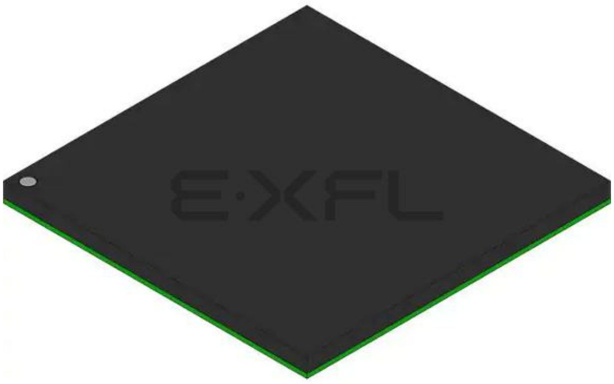


Welcome to E-XFL.COM

Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Active
Core Processor	PowerPC 603e
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	200MHz
Co-Processors/DSP	-
RAM Controllers	SDRAM
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	-
SATA	-
USB	-
Voltage - I/O	3.3V
Operating Temperature	-40°C ~ 105°C (TA)
Security Features	-
Package / Case	357-BBGA
Supplier Device Package	357-PBGA (25x25)
Purchase URL	https://www.e-xfl.com/pro/item?MUrl=&PartUrl=mpc8241tvr200d

Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

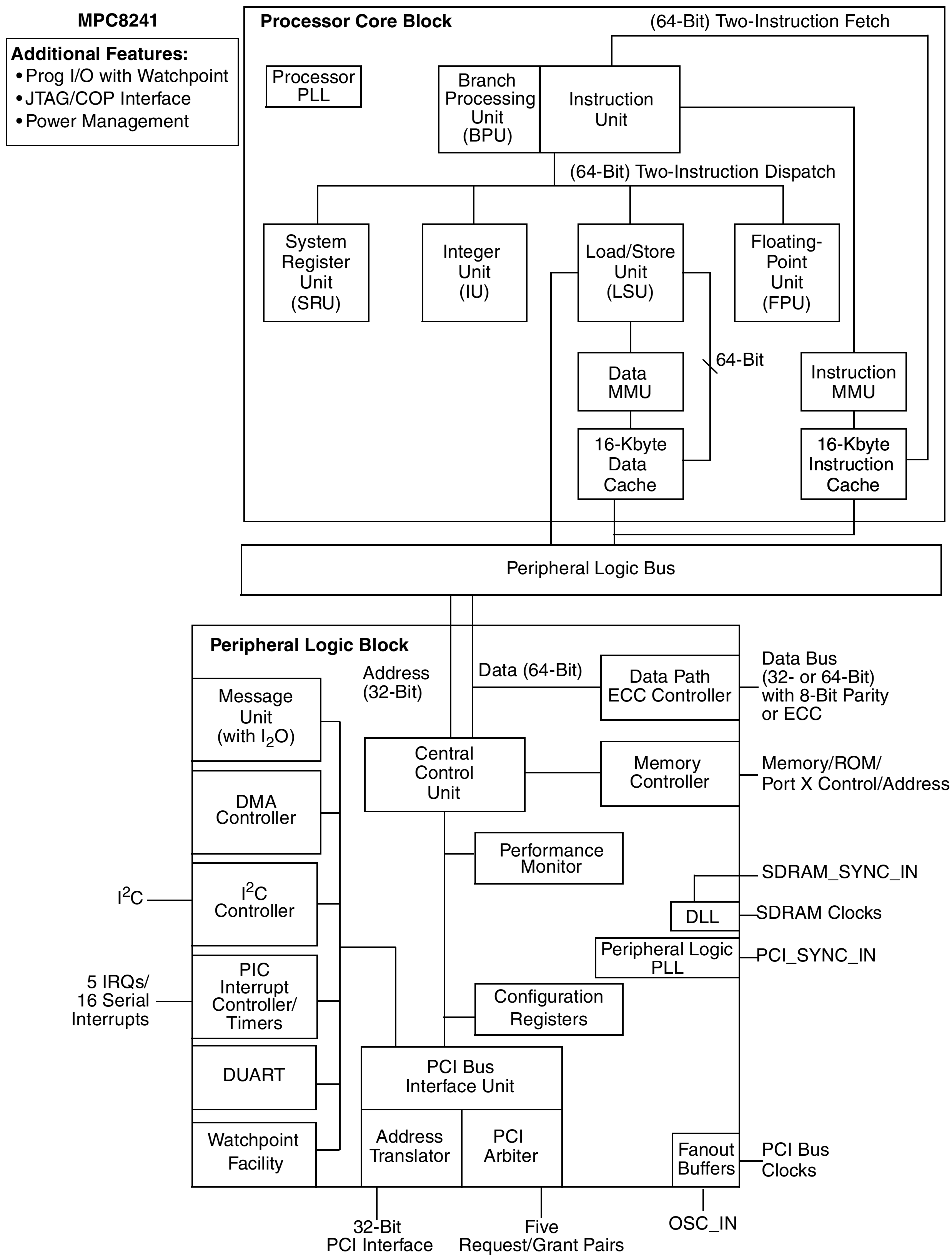


Figure 1. MPC8241 Block Diagram

4 Electrical and Thermal Characteristics

This section provides the AC and DC electrical specifications and thermal characteristics for the MPC8241.

4.1 DC Electrical Characteristics

This section covers ratings, conditions, and other characteristics.

4.1.1 Absolute Maximum Ratings

This section describes the MPC8241 DC electrical characteristics. Table 1 provides the absolute maximum ratings.

Table 1. Absolute Maximum Ratings

Characteristic [1]	Symbol	Range	Unit
Supply voltage—CPU core and peripheral logic	V_{DD}	–0.3 to 2.1	V
Supply voltage—memory bus drivers, PCI and standard I/O buffers	$GV_{DD}\_OV_{DD}$	–0.3 to 3.6	V
Supply voltage—PLLs	$AV_{DD}/AV_{DD}2$	–0.3 to 2.1	V
Supply voltage—PCI reference	LV_{DD}	–0.3 to 5.4	V
Input voltage [2]	V_{in}	–0.3 to 3.6	V
Operational die-junction temperature range	T_j	0 to 105	•C
Storage temperature range	T_{stg}	–55 to 150	•C

Notes:

1. Table 2 provides functional and tested operating conditions. Absolute maximum ratings are stress ratings only, and functional operation at the maximums is not guaranteed. Stresses beyond those listed may affect device reliability or cause permanent damage to the device.
2. PCI inputs with LV_{DD} = 5 V ± 5% V DC may be correspondingly stressed at voltages exceeding LV_{DD} + 0.5 V DC.

4.1.2 Recommended Operating Conditions

Table 2 provides the recommended operating conditions for the MPC8241.

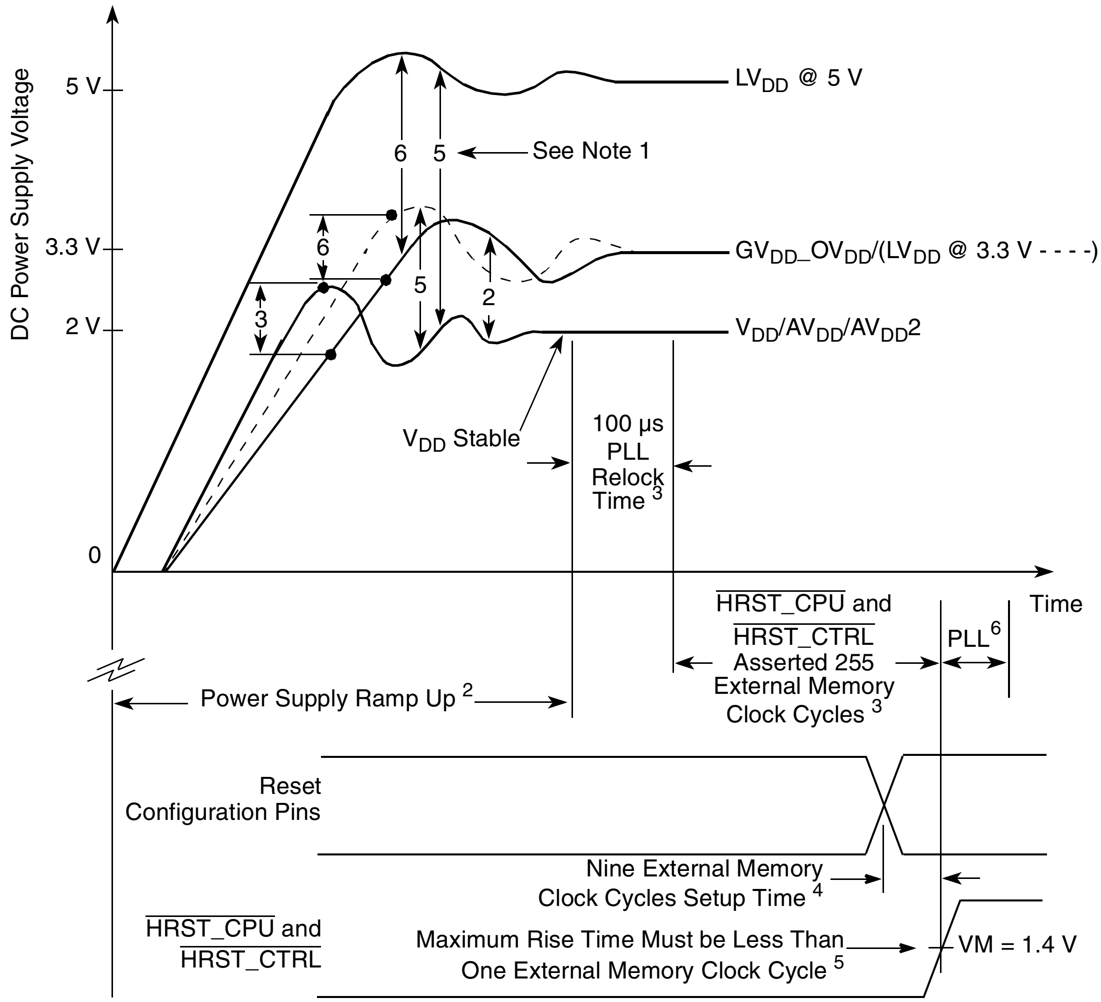
Table 2. Recommended Operating Conditions [1]

Characteristic		Symbol	Recommended Value	Unit	Notes
Supply voltage		V_{DD}	1.8 ± 100 mV	V	2
I/O buffer supply for PCI and standard; supply voltages for memory bus drivers		$GV_{DD}\_OV_{DD}$	3.3 ± 0.3	V	2
CPU PLL supply voltage		AV_{DD}	1.8 ± 100 mV		2
PLL supply voltage—peripheral logic		$AV_{DD}2$	1.8 ± 100 mV	V	2
PCI reference		LV_{DD}	5.0 ± 5%	V	4, 5, 6
			3.3 ± 0.3	V	5, 6, 7
Input voltage	PCI inputs	V_{in}	0 to 3.6 or 5.75	V	4, 7
	All other inputs		0 to 3.6	V	8
Die-junction temperature		T_j	0 to 105	•C	

Notes:

1. Freescale has tested these operating conditions and recommends them. Proper device operation outside of these conditions is not guaranteed.
2. **Caution:** $GV_{DD}\_OV_{DD}$ must not exceed $V_{DD}/AV_{DD}/AV_{DD}2$ by more than 1.8 V at any time including during power-on reset. Note that $GV_{DD}\_OV_{DD}$ pins are all shorted together: This limit may be exceeded for a maximum of 20 ms during power-on reset and power-down sequences. Connections should not be made to individual PWRRING pins.
3. **Caution:** $V_{DD}/AV_{DD}/AV_{DD}2$ must not exceed $GV_{DD}\_OV_{DD}$ by more than 0.6 V at any time, including during power-on reset. This limit may be exceeded for a maximum of 20 ms during power-on reset and power-down sequences.
4. PCI pins are designed to withstand LV_{DD} + 0.5 V DC when LV_{DD} is connected to a 5.0 V DC power supply.
5. **Caution:** LV_{DD} must not exceed $V_{DD}/AV_{DD}/AV_{DD}2$ by more than 5.4 V at any time, including during power-on reset. This limit may be exceeded for a maximum of 20 ms during power-on reset and power-down sequences.
6. **Caution:** LV_{DD} must not exceed $GV_{DD}\_OV_{DD}$ by more than 3.0 V at any time, including during power-on reset. This limit may be exceeded for a maximum of 20 ms during power-on reset and power-down sequences.
7. PCI pins are designed to withstand LV_{DD} + 0.5 V DC when LV_{DD} is connected to a 3.3 V DC power supply.
8. **Caution:** Input voltage (V_{in}) must not be greater than the supply voltage ($V_{DD}/AV_{DD}/AV_{DD}2$) by more than 2.5 V at all times including during power-on reset. Input voltage (V_{in}) must not be greater than $GV_{DD}\_OV_{DD}$ by more than 0.6 V at all times including during power-on reset.

Figure 2 shows supply voltage sequencing and separation cautions.



Notes:

1. Numbers associated with waveform separations correspond to caution numbers listed in Table 2.
2. See the Cautions section of Table 2 for details on this topic.
3. Refer to Table 8 for details on PLL relock and reset signal assertion timing requirements.
4. Refer to Table 10 for details on reset configuration pin setup timing requirements.
5. HRST_CPU/HRST_CTRL must transition from a logic 0 to a logic 1 in less than one SDRAM_SYNC_IN clock cycle for the device to be in the nonreset state.
6. PLL_CFG signals must be driven on reset and must be held for at least 25 clock cycles after the negation of HRST_CTRL and HRST_CPU negate in order to be latched.

Figure 2. Supply Voltage Sequencing and Separation Cautions

Figure 3 shows the undershoot and overshoot voltage of the memory interface.

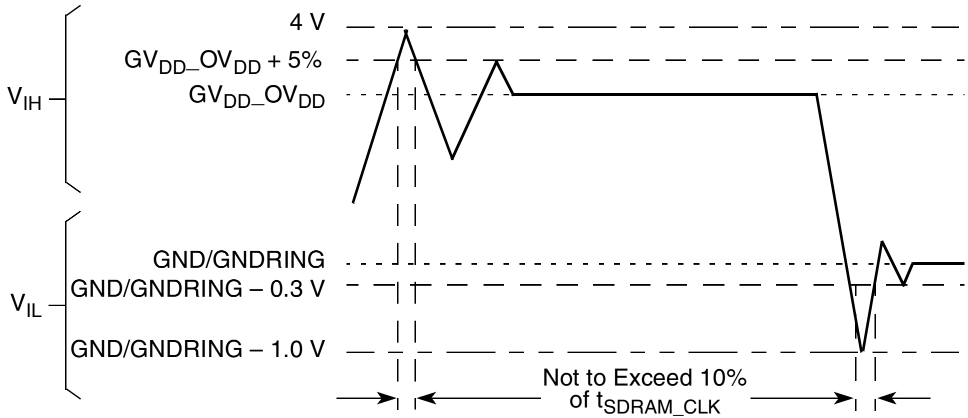


Figure 3. Overshoot/Undershoot Voltage

Figure 4 and Figure 5 show the undershoot and overshoot voltage of the PCI interface for the 3.3- and 5-V signals, respectively.

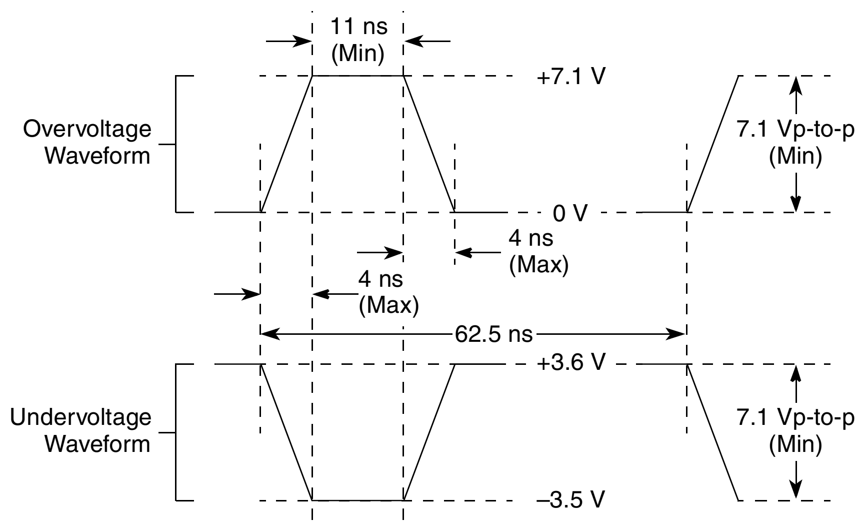


Figure 4. Maximum AC Waveforms for 3.3-V Signaling

Table 10. Input AC Timing Specifications (continued)

Num	Characteristic	Min	Max	Unit	Notes
10b0	Tap 0, register offset <0x77>, bits 5:4 = 0b00	2.6	—	ns	2, 3, 6
10b1	Tap 1, register offset <0x77>, bits 5:4 = 0b01	1.9	—		
10b2	Tap 2, register offset <0x77>, bits 5:4 = 0b10 (default)	1.2	—		
10b3	Tap 3, register offset <0x77>, bits 5:4 = 0b11	0.5	—		
10c	PIC miscellaneous debug input signals valid to *sys_logic_clk* (input setup)	3.0	—	ns	2, 3
10d	I^2C input signals valid to *sys_logic_clk* (input setup)	3.0	—	ns	2, 3
10e	Mode select inputs valid to $\overline{\text{HRST\_CPU}}/\overline{\text{HRST\_CTRL}}$ (input setup)	$9 \times t_{CLK}$	—	ns	2, 3–5
11	T_{os}—SDRAM_SYNC_IN to *sys_logic_clk* offset time	0.4	1.0	ns	7
11a	*sys_logic_clk* to memory signal inputs invalid (input hold)				
11a0	Tap 0, register offset <0x77>, bits 5:4 = 0b00	0	—	ns	2, 3, 6
11a1	Tap 1, register offset <0x77>, bits 5:4 = 0b01	0.7	—		
11a2	Tap 2, register offset <0x77>, bits 5:4 = 0b10 (default)	1.4	—		
11a3	Tap 3, register offset <0x77>, bits 5:4 = 0b11	2.1	—		
11b	$\overline{\text{HRST\_CPU}}/\overline{\text{HRST\_CTRL}}$ to mode select inputs invalid (input hold)	0	—	ns	2, 3, 5
11c	PCI_SYNC_IN to inputs invalid (input hold)	1.0	—	ns	1, 2, 3

Notes:

1. All PCI signals are measured from $GV_{DD}\_OV_{DD}/2$ of the rising edge of PCI_SYNC_IN to $0.4 \times GV_{DD}\_OV_{DD}$ of the signal in question for 3.3-V PCI signaling levels. See Figure 12.
2. All memory and related interface input signal specifications are measured from the TTL level (0.8 or 2.0 V) of the signal in question to the VM = 1.4 V of the rising edge of the memory bus clock. *sys_logic_clk*. *sys_logic_clk* is the same as PCI_SYNC_IN in 1:1 mode, but is twice the frequency in 2:1 mode (processor/memory bus clock rising edges occur on every rising and falling edge of PCI_SYNC_IN). See Figure 11.
3. Input timings are measured at the pin.
4. t_{CLK} is the time of one SDRAM_SYNC_IN clock cycle.
5. All mode select input signals specifications are measured from the TTL level (0.8 or 2.0 V) of the signal in question to the VM = 1.4 V of the rising edge of the $\overline{\text{HRST\_CPU}}/\overline{\text{HRST\_CTRL}}$ signal. See Figure 13.
6. The memory interface input setup and hold times are programmable to four possible combinations by programming bits 5:4 of register offset <0x77> to select the desired input setup and hold times.
7. T_{os} represents a timing adjustment for SDRAM_SYNC_IN with respect to *sys_logic_clk*. Due to the internal delay present on the SDRAM_SYNC_IN signal with respect to the *sys_logic_clk* inputs to the DLL, the resulting SDRAM clocks become offset by the delay amount. The feedback trace length of SDRAM_SYNC_OUT to SDRAM_SYNC_IN must be shortened to accommodate this range relative to the SDRAM clock output trace lengths to maintain phase-alignment of the memory clocks with respect to *sys_logic_clk*. It is recommended that the length of SDRAM_SYNC_OUT to SDRAM_SYNC_IN be shortened by 0.7 ns because that is the midpoint of the range of T_{os} and allows the impact from the range of T_{os} to be reduced. Additional analyses of trace lengths and SDRAM loading must be performed to optimize timing. For details on trace measurements and the T_{os} problem, refer to the Freescale application note AN2164, *MPC8245/MPC8241 Memory Clock Design Guidelines*.

Figure 13 shows the input timing diagram for mode select signals.

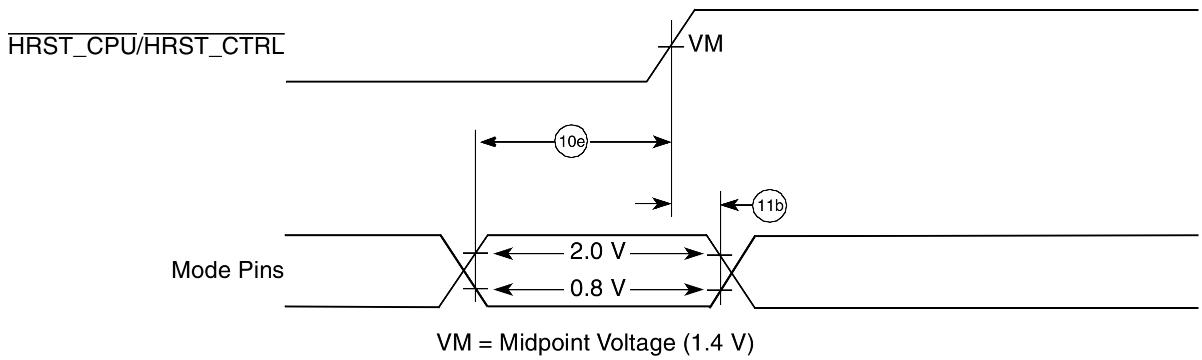


Figure 13. Input Timing Diagram for Mode Select Signals

4.5.3 Output AC Timing Specification

Table 11 provides the processor bus AC timing specifications for the MPC8241 at recommended operating conditions (see Table 2) with LV_{DD} = 3.3 V ± 0.3 V (see Figure 11). All output timings assume a purely resistive 50-Ω load (see Figure 14). Output timings are measured at the pin; time-of-flight delays must be added for trace lengths, vias, and connectors in the system. These specifications are for the default driver strengths that Table 4 indicates.

Table 11. Output AC Timing Specifications

Num	Characteristic	Min	Max	Unit	Notes
12a	PCI_SYNC_IN to output valid, see Figure 15				
12a0	Tap 0, PCI_HOLD_DEL = 00, [$\overline{MCP}$,CKE] = 11, 66 MHz PCI (default)	—	6.0	ns	1, 3
12a1	Tap 1, PCI_HOLD_DEL = 01, [$\overline{MCP}$,CKE] = 10	—	6.5		
12a2	Tap 2, PCI_HOLD_DEL = 10, [$\overline{MCP}$,CKE] = 01, 33 MHz PCI	—	7.0		
12a3	Tap 3, PCI_HOLD_DEL = 11, [$\overline{MCP}$,CKE] = 00	—	7.5		
12b	*sys_logic_clk* to output valid (memory address, control, and data signals)	—	4.5	ns	2
12c	*sys_logic_clk* to output valid (for all others)	—	7.0	ns	2
12d	*sys_logic_clk* to output valid (for I^2C)	—	5.0	ns	2
12e	*sys_logic_clk* to output valid (ROM/Flash/Port X)	—	6.0	ns	2
13a	Output hold (PCI), see Figure 15				
13a0	Tap 0, PCI_HOLD_DEL = 00, [$\overline{MCP}$,CKE] = 11, 66 MHz PCI (default)	2.0	—	ns	1, 3, 4
13a1	Tap 1, PCI_HOLD_DEL = 01, [$\overline{MCP}$,CKE] = 10	2.5	—		
13a2	Tap 2, PCI_HOLD_DEL = 10, [$\overline{MCP}$,CKE] = 01, 33 MHz PCI	3.0	—		
13a3	Tap 3, PCI_HOLD_DEL = 11, [$\overline{MCP}$,CKE] = 00	3.5	—		
13b	Output hold (all others)	1.0	—	ns	2
14a	PCI_SYNC_IN to output high impedance (for PCI)	—	14.0	ns	1, 3

Table 11. Output AC Timing Specifications (continued)

Num	Characteristic	Min	Max	Unit	Notes
14b	*sys_logic_clk* to output high impedance (for all others)	—	4.0	ns	2

Notes:

1. All PCI signals are measured from $GV_{DD}\_OV_{DD}/2$ of the rising edge of PCI_SYNC_IN to $0.285 \times GV_{DD}\_OV_{DD}$ or $0.615 \times GV_{DD}\_OV_{DD}$ of the signal in question for 3.3 V PCI signaling levels. See Figure 12.
2. All memory and related interface output signal specifications are specified from the VM = 1.4 V of the rising edge of the memory bus clock, *sys_logic_clk* to the TTL level (0.8 or 2.0 V) of the signal in question. *sys_logic_clk* is the same as PCI_SYNC_IN in 1:1 mode, but is twice the frequency in 2:1 mode (processor/memory bus clock rising edges occur on every rising and falling edge of PCI_SYNC_IN). See Figure 11.
3. PCI bused signals are composed of the following signals: $\overline{\text{LOCK}}$, $\overline{\text{IRDY}}$, $\overline{\text{C/BE}}$[3:0], PAR, $\overline{\text{TRDY}}$, $\overline{\text{FRAME}}$, $\overline{\text{STOP}}$, $\overline{\text{DEVSEL}}$, $\overline{\text{PERR}}$, $\overline{\text{SERR}}$, AD[31:0], $\overline{\text{REQ}}$[4:0], $\overline{\text{GNT}}$[4:0], $\overline{\text{IDSEL}}$, and $\overline{\text{INTA}}$.
4. To meet minimum output hold specifications relative to PCI_SYNC_IN for both 33- and 66-MHz PCI systems, the MPC8241 has a programmable output hold delay for PCI signals (the PCI_SYNC_IN to output valid timing is also affected). The initial value of the output hold delay is determined by the values on the $\overline{\text{MCP}}$ and CKE reset configuration signals; the values on these two signals are inverted and subsequently stored as the initial settings of PCI_HOLD_DEL = PMCR2[5, 4] (power management configuration register 2 <0x72>), respectively. Because $\overline{\text{MCP}}$ and CKE have internal pull-up resistors, the default value of PCI_HOLD_DEL after reset is 0b00. Additional output hold delay values are available by programming the PCI_HOLD_DEL value of the PMCR2 configuration register. See Figure 15 for PCI_HOLD_DEL effect on output valid and hold time.

Figure 14 provides the AC test load for the MPC8241.

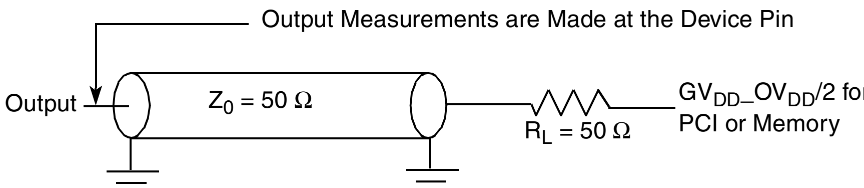


Figure 14. AC Test Load for the MPC8241

Table 13. I^2C AC Electrical Specifications (continued)

All values refer to V_{IH} (min) and V_{IL} (max) levels (see Table 12).

Parameter	Symbol [1]	Min	Max	Unit
Noise margin at the HIGH level for each connected device (including hysteresis)	V_{NH}	$0.2 \times OV_{DD}$	—	V

Note:

1. The symbols used for timing specifications herein follow the pattern of $t_{\text{(first two letters of functional block)(signal)(state) (reference)(state)}}$ for inputs and $t_{\text{(first two letters of functional block)(reference)(state)(signal)(state)}}$ for outputs. For example, t_{I2DVKH} symbolizes I^2C timing (I2) with respect to the time data input signals (D) reach the valid state (V) relative to the t_{I2C} clock reference (K) going to the high (H) state or setup time. Also, t_{I2SXKL} symbolizes I^2C timing (I2) for the time that the data with respect to the start condition (S) went invalid (X) relative to the t_{I2C} clock reference (K) going to the low (L) state or hold time. Also, t_{I2PVKH} symbolizes I^2C timing (I2) for the time that the data with respect to the stop condition (P) reaching the valid state (V) relative to the t_{I2C} clock reference (K) going to the high (H) state or setup time. For rise and fall times, the latter convention is used with the appropriate letter: R (rise) or F (fall).
2. As a transmitter, the MPC8245 provides a delay time of at least 300 ns for the SDA signal (referred to the Vihmin of the SCL signal) to bridge the undefined region of the falling edge of SCL to avoid the unintended generation of a Start or Stop condition. When the MPC8245 acts as the I^2C bus master while transmitting, it drives both SCL and SDA. As long as the load on SCL and SDA is balanced, the MPC8245 does not cause an unintended generation of a Start or Stop condition. Therefore, the 300 ns SDA output delay time is not a concern. If, under some rare condition, the 300 ns SDA output delay time is required for the MPC8245 as transmitter, the following setting is recommended for the FDR bit field of the I2CFDR register to ensure both the desired I^2C SCL clock frequency and SDA output delay time are achieved. It is assumed that the desired I^2C SCL clock frequency is 400 KHz and the digital filter sampling rate register (DFFSR bits in I2CFDR) is programmed with its default setting of 0x10 (decimal 16):

SDRAM Clock Frequency	100 MHz	133 MHz
FDR Bit Setting	0x00	0x2A
Actual FDR Divider Selected	384	896
Actual I^2C SCL Frequency Generated	260.4 KHz	148.4 KHz

   For details on I^2C frequency calculation, refer to the application note AN2919 “Determining the I^2C Frequency Divider Ratio for SCL”.
3. The maximum t_{I2DXKL} has only to be met if the device does not stretch the LOW period (t_{I2CL}) of the SCL signal.
4. Guaranteed by design

Figure 16 provides the AC test load for the I^2C.

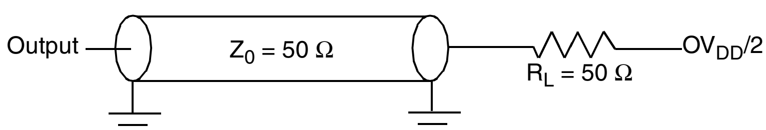


Figure 16. I^2C AC Test Load

Figure 17 shows the AC timing diagram for the I²C bus.

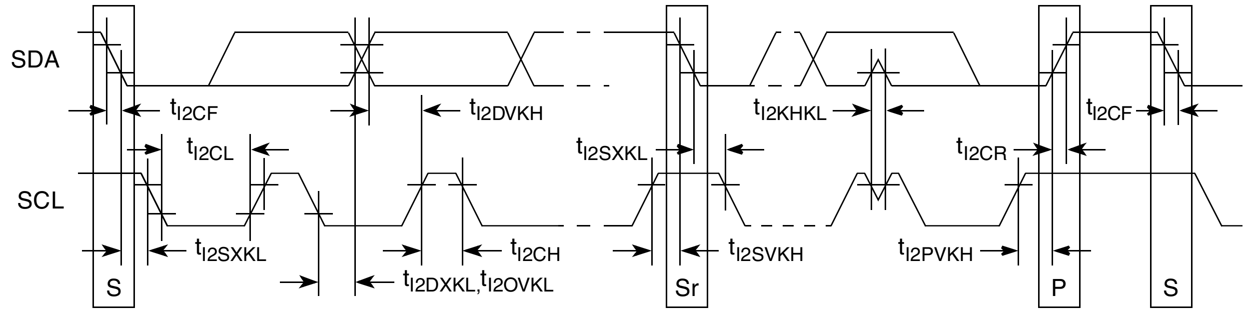


Figure 17. I²C Bus AC Timing Diagram

4.7 PIC Serial Interrupt Mode AC Timing Specifications

Table 14 provides the PIC serial interrupt mode AC timing specifications for the MPC8241 at recommended operating conditions (see Table 2) with $GV_{DD\_}OV_{DD}$ = 3.3 V ± 5% and LV_{DD} = 3.3 V ± 0.3 V.

Table 14. PIC Serial Interrupt Mode AC Timing Specifications

Num	Characteristic	Min	Max	Unit	Notes
1	S_CLK frequency	1/14 SDRAM_SYNC_IN	1/2 SDRAM_SYNC_IN	MHz	1
2	S_CLK duty cycle	40	60	%	—
3	S_CLK output valid time	—	6	ns	—
4	Output hold time	0	—	ns	—
5	$\overline{\text{S\_FRAME}}$, S_RST output valid time	—	1 *sys_logic_clk* period + 6	ns	2
6	S_INT input setup time to S_CLK	1 *sys_logic_clk* period + 2	—	ns	2
7	S_INT inputs invalid (hold time) to S_CLK	—	0	ns	2

Notes:

1. See the *MPC8245 Integrated Processor Reference Manual* for a description of the PIC interrupt control register (ICR) and S_CLK frequency programming.
2. S_RST, $\overline{\text{S\_FRAME}}$, and S_INT shown in Figure 18 and Figure 19, depict timing relationships to *sys_logic_clk* and S_CLK and do not describe functional relationships between S_RST, $\overline{\text{S\_FRAME}}$, and S_INT. The *MPC8245 Integrated Processor Reference Manual* describes the functional relationships between these signals.
3. The *sys_logic_clk* waveform is the clocking signal of the internal peripheral logic from the output of the peripheral logic PLL; *sys_logic_clk* is the same as SDRAM_SYNC_IN when the SDRAM_SYNC_OUT to SDRAM_SYNC_IN feedback loop is implemented and the DLL is locked. See the *MPC8245 Integrated Processor Reference Manual* for a complete clocking description.

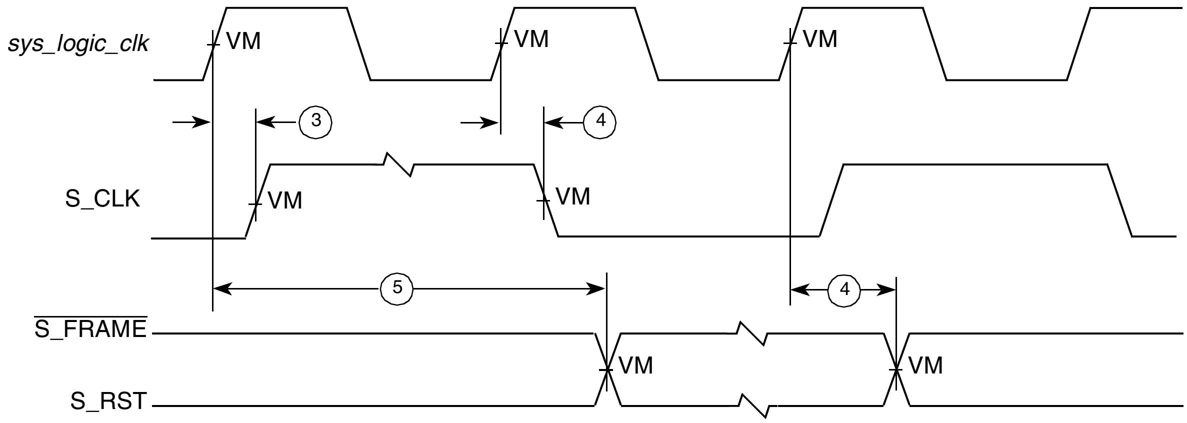


Figure 18. PIC Serial Interrupt Mode Output Timing Diagram

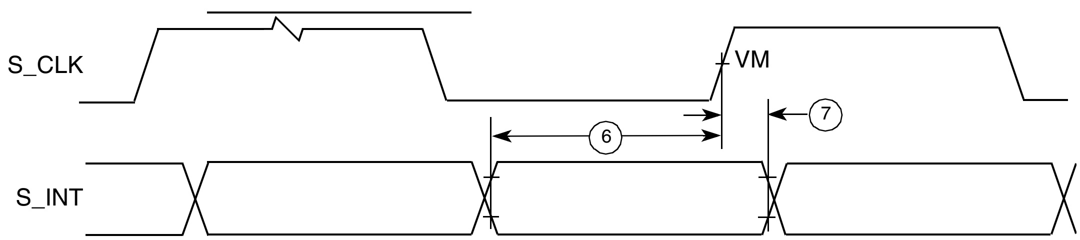


Figure 19. PIC Serial Interrupt Mode Input Timing Diagram

4.7.1 IEEE 1149.1 (JTAG) AC Timing Specifications

Table 15 provides the JTAG AC timing specifications for the MPC8241 while in the JTAG operating mode at recommended operating conditions (see Table 2) with LV_{DD} = 3.3 V ± 0.3 V. Timings are independent of the system clock (PCI_SYNC_IN).

Table 15. JTAG AC Timing Specification (Independent of PCI_SYNC_IN)

Num	Characteristic	Min	Max	Unit	Notes
	TCK frequency of operation	0	25	MHz	—
1	TCK cycle time	40	—	ns	—
2	TCK clock pulse width measured at 1.5 V	20	—	ns	—
3	TCK rise and fall times	0	3	ns	—
4	$\overline{\text{TRST}}$ setup time to TCK falling edge	10	—	ns	1
5	$\overline{\text{TRST}}$ assert time	10	—	ns	—
6	Input data setup time	5	—	ns	2
7	Input data hold time	15	—	ns	2
8	TCK to output data valid	0	30	ns	3
9	TCK to output high impedance	0	30	ns	3
10	TMS, TDI data setup time	5	—	ns	—

Table 16. MPC8241 Pinout Listing (continued)

Signal Name	Package Pin Number	Pin Type	Power Supply	Output Driver Type	Notes
TMS	T18	Input	GV_{DD}_OV_{DD}	—	6, 13
$\overline{\text{TRST}}$	R16	Input	GV_{DD}_OV_{DD}	—	6, 13
Power and Ground Signals					
GNDRING/GND	F07 F08 F09 F10 F11 F12 F13 G07 G08 G09 G10 G11 G12 G13 H07 H08 H09 H10 H11 H12 H13 J07 J08 J09 J10 J11 J12 J13 K07 K08 K09 K10 K11 K12 K13 L07 L08 L09 L10 L11 L12 L13 M07 M08 M09 M10 M11 M12 M13 N07 N08 N09 N10 N11 N12 N13 P08 P09 P10 P11 P12 P13 R15	Ground	—	—	17
LV_{DD}	R18 U18 T1 U4 T6 W11 T14	Reference voltage 3.3 V, 5.0 V	LV_{DD}	—	—
GV_{DD}_OV_{DD}/PWRRING	D09 D10 D11 E06 E07 E08 E09 E10 E11 E12 E13 E14 F06 F14 G06 G14 H06 H14 J06 J14 K06 K14 L06 L14 M06 M14 N06 N14 P06 P07 P14 R08 R09 R10 R11 R12	Power for memory drivers and PCI/Stnd 3.3 V	GV_{DD}_OV_{DD}	—	18
V_{DD}	F03 H3 L5 N4 P5 V5 U8 W12 W16 R13 P19 L19 H19 F19 F15 C15 A13 A8 B5 A2	Power for core 1.8 V	V_{DD}	—	—
No Connect	N5 W2 B1	—	—	—	—
AV_{DD}	M5	Power for PLL (CPU core logic) 1.8 V	AV_{DD}	—	—
$AV_{DD}2$	R14	Power for PLL (peripheral logic) 1.8 V	$AV_{DD}2$	—	—
Debug/Manufacturing Pins					
DA0/$\overline{\text{QACK}}$	A3	Output	GV_{DD}_OV_{DD}	DRV_STD_MEM	5, 11, 12
DA1/CKO	L1	Output	GV_{DD}_OV_{DD}	DRV_STD_MEM	5
DA2	R5	Output	GV_{DD}_OV_{DD}	DRV_PCI	19
DA3/PCI_CLK4	V17	Output	GV_{DD}_OV_{DD}	DRV_PCI_CLK	5
DA4/$\overline{\text{REQ4}}$	W13	I/O	GV_{DD}_OV_{DD}	—	5, 6
DA5/$\overline{\text{GNT4}}$	T11	Output	GV_{DD}_OV_{DD}	DRV_PCI	2, 4, 5

Table 17. PLL Configurations (166- and 200-MHz) (continued)

Ref [2]	PLL_CFG [0:4] [1]	166 MHz-Part [2]			200-MHz Part [2]			Multipliers	
		PCI Clock Input (PCI_ SYNC_IN) Range [3] (MHz)	Peripheral Logic/ Mem Bus Clock Range (MHz)	CPU Clock Range (MHz)	PCI Clock Input (PCI_ SYNC_IN) Range [3] (MHz)	Peripheral Logic/ Mem Bus Clock Range (MHz)	CPU Clock Range (MHz)	PCI-to-Mem (Mem VCO)	Mem-to-CPU (CPU VCO)
1E	11110[14]	Not usable			Not usable			Off	Off
1F	11111[14]	Not usable			Not usable			Off	Off

Notes:

1. PLL_CFG[0:4] settings not listed are reserved. Bits 7–4 of register offset <0xE2> contain the PLL_CFG[0:4] setting value. Note the impact of the relevant revisions for mode 7.
2. Range values are shown rounded down to the nearest whole number (decimal place accuracy removed) for clarity.
3. Limited by maximum PCI input frequency (66 MHz).
4. Limited by minimum CPU VCO frequency (300 MHz).
5. Limited by maximum CPU operating frequency.
6. In PLL bypass mode, the PCI_SYNC_IN input signal clocks the internal processor directly, the peripheral logic PLL is disabled, and the bus mode is set for 1:1 (PCI:Mem) mode operation. This mode is intended for hardware modeling. The AC timing specifications in this document do not apply in PLL bypass mode.
7. Limited by minimum CPU operating frequency (100 MHz).
8. Limited due to maximum memory VCO frequency (352 MHz).
9. In dual PLL bypass mode, the PCI_SYNC_IN input signal clocks the internal peripheral logic directly, the peripheral logic PLL is disabled, and the bus mode is set for 1:1 (PCI_SYNC_IN:Mem) mode operation. In this mode, the OSC_IN input signal clocks the internal processor directly in 1:1 (OSC_IN:CPU) mode operation, and the processor PLL is disabled. The PCI_SYNC_IN and OSC_IN input clocks must be externally synchronized. This mode is intended for hardware modeling. The AC timing specifications in this document do not apply in dual PLL bypass mode.
10. Limited by maximum CPU VCO frequency (704 MHz).
11. Limited by maximum system memory interface operating frequency (83 MHz @ 166 MHz CPU bus speed).
12. Limited by maximum system memory interface operating frequency (100 MHz @ 200 MHz CPU bus speed).
13. Limited by minimum memory VCO frequency (132 MHz).
14. In clock off mode, no clocking occurs inside the MPC8241, regardless of the PCI_SYNC_IN input.

Table 18. PLL Configurations (266-MHz Parts)

Ref [2]	PLL_ CFG[0:4] [10,11]	266-MHz Part [9]			Multipliers	
		PCI Clock Input (PCI_SYNC_IN) Range [1] (MHz)	Periph Logic/ Mem Bus Clock Range (MHz)	CPU Clock Range (MHz)	PCI-to-Mem (Mem VCO)	Mem-to-CPU (CPU VCO)
0	00000	25–35[5]	75–105	188–263	3 (2)	2.5 (2)
1	00001	25–29[5]	75–88	225–264	3 (2)	3 (2)
2	00010	50[15]–59[5]	50–59	225–266	1 (4)	4.5 (2)
3	00011[12]	50[14]–66[1]	50–66	100–133	1 (Bypass)	2 (4)
4	00100	25–44[4]	50–88	100–176	2 (4)	2 (4)

Table 18. PLL Configurations (266-MHz Parts) (continued)

Ref [2]	PLL_ CFG[0:4] [10,11]	266-MHz Part [9]			Multipliers	
		PCI Clock Input (PCI_SYNC_IN) Range [1] (MHz)	Periph Logic/ Mem Bus Clock Range (MHz)	CPU Clock Range (MHz)	PCI-to-Mem (Mem VCO)	Mem-to-CPU (CPU VCO)
1F	11111[8]	Not usable			Off	Off

Notes:

1. Limited by maximum PCI input frequency (66 MHz).
2. Note the impact of the relevant revisions for modes 7 and 1E.
3. Limited by minimum memory VCO frequency (132 MHz).
4. Limited due to maximum memory VCO frequency (352 MHz).
5. Limited by maximum CPU operating frequency.
6. Limited by minimum CPU VCO frequency (300 MHz).
7. Limited by maximum CPU VCO frequency (704 MHz).
8. In clock off mode, no clocking occurs inside the MPC8241, regardless of the PCI_SYNC_IN input.
9. Range values are shown rounded down to the nearest whole number (decimal place accuracy removed) for clarity.
10. PLL_CFG[0:4] settings that are not listed are reserved.
11. Bits 7–4 of register offset <0xE2> contain the PLL_CFG[0:4] setting value.
12. In PLL bypass mode, the PCI_SYNC_IN input signal clocks the internal processor directly, the peripheral logic PLL is disabled, and the bus mode is set for 1:1 (PCI:Mem) mode operation. This mode is intended for hardware modeling. The AC timing specifications in this document do not apply in PLL bypass mode.
13. In dual PLL bypass mode, the PCI_SYNC_IN input signal clocks the internal peripheral logic directly, the peripheral logic PLL is disabled, and the bus mode is set for 1:1 (PCI_SYNC_IN:Mem) mode operation. In this mode, the OSC_IN input signal clocks the internal processor directly in 1:1 (OSC_IN:CPU) mode operation and the processor PLL is disabled. The PCI_SYNC_IN and OSC_IN input clocks must be externally synchronized. This mode is intended for hardware modeling. The AC timing specifications in this document do not apply in dual PLL bypass mode.
14. Limited by minimum CPU operating frequency (100 MHz).
15. Limited by minimum memory bus frequency (50 MHz).

7 System Design Information

This section provides electrical and thermal design recommendations for successful application of the MPC8241.

7.1 PLL Power Supply Filtering

The AV_{DD} and $AV_{DD}2$ power signals on the MPC8241 provide power to the peripheral logic/memory bus PLL and the MPC603e processor PLL. To ensure stability of the internal clocks, the power supplied to the AV_{DD} and $AV_{DD}2$ input signals should be filtered of any noise in the 500 kHz to 10 MHz resonant frequency range of the PLLs. Two separate circuits similar to the one shown in Figure 26 using surface mount capacitors with minimum effective series inductance (ESL) is recommended for AV_{DD} and $AV_{DD}2$ power signal pins. In *High Speed Digital Design: A Handbook of Black Magic* (Prentice Hall, 1993), Dr. Howard Johnson recommends using multiple small capacitors of equal value instead of multiple values.

7.6 JTAG Configuration Signals

Boundary scan testing is enabled through the JTAG interface signals. The $\overline{\text{TRST}}$ signal is optional in the IEEE 1149.1 specification, but is provided on all processors that implement the PowerPC architecture. While the TAP controller can be forced to the reset state using only the TCK and TMS signals, more reliable power-on reset performance will be obtained if the $\overline{\text{TRST}}$ signal is asserted during power-on reset. Because the JTAG interface is also used for accessing the common on-chip processor (COP) function, simply tying $\overline{\text{TRST}}$ to $\overline{\text{HRESET}}$ is not practical.

The COP function of these processors allows a remote computer system (typically, a PC with dedicated hardware and debugging software) to access and control the internal operations of the processor. The COP interface connects primarily through the JTAG port, with additional status monitoring signals. The COP port must independently assert $\overline{\text{HRESET}}$ or $\overline{\text{TRST}}$ to control the processor. If the target system has independent reset sources, such as voltage monitors, watchdog timers, power supply failures, or push-button switches, the COP reset signals must be merged into these signals with logic.

The arrangement shown in Figure 27 allows the COP port to independently assert $\overline{\text{HRESET}}$ or $\overline{\text{TRST}}$, while ensuring that the target can drive $\overline{\text{HRESET}}$ as well. If the JTAG interface and COP header will not be used, $\overline{\text{TRST}}$ should be tied to $\overline{\text{HRESET}}$ through a 0-Ω isolation resistor so that it is asserted when the system reset signal ($\overline{\text{HRESET}}$) is asserted, ensuring that the JTAG scan chain is initialized during power-on. Although Freescale recommends that the COP header be designed into the system as shown in Figure 27, if this is not possible, the isolation resistor will allow future access to $\overline{\text{TRST}}$ in the case where a JTAG interface may need to be wired onto the system in debug situations.

The COP interface has a standard header for connection to the target system, based on the 0.025" square-post, 0.100" centered header assembly (often called a Berg header). Typically, pin 14 is removed as a connector key.

There is no standardized way to number the COP header shown in Figure 27. Consequently, different emulator vendors number the pins differently. Some pins are numbered top-to-bottom and left-to-right while others use left-to-right then top-to-bottom and still others number the pins counter clockwise from pin 1 (as with an IC). Regardless of the numbering, the signal placement recommended in Figure 27 is common to all known emulators.

7.7 Thermal Management

This section provides thermal management information for the plastic ball grid array (PBGA) package for air-cooled applications. Depending on the application environment and the operating frequency, a heat sink may be required to maintain junction temperature within specifications. Proper thermal control design primarily depends on the system-level design: heat sink, airflow, and thermal interface material. To reduce the die-junction temperature, heat sinks can be attached to the package by several methods: adhesive, spring clip to holes in the printed-circuit board or package, or mounting clip and screw assembly (see Figure 28).

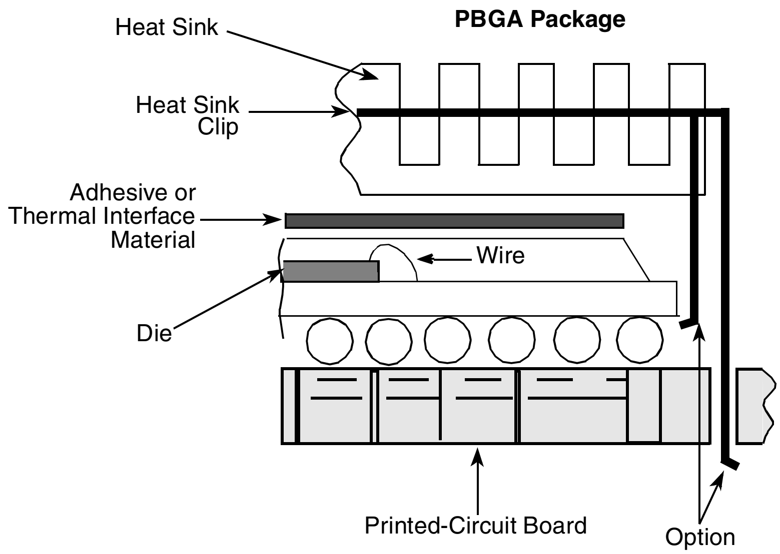


Figure 28. Package Exploded Cross-Sectional View with Several Heat Sink Options

Figure 29 depicts the die junction-to-ambient thermal resistance for four typical cases:

- A heat sink is not attached to the PBGA package and a high board-level thermal loading from adjacent components exists (label used—1s).
- A heat sink is not attached to the PBGA package and a low board-level thermal loading from adjacent components exists (label used—2s2p).
- A large heat sink (cross cut extrusion, 38 × 38 × 16.5 mm) is attached to the PBGA package and a high board-level thermal loading from adjacent components exists (label used—1s/sink).
- A large heat sink (cross cut extrusion, 38 × 38 × 16.5 mm) is attached to the PBGA package and a low board-level thermal loading from adjacent components exists (label used—2s2p/sink).

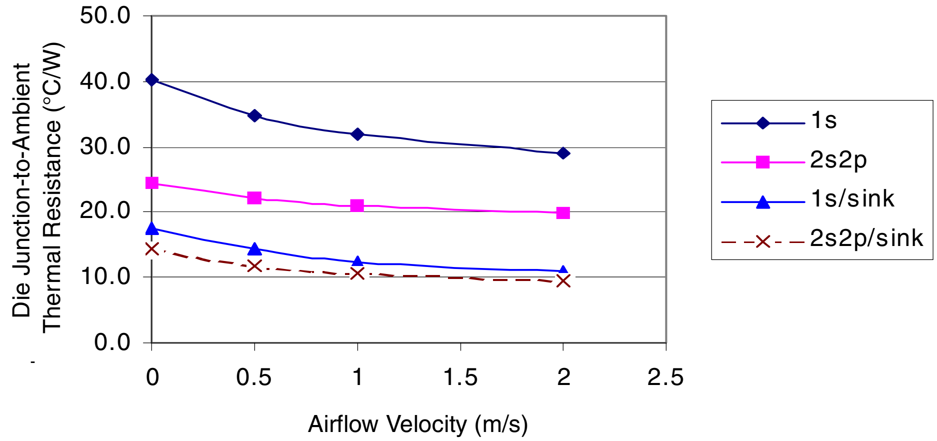


Figure 29. Die Junction-to-Ambient Resistance

The board designer can choose among several types of heat sinks to place on the MPC8241. Several commercially available heat sinks for the MPC8241 are provided by the following vendors:

Aavid Thermalloy 80 Commercial St. Concord, NH 03301 Internet: www.aavidthermalloy.com	603-224-9988
Alpha Novatech 473 Sapena Ct. #15 Santa Clara, CA 95054 Internet: www.alphanovatech.com	408-749-7601
International Electronic Research Corporation (IERC) 413 North Moss St. Burbank, CA 91502 Internet: www.ctscorp.com	818-842-7277
Tyco Electronics Chip Coolers™ P.O. Box 3668 Harrisburg, PA 17105-3668 Internet: www.chipcoolers.com	800-522-6752
Wakefield Engineering 33 Bridge St. Pelham, NH 03076 Internet: www.wakefield.com	603-635-5102

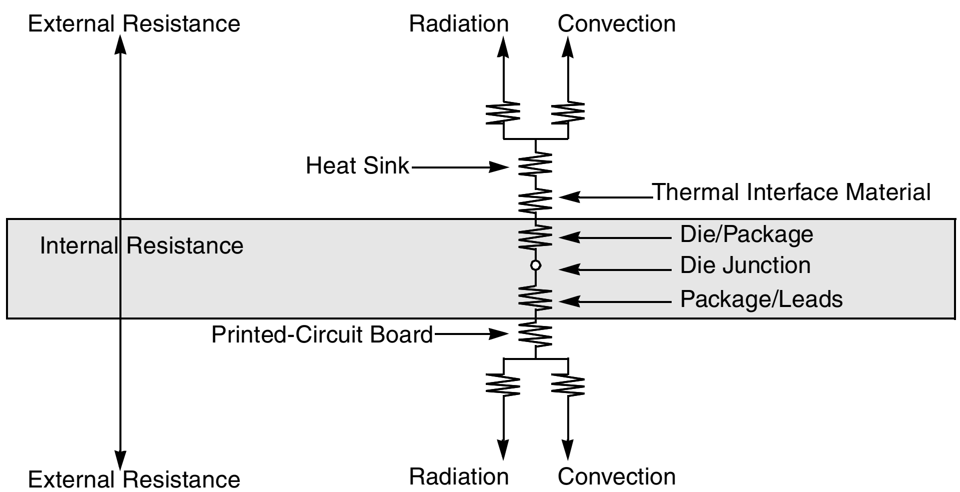
Selection of an appropriate heat sink depends on thermal performance at a given air velocity, spatial volume, mass, attachment method, assembly, and cost. Other heat sinks offered by Aavid Thermalloy, Alpha Novatech, IERC, Chip Coolers, and Wakefield Engineering offer different heat sink-to-ambient thermal resistances, and may or may not need airflow.

7.7.1 Internal Package Conduction Resistance

For the PBGA, die-up, packaging technology, shown in Figure 28, the intrinsic conduction thermal resistance paths are as follows:

- The die junction-to-case thermal resistance
- The die junction-to-ball thermal resistance

Figure 30 depicts the primary heat transfer path for a package with an attached heat sink mounted to a printed-circuit board.



(Note the internal versus external package resistance)

Figure 30. PBGA Package with Heat Sink Mounted to a Printed-Circuit Board

For this die-up, wire-bond PBGA package, heat generated on the active side of the chip is conducted mainly through the mold cap, the heat sink attach material (or thermal interface material), and finally through the heat sink where forced-air convection removes it.

7.7.2 Adhesives and Thermal Interface Materials

A thermal interface material should be used between the top of the mold cap and the bottom of the heat sink minimizes thermal contact resistance. For applications that attach the heat sink by a spring clip mechanism, Figure 31 shows the thermal performance of three thin-sheet thermal-interface materials (silicone, graphite/oil, floroether oil), a bare joint, and a joint with thermal grease as a function of contact pressure. As shown, the performance of these thermal interface materials improves with increasing contact pressure. Thermal grease significantly reduces the interface thermal resistance. That is, the bare joint offers a thermal resistance approximately seven times greater than the thermal grease joint.

A spring clip attaches heat sinks to holes in the printed-circuit board (see Figure 28). Therefore, the synthetic grease offers the best thermal performance, considering the low interface pressure. The selection of any thermal interface material depends on factors such as thermal performance requirements, manufacturability, service temperature, dielectric properties, and cost.

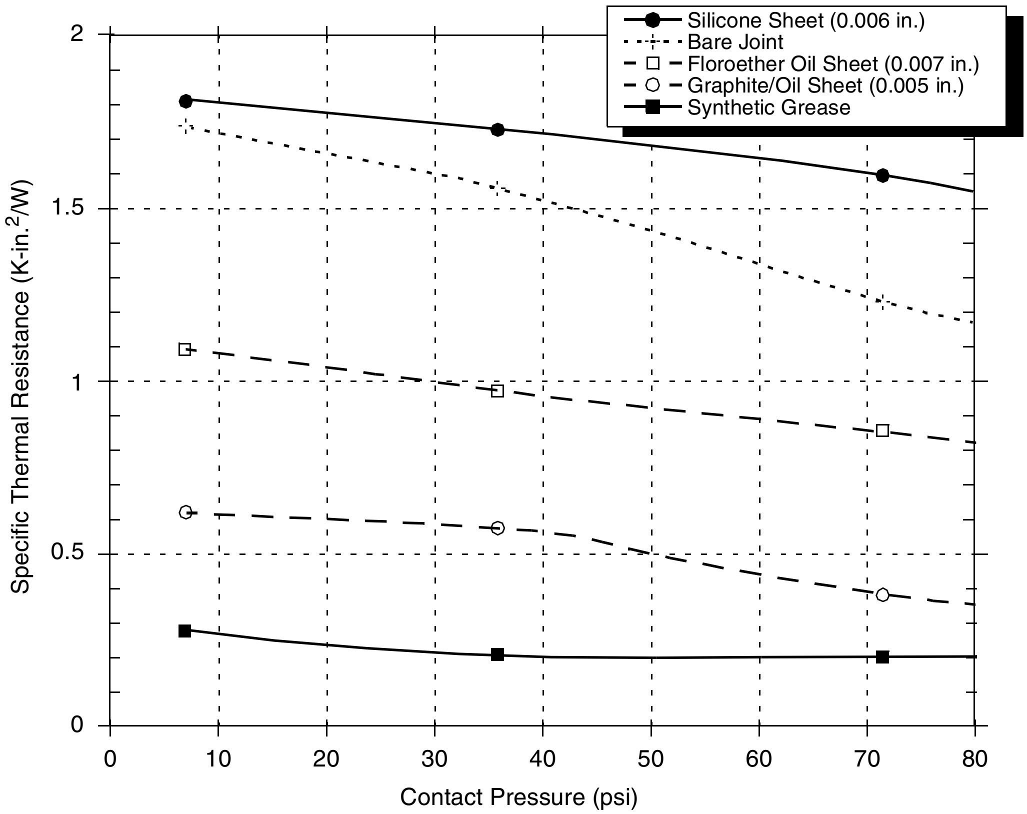


Figure 31. Thermal Performance of Select Thermal Interface Material

The board designer can choose among several types of thermal interface. Heat sink adhesive materials are selected on the basis of high conductivity and adequate mechanical strength to meet equipment shock/vibration requirements. Several commercially-available thermal interfaces and adhesive materials are provided by the following vendors:

The Bergquist Company 18930 West 78th St. Chanhassen, MN 55317 Internet: www.bergquistcompany.com	800-347-4572
Chomerics, Inc. 77 Dragon Ct. Woburn, MA 01888-4014 Internet: www.chomerics.com	781-935-4850
Dow-Corning Corporation Dow-Corning Electronic Materials 2200 W. Salzburg Rd. Midland, MI 48686-0997 Internet: www.dow.com	800-248-2481

8.1 Part Numbers Fully Addressed by This Document

Table 19 provides the Freescale part numbering nomenclature for the MPC8241. Note that the individual part numbers correspond to a maximum processor core frequency. For available frequencies, contact your local Freescale sales office. In addition to the processor frequency, the part numbering scheme also includes an application modifier that may specify special application conditions. Each part number also contains a revision code that refers to the die mask revision number. Read the Revision ID register at address offset 0x08 to determine the revision level.

Table 19. Part Numbering Nomenclature

MPC	*nnnn*	L	*xx*	*nnn*	*x*
Product Code	**Part Identifier**	**Process Descriptor**	**Package** [1]	**Processor Frequency** [2] **(MHz)**	**Revision Level**
MPC	8241	L = Standard spec. 0° to 105°C	ZQ = thick substrate and thick mold cap PBGA (two layers)	166, 200 1.8 V ± 100 mV	D:1.4 = Rev. ID:0x14
			ZQ = thick substrate and thick mold cap PBGA (four layers, thermally enhanced)	266 1.8 V ± 100 mV	
			VR = Lead-free version of package	166, 200, 266 1.8 V ± 100 mV	

Notes:
1. See Section 5, "Package Description," for more information on available package types.
2. Processor core frequencies supported by parts addressed by this specification only. Not all parts described in this specification support all core frequencies. Additionally, parts addressed by hardware specifications addendums may support other maximum core frequencies.

8.2 Part Numbers Not Fully Addressed by This Document

Parts with application modifiers or revision levels not fully addressed in this specification document are described in separate hardware specifications addendums that supplement and supersede this document (see Table 20).

Table 20. Part Numbers Addressed by MPC8241TXXPNS Series (Document No. MPC8241ECSO1AD))

MPC	*nnnn*	T	*xx*	*nnn*	*x*	
Product Code	**Part Identifier**	**Process Descriptor**	**Package** [1]	**Processor Frequency** [2] **(MHz)**	**Revision Level**	**Processor Version Register Value**