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Understanding <u>Embedded - FPGAs (Field Programmable Gate Array)</u>

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	132
Number of Logic Elements/Cells	1320
Total RAM Bits	-
Number of I/O	171
Number of Gates	16000
Voltage - Supply	3V ~ 3.6V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	208-BFQFP
Supplier Device Package	208-PQFP (28x28)
Purchase URL	https://www.e-xfl.com/product-detail/intel/epf6016aqc208-1

Email: info@E-XFL.COM

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General Description

The Altera® FLEX 6000 programmable logic device (PLD) family provides a low-cost alternative to high-volume gate array designs. FLEX 6000 devices are based on the OptiFLEX architecture, which minimizes die size while maintaining high performance and routability. The devices have reconfigurable SRAM elements, which give designers the flexibility to quickly change their designs during prototyping and design testing. Designers can also change functionality during operation via in-circuit reconfiguration.

FLEX 6000 devices are reprogrammable, and they are 100% tested prior to shipment. As a result, designers are not required to generate test vectors for fault coverage purposes, allowing them to focus on simulation and design verification. In addition, the designer does not need to manage inventories of different gate array designs. FLEX 6000 devices are configured on the board for the specific functionality required.

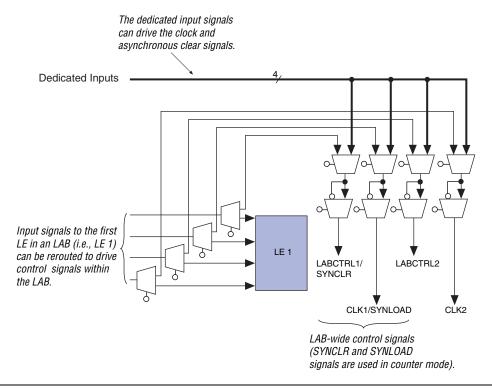
Table 3 shows FLEX 6000 performance for some common designs. All performance values shown were obtained using Synopsys DesignWare or LPM functions. Special design techniques are not required to implement the applications; the designer simply infers or instantiates a function in a Verilog HDL, VHDL, Altera Hardware Description Language (AHDL), or schematic design file.

Application	LEs Used	LEs Used Performance					
		-1 Speed Grade	-2 Speed Grade	-3 Speed Grade			
16-bit loadable counter	16	172	153	133	MHz		
16-bit accumulator	16	172	153	133	MHz		
24-bit accumulator	24	136	123	108	MHz		
16-to-1 multiplexer (pin-to-pin) (1)	10	12.1	13.4	16.6	ns		
16 × 16 multiplier with a 4-stage pipeline	592	84	67	58	MHz		

Note:

(1) This performance value is measured as a pin-to-pin delay.

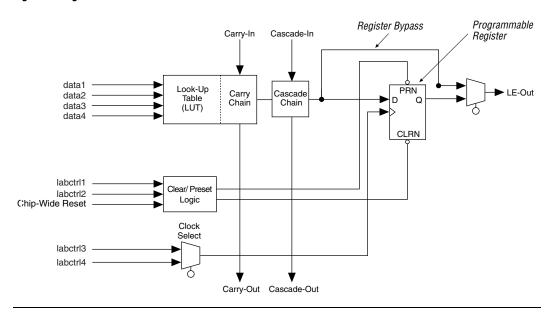
Figure 3. LAB Control Signals



Logic Element

An LE, the smallest unit of logic in the FLEX 6000 architecture, has a compact size that provides efficient logic usage. Each LE contains a four-input LUT, which is a function generator that can quickly implement any function of four variables. An LE contains a programmable flipflop, carry and cascade chains. Additionally, each LE drives both the local and the FastTrack Interconnect. See Figure 4.

Figure 4. Logic Element



The programmable flipflop in the LE can be configured for D, T, JK, or SR operation. The clock and clear control signals on the flipflop can be driven by global signals, general-purpose I/O pins, or any internal logic. For combinatorial functions, the flipflop is bypassed and the output of the LUT drives the outputs of the LE. The LE output can drive both the local interconnect and the FastTrack Interconnect.

The FLEX 6000 architecture provides two types of dedicated high-speed data paths that connect adjacent LEs without using local interconnect paths: carry chains and cascade chains. A carry chain supports high-speed arithmetic functions such as counters and adders, while a cascade chain implements wide-input functions such as equivalent comparators with minimum delay. Carry and cascade chains connect LEs 2 through 10 in an LAB and all LABs in the same half of the row. Because extensive use of carry and cascade chains can reduce routing flexibility, these chains should be limited to speed-critical portions of a design.

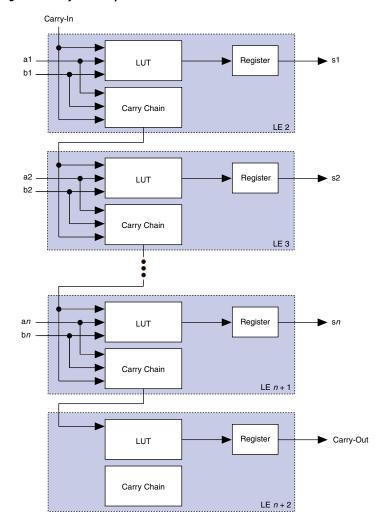
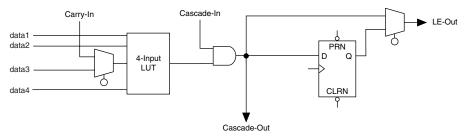


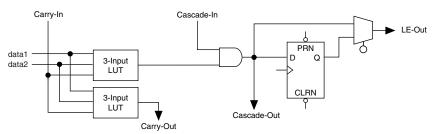
Figure 5. Carry Chain Operation

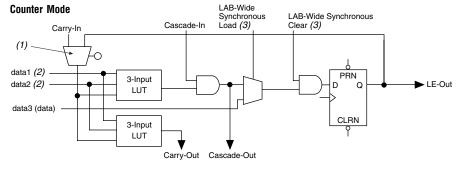
Figure 7. LE Operating Modes

Normal Mode



Arithmetic Mode





Notes:

- (1) The register feedback multiplexer is available on LE 2 of each LAB.
- (2) The data1 and data2 input signals can supply a clock enable, up or down control, or register feedback signals for all LEs other than the second LE in an LAB.
- (3) The LAB-wide synchronous clear and LAB-wide synchronous load affect all registers in an LAB.

The FastTrack Interconnect consists of column and row interconnect channels that span the entire device. Each row of LABs is served by a dedicated row interconnect, which routes signals between LABs in the same row, and also routes signals from I/O pins to LABs. Additionally, the local interconnect routes signals between LEs in the same LAB and in adjacent LABs. The column interconnect routes signals between rows and routes signals from I/O pins to rows.

LEs 1 through 5 of an LAB drive the local interconnect to the right, while LEs 6 through 10 drive the local interconnect to the left. The DATA1 and DATA3 inputs of each LE are driven by the local interconnect to the left; DATA2 and DATA4 are driven by the local interconnect to the right. The local interconnect also routes signals from LEs to I/O pins. Figure 9 shows an overview of the FLEX 6000 interconnect architecture. LEs in the first and last columns have drivers on both sides so that all LEs in the LAB can drive I/O pins via the local interconnect.

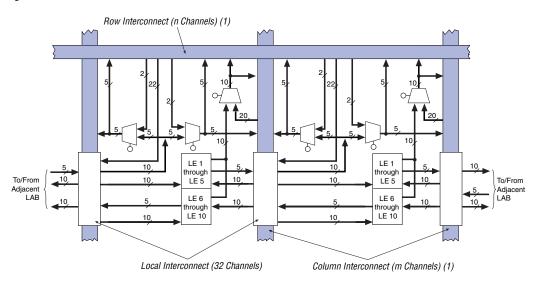


Figure 9. FastTrack Interconnect Architecture

Note:

(1) For EPF6010A, EPF6016, and EPF6016A devices, *n* = 144 channels and *m* = 20 channels; for EPF6024A devices, *n* = 186 channels and *m* = 30 channels.

A row channel can be driven by an LE or by one of two column channels. These three signals feed a 3-to-1 multiplexer that connects to six specific row channels. Row channels drive into the local interconnect via multiplexers.

Each column of LABs is served by a dedicated column interconnect. The LEs in an LAB can drive the column interconnect. The LEs in an LAB, a column IOE, or a row interconnect can drive the column interconnect. The column interconnect can then drive another row's interconnect to route the signals to other LABs in the device. A signal from the column interconnect must be routed to the row interconnect before it can enter an LAB.

Each LE has a FastTrack Interconnect output and a local output. The FastTrack interconnect output can drive six row and two column lines directly; the local output drives the local interconnect. Each local interconnect channel driven by an LE can drive four row and two column channels. This feature provides additional flexibility, because each LE can drive any of ten row lines and four column lines.

In addition, LEs can drive global control signals. This feature is useful for distributing internally generated clock, asynchronous clear, and asynchronous preset signals. A pin-driven global signal can also drive data signals, which is useful for high-fan-out data signals.

Each LAB drives two groups of local interconnects, which allows an LE to drive two LABs, or 20 LEs, via the local interconnect. The row-to-local multiplexers are used more efficiently, because the multiplexers can now drive two LABs. Figure 10 shows how an LAB connects to row and column interconnects.

Each LE FastTrack Interconnect output can drive six row channels. Each local channel driven by an LE can Each LE output signal driving drive two column the FastTrack Interconnect can channels. drive two column channels. At each intersection, four row channels can Row drive column channels. Interconnect Each local channel driven by an LE can drive four row channels. Row interconnect drives the local interconnect. From Adjacent Local Interconnect Local Interconnect Column Interconnect Any column channel can drive six row channels.

An LE can be driven by any signal from two local interconnect areas.

Figure 10. LAB Connections to Row & Column Interconnects

For improved routability, the row interconnect consists of full-length and half-length channels. The full-length channels connect to all LABs in a row; the half-length channels connect to the LABs in half of the row. In addition to providing a predictable, row-wide interconnect, this architecture provides increased routing resources. Two neighboring LABs can be connected using a half-length channel, which saves the other half of the channel for the other half of the row. One-third of the row channels are half-length channels.

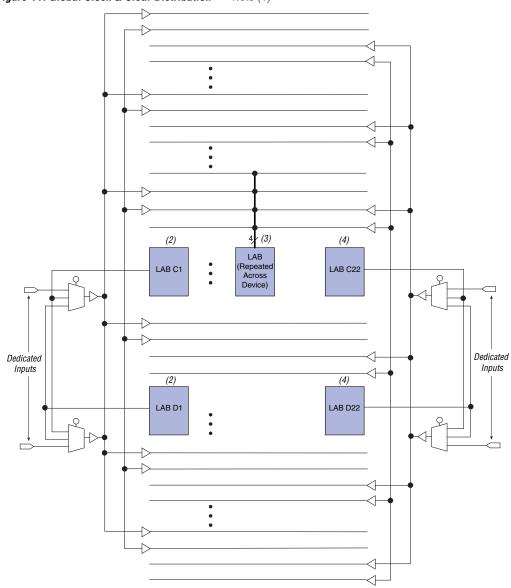


Figure 11. Global Clock & Clear Distribution Note (1)

Notes:

- The global clock and clear distribution signals are shown for EPF6016 and EPF6016A devices. In EPF6010A devices, LABs in rows B and C drive global signals. In EPF6024A devices, LABs in rows C and E drive global signals. The local interconnect from LABs C1 and D1 can drive two global control signals on the left side.
- (2)
- Global signals drive into every LAB as clock, asynchronous clear, preset, and data signals. (3)
- The local interconnect from LABs C22 and D22 can drive two global control signals on the right side.

Each IOE can drive two column interconnect channels. Each IOE data and OE signal is driven to a local interconnect. IOE IOE FastFLEX I/O: An LE can drive a pin through a local interconnect for faster clock-to-output times. LAB Any LE can drive a pin through the row Column Interconnect and local interconnect. Row Interconnect

Figure 14. IOE Connection to Column Interconnect

SameFrame Pin-Outs

3.3-V FLEX 6000 devices support the SameFrame pin-out feature for FineLine BGA packages. The SameFrame pin-out feature is the arrangement of balls on FineLine BGA packages such that the lower-ball-count packages form a subset of the higher-ball-count packages. SameFrame pin-outs provide the flexibility to migrate not only from device to device within the same package, but also from one package to another. A given printed circuit board (PCB) layout can support multiple device density/package combinations. For example, a single board layout can support an EPF6016A device in a 100-pin FineLine BGA package or an EPF6024A device in a 256-pin FineLine BGA package.

The Altera software packages provide support to design PCBs with SameFrame pin-out devices. Devices can be defined for present and future use. The Altera software packages generate pin-outs describing how to lay out a board to take advantage of this migration (see Figure 15).

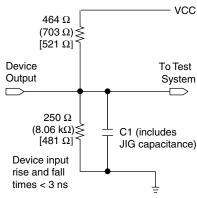
Symbol	Parameter	Min	Max	Unit
t _{JCP}	TCK clock period	100		ns
t _{JCH}	TCK clock high time	50		ns
t _{JCL}	TCK clock low time	50		ns
t _{JPSU}	JTAG port setup time	20		ns
t _{JPH}	JTAG port hold time	45		ns
t _{JPCO}	JTAG port clock-to-output		25	ns
t _{JPZX}	JTAG port high impedance to valid output		25	ns
t _{JPXZ}	JTAG port valid output to high impedance		25	ns
t _{JSSU}	Capture register setup time	20		ns
t _{JSH}	Capture register hold time	45		ns
t _{JSCO}	Update register clock-to-output		35	ns
t _{JSZX}	Update register high impedance to valid output		35	ns
t _{JSXZ}	Update register valid output to high impedance		35	ns

Generic Testing

Each FLEX 6000 device is functionally tested. Complete testing of each configurable SRAM bit and all logic functionality ensures 100% configuration yield. AC test measurements for FLEX 6000 devices are made under conditions equivalent to those shown in Figure 17. Multiple test patterns can be used to configure devices during all stages of the production flow.

Figure 17. AC Test Conditions

Power supply transients can affect AC measurements. Simultaneous transitions of multiple outputs should be avoided for accurate measurement. Threshold tests must not be performed under AC conditions. Large-amplitude, fast-ground-current transients normally occur as the device outputs discharge the load capacitances. When these transients flow through the parasitic inductance between the device ground pin and the test system ground, significant reductions in observable noise immunity can result. Numbers without parentheses are for 5.0-V devices or outputs. Numbers in parentheses are for 3.3-V devices or outputs. Numbers in brackets are for 2.5-V devices or outputs.



Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V _{IH}	High-level input voltage		1.7		5.75	٧
V _{IL}	Low-level input voltage		-0.5		0.8	٧
V _{OH}	3.3-V high-level TTL output voltage	$I_{OH} = -8 \text{ mA DC}, V_{CCIO} = 3.00 \text{ V}$ (7)	2.4			V
	3.3-V high-level CMOS output voltage	$I_{OH} = -0.1 \text{ mA DC}, V_{CCIO} = 3.00 \text{ V}$ (7)	V _{CCIO} - 0.2			V
	2.5-V high-level output voltage	$I_{OH} = -100 \mu A DC, V_{CCIO} = 2.30 V (7)$	2.1			٧
		I _{OH} = -1 mA DC, V _{CCIO} = 2.30 V (7)	2.0			٧
		$I_{OH} = -2 \text{ mA DC}, V_{CCIO} = 2.30 \text{ V}$ (7)	1.7			٧
V _{OL}	3.3-V low-level TTL output voltage	$I_{OL} = 8 \text{ mA DC}, V_{CCIO} = 3.00 \text{ V } (8)$			0.45	V
	3.3-V low-level CMOS output voltage	$I_{OL} = 0.1 \text{ mA DC}, V_{CCIO} = 3.00 \text{ V } (8)$			0.2	V
	2.5-V low-level output voltage	I _{OL} = 100 μA DC, V _{CCIO} = 2.30 V (8)			0.2	٧
		I _{OL} = 1 mA DC, V _{CCIO} = 2.30 V (8)			0.4	٧
		I _{OL} = 2 mA DC, V _{CCIO} = 2.30 V (8)			0.7	٧
I _I	Input pin leakage current	V _I = 5.3 V to ground (8)	-10		10	μΑ
I _{OZ}	Tri-stated I/O pin leakage current	$V_O = 5.3 \text{ V to ground } (8)$	-10		10	μΑ
I _{CC0}	V _{CC} supply current (standby)	V _I = ground, no load		0.5	5	mA

Table 1	Table 18. FLEX 6000 3.3-V Device CapacitanceNote (9)									
Symbol	Parameter	Conditions	Min	Max	Unit					
C _{IN}	Input capacitance for I/O pin	V _{IN} = 0 V, f = 1.0 MHz		8	pF					
C _{INCLK}	Input capacitance for dedicated input	$V_{IN} = 0 V$, $f = 1.0 MHz$		12	pF					
C _{OUT}	Output capacitance	V _{OUT} = 0 V, f = 1.0 MHz		8	pF					

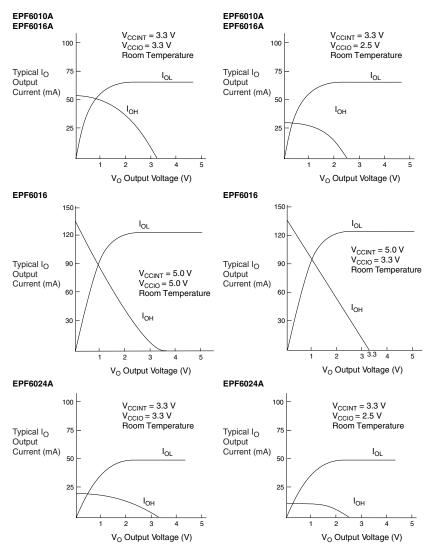
Notes to tables:

- (1) See the Operating Requirements for Altera Devices Data Sheet.
- (2) The minimum DC input voltage is -0.5 V. During transitions, the inputs may undershoot to -2.0 V or overshoot to 5.75 V for input currents less than 100 mA and periods shorter than 20 ns.
- (3) Numbers in parentheses are for industrial-temperature-range devices.

- (4) Maximum V_{CC} rise time is 100 ms. V_{CC} must rise monotonically.
 (5) Typical values are for T_A = 25° C and V_{CC} = 3.3 V.
 (6) These values are specified under Table 16 on page 33.
 (7) The I_{OH} parameter refers to high-level TTL or CMOS output current.
- (8) The I_{OL} parameter refers to low-level TTL, PCI, or CMOS output current. This parameter applies to open-drain pins as well as output pins.
- (9) Capacitance is sample-tested only.

Figure 18 shows the typical output drive characteristics of 5.0-V and 3.3-V FLEX 6000 devices with 5.0-V, 3.3-V, and 2.5-V $V_{\rm CCIO}$. When $V_{\rm CCIO}=5.0$ V on EPF6016 devices, the output driver is compliant with the *PCI Local Bus Specification, Revision* 2.2 for 5.0-V operation. When $V_{\rm CCIO}=3.3$ V on the EPF6010A and EPF6016A devices, the output driver is compliant with the *PCI Local Bus Specification, Revision* 2.2 for 3.3-V operation.

Figure 18. Output Drive Characteristics



Timing Model

The continuous, high-performance FastTrack Interconnect routing resources ensure predictable performance and accurate simulation and timing analysis. This predictable performance contrasts with that of FPGAs, which use a segmented connection scheme and therefore have unpredictable performance.

Device performance can be estimated by following the signal path from a source, through the interconnect, to the destination. For example, the registered performance between two LEs on the same row can be calculated by adding the following parameters:

- LE register clock-to-output delay ($t_{CO} + t_{REG_TO_OUT}$)
- Routing delay $(t_{ROW} + t_{LOCAL})$
- LE LUT delay ($t_{DATA_TO_REG}$)
- LE register setup time (t_{SU})

The routing delay depends on the placement of the source and destination LEs. A more complex registered path may involve multiple combinatorial LEs between the source and destination LEs.

Timing simulation and delay prediction are available with the Simulator and Timing Analyzer, or with industry-standard EDA tools. The Simulator offers both pre-synthesis functional simulation to evaluate logic design accuracy and post-synthesis timing simulation with 0.1-ns resolution. The Timing Analyzer provides point-to-point timing delay information, setup and hold time analysis, and device-wide performance analysis.

Figure 19 shows the overall timing model, which maps the possible routing paths to and from the various elements of the FLEX 6000 device.

Figure 19. FLEX 6000 Timing Model

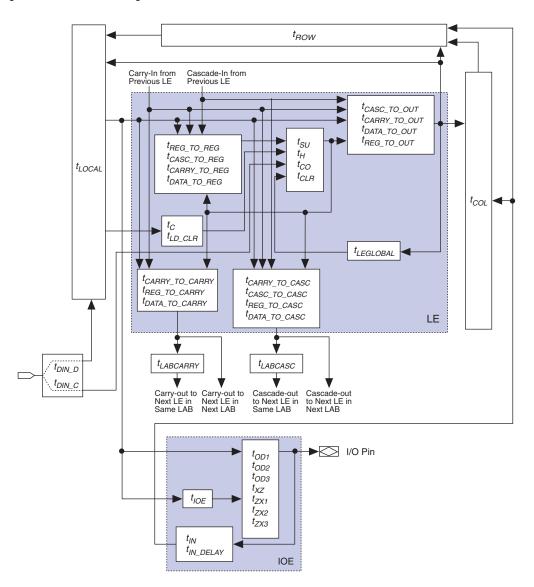


Table 23. Ex	Table 23. External Timing Parameters							
Symbol	Parameter	Conditions						
t _{INSU}	Setup time with global clock at LE register	(8)						
t _{INH}	Hold time with global clock at LE register	(8)						
t _{оитсо}	Clock-to-output delay with global clock with LE register using FastFLEX I/O pin	(8)						

Notes to tables:

- Microparameters are timing delays contributed by individual architectural elements and cannot be measured explicitly.
- (2) Operating conditions:
 - $\hat{V_{CCIO}} = \widecheck{5}.0~V \pm 5\%$ for commercial use in 5.0-V FLEX 6000 devices.
 - $V_{CCIO} = 5.0 \text{ V} \pm 10\%$ for industrial use in 5.0-V FLEX 6000 devices.
 - $V_{CCIO} = 3.3 \text{ V} \pm 10\%$ for commercial or industrial use in 3.3-V FLEX 6000 devices.
- (3) Operating conditions:
 - $\hat{V_{CCIO}} = 3.3 \text{ V} \pm 10\%$ for commercial or industrial use in 5.0-V FLEX 6000 devices.
 - V_{CCIO} = 2.5 V ±0.2 V for commercial or industrial use in 3.3-V FLEX 6000 devices.
- (4) Operating conditions:
 - $V_{\text{CCIO}} = 2.5 \text{ V}, 3.3 \text{ V}, \text{ or } 5.0 \text{ V}.$
- (5) These parameters are worst-case values for typical applications. Post-compilation timing simulation and timing analysis are required to determine actual worst-case performance.
- (6) This timing parameter shows the delay of a register-to-register test pattern and is used to determine speed grades. There are 12 LEs, including source and destination registers. The row and column interconnects between the registers vary in length.
- 7) This timing parameter is shown for reference and is specified by characterization.
- (8) This timing parameter is specified by characterization.

Tables 24 through 28 show the timing information for EPF6010A and EPF6016A devices.

Parameter	Speed Grade								
	-1		-	-2		3	1		
	Min	Max	Min	Max	Min	Max			
treg_to_reg		1.2		1.3		1.7	ns		
t _{CASC_TO_REG}		0.9		1.0		1.2	ns		
t _{CARRY_TO_REG}		0.9		1.0		1.2	ns		
t _{DATA_TO_REG}		1.1		1.2		1.5	ns		
t _{CASC_TO_OUT}		1.3		1.4		1.8	ns		
t _{CARRY_TO_OUT}		1.6		1.8		2.3	ns		
^t DATA_TO_OUT		1.7		2.0		2.5	ns		
t _{REG_TO_OUT}		0.4		0.4		0.5	ns		
t _{SU}	0.9		1.0		1.3		ns		
t _H	1.4		1.7		2.1		ns		

Parameter	Speed Grade								
	-	1	-2		-3]		
	Min	Max	Min	Max	Min	Max			
t _{co}		0.3		0.4		0.4	ns		
t _{CLR}		0.4		0.4		0.5	ns		
t _C		1.8		2.1		2.6	ns		
t _{LD_CLR}		1.8		2.1		2.6	ns		
tCARRY_TO_CARRY		0.1		0.1		0.1	ns		
tREG_TO_CARRY		1.6		1.9		2.3	ns		
tDATA_TO_CARRY		2.1		2.5		3.0	ns		
tCARRY_TO_CASC		1.0		1.1		1.4	ns		
tcasc_to_casc		0.5		0.6		0.7	ns		
tREG_TO_CASC		1.4		1.7		2.1	ns		
t _{DATA_TO_CASC}		1.1		1.2		1.5	ns		
^t ch	2.5		3.0		3.5		ns		
^t CL	2.5		3.0		3.5		ns		

Parameter	Speed Grade								
	-1		-2		-3		1		
	Min	Max	Min	Max	Min	Max			
t _{OD1}		1.9		2.2		2.7	ns		
t _{OD2}		4.1		4.8		5.8	ns		
t _{OD3}		5.8		6.8		8.3	ns		
t_{XZ}		1.4		1.7		2.1	ns		
t _{XZ1}		1.4		1.7		2.1	ns		
t _{XZ2}		3.6		4.3		5.2	ns		
t _{XZ3}		5.3		6.3		7.7	ns		
t _{IOE}		0.5		0.6		0.7	ns		
t _{IN}		3.6		4.1		5.1	ns		
tin delay		4.8		5.4		6.7	ns		

Parameter	Speed Grade								
	-	1		2	-;	3			
	Min	Max	Min	Max	Min	Max			
t _{LOCAL}		0.7		0.7		1.0	ns		
t _{ROW}		2.9		3.2		3.2	ns		
t _{COL}		1.2		1.3		1.4	ns		
t _{DIN_D}		5.4		5.7		6.4	ns		
t _{DIN_C}		4.3		5.0		6.1	ns		
t LEGLOBAL		2.6		3.0		3.7	ns		
t _{LABCARRY}		0.7		0.8		0.9	ns		
t _{LABCASC}		1.3		1.4		1.8	ns		

Table 27. External Reference Timing Parameters for EPF6010A & EPF6016A Devices									
Parameter	Parameter Device Speed Grade						Unit		
		-1 -2 -3							
		Min	Max	Min	Max	Min	Max		
t ₁	EPF6010A		37.6		43.6		53.7	ns	
	EPF6016A		38.0		44.0		54.1	ns	

Table 28. Externa	Table 28. External Timing Parameters for EPF6010A & EPF6016A Devices									
Parameter			Speed (Grade			Unit			
	-1	I	-2	l.	-;	3				
	Min	Max	Min	Max	Min	Max				
t _{INSU}	2.1 (1)		2.4 (1)		3.3 (1)		ns			
t _{INH}	0.2 (2)		0.3 (2)		0.1 (2)		ns			
t _{оитсо}	2.0	7.1	2.0	8.2	2.0	10.1	ns			

Notes:

Setup times are longer when the *Increase Input Delay* option is turned on. The setup time values are shown with the *Increase Input Delay* option turned off.
 Hold time is zero when the *Increase Input Delay* option is turned on.

Parameter	Speed Grade				
	-2		-3		1
	Min	Max	Min	Max	
OD3		4.7		5.2	ns
XZ		2.3		2.8	ns
ZX1		2.3		2.8	ns
ZX2		4.6		5.1	ns
ZX3		4.7		5.2	ns
IOE		0.5		0.6	ns
^t in		3.3		4.0	ns
t _{IN DELAY}		4.6		5.6	ns

Parameter	Speed Grade					
	-2		-3]	
	Min	Max	Min	Max		
t _{LOCAL}		0.8		1.0	ns	
t _{ROW}		2.9		3.3	ns	
t _{COL}		2.3		2.5	ns	
t _{DIN_D}		4.9		6.0	ns	
t _{DIN_C}		4.8		6.0	ns	
t _{LEGLOBAL}		3.1		3.9	ns	
t _{LABCARRY}		0.4		0.5	ns	
t _{LABCASC}		0.8		1.0	ns	

Table 32. External Referen	nce Timing Parai	meters for EPF60	116 Devices		
Parameter		Unit			
		-2		-3	
	Min	Max	Min	Max	
t ₁		53.0		65.0	ns
t _{DRR}		16.0		20.0	ns

Table 33. External Timing Parameters for EPF6016 Devices						
Parameter		Unit				
	-2		-3		1	
	Min	Max	Min	Max		
t _{INSU}	3.2		4.1		ns	
t _{INH}	0.0		0.0		ns	
t _{оитсо}	2.0	7.9	2.0	9.9	ns	

Tables 34 through 38 show the timing information for EPF6024A devices.

Parameter	Speed Grade							
	-1		-2		-3			
	Min	Max	Min	Max	Min	Max		
t _{REG_TO_REG}		1.2		1.3		1.6	ns	
t _{CASC_TO_REG}		0.7		0.8		1.0	ns	
t _{CARRY_TO_REG}		1.6		1.8		2.2	ns	
t _{DATA_TO_REG}		1.3		1.4		1.7	ns	
t _{CASC_TO_OUT}		1.2		1.3		1.6	ns	
t _{CARRY_TO_OUT}		2.0		2.2		2.6	ns	
t _{DATA_TO_OUT}		1.8		2.1		2.6	ns	
t _{REG_TO_OUT}		0.3		0.3		0.4	ns	
t _{SU}	0.9		1.0		1.2		ns	
t _H	1.3		1.4		1.7		ns	
t_{CO}		0.2		0.3		0.3	ns	
t _{CLR}		0.3		0.3		0.4	ns	
t_C		1.9		2.1		2.5	ns	
t _{LD_CLR}		1.9		2.1		2.5	ns	
t _{CARRY_TO_CARRY}		0.2		0.2		0.3	ns	
t _{REG_TO_CARRY}		1.4		1.6		1.9	ns	
t _{DATA_TO_CARRY}		1.3		1.4		1.7	ns	
t _{CARRY_TO_CASC}		1.1		1.2		1.4	ns	
t _{CASC_TO_CASC}		0.7		0.8		1.0	ns	
t _{REG_TO_CASC}		1.4		1.6		1.9	ns	
t _{DATA_TO_CASC}		1.0		1.1		1.3	ns	
t _{CH}	2.5		3.0		3.5		ns	
t_{CL}	2.5		3.0		3.5	_	ns	