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# Understanding <u>Embedded - FPGAs (Field Programmable Gate Array)</u>

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

## **Applications of Embedded - FPGAs**

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	132
Number of Logic Elements/Cells	1320
Total RAM Bits	-
Number of I/O	171
Number of Gates	16000
Voltage - Supply	3V ~ 3.6V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	208-BFQFP
Supplier Device Package	208-PQFP (28x28)
Purchase URL	https://www.e-xfl.com/product-detail/intel/epf6016aqc208-2

Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

# Functional Description

The FLEX 6000 OptiFLEX architecture consists of logic elements (LEs). Each LE includes a 4-input look-up table (LUT), which can implement any 4-input function, a register, and dedicated paths for carry and cascade chain functions. Because each LE contains a register, a design can be easily pipelined without consuming more LEs. The specified gate count for FLEX 6000 devices includes all LUTs and registers.

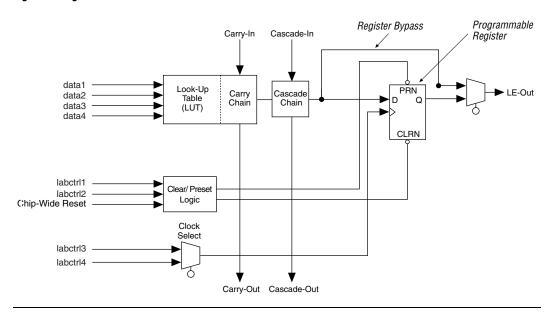
LEs are combined into groups called logic array blocks (LABs); each LAB contains 10 LEs. The Altera software automatically places related LEs into the same LAB, minimizing the number of required interconnects. Each LAB can implement a medium-sized block of logic, such as a counter or multiplexer.

Signal interconnections within FLEX 6000 devices—and to and from device pins—are provided via the routing structure of the FastTrack Interconnect. The routing structure is a series of fast, continuous row and column channels that run the entire length and width of the device. Any LE or pin can feed or be fed by any other LE or pin via the FastTrack Interconnect. See "FastTrack Interconnect" on page 17 of this data sheet for more information.

Each I/O pin is fed by an I/O element (IOE) located at the end of each row and column of the FastTrack Interconnect. Each IOE contains a bidirectional I/O buffer. Each IOE is placed next to an LAB, where it can be driven by the local interconnect of that LAB. This feature allows fast clock-to-output times of less than 8 ns when a pin is driven by any of the 10 LEs in the adjacent LAB. Also, any LE can drive any pin via the row and column interconnect. I/O pins can drive the LE registers via the row and column interconnect, providing setup times as low as 2 ns and hold times of 0 ns. IOEs provide a variety of features, such as JTAG BST support, slew-rate control, and tri-state buffers.

Figure 1 shows a block diagram of the FLEX 6000 OptiFLEX architecture. Each group of ten LEs is combined into an LAB, and the LABs are arranged into rows and columns. The LABs are interconnected by the FastTrack Interconnect. IOEs are located at the end of each FastTrack Interconnect row and column.

Figure 4. Logic Element



The programmable flipflop in the LE can be configured for D, T, JK, or SR operation. The clock and clear control signals on the flipflop can be driven by global signals, general-purpose I/O pins, or any internal logic. For combinatorial functions, the flipflop is bypassed and the output of the LUT drives the outputs of the LE. The LE output can drive both the local interconnect and the FastTrack Interconnect.

The FLEX 6000 architecture provides two types of dedicated high-speed data paths that connect adjacent LEs without using local interconnect paths: carry chains and cascade chains. A carry chain supports high-speed arithmetic functions such as counters and adders, while a cascade chain implements wide-input functions such as equivalent comparators with minimum delay. Carry and cascade chains connect LEs 2 through 10 in an LAB and all LABs in the same half of the row. Because extensive use of carry and cascade chains can reduce routing flexibility, these chains should be limited to speed-critical portions of a design.

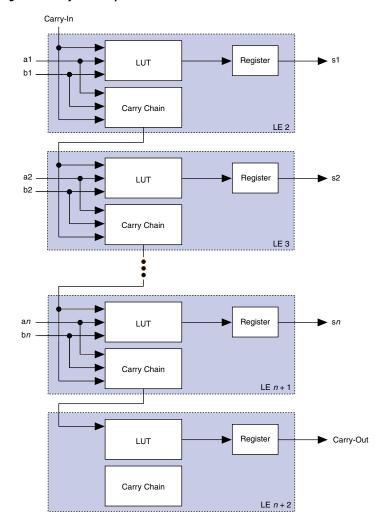


Figure 5. Carry Chain Operation

#### **Normal Mode**

The normal mode is suitable for general logic applications, combinatorial functions, or wide decoding functions that can take advantage of a cascade chain. In normal mode, four data inputs from the LAB local interconnect and the carry-in are inputs to a 4-input LUT. The Altera software automatically selects the carry-in or the DATA3 signal as one of the inputs to the LUT. The LUT output can be combined with the cascade-in signal to form a cascade chain through the cascade-out signal.

#### Arithmetic Mode

The arithmetic mode is ideal for implementing adders, accumulators, and comparators. An LE in arithmetic mode uses two 3-input LUTs. One LUT computes a 3-input function; the other generates a carry output. As shown in Figure 7, the first LUT uses the carry-in signal and two data inputs from the LAB local interconnect to generate a combinatorial or registered output. For example, when implementing an adder, this output is the sum of three signals: DATA1, DATA2, and carry-in. The second LUT uses the same three signals to generate a carry-out signal, thereby creating a carry chain. The arithmetic mode also supports simultaneous use of the cascade chain.

The Altera software implements logic functions to use the arithmetic mode automatically where appropriate; the designer does not have to decide how the carry chain will be used.

### **Counter Mode**

The counter mode offers counter enable, synchronous up/down control, synchronous clear, and synchronous load options. The counter enable and synchronous up/down control signals are generated from the data inputs of the LAB local interconnect. The synchronous clear and synchronous load options are LAB-wide signals that affect all registers in the LAB. Consequently, if any of the LEs in a LAB use counter mode, other LEs in that LAB must be used as part of the same counter or be used for a combinatorial function. In addition, the Altera software automatically places registers that are not in the counter into other LABs.

The counter mode uses two 3-input LUTs: one generates the counter data and the other generates the fast carry bit. A 2-to-1 multiplexer provides synchronous loading, and another AND gate provides synchronous clearing. If the cascade function is used by an LE in counter mode, the synchronous clear or load will override any signal carried on the cascade chain. The synchronous clear overrides the synchronous load.

A row channel can be driven by an LE or by one of two column channels. These three signals feed a 3-to-1 multiplexer that connects to six specific row channels. Row channels drive into the local interconnect via multiplexers.

Each column of LABs is served by a dedicated column interconnect. The LEs in an LAB can drive the column interconnect. The LEs in an LAB, a column IOE, or a row interconnect can drive the column interconnect. The column interconnect can then drive another row's interconnect to route the signals to other LABs in the device. A signal from the column interconnect must be routed to the row interconnect before it can enter an LAB.

Each LE has a FastTrack Interconnect output and a local output. The FastTrack interconnect output can drive six row and two column lines directly; the local output drives the local interconnect. Each local interconnect channel driven by an LE can drive four row and two column channels. This feature provides additional flexibility, because each LE can drive any of ten row lines and four column lines.

In addition, LEs can drive global control signals. This feature is useful for distributing internally generated clock, asynchronous clear, and asynchronous preset signals. A pin-driven global signal can also drive data signals, which is useful for high-fan-out data signals.

Each LAB drives two groups of local interconnects, which allows an LE to drive two LABs, or 20 LEs, via the local interconnect. The row-to-local multiplexers are used more efficiently, because the multiplexers can now drive two LABs. Figure 10 shows how an LAB connects to row and column interconnects.

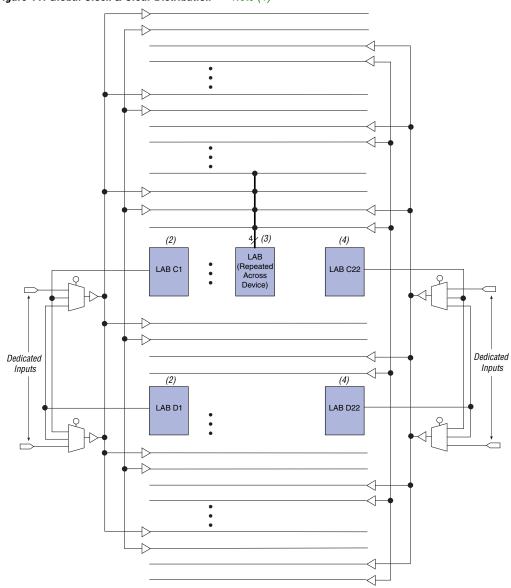


Figure 11. Global Clock & Clear Distribution Note (1)

### Notes:

- The global clock and clear distribution signals are shown for EPF6016 and EPF6016A devices. In EPF6010A devices, LABs in rows B and C drive global signals. In EPF6024A devices, LABs in rows C and E drive global signals. The local interconnect from LABs C1 and D1 can drive two global control signals on the left side.
- (2)
- Global signals drive into every LAB as clock, asynchronous clear, preset, and data signals. (3)
- The local interconnect from LABs C22 and D22 can drive two global control signals on the right side.

Each IOE can drive two column interconnect channels. Each IOE data and OE signal is driven to a local interconnect. IOE IOE FastFLEX I/O: An LE can drive a pin through a local interconnect for faster clock-to-output times. LAB Any LE can drive a pin through the row Column Interconnect and local interconnect. Row Interconnect

Figure 14. IOE Connection to Column Interconnect

# SameFrame Pin-Outs

3.3-V FLEX 6000 devices support the SameFrame pin-out feature for FineLine BGA packages. The SameFrame pin-out feature is the arrangement of balls on FineLine BGA packages such that the lower-ball-count packages form a subset of the higher-ball-count packages. SameFrame pin-outs provide the flexibility to migrate not only from device to device within the same package, but also from one package to another. A given printed circuit board (PCB) layout can support multiple device density/package combinations. For example, a single board layout can support an EPF6016A device in a 100-pin FineLine BGA package or an EPF6024A device in a 256-pin FineLine BGA package.

The Altera software packages provide support to design PCBs with SameFrame pin-out devices. Devices can be defined for present and future use. The Altera software packages generate pin-outs describing how to lay out a board to take advantage of this migration (see Figure 15).

Figure 15. SameFrame Pin-Out Example

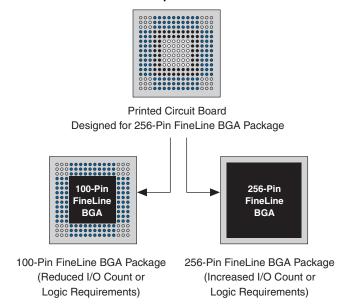


Table 6 lists the 3.3-V FLEX 6000 devices with the Same Frame pin-out feature.

Table 6. 3.3-V FLEX 6000 Devices with SameFrame Pin-Outs							
Device 100-Pin FineLine BGA 256-Pin FineLine BGA							
EPF6016A	V	v					
EPF6024A		V					

# Output Configuration

This section discusses slew-rate control, the MultiVolt I/O interface, power sequencing, and hot-socketing for FLEX 6000 devices.

### **Slew-Rate Control**

The output buffer in each IOE has an adjustable output slew-rate that can be configured for low-noise or high-speed performance. A slower slew-rate reduces system noise and adds a maximum delay of 6.8 ns. The fast slew-rate should be used for speed-critical outputs in systems that are adequately protected against noise. Designers can specify the slew-rate on a pin-by-pin basis during design entry or assign a default slew rate to all pins on a device-wide basis. The slew-rate setting affects only the falling edge of the output.

Open-drain output pins on 5.0-V or 3.3-V FLEX 6000 devices (with a pull-up resistor to the 5.0-V supply) can drive 5.0-V CMOS input pins that require a  $V_{\rm IH}$  of 3.5 V. When the open-drain pin is active, it will drive low. When the pin is inactive, the trace will be pulled up to 5.0 V by the resistor. The open-drain pin will only drive low or tri-state; it will never drive high. The rise time is dependent on the value of the pull-up resistor and load impedance. The  $I_{\rm OL}$  current specification should be considered when selecting a pull-up resistor.

Output pins on 5.0-V FLEX 6000 devices with  $V_{CCIO}$  = 3.3 V or 5.0 V (with a pull-up resistor to the 5.0-V supply) can also drive 5.0-V CMOS input pins. In this case, the pull-up transistor will turn off when the pin voltage exceeds 3.3 V. Therefore, the pin does not have to be open-drain.

## **Power Sequencing & Hot-Socketing**

Because FLEX 6000 family devices can be used in a mixed-voltage environment, they have been designed specifically to tolerate any possible power-up sequence. The  $\rm V_{CCIO}$  and  $\rm V_{CCINT}$  power planes can be powered in any order.

Signals can be driven into 3.3-V FLEX 6000 devices before and during power up without damaging the device. Additionally, FLEX 6000 devices do not drive out during power up. Once operating conditions are reached, FLEX 6000 devices operate as specified by the user.

# IEEE Std. 1149.1 (JTAG) Boundary-Scan Support

All FLEX 6000 devices provide JTAG BST circuitry that comply with the IEEE Std. 1149.1-1990 specification. Table 8 shows JTAG instructions for FLEX 6000 devices. JTAG BST can be performed before or after configuration, but not during configuration (except when you disable JTAG support in user mode).

See Application Note 39 (IEEE 1149.1 (JTAG) Boundary-Scan Testing in Altera Devices) for more information on JTAG BST circuitry.

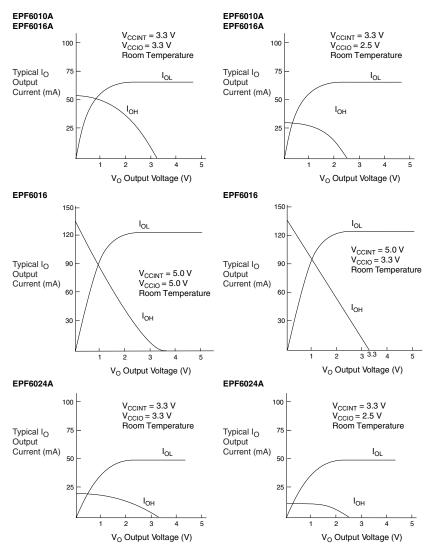
Table 8. FLEX 6000	Table 8. FLEX 6000 JTAG Instructions				
JTAG Instruction	Description				
SAMPLE/PRELOAD	Allows a snapshot of the signals at the device pins to be captured and examined during normal device operation, and permits an initial data pattern to be output at the device pins.				
EXTEST	Allows the external circuitry and board-level interconnections to be tested by forcing a test pattern at the output pins and capturing test result at the input pins.				
BYPASS	Places the 1-bit bypass register between the TDI and TDO pins, which allows the BST data to pass synchronously through the selected device to adjacent devices during normal device operation.				

Table 1	Table 15. FLEX 6000 3.3-V Device Absolute Maximum Ratings Note (1)					
Symbol	Parameter	Conditions	Min	Max	Unit	
V <sub>CC</sub>	Supply voltage	With respect to ground (2)	-0.5	4.6	V	
V <sub>I</sub>	DC input voltage		-2.0	5.75	٧	
I <sub>OUT</sub>	DC output current, per pin		-25	25	mA	
T <sub>STG</sub>	Storage temperature	No bias	-65	150	° C	
T <sub>AMB</sub>	Ambient temperature	Under bias	-65	135	° C	
T <sub>J</sub>	Junction temperature	PQFP, PLCC, and BGA packages		135	° C	

Table 1	6. FLEX 6000 3.3-V Device Rec	ommended Operating Condition	ons		
Symbol	Parameter	Conditions	Min	Max	Unit
V <sub>CCINT</sub>	Supply voltage for internal logic and input buffers	(3), (4)	3.00 (3.00)	3.60 (3.60)	V
V <sub>CCIO</sub>	Supply voltage for output buffers, 3.3-V operation	(3), (4)	3.00 (3.00)	3.60 (3.60)	V
	Supply voltage for output buffers, 2.5-V operation	(3), (4)	2.30 (2.30)	2.70 (2.70)	V
VI	Input voltage		-0.5	5.75	٧
V <sub>O</sub>	Output voltage		0	V <sub>CCIO</sub>	٧
$T_J$	Operating temperature	For commercial use	0	85	° C
		For industrial use	-40	100	°C
t <sub>R</sub>	Input rise time			40	ns
t <sub>F</sub>	Input fall time			40	ns

Figure 18 shows the typical output drive characteristics of 5.0-V and 3.3-V FLEX 6000 devices with 5.0-V, 3.3-V, and 2.5-V  $V_{\rm CCIO}$ . When  $V_{\rm CCIO}=5.0$  V on EPF6016 devices, the output driver is compliant with the *PCI Local Bus Specification, Revision* 2.2 for 5.0-V operation. When  $V_{\rm CCIO}=3.3$  V on the EPF6010A and EPF6016A devices, the output driver is compliant with the *PCI Local Bus Specification, Revision* 2.2 for 3.3-V operation.

Figure 18. Output Drive Characteristics



## **Timing Model**

The continuous, high-performance FastTrack Interconnect routing resources ensure predictable performance and accurate simulation and timing analysis. This predictable performance contrasts with that of FPGAs, which use a segmented connection scheme and therefore have unpredictable performance.

Device performance can be estimated by following the signal path from a source, through the interconnect, to the destination. For example, the registered performance between two LEs on the same row can be calculated by adding the following parameters:

- LE register clock-to-output delay ( $t_{CO} + t_{REG\_TO\_OUT}$ )
- Routing delay  $(t_{ROW} + t_{LOCAL})$
- LE LUT delay ( $t_{DATA\_TO\_REG}$ )
- LE register setup time  $(t_{SU})$

The routing delay depends on the placement of the source and destination LEs. A more complex registered path may involve multiple combinatorial LEs between the source and destination LEs.

Timing simulation and delay prediction are available with the Simulator and Timing Analyzer, or with industry-standard EDA tools. The Simulator offers both pre-synthesis functional simulation to evaluate logic design accuracy and post-synthesis timing simulation with 0.1-ns resolution. The Timing Analyzer provides point-to-point timing delay information, setup and hold time analysis, and device-wide performance analysis.

Figure 19 shows the overall timing model, which maps the possible routing paths to and from the various elements of the FLEX 6000 device.

Symbol	Parameter	Conditions	
t <sub>OD1</sub>	Output buffer and pad delay, slow slew rate = off, V <sub>CCIO</sub> = V <sub>CCINT</sub>	C1 = 35 pF (2)	
t <sub>OD2</sub>	Output buffer and pad delay, slow slew rate = off, V <sub>CCIO</sub> = low voltage	C1 = 35 pF (3)	
t <sub>OD3</sub>	Output buffer and pad delay, slow slew rate = on	C1 = 35 pF (4)	
$t_{XZ}$	Output buffer disable delay	C1 = 5 pF	
t <sub>ZX1</sub>	Output buffer enable delay, slow slew rate = off, V <sub>CCIO</sub> = V <sub>CCINT</sub>	C1 = 35 pF (2)	
$t_{ZX2}$	Output buffer enable delay, slow slew rate = off, V <sub>CCIO</sub> = low voltage	C1 = 35 pF (3)	
t <sub>ZX3</sub>	IOE output buffer enable delay, slow slew rate = on	C1 = 35 pF (4)	
t <sub>IOE</sub>	Output enable control delay		
t <sub>IN</sub>	Input pad and buffer to FastTrack Interconnect delay		
t <sub>IN_DELAY</sub>	Input pad and buffer to FastTrack Interconnect delay with additional delay turned on		

Table 21. Int	Table 21. Interconnect Timing Microparameters Note (1)					
Symbol	Parameter	Conditions				
t <sub>LOCAL</sub>	LAB local interconnect delay					
t <sub>ROW</sub>	Row interconnect routing delay	(5)				
t <sub>COL</sub>	Column interconnect routing delay	(5)				
t <sub>DIN_D</sub>	Dedicated input to LE data delay	(5)				
t <sub>DIN_C</sub>	Dedicated input to LE control delay					
t <sub>LEGLOBAL</sub>	LE output to LE control via internally-generated global signal delay	(5)				
t <sub>LABCARRY</sub>	Routing delay for the carry-out of an LE driving the carry-in signal of a different LE in a different LAB					
t <sub>LABCASC</sub>	Routing delay for the cascade-out signal of an LE driving the cascade-in signal of a different LE in a different LAB					

Table 22. Ex	ternal Reference Timing Parameters	
Symbol	Parameter	Conditions
t <sub>1</sub>	Register-to-register test pattern	(6)
t <sub>DRR</sub>	Register-to-register delay via 4 LEs, 3 row interconnects, and 4 local interconnects	(7)

Table 23. Ex	ternal Timing Parameters	
Symbol	Parameter	Conditions
t <sub>INSU</sub>	Setup time with global clock at LE register	(8)
t <sub>INH</sub>	Hold time with global clock at LE register	(8)
t <sub>оитсо</sub>	Clock-to-output delay with global clock with LE register using FastFLEX I/O pin	(8)

#### *Notes to tables:*

- Microparameters are timing delays contributed by individual architectural elements and cannot be measured explicitly.
- (2) Operating conditions:
  - $\hat{V_{CCIO}} = \widecheck{5}.0~V \pm 5\%$  for commercial use in 5.0-V FLEX 6000 devices.
  - $V_{CCIO} = 5.0 \text{ V} \pm 10\%$  for industrial use in 5.0-V FLEX 6000 devices.
  - $V_{CCIO} = 3.3 \text{ V} \pm 10\%$  for commercial or industrial use in 3.3-V FLEX 6000 devices.
- (3) Operating conditions:
  - $\hat{V_{CCIO}} = 3.3 \text{ V} \pm 10\%$  for commercial or industrial use in 5.0-V FLEX 6000 devices.
  - $V_{CCIO}$  = 2.5 V ±0.2 V for commercial or industrial use in 3.3-V FLEX 6000 devices.
- (4) Operating conditions:
  - $V_{\text{CCIO}} = 2.5 \text{ V}, 3.3 \text{ V}, \text{ or } 5.0 \text{ V}.$
- (5) These parameters are worst-case values for typical applications. Post-compilation timing simulation and timing analysis are required to determine actual worst-case performance.
- (6) This timing parameter shows the delay of a register-to-register test pattern and is used to determine speed grades. There are 12 LEs, including source and destination registers. The row and column interconnects between the registers vary in length.
- 7) This timing parameter is shown for reference and is specified by characterization.
- (8) This timing parameter is specified by characterization.

Tables 24 through 28 show the timing information for EPF6010A and EPF6016A devices.

Parameter			Speed	Grade			Unit
	-	1	-	-2		-3	
	Min	Max	Min	Max	Min	Max	
treg_to_reg		1.2		1.3		1.7	ns
t <sub>CASC_TO_REG</sub>		0.9		1.0		1.2	ns
t <sub>CARRY_TO_REG</sub>		0.9		1.0		1.2	ns
t <sub>DATA_TO_REG</sub>		1.1		1.2		1.5	ns
t <sub>CASC_TO_OUT</sub>		1.3		1.4		1.8	ns
t <sub>CARRY_TO_OUT</sub>		1.6		1.8		2.3	ns
<sup>t</sup> DATA_TO_OUT		1.7		2.0		2.5	ns
t <sub>REG_TO_OUT</sub>		0.4		0.4		0.5	ns
t <sub>SU</sub>	0.9		1.0		1.3		ns
t <sub>H</sub>	1.4		1.7		2.1		ns

Parameter	Speed Grade							
	-	1		2	-3		1	
	Min	Max	Min	Max	Min	Max		
t <sub>co</sub>		0.3		0.4		0.4	ns	
t <sub>CLR</sub>		0.4		0.4		0.5	ns	
t <sub>C</sub>		1.8		2.1		2.6	ns	
t <sub>LD_CLR</sub>		1.8		2.1		2.6	ns	
tCARRY_TO_CARRY		0.1		0.1		0.1	ns	
tREG_TO_CARRY		1.6		1.9		2.3	ns	
tDATA_TO_CARRY		2.1		2.5		3.0	ns	
tCARRY_TO_CASC		1.0		1.1		1.4	ns	
t <sub>CASC_TO_CASC</sub>		0.5		0.6		0.7	ns	
tREG_TO_CASC		1.4		1.7		2.1	ns	
t <sub>DATA_TO_CASC</sub>		1.1		1.2		1.5	ns	
<sup>t</sup> ch	2.5		3.0		3.5		ns	
<sup>t</sup> CL	2.5		3.0		3.5		ns	

Parameter	Speed Grade							
	-	-1		-2		3	1	
	Min	Max	Min	Max	Min	Max		
t <sub>OD1</sub>		1.9		2.2		2.7	ns	
t <sub>OD2</sub>		4.1		4.8		5.8	ns	
t <sub>OD3</sub>		5.8		6.8		8.3	ns	
$t_{XZ}$		1.4		1.7		2.1	ns	
t <sub>XZ1</sub>		1.4		1.7		2.1	ns	
t <sub>XZ2</sub>		3.6		4.3		5.2	ns	
t <sub>XZ3</sub>		5.3		6.3		7.7	ns	
t <sub>IOE</sub>		0.5		0.6		0.7	ns	
t <sub>IN</sub>		3.6		4.1		5.1	ns	
tin delay		4.8		5.4		6.7	ns	

Parameter	Speed Grade						
	-2		-	3	=		
	Min	Max	Min	Max			
OD3		4.7		5.2	ns		
XZ		2.3		2.8	ns		
ZX1		2.3		2.8	ns		
ZX2		4.6		5.1	ns		
ZX3		4.7		5.2	ns		
IOE		0.5		0.6	ns		
<sup>t</sup> in		3.3		4.0	ns		
t <sub>IN DELAY</sub>		4.6		5.6	ns		

Parameter	Speed Grade						
	-2		-	3			
	Min	Max	Min	Max			
t <sub>LOCAL</sub>		0.8		1.0	ns		
t <sub>ROW</sub>		2.9		3.3	ns		
t <sub>COL</sub>		2.3		2.5	ns		
t <sub>DIN_D</sub>		4.9		6.0	ns		
t <sub>DIN_C</sub>		4.8		6.0	ns		
t <sub>LEGLOBAL</sub>		3.1		3.9	ns		
t <sub>LABCARRY</sub>		0.4		0.5	ns		
t <sub>LABCASC</sub>		0.8		1.0	ns		

Table 32. External Reference Timing Parameters for EPF6016 Devices							
Parameter		Unit					
	Min	Max	Min	Max			
t <sub>1</sub>		53.0		65.0	ns		
t <sub>DRR</sub>		16.0		20.0	ns		

Table 33. External Timing Parameters for EPF6016 Devices								
Parameter		Speed Grade						
		-2						
	Min	Max	Min	Max				
t <sub>INSU</sub>	3.2		4.1		ns			
t <sub>INH</sub>	0.0		0.0		ns			
t <sub>оитсо</sub>	2.0	7.9	2.0	9.9	ns			

Tables 34 through 38 show the timing information for EPF6024A devices.

Parameter	Speed Grade							
	-1		-2		-3			
	Min	Max	Min	Max	Min	Max		
t <sub>REG_TO_REG</sub>		1.2		1.3		1.6	ns	
t <sub>CASC_TO_REG</sub>		0.7		0.8		1.0	ns	
t <sub>CARRY_TO_REG</sub>		1.6		1.8		2.2	ns	
t <sub>DATA_TO_REG</sub>		1.3		1.4		1.7	ns	
t <sub>CASC_TO_OUT</sub>		1.2		1.3		1.6	ns	
t <sub>CARRY_TO_OUT</sub>		2.0		2.2		2.6	ns	
t <sub>DATA_TO_OUT</sub>		1.8		2.1		2.6	ns	
t <sub>REG_TO_OUT</sub>		0.3		0.3		0.4	ns	
t <sub>SU</sub>	0.9		1.0		1.2		ns	
t <sub>H</sub>	1.3		1.4		1.7		ns	
$t_{CO}$		0.2		0.3		0.3	ns	
t <sub>CLR</sub>		0.3		0.3		0.4	ns	
$t_C$		1.9		2.1		2.5	ns	
t <sub>LD_CLR</sub>		1.9		2.1		2.5	ns	
t <sub>CARRY_TO_CARRY</sub>		0.2		0.2		0.3	ns	
t <sub>REG_TO_CARRY</sub>		1.4		1.6		1.9	ns	
t <sub>DATA_TO_CARRY</sub>		1.3	_	1.4		1.7	ns	
t <sub>CARRY_TO_CASC</sub>		1.1		1.2		1.4	ns	
t <sub>CASC_TO_CASC</sub>		0.7		0.8		1.0	ns	
t <sub>REG_TO_CASC</sub>		1.4		1.6		1.9	ns	
t <sub>DATA_TO_CASC</sub>		1.0		1.1		1.3	ns	
t <sub>CH</sub>	2.5		3.0		3.5		ns	
t <sub>CL</sub>	2.5		3.0		3.5		ns	

Parameter	Speed Grade							
	-1		-2		-3		1	
	Min	Max	Min	Max	Min	Max		
t <sub>OD1</sub>		1.9		2.1		2.5	ns	
t <sub>OD2</sub>		4.0		4.4		5.3	ns	
t <sub>OD3</sub>		7.0		7.8		9.3	ns	
$t_{XZ}$		4.3		4.8		5.8	ns	
$t_{XZ1}$		4.3		4.8		5.8	ns	
t <sub>XZ2</sub>		6.4		7.1		8.6	ns	
t <sub>XZ3</sub>		9.4		10.5		12.6	ns	
<sup>t</sup> IOE		0.5		0.6		0.7	ns	
İN		3.3		3.7		4.4	ns	
t <sub>IN DELAY</sub>		5.3		5.9		7.0	ns	

Parameter	Speed Grade							
	-1		-2		-3			
	Min	Max	Min	Max	Min	Max	1	
t <sub>LOCAL</sub>		0.8		0.8		1.1	ns	
t <sub>ROW</sub>		3.0		3.1		3.3	ns	
t <sub>COL</sub>		3.0		3.2		3.4	ns	
t <sub>DIN_D</sub>		5.4		5.6		6.2	ns	
t <sub>DIN_C</sub>		4.6		5.1		6.1	ns	
t <sub>LEGLOBAL</sub>		3.1		3.5		4.3	ns	
t <sub>LABCARRY</sub>		0.6		0.7		0.8	ns	
t <sub>LABCASC</sub>		0.3		0.3		0.4	ns	

Table 37. External Reference Timing Parameters for EPF6024A Devices									
Parameter		Speed Grade Unit							
	-1 -2 -3								
	Min	Max	Min	Max	Min	Max			
t <sub>1</sub>		45.0		50.0		60.0	ns		

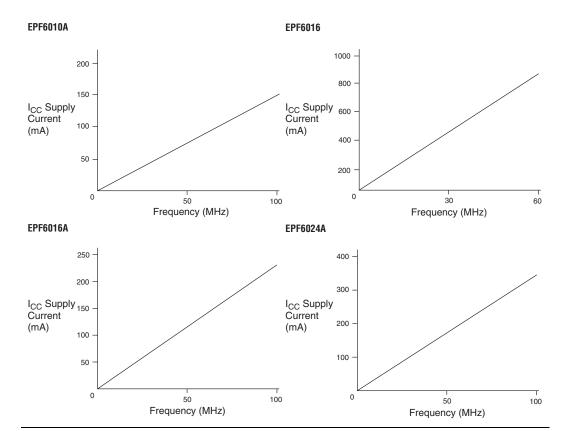


Figure 20. I<sub>CCACTIVE</sub> vs. Operating Frequency

# Device Configuration & Operation

The FLEX 6000 architecture supports several configuration schemes to load a design into the device(s) on the circuit board. This section summarizes the device operating modes and available device configuration schemes.

See Application Note 116 (Configuring APEX 20K, FLEX 10K & FLEX 6000 Devices) for detailed information on configuring FLEX 6000 devices, including sample schematics, timing diagrams, configuration options, pins names, and timing parameters.

# Device Pin-Outs

See the Altera web site (http://www.altera.com) or the *Altera Digital Library* for pin-out information.