Welcome to E-XFL.COM

Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Active
Number of LABs/CLBs	-
Number of Logic Elements/Cells	-
Total RAM Bits	110592
Number of I/O	119
Number of Gates	600000
Voltage - Supply	1.425V ~ 1.575V
Mounting Type	Surface Mount
Operating Temperature	-55°C ~ 100°C (TJ)
Package / Case	256-LBGA
Supplier Device Package	256-FPBGA (17x17)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/afs600-fg256k

Related Documents

Datasheet

Core8051
www.microsemi.com/soc/ipdocs/Core8051_DS.pdf

Application Notes

Fusion FlashROM
http://www.microsemi.com/soc/documents/Fusion_FROM_AN.pdf
Fusion SRAM/FIFO Blocks
http://www.microsemi.com/soc/documents/Fusion_RAM_FIFO_AN.pdf
Using DDR in Fusion Devices
http://www.microsemi.com/soc/documents/Fusion_DDR_AN.pdf
Fusion Security
http://www.microsemi.com/soc/documents/Fusion_Security_AN.pdf
Using Fusion RAM as Multipliers
http://www.microsemi.com/soc/documents/Fusion_Multipliers_AN.pdf

Handbook

Cortex-M1 Handbook
www.microsemi.com/soc/documents/CortexM1_HB.pdf

User’s Guides

Fusion FPGA Fabric User’s Guide
http://www.microsemi.com/soc/documents/Fusion_UG.pdf
Designer User's Guide
http://www.microsemi.com/soc/documents/designer_UG.pdf
Fusion FPGA Fabric User’s Guide
http://www.microsemi.com/soc/documents/Fusion_UG.pdf
IGLOO, ProASIC3, SmartFusion, and Fusion Macro Library Guide
http://www.microsemi.com/soc/documents/pa3_libguide_ug.pdf
SmartGen, FlashROM, Flash Memory System Builder, and Analog System Builder User's Guide
http://www.microsemi.com/soc/documents/genguide_ug.pdf

White Papers

Fusion Technology
http://www.microsemi.com/soc/documents/Fusion_Tech_WP.pdf

Clocking Resources

The Fusion family has a robust collection of clocking peripherals, as shown in the block diagram in Figure 2-16. These on-chip resources enable the creation, manipulation, and distribution of many clock signals. The Fusion integrated RC oscillator produces a 100 MHz clock source with no external components. For systems requiring more precise clock signals, the Fusion family supports an on-chip crystal oscillator circuit. The integrated PLLs in each Fusion device can use the RC oscillator, crystal oscillator, or another on-chip clock signal as a source. These PLLs offer a variety of capabilities to modify the clock source (multiply, divide, synchronize, advance, or delay). Utilizing the CCC found in the popular ProASIC3 family, Fusion incorporates six CCC blocks. The CCCs allow access to Fusion global and local clock distribution nets, as described in the "Global Resources (VersaNets)" section on page 2-11.

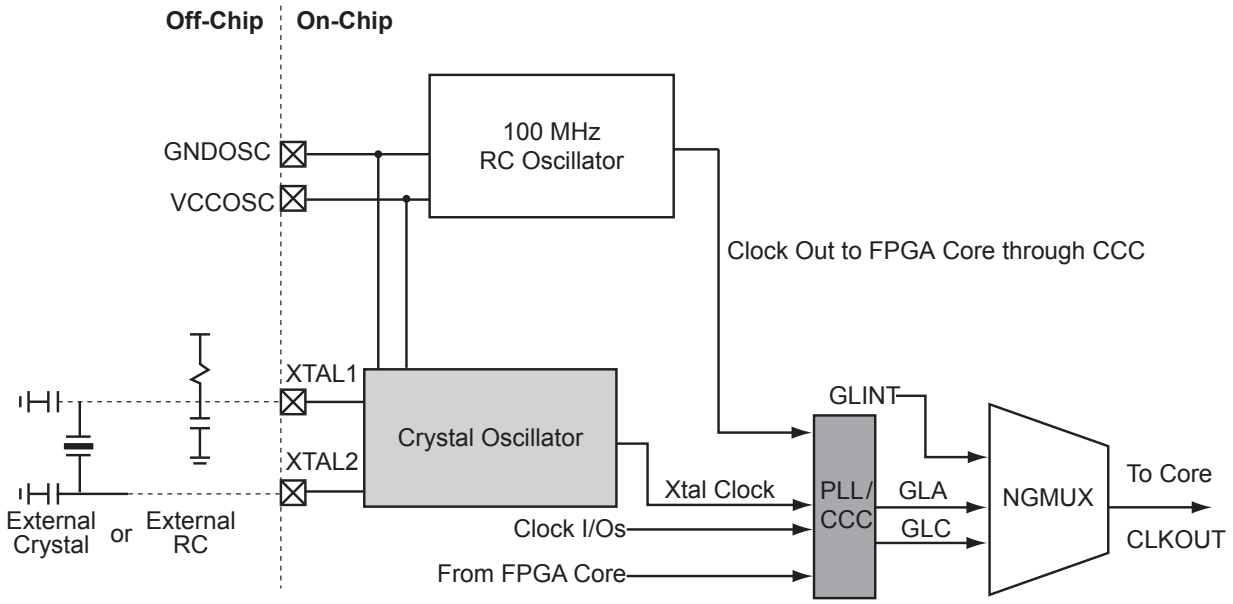

Figure 2-16 • **Fusion Clocking Options**

Global Buffers with No Programmable Delays

The CLKBUF and CLKBUF_LVPECL/LVDS macros are composite macros that include an I/O macro driving a global buffer, hardwired together (Figure 2-20).

The CLKINT macro provides a global buffer function driven by the FPGA core.

The CLKBUF, CLKBUF_LVPECL/LVDS, and CLKINT macros are pass-through clock sources and do not use the PLL or provide any programmable delay functionality.

Many specific CLKBUF macros support the wide variety of single-ended and differential I/O standards supported by Fusion devices. The available CLKBUF macros are described in the *IGLOO, ProASIC3, SmartFusion, and Fusion Macro Library Guide*.

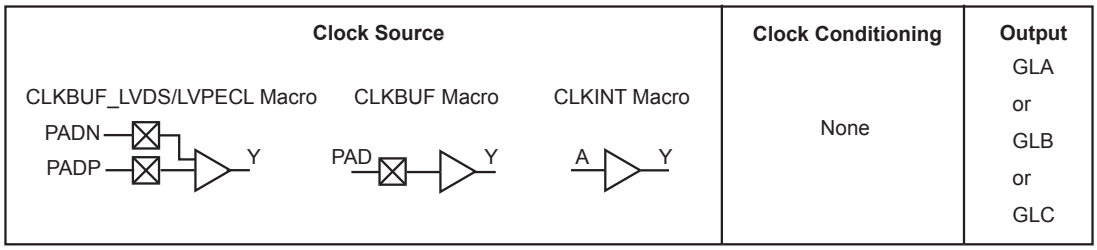

Figure 2-20 • **Global Buffers with No Programmable Delay**

Access to the FB is controlled by the BUSY signal. The BUSY output is synchronous to the CLK signal. FB operations are only accepted in cycles where BUSY is logic 0.

Write Operation

Write operations are initiated with the assertion of the WEN signal. Figure 2-35 illustrates the multiple Write operations.

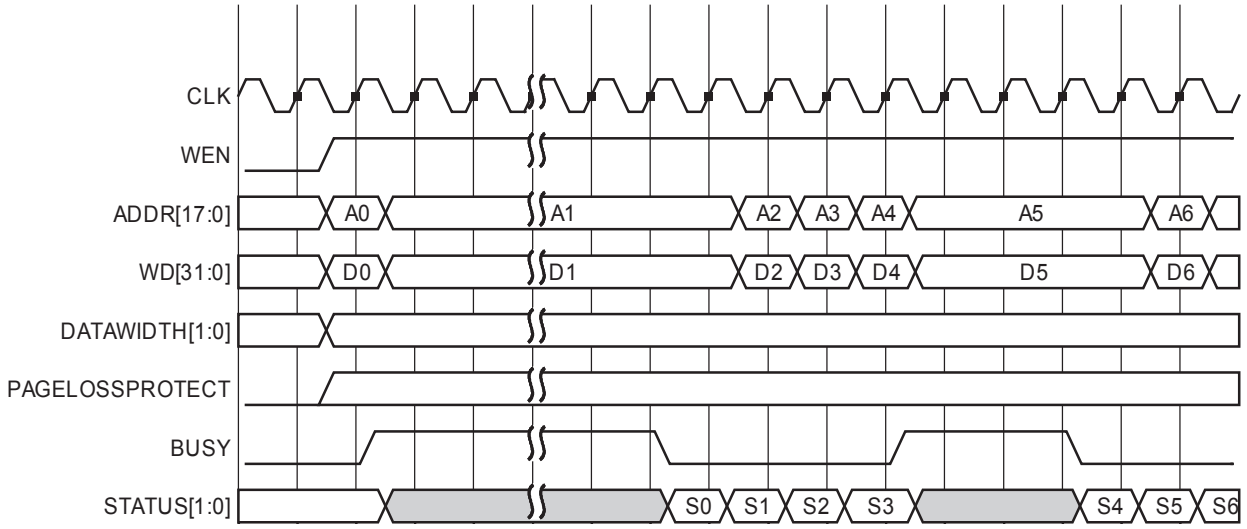

Figure 2-35 • **FB Write Waveform**

When a Write operation is initiated to a page that is currently not in the Page Buffer, the FB control logic will issue a BUSY signal to the user interface while the page is loaded from the FB Array into the Page Buffer. A Copy Page operation takes no less than 55 cycles and could take more if a Write or Unprotect Page operation is started while the NVM is busy pre-fetching a block. The basic operation is to read a block from the array into the block register (5 cycles) and then write the block register to the page buffer (1 cycle) and if necessary, when the copy is complete, reading the block being written from the page buffer into the block buffer (1 cycle). A page contains 9 blocks, so 9 blocks multiplied by 6 cycles to read/write each block, plus 1 is 55 cycles total. Subsequent writes to the same block of the page will incur no busy cycles. A write to another block in the page will assert BUSY for four cycles (five cycles when PIPE is asserted), to allow the data to be written to the Page Buffer and have the current block loaded into the Block Buffer.

Write operations are considered successful as long as the STATUS output is '00'. A non-zero STATUS indicates that an error was detected during the operation and the write was not performed. Note that the STATUS output is "sticky"; it is unchanged until another operation is started.

Only one word can be written at a time. Write word width is controlled by the DATAWIDTH bus. Users are responsible for keeping track of the contents of the Page Buffer and when to program it to the array. Just like a regular RAM, writing to random addresses is possible. Users can write into the Page Buffer in any order but will incur additional BUSY cycles. It is not necessary to modify the entire Page Buffer before saving it to nonvolatile memory.

Write errors include the following:

1. Attempting to write a page that is Overwrite Protected (STATUS = '01'). The write is not performed.
2. Attempting to write to a page that is not in the Page Buffer when Page Loss Protection is enabled (STATUS = '11'). The write is not performed.

FlashROM Characteristics

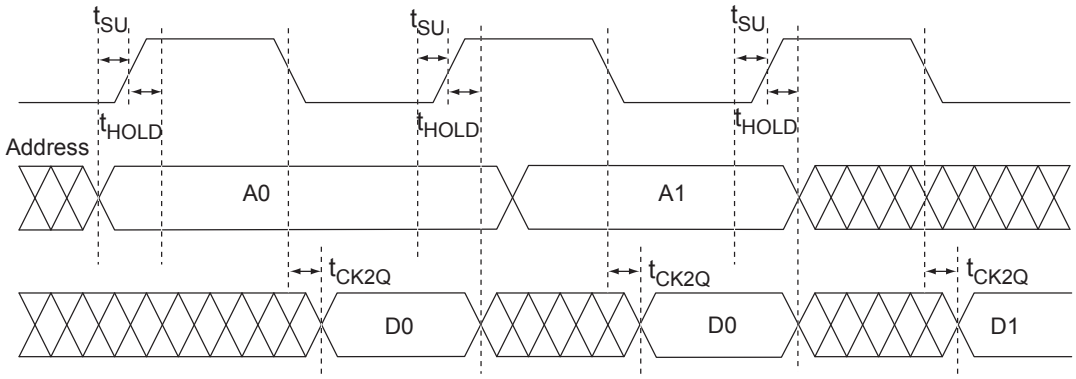

Figure 2-46 • **FlashROM Timing Diagram**

Table 2-25 • **FlashROM Access Time, Extended Temperature Conditions: T_J = 100°C, Worst-Case VCC = 1.425 V**

Parameter	Description	–2	–1	Std.	Units
t_{SU}	Address Setup Time	0.55	0.63	0.74	ns
t_{HOLD}	Address Hold Time	0.00	0.00	0.00	ns
t_{CK2Q}	Clock to Out	16.73	19.06	22.41	ns
FMAX	Maximum Clock frequency	40.00	40.00	40.00	MHz

Gate Driver

The Fusion Analog Quad includes a Gate Driver connected to the Quad's AG pin (Figure 2-73). Designed to work with external p- or n-channel MOSFETs, the Gate driver is a configurable current sink or source and requires an external pull-up or pull-down resistor. The AG supports 4 selectable gate drive levels: 1 µA, 3 µA, 10 µA, and 30 µA (Figure 2-74 on page 2-91). The AG also supports a High Current Drive mode in which it can sink 20 mA; in this mode the switching rate is approximately 1.3 MHz with 100 ns turn-on time and 600 ns turn-off time. Modeled on an open-drain-style output, it does not output a voltage level without an appropriate pull-up or pull-down resistor. If 1 V is forced on the drain, the current sinking/sourcing will exceed the ability of the transistor, and the device could be damaged.

The AG pad is turned on via the corresponding GDON*x* pin in the Analog Block macro, where *x* is the number of the corresponding Analog Quad for the AG pad to be enabled (GDON0 to GDON9).

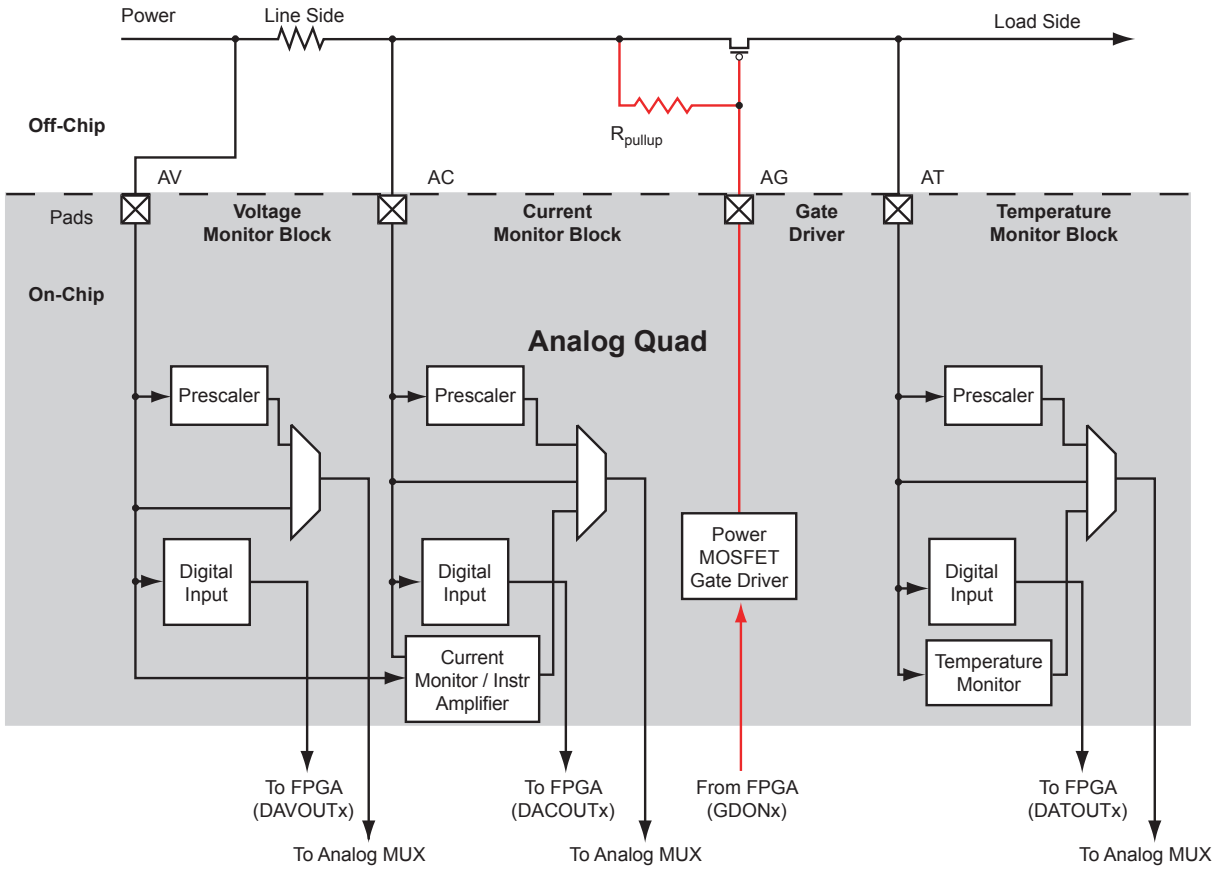

Figure 2-73 • **Gate Driver**

The gate-to-source voltage (V_{gs}) of the external MOSFET is limited to the programmable drive current times the external pull-up or pull-down resistor value (EQ 5).

$$V_{gs} \leq I_g \times (R_{pullup} \text{ or } R_{pulldown})$$

EQ 5

The rate at which the gate voltage of the external MOSFET slews is determined by the current, I_g, sourced or sunk by the AG pin and the gate-to-source capacitance, C_{GS}, of the external MOSFET. As an approximation, the slew rate is given by EQ 6.

$$dv/dt = I_g / C_{GS}$$

EQ 6

ADC Description

The Fusion ADC is a 12-bit SAR ADC. It offers a wide variety of features for different use models. Figure 2-79 shows a block diagram of the Fusion ADC.

- Configurable resolution: 8-bit, 10-bit, and 12-bit mode
- DNL: 0.6 LSB for 10-bit mode
- INL: 0.4 LSB for 10-bit mode
- No missing code
- Internal VAREF = 2.56 V
- Maximum Sample Rate = 600 Ksps
- Power-up calibration and dynamic calibration after every sample to compensate for temperature drift over time

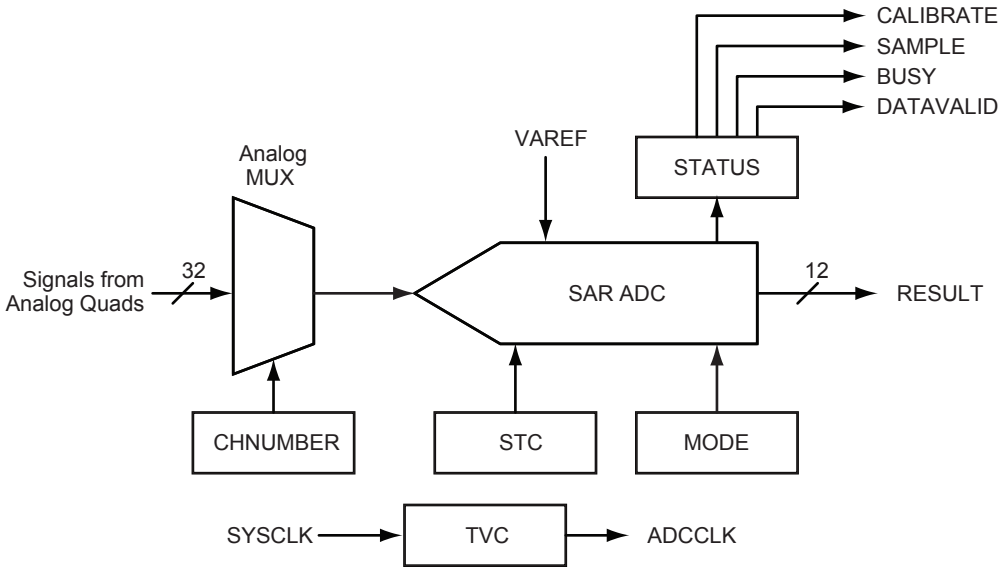

Figure 2-79 • **ADC Simplified Block Diagram**

ADC Theory of Operation

An analog-to-digital converter is used to capture discrete samples of a continuous analog voltage and provide a discrete binary representation of the signal. Analog-to-digital converters are generally characterized in three ways:

- Input voltage range
- Resolution
- Bandwidth or conversion rate

The input voltage range of an ADC is determined by its reference voltage (VREF). Fusion devices include an internal 2.56 V reference, or the user can supply an external reference of up to 3.3 V. The following examples use the internal 2.56 V reference, so the full-scale input range of the ADC is 0 to 2.56 V.

The resolution (LSB) of the ADC is a function of the number of binary bits in the converter. The ADC approximates the value of the input voltage using 2n steps, where n is the number of bits in the converter. Each step therefore represents VREF÷ 2n volts. In the case of the Fusion ADC configured for 12-bit operation, the LSB is 2.56 V / 4096 = 0.625 mV.

Finally, bandwidth is an indication of the maximum number of conversions the ADC can perform each second. The bandwidth of an ADC is constrained by its architecture and several key performance characteristics.

Typical Performance Characteristics

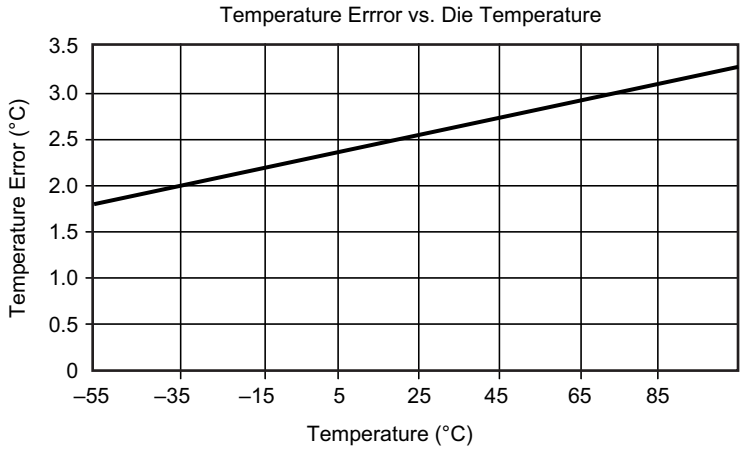

Figure 2-93 • **Temperature Error**

Table 2-48 • **Temperature vs. Average Fitted Error**

Temperature (°C)	Averaged Fitted Error
100	3.2469
90	3.1559
80	3.0649
70	2.9739
60	2.8829
50	2.7919
40	2.7009
30	2.6099
20	2.5189
10	2.4279
0	2.3369
–10	2.2459
–20	2.1549
–30	2.0639
–40	1.9729
–55	1.8364

Table 2-85 • **Fusion Pro I/O Attributes vs. I/O Standard Applications**

I/O Standards	SLEW (output only)	OUT_DRIVE (output only)	SKEW (all macros with OE)	RES_PULL	OUT_LOAD (output only)	COMBINE_REGISTER	IN_DELAY (input only)	IN_DELAY_VAL (input only)	SCHMITT_TRIGGER (input only)	HOT_SWAPPABLE
LVTTL/LVCMOS 3.3 V	3	3	3	3	3	3	3	3	3	3
LVCMOS 2.5 V	3	3	3	3	3	3	3	3	3	3
LVCMOS 2.5/5.0 V	3	3	3	3	3	3	3	3	3	3
LVCMOS 1.8 V	3	3	3	3	3	3	3	3	3	3
LVCMOS 1.5 V	3	3	3	3	3	3	3	3	3	3
PCI (3.3 V)			3		3	3	3	3		
PCI-X (3.3 V)	3		3		3	3	3	3		
GTL+ (3.3 V)			3		3	3	3	3		3
GTL+ (2.5 V)			3		3	3	3	3		3
GTL (3.3 V)			3		3	3	3	3		3
GTL (2.5 V)			3		3	3	3	3		3
HSTL Class I			3		3	3	3	3		3
HSTL Class II			3		3	3	3	3		3
SSTL2 Class I and II			3		3	3	3	3		3
SSTL3 Class I and II			3		3	3	3	3		3
LVDS, B-LVDS, M-LVDS			3			3	3	3		3
LVPECL						3	3	3		3

Overview of I/O Performance

Summary of I/O DC Input and Output Levels – Default I/O Software Settings

Table 2-86 • Summary of Maximum and Minimum DC Input and Output Levels Applicable to Extended Temperature Conditions Applicable to Pro I/Os

I/O Standard	Drive Strength	Slew Rate	VIL		VIH		VOL	VOH	IOL	IOH
			Min. V	Max. V	Min. V	Max. V	Max. V	Min. V	mA	mA
3.3 V LVTTL / 3.3 V LVCMOS	12 mA	High	–0.3	0.8	2	3.6	0.4	2.4	12	12
2.5 V LVCMOS	12 mA	High	–0.3	0.7	1.7	3.6	0.7	1.7	12	12
1.8 V LVCMOS	12 mA	High	–0.3	0.35 * VCCI	0.65 * VCCI	3.6	0.45	VCCI – 0.45	12	12
1.5 V LVCMOS	12 mA	High	–0.3	0.35 * VCCI	0.65 * VCCI	3.6	0.25 * VCCI	0.75 * VCCI	12	12
3.3 V PCI	Per PCI Specification									
3.3 V PCI-X	Per PCI-X Specification									
3.3 V GTL	20 mA[2]	High	–0.3	VREF – 0.05	VREF + 0.05	3.6	0.4	–	20	20
2.5 V GTL	20 mA[2]	High	–0.3	VREF – 0.05	VREF + 0.05	3.6	0.4	–	20	20
3.3 V GTL+	35 mA	High	–0.3	VREF – 0.1	VREF + 0.1	3.6	0.6	–	35	35
2.5 V GTL+	33 mA	High	–0.3	VREF – 0.1	VREF + 0.1	3.6	0.6	–	33	33
HSTL (I)	8 mA	High	–0.3	VREF – 0.1	VREF + 0.1	3.6	0.4	VCCI – 0.4	8	8
HSTL (II)	15 mA[2]	High	–0.3	VREF – 0.1	VREF + 0.1	3.6	0.4	VCCI – 0.4	15	15
SSTL2 (I)	15 mA	High	–0.3	VREF – 0.2	VREF + 0.2	3.6	0.54	VCCI – 0.62	15	15
SSTL2 (II)	18 mA	High	–0.3	VREF – 0.2	VREF + 0.2	3.6	0.35	VCCI – 0.43	18	18
SSTL3 (I)	14 mA	High	–0.3	VREF – 0.2	VREF + 0.2	3.6	0.7	VCCI – 1.1	14	14
SSTL3 (II)	21 mA	High	–0.3	VREF – 0.2	VREF + 0.2	3.6	0.5	VCCI – 0.9	21	21

Notes:

1. *Currents are measured at 100°C junction temperature.*
2. *Output drive strength is below JEDEC specification.*
3. *Output slew rate can be extracted by the IBIS models.*

Table 2-87 • Summary of Maximum and Minimum DC Input and Output Levels Applicable to Extended Temperature Conditions Applicable to Advanced I/Os

I/O Standard	Drive Strength	Slew Rate	VIL		VIH		VOL	VOH	I_{OL}	I_{OH}
			Min. V	Max. V	Min. V	Max. V	Max. V	Min. V	mA	mA
3.3 V LVTTL / 3.3 V LVCMOS	12 mA	High	–0.3	0.8	2	3.6	0.4	2.4	12	12
2.5 V LVCMOS	12 mA	High	–0.3	0.7	1.7	2.7	0.7	1.7	12	12
1.8 V LVCMOS	12 mA	High	–0.3	0.35 * VCCI	0.65 * VCCI	1.9	0.45	VCCI – 0.45	12	12
1.5 V LVCMOS	12 mA	High	–0.3	0.35 * VCCI	0.65 * VCCI	1.575	0.25 * VCCI	0.75 * VCCI	12	12
3.3 V PCI	Per PCI specifications									
3.3 V PCI-X	Per PCI-X specifications									

Note: Currents are measured at 100°C junction temperature.

Table 2-94 • **I/O Output Buffer Maximum Resistances [1] (continued)**

Standard	Drive Strength	$R_{PULL-DOWN}$ (ohms) [2]	$R_{PULL-UP}$ (ohms) [3]
HSTL (I)	8 mA	50	50
HSTL (II)	15 mA	25	25
SSTL2 (I)	17 mA	27	31
SSTL2 (II)	21 mA	13	15
SSTL3 (I)	16 mA	44	69
SSTL3 (II)	24 mA	18	32
Applicable to Advanced I/O Banks			
Standard	**Drive Strength**	**RPULL-DOWN (Ω)[2]**	**RPULL-UP (Ω)[3]**
3.3 V LVTTL / 3.3 V LVCMOS	2 mA	100	300
	4 mA	100	300
	6 mA	50	150
	8 mA	50	150
	12 mA	25	75
	16 mA	17	50
	24 mA	11	33
2.5 V LVCMOS	2 mA	100	200
	4 mA	100	200
	6 mA	50	100
	8 mA	50	100
	12 mA	25	50
	16 mA	20	40
	24 mA	11	22
1.8 V LVCMOS	2 mA	200	225
	4 mA	100	112
	6 mA	50	56
	8 mA	50	56
	12 mA	20	22
	16 mA	20	22

Notes:

1. *These maximum values are provided for informational reasons only. Minimum output buffer resistance values depend on VCC, drive strength selection, temperature, and process. For board design considerations and detailed output buffer resistances, use the corresponding IBIS models located on the Microsemi SoC Products Group website: www.microsemi.com/soc/techdocs/models/ibis.html.*
2. *R(PULL-DOWN-MAX) = VOLspec / IOLspec*
3. *R(PULL-UP-MAX) = (VCCImax – VOHspec) / IOHspec*

Timing Characteristics

Table 2-107 • 2.5 V LVCMOS Low Slew, Extended Temperature Case Conditions: T_J = 100°C, Worst Case VCC = 1.425 V, Worst Case VCCI = 2.3 V
Applicable to Pro I/O Banks

Drive Strength	Speed Grade	t_{DOUT}	t_{DP}	t_{DIN}	t_{PY}	t_{PYS}	t_{EOUT}	t_{ZL}	t_{ZH}	t_{LZ}	t_{HZ}	t_{ZLS}	t_{ZHS}	Units
4 mA	Std.	0.68	12.66	0.05	1.59	1.75	0.44	12.89	12.24	2.87	2.32	15.25	14.60	ns
	–1	0.58	10.77	0.04	1.36	1.49	0.38	10.97	10.42	2.44	1.97	12.97	12.42	ns
	–2	0.51	9.45	0.03	1.19	1.31	0.33	9.63	9.14	2.14	1.73	11.39	10.90	ns
8 mA	Std.	0.68	9.21	0.05	1.59	1.75	0.44	9.38	8.45	3.27	3.09	11.74	10.81	ns
	–1	0.58	7.83	0.04	1.36	1.49	0.38	7.98	7.19	2.78	2.63	9.98	9.19	ns
	–2	0.51	6.88	0.03	1.19	1.31	0.33	7.00	6.31	2.44	2.31	8.76	8.07	ns
12 mA	Std.	0.68	7.14	0.05	1.59	1.75	0.44	7.28	6.44	3.55	3.58	9.63	8.80	ns
	–1	0.58	6.08	0.04	1.36	1.49	0.38	6.19	5.48	3.02	3.04	8.20	7.48	ns
	–2	0.51	5.33	0.03	1.19	1.31	0.33	5.43	4.81	2.65	2.67	7.19	6.57	ns
16 mA	Std.	0.68	6.65	0.05	1.59	1.75	0.44	6.77	6.04	3.61	3.71	9.13	8.40	ns
	–1	0.58	5.66	0.04	1.36	1.49	0.38	5.76	5.14	3.07	3.16	7.77	7.14	ns
	–2	0.51	4.97	0.03	1.19	1.31	0.33	5.06	4.51	2.69	2.77	6.82	6.27	ns
24 mA	Std.	0.68	6.25	0.05	1.59	1.75	0.44	6.37	6.02	3.69	4.22	8.73	8.37	ns
	–1	0.58	5.32	0.04	1.36	1.49	0.38	5.42	5.12	3.14	3.59	7.43	7.12	ns
	–2	0.51	4.67	0.03	1.19	1.31	0.33	4.76	4.49	2.75	3.15	6.52	6.25	ns

Note: For the derating values at specific junction temperature and voltage supply levels, refer to Table 3-7 on page 3-10.

Table 2-158 • LVDS AC Waveforms, Measuring Points, and Capacitive Loads

Input Low (V)	Input High (V)	Measuring Point* (V)	V_{REF} (typ.) (V)
1.075	1.325	Cross point	–

*Note: *Measuring point = V_{trip}. See Table 2-89 on page 2-166 for a complete table of trip points.*

Timing Characteristics

Table 2-159 • LVDS
Extended Temperature Case Conditions: T_J = 100°C, Worst Case VCC = 1.425 V, Worst Case VCCI = 2.3 V
Applicable to Pro I/O Banks

Speed Grade	t_{DOUT}	t_{DP}	t_{DIN}	t_{PY}	Units
Std.	0.68	1.98	0.05	1.97	ns
–1	0.58	1.69	0.04	1.68	ns
–2	0.51	1.48	0.03	1.47	ns

Note: For the derating values at specific junction temperature and voltage supply levels, refer to Table 3-7 on page 3-10.

Table 2-160 • LVDS
Extended Temperature Case Conditions: T_J = 100°C, Worst Case VCC = 1.425 V, Worst Case VCCI = 2.3 V
Applicable to Advanced I/O Banks

Speed Grade	t_{DOUT}	t_{DP}	t_{DIN}	t_{PY}	Units
Std.	0.68	1.98	0.05	1.75	ns
–1	0.58	1.69	0.04	1.49	ns
–2	0.51	1.48	0.03	1.31	ns

Note: For the derating values at specific junction temperature and voltage supply levels, refer to Table 3-7 on page 3-10.

B-LVDS/M-LVDS

Bus LVDS (B-LVDS) and Multipoint LVDS (M-LVDS) specifications extend the existing LVDS standard to high-performance multipoint bus applications. Multidrop and multipoint bus configurations can contain any combination of drivers, receivers, and transceivers. Microsemi LVDS drivers provide the higher drive current required by B-LVDS and M-LVDS to accommodate the loading. The driver requires series terminations for better signal quality and to control voltage swing. Termination is also required at both ends of the bus, since the driver can be located anywhere on the bus. These configurations can be implemented using TRIBUF_LVDS and BIBUF_LVDS macros along with appropriate terminations. Multipoint designs using Microsemi LVDS macros can achieve up to 200 MHz with a maximum of 20 loads. A sample application is given in Figure 2-133 on page 2-209. The input and output buffer delays are available in the LVDS section in Table 2-161 on page 2-210.

Refer to the "User I/O Naming Convention" section on page 2-159 for a description of naming of global pins.

JTAG Pins

Fusion devices have a separate bank for the dedicated JTAG pins. The JTAG pins can be run at any voltage from 1.5 V to 3.3 V (nominal). VCC must also be powered for the JTAG state machine to operate, even if the device is in bypass mode; VJTAG alone is insufficient. Both VJTAG and VCC to the Fusion part must be supplied to allow JTAG signals to transition the Fusion device.

Isolating the JTAG power supply in a separate I/O bank gives greater flexibility with supply selection and simplifies power supply and PCB design. If the JTAG interface is neither used nor planned to be used, the VJTAG pin together with the TRST pin could be tied to GND.

TCK Test Clock

Test clock input for JTAG boundary scan, ISP, and UJTAG. The TCK pin does not have an internal pull-up/-down resistor. If JTAG is not used, Microsemi recommends tying off TCK to GND or VJTAG through a resistor placed close to the FPGA pin. This prevents JTAG operation in case TMS enters an undesired state.

Note that to operate at all VJTAG voltages, 500 Ω to 1 kΩ will satisfy the requirements. Refer to Table 2-174 for more information. TDITest Data Input

***Table 2-174* • Recommended Tie-Off Values for the TCK and TRST Pins**

VJTAG	Tie-Off Resistance[2, 3]
VJTAG at 3.3 V	200 Ω to 1 kΩ
VJTAG at 2.5 V	200 Ω to 1 kΩ
VJTAG at 1.8 V	500 Ω to 1 kΩ
VJTAG at 1.5 V	500 Ω to 1 kΩ

Notes:

1. *Equivalent parallel resistance if more than one device is on JTAG chain.*
2. *The TCK pin can be pulled up/down.*
3. *The TRST pin can only be pulled down.*

Serial input for JTAG boundary scan, ISP, and UJTAG usage. There is an internal weak pull-up resistor on the TDI pin.

TDO Test Data Output

Serial output for JTAG boundary scan, ISP, and UJTAG usage.

TMS Test Mode Select

The TMS pin controls the use of the IEEE1532 boundary scan pins (TCK, TDI, TDO, TRST). There is an internal weak pull-up resistor on the TMS pin.

TRST Boundary Scan Reset Pin

The TRST pin functions as an active low input to asynchronously initialize (or reset) the boundary scan circuitry. There is an internal weak pull-up resistor on the TRST pin. If JTAG is not used, an external pull-down resistor could be included to ensure the TAP is held in reset mode. The resistor values must be chosen from Table 2-174 and must satisfy the parallel resistance value requirement. The values in Table 2-174 correspond to the resistor recommended when a single device is used and to the equivalent parallel resistor when multiple devices are connected via a JTAG chain.

In critical applications, an upset in the JTAG circuit could allow entering an undesired JTAG state. In such cases, Microsemi recommends tying off TRST to GND through a resistor placed close to the FPGA pin.

Note that to operate at all VJTAG voltages, 500 Ω to 1 kΩ will satisfy the requirements.

3 – DC and Power Characteristics

General Specifications

Operating Conditions

Stresses beyond those listed in Table 3-1 may cause permanent damage to the device.

Exposure to absolute maximum rated conditions for extended periods may affect device reliability. Devices should not be operated outside the recommended operating ranges specified in Table 3-2 on page 3-3.

Table 3-1 • **Absolute Maximum Ratings**

Symbol	Parameter	Limit	Units
VCC	DC core supply voltage	–0.3 to 1.65	V
VJTAG	JTAG DC voltage	–0.3 to 3.75	V
VPUMP	Programming voltage	–0.3 to 3.75	V
VCCPLL	Analog power supply (PLL)	–0.3 to 1.65	V
VCCI	DC I/O output buffer supply voltage	–0.3 to 3.75	V
VI	I/O input voltage [1]	–0.3 V to 3.6 V (when I/O hot insertion mode is enabled) –0.3 V to (VCCI + 1 V) or 3.6 V, whichever voltage is lower (when I/O hot-insertion mode is disabled)	V
VCC33A	+3.3 V power supply	–0.3 to 3.75 [2]	V
VCC33PMP	+3.3 V power supply	–0.3 to 3.75 [2]	V
VAREF	Voltage reference for ADC	–0.3 to 3.75	V
VCC15A	Digital power supply for the analog system	–0.3 to 1.65	V
VCCNVM	Embedded flash power supply	–0.3 to 1.65	V
VCCOSC	Oscillator power supply	–0.3 to 3.75	V

Notes:

1. *The device should be operated within the limits specified by the datasheet. During transitions, the input signal may undershoot or overshoot according to the limits shown in Table 3-4 on page 3-5.*
2. *Analog data not valid beyond 3.65 V.*
3. *The high current mode has a maximum power limit of 15 mW. Appropriate current limit resistors must be used, based on voltage on the pad.*
4. *For flash programming and retention maximum limits, refer to Table 3-5 on page 3-5. For recommended operating limits refer to Table 3-2 on page 3-3.*
5. *Negative input is not supported between –40°C and –55°C.*
6. *Positive input is not supported between –40°C and –55°C.*

Thermal Characteristics

Introduction

The temperature variable in the Microsemi Designer software refers to the junction temperature, not the ambient, case, or board temperatures. This is an important distinction because dynamic and static power consumption will cause the chip's junction temperature to be higher than the ambient, case, or board temperatures. EQ 1 through EQ 3 give the relationship between thermal resistance, temperature gradient, and power.

$$\theta_{JA} = \frac{T_J - \theta_A}{P}$$

EQ 1

$$\theta_{JB} = \frac{T_J - T_B}{P}$$

EQ 2

$$\theta_{JC} = \frac{T_J - T_C}{P}$$

EQ 3

where

θ_{JA} = Junction-to-air thermal resistance
θ_{JB} = Junction-to-board thermal resistance
θ_{JC} = Junction-to-case thermal resistance
T_J = Junction temperature
T_A = Ambient temperature
T_B = Board temperature (measured 1.0 mm away from the package edge)
T_C = Case temperature
P = Total power dissipated by the device

Table 3-6 • **Package Thermal Resistance**

Product	θ_{JA}			θ_{JC}	θ_{JB}	Units
	Still Air	1.0 m/s	2.5 m/s			
AFS600-FG256	28.9	25.2	23.5	6.8	19.9	°C/W
AFS1500-FG256	23.3	19.6	18.0	4.3	14.2	°C/W
AFS600-FG484	21.8	18.2	16.7	7.7	16.8	°C/W
AFS1500-FG484	21.6	16.8	15.2	5.6	14.9	°C/W
AFS1500-FG676	TBD	TBD	TBD	TBD	TBD	°C/W

Table 3-9 • **AFS600 Quiescent Supply Current Characteristics**

Parameter	Description	Conditions	Temp.	Min	Typ	Max	Unit
ICC[1]	1.5 V quiescent current	Operational standby[4], VCC = 1.575 V	T_J = 25°C		13	25	mA
			T_J = 85°C		20	45	mA
			T_J=100°C		25	75	mA
		Standby mode[5] or Sleep mode[6], VCC = 0 V			0	0	µA
ICC33[2]	3.3 V analog supplies current	Operational standby[4], VCC33 = 3.63 V	T_J = 25°C		9.8	13	mA
			T_J = 85°C		10.7	14	mA
			T_J = 100°C		10.8	15	mA
		Operational standby, only Analog Quad and –3.3 V output ON, VCC33 = 3.63 V	T_J = 25°C		0.31	2	mA
			T_J = 85°C		0.35	2	mA
			T_J = 100°C		0.45	2	mA
		Standby mode[5], VCC33 = 3.63 V	T_J = 25°C		2.8	3.6	mA
			T_J = 85°C		2.9	4	mA
			T_J = 100°C		3.5	6	mA
		Sleep mode[6], VCC33 = 3.63 V	T_J = 25°C		17	19	µA
			T_J = 85°C		18	20	µA
			T_J = 100°C		24	25	µA
ICCI[3]	I/O quiescent current	Operational standby,[4] VCCIx = 3.63 V	T_J = 25°C		417	648	µA
			T_J = 85°C		417	648	µA
			T_J = 100°C		417	649	µA
I_{JTAG}	JTAG I/O quiescent current	Operational standby,[4] VJTAG = 3.63 V	T_J = 25°C		80	100	µA
			T_J = 85°C		80	100	µA
			T_J = 100°C		80	100	µA
		Standby mode[5] or Sleep mode[6], VJTAG = 0 V			0	0	µA

Notes:

1. *ICC is the 1.5 V power supplies, ICC and ICC15A.*
2. *ICC33A includes ICC33A, ICC33PMP, and ICCOSC.*
3. *ICCI includes all ICCI0, ICCI1, ICCI2, and ICCI4.*
4. *Operational standby is when the Fusion device is powered up, all blocks are used, no I/O is toggling, Voltage Regulator is loaded with 200 mA, VCC33PMP is ON, XTAL is ON, and ADC is ON.*
5. *XTAL is configured as high gain, VCC = VJTAG = VPUMP = 0 V.*
6. *Sleep Mode, VCC = VJTAG = VPUMP = 0 V. Sleep mode is not supported between –40°C and –55°C.*

Static Power Consumption of Various Internal Resources

***Table 3-13* • Different Components Contributing to the Static Power Consumption in Fusion Devices**

Parameter	Definition	Power Supply		Device-Specific Static Contributions		Units
		Name	Setting	AFS1500	AFS600	
PDC1	Core static power contribution in operating mode	VCC	1.5 V	18	7.5	mW
PDC2	Device static power contribution in sleep mode*	VCC33A	3.3 V	0.66		mW
PDC3	Device static power contribution in standby mode	VCC33A	3.3 V	0.03		mW
PDC4	NVM static power contribution	VCC	1.5 V	1.19		mW
PDC5	Analog Block static power contribution of ADC	VCC33A	3.3 V	8.25		mW
PDC6	Analog Block static power contribution per Quad	VCC33A	3.3 V	3.3		mW
PDC7	Static contribution per input pin – standard dependent contribution	VCCI	See Table 3-10 on page 3-15			
PDC8	Static contribution per output pin – standard dependent contribution	VCCI	See Table 3-11 on page 3-17			
PDC9	Static contribution for PLL	VCC	1.5 V	2.55		mW

*Note: *Sleep mode is not supported between –40°C and –55°C.*

Power Calculation Methodology

This section describes a simplified method to estimate power consumption of an application. For more accurate and detailed power estimations, use the SmartPower tool in the Libero SoC software.

The power calculation methodology described below uses the following variables:

- The number of PLLs as well as the number and the frequency of each output clock generated
- The number of combinatorial and sequential cells used in the design
- The internal clock frequencies
- The number and the standard of I/O pins used in the design
- The number of RAM blocks used in the design
- The number of NVM blocks used in the design
- The number of Analog Quads used in the design
- Toggle rates of I/O pins as well as VersaTiles—guidelines are provided in Table 3-14 on page 3-23.
- Enable rates of output buffers—guidelines are provided for typical applications in Table 3-15 on page 3-23.
- Read rate and write rate to the RAM—guidelines are provided for typical applications in Table 3-15 on page 3-23.
- Read rate to the NVM blocks

The calculation should be repeated for each clock domain defined in the design.

FG484		
Pin Number	AFS600 Function	AFS1500 Function
E9	NC	IO08PDB0V1
E10	GND	GND
E11	IO15NDB1V0	IO22NDB1V0
E12	IO15PDB1V0	IO22PDB1V0
E13	GND	GND
E14	NC	IO32PPB1V1
E15	NC	IO36NPB1V2
E16	VCCIB1	VCCIB1
E17	GND	GND
E18	NC	IO47NPB2V0
E19	IO33PDB2V0	IO49PDB2V0
E20	VCCIB2	VCCIB2
E21	IO32NDB2V0	IO46NDB2V0
E22	GBC2/IO32PDB2V0	GBC2/IO46PDB2V0
F1	IO80NDB4V0	IO118NDB4V0
F2	IO80PDB4V0	IO118PDB4V0
F3	NC	IO119NSB4V0
F4	IO84NDB4V0	IO124NDB4V0
F5	GND	GND
F6	VCOMPLA	VCOMPLA
F7	VCCPLA	VCCPLA
F8	VCCIB0	VCCIB0
F9	IO08NDB0V1	IO12NDB0V1
F10	IO08PDB0V1	IO12PDB0V1
F11	VCCIB0	VCCIB0
F12	VCCIB1	VCCIB1
F13	IO22NDB1V0	IO30NDB1V1
F14	IO22PDB1V0	IO30PDB1V1
F15	VCCIB1	VCCIB1
F16	NC	IO36PPB1V2
F17	NC	IO38NPB1V2
F18	GND	GND
F19	IO33NDB2V0	IO49NDB2V0
F20	IO34PDB2V0	IO50PDB2V0
F21	IO34NDB2V0	IO50NDB2V0

FG484		
Pin Number	AFS600 Function	AFS1500 Function
F22	IO35PDB2V0	IO51PDB2V0
G1	IO77PDB4V0	IO115PDB4V0
G2	GND	GND
G3	IO78NDB4V0	IO116NDB4V0
G4	IO78PDB4V0	IO116PDB4V0
G5	VCCIB4	VCCIB4
G6	NC	IO117PDB4V0
G7	VCCIB4	VCCIB4
G8	GND	GND
G9	IO04NDB0V0	IO06NDB0V1
G10	IO04PDB0V0	IO06PDB0V1
G11	IO12NDB0V1	IO16NDB0V2
G12	IO12PDB0V1	IO16PDB0V2
G13	NC	IO28NDB1V1
G14	NC	IO28PDB1V1
G15	GND	GND
G16	NC	IO38PPB1V2
G17	NC	IO53PDB2V0
G18	VCCIB2	VCCIB2
G19	IO36PDB2V0	IO52PDB2V0
G20	IO36NDB2V0	IO52NDB2V0
G21	GND	GND
G22	IO35NDB2V0	IO51NDB2V0
H1	IO77NDB4V0	IO115NDB4V0
H2	IO76PDB4V0	IO113PDB4V0
H3	VCCIB4	VCCIB4
H4	IO79NDB4V0	IO114NDB4V0
H5	IO79PDB4V0	IO114PDB4V0
H6	NC	IO117NDB4V0
H7	GND	GND
H8	VCC	VCC
H9	VCCIB0	VCCIB0
H10	GND	GND
H11	VCCIB0	VCCIB0
H12	VCCIB1	VCCIB1

Revision	Changes	Page
Revision 1 (continued)	An incomplete, duplicate sentence was removed from the end of the "GNDAQ Ground (analog quiet)" pin description (SAR 38706).	2-222
	Information about configuration of unused I/Os was added to the "User Pins" section (SAR 34903).	2-224
	The following information was added to the pin description for "XTAL1 Crystal Oscillator Circuit Input" and "XTAL2 Crystal Oscillator Circuit Input" (SAR 34900): In the case where the Crystal Oscillator block is not used, the XTAL1 pin should be connected to GND and the XTAL2 pin should be left floating.	2-226
	The input resistance to ground value in Table 3-3 • Input Resistance of Analog Pads for Analog Input (direct input to ADC), was corrected from 1 MΩ (typical) to 2 kΩ (typical) (SAR 38707).	3-5
	The reference to guidelines for global spines and VersaTile rows, given in the "Global Clock Dynamic Contribution—PCLOCK" section, was corrected to the "Spine Architecture" section of the Global Resources chapter in the *Fusion FPGA Fabric User's Guide* (SAR 34740).	3-20
	Package names used in the "Pin Assignments" section were revised to match standards given in *Package Mechanical Drawings* (SAR 38711).	4-1