUS}).

The clock scheme will be FLL engaged, internal (FEI). So

$$f_{ICGOUT} = (f_{IRG} / 7) * P * N / R ; P = 64, f_{IRG} = 243 \text{ kHz} \quad \text{Eqn. 10-5}$$

Solving for N / R gives:

$$N / R = 10.8 \text{ MHz} / (243/7 \text{ kHz} * 64) = 4.86 ; \text{ We can choose } N = 10 \text{ and } R = 2. \quad \text{Eqn. 10-6}$$

A trim procedure will be required to hone the frequency to exactly 5.4 MHz. An example of the trim procedure is shown in example #4.

The values needed in each register to set up the desired operation are:

ICGC1 = \$28 (%00101000)

Bit 7	HGO	0	Configures oscillator for low power
Bit 6	RANGE	0	Configures oscillator for low-frequency range; FLL prescale factor is 64
Bit 5	REFS	1	Oscillator using crystal or resonator requested (bit is really a don't care)
Bits 4:3	CLKS	01	FLL engaged, internal reference clock mode
Bit 2	OSCSTEN	0	Disables the oscillator
Bit 1	LOCD	0	Loss-of-clock enabled
Bit 0		0	Unimplemented or reserved, always reads zero

ICGC2 = \$31 (%00110001)

Bit 7	LOLRE	0	Generates an interrupt request on loss of lock
Bit 6:4	MFD	011	Sets the MFD multiplication factor to 10
Bit 3	LOCRE	0	Generates an interrupt request on loss of clock
Bit 2:0	RFD	001	Sets the RFD division factor to ÷2

ICGS1 = \$xx

This is read only except for clearing interrupt flag

ICGS2 = \$xx

This is read only; good idea to read this before performing time critical operations

ICGFLTLU/L = \$xx

Not used in this example

When center-aligned PWM operation is specified, the counter counts upward from 0x0000 through its terminal count and then counts downward to 0x0000 where it returns to up-counting. Both 0x0000 and the terminal count value (value in TPMxMODH:TPMxMODL) are normal length counts (one timer clock period long).

An interrupt flag and enable are associated with the main 16-bit counter. The timer overflow flag (TOF) is a software-accessible indication that the timer counter has overflowed. The enable signal selects between software polling (TOIE = 0) where no hardware interrupt is generated, or interrupt-driven operation (TOIE = 1) where a static hardware interrupt is automatically generated whenever the TOF flag is 1.

The conditions that cause TOF to become set depend on the counting mode (up or up/down). In up-counting mode, the main 16-bit counter counts from 0x0000 through 0xFFFF and overflows to 0x0000 on the next counting clock. TOF becomes set at the transition from 0xFFFF to 0x0000. When a modulus limit is set, TOF becomes set at the transition from the value set in the modulus register to 0x0000. When the main 16-bit counter is operating in up-/down-counting mode, the TOF flag gets set as the counter changes direction at the transition from the value set in the modulus register and the next lower count value. This corresponds to the end of a PWM period. (The 0x0000 count value corresponds to the center of a period.)

Because the HCS08 MCU is an 8-bit architecture, a coherency mechanism is built into the timer counter for read operations. Whenever either byte of the counter is read (TPMxCNTH or TPMxCNTL), both bytes are captured into a buffer so when the other byte is read, the value will represent the other byte of the count at the time the first byte was read. The counter continues to count normally, but no new value can be read from either byte until both bytes of the old count have been read.

The main timer counter can be reset manually at any time by writing any value to either byte of the timer count TPMxCNTH or TPMxCNTL. Resetting the counter in this manner also resets the coherency mechanism in case only one byte of the counter was read before resetting the count.

11.4.2 Channel Mode Selection

Provided CPWMS = 0 (center-aligned PWM operation is not specified), the MSnB and MSnA control bits in the channel n status and control registers determine the basic mode of operation for the corresponding channel. Choices include input capture, output compare, and buffered edge-aligned PWM.

11.4.2.1 Input Capture Mode

With the input capture function, the TPM can capture the time at which an external event occurs. When an active edge occurs on the pin of an input capture channel, the TPM latches the contents of the TPM counter into the channel value registers (TPMxCnVH:TPMxCnVL). Rising edges, falling edges, or any edge may be chosen as the active edge that triggers an input capture.

When either byte of the 16-bit capture register is read, both bytes are latched into a buffer to support coherent 16-bit accesses regardless of order. The coherency sequence can be manually reset by writing to the channel status/control register (TPMxCnSC).

An input capture event sets a flag bit (CHnF) that can optionally generate a CPU interrupt request.

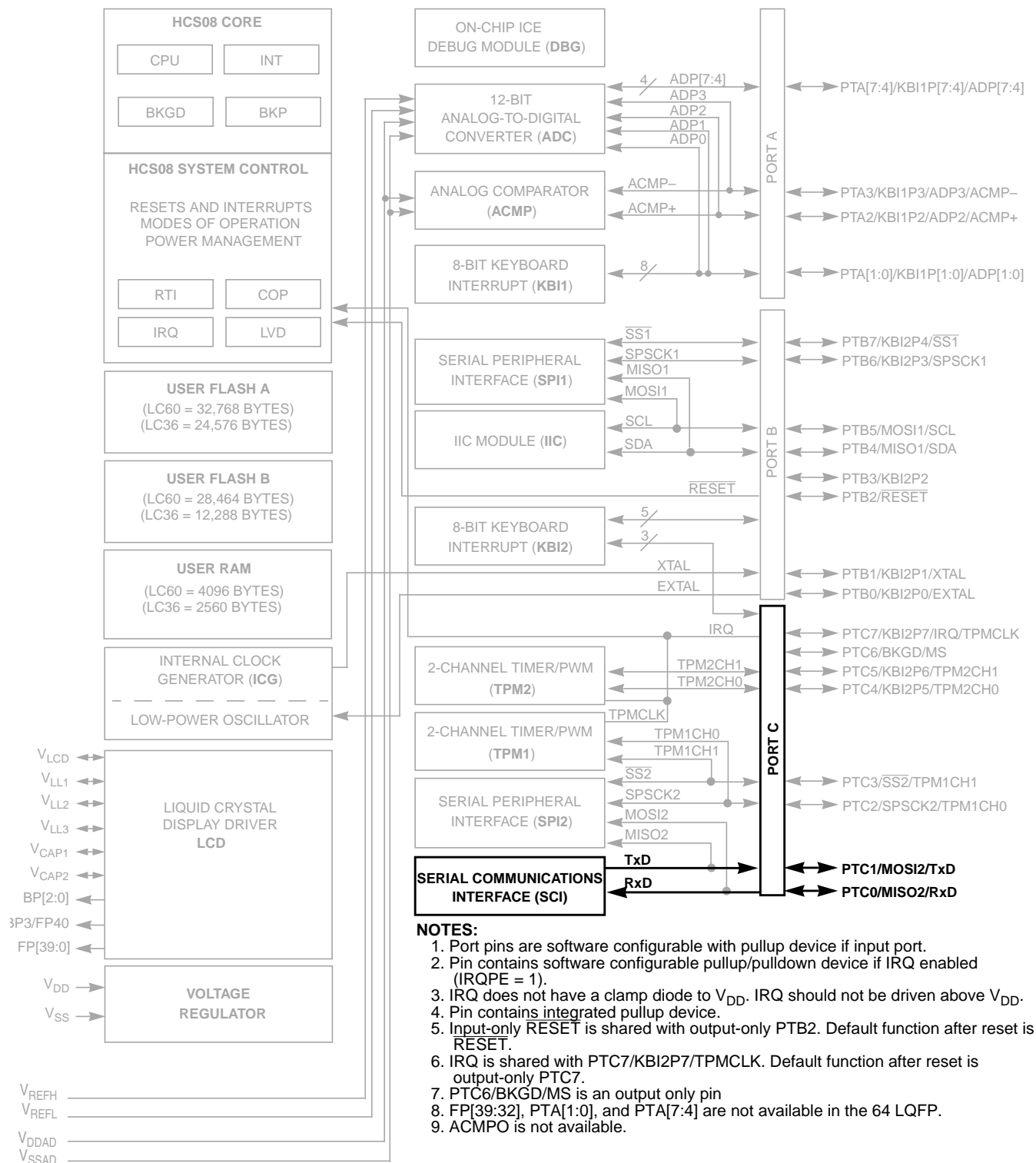


Figure 12-1. MC9S08LC60 Series Block Diagram Highlighting SCI Block and Pins

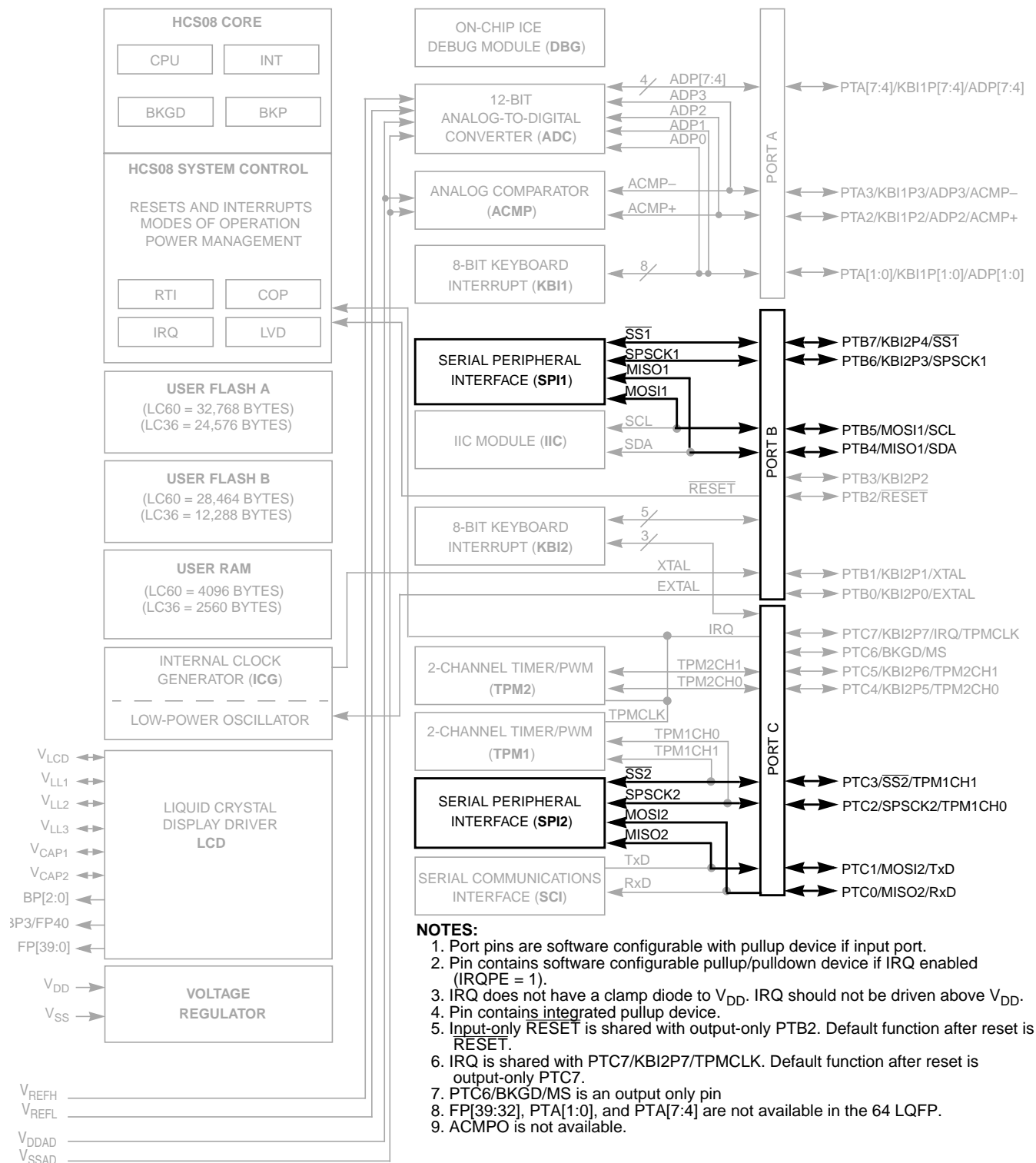


Figure 13-1. MC9S08LC60 Series Block Diagram Highlighting SPI Blocks and Pins

Chapter 16

Analog Comparator (S08ACMPV2)

16.1 Introduction

The analog comparator module (ACMP) provides a circuit for comparing two analog input voltages or for comparing one analog input voltage to an internal reference voltage. The comparator circuit is designed to operate across the full range of the supply voltage (rail to rail operation).

16.1.1 ACMP/TPM1 Configuration Information

The ACMP module can be configured to connect the output of the analog comparator to TPM1 input capture channel 0 by setting ACIC in SOPT2. With ACIC set, the TPM1CH0 pin is not available externally regardless of the configuration of the TPM1 module

Figure 16-1 shows the MC9S08LC60 Series block diagram with the ACMP highlighted.

16.1.2 AMCPO Availability

For the MC9S08LC60 Series, the AMCPO pin is not available, so the ACOPE bit in the ACMPSC register is reserved and does not have any effect.

Table A-7. 3 Volt 12-bit ADC Operating Conditions

Characteristic	Conditions	Symbol	Min	Typ ⁽¹⁾	Max	Unit	Comment
Analog Source Resistance	12 bit mode $f_{ADCK} > 4\text{MHz}$ $f_{ADCK} < 4\text{MHz}$	R_{AS}	—	—	2	k Ω	External to MCU
	10 bit mode $f_{ADCK} > 4\text{MHz}$ $f_{ADCK} < 4\text{MHz}$		—	—	5		
	8 bit mode (all valid f_{ADCK})		—	—	10		
ADC Conversion Clock Freq.	High Speed (ADLPC=0)	f_{ADCK}	0.4	—	8.0	MHz	
	Low Power (ADLPC=1)		0.4	—	4.0		

¹ Typical values assume $V_{DDAD} = 3.0\text{V}$, $\text{Temp} = 25^\circ\text{C}$, $f_{ADCK} = 1.0\text{MHz}$ unless otherwise stated. Typical values are for reference only and are not tested in production.

² DC potential difference.

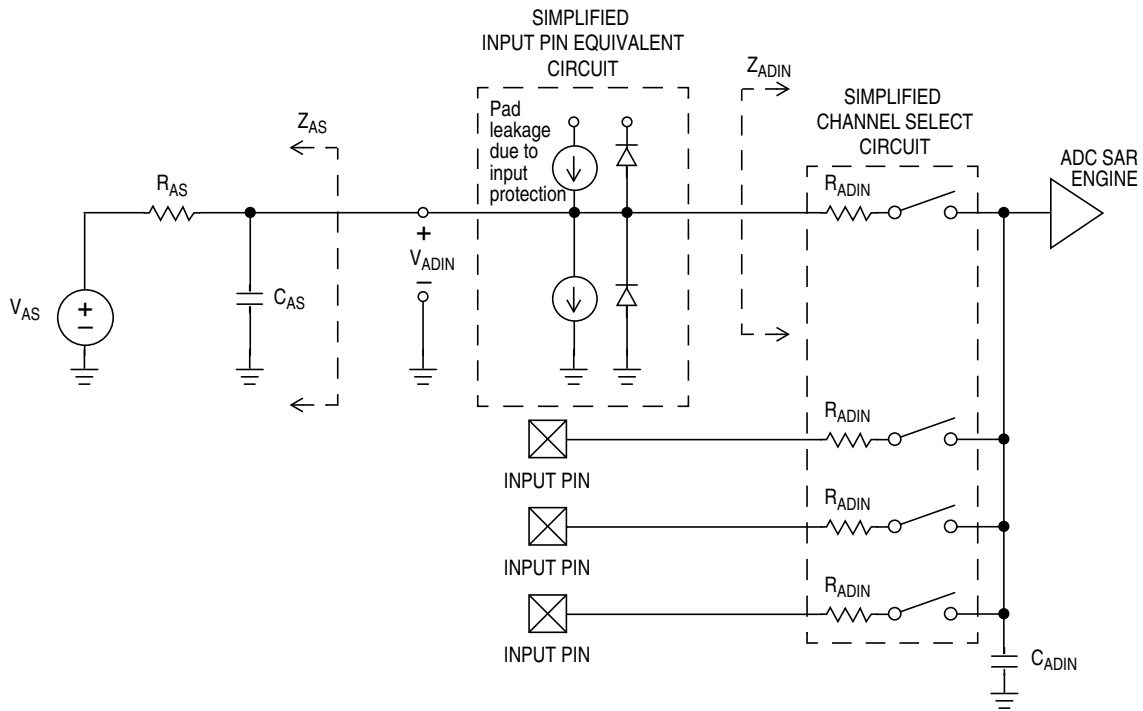
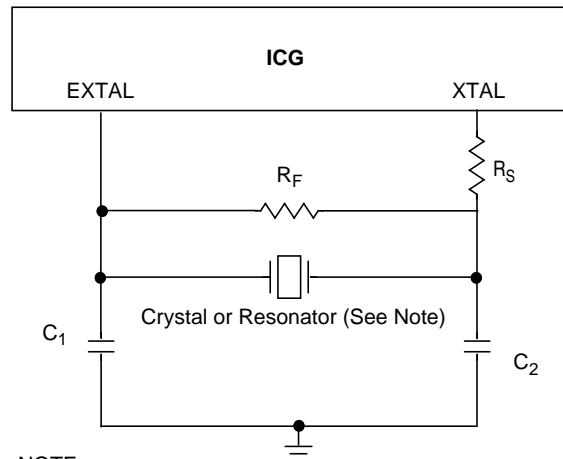


Figure A-10. ADC Input Impedance Equivalency Diagram

Table A-8. 3 Volt 12-bit ADC Characteristics ($V_{REFH} = V_{DDAD}$, $V_{REFL} = V_{SSAD}$)

Characteristic	Conditions	C	Symbol	Min	Typ ⁽¹⁾	Max	Unit	Comment
Supply Current ADLPC=1 ADLSMP=1 ADCO=1		T	I_{DDAD}	—	120	—	μA	

A.9 Internal Clock Generation Module Characteristics



NOTE:
Use fundamental mode crystal or ceramic resonator only.

Table A-11. ICG DC Electrical Specifications (Temperature Range = 0 to 70°C Ambient)

Characteristic	Symbol	Min	Typ ⁽¹⁾	Max	Unit
Load capacitors	C_1 C_2	See Note ⁽²⁾			
Feedback resistor Low range (32k to 100 kHz) High range (1M – 16 MHz)	R_F		10 1		MΩ MΩ
Series resistor Low range (32 kHz to 100 kHz) High range (1 MHz to 16 MHz)	R_S	0 0	0 0	10 0	kΩ

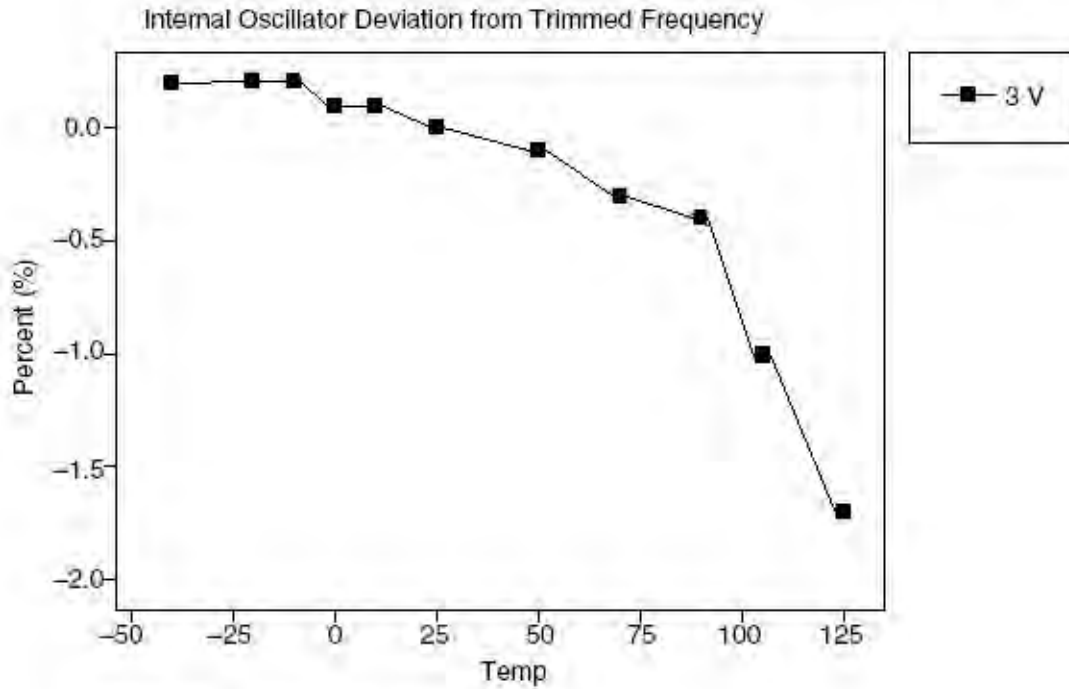
¹ Data in Typical column was characterized at 3.0 V, 25°C or is typical recommended value.

² See crystal or resonator manufacturer's recommendation.

A.9.1 ICG Frequency Specifications

Table A-12. ICG Frequency Specifications
($V_{DDA} = V_{DDA}(\min)$ to $V_{DDA}(\max)$, Temperature Range = 0 to 70°C Ambient)

Characteristic	Symbol	Min	Typical	Max	Unit
Oscillator crystal or resonator (REFS = 1) (Fundamental mode crystal or ceramic resonator)					
Low range	f_{lo}	32	—	100	kHz
High range, FLL bypassed external (CLKS = 10)	f_{hi_byp}	1	—	16	MHz
High range, FLL engaged external (CLKS = 11)	f_{hi_eng}	2	—	10	MHz
Input clock frequency (CLKS = 11, REFS = 0)					
Low range	f_{lo}	32	—	100	kHz
High range	f_{hi_eng}	2	—	10	MHz
Input clock frequency (CLKS = 10, REFS = 0)	f_{Extal}	0	—	40	MHz
Internal reference frequency (untrimmed)	$f_{ICGIRCLK}$	182.25	243	303.75	kHz



Device trimmed at 25°C at 3.0 V.

Figure A-11. Internal Oscillator Deviation from Trimmed Frequency

A.10 AC Characteristics

This section describes ac timing characteristics for each peripheral system. For detailed information about how clocks for the bus are generated, see Chapter 7, “Internal Clock Generator (ICG) Module.”

A.10.1 Control Timing

Table A-13. Control Timing

Parameter	Symbol	Min	Typical	Max	Unit
Bus frequency ($t_{cyc} = 1/f_{Bus}$)	f_{Bus}	dc	—	20	MHz
Real-time interrupt internal oscillator period	t_{RTI}	700		1300	μ s
External reset pulse width ⁽¹⁾	t_{extrst}	1.5 x f_{Self_reset}		—	ns
Reset low drive ⁽²⁾	t_{rstdrv}	34 x f_{Self_reset}		—	ns
Active background debug mode latch setup time	t_{MSSU}	25		—	ns
Active background debug mode latch hold time	t_{MSH}	25		—	ns
IRQ pulse width ⁽³⁾	t_{ILIH}	1.5 x t_{cyc}		—	ns
Port rise and fall time (load = 50 pF) ⁽⁴⁾	t_{Rise}, t_{Fall}	—	3		ns
Slew rate control disabled		—	30		
Slew rate control enabled					

- ¹ This is the shortest pulse that is guaranteed to be recognized as a reset pin request. Shorter pulses are not guaranteed to override reset requests from internal sources.
- ² When any reset is initiated, internal circuitry drives the reset pin low for about 34 cycles of f_{Self_reset} and then samples the level on the reset pin about 38 cycles later to distinguish external reset requests from internal requests.
- ³ This is the minimum pulse width that is guaranteed to pass through the pin synchronization circuitry. Shorter pulses may or may not be recognized. In stop mode, the synchronizer is bypassed so shorter pulses can be recognized in that case.
- ⁴ Timing is shown with respect to 20% V_{DD} and 80% V_{DD} levels. Temperature range -40°C to 85°C .

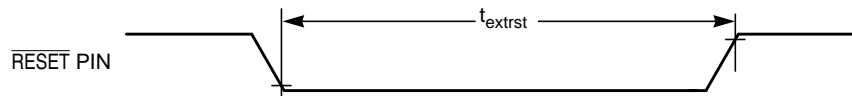


Figure A-12. Reset Timing

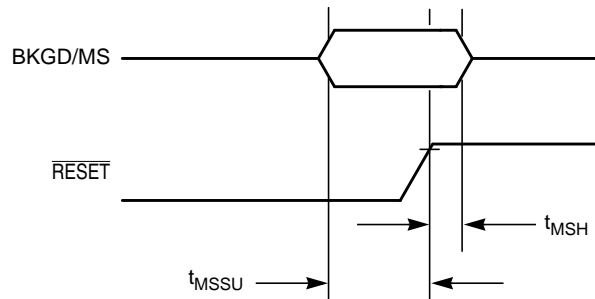


Figure A-13. Active Background Debug Mode Latch Timing