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Details

Product Status	Active
Core Processor	S08
Core Size	8-Bit
Speed	40MHz
Connectivity	I ² C, LINbus, SCI, SPI
Peripherals	LVD, POR, PWM, WDT
Number of I/O	12
Program Memory Size	16KB (16K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	1K x 8
Voltage - Supply (Vcc/Vdd)	2.7V ~ 5.5V
Data Converters	A/D 8x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 150°C (TA)
Mounting Type	Surface Mount
Package / Case	16-TSSOP (0.173", 4.40mm Width)
Supplier Device Package	16-TSSOP
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/s9s08sg16e1wtg

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Chapter 2 Pins and Connections

Typically, application systems have two separate capacitors across the power pins. In this case, there should be a bulk electrolytic capacitor, such as a $10-\mu$ F tantalum capacitor, to provide bulk charge storage for the overall system and a $0.1-\mu$ F ceramic bypass capacitor located as near to the MCU power pins as practical to suppress high-frequency noise. Each pin must have a bypass capacitor for best noise suppression.

 V_{DDA} and V_{SSA} are the analog power supply pins for MCU. This voltage source supplies power to the ADC module. A 0.1 µF ceramic bypass capacitor should be located as near to the MCU power pins as practical to suppress high-frequency noise. The V_{REFH} and V_{REFL} pins are the voltage reference high and voltage reference low inputs, respectively for the ADC module. For this MCU, V_{DDA} shares the V_{REFH} pin and these pins are available only in the 28-pin packages. In the 16-pin and 20-pin packages, they are double bonded to the V_{DD} pin. For this MCU, V_{SSA} shares the V_{REFL} pin and these pins are available only in the 20-pin packages, they are double bonded to the V_{SS} pin.

2.2.2 Oscillator (XOSC)

Immediately after reset, the MCU uses an internally generated clock provided by the clock source generator (ICS) module. For more information on the ICS, see Chapter 11, "Internal Clock Source (S08ICSV2)."

The oscillator (XOSC) in this MCU is a Pierce oscillator that can accommodate a crystal or ceramic resonator. Rather than a crystal or ceramic resonator, an external oscillator can be connected to the EXTAL input pin.

Refer to Figure 2-4 for the following discussion. R_S (when used) and R_F should be low-inductance resistors such as carbon composition resistors. Wire-wound resistors, and some metal film resistors, have too much inductance. C1 and C2 normally should be high-quality ceramic capacitors that are specifically designed for high-frequency applications.

 R_F is used to provide a bias path to keep the EXTAL input in its linear range during crystal startup; its value is not generally critical. Typical systems use 1 M Ω to 10 M Ω . Higher values are sensitive to humidity and lower values reduce gain and (in extreme cases) could prevent startup.

C1 and C2 are typically in the 5-pF to 25-pF range and are chosen to match the requirements of a specific crystal or resonator. Be sure to take into account printed circuit board (PCB) capacitance and MCU pin capacitance when selecting C1 and C2. The crystal manufacturer typically specifies a load capacitance which is the series combination of C1 and C2 (which are usually the same size). As a first-order approximation, use 10 pF as an estimate of combined pin and PCB capacitance for each oscillator pin (EXTAL and XTAL).

2.2.3 **RESET**

RESET is a dedicated pin with open-drain drive containing an internal pull-up device. Internal power-on reset and low-voltage reset circuitry typically make external reset circuitry unnecessary. This pin is normally connected to the standard 6-pin background debug connector so a development system can directly reset the MCU system. If desired, a manual external reset can be added by supplying a simple switch to ground (pull reset pin low to force a reset).



4.5.1 Features

Features of the FLASH memory include:

- FLASH size
 - MC9S08SG32: 32,768 bytes (64 pages of 512 bytes each)
 - MC9S08SG16: 16,384 bytes (32 pages of 512 bytes each)
- Single power supply program and erase
- Command interface for fast program and erase operation
- Up to 100,000 program/erase cycles at typical voltage and temperature
- Flexible block protection and vector redirection
- Security feature for FLASH and RAM
- Auto power-down for low-frequency read accesses

4.5.2 Program and Erase Times

Before any program or erase command can be accepted, the FLASH clock divider register (FCDIV) must be written to set the internal clock for the FLASH module to a frequency (f_{FCLK}) between 150 kHz and 200 kHz (see Section 4.7.1, "FLASH Clock Divider Register (FCDIV)"). This register can be written only once, so normally this write is done during reset initialization. FCDIV cannot be written if the access error flag, FACCERR in FSTAT, is set. The user must ensure that FACCERR is not set before writing to the FCDIV register. One period of the resulting clock ($1/f_{FCLK}$) is used by the command processor to time program and erase pulses. An integer number of these timing pulses are used by the command processor to complete a program or erase command.

Table 4-5 shows program and erase times. The bus clock frequency and FCDIV determine the frequency of FCLK (f_{FCLK}). The time for one cycle of FCLK is $t_{FCLK} = 1/f_{FCLK}$. The times are shown as a number of cycles of FCLK and as an absolute time for the case where $t_{FCLK} = 5 \ \mu$ s. Program and erase times shown include overhead for the command state machine and enabling and disabling of program and erase voltages.

Parameter	Cycles of FCLK	Time if FCLK = 200 kHz
Byte program	9	45 μs
Byte program (burst)	4	20 μs ¹
Page erase	4000	20 ms
Mass erase	20,000	100 ms

Table 4-5. Program and Erase Times

¹ Excluding start/end overhead

Chapter 4 Memory



4.5.4 Burst Program Execution

The burst program command is used to program sequential bytes of data in less time than would be required using the standard program command. This is possible because the high voltage to the FLASH array does not need to be disabled between program operations. Ordinarily, when a program or erase command is issued, an internal charge pump associated with the FLASH memory must be enabled to supply high voltage to the array. Upon completion of the command, the charge pump is turned off. When a burst program command is issued, the charge pump is enabled and then remains enabled after completion of the burst program operation if these two conditions are met:

- The next burst program command has been queued before the current program operation has completed.
- The next sequential address selects a byte on the same physical row as the current byte being programmed. A row of FLASH memory consists of 64 bytes. A byte within a row is selected by addresses A5 through A0. A new row begins when addresses A5 through A0 are all zero.



5.6 Low-Voltage Detect (LVD) System

The MC9S08SG32 Series includes a system to protect against low voltage conditions in order to protect memory contents and control MCU system states during supply voltage variations. The system is comprised of a power-on reset (POR) circuit and a LVD circuit with trip voltages for warning and detection. The LVD circuit is enabled when LVDE in SPMSC1 is set to 1. The LVD is disabled upon entering any of the stop modes unless LVDSE is set in SPMSC1. If LVDSE and LVDE are both set, then the MCU cannot enter stop2, and the current consumption in stop3 with the LVD enabled will be higher.

5.6.1 Power-On Reset Operation

When power is initially applied to the MCU, or when the supply voltage drops below the power-on reset rearm voltage level, V_{POR} , the POR circuit will cause a reset condition. As the supply voltage rises, the LVD circuit will hold the MCU in reset until the supply has risen above the low voltage detection low threshold, V_{LVDL} . Both the POR bit and the LVD bit in SRS are set following a POR.

5.6.2 Low-Voltage Detection (LVD) Reset Operation

The LVD can be configured to generate a reset upon detection of a low voltage condition by setting LVDRE to 1. The low voltage detection threshold is determined by the LVDV bit. After an LVD reset has occurred, the LVD system will hold the MCU in reset until the supply voltage has risen above the low voltage detection threshold. The LVD bit in the SRS register is set following either an LVD reset or POR.

5.6.3 Low-Voltage Warning (LVW) Interrupt Operation

The LVD system has a low voltage warning flag to indicate to the user that the supply voltage is approaching the low voltage condition. When a low voltage warning condition is detected and is configured for interrupt operation (LVWIE set to 1), LVWF in SPMSC1 will be set and an LVW interrupt request will occur.

5.7 Reset, Interrupt, and System Control Registers and Control Bits

One 8-bit register in the direct page register space and eight 8-bit registers in the high-page register space are related to reset and interrupt systems.

Refer to Table 4-2 and Table 4-3 in Chapter 4, "Memory," of this data sheet for the absolute address assignments for all registers. This section refers to registers and control bits only by their names. A Freescale-provided equate or header file is used to translate these names into the appropriate absolute addresses.

Some control bits in the SOPT1 and SPMSC2 registers are related to modes of operation. Although brief descriptions of these bits are provided here, the related functions are discussed in greater detail in Chapter 3, "Modes of Operation."



Chapter 5 Resets, Interrupts, and General System Control



Chapter 9 Analog-to-Digital Converter (S08ADC10V1)

ADCH	Input Select
00100	AD4
00101	AD5
00110	AD6
00111	AD7
01000	AD8
01001	AD9
01010	AD10
01011	AD11
01100	AD12
01101	AD13
01110	AD14
01111	AD15

ADCH	Input Select
10100	AD20
10101	AD21
10110	AD22
10111	AD23
11000	AD24
11001	AD25
11010	AD26
11011	AD27
11100	Reserved
11101	V _{REFH}

V_{REFL}

Module disabled

11110

11111

Table 9-4. Input Channel Select (continued)

9.3.2 Status and Control Register 2 (ADCSC2)

The ADCSC2 register controls the compare function, conversion trigger, and conversion active of the ADC module.



¹ Bits 1 and 0 are reserved bits that must always be written to 0.

Figure 9-4. Status and Control Register 2 (ADCSC2)

Table 9-5. ADCSC2 Register Field Descriptions

Field	Description
7 ADACT	 Conversion Active — Indicates that a conversion is in progress. ADACT is set when a conversion is initiated and cleared when a conversion is completed or aborted. 0 Conversion not in progress 1 Conversion in progress
6 ADTRG	 Conversion Trigger Select — Selects the type of trigger used for initiating a conversion. Two types of triggers are selectable: software trigger and hardware trigger. When software trigger is selected, a conversion is initiated following a write to ADCSC1. When hardware trigger is selected, a conversion is initiated following the assertion of the ADHWT input. 0 Software trigger selected 1 Hardware trigger selected



- Average the result by converting the analog input many times in succession and dividing the sum of the results. Four samples are required to eliminate the effect of a 1LSB, one-time error.
- Reduce the effect of synchronous noise by operating off the asynchronous clock (ADACK) and averaging. Noise that is synchronous to ADCK cannot be averaged out.

9.6.2.4 Code Width and Quantization Error

The ADC quantizes the ideal straight-line transfer function into 1024 steps (in 10-bit mode). Each step ideally has the same height (1 code) and width. The width is defined as the delta between the transition points to one code and the next. The ideal code width for an N bit converter (in this case N can be 8 or 10), defined as 1LSB, is:

$1LSB = (V_{REFH} - V_{REFL}) / 2^{N}$ Eqn. 9-2

There is an inherent quantization error due to the digitization of the result. For 8-bit or 10-bit conversions the code transitions when the voltage is at the midpoint between the points where the straight line transfer function is exactly represented by the actual transfer function. Therefore, the quantization error will be \pm 1/2LSB in 8- or 10-bit mode. As a consequence, however, the code width of the first (0x000) conversion is only 1/2LSB and the code width of the last (0xFF or 0x3FF) is 1.5LSB.

9.6.2.5 Linearity Errors

The ADC may also exhibit non-linearity of several forms. Every effort has been made to reduce these errors but the system should be aware of them because they affect overall accuracy. These errors are:

- Zero-scale error (E_{ZS}) (sometimes called offset) This error is defined as the difference between the actual code width of the first conversion and the ideal code width (1/2LSB). If the first conversion is 0x001, then the difference between the actual 0x001 code width and its ideal (1LSB) is used.
- Full-scale error (E_{FS}) This error is defined as the difference between the actual code width of the last conversion and the ideal code width (1.5LSB). If the last conversion is 0x3FE, then the difference between the actual 0x3FE code width and its ideal (1LSB) is used.
- Differential non-linearity (DNL) This error is defined as the worst-case difference between the actual code width and the ideal code width for all conversions.
- Integral non-linearity (INL) This error is defined as the highest-value the (absolute value of the) running sum of DNL achieves. More simply, this is the worst-case difference of the actual transition voltage to a given code and its corresponding ideal transition voltage, for all codes.
- Total unadjusted error (TUE) This error is defined as the difference between the actual transfer function and the ideal straight-line transfer function and includes all forms of error.

9.6.2.6 Code Jitter, Non-Monotonicity, and Missing Codes

Analog-to-digital converters are susceptible to three special forms of error. These are code jitter, non-monotonicity, and missing codes.

Code jitter is when, at certain points, a given input voltage converts to one of two values when sampled repeatedly. Ideally, when the input voltage is infinitesimally smaller than the transition voltage, the





11.3.2 ICS Control Register 2 (ICSC2)



Figure 11-4. ICS Control Register 2 (ICSC2)

Table 11-3.	ICS	Control	Register 2	2 Field	Descriptions
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Field	Description
7:6 BDIV	Bus Frequency Divider — Selects the amount to divide down the clock source selected by the CLKS bits. This controls the bus frequency. 00 Encoding 0 — Divides selected clock by 1 01 Encoding 1 — Divides selected clock by 2 (reset default) 10 Encoding 2 — Divides selected clock by 4 11 Encoding 3 — Divides selected clock by 8
5 RANGE	 Frequency Range Select — Selects the frequency range for the external oscillator. 1 High frequency range selected for the external oscillator 0 Low frequency range selected for the external oscillator
4 HGO	 High Gain Oscillator Select — The HGO bit controls the external oscillator mode of operation. Configure external oscillator for high gain operation Configure external oscillator for low power operation
3 LP	 Low Power Select — The LP bit controls whether the FLL is disabled in FLL bypassed modes. 1 FLL is disabled in bypass modes unless BDM is active 0 FLL is not disabled in bypass mode
2 EREFS	 External Reference Select — The EREFS bit selects the source for the external reference clock. 1 Oscillator requested 0 External Clock Source requested
1 ERCLKEN	 External Reference Enable — The ERCLKEN bit enables the external reference clock for use as ICSERCLK. 1 ICSERCLK active 0 ICSERCLK inactive
0 EREFSTEN	 External Reference Stop Enable — The EREFSTEN bit controls whether or not the external reference clock remains enabled when the ICS enters stop mode. 1 External reference clock stays enabled in stop if ERCLKEN is set or if ICS is in FEE, FBE, or FBELP mode before entering stop 0 External reference clock is disabled in stop



13.1.3 Block Diagram

The block diagram for the RTC module is shown in Figure 13-2.



Figure 13-2. Real-Time Counter (RTC) Block Diagram

13.2 External Signal Description

The RTC does not include any off-chip signals.

13.3 Register Definition

The RTC includes a status and control register, an 8-bit counter register, and an 8-bit modulo register.

Refer to the direct-page register summary in the memory section of this document for the absolute address assignments for all RTC registers. This section refers to registers and control bits only by their names and relative address offsets.

Table 13-1 is a summary of RTC registers.

Table 13-1	. RTC	Register	Summary
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Name		7	6	5	4	3	2	1	0
RTCSC	R	RTIF	RTC	LKS	RTIE		RT	CPS	
	W								
PTCONT	R				RTCC	NT			
KICCINI	W								
PTCMOD	R		BTOMOD						
		RICMOD							

Chapter 13 Real-Time	Counter (S08RTC)	(1)				
Internal 1-kHz Clock Source						
RTC Clock (RTCPS = 0xA)						
RTCCNT	0x52	0x53	0x54	0x55	0x00	0x01
RTIF						
RTCMOD			0x	55		

Figure 13-6. RTC Counter Overflow Example

In the example of Figure 13-6, the selected clock source is the 1-kHz internal oscillator clock source. The prescaler (RTCPS) is set to 0xA or divide-by-4. The modulo value in the RTCMOD register is set to 0x55. When the counter, RTCCNT, reaches the modulo value of 0x55, the counter overflows to 0x00 and continues counting. The real-time interrupt flag, RTIF, sets when the counter value changes from 0x55 to 0x00. A real-time interrupt is generated when RTIF is set, if RTIE is set.

13.5 Initialization/Application Information

This section provides example code to give some basic direction to a user on how to initialize and configure the RTC module. The example software is implemented in C language.

The example below shows how to implement time of day with the RTC using the 1-kHz clock source to achieve the lowest possible power consumption. Because the 1-kHz clock source is not as accurate as a crystal, software can be added for any adjustments. For accuracy without adjustments at the expense of additional power consumption, the external clock (ERCLK) or the internal clock (IRCLK) can be selected with appropriate prescaler and modulo values.

```
/* Initialize the elapsed time counters */
Seconds = 0;
Minutes = 0;
Hours = 0;
Days=0;
/* Configure RTC to interrupt every 1 second from 1-kHz clock source */
RTCMOD.byte = 0 \times 00;
RTCSC.byte = 0x1F;
Function Name : RTC_ISR
Notes : Interrupt service routine for RTC module.
                                              * * * * * * * * * * * * * * * * * /
#pragma TRAP_PROC
void RTC_ISR(void)
{
        /* Clear the interrupt flag */
```



Table 14-5. SCIS1 Field Descriptions (continued)

Field	Description
1 FE	 Framing Error Flag — FE is set at the same time as RDRF when the receiver detects a logic 0 where the stop bit was expected. This suggests the receiver was not properly aligned to a character frame. To clear FE, read SCIS1 with FE = 1 and then read the SCI data register (SCID). 0 No framing error detected. This does not guarantee the framing is correct. 1 Framing error.
0 PF	 Parity Error Flag — PF is set at the same time as RDRF when parity is enabled (PE = 1) and the parity bit in the received character does not agree with the expected parity value. To clear PF, read SCIS1 and then read the SCI data register (SCID). 0 No parity error. 1 Parity error.

14.2.5 SCI Status Register 2 (SCIS2)

This register has one read-only status flag.



Figure 14-9. SCI Status Register 2 (SCIS2)

Table 14-6. SCIS2 Field Descriptions

Field	Description
7 LBKDIF	 LIN Break Detect Interrupt Flag — LBKDIF is set when the LIN break detect circuitry is enabled and a LIN break character is detected. LBKDIF is cleared by writing a "1" to it. 0 No LIN break character has been detected. 1 LIN break character has been detected.
6 RXEDGIF	 RxD Pin Active Edge Interrupt Flag — RXEDGIF is set when an active edge (falling if RXINV = 0, rising if RXINV=1) on the RxD pin occurs. RXEDGIF is cleared by writing a "1" to it. 0 No active edge on the receive pin has occurred. 1 An active edge on the receive pin has occurred.
4 RXINV ¹	 Receive Data Inversion — Setting this bit reverses the polarity of the received data input. 0 Receive data not inverted 1 Receive data inverted
3 RWUID	 Receive Wake Up Idle Detect— RWUID controls whether the idle character that wakes up the receiver sets the IDLE bit. 0 During receive standby state (RWU = 1), the IDLE bit does not get set upon detection of an idle character. 1 During receive standby state (RWU = 1), the IDLE bit gets set upon detection of an idle character.
2 BRK13	Break Character Generation Length — BRK13 is used to select a longer transmitted break character length. Detection of a framing error is not affected by the state of this bit. 0 Break character is transmitted with length of 10 bit times (11 if M = 1) 1 Break character is transmitted with length of 13 bit times (14 if M = 1)



characters. At the end of a message, or at the beginning of the next message, all receivers automatically force RWU to 0 so all receivers wake up in time to look at the first character(s) of the next message.

14.3.3.2.1 Idle-Line Wakeup

When WAKE = 0, the receiver is configured for idle-line wakeup. In this mode, RWU is cleared automatically when the receiver detects a full character time of the idle-line level. The M control bit selects 8-bit or 9-bit data mode that determines how many bit times of idle are needed to constitute a full character time (10 or 11 bit times because of the start and stop bits).

When RWU is one and RWUID is zero, the idle condition that wakes up the receiver does not set the IDLE flag. The receiver wakes up and waits for the first data character of the next message which will set the RDRF flag and generate an interrupt if enabled. When RWUID is one, any idle condition sets the IDLE flag and generates an interrupt if enabled, regardless of whether RWU is zero or one.

The idle-line type (ILT) control bit selects one of two ways to detect an idle line. When ILT = 0, the idle bit counter starts after the start bit so the stop bit and any logic 1s at the end of a character count toward the full character time of idle. When ILT = 1, the idle bit counter does not start until after a stop bit time, so the idle detection is not affected by the data in the last character of the previous message.

14.3.3.2.2 Address-Mark Wakeup

When WAKE = 1, the receiver is configured for address-mark wakeup. In this mode, RWU is cleared automatically when the receiver detects a logic 1 in the most significant bit of a received character (eighth bit in M = 0 mode and ninth bit in M = 1 mode).

Address-mark wakeup allows messages to contain idle characters but requires that the MSB be reserved for use in address frames. The logic 1 MSB of an address frame clears the RWU bit before the stop bit is received and sets the RDRF flag. In this case the character with the MSB set is received even though the receiver was sleeping during most of this character time.

14.3.4 Interrupts and Status Flags

The SCI system has three separate interrupt vectors to reduce the amount of software needed to isolate the cause of the interrupt. One interrupt vector is associated with the transmitter for TDRE and TC events. Another interrupt vector is associated with the receiver for RDRF, IDLE, RXEDGIF and LBKDIF events, and a third vector is used for OR, NF, FE, and PF error conditions. Each of these ten interrupt sources can be separately masked by local interrupt enable masks. The flags can still be polled by software when the local masks are cleared to disable generation of hardware interrupt requests.

The SCI transmitter has two status flags that optionally can generate hardware interrupt requests. Transmit data register empty (TDRE) indicates when there is room in the transmit data buffer to write another transmit character to SCID. If the transmit interrupt enable (TIE) bit is set, a hardware interrupt will be requested whenever TDRE = 1. Transmit complete (TC) indicates that the transmitter is finished transmitting all data, preamble, and break characters and is idle with TxD at the inactive level. This flag is often used in systems with modems to determine when it is safe to turn off the modem. If the transmit complete interrupt enable (TCIE) bit is set, a hardware TC = 1.



Chapter 14 Serial Communications Interface (S08SCIV4)



Chapter 15 Serial Peripheral Interface (S08SPIV3)



NOTE

- PTC7-PTC0 and PTA7-PTA6 are not available on 16-pin Packages
- PTC7-PTC4 and PTA7-PTA6 are not available on 20-pin Packages
- + For the 16-pin and 20-pin packages: $V_{\mbox{DDA}}\!/V_{\mbox{REFH}}$ and $V_{\mbox{SSA}}\!/V_{\mbox{REFL}}$ are
- double bonded to V_{DD} and V_{SS} respectively.

 Δ = Pin can be enabled as part of the ganged output drive feature

Figure 15-1. MC9S08SG32 Series Block Diagram Highlighting SPI Block and Pin



Table 15-1. SPIC1 Field Descriptions (continued)

Field	Description
4 MSTR	Master/Slave Mode Select 0 SPI module configured as a slave SPI device 1 SPI module configured as a master SPI device
3 CPOL	 Clock Polarity — This bit effectively places an inverter in series with the clock signal from a master SPI or to a slave SPI device. Refer to Section 15.5.1, "SPI Clock Formats" for more details. 0 Active-high SPI clock (idles low) 1 Active-low SPI clock (idles high)
2 CPHA	 Clock Phase — This bit selects one of two clock formats for different kinds of synchronous serial peripheral devices. Refer to Section 15.5.1, "SPI Clock Formats" for more details. 0 First edge on SPSCK occurs at the middle of the first cycle of an 8-cycle data transfer 1 First edge on SPSCK occurs at the start of the first cycle of an 8-cycle data transfer
1 SSOE	Slave Select Output Enable — This bit is used in combination with the mode fault enable (MODFEN) bit in SPCR2 and the master/slave (MSTR) control bit to determine the function of the SS pin as shown in Table 15-2.
0 LSBFE	 LSB First (Shifter Direction) 0 SPI serial data transfers start with most significant bit 1 SPI serial data transfers start with least significant bit

Table 15-2. SS Pin Function

MODFEN	SSOE	Master Mode	Slave Mode
0	0	General-purpose I/O (not SPI)	Slave select input
0	1	General-purpose I/O (not SPI)	Slave select input
1	0	SS input for mode fault	Slave select input
1	1	Automatic SS output	Slave select input

NOTE

Ensure that the SPI should not be disabled (SPE=0) at the same time as a bit change to the CPHA bit. These changes should be performed as separate operations or unexpected behavior may occur.

15.4.2 SPI Control Register 2 (SPIC2)

This read/write register is used to control optional features of the SPI system. Bits 7, 6, 5, and 2 are not implemented and always read 0.





CLKSB:CLKSA	TPM Clock Source to Prescaler Input
00	No clock selected (TPM counter disabled)
01	Bus rate clock
10	Fixed system clock
11	External source

Table 10-0. I FW Clock Source Selection	Table	16-8.	TPM	Clock	Source	Selection
---	-------	-------	-----	-------	--------	-----------

The bus rate clock is the main system bus clock for the MCU. This clock source requires no synchronization because it is the clock that is used for all internal MCU activities including operation of the CPU and buses.

In MCUs that have no PLL and FLL or the PLL and FLL are not engaged, the fixed system clock source is the same as the bus-rate-clock source, and it does not go through a synchronizer. When a PLL or FLL is present and engaged, a synchronizer is required between the crystal divided-by two clock source and the timer counter so counter transitions will be properly aligned to bus-clock transitions. A synchronizer will be used at chip level to synchronize the crystal-related source clock to the bus clock.

The external clock source may be connected to any TPM channel pin. This clock source always has to pass through a synchronizer to assure that counter transitions are properly aligned to bus clock transitions. The bus-rate clock drives the synchronizer; therefore, to meet Nyquist criteria even with jitter, the frequency of the external clock source must not be faster than the bus rate divided-by four. With ideal clocks the external clock can be as fast as bus clock divided by four.

When the external clock source shares the TPM channel pin, this pin should not be used for other channel timing functions. For example, it would be ambiguous to configure channel 0 for input capture when the TPM channel 0 pin was also being used as the timer external clock source. (It is the user's responsibility to avoid such settings.) The TPM channel could still be used in output compare mode for software timing functions (pin controls set not to affect the TPM channel pin).

16.4.1.2 Counter Overflow and Modulo Reset

An interrupt flag and enable are associated with the 16-bit main counter. The flag (TOF) is a software-accessible indication that the timer counter has overflowed. The enable signal selects between software polling (TOIE=0) where no hardware interrupt is generated, or interrupt-driven operation (TOIE=1) where a static hardware interrupt is generated whenever the TOF flag is equal to one.

The conditions causing TOF to become set depend on whether the TPM is configured for center-aligned PWM (CPWMS=1). In the simplest mode, there is no modulus limit and the TPM is not in CPWMS=1 mode. In this case, the 16-bit timer counter counts from 0x0000 through 0xFFFF and overflows to 0x0000 on the next counting clock. TOF becomes set at the transition from 0xFFFF to 0x0000. When a modulus limit is set, TOF becomes set at the transition from the value set in the modulus register to 0x0000. When the TPM is in center-aligned PWM mode (CPWMS=1), the TOF flag gets set as the counter changes direction at the end of the count value set in the modulus register (that is, at the transition from the value set in the modulus register to the next lower count value). This corresponds to the end of a PWM period (the 0x0000 count value corresponds to the center of a period).

In output compare mode, values are transferred to the corresponding timer channel registers only after both 8-bit halves of a 16-bit register have been written and according to the value of CLKSB:CLKSA bits, so:

- If (CLKSB:CLKSA = 0:0), the registers are updated when the second byte is written
- If (CLKSB:CLKSA not = 0:0), the registers are updated at the next change of the TPM counter (end of the prescaler counting) after the second byte is written.

The coherency sequence can be manually reset by writing to the channel status/control register (TPMxCnSC).

An output compare event sets a flag bit (CHnF) which may optionally generate a CPU-interrupt request.

16.4.2.3 Edge-Aligned PWM Mode

This type of PWM output uses the normal up-counting mode of the timer counter (CPWMS=0) and can be used when other channels in the same TPM are configured for input capture or output compare functions. The period of this PWM signal is determined by the value of the modulus register (TPMxMODH:TPMxMODL) plus 1. The duty cycle is determined by the setting in the timer channel register (TPMxCnVH:TPMxCnVL). The polarity of this PWM signal is determined by the setting in the ELSnA control bit. 0% and 100% duty cycle cases are possible.

The output compare value in the TPM channel registers determines the pulse width (duty cycle) of the PWM signal (Figure 16-15). The time between the modulus overflow and the output compare is the pulse width. If ELSnA=0, the counter overflow forces the PWM signal high, and the output compare forces the PWM signal low. If ELSnA=1, the counter overflow forces the PWM signal low, and the output compare forces the PWM signal high.



Figure 16-15. PWM Period and Pulse Width (ELSnA=0)

When the channel value register is set to 0x0000, the duty cycle is 0%. 100% duty cycle can be achieved by setting the timer-channel register (TPMxCnVH:TPMxCnVL) to a value greater than the modulus setting. This implies that the modulus setting must be less than 0xFFFF in order to get 100% duty cycle.

Because the TPM may be used in an 8-bit MCU, the settings in the timer channel registers are buffered to ensure coherent 16-bit updates and to avoid unexpected PWM pulse widths. Writes to any of the registers TPMxCnVH and TPMxCnVL, actually write to buffer registers. In edge-aligned PWM mode, values are transferred to the corresponding timer-channel registers according to the value of CLKSB:CLKSA bits, so:

- If (CLKSB:CLKSA = 0:0), the registers are updated when the second byte is written
- If (CLKSB:CLKSA not = 0:0), the registers are updated after the both bytes were written, and the TPM counter changes from (TPMxMODH:TPMxMODL 1) to (TPMxMODH:TPMxMODL). If



Chapter 16 Timer/PWM Module (S08TPMV3)

EPWM mode TPMxMODH:TPMxMODL = 0x0007 TPMxCnVH:TPMxCnVL = 0x0005

	-													
RESET (active low)														
BUS CLOCK	huuu	ŲŬ		ᠾ		ΠŪ		Ц		$\prod_{i=1}^{n}$			Л	Π
TPMxCNTH:TPMxCNTL		i		0		1	2	3 4	5	6	7	0 1	2	<u> </u>
	i.	i				1			I.					
CLKSB:CLKSA BITS		I	00						1	01	1			_
-		1							1					
MSnB:MSnA BITS	00	 			10				1					
ELSnB:ELSnA BITS	00				10						 			
TPMv2 TPMxCHn									1					_
TPMv3 TPMxCHn				(—
CHnF BIT (in TPMv2 and TPMv3)														—



EPWM mode					
TPMxMODH:TPMxMODL	_ = 0x0007				
TPMxCnVH:TPMxCnVL =	= 0x0005				
RESET (active low)					
-					
BUS CLOCK				ΠΠΠ	
TPMxCNTH:TPMxCNTL		0	1 2 3 4	5 6 7	0 1 2
			1	1	1
CLKSB:CLKSA BITS		00		01	I
-				1	1
MSnB:MSnA BITS	00	10		1	l
		04		4	
ELSUB:ELSUA BITS	00	01		 	I
TPMv2 TPMxCHn				[1
•				1	ļ
TPMv3 TPMxCHn					
(in TPMv2 and TPMv3)				۱.	

Figure 16-18. Generation of low-true EPWM signal by TPM v2 and v3 after the reset

The following procedure can be used in TPM v3 (when the channel pin is also a port pin) to emulate the high-true EPWM generated by TPM v2 after the reset.



Chapter 17 Development Support

When no debugger pod is connected to the 6-pin BDM interface connector, the internal pullup on BKGD chooses normal operating mode. When a debug pod is connected to BKGD it is possible to force the MCU into active background mode after reset. The specific conditions for forcing active background depend upon the HCS08 derivative (refer to the introduction to this Development Support section). It is not necessary to reset the target MCU to communicate with it through the background debug interface.

17.2.2 Communication Details

The BDC serial interface requires the external controller to generate a falling edge on the BKGD pin to indicate the start of each bit time. The external controller provides this falling edge whether data is transmitted or received.

BKGD is a pseudo-open-drain pin that can be driven either by an external controller or by the MCU. Data is transferred MSB first at 16 BDC clock cycles per bit (nominal speed). The interface times out if 512 BDC clock cycles occur between falling edges from the host. Any BDC command that was in progress when this timeout occurs is aborted without affecting the memory or operating mode of the target MCU system.

The custom serial protocol requires the debug pod to know the target BDC communication clock speed.

The clock switch (CLKSW) control bit in the BDC status and control register allows the user to select the BDC clock source. The BDC clock source can either be the bus or the alternate BDC clock source.

The BKGD pin can receive a high or low level or transmit a high or low level. The following diagrams show timing for each of these cases. Interface timing is synchronous to clocks in the target BDC, but asynchronous to the external host. The internal BDC clock signal is shown for reference in counting cycles.