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### **Understanding Embedded - FPGAs (Field Programmable Gate Array)**

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

### **Applications of Embedded - FPGAs**

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

#### **Details**

Product Status	Active
Number of LABs/CLBs	88
Number of Logic Elements/Cells	880
Total RAM Bits	-
Number of I/O	71
Number of Gates	10000
Voltage - Supply	3V ~ 3.6V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	100-TQFP
Supplier Device Package	100-TQFP (14x14)
Purchase URL	<a href="https://www.e-xfl.com/pro/item?MUrl=&amp;PartUrl=epf6010atc100-1">https://www.e-xfl.com/pro/item?MUrl=&amp;PartUrl=epf6010atc100-1</a>

## ...and More Features

- Powerful I/O pins
  - Individual tri-state output enable control for each pin
  - Programmable output slew-rate control to reduce switching noise
  - Fast path from register to I/O pin for fast clock-to-output time
- Flexible interconnect
  - FastTrack<sup>®</sup> Interconnect continuous routing structure for fast, predictable interconnect delays
  - Dedicated carry chain that implements arithmetic functions such as fast adders, counters, and comparators (automatically used by software tools and megafunctions)
  - Dedicated cascade chain that implements high-speed, high-fan-in logic functions (automatically used by software tools and megafunctions)
  - Tri-state emulation that implements internal tri-state networks
  - Four low-skew global paths for clock, clear, preset, or logic signals
- Software design support and automatic place-and-route provided by Altera’s development system for Windows-based PCs, Sun SPARCstations, and HP 9000 Series 700/800
- Flexible package options
  - Available in a variety of packages with 100 to 256 pins, including the innovative FineLine BGA<sup>™</sup> packages (see [Table 2](#))
  - SameFrame<sup>™</sup> pin-compatibility (with other FLEX<sup>®</sup> 6000 devices) across device densities and pin counts
  - Thin quad flat pack (TQFP), plastic quad flat pack (PQFP), and ball-grid array (BGA) packages (see [Table 2](#))
  - Footprint- and pin-compatibility with other FLEX 6000 devices in the same package
- Additional design entry and simulation support provided by EDIF 2.0.0 and 3.0.0 netlist files, the library of parameterized modules (LPM), Verilog HDL, VHDL, DesignWare components, and other interfaces to popular EDA tools from manufacturers such as Cadence, Exemplar Logic, Mentor Graphics, OrCAD, Synopsys, Synplicity, VeriBest, and Viewlogic

Device	100-Pin TQFP	100-Pin FineLine BGA	144-Pin TQFP	208-Pin PQFP	240-Pin PQFP	256-Pin BGA	256-pin FineLine BGA
EPF6010A	71		102				
EPF6016			117	171	199	204	
EPF6016A	81	81	117	171			171
EPF6024A			117	171	199	218	219

Table 4 shows FLEX 6000 performance for more complex designs.

Application	LEs Used	Performance			Units
		-1 Speed Grade	-2 Speed Grade	-3 Speed Grade	
8-bit, 16-tap parallel finite impulse response (FIR) filter	599	94	80	72	MSPS
8-bit, 512-point fast Fourier transform (FFT) function	1,182	75 63	89 53	109 43	$\mu$ S MHz
16450 universal asynchronous receiver/transmitter (UART)	487	36	30	25	MHz
PCI bus target with zero wait states	609	56	49	42	MHz

**Note:**

(1) The applications in this table were created using Altera MegaCore™ functions.

FLEX 6000 devices are supported by Altera development systems; a single, integrated package that offers schematic, text (including AHDL), and waveform design entry, compilation and logic synthesis, full simulation and worst-case timing analysis, and device configuration. The Altera software provides EDIF 2.0.0 and 3.0.0, LPM, VHDL, Verilog HDL, and other interfaces for additional design entry and simulation support from other industry-standard PC- and UNIX workstation-based EDA tools.

The Altera software works easily with common gate array EDA tools for synthesis and simulation. For example, the Altera software can generate Verilog HDL files for simulation with tools such as Cadence Verilog-XL. Additionally, the Altera software contains EDA libraries that use device-specific features such as carry chains which are used for fast counter and arithmetic functions. For instance, the Synopsys Design Compiler library supplied with the Altera development systems include DesignWare functions that are optimized for the FLEX 6000 architecture.

The Altera development system runs on Windows-based PCs, Sun SPARCstations, and HP 9000 Series 700/800.

f See the *MAX+PLUS II Programmable Logic Development System & Software Data Sheet* and the *Quartus Programmable Logic Development System & Software Data Sheet* for more information.

## Functional Description

The FLEX 6000 OptiFLEX architecture consists of logic elements (LEs). Each LE includes a 4-input look-up table (LUT), which can implement any 4-input function, a register, and dedicated paths for carry and cascade chain functions. Because each LE contains a register, a design can be easily pipelined without consuming more LEs. The specified gate count for FLEX 6000 devices includes all LUTs and registers.

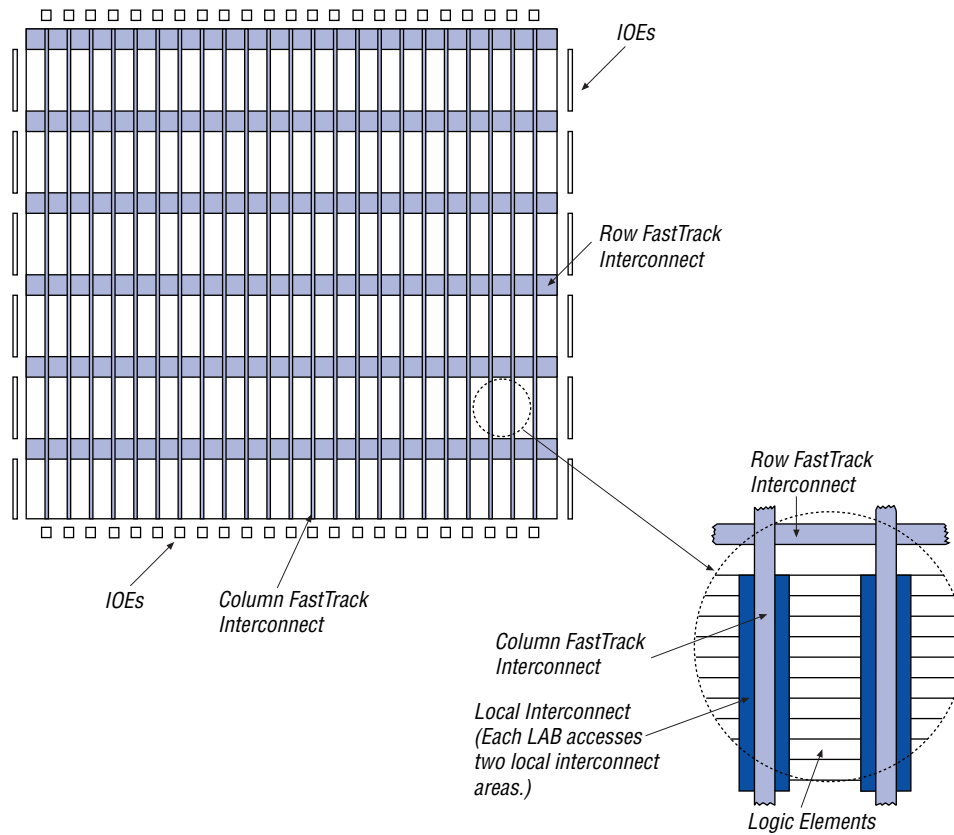
LEs are combined into groups called logic array blocks (LABs); each LAB contains 10 LEs. The Altera software automatically places related LEs into the same LAB, minimizing the number of required interconnects. Each LAB can implement a medium-sized block of logic, such as a counter or multiplexer.

Signal interconnections within FLEX 6000 devices—and to and from device pins—are provided via the routing structure of the FastTrack Interconnect. The routing structure is a series of fast, continuous row and column channels that run the entire length and width of the device. Any LE or pin can feed or be fed by any other LE or pin via the FastTrack Interconnect. See “FastTrack Interconnect” on [page 17](#) of this data sheet for more information.

Each I/O pin is fed by an I/O element (IOE) located at the end of each row and column of the FastTrack Interconnect. Each IOE contains a bidirectional I/O buffer. Each IOE is placed next to an LAB, where it can be driven by the local interconnect of that LAB. This feature allows fast clock-to-output times of less than 8 ns when a pin is driven by any of the 10 LEs in the adjacent LAB. Also, any LE can drive any pin via the row and column interconnect. I/O pins can drive the LE registers via the row and column interconnect, providing setup times as low as 2 ns and hold times of 0 ns. IOEs provide a variety of features, such as JTAG BST support, slew-rate control, and tri-state buffers.

[Figure 1](#) shows a block diagram of the FLEX 6000 OptiFLEX architecture. Each group of ten LEs is combined into an LAB, and the LABs are arranged into rows and columns. The LABs are interconnected by the FastTrack Interconnect. IOEs are located at the end of each FastTrack Interconnect row and column.

**Figure 1. OptiFLEX Architecture Block Diagram**

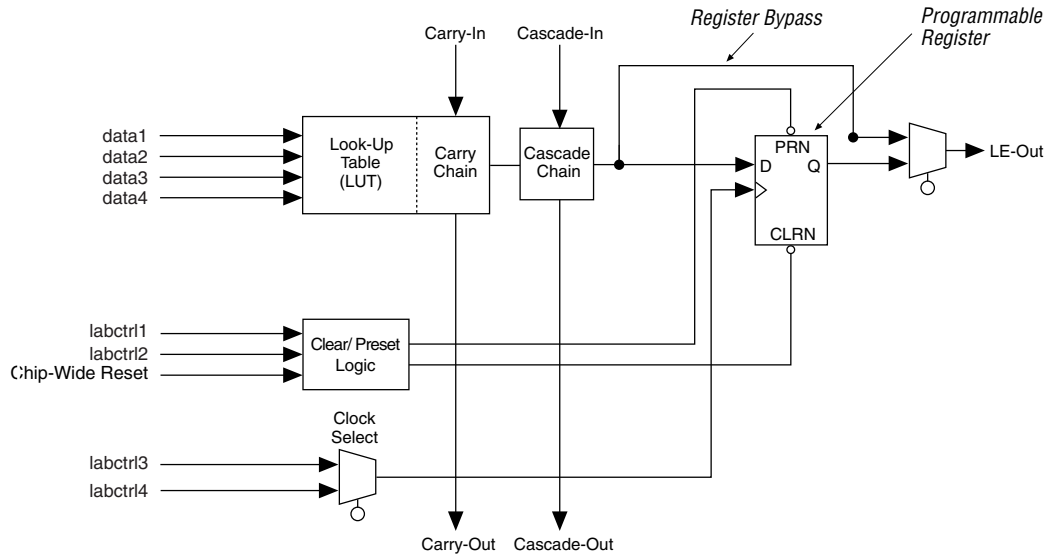


FLEX 6000 devices provide four dedicated, global inputs that drive the control inputs of the flipflops to ensure efficient distribution of high-speed, low-skew control signals. These inputs use dedicated routing channels that provide shorter delays and lower skews than the FastTrack Interconnect. These inputs can also be driven by internal logic, providing an ideal solution for a clock divider or an internally generated asynchronous clear signal that clears many registers in the device. The dedicated global routing structure is built into the device, eliminating the need to create a clock tree.

### Logic Array Block

An LAB consists of ten LEs, their associated carry and cascade chains, the LAB control signals, and the LAB local interconnect. The LAB provides the coarse-grained structure of the FLEX 6000 architecture, and facilitates efficient routing with optimum device utilization and high performance.

Figure 4. Logic Element



The programmable flipflop in the LE can be configured for D, T, JK, or SR operation. The clock and clear control signals on the flipflop can be driven by global signals, general-purpose I/O pins, or any internal logic. For combinatorial functions, the flipflop is bypassed and the output of the LUT drives the outputs of the LE. The LE output can drive both the local interconnect and the FastTrack Interconnect.

The FLEX 6000 architecture provides two types of dedicated high-speed data paths that connect adjacent LEs without using local interconnect paths: carry chains and cascade chains. A carry chain supports high-speed arithmetic functions such as counters and adders, while a cascade chain implements wide-input functions such as equivalent comparators with minimum delay. Carry and cascade chains connect LEs 2 through 10 in an LAB and all LABs in the same half of the row. Because extensive use of carry and cascade chains can reduce routing flexibility, these chains should be limited to speed-critical portions of a design.

**Normal Mode**

The normal mode is suitable for general logic applications, combinatorial functions, or wide decoding functions that can take advantage of a cascade chain. In normal mode, four data inputs from the LAB local interconnect and the carry-in are inputs to a 4-input LUT. The Altera software automatically selects the carry-in or the DATA3 signal as one of the inputs to the LUT. The LUT output can be combined with the cascade-in signal to form a cascade chain through the cascade-out signal.

**Arithmetic Mode**

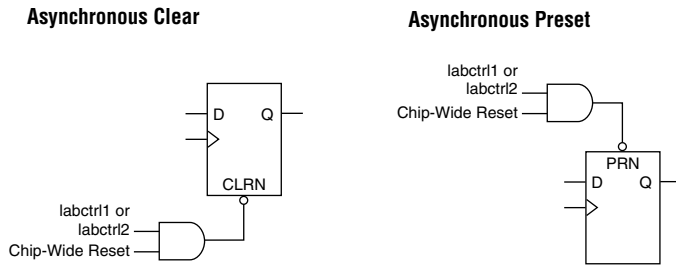
The arithmetic mode is ideal for implementing adders, accumulators, and comparators. An LE in arithmetic mode uses two 3-input LUTs. One LUT computes a 3-input function; the other generates a carry output. As shown in [Figure 7](#), the first LUT uses the carry-in signal and two data inputs from the LAB local interconnect to generate a combinatorial or registered output. For example, when implementing an adder, this output is the sum of three signals: DATA1, DATA2, and carry-in. The second LUT uses the same three signals to generate a carry-out signal, thereby creating a carry chain. The arithmetic mode also supports simultaneous use of the cascade chain.

The Altera software implements logic functions to use the arithmetic mode automatically where appropriate; the designer does not have to decide how the carry chain will be used.

**Counter Mode**

The counter mode offers counter enable, synchronous up/down control, synchronous clear, and synchronous load options. The counter enable and synchronous up/down control signals are generated from the data inputs of the LAB local interconnect. The synchronous clear and synchronous load options are LAB-wide signals that affect all registers in the LAB. Consequently, if any of the LEs in a LAB use counter mode, other LEs in that LAB must be used as part of the same counter or be used for a combinatorial function. In addition, the Altera software automatically places registers that are not in the counter into other LABs.

The counter mode uses two 3-input LUTs: one generates the counter data and the other generates the fast carry bit. A 2-to-1 multiplexer provides synchronous loading, and another AND gate provides synchronous clearing. If the cascade function is used by an LE in counter mode, the synchronous clear or load will override any signal carried on the cascade chain. The synchronous clear overrides the synchronous load.

**Figure 8. LE Clear & Preset Modes****Asynchronous Clear**

The flipflop can be cleared by either LABCTRL1 or LABCTRL2.

**Asynchronous Preset**

An asynchronous preset is implemented with an asynchronous clear. The Altera software provides preset control by using the clear and inverting the input and output of the register. Inversion control is available for the inputs to both LEs and IOEs. Therefore, this technique can be used when a register drives logic or drives a pin.

In addition to the two clear and preset modes, FLEX 6000 devices provide a chip-wide reset pin (DEV\_CLRn) that can reset all registers in the device. The option to use this pin is set in the Altera software before compilation. The chip-wide reset overrides all other control signals. Any register with an asynchronous preset will be preset when the chip-wide reset is asserted because of the inversion technique used to implement the asynchronous preset.

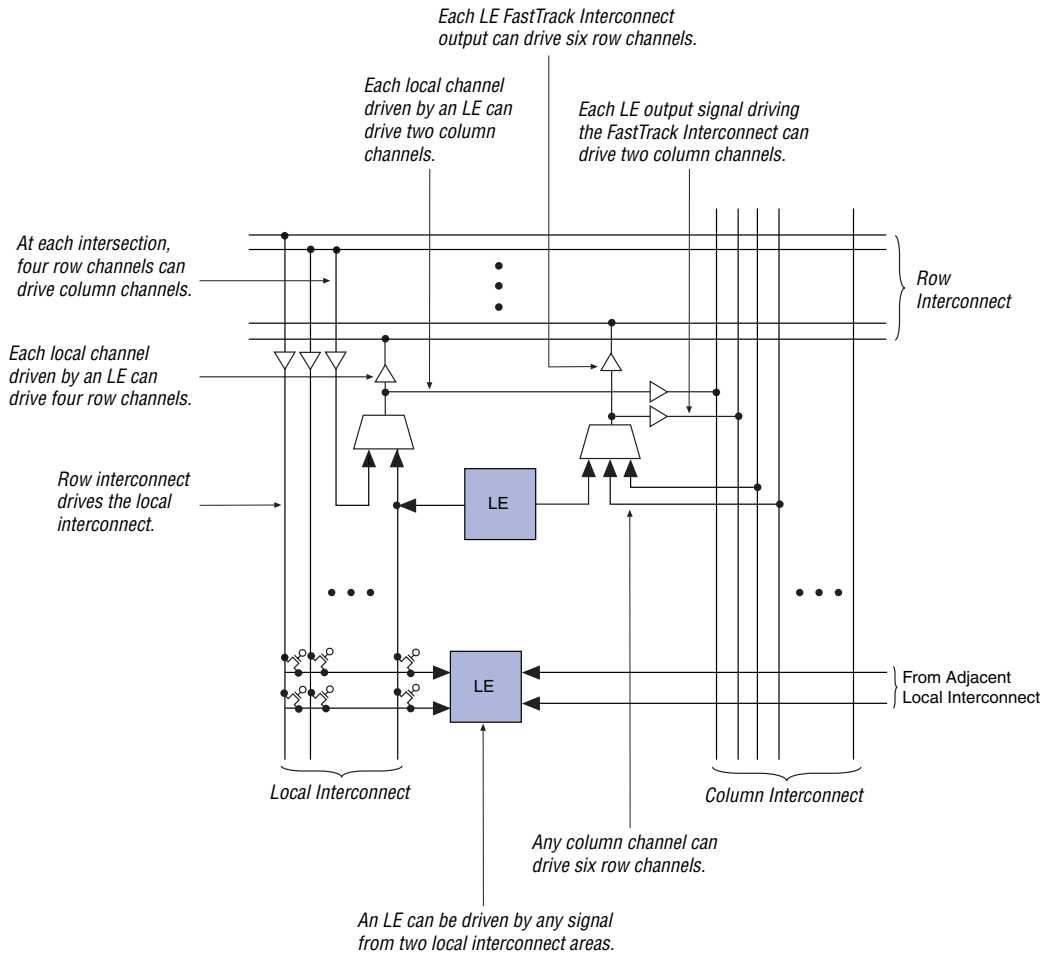
The Altera software can use a programmable NOT-gate push-back technique to emulate simultaneous preset and clear or asynchronous load. However, this technique uses an additional three LEs per register.

**FastTrack Interconnect**

In the FLEX 6000 OptiFLEX architecture, connections between LEs and device I/O pins are provided by the FastTrack Interconnect, a series of continuous horizontal and vertical routing channels that traverse the device. This global routing structure provides predictable performance, even for complex designs. In contrast, the segmented routing in FPGAs requires switch matrices to connect a variable number of routing paths, increasing the delays between logic resources and reducing performance.



**Figure 10. LAB Connections to Row & Column Interconnects**



For improved routability, the row interconnect consists of full-length and half-length channels. The full-length channels connect to all LABs in a row; the half-length channels connect to the LABs in half of the row. In addition to providing a predictable, row-wide interconnect, this architecture provides increased routing resources. Two neighboring LABs can be connected using a half-length channel, which saves the other half of the channel for the other half of the row. One-third of the row channels are half-length channels.

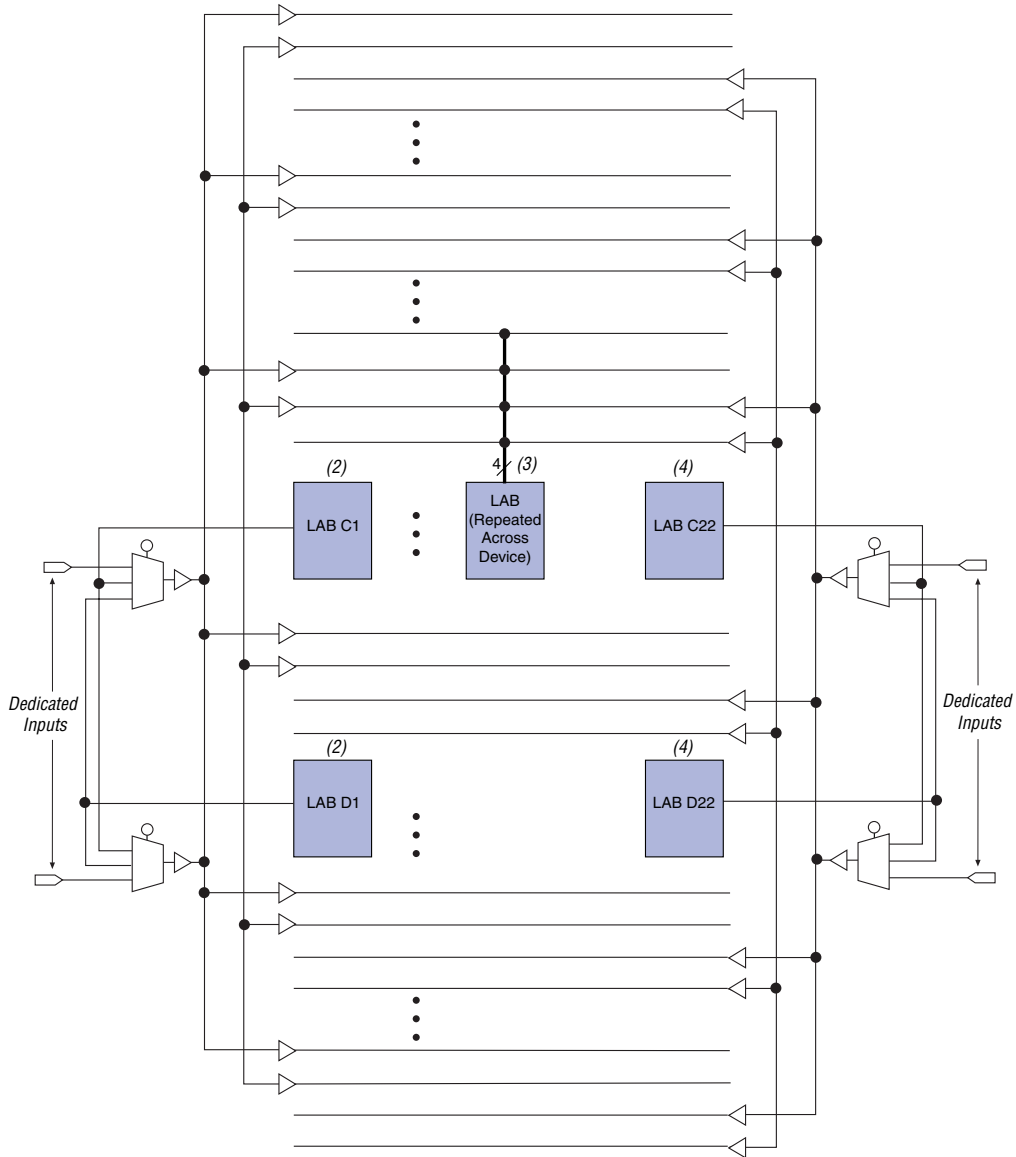
Table 5 summarizes the FastTrack Interconnect resources available in each FLEX 6000 device.

<b>Table 5. FLEX 6000 FastTrack Interconnect Resources</b>				
<b>Device</b>	<b>Rows</b>	<b>Channels per Row</b>	<b>Columns</b>	<b>Channels per Column</b>
EPF6010A	4	144	22	20
EPF6016 EPF6016A	6	144	22	20
EPF6024A	7	186	28	30

In addition to general-purpose I/O pins, FLEX 6000 devices have four dedicated input pins that provide low-skew signal distribution across the device. These four inputs can be used for global clock and asynchronous clear control signals. These signals are available as control signals for all LEs in the device. The dedicated inputs can also be used as general-purpose data inputs because they can feed the local interconnect of each LAB in the device. Using dedicated inputs to route data signals provides a fast path for high fan-out signals.

The local interconnect from LABs located at either end of two rows can drive a global control signal. For instance, in an EPF6016 device, LABs C1, D1, C22, and D22 can all drive global control signals. When an LE drives a global control signal, the dedicated input pin that drives that signal cannot be used. Any LE in the device can drive a global control signal by driving the FastTrack Interconnect into the appropriate LAB. To minimize delay, however, the Altera software places the driving LE in the appropriate LAB. The LE-driving-global signal feature is optimized for speed for control signals; regular data signals are better routed on the FastTrack Interconnect and do not receive any advantage from being routed on global signals. This LE-driving-global control signal feature is controlled by the designer and is not used automatically by the Altera software. See Figure 11.

Figure 11. Global Clock & Clear Distribution *Note (1)*

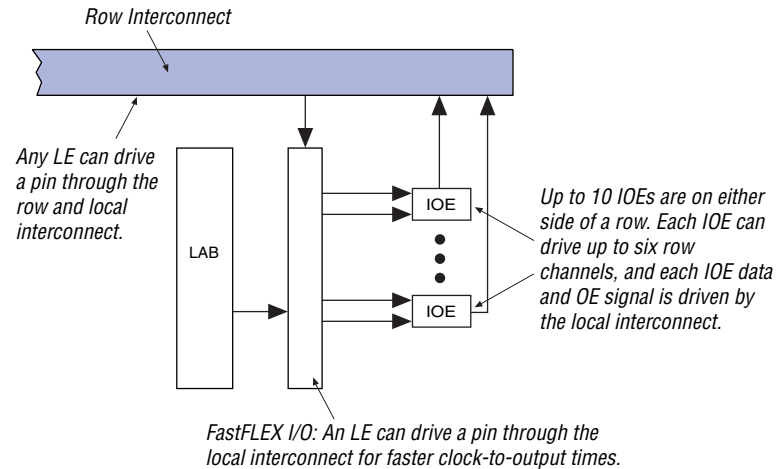


**Notes:**

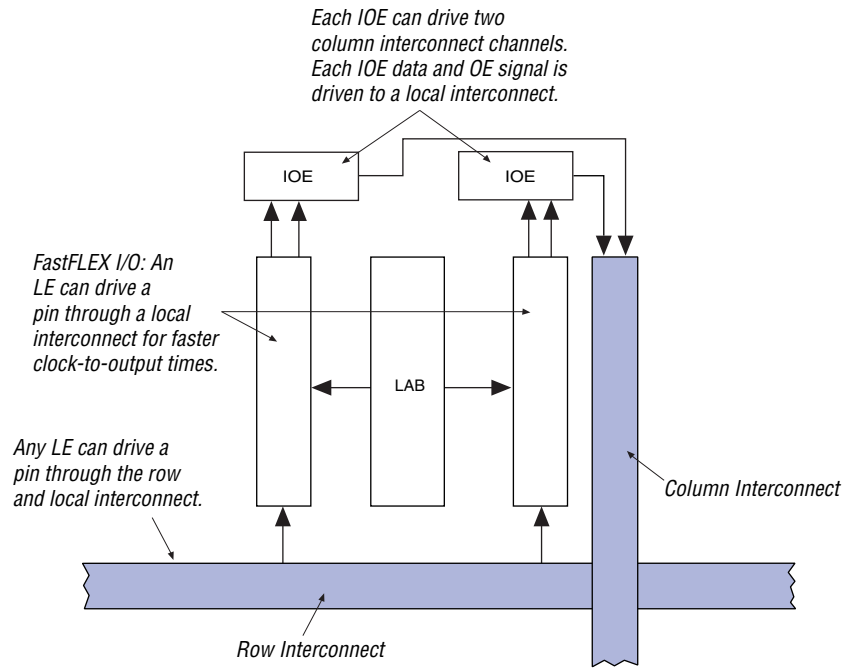
- (1) The global clock and clear distribution signals are shown for EPF6016 and EPF6016A devices. In EPF6010A devices, LABs in rows B and C drive global signals. In EPF6024A devices, LABs in rows C and E drive global signals.
- (2) The local interconnect from LABs C1 and D1 can drive two global control signals on the left side.
- (3) Global signals drive into every LAB as clock, asynchronous clear, preset, and data signals.
- (4) The local interconnect from LABs C22 and D22 can drive two global control signals on the right side.

Each IOE drives a row or column interconnect when used as an input or bidirectional pin. A row IOE can drive up to six row lines; a column IOE can drive up to two column lines. The input path from the I/O pad to the FastTrack Interconnect has a programmable delay element that can be used to guarantee a zero hold time. Depending on the placement of the IOE relative to what it is driving, the designer may choose to turn on the programmable delay to ensure a zero hold time. Figure 13 shows how an IOE connects to a row interconnect, and Figure 14 shows how an IOE connects to a column interconnect.

**Figure 13. IOE Connection to Row Interconnect**



**Figure 14. IOE Connection to Column Interconnect**



## SameFrame Pin-Outs

3.3-V FLEX 6000 devices support the SameFrame pin-out feature for FineLine BGA packages. The SameFrame pin-out feature is the arrangement of balls on FineLine BGA packages such that the lower-ball-count packages form a subset of the higher-ball-count packages. SameFrame pin-outs provide the flexibility to migrate not only from device to device within the same package, but also from one package to another. A given printed circuit board (PCB) layout can support multiple device density/package combinations. For example, a single board layout can support an EPF6016A device in a 100-pin FineLine BGA package or an EPF6024A device in a 256-pin FineLine BGA package.

The Altera software packages provide support to design PCBs with SameFrame pin-out devices. Devices can be defined for present and future use. The Altera software packages generate pin-outs describing how to lay out a board to take advantage of this migration (see [Figure 15](#)).

Open-drain output pins on 5.0-V or 3.3-V FLEX 6000 devices (with a pull-up resistor to the 5.0-V supply) can drive 5.0-V CMOS input pins that require a  $V_{IH}$  of 3.5 V. When the open-drain pin is active, it will drive low. When the pin is inactive, the trace will be pulled up to 5.0 V by the resistor. The open-drain pin will only drive low or tri-state; it will never drive high. The rise time is dependent on the value of the pull-up resistor and load impedance. The  $I_{OL}$  current specification should be considered when selecting a pull-up resistor.

Output pins on 5.0-V FLEX 6000 devices with  $V_{CCIO} = 3.3$  V or 5.0 V (with a pull-up resistor to the 5.0-V supply) can also drive 5.0-V CMOS input pins. In this case, the pull-up transistor will turn off when the pin voltage exceeds 3.3 V. Therefore, the pin does not have to be open-drain.

### Power Sequencing & Hot-Socketing

Because FLEX 6000 family devices can be used in a mixed-voltage environment, they have been designed specifically to tolerate any possible power-up sequence. The  $V_{CCIO}$  and  $V_{CCINT}$  power planes can be powered in any order.

Signals can be driven into 3.3-V FLEX 6000 devices before and during power up without damaging the device. Additionally, FLEX 6000 devices do not drive out during power up. Once operating conditions are reached, FLEX 6000 devices operate as specified by the user.

## IEEE Std. 1149.1 (JTAG) Boundary-Scan Support

All FLEX 6000 devices provide JTAG BST circuitry that comply with the IEEE Std. 1149.1-1990 specification. [Table 8](#) shows JTAG instructions for FLEX 6000 devices. JTAG BST can be performed before or after configuration, but not during configuration (except when you disable JTAG support in user mode).

- 1 See [Application Note 39 \(IEEE 1149.1 \(JTAG\) Boundary-Scan Testing in Altera Devices\)](#) for more information on JTAG BST circuitry.

JTAG Instruction	Description
SAMPLE/PRELOAD	Allows a snapshot of the signals at the device pins to be captured and examined during normal device operation, and permits an initial data pattern to be output at the device pins.
EXTEST	Allows the external circuitry and board-level interconnections to be tested by forcing a test pattern at the output pins and capturing test result at the input pins.
BYPASS	Places the 1-bit bypass register between the TDI and TDO pins, which allows the BST data to pass synchronously through the selected device to adjacent devices during normal device operation.

**Table 17. FLEX 6000 3.3-V Device DC Operating Conditions** Notes (5), (6)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V <sub>IH</sub>	High-level input voltage		1.7		5.75	V
V <sub>IL</sub>	Low-level input voltage		-0.5		0.8	V
V <sub>OH</sub>	3.3-V high-level TTL output voltage	I <sub>OH</sub> = -8 mA DC, V <sub>CCIO</sub> = 3.00 V (7)	2.4			V
	3.3-V high-level CMOS output voltage	I <sub>OH</sub> = -0.1 mA DC, V <sub>CCIO</sub> = 3.00 V (7)	V <sub>CCIO</sub> - 0.2			V
	2.5-V high-level output voltage	I <sub>OH</sub> = -100 μA DC, V <sub>CCIO</sub> = 2.30 V (7)	2.1			V
		I <sub>OH</sub> = -1 mA DC, V <sub>CCIO</sub> = 2.30 V (7)	2.0			V
		I <sub>OH</sub> = -2 mA DC, V <sub>CCIO</sub> = 2.30 V (7)	1.7			V
V <sub>OL</sub>	3.3-V low-level TTL output voltage	I <sub>OL</sub> = 8 mA DC, V <sub>CCIO</sub> = 3.00 V (8)			0.45	V
	3.3-V low-level CMOS output voltage	I <sub>OL</sub> = 0.1 mA DC, V <sub>CCIO</sub> = 3.00 V (8)			0.2	V
	2.5-V low-level output voltage	I <sub>OL</sub> = 100 μA DC, V <sub>CCIO</sub> = 2.30 V (8)			0.2	V
		I <sub>OL</sub> = 1 mA DC, V <sub>CCIO</sub> = 2.30 V (8)			0.4	V
I <sub>OL</sub> = 2 mA DC, V <sub>CCIO</sub> = 2.30 V (8)				0.7	V	
I <sub>I</sub>	Input pin leakage current	V <sub>I</sub> = 5.3 V to ground (8)	-10		10	μA
I <sub>OZ</sub>	Tri-stated I/O pin leakage current	V <sub>O</sub> = 5.3 V to ground (8)	-10		10	μA
I <sub>CC0</sub>	V <sub>CC</sub> supply current (standby)	V <sub>I</sub> = ground, no load		0.5	5	mA

**Table 18. FLEX 6000 3.3-V Device Capacitance** Note (9)

Symbol	Parameter	Conditions	Min	Max	Unit
C <sub>IN</sub>	Input capacitance for I/O pin	V <sub>IN</sub> = 0 V, f = 1.0 MHz		8	pF
C <sub>INCLK</sub>	Input capacitance for dedicated input	V <sub>IN</sub> = 0 V, f = 1.0 MHz		12	pF
C <sub>OUT</sub>	Output capacitance	V <sub>OUT</sub> = 0 V, f = 1.0 MHz		8	pF

**Notes to tables:**

- (1) See the *Operating Requirements for Altera Devices Data Sheet*.
- (2) The minimum DC input voltage is -0.5 V. During transitions, the inputs may undershoot to -2.0 V or overshoot to 5.75 V for input currents less than 100 mA and periods shorter than 20 ns.
- (3) Numbers in parentheses are for industrial-temperature-range devices.
- (4) Maximum V<sub>CC</sub> rise time is 100 ms. V<sub>CC</sub> must rise monotonically.
- (5) Typical values are for T<sub>A</sub> = 25° C and V<sub>CC</sub> = 3.3 V.
- (6) These values are specified under [Table 16 on page 33](#).
- (7) The I<sub>OH</sub> parameter refers to high-level TTL or CMOS output current.
- (8) The I<sub>OL</sub> parameter refers to low-level TTL, PCI, or CMOS output current. This parameter applies to open-drain pins as well as output pins.
- (9) Capacitance is sample-tested only.

## Timing Model

The continuous, high-performance FastTrack Interconnect routing resources ensure predictable performance and accurate simulation and timing analysis. This predictable performance contrasts with that of FPGAs, which use a segmented connection scheme and therefore have unpredictable performance.

Device performance can be estimated by following the signal path from a source, through the interconnect, to the destination. For example, the registered performance between two LEs on the same row can be calculated by adding the following parameters:

- LE register clock-to-output delay ( $t_{CO} + t_{REG\_TO\_OUT}$ )
- Routing delay ( $t_{ROW} + t_{LOCAL}$ )
- LE LUT delay ( $t_{DATA\_TO\_REG}$ )
- LE register setup time ( $t_{SU}$ )

The routing delay depends on the placement of the source and destination LEs. A more complex registered path may involve multiple combinatorial LEs between the source and destination LEs.

Timing simulation and delay prediction are available with the Simulator and Timing Analyzer, or with industry-standard EDA tools. The Simulator offers both pre-synthesis functional simulation to evaluate logic design accuracy and post-synthesis timing simulation with 0.1-ns resolution. The Timing Analyzer provides point-to-point timing delay information, setup and hold time analysis, and device-wide performance analysis.

Figure 19 shows the overall timing model, which maps the possible routing paths to and from the various elements of the FLEX 6000 device.



<b>Table 20. IOE Timing Microparameters</b> <i>Note (1)</i>		
<b>Symbol</b>	<b>Parameter</b>	<b>Conditions</b>
$t_{OD1}$	Output buffer and pad delay, slow slew rate = off, $V_{CCIO} = V_{CCINT}$	C1 = 35 pF (2)
$t_{OD2}$	Output buffer and pad delay, slow slew rate = off, $V_{CCIO} = \text{low voltage}$	C1 = 35 pF (3)
$t_{OD3}$	Output buffer and pad delay, slow slew rate = on	C1 = 35 pF (4)
$t_{XZ}$	Output buffer disable delay	C1 = 5 pF
$t_{ZX1}$	Output buffer enable delay, slow slew rate = off, $V_{CCIO} = V_{CCINT}$	C1 = 35 pF (2)
$t_{ZX2}$	Output buffer enable delay, slow slew rate = off, $V_{CCIO} = \text{low voltage}$	C1 = 35 pF (3)
$t_{ZX3}$	IOE output buffer enable delay, slow slew rate = on	C1 = 35 pF (4)
$t_{IOE}$	Output enable control delay	
$t_{IN}$	Input pad and buffer to FastTrack Interconnect delay	
$t_{IN\_DELAY}$	Input pad and buffer to FastTrack Interconnect delay with additional delay turned on	

<b>Table 21. Interconnect Timing Microparameters</b> <i>Note (1)</i>		
<b>Symbol</b>	<b>Parameter</b>	<b>Conditions</b>
$t_{LOCAL}$	LAB local interconnect delay	
$t_{ROW}$	Row interconnect routing delay	(5)
$t_{COL}$	Column interconnect routing delay	(5)
$t_{DIN\_D}$	Dedicated input to LE data delay	(5)
$t_{DIN\_C}$	Dedicated input to LE control delay	
$t_{LEGLOBAL}$	LE output to LE control via internally-generated global signal delay	(5)
$t_{LABCARRY}$	Routing delay for the carry-out of an LE driving the carry-in signal of a different LE in a different LAB	
$t_{LABCASC}$	Routing delay for the cascade-out signal of an LE driving the cascade-in signal of a different LE in a different LAB	

<b>Table 22. External Reference Timing Parameters</b>		
<b>Symbol</b>	<b>Parameter</b>	<b>Conditions</b>
$t_1$	Register-to-register test pattern	(6)
$t_{DRR}$	Register-to-register delay via 4 LEs, 3 row interconnects, and 4 local interconnects	(7)

Parameter	Speed Grade						Unit
	-1		-2		-3		
	Min	Max	Min	Max	Min	Max	
$t_{CO}$		0.3		0.4		0.4	ns
$t_{CLR}$		0.4		0.4		0.5	ns
$t_C$		1.8		2.1		2.6	ns
$t_{LD\_CLR}$		1.8		2.1		2.6	ns
$t_{CARRY\_TO\_CARRY}$		0.1		0.1		0.1	ns
$t_{REG\_TO\_CARRY}$		1.6		1.9		2.3	ns
$t_{DATA\_TO\_CARRY}$		2.1		2.5		3.0	ns
$t_{CARRY\_TO\_CASC}$		1.0		1.1		1.4	ns
$t_{CASC\_TO\_CASC}$		0.5		0.6		0.7	ns
$t_{REG\_TO\_CASC}$		1.4		1.7		2.1	ns
$t_{DATA\_TO\_CASC}$		1.1		1.2		1.5	ns
$t_{CH}$	2.5		3.0		3.5		ns
$t_{CL}$	2.5		3.0		3.5		ns

Parameter	Speed Grade						Unit
	-1		-2		-3		
	Min	Max	Min	Max	Min	Max	
$t_{OD1}$		1.9		2.2		2.7	ns
$t_{OD2}$		4.1		4.8		5.8	ns
$t_{OD3}$		5.8		6.8		8.3	ns
$t_{XZ}$		1.4		1.7		2.1	ns
$t_{XZ1}$		1.4		1.7		2.1	ns
$t_{XZ2}$		3.6		4.3		5.2	ns
$t_{XZ3}$		5.3		6.3		7.7	ns
$t_{IOE}$		0.5		0.6		0.7	ns
$t_{IN}$		3.6		4.1		5.1	ns
$t_{IN\_DELAY}$		4.8		5.4		6.7	ns

**Table 26. Interconnect Timing Microparameters for EPF6010A & EPF6016A Devices**

Parameter	Speed Grade						Unit
	-1		-2		-3		
	Min	Max	Min	Max	Min	Max	
$t_{LOCAL}$		0.7		0.7		1.0	ns
$t_{ROW}$		2.9		3.2		3.2	ns
$t_{COL}$		1.2		1.3		1.4	ns
$t_{DIN\_D}$		5.4		5.7		6.4	ns
$t_{DIN\_C}$		4.3		5.0		6.1	ns
$t_{LEGLOBAL}$		2.6		3.0		3.7	ns
$t_{LABCARRY}$		0.7		0.8		0.9	ns
$t_{LABCASC}$		1.3		1.4		1.8	ns

**Table 27. External Reference Timing Parameters for EPF6010A & EPF6016A Devices**

Parameter	Device	Speed Grade						Unit
		-1		-2		-3		
		Min	Max	Min	Max	Min	Max	
$t_1$	EPF6010A		37.6		43.6		53.7	ns
	EPF6016A		38.0		44.0		54.1	ns

**Table 28. External Timing Parameters for EPF6010A & EPF6016A Devices**

Parameter	Speed Grade						Unit
	-1		-2		-3		
	Min	Max	Min	Max	Min	Max	
$t_{INSU}$	2.1 (1)		2.4 (1)		3.3 (1)		ns
$t_{INH}$	0.2 (2)		0.3 (2)		0.1 (2)		ns
$t_{OUTCO}$	2.0	7.1	2.0	8.2	2.0	10.1	ns

**Notes:**

- (1) Setup times are longer when the *Increase Input Delay* option is turned on. The setup time values are shown with the *Increase Input Delay* option turned off.
- (2) Hold time is zero when the *Increase Input Delay* option is turned on.

**Table 30. IOE Timing Microparameters for EPF6016 Devices**

Parameter	Speed Grade				Unit
	-2		-3		
	Min	Max	Min	Max	
$t_{OD3}$		4.7		5.2	ns
$t_{XZ}$		2.3		2.8	ns
$t_{ZX1}$		2.3		2.8	ns
$t_{ZX2}$		4.6		5.1	ns
$t_{ZX3}$		4.7		5.2	ns
$t_{IOE}$		0.5		0.6	ns
$t_{IN}$		3.3		4.0	ns
$t_{IN\_DELAY}$		4.6		5.6	ns

**Table 31. Interconnect Timing Microparameters for EPF6016 Devices**

Parameter	Speed Grade				Unit
	-2		-3		
	Min	Max	Min	Max	
$t_{LOCAL}$		0.8		1.0	ns
$t_{ROW}$		2.9		3.3	ns
$t_{COL}$		2.3		2.5	ns
$t_{DIN\_D}$		4.9		6.0	ns
$t_{DIN\_C}$		4.8		6.0	ns
$t_{LEGLOBAL}$		3.1		3.9	ns
$t_{LABCARRY}$		0.4		0.5	ns
$t_{LABCASC}$		0.8		1.0	ns

**Table 32. External Reference Timing Parameters for EPF6016 Devices**

Parameter	Speed Grade				Unit
	-2		-3		
	Min	Max	Min	Max	
$t_1$		53.0		65.0	ns
$t_{DRR}$		16.0		20.0	ns

## Operating Modes

The FLEX 6000 architecture uses SRAM configuration elements that require configuration data to be loaded every time the circuit powers up. This process of physically loading the SRAM data into a FLEX 6000 device is known as configuration. During initialization—a process that occurs immediately after configuration—the device resets registers, enables I/O pins, and begins to operate as a logic device. The I/O pins are tri-stated during power-up, and before and during configuration. The configuration and initialization processes of a device are referred to as *command mode*; normal device operation is called *user mode*.

SRAM configuration elements allow FLEX 6000 devices to be reconfigured in-circuit by loading new configuration data into the device. Real-time reconfiguration is performed by forcing the device into command mode with a device pin, loading different configuration data, reinitializing the device, and resuming user-mode operation. The entire reconfiguration process requires less than 100 ms and is used to dynamically reconfigure an entire system. Also, in-field system upgrades can be performed by distributing new configuration files.

## Configuration Schemes

The configuration data for a FLEX 6000 device can be loaded with one of three configuration schemes, which is chosen on the basis of the target application. An EPC1 or EPC1441 configuration device or intelligent controller can be used to control the configuration of a FLEX 6000 device, allowing automatic configuration on system power-up.

Multiple FLEX 6000 devices can be configured in any of the three configuration schemes by connecting the configuration enable input (nCE) and configuration enable output (nCEO) pins on each device.

Table 40 shows the data sources for each configuration scheme.

<b>Configuration Scheme</b>	<b>Data Source</b>
Configuration device	EPC1 or EPC1441 configuration device
Passive serial (PS)	BitBlaster™, ByteBlasterMV™, or MasterBlaster™ download cables, or serial data source
Passive serial asynchronous (PSA)	BitBlaster, ByteBlasterMV, or MasterBlaster download cables, or serial data source