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#### Details

Product Status	Active
Core Processor	S08
Core Size	8-Bit
Speed	40MHz
Connectivity	I <sup>2</sup> C, LINbus, SCI, SPI
Peripherals	LVD, POR, PWM, WDT
Number of I/O	22
Program Memory Size	32KB (32K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	1K x 8
Voltage - Supply (Vcc/Vdd)	2.7V ~ 5.5V
Data Converters	A/D 16x10b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 105°C (TA)
Mounting Type	Surface Mount
Package / Case	28-TSSOP (0.173", 4.40mm Width)
Supplier Device Package	28-TSSOP
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/s9s08sg32e1vtlr

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**Section Number** 

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Title

Page



**Chapter 2 Pins and Connections** 







Parinharal	Mode			
Feripiteral	Stop2	Stop3		
CPU	Off	Standby		
RAM	Standby	Standby		
FLASH	Off	Standby		
Parallel Port Registers	Off	Standby		
ADC	Off	Optionally On <sup>1</sup>		
ACMP	Off	Optionally On <sup>2</sup>		
BDM	Off <sup>3</sup>	Optionally On		
ICS	Off	Optionally On <sup>4</sup>		
IIC	Off	Standby		
LVD/LVW	Off <sup>5</sup>	Optionally On		
MTIM	Off	Standby		
RTC	Optionally On	Optionally On		
SCI	Off	Standby		
SPI	Off	Standby		
ТРМ	Off	Standby		
Voltage Regulator	Standby	Optionally On <sup>6</sup>		
XOSC	Off	Optionally On <sup>7</sup>		
I/O Pins	States Held	States Held		

#### Table 3-2. Stop Mode Behavior

<sup>1</sup> Requires the asynchronous ADC clock and LVD to be enabled, else in standby.

<sup>2</sup> Requires the LVD to be enabled when compare to internal band-up reference option is enabled.

 $^3\,$  If ENBDM is set when entering stop2, the MCU will actually enter stop3.

<sup>4</sup> IRCLKEN and IREFSTEN set in ICSC1, else in standby.

- <sup>5</sup> If LVDSE is set when entering stop2, the MCU will actually enter stop3.
- <sup>6</sup> Voltage regulator will be on if BDM is enabled or if LVD is enabled when entering stop3.

<sup>7</sup> ERCLKEN and EREFSTEN set in ICSC2, else in standby. For high frequency range (RANGE in ICSC2 set) requires the LVD to also be enabled in stop3.



# 4.4 RAM

The MC9S08SG32 Series includes static RAM. The locations in RAM below 0x0100 can be accessed using the more efficient direct addressing mode, and any single bit in this area can be accessed with the bit manipulation instructions (BCLR, BSET, BRCLR, and BRSET). Locating the most frequently accessed program variables in this area of RAM is preferred.

The RAM retains data when the MCU is in low-power wait, stop2, or stop3 mode. At power-on the contents of RAM are uninitialized. RAM data is unaffected by any reset provided that the supply voltage does not drop below the minimum value for RAM retention ( $V_{RAM}$ ).

For compatibility with M68HC05 MCUs, the HCS08 resets the stack pointer to 0x00FF. In the MC9S08SG32 Series, it is usually best to reinitialize the stack pointer to the top of the RAM so the direct page RAM can be used for frequently accessed RAM variables and bit-addressable program variables. Include the following 2-instruction sequence in your reset initialization routine (where RamLast is equated to the highest address of the RAM in the Freescale Semiconductor-provided equate file).

LDHX #RamLast+1 ;point one past RAM TXS ;SP<-(H:X-1)

When security is enabled, the RAM is considered a secure memory resource and is not accessible through BDM or through code executing from non-secure memory. See Section 4.6, "Security", for a detailed description of the security feature.

# 4.5 FLASH

The FLASH memory is intended primarily for program storage. In-circuit programming allows the operating program to be loaded into the FLASH memory after final assembly of the application product. It is possible to program the entire array through the single-wire background debug interface. Because no special voltages are needed for FLASH erase and programming operations, in-application programming is also possible through other software-controlled communication paths. For a more detailed discussion of in-circuit and in-application programming, refer to the *HCS08 Family Reference Manual, Volume I,* Freescale Semiconductor document order number HCS08RMv1/D.



The COP counter is initialized by the first writes to the SOPT1 and SOPT2 registers after any system reset. Subsequent writes to SOPT1 and SOPT2 have no effect on COP operation. Even if the application will use the reset default settings of COPT, COPCLKS, and COPW bits, the user should write to the write-once SOPT1 and SOPT2 registers during reset initialization to lock in the settings. This will prevent accidental changes if the application program gets lost.

The write to SRS that services (clears) the COP counter should not be placed in an interrupt service routine (ISR) because the ISR could continue to be executed periodically even if the main application program fails.

If the bus clock source is selected, the COP counter does not increment while the MCU is in background debug mode or while the system is in stop mode. The COP counter resumes when the MCU exits background debug mode or stop mode.

If the 1-kHz clock source is selected, the COP counter is re-initialized to zero upon entry to either background debug mode or stop mode and begins from zero upon exit from background debug mode or stop mode.

# 5.5 Interrupts

Interrupts provide a way to save the current CPU status and registers, execute an interrupt service routine (ISR), and then restore the CPU status so processing resumes where it left off before the interrupt. Other than the software interrupt (SWI), which is a program instruction, interrupts are caused by hardware events such as an edge on a pin interrupt or a timer-overflow event. The debug module can also generate an SWI under certain circumstances.

If an event occurs in an enabled interrupt source, an associated read-only status flag will become set. The CPU will not respond unless the local interrupt enable is a 1 to enable the interrupt and the I bit in the CCR is 0 to allow interrupts. The global interrupt mask (I bit) in the CCR is initially set after reset which prevents all maskable interrupt sources. The user program initializes the stack pointer and performs other system setup before clearing the I bit to allow the CPU to respond to interrupts.

When the CPU receives a qualified interrupt request, it completes the current instruction before responding to the interrupt. The interrupt sequence obeys the same cycle-by-cycle sequence as the SWI instruction and consists of:

- Saving the CPU registers on the stack
- Setting the I bit in the CCR to mask further interrupts
- Fetching the interrupt vector for the highest-priority interrupt that is currently pending
- Filling the instruction queue with the first three bytes of program information starting from the address fetched from the interrupt vector locations

While the CPU is responding to the interrupt, the I bit is automatically set to avoid the possibility of another interrupt interrupting the ISR itself (this is called nesting of interrupts). Normally, the I bit is restored to 0 when the CCR is restored from the value stacked on entry to the ISR. In rare cases, the I bit can be cleared inside an ISR (after clearing the status flag that generated the interrupt) so that other interrupts can be serviced without waiting for the first service routine to finish. This practice is not recommended for anyone

#### Chapter 7 Central Processor Unit (S08CPUV3)

#### Table 7-2. Instruction Set Summary (Sheet 9 of 9)

Source		ess de		es	Cvc-by-Cvc	Affecton CCR	
Form	Operation	Addr Moe	Object Code	Cycl	Details	<b>V</b> 1 1 <b>H</b>	INZC
TXS	Transfer Index Reg. to SP SP $\leftarrow$ (H:X) – \$0001	INH	94	2	fp	-11-	
WAIT	Enable Interrupts; Wait for Interrupt I bit $\leftarrow$ 0; Halt CPU	INH	8F	2+	fp	- 1 1 -	0

Source Form: Everything in the source forms columns, except expressions in *italic characters*, is literal information which must appear in the assembly source file exactly as shown. The initial 3- to 5-letter mnemonic and the characters (#, () and +) are always a literal characters.

*n* Any label or expression that evaluates to a single integer in the range 0-7.

opr8i Any label or expression that evaluates to an 8-bit immediate value.

*opr16i* Any label or expression that evaluates to a 16-bit immediate value.

opr8a Any label or expression that evaluates to an 8-bit direct-page address (\$00xx).

opr16a Any label or expression that evaluates to a 16-bit address.

oprx8 Any label or expression that evaluates to an unsigned 8-bit value, used for indexed addressing.

oprx16 Any label or expression that evaluates to a 16-bit value, used for indexed addressing.

rel Any label or expression that refers to an address that is within -128 to +127 locations from the start of the next instruction.

#### **Operation Symbols:**

- A Accumulator
- CCR Condition code register
- H Index register high byte
- M Memory location
- n Any bit
- opr Operand (one or two bytes)
- PC Program counter
- PCH Program counter high byte
- PCL Program counter low byte
- *rel* Relative program counter offset byte
- SP Stack pointer
- SPL Stack pointer low byte
- X Index register low byte
- & Logical AND
- Logical OR
- Eugical EXCLUSIVE OR
- () Contents of
- + Add
- Subtract, Negation (two's complement)
- × Multiply
- + Divide
- # Immediate value
- ← Loaded with
- : Concatenated with

#### CCR Bits:

- V Overflow bit
- H Half-carry bit
- I Interrupt mask
- N Negative bit
- Z Zero bit
- C Carry/borrow bit

- Addressing Modes:
- DIR Direct addressing mode
- EXT Extended addressing mode
- IMM Immediate addressing mode
- INH Inherent addressing mode
- IX Indexed, no offset addressing mode
- IX1 Indexed, 8-bit offset addressing mode
- IX2 Indexed, 16-bit offset addressing mode
- IX+ Indexed, no offset, post increment addressing mode
- IX1+ Indexed, 8-bit offset, post increment addressing mode
- REL Relative addressing mode
- SP1 Stack pointer, 8-bit offset addressing mode
- SP2 Stack pointer 16-bit offset addressing mode

#### Cycle-by-Cycle Codes:

- Free cycle. This indicates a cycle where the CPU does not require use of the system buses. An f cycle is always one cycle of the system bus clock and is always a read cycle.
- p Program fetch; read from next consecutive location in program
- memory
- r Read 8-bit operand
- s Push (write) one byte onto stack
- u Pop (read) one byte from stack
- v Read vector from \$FFxx (high byte first)
- w Write 8-bit operand

#### CCR Effects:

- \$\$ Set or cleared
- Not affected
- U Undefined



are too fast, the clock must be divided to the appropriate frequency. This divider is specified by the ADIV bits and can be divide-by 1, 2, 4, or 8.

### 9.4.2 Input Select and Pin Control

The pin control registers (APCTLx) disable the digital interface to the I/O of the pins used as analog inputs. When a pin control register bit is set, the following conditions are forced for the associated MCU pin:

- The output buffer is forced to its high impedance state.
- The input buffer is disabled. A read of the I/O port returns a zero for any pin with its input buffer disabled.
- The pullup is disabled.

### 9.4.3 Hardware Trigger

The ADC module has a selectable asynchronous hardware conversion trigger, ADHWT, that is enabled when the ADTRG bit is set. This source is not available on all MCUs. Consult the module introduction for information on the ADHWT source specific to this MCU.

When ADHWT source is available and hardware trigger is enabled (ADTRG=1), a conversion is initiated on the rising edge of ADHWT. If a conversion is in progress when a rising edge occurs, the rising edge is ignored. In continuous convert configuration, only the initial rising edge to launch continuous conversions is observed. The hardware trigger function operates in conjunction with any of the conversion modes and configurations.

### 9.4.4 Conversion Control

Conversions can be performed in either 10-bit mode or 8-bit mode as determined by the MODE bits. Conversions can be initiated by either a software or hardware trigger. In addition, the ADC module can be configured for low power operation, long sample time, continuous conversion, and automatic compare of the conversion result to a software determined compare value.

### 9.4.4.1 Initiating Conversions

A conversion is initiated:

- Following a write to ADCSC1 (with ADCH bits not all 1s) if software triggered operation is selected.
- Following a hardware trigger (ADHWT) event if hardware triggered operation is selected.
- Following the transfer of the result to the data registers when continuous conversion is enabled.

If continuous conversions are enabled, a new conversion is automatically initiated after the completion of the current conversion. In software triggered operation, continuous conversions begin after ADCSC1 is written and continue until aborted. In hardware triggered operation, continuous conversions begin after a hardware trigger event and continue until aborted.



approximation algorithm is performed to determine the digital value of the analog signal. The result of the conversion is transferred to ADCRH and ADCRL upon completion of the conversion algorithm.

If the bus frequency is less than the  $f_{ADCK}$  frequency, precise sample time for continuous conversions cannot be guaranteed when short sample is enabled (ADLSMP=0). If the bus frequency is less than 1/11th of the  $f_{ADCK}$  frequency, precise sample time for continuous conversions cannot be guaranteed when long sample is enabled (ADLSMP=1).

The maximum total conversion time for different conditions is summarized in Table 9-13.

Conversion Type	ADICLK	ADLSMP	Max Total Conversion Time
Single or first continuous 8-bit	0x, 10	0	20 ADCK cycles + 5 bus clock cycles
Single or first continuous 10-bit	0x, 10	0	23 ADCK cycles + 5 bus clock cycles
Single or first continuous 8-bit	0x, 10	1	40 ADCK cycles + 5 bus clock cycles
Single or first continuous 10-bit	0x, 10	1	43 ADCK cycles + 5 bus clock cycles
Single or first continuous 8-bit	11	0	5 $\mu$ s + 20 ADCK + 5 bus clock cycles
Single or first continuous 10-bit	11	0	5 $\mu$ s + 23 ADCK + 5 bus clock cycles
Single or first continuous 8-bit	11	1	5 $\mu$ s + 40 ADCK + 5 bus clock cycles
Single or first continuous 10-bit	11	1	5 $\mu$ s + 43 ADCK + 5 bus clock cycles
Subsequent continuous 8-bit; $f_{BUS} \ge f_{ADCK}$	xx	0	17 ADCK cycles
Subsequent continuous 10-bit; $f_{BUS} \ge f_{ADCK}$	xx	0	20 ADCK cycles
Subsequent continuous 8-bit; $f_{BUS} \ge f_{ADCK}/11$	xx	1	37 ADCK cycles
Subsequent continuous 10-bit; $f_{BUS} \ge f_{ADCK}/11$	xx	1	40 ADCK cycles

Table 9-13. Total Conversion Time vs. Control Conditions

The maximum total conversion time is determined by the clock source chosen and the divide ratio selected. The clock source is selectable by the ADICLK bits, and the divide ratio is specified by the ADIV bits. For example, in 10-bit mode, with the bus clock selected as the input clock source, the input clock divide-by-1 ratio selected, and a bus frequency of 8 MHz, then the conversion time for a single conversion is:

Conversion time =  $\frac{23 \text{ ADCK cyc}}{8 \text{ MHz/1}}$  +  $\frac{5 \text{ bus cyc}}{8 \text{ MHz}}$  = 3.5 µs

Number of bus cycles =  $3.5 \ \mu s \ x \ 8 \ MHz = 28 \ cycles$ 

### NOTE

The ADCK frequency must be between  $f_{ADCK}$  minimum and  $f_{ADCK}$  maximum to meet ADC specifications.

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ICR (hex)	SCL Divider	SDA Hold Value	SCL Hold (Start) Value	SCL Hold (Stop) Value
00	20	7	6	11
01	22	7	7	12
02	24	8	8	13
03	26	8	9	14
04	28	9	10	15
05	30	9	11	16
06	34	10	13	18
07	40	10	16	21
08	28	7	10	15
09	32	7	12	17
0A	36	9	14	19
0B	40	9	16	21
0C	44	11	18	23
0D	48	11	20	25
0E	56	13	24	29
0F	68	13	30	35
10	48	9	18	25
11	56	9	22	29
12	64	13	26	33
13	72	13	30	37
14	80	17	34	41
15	88	17	38	45
16	104	21	46	53
17	128	21	58	65
18	80	9	38	41
19	96	9	46	49
1A	112	17	54	57
1B	128	17	62	65
1C	144	25	70	73
1D	160	25	78	81
1E	192	33	94	97
1F	240	33	118	121

Table 10-	5. IIC	Divider	and	Hold	Values

ICR (hex)	SCL Divider	SDA Hold Value	SCL Hold (Start) Value	SCL Hold (Stop) Value
20	160	17	78	81
21	192	17	94	97
22	224	33	110	113
23	256	33	126	129
24	288	49	142	145
25	320	49	158	161
26	384	65	190	193
27	480	65	238	241
28	320	33	158	161
29	384	33	190	193
2A	448	65	222	225
2B	512	65	254	257
2C	576	97	286	289
2D	640	97	318	321
2E	768	129	382	385
2F	960	129	478	481
30	640	65	318	321
31	768	65	382	385
32	896	129	446	449
33	1024	129	510	513
34	1152	193	574	577
35	1280	193	638	641
36	1536	257	766	769
37	1920	257	958	961
38	1280	129	638	641
39	1536	129	766	769
3A	1792	257	894	897
3B	2048	257	1022	1025
3C	2304	385	1150	1153
3D	2560	385	1278	1281
3E	3072	513	1534	1537
3F	3840	513	1918	1921

Chapter 13 Real-Time	hapter 13 Real-Time Counter (S08RTCV1)						
Internal 1-kHz Clock Source							
RTC Clock (RTCPS = 0xA)							
RTCCNT	0x52	0x53	0x54	0x55	0x00	0x01	
RTIF							
RTCMOD			0x	55			

Figure 13-6. RTC Counter Overflow Example

In the example of Figure 13-6, the selected clock source is the 1-kHz internal oscillator clock source. The prescaler (RTCPS) is set to 0xA or divide-by-4. The modulo value in the RTCMOD register is set to 0x55. When the counter, RTCCNT, reaches the modulo value of 0x55, the counter overflows to 0x00 and continues counting. The real-time interrupt flag, RTIF, sets when the counter value changes from 0x55 to 0x00. A real-time interrupt is generated when RTIF is set, if RTIE is set.

# 13.5 Initialization/Application Information

This section provides example code to give some basic direction to a user on how to initialize and configure the RTC module. The example software is implemented in C language.

The example below shows how to implement time of day with the RTC using the 1-kHz clock source to achieve the lowest possible power consumption. Because the 1-kHz clock source is not as accurate as a crystal, software can be added for any adjustments. For accuracy without adjustments at the expense of additional power consumption, the external clock (ERCLK) or the internal clock (IRCLK) can be selected with appropriate prescaler and modulo values.

```
/* Initialize the elapsed time counters */
Seconds = 0;
Minutes = 0;
Hours = 0;
Days=0;
/* Configure RTC to interrupt every 1 second from 1-kHz clock source */
RTCMOD.byte = 0 \times 00;
RTCSC.byte = 0x1F;
Function Name : RTC_ISR
Notes : Interrupt service routine for RTC module.
                                              * * * * * * * * * * * * * * * * * /
#pragma TRAP_PROC
void RTC_ISR(void)
{
        /* Clear the interrupt flag */
```

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Table 14-3. SC	CIC1 Field	Descriptions (	(continued)
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Field	Description
3 WAKE	<ul> <li>Receiver Wakeup Method Select — Refer to Section 14.3.3.2, "Receiver Wakeup Operation" for more information.</li> <li>0 Idle-line wakeup.</li> <li>1 Address-mark wakeup.</li> </ul>
2 ILT	Idle Line Type Select — Setting this bit to 1 ensures that the stop bit and logic 1 bits at the end of a character do not count toward the 10 or 11 bit times of logic high level needed by the idle line detection logic. Refer to Section 14.3.3.2.1, "Idle-Line Wakeup" for more information.         0       Idle character bit count starts after start bit.         1       Idle character bit count starts after stop bit.
1 PE	<ul> <li>Parity Enable — Enables hardware parity generation and checking. When parity is enabled, the most significant bit (MSB) of the data character (eighth or ninth data bit) is treated as the parity bit.</li> <li>0 No hardware parity generation or checking.</li> <li>1 Parity enabled.</li> </ul>
0 PT	<ul> <li>Parity Type — Provided parity is enabled (PE = 1), this bit selects even or odd parity. Odd parity means the total number of 1s in the data character, including the parity bit, is odd. Even parity means the total number of 1s in the data character, including the parity bit, is even.</li> <li>0 Even parity.</li> <li>1 Odd parity.</li> </ul>

# 14.2.3 SCI Control Register 2 (SCIC2)

This register can be read or written at any time.



### Figure 14-7. SCI Control Register 2 (SCIC2)

### Table 14-4. SCIC2 Field Descriptions

Field	Description		
7 TIE	Transmit Interrupt Enable (for TDRE)00111Hardware interrupt requested when TDRE flag is 1.		
6 TCIE	<ul> <li>Transmission Complete Interrupt Enable (for TC)</li> <li>0 Hardware interrupts from TC disabled (use polling).</li> <li>1 Hardware interrupt requested when TC flag is 1.</li> </ul>		
5 RIE	Receiver Interrupt Enable (for RDRF)0Hardware interrupts from RDRF disabled (use polling).1Hardware interrupt requested when RDRF flag is 1.		
4 ILIE	Idle Line Interrupt Enable (for IDLE)00111Hardware interrupt requested when IDLE flag is 1.		



characters. At the end of a message, or at the beginning of the next message, all receivers automatically force RWU to 0 so all receivers wake up in time to look at the first character(s) of the next message.

### 14.3.3.2.1 Idle-Line Wakeup

When WAKE = 0, the receiver is configured for idle-line wakeup. In this mode, RWU is cleared automatically when the receiver detects a full character time of the idle-line level. The M control bit selects 8-bit or 9-bit data mode that determines how many bit times of idle are needed to constitute a full character time (10 or 11 bit times because of the start and stop bits).

When RWU is one and RWUID is zero, the idle condition that wakes up the receiver does not set the IDLE flag. The receiver wakes up and waits for the first data character of the next message which will set the RDRF flag and generate an interrupt if enabled. When RWUID is one, any idle condition sets the IDLE flag and generates an interrupt if enabled, regardless of whether RWU is zero or one.

The idle-line type (ILT) control bit selects one of two ways to detect an idle line. When ILT = 0, the idle bit counter starts after the start bit so the stop bit and any logic 1s at the end of a character count toward the full character time of idle. When ILT = 1, the idle bit counter does not start until after a stop bit time, so the idle detection is not affected by the data in the last character of the previous message.

### 14.3.3.2.2 Address-Mark Wakeup

When WAKE = 1, the receiver is configured for address-mark wakeup. In this mode, RWU is cleared automatically when the receiver detects a logic 1 in the most significant bit of a received character (eighth bit in M = 0 mode and ninth bit in M = 1 mode).

Address-mark wakeup allows messages to contain idle characters but requires that the MSB be reserved for use in address frames. The logic 1 MSB of an address frame clears the RWU bit before the stop bit is received and sets the RDRF flag. In this case the character with the MSB set is received even though the receiver was sleeping during most of this character time.

### 14.3.4 Interrupts and Status Flags

The SCI system has three separate interrupt vectors to reduce the amount of software needed to isolate the cause of the interrupt. One interrupt vector is associated with the transmitter for TDRE and TC events. Another interrupt vector is associated with the receiver for RDRF, IDLE, RXEDGIF and LBKDIF events, and a third vector is used for OR, NF, FE, and PF error conditions. Each of these ten interrupt sources can be separately masked by local interrupt enable masks. The flags can still be polled by software when the local masks are cleared to disable generation of hardware interrupt requests.

The SCI transmitter has two status flags that optionally can generate hardware interrupt requests. Transmit data register empty (TDRE) indicates when there is room in the transmit data buffer to write another transmit character to SCID. If the transmit interrupt enable (TIE) bit is set, a hardware interrupt will be requested whenever TDRE = 1. Transmit complete (TC) indicates that the transmitter is finished transmitting all data, preamble, and break characters and is idle with TxD at the inactive level. This flag is often used in systems with modems to determine when it is safe to turn off the modem. If the transmit complete interrupt enable (TCIE) bit is set, a hardware TC = 1.



# 15.3 Modes of Operation

### 15.3.1 SPI in Stop Modes

The SPI is disabled in all stop modes, regardless of the settings before executing the STOP instruction. During stop2 mode, the SPI module will be fully powered down. Upon wake-up from stop2 mode, the SPI module will be in the reset state. During stop3 mode, clocks to the SPI module are halted. No registers are affected. If stop3 is exited with a reset, the SPI will be put into its reset state. If stop3 is exited with an interrupt, the SPI continues from the state it was in when stop3 was entered.

# 15.4 Register Definition

The SPI has five 8-bit registers to select SPI options, control baud rate, report SPI status, and for transmit/receive data.

Refer to the direct-page register summary in the Memory chapter of this data sheet for the absolute address assignments for all SPI registers. This section refers to registers and control bits only by their names, and a Freescale-provided equate or header file is used to translate these names into the appropriate absolute addresses.

### 15.4.1 SPI Control Register 1 (SPIC1)

This read/write register includes the SPI enable control, interrupt enables, and configuration options.



Figure 15-5. SPI Control Register 1 (SPIC1)

Table 15-1	. SPIC1	Field	Descriptions
------------	---------	-------	--------------

Field	Description
7 SPIE	<ul> <li>SPI Interrupt Enable (for SPRF and MODF) — This is the interrupt enable for SPI receive buffer full (SPRF) and mode fault (MODF) events.</li> <li>Interrupts from SPRF and MODF inhibited (use polling)</li> <li>When SPRF or MODF is 1, request a hardware interrupt</li> </ul>
6 SPE	<ul> <li>SPI System Enable — Disabling the SPI halts any transfer that is in progress, clears data buffers, and initializes internal state machines. SPRF is cleared and SPTEF is set to indicate the SPI transmit data buffer is empty.</li> <li>SPI system inactive</li> <li>SPI system enabled</li> </ul>
5 SPTIE	<ul> <li>SPI Transmit Interrupt Enable — This is the interrupt enable bit for SPI transmit buffer empty (SPTEF).</li> <li>Interrupts from SPTEF inhibited (use polling)</li> <li>When SPTEF is 1, hardware interrupt requested</li> </ul>



When the TPM is configured for center-aligned PWM (and ELSnB:ELSnA not = 0:0), the data direction for all channels in this TPM are overridden, the TPMxCHn pins are forced to be outputs controlled by the TPM, and the ELSnA bits control the polarity of each TPMxCHn output. If ELSnB:ELSnA=1:0, the corresponding TPMxCHn pin is cleared when the timer counter is counting up, and the channel value register matches the timer counter; the TPMxCHn pin is set when the timer counter is counting down, and the channel value register matches the timer counter. If ELSnA=1, the corresponding TPMxCHn pin is set when the timer counter; the TPMxCHn pin is cleared when the channel value register matches the timer counter is counting up and the channel value register matches the timer counter; the TPMxCHn pin is cleared when the timer counter is the timer counter; the timer counter is counting up and the channel value register matches the timer counter is counting the timer counter; the timer counter is counter is counter; the timer counter is counter is counter.

TPMxMODH:TPMxMODL = 0x0008 TPMxCnVH:TPMxCnVL = 0x0005





TPMxMODH:TPMxMODL = 0x0008 TPMxCnVH:TPMxCnVL = 0x0005



Figure 16-6. Low-True Pulse of a Center-Aligned PWM







In input capture mode, reading either byte (TPMxCnVH or TPMxCnVL) latches the contents of both bytes into a buffer where they remain latched until the other half is read. This latching mechanism also resets (becomes unlatched) when the TPMxCnSC register is written (whether BDM mode is active or not). Any write to the channel registers will be ignored during the input capture mode.

When BDM is active, the coherency mechanism is frozen (unless reset by writing to TPMxCnSC register) such that the buffer latches remain in the state they were in when the BDM became active, even if one or both halves of the channel register are read while BDM is active. This assures that if the user was in the middle of reading a 16-bit register when BDM became active, it will read the appropriate value from the other half of the 16-bit value after returning to normal execution. The value read from the TPMxCnVH and TPMxCnVL registers in BDM mode is the value of these registers and not the value of their read buffer.

In output compare or PWM modes, writing to either byte (TPMxCnVH or TPMxCnVL) latches the value into a buffer. After both bytes are written, they are transferred as a coherent 16-bit value into the timer-channel registers according to the value of CLKSB:CLKSA bits and the selected mode, so:

- If (CLKSB:CLKSA = 0:0), then the registers are updated when the second byte is written.
- If (CLKSB:CLKSA not = 0:0 and in output compare mode) then the registers are updated after the second byte is written and on the next change of the TPM counter (end of the prescaler counting).
- If (CLKSB:CLKSA not = 0:0 and in EPWM or CPWM modes), then the registers are updated after the both bytes were written, and the TPM counter changes from (TPMxMODH:TPMxMODL 1) to (TPMxMODH:TPMxMODL). If the TPM counter is a free-running counter then the update is made when the TPM counter changes from 0xFFFE to 0xFFFF.

The latching mechanism may be manually reset by writing to the TPMxCnSC register (whether BDM mode is active or not). This latching mechanism allows coherent 16-bit writes in either big-endian or little-endian order which is friendly to various compiler implementations.

When BDM is active, the coherency mechanism is frozen such that the buffer latches remain in the state they were in when the BDM became active even if one or both halves of the channel register are written while BDM is active. Any write to the channel registers bypasses the buffer latches and directly write to the channel register while BDM is active. The values written to the channel register while BDM is active



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are used for PWM & output compare operation once normal execution resumes. Writes to the channel registers while BDM is active do not interfere with partial completion of a coherency sequence. After the coherency mechanism has been fully exercised, the channel registers are updated using the buffered values written (while BDM was not active) by the user.

# 16.4 Functional Description

All TPM functions are associated with a central 16-bit counter which allows flexible selection of the clock source and prescale factor. There is also a 16-bit modulo register associated with the main counter.

The CPWMS control bit chooses between center-aligned PWM operation for all channels in the TPM (CPWMS=1) or general purpose timing functions (CPWMS=0) where each channel can independently be configured to operate in input capture, output compare, or edge-aligned PWM mode. The CPWMS control bit is located in the main TPM status and control register because it affects all channels within the TPM and influences the way the main counter operates. (In CPWM mode, the counter changes to an up/down mode rather than the up-counting mode used for general purpose timer functions.)

The following sections describe the main counter and each of the timer operating modes (input capture, output compare, edge-aligned PWM, and center-aligned PWM). Because details of pin operation and interrupt activity depend upon the operating mode, these topics will be covered in the associated mode explanation sections.

# 16.4.1 Counter

All timer functions are based on the main 16-bit counter (TPMxCNTH:TPMxCNTL). This section discusses selection of the clock source, end-of-count overflow, up-counting vs. up/down counting, and manual counter reset.

### 16.4.1.1 Counter Clock Source

The 2-bit field, CLKSB:CLKSA, in the timer status and control register (TPMxSC) selects one of three possible clock sources or OFF (which effectively disables the TPM). See Table 16-4. After any MCU reset, CLKSB:CLKSA=0:0 so no clock source is selected, and the TPM is in a very low power state. These control bits may be read or written at any time and disabling the timer (writing 00 to the CLKSB:CLKSA field) does not affect the values in the counter or other timer registers.



**Chapter 17 Development Support** 

### 17.4.1.1 BDC Status and Control Register (BDCSCR)

This register can be read or written by serial BDC commands (READ\_STATUS and WRITE\_CONTROL) but is not accessible to user programs because it is not located in the normal memory map of the MCU.



= Unimplemented or Reserved

### Figure 17-5. BDC Status and Control Register (BDCSCR)

### Table 17-2. BDCSCR Register Field Descriptions

Field	Description					
7 ENBDM	<ul> <li>Enable BDM (Permit Active Background Mode) — Typically, this bit is written to 1 by the debug host shortly after the beginning of a debug session or whenever the debug host resets the target and remains 1 until a normal reset clears it.</li> <li>0 BDM cannot be made active (non-intrusive commands still allowed)</li> <li>1 BDM can be made active to allow active background mode commands</li> </ul>					
6 BDMACT	<ul> <li>Background Mode Active Status — This is a read-only status bit.</li> <li>0 BDM not active (user application program running)</li> <li>1 BDM active and waiting for serial commands</li> </ul>					
5 BKPTEN	<ul> <li>BDC Breakpoint Enable — If this bit is clear, the BDC breakpoint is disabled and the FTS (force tag select) control bit and BDCBKPT match register are ignored.</li> <li>0 BDC breakpoint disabled</li> <li>1 BDC breakpoint enabled</li> </ul>					
4 FTS	<ul> <li>Force/Tag Select — When FTS = 1, a breakpoint is requested whenever the CPU address bus matches the BDCBKPT match register. When FTS = 0, a match between the CPU address bus and the BDCBKPT register causes the fetched opcode to be tagged. If this tagged opcode ever reaches the end of the instruction queue, the CPU enters active background mode rather than executing the tagged opcode.</li> <li>0 Tag opcode at breakpoint address and enter active background mode if CPU attempts to execute that instruction</li> <li>1 Breakpoint match forces active background mode at next instruction boundary (address need not be an opcode)</li> </ul>					
3 CLKSW	Select Source for BDC Communications Clock — CLKSW defaults to 0, which selects the alternate BDC clock source. 0 Alternate BDC clock source 1 MCU bus clock					



			Symbol	V <sub>DD</sub> (V)	Typ <sup>1</sup>	Max <sup>2</sup>	Unit	Temp Rated	
#	С	Parameter						Standard	AEC Grade 0
		Stop2 mode supply current			•				•
5	С	-40°C (C,M, and V suffix)		5	0.94	_	μA	•	—
	Р	25°C (All parts)			1.25	-	μA	•	_
	P <sup>5</sup>	85°C (C suffix only)			13.4	30	μA	•	_
	P <sup>5</sup>	105°C (V suffix only)			30	65	μA	•	_
	P <sup>5</sup>	125°C (M suffix only)	S2I <sub>DD</sub>		65	120	μA	•	_
	С	–40°C (C,M, and V suffix)		3	0.83	-	μA	•	_
	Р	25°C (All parts)			1.1	-	μA	•	_
	P <sup>5</sup>	85°C (C suffix only)			11.5	25	μA	•	_
	P <sup>5</sup>	105°C (V suffix only)			25	55	μA	•	_
	P <sup>5</sup>	125°C (M suffix only)			57	100	μA	•	_
6	с	RTC adder to stop2 or stop3 <sup>6</sup>	S23I <sub>DDR</sub>	5	300	500	nA	•	_
0				3	300	500	nA	•	_
7	с	LVD adder to stop3 (LVDE = LVDSE = 1)	S3I <sub>DDLVD</sub> -	5	110	180	μA	•	_
				3	90	160	μA	•	_
8	с	Adder to stop3 for oscillator enabled <sup>7</sup> (EREFSTEN =1)	S3I <sub>DDOS</sub>	5,3	5	8	μA	•	_

Table A-7.	Supply	Current	Characteristics	(continued)
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<sup>1</sup> Typical values are based on characterization data at 25°C. See Figure A-5 through Figure A-7 for typical curves across temperature and voltage.

- <sup>2</sup> Max values in this column apply for the full operating temperature range of the device unless otherwise noted.
- <sup>3</sup> All modules except ADC active, ICS configured for FBELP, and does not include any dc loads on port pins
- <sup>4</sup> All modules except ADC active, ICS configured for FEI, and does not include any dc loads on port pins
- <sup>5</sup> Stop Currents are tested in production for 25 Con all parts. Tests at other temperatures depend upon the part number suffix and maturity of the product. Freescale may eliminate a test insertion at a particular temperature from the production test flow once sufficient data has been collected and is approved.

<sup>6</sup> Most customers are expected to find that auto-wakeup from stop2 or stop3 can be used instead of the higher current wait mode.

<sup>7</sup> Values given under the following conditions: low range operation (RANGE = 0) with a 32.768kHz crystal and low power mode (HGO = 0).



### Appendix A Electrical Characteristics



Figure A-5. Typical Run  $I_{DD}$  vs. Bus Frequency ( $V_{DD}$  = 5V)



Figure A-6. Typical Run and Wait  $I_{DD}$  vs. Temperature (V<sub>DD</sub> = 5V; f<sub>bus</sub> = 8MHz)



# A.13 Flash Specifications

This section provides details about program/erase times and program-erase endurance for the Flash memory.

Program and erase operations do not require any special power sources other than the normal  $V_{DD}$  supply. For more detailed information about program/erase operations, see the Memory section.

								Temp Rated	
#	С	Characteristic	Symbol	Min	Typical	Max	Unit	Standard	AEC Grade 0
1	_	Supply voltage for program/erase	V <sub>prog/era</sub> se	2.7		5.5	V	•	•
2	_	Supply voltage for read operation	V <sub>Read</sub>	2.7	—	5.5	V	•	•
3	_	Internal FCLK frequency <sup>1</sup>	f <sub>FCLK</sub>	150	—	200	kHz	•	•
4	_	Internal FCLK period (1/f <sub>FCLK</sub> )	t <sub>Fcyc</sub>	5	—	6.67	μs	•	•
5	_	Byte program time (random location) <sup>2</sup>	t <sub>prog</sub>		9	t <sub>Fcyc</sub>	•	•	
6	_	Byte program time (burst mode) <sup>2</sup>	t <sub>Burst</sub>		4	t <sub>Fcyc</sub>	•	•	
7	_	Page erase time <sup>2</sup>	t <sub>Page</sub>		4000	t <sub>Fcyc</sub>	•	•	
8	_	Mass erase time <sup>2</sup>	t <sub>Mass</sub>		20,000	t <sub>Fcyc</sub>			
9	С	Program/erase endurance <sup>3</sup>	n <sub>FLPE</sub>				cycles		
		$T_L$ to $T_H = -40^{\circ}C$ to +125°C		10,000	-	—		•	—
		$T_L$ to $T_H = -40^{\circ}C$ to +150°C		10,000	-	—		_	•
		T = 25°C		10,000	100,000	—		•	•
10	С	Data retention <sup>4</sup>	t <sub>D_ret</sub>	15	100		years	•	•

Table A-16. Flash Characteristics

<sup>1</sup> The frequency of this clock is controlled by a software setting.

<sup>2</sup> These values are hardware state machine controlled. User code does not need to count cycles. This information supplied for calculating approximate time to program and erase.

<sup>3</sup> **Typical endurance for Flash** is based upon the intrinsic bit cell performance. For additional information on how Freescale defines typical endurance, please refer to Engineering Bulletin EB619/D, *Typical Endurance for Nonvolatile Memory*.

<sup>4</sup> Typical data retention values are based on intrinsic capability of the technology measured at high temperature and de-rated to 25°C using the Arrhenius equation. For additional information on how Freescale defines typical data retention, please refer to Engineering Bulletin EB618/D, *Typical Data Retention for Nonvolatile Memory.*