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Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Obsolete
Core Processor	PowerPC
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	400MHz
Co-Processors/DSP	-
RAM Controllers	-
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	-
SATA	-
USB	-
Voltage - I/O	2.5V, 3.3V
Operating Temperature	0°C ~ 105°C (TA)
Security Features	-
Package / Case	360-BCBGA, FCCBGA
Supplier Device Package	360-CBGA (25x25)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mpc755crx400te

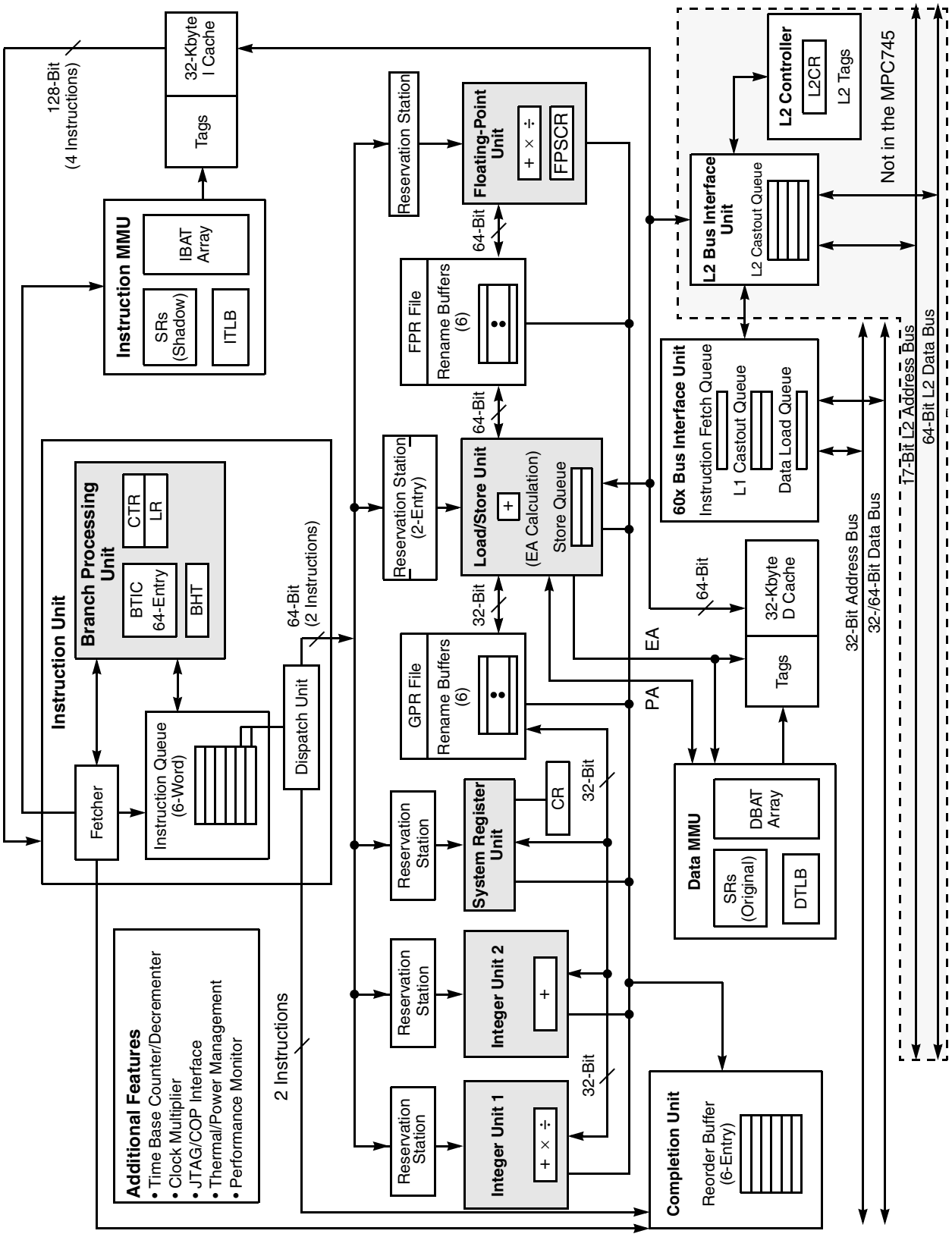


Figure 1. MPC755 Block Diagram

2 Features

This section summarizes features of the MPC755 implementation of the PowerPC architecture. Major features of the MPC755 are as follows:

- Branch processing unit
  - Four instructions fetched per clock
  - One branch processed per cycle (plus resolving two speculations)
  - Up to one speculative stream in execution, one additional speculative stream in fetch
  - 512-entry branch history table (BHT) for dynamic prediction
  - 64-entry, four-way set-associative branch target instruction cache (BTIC) for eliminating branch delay slots
- Dispatch unit
  - Full hardware detection of dependencies (resolved in the execution units)
  - Dispatch two instructions to six independent units (system, branch, load/store, fixed-point unit 1, fixed-point unit 2, floating-point)
  - Serialization control (predispatch, postdispatch, execution serialization)
- Decode
  - Register file access
  - Forwarding control
  - Partial instruction decode
- Completion
  - Six-entry completion buffer
  - Instruction tracking and peak completion of two instructions per cycle
  - Completion of instructions in program order while supporting out-of-order instruction execution, completion serialization, and all instruction flow changes
- Fixed point units (FXUs) that share 32 GPRs for integer operands
  - Fixed Point Unit 1 (FXU1)—multiply, divide, shift, rotate, arithmetic, logical
  - Fixed Point Unit 2 (FXU2)—shift, rotate, arithmetic, logical
  - Single-cycle arithmetic, shifts, rotates, logical
  - Multiply and divide support (multi-cycle)
  - Early out multiply
- Floating-point unit and a 32-entry FPR file
  - Support for IEEE standard 754 single- and double-precision floating-point arithmetic
  - Hardware support for divide
  - Hardware support for denormalized numbers
  - Single-entry reservation station
  - Supports non-IEEE mode for time-critical operations
  - Three-cycle latency, one-cycle throughput, single-precision multiply-add

  - Three-cycle latency, one-cycle throughput, double-precision add
  - Four-cycle latency, two-cycle throughput, double-precision multiply-add
- System unit
  - Executes CR logical instructions and miscellaneous system instructions
  - Special register transfer instructions
- Load/store unit
  - One-cycle load or store cache access (byte, half-word, word, double word)
  - Effective address generation
  - Hits under misses (one outstanding miss)
  - Single-cycle unaligned access within double-word boundary
  - Alignment, zero padding, sign extend for integer register file
  - Floating-point internal format conversion (alignment, normalization)
  - Sequencing for load/store multiples and string operations
  - Store gathering
  - Cache and TLB instructions
  - Big- and little-endian byte addressing supported
- Level 1 cache structure
  - 32K, 32-byte line, eight-way set-associative instruction cache (iL1)
  - 32K, 32-byte line, eight-way set-associative data cache (dL1)
  - Cache locking for both instruction and data caches, selectable by group of ways
  - Single-cycle cache access
  - Pseudo least-recently-used (PLRU) replacement
  - Copy-back or write-through data cache (on a page per page basis)
  - MEI data cache coherency maintained in hardware
  - Nonblocking instruction and data cache (one outstanding miss under hits)
  - No snooping of instruction cache
- Level 2 (L2) cache interface (not implemented on MPC745)
  - Internal L2 cache controller and tags; external data SRAMs
  - 256K, 512K, and 1 Mbyte two-way set-associative L2 cache support
  - Copy-back or write-through data cache (on a page basis, or for all L2)
  - Instruction-only mode and data-only mode
  - 64-byte (256K/512K) or 128-byte (1M) sectored line size
  - Supports flow through (register-buffer) synchronous BurstRAMs, pipelined (register-register) synchronous BurstRAMs (3-1-1-1 or strobeless 4-1-1-1) and pipelined (register-register) late write synchronous BurstRAMs
  - L2 configurable to cache, private memory, or split cache/private memory
  - Core-to-L2 frequency divisors of ÷1, ÷1.5, ÷2, ÷2.5, and ÷3 supported
  - 64-bit data bus

Figure 2 shows the allowable undershoot and overshoot voltage on the MPC755.

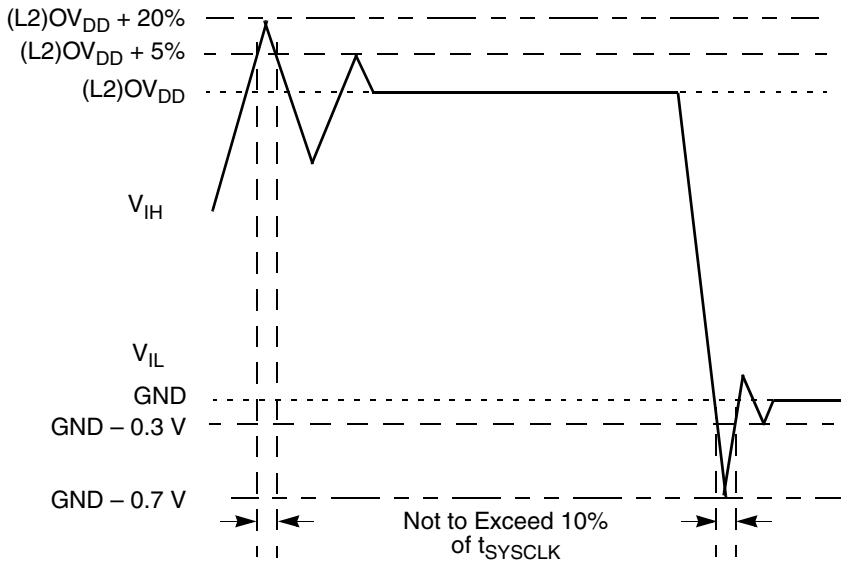


Figure 2. Overshoot/Undershoot Voltage

The MPC755 provides several I/O voltages to support both compatibility with existing systems and migration to future systems. The MPC755 core voltage must always be provided at nominal 2.0 V (see Table 3 for actual recommended core voltage). Voltage to the L2 I/Os and processor interface I/Os are provided through separate sets of supply pins and may be provided at the voltages shown in Table 2. The input voltage threshold for each bus is selected by sampling the state of the voltage select pins BVSEL and L2VSEL during operation. These signals must remain stable during part operation and cannot change. The output voltage will swing from GND to the maximum voltage applied to the OV_{DD} or $L2OV_{DD}$ power pins.

Table 2 describes the input threshold voltage setting.

Table 2. Input Threshold Voltage Setting

Part Revision	BVSEL Signal	Processor Bus Interface Voltage	L2VSEL Signal	L2 Bus Interface Voltage
E	0	Not Available	0	Not Available
	1	2.5 V/3.3 V	1	2.5 V/3.3 V

Caution: The input threshold selection must agree with the OV_{DD}/$L2OV_{DD}$ voltages supplied.

Note: The input threshold settings above are different for all revisions prior to Rev. 2.8 (Rev. E). For more information, refer to Section 10.2, "Part Numbers Not Fully Addressed by This Document."

Table 4. Package Thermal Characteristics [6]

Characteristic	Symbol	Value			Unit	Notes
		MPC755 CBGA	MPC755 PBGA	MPC745 PBGA		
Junction-to-ambient thermal resistance, natural convection	$R_{\theta JA}$	24	31	34	°C/W	1, 2
Junction-to-ambient thermal resistance, natural convection, four-layer (2s2p) board	$R_{\theta JMA}$	17	25	26	°C/W	1, 3
Junction-to-ambient thermal resistance, 200 ft/min airflow, single-layer (1s) board	$R_{\theta JMA}$	18	25	27	°C/W	1, 3
Junction-to-ambient thermal resistance, 200 ft/min airflow, four-layer (2s2p) board	$R_{\theta JMA}$	14	21	22	°C/W	1, 3
Junction-to-board thermal resistance	$R_{\theta JB}$	8	17	17	°C/W	4
Junction-to-case thermal resistance	$R_{\theta JC}$	<0.1	<0.1	<0.1	°C/W	5

Notes:

1. Junction temperature is a function of on-chip power dissipation, package thermal resistance, mounting site (board) temperature, ambient temperature, air flow, power dissipation of other components on the board, and board thermal resistance.
2. Per SEMI G38-87 and JEDEC JESD51-2 with the single layer board horizontal.
3. Per JEDEC JESD51-6 with the board horizontal.
4. Thermal resistance between the die and the printed circuit board per JEDEC JESD51-8. Board temperature is measured on the top surface of the board near the package.
5. Thermal resistance between the die and the case top surface as measured by the cold plate method (MIL SPEC-883 Method 1012.1) with the calculated case temperature. The actual value of $R_{\theta JC}$ for the part is less than 0.1°C/W.
6. Refer to Section 8.8, "Thermal Management Information," for more details about thermal management.

The MPC755 incorporates a thermal management assist unit (TAU) composed of a thermal sensor, digital-to-analog converter, comparator, control logic, and dedicated special-purpose registers (SPRs). See the *MPC750 RISC Microprocessor Family User's Manual* for more information on the use of this feature. Specifications for the thermal sensor portion of the TAU are found in Table 5.

Table 5. Thermal Sensor Specifications

At recommended operating conditions (see Table 3)

Characteristic	Min	Max	Unit	Notes
Temperature range	0	127	°C	1
Comparator settling time	20	—	µs	2, 3
Resolution	4	—	°C	3
Accuracy	–12	+12	°C	3

Notes:

1. The temperature is the junction temperature of the die. The thermal assist unit's raw output does not indicate an absolute temperature, but must be interpreted by software to derive the absolute junction temperature. For information about the use and calibration of the TAU, see Freescale Application Note AN1800/D, *Programming the Thermal Assist Unit in the MPC750 Microprocessor.*
2. The comparator settling time value must be converted into the number of CPU clocks that need to be written into the THRM3 SPR.
3. Guaranteed by design and characterization.

Table 6 provides the DC electrical characteristics for the MPC755.

Table 6. DC Electrical Specifications

At recommended operating conditions (see Table 3)

Characteristic	Nominal Bus Voltage [1]	Symbol	Min	Max	Unit	Notes
Input high voltage (all inputs except SYSCLK)	2.5	V_{IH}	1.6	(L2)OV_{DD} + 0.3	V	2, 3
	3.3	V_{IH}	2.0	(L2)OV_{DD} + 0.3	V	2, 3
Input low voltage (all inputs except SYSCLK)	2.5	V_{IL}	–0.3	0.6	V	2
	3.3	V_{IL}	–0.3	0.8	V	
SYSCLK input high voltage	2.5	KV_{IH}	1.8	OV_{DD} + 0.3	V	
	3.3	KV_{IH}	2.4	OV_{DD} + 0.3	V	
SYSCLK input low voltage	2.5	KV_{IL}	–0.3	0.4	V	
	3.3	KV_{IL}	–0.3	0.4	V	
Input leakage current, V_{in} = L2OV_{DD}/OV_{DD}		I_{in}	—	10	µA	2, 3
High-Z (off-state) leakage current, V_{in} = L2OV_{DD}/OV_{DD}		I_{TSI}	—	10	µA	2, 3, 5
Output high voltage, I_{OH} = –6 mA	2.5	V_{OH}	1.7	—	V	
	3.3	V_{OH}	2.4	—	V	
Output low voltage, I_{OL} = 6 mA	2.5	V_{OL}	—	0.45	V	
	3.3	V_{OL}	—	0.4	V	

4.2 AC Electrical Characteristics

This section provides the AC electrical characteristics for the MPC755. After fabrication, functional parts are sorted by maximum processor core frequency as shown in Section 4.2.1, "Clock AC Specifications," and tested for conformance to the AC specifications for that frequency. The processor core frequency is determined by the bus (SYSCLK) frequency and the settings of the PLL_CFG[0:3] signals. Parts are sold by maximum processor core frequency; see Section 10, "Ordering Information."

4.2.1 Clock AC Specifications

Table 8 provides the clock AC timing specifications as defined in Figure 3.

Table 8. Clock AC Timing Specifications

At recommended operating conditions (see Table 3)

Characteristic	Symbol	Maximum Processor Core Frequency						Unit	Notes
		300 MHz		350 MHz		400 MHz			
		Min	Max	Min	Max	Min	Max		
Processor frequency	f_{core}	200	300	200	350	200	400	MHz	1
VCO frequency	f_{VCO}	400	600	400	700	400	800	MHz	1
SYSCLK frequency	f_{SYSCLK}	25	100	25	100	25	100	MHz	1
SYSCLK cycle time	t_{SYSCLK}	10	40	10	40	10	40	ns	
SYSCLK rise and fall time	t_{KR}, t_{KF}	—	2.0	—	2.0	—	2.0	ns	2
	t_{KR}, t_{KF}	—	1.4	—	1.4	—	1.4	ns	2
SYSCLK duty cycle measured at $OV_{DD}/2$	t_{KHKL}/t_{SYSCLK}	40	60	40	60	40	60	%	3
SYSCLK jitter		—	±150	—	±150	—	±150	ps	3, 4
Internal PLL relock time		—	100	—	100	—	100	μs	3, 5

Notes:

1. **Caution:** The SYSCLK frequency and PLL_CFG[0:3] settings must be chosen such that the resulting SYSCLK (bus) frequency, CPU (core) frequency, and PLL (VCO) frequency do not exceed their respective maximum or minimum operating frequencies. Refer to the PLL_CFG[0:3] signal description in Section 8.1, "PLL Configuration," for valid PLL_CFG[0:3] settings.
2. Rise and fall times measurements are now specified in terms of slew rates, rather than time to account for selectable I/O bus interface levels. The minimum slew rate of 1 V/ns is equivalent to a 2 ns maximum rise/fall time measured at 0.4 and 2.4 V (OV_{DD} = 3.3 V) or a rise/fall time of 1 ns measured at 0.4 and 1.8 V (OV_{DD} = 2.5 V).
3. Timing is guaranteed by design and characterization.
4. This represents total input jitter—short term and long term combined—and is guaranteed by design.
5. Relock timing is guaranteed by design and characterization. PLL-relock time is the maximum amount of time required for PLL lock after a stable V_{DD} and SYSCLK are reached during the power-on reset sequence. This specification also applies when the PLL has been disabled and subsequently re-enabled during sleep mode. Also note that $\overline{\text{HRESET}}$ must be held asserted for a minimum of 255 bus clocks after the PLL-relock time during the power-on reset sequence.

Table 11. L2CLK Output AC Timing Specification

At recommended operating conditions (see Table 3)

Parameter	Symbol	All Speed Grades		Unit	Notes
		Min	Max		
L2CLK frequency	f_{L2CLK}	80	450	MHz	1, 4
L2CLK cycle time	t_{L2CLK}	2.5	12.5	ns	
L2CLK duty cycle	t_{CHCL}/t_{L2CLK}	45	55	%	2, 7
Internal DLL-relock time		640	—	L2CLK	3, 7
DLL capture window		0	10	ns	5, 7
L2CLK_OUT output-to-output skew	t_{L2CSKW}	—	50	ps	6, 7
L2CLK_OUT output jitter		—	±150	ps	6, 7

Notes:

1. L2CLK outputs are L2CLK_OUTA, L2CLK_OUTB, L2CLK_OUT, and L2SYNC_OUT pins. The L2CLK frequency-to-core frequency settings must be chosen such that the resulting L2CLK frequency and core frequency do not exceed their respective maximum or minimum operating frequencies. The maximum L2LCK frequency will be system dependent. L2CLK_OUTA and L2CLK_OUTB must have equal loading.
2. The nominal duty cycle of the L2CLK is 50% measured at midpoint voltage.
3. The DLL-relock time is specified in terms of L2CLK periods. The number in the table must be multiplied by the period of L2CLK to compute the actual time duration in ns. Relock timing is guaranteed by design and characterization.
4. The L2CR[L2SL] bit should be set for L2CLK frequencies less than 110 MHz. This adds more delay to each tap of the DLL.
5. Allowable skew between L2SYNC_OUT and L2SYNC_IN.
6. This output jitter number represents the maximum delay of one tap forward or one tap back from the current DLL tap as the phase comparator seeks to minimize the phase difference between L2SYNC_IN and the internal L2CLK. This number must be comprehended in the L2 timing analysis. The input jitter on SYSCLK affects L2CLK_OUT and the L2 address/data/control signals equally and, therefore, is already comprehended in the AC timing and does not have to be considered in the L2 timing analysis.
7. Guaranteed by design.

Figure 15 provides the test access port timing diagram.

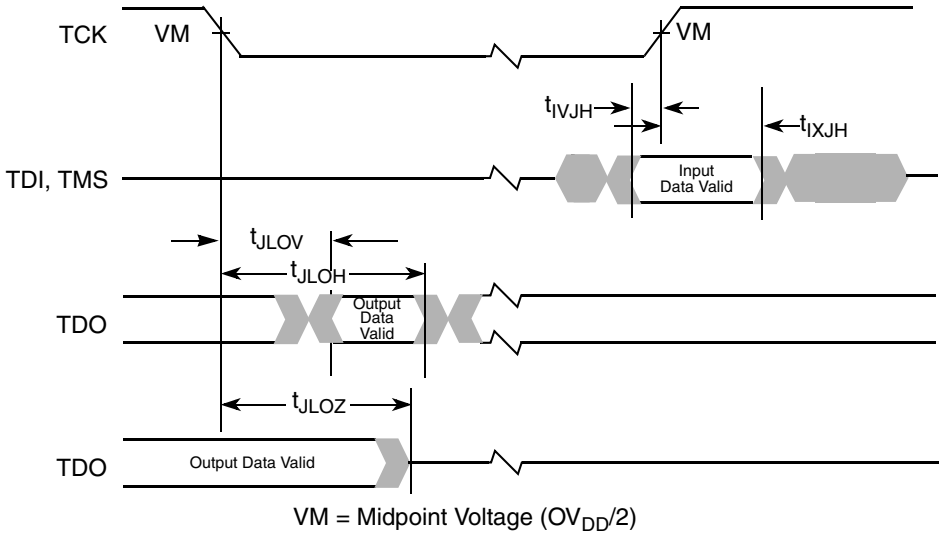


Figure 15. Test Access Port Timing Diagram

6 Pinout Listings

Table 14 provides the pinout listing for the MPC745, 255 PBGA package.

Table 14. Pinout Listing for the MPC745, 255 PBGA Package

Signal Name	Pin Number	Active	I/O	I/F Voltage [1]	Notes
A[0:31]	C16, E4, D13, F2, D14, G1, D15, E2, D16, D4, E13, G2, E15, H1, E16, H2, F13, J1, F14, J2, F15, H3, F16, F4, G13, K1, G15, K2, H16, M1, J15, P1	High	I/O	OV_{DD}	
$\overline{\text{AACK}}$	L2	Low	Input	OV_{DD}	
$\overline{\text{ABB}}$	K4	Low	I/O	OV_{DD}	
AP[0:3]	C1, B4, B3, B2	High	I/O	OV_{DD}	
$\overline{\text{ARTRY}}$	J4	Low	I/O	OV_{DD}	
AV_{DD}	A10	—	—	2.0 V	
$\overline{\text{BG}}$	L1	Low	Input	OV_{DD}	
$\overline{\text{BR}}$	B6	Low	Output	OV_{DD}	
BVSEL	B1	High	Input	OV_{DD}	3, 4, 5
$\overline{\text{CI}}$	E1	Low	Output	OV_{DD}	
$\overline{\text{CKSTP\_IN}}$	D8	Low	Input	OV_{DD}	
$\overline{\text{CKSTP\_OUT}}$	A6	Low	Output	OV_{DD}	
CLK_OUT	D7	—	Output	OV_{DD}	
$\overline{\text{DBB}}$	J14	Low	I/O	OV_{DD}	
$\overline{\text{DBG}}$	N1	Low	Input	OV_{DD}	
$\overline{\text{DBDIS}}$	H15	Low	Input	OV_{DD}	
$\overline{\text{DBWO}}$	G4	Low	Input	OV_{DD}	
DH[0:31]	P14, T16, R15, T15, R13, R12, P11, N11, R11, T12, T11, R10, P9, N9, T10, R9, T9, P8, N8, R8, T8, N7, R7, T7, P6, N6, R6, T6, R5, N5, T5, T4	High	I/O	OV_{DD}	
DL[0:31]	K13, K15, K16, L16, L15, L13, L14, M16, M15, M13, N16, N15, N13, N14, P16, P15, R16, R14, T14, N10, P13, N12, T13, P3, N3, N4, R3, T1, T2, P4, T3, R4	High	I/O	OV_{DD}	
DP[0:7]	M2, L3, N2, L4, R1, P2, M4, R2	High	I/O	OV_{DD}	
$\overline{\text{DRTRY}}$	G16	Low	Input	OV_{DD}	
$\overline{\text{GBL}}$	F1	Low	I/O	OV_{DD}	
GND	C5, C12, E3, E6, E8, E9, E11, E14, F5, F7, F10, F12, G6, G8, G9, G11, H5, H7, H10, H12, J5, J7, J10, J12, K6, K8, K9, K11, L5, L7, L10, L12, M3, M6, M8, M9, M11, M14, P5, P12	—	—	GND	
$\overline{\text{HRESET}}$	A7	Low	Input	OV_{DD}	

Table 14. Pinout Listing for the MPC745, 255 PBGA Package (continued)

Signal Name	Pin Number	Active	I/O	I/F Voltage [1]	Notes
VOLTDET	F3	High	Output	—	6

Notes:

1. OV_{DD} supplies power to the processor bus, JTAG, and all control signals; and V_{DD} supplies power to the processor core and the PLL (after filtering to become AV_{DD}). These columns serve as a reference for the nominal voltage supported on a given signal as selected by the BVSEL pin configuration of Table 2 and the voltage supplied. For actual recommended value of V_{in} or supply voltages, see Table 3.
2. These are test signals for factory use only and must be pulled up to OV_{DD} for normal machine operation.
3. This pin must be pulled up to OV_{DD} for proper operation of the processor interface. To allow for future I/O voltage changes, provide the option to connect BVSEL independently to either OV_{DD} or GND.
4. Uses 1 of 15 existing no connects in the MPC740, 255 BGA package.
5. Internal pull-up on die.
6. Internally tied to GND in the MPC745, 255 BGA package to indicate to the power supply that a low-voltage processor is present. This signal is not a power supply input.

   Caution: This differs from the MPC755, 360 BGA package.

Table 15 provides the pinout listing for the MPC755, 360 PBGA and CBGA packages.

Table 15. Pinout Listing for the MPC755, 360 BGA Package

Signal Name	Pin Number	Active	I/O	I/F Voltage [1]	Notes
A[0:31]	A13, D2, H11, C1, B13, F2, C13, E5, D13, G7, F12, G3, G6, H2, E2, L3, G5, L4, G4, J4, H7, E1, G2, F3, J7, M3, H3, J2, J6, K3, K2, L2	High	I/O	OV_{DD}	
$\overline{\text{AACK}}$	N3	Low	Input	OV_{DD}	
$\overline{\text{ABB}}$	L7	Low	I/O	OV_{DD}	
AP[0:3]	C4, C5, C6, C7	High	I/O	OV_{DD}	
$\overline{\text{ARTRY}}$	L6	Low	I/O	OV_{DD}	
AV_{DD}	A8	—	—	2.0 V	
$\overline{\text{BG}}$	H1	Low	Input	OV_{DD}	
$\overline{\text{BR}}$	E7	Low	Output	OV_{DD}	
BVSEL	W1	High	Input	OV_{DD}	3, 5, 6
$\overline{\text{CI}}$	C2	Low	Output	OV_{DD}	
$\overline{\text{CKSTP\_IN}}$	B8	Low	Input	OV_{DD}	
$\overline{\text{CKSTP\_OUT}}$	D7	Low	Output	OV_{DD}	
CLK_OUT	E3	—	Output	OV_{DD}	
$\overline{\text{DBB}}$	K5	Low	I/O	OV_{DD}	
$\overline{\text{DBDIS}}$	G1	Low	Input	OV_{DD}	
$\overline{\text{DBG}}$	K1	Low	Input	OV_{DD}	
$\overline{\text{DBWO}}$	D1	Low	Input	OV_{DD}	

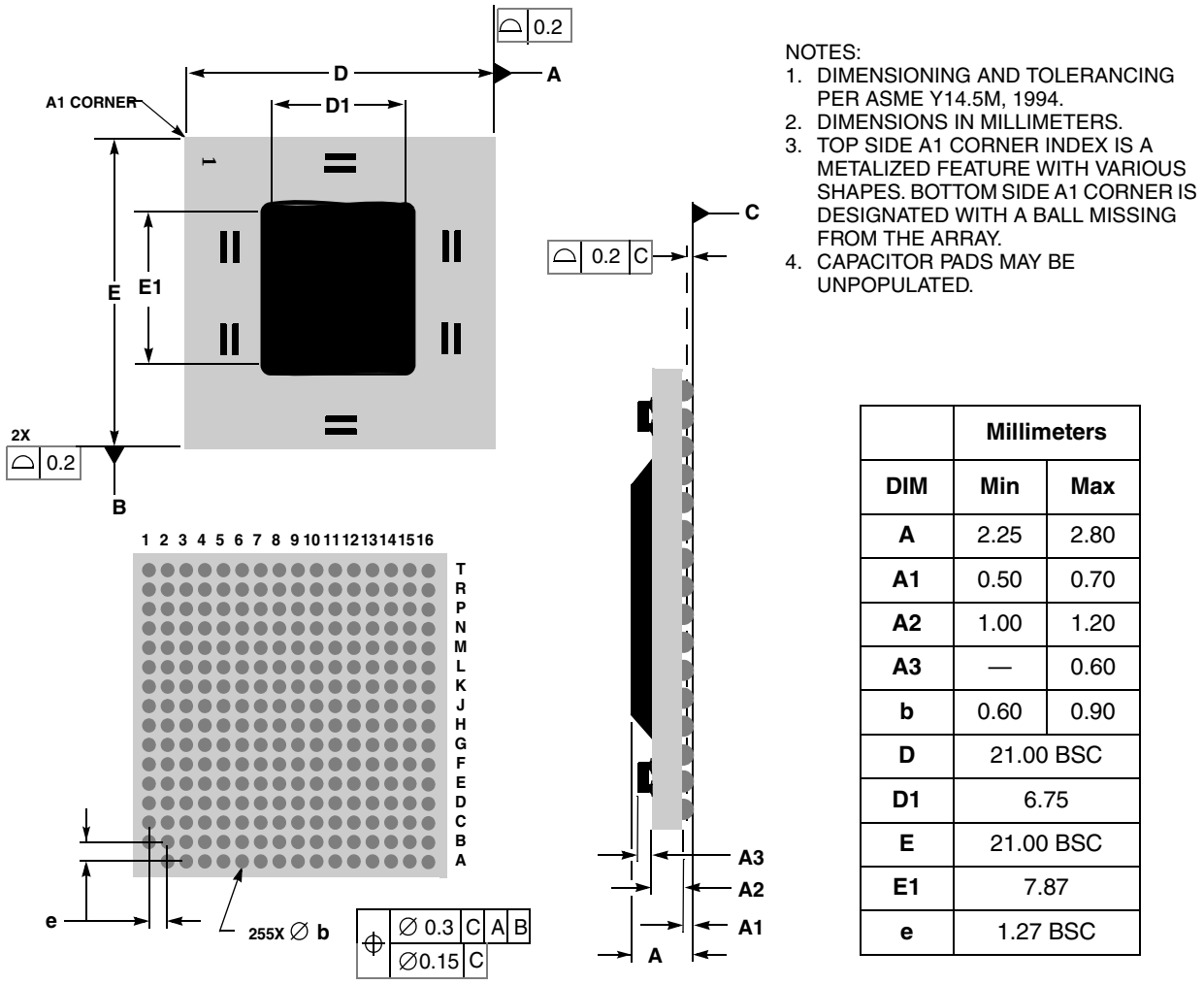
7.1 Package Parameters for the MPC745 PBGA

The package parameters are as provided in the following list. The package type is 21 × 21 mm, 255-lead plastic ball grid array (PBGA).

Package outline	21 × 21 mm
Interconnects	255 (16 × 16 ball array – 1)
Pitch	1.27 mm (50 mil)
Minimum module height	2.25 mm
Maximum module height	2.80 mm
Ball diameter (typical)	0.75 mm (29.5 mil)

7.2 Mechanical Dimensions for the MPC745 PBGA

Figure 18 provides the mechanical dimensions and bottom surface nomenclature for the MPC745, 255 PBGA package.



	Millimeters	
DIM	Min	Max
A	2.25	2.80
A1	0.50	0.70
A2	1.00	1.20
A3	—	0.60
b	0.60	0.90
D	21.00 BSC	
D1	6.75	
E	21.00 BSC	
E1	7.87	
e	1.27 BSC	

Figure 18. Mechanical Dimensions and Bottom Surface Nomenclature for the MPC745, 255 PBGA Package

Figure 22 describes the driver impedance measurement circuit described above.

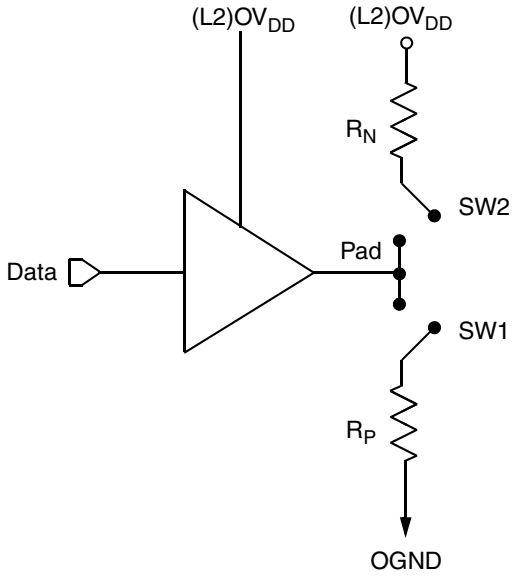


Figure 22. Driver Impedance Measurement Circuit

Alternately, the following is another method to determine the output impedance of the MPC755. A voltage source, V_{force}, is connected to the output of the MPC755 as shown in Figure 23. Data is held low, the voltage source is set to a value that is equal to (L2)OV_{DD}/2 and the current sourced by V_{force} is measured. The voltage drop across the pull-down device, which is equal to (L2)OV_{DD}/2, is divided by the measured current to determine the output impedance of the pull-down device, R_N. Similarly, the impedance of the pull-up device is determined by dividing the voltage drop of the pull-up, (L2)OV_{DD}/2, by the current sank by the pull-up when the data is high and V_{force} is equal to (L2)OV_{DD}/2. This method can be employed with either empirical data from a test setup or with data from simulation models, such as IBIS.

R_P and R_N are designed to be close to each other in value. Then $Z_0 = (R_P + R_N)/2$.

Figure 23 describes the alternate driver impedance measurement circuit.

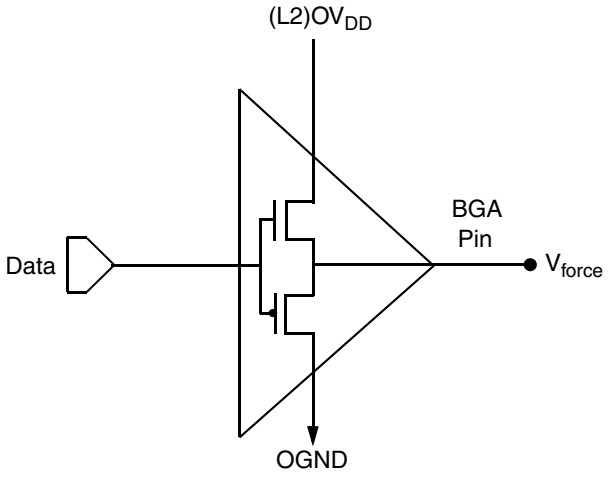


Figure 23. Alternate Driver Impedance Measurement Circuit

should be left unconnected by the system. If all parity generation is disabled through HID0, then all parity checking should also be disabled through HID0, and all parity pins may be left unconnected by the system.

The L2 interface does not require pull-up resistors.

8.7 JTAG Configuration Signals

Boundary scan testing is enabled through the JTAG interface signals. The $\overline{\text{TRST}}$ signal is optional in the IEEE 1149.1 specification, but is provided on all processors that implement the PowerPC architecture. While it is possible to force the TAP controller to the reset state using only the TCK and TMS signals, more reliable power-on reset performance will be obtained if the $\overline{\text{TRST}}$ signal is asserted during power-on reset. Because the JTAG interface is also used for accessing the common on-chip processor (COP) function, simply tying $\overline{\text{TRST}}$ to $\overline{\text{HRESET}}$ is not practical.

The COP function of these processors allows a remote computer system (typically, a PC with dedicated hardware and debugging software) to access and control the internal operations of the processor. The COP interface connects primarily through the JTAG port of the processor, with some additional status monitoring signals. The COP port requires the ability to independently assert $\overline{\text{HRESET}}$ or $\overline{\text{TRST}}$ in order to fully control the processor. If the target system has independent reset sources, such as voltage monitors, watchdog timers, power supply failures, or push-button switches, then the COP reset signals must be merged into these signals with logic.

The arrangement shown in Figure 24 allows the COP port to independently assert $\overline{\text{HRESET}}$ or $\overline{\text{TRST}}$, while ensuring that the target can drive $\overline{\text{HRESET}}$ as well. If the JTAG interface and COP header will not be used, $\overline{\text{TRST}}$ should be tied to $\overline{\text{HRESET}}$ through a 0-Ω isolation resistor so that it is asserted when the system reset signal ($\overline{\text{HRESET}}$) is asserted ensuring that the JTAG scan chain is initialized during power-on. While Freescale recommends that the COP header be designed into the system as shown in Figure 24, if this is not possible, the isolation resistor will allow future access to $\overline{\text{TRST}}$ in the case where a JTAG interface may need to be wired onto the system in debug situations.

The COP header shown in Figure 24 adds many benefits—breakpoints, watchpoints, register and memory examination/modification, and other standard debugger features are possible through this interface—and can be as inexpensive as an unpopulated footprint for a header to be added when needed.

The COP interface has a standard header for connection to the target system, based on the 0.025" square-post 0.100" centered header assembly (often called a Berg header). The connector typically has pin 14 removed as a connector key.

There is no standardized way to number the COP header shown in Figure 24; consequently, many different pin numbers have been observed from emulator vendors. Some are numbered top-to-bottom then left-to-right, while others use left-to-right then top-to-bottom, while still others number the pins counter clockwise from pin 1 (as with an IC). Regardless of the numbering, the signal placement recommended in Figure 25 is common to all known emulators.

The $\overline{\text{QACK}}$ signal shown in Figure 24 is usually connected to the PCI bridge chip in a system and is an input to the MPC755 informing it that it can go into the quiescent state. Under normal operation this occurs during a low-power mode selection. In order for COP to work, the MPC755 must see this signal asserted (pulled down). While shown on the COP header, not all emulator products drive this signal. If the product does not, a pull-down resistor can be populated to assert this signal. Additionally, some emulator products implement open-drain type outputs and can only drive $\overline{\text{QACK}}$ asserted; for these tools, a pull-up resistor can be implemented to ensure this signal is deasserted when it is not being driven by the tool. Note that the pull-up and pull-down resistors on the $\overline{\text{QACK}}$ signal are mutually exclusive and it is never necessary to populate both in a system. To preserve correct power-down operation, $\overline{\text{QACK}}$ should be merged via logic so that it also can be driven by the PCI bridge.

8.8 Thermal Management Information

This section provides thermal management information for air-cooled applications. Proper thermal control design is primarily dependent on the system-level design—the heat sink, airflow, and thermal interface material. To reduce the die-junction temperature, heat sinks may be attached to the package by several methods—adhesive, spring clip to holes in the printed-circuit board or package, and mounting clip and screw assembly; see Figure 25. This spring force should not exceed 5.5 pounds (2.5 kg) of force.

Figure 25 describes the package exploded cross-sectional view with several heat sink options.

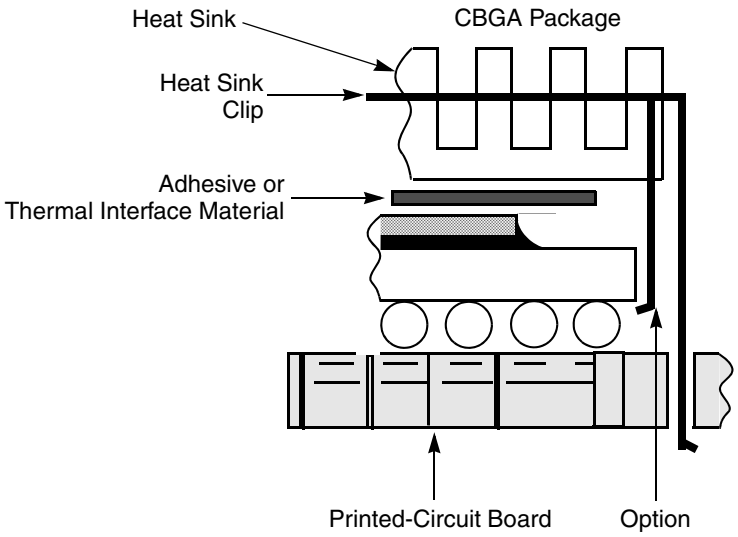


Figure 25. Package Exploded Cross-Sectional View with Several Heat Sink Options

The board designer can choose between several types of heat sinks to place on the MPC755. There are several commercially-available heat sinks for the MPC755 provided by the following vendors:

Aavid Thermalloy 80 Commercial St. Concord, NH 03301 Internet: www.aavidthermalloy.com	603-224-9988
Alpha Novatech 473 Sapena Ct. #15 Santa Clara, CA 95054 Internet: www.alphanovatech.com	408-749-7601
International Electronic Research Corporation (IERC) 413 North Moss St. Burbank, CA 91502 Internet: www.ctscorp.com	818-842-7277
Tyco Electronics Chip Coolers™ P.O. Box 3668 Harrisburg, PA 17105-3668 Internet: www.chipcoolers.com	800-522-6752
Wakefield Engineering 33 Bridge St. Pelham, NH 03076 Internet: www.wakefield.com	603-635-5102

Ultimately, the final selection of an appropriate heat sink depends on many factors, such as thermal performance at a given air velocity, spatial volume, mass, attachment method, assembly, and cost.

8.8.1 Internal Package Conduction Resistance

For the exposed-die packaging technology, shown in Table 4, the intrinsic conduction thermal resistance paths are as follows:

- The die junction-to-case (or top-of-die for exposed silicon) thermal resistance
- The die junction-to-ball thermal resistance

Figure 26 depicts the primary heat transfer path for a package with an attached heat sink mounted to a printed-circuit board.

Heat generated on the active side of the chip is conducted through the silicon, then through the heat sink attach material (or thermal interface material), and finally to the heat sink where it is removed by forced-air convection.

Since the silicon thermal resistance is quite small, for a first-order analysis, the temperature drop in the silicon may be neglected. Thus, the heat sink attach material and the heat sink conduction/convective thermal resistances are the dominant terms.

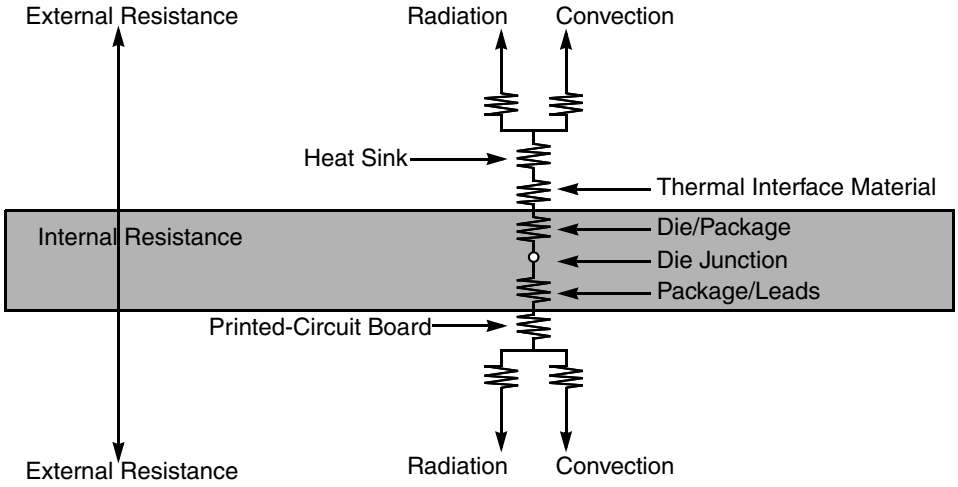


Figure 26. C4 Package with Heat Sink Mounted to a Printed-Circuit Board

8.8.2 Adhesives and Thermal Interface Materials

A thermal interface material is recommended at the package lid-to-heat sink interface to minimize the thermal contact resistance. For those applications where the heat sink is attached by spring clip mechanism, Figure 27 shows the thermal performance of three thin-sheet thermal-interface materials (silicone, graphite/oil, floroether oil), a bare joint, and a joint with thermal grease as a function of contact pressure. As shown, the performance of these thermal interface materials improves with increasing contact pressure. The use of thermal grease significantly reduces the interface thermal resistance. That is, the bare joint results in a thermal resistance approximately seven times greater than the thermal grease joint.

Heat sinks are attached to the package by means of a spring clip to holes in the printed-circuit board (see Figure 25). This spring force should not exceed 5.5 pounds of force. Therefore, the synthetic grease offers the best thermal performance, considering the low interface pressure. Of course, the selection of any thermal interface material depends on many factors—thermal performance requirements, manufacturability, service temperature, dielectric properties, cost, etc.

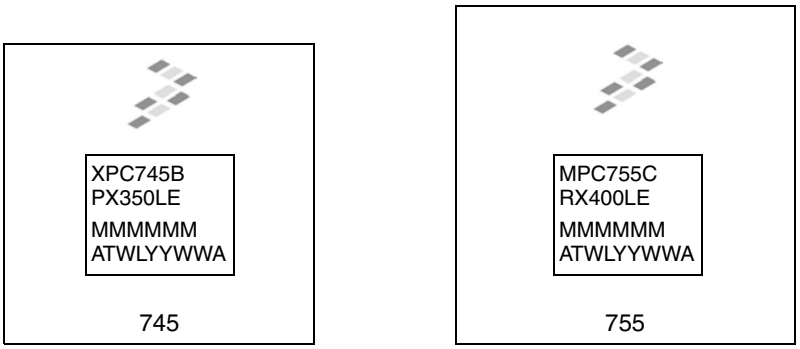
Figure 27 describes the thermal performance of select thermal interface materials.

Table 19. Document Revision History (continued)

Revision	Date	Substantive Change(s)
4	—	Added 450 MHz speed bin.
		Changed Table 16 to show 450 MHz part in example.
		Added row for 433 and 450 MHz core frequencies to Table 17.
		In Section 1.8.8, revised the heat sink vendor list.
		In Section 1.8.8.2, revised the interface vendor list.
3	—	Updated format and thermal resistance specifications of Table 4.
		Reformatted Tables 9, 10, 11, and 12.
		Added dimensions A3, D1, and E1 to Figures 18, 19, and 20.
		Revised Section 1.8.7 and Figure 25, removed Figure 26 and Table 19 (information now included in Figure 25).
		Reformatted Section 1.10.
		Clarified address bus and address attribute pull-up recommendations in Section 1.8.7.
		Clarified Table 2.
		Updated voltage sequencing requirements in Table 1 and removed Section 1.8.3.
2	—	1.8 V/2.0 V mode no longer supported; added 2.5 V support.
		Removed 1.8 V/2.0 V mode data from Tables 2, 3, and 6.
		Added 2.5 V mode data to Tables 2, 3, and 6.
		Extended recommended operating voltage (down to 1.8 V) for V_{DD}, AV_{DD}, and $L2AV_{DD}$ for 300 and 350 MHz parts in Table 3.
		Updated Table 7 and test conditions for power consumption specifications.
		Corrected Note 6 of Table 9 to include $\overline{\text{TLBISYNC}}$ as a mode-select signal.
		Updated AC timing specifications in Table 10.
		Updated AC timing specifications in Table 12.
		Corrected AC timing specifications in Table 13.
		Added L1_TSTCLK, L2_TSTCLK, and $\overline{\text{LSSD\_MODE}}$ pull-up requirements to Section 1.8.6.
		Corrected Figure 22.

10.3 Part Marking

Parts are marked as the example shown in Figure 29.



Notes:

MMMMMM is the 6-digit mask number.
ATWLYYWWA is the traceability code.
CCCCC is the country of assembly. This space is left blank if parts are assembled in the United States.

Figure 29. Part Marking for BGA Device

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