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### **Understanding Embedded - FPGAs (Field Programmable Gate Array)**

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

### **Applications of Embedded - FPGAs**

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

#### **Details**

Product Status	Active
Number of LABs/CLBs	196
Number of Logic Elements/Cells	1960
Total RAM Bits	-
Number of I/O	218
Number of Gates	24000
Voltage - Supply	3V ~ 3.6V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 100°C (TJ)
Package / Case	256-BBGA
Supplier Device Package	256-BGA (27x27)
Purchase URL	<a href="https://www.e-xfl.com/pro/item?MUrl=&amp;PartUrl=epf6024abi256-2">https://www.e-xfl.com/pro/item?MUrl=&amp;PartUrl=epf6024abi256-2</a>

## ...and More Features

- Powerful I/O pins
  - Individual tri-state output enable control for each pin
  - Programmable output slew-rate control to reduce switching noise
  - Fast path from register to I/O pin for fast clock-to-output time
- Flexible interconnect
  - FastTrack<sup>®</sup> Interconnect continuous routing structure for fast, predictable interconnect delays
  - Dedicated carry chain that implements arithmetic functions such as fast adders, counters, and comparators (automatically used by software tools and megafunctions)
  - Dedicated cascade chain that implements high-speed, high-fan-in logic functions (automatically used by software tools and megafunctions)
  - Tri-state emulation that implements internal tri-state networks
  - Four low-skew global paths for clock, clear, preset, or logic signals
- Software design support and automatic place-and-route provided by Altera’s development system for Windows-based PCs, Sun SPARCstations, and HP 9000 Series 700/800
- Flexible package options
  - Available in a variety of packages with 100 to 256 pins, including the innovative FineLine BGA<sup>™</sup> packages (see [Table 2](#))
  - SameFrame<sup>™</sup> pin-compatibility (with other FLEX<sup>®</sup> 6000 devices) across device densities and pin counts
  - Thin quad flat pack (TQFP), plastic quad flat pack (PQFP), and ball-grid array (BGA) packages (see [Table 2](#))
  - Footprint- and pin-compatibility with other FLEX 6000 devices in the same package
- Additional design entry and simulation support provided by EDIF 2.0.0 and 3.0.0 netlist files, the library of parameterized modules (LPM), Verilog HDL, VHDL, DesignWare components, and other interfaces to popular EDA tools from manufacturers such as Cadence, Exemplar Logic, Mentor Graphics, OrCAD, Synopsys, Synplicity, VeriBest, and Viewlogic

Device	100-Pin TQFP	100-Pin FineLine BGA	144-Pin TQFP	208-Pin PQFP	240-Pin PQFP	256-Pin BGA	256-pin FineLine BGA
EPF6010A	71		102				
EPF6016			117	171	199	204	
EPF6016A	81	81	117	171			171
EPF6024A			117	171	199	218	219

## General Description

The Altera® FLEX 6000 programmable logic device (PLD) family provides a low-cost alternative to high-volume gate array designs. FLEX 6000 devices are based on the OptiFLEX architecture, which minimizes die size while maintaining high performance and routability. The devices have reconfigurable SRAM elements, which give designers the flexibility to quickly change their designs during prototyping and design testing. Designers can also change functionality during operation via in-circuit reconfiguration.

FLEX 6000 devices are reprogrammable, and they are 100% tested prior to shipment. As a result, designers are not required to generate test vectors for fault coverage purposes, allowing them to focus on simulation and design verification. In addition, the designer does not need to manage inventories of different gate array designs. FLEX 6000 devices are configured on the board for the specific functionality required.

Table 3 shows FLEX 6000 performance for some common designs. All performance values shown were obtained using Synopsys DesignWare or LPM functions. Special design techniques are not required to implement the applications; the designer simply infers or instantiates a function in a Verilog HDL, VHDL, Altera Hardware Description Language (AHDL), or schematic design file.

Application	LEs Used	Performance			Units
		-1 Speed Grade	-2 Speed Grade	-3 Speed Grade	
16-bit loadable counter	16	172	153	133	MHz
16-bit accumulator	16	172	153	133	MHz
24-bit accumulator	24	136	123	108	MHz
16-to-1 multiplexer (pin-to-pin) (1)	10	12.1	13.4	16.6	ns
16 × 16 multiplier with a 4-stage pipeline	592	84	67	58	MHz

**Note:**

(1) This performance value is measured as a pin-to-pin delay.

## Functional Description

The FLEX 6000 OptiFLEX architecture consists of logic elements (LEs). Each LE includes a 4-input look-up table (LUT), which can implement any 4-input function, a register, and dedicated paths for carry and cascade chain functions. Because each LE contains a register, a design can be easily pipelined without consuming more LEs. The specified gate count for FLEX 6000 devices includes all LUTs and registers.

LEs are combined into groups called logic array blocks (LABs); each LAB contains 10 LEs. The Altera software automatically places related LEs into the same LAB, minimizing the number of required interconnects. Each LAB can implement a medium-sized block of logic, such as a counter or multiplexer.

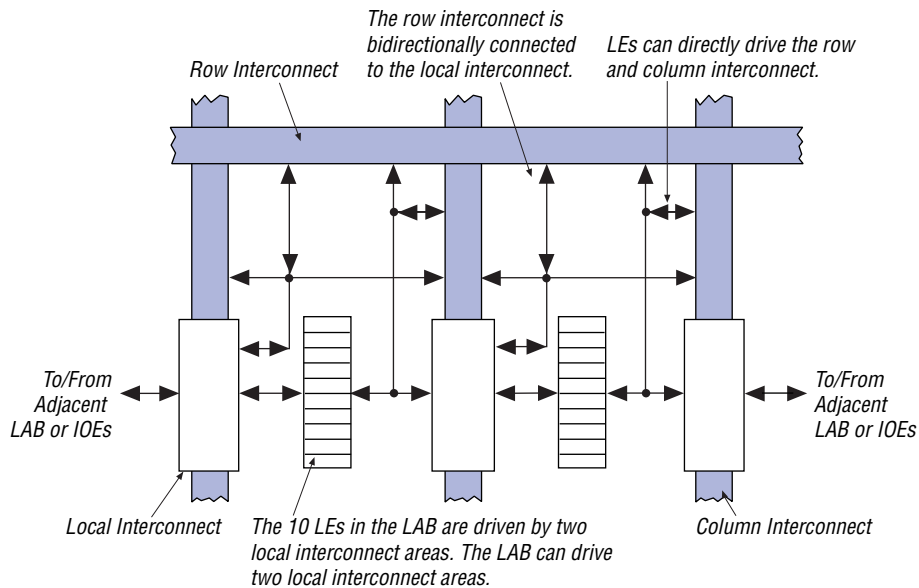
Signal interconnections within FLEX 6000 devices—and to and from device pins—are provided via the routing structure of the FastTrack Interconnect. The routing structure is a series of fast, continuous row and column channels that run the entire length and width of the device. Any LE or pin can feed or be fed by any other LE or pin via the FastTrack Interconnect. See “FastTrack Interconnect” on [page 17](#) of this data sheet for more information.

Each I/O pin is fed by an I/O element (IOE) located at the end of each row and column of the FastTrack Interconnect. Each IOE contains a bidirectional I/O buffer. Each IOE is placed next to an LAB, where it can be driven by the local interconnect of that LAB. This feature allows fast clock-to-output times of less than 8 ns when a pin is driven by any of the 10 LEs in the adjacent LAB. Also, any LE can drive any pin via the row and column interconnect. I/O pins can drive the LE registers via the row and column interconnect, providing setup times as low as 2 ns and hold times of 0 ns. IOEs provide a variety of features, such as JTAG BST support, slew-rate control, and tri-state buffers.

[Figure 1](#) shows a block diagram of the FLEX 6000 OptiFLEX architecture. Each group of ten LEs is combined into an LAB, and the LABs are arranged into rows and columns. The LABs are interconnected by the FastTrack Interconnect. IOEs are located at the end of each FastTrack Interconnect row and column.

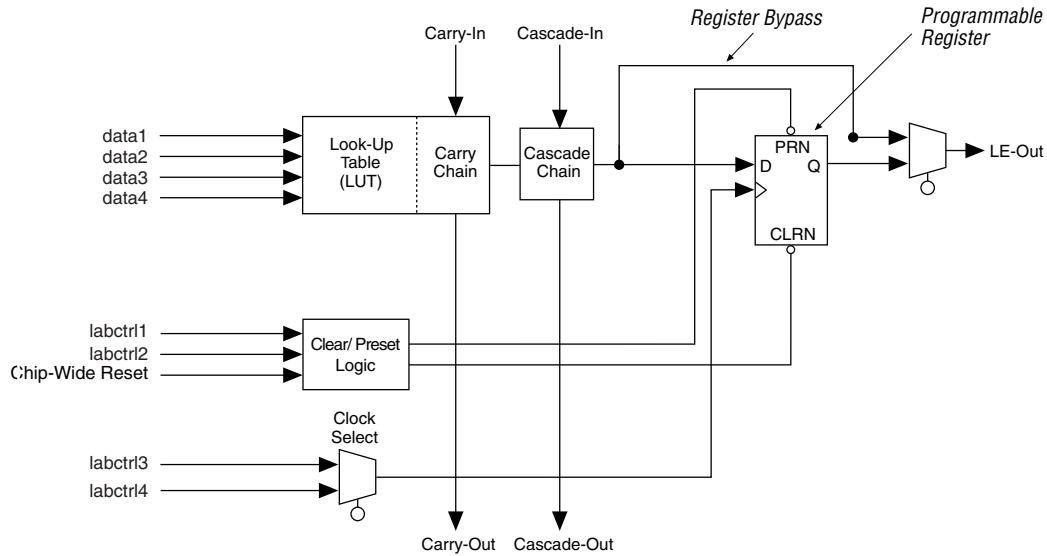
The interleaved LAB structure—an innovative feature of the FLEX 6000 architecture—allows each LAB to drive two local interconnects. This feature minimizes the use of the FastTrack Interconnect, providing higher performance. An LAB can drive 20 LEs in adjacent LABs via the local interconnect, which maximizes fitting flexibility while minimizing die size. See [Figure 2](#).

**Figure 2. Logic Array Block**



In most designs, the registers only use global clock and clear signals. However, in some cases, other clock or asynchronous clear signals are needed. In addition, counters may also have synchronous clear or load signals. In a design that uses non-global clock and clear signals, inputs from the first LE in an LAB are re-routed to drive the control signals for that LAB. See [Figure 3](#).

Figure 4. Logic Element



The programmable flipflop in the LE can be configured for D, T, JK, or SR operation. The clock and clear control signals on the flipflop can be driven by global signals, general-purpose I/O pins, or any internal logic. For combinatorial functions, the flipflop is bypassed and the output of the LUT drives the outputs of the LE. The LE output can drive both the local interconnect and the FastTrack Interconnect.

The FLEX 6000 architecture provides two types of dedicated high-speed data paths that connect adjacent LEs without using local interconnect paths: carry chains and cascade chains. A carry chain supports high-speed arithmetic functions such as counters and adders, while a cascade chain implements wide-input functions such as equivalent comparators with minimum delay. Carry and cascade chains connect LEs 2 through 10 in an LAB and all LABs in the same half of the row. Because extensive use of carry and cascade chains can reduce routing flexibility, these chains should be limited to speed-critical portions of a design.

Figure 5. Carry Chain Operation

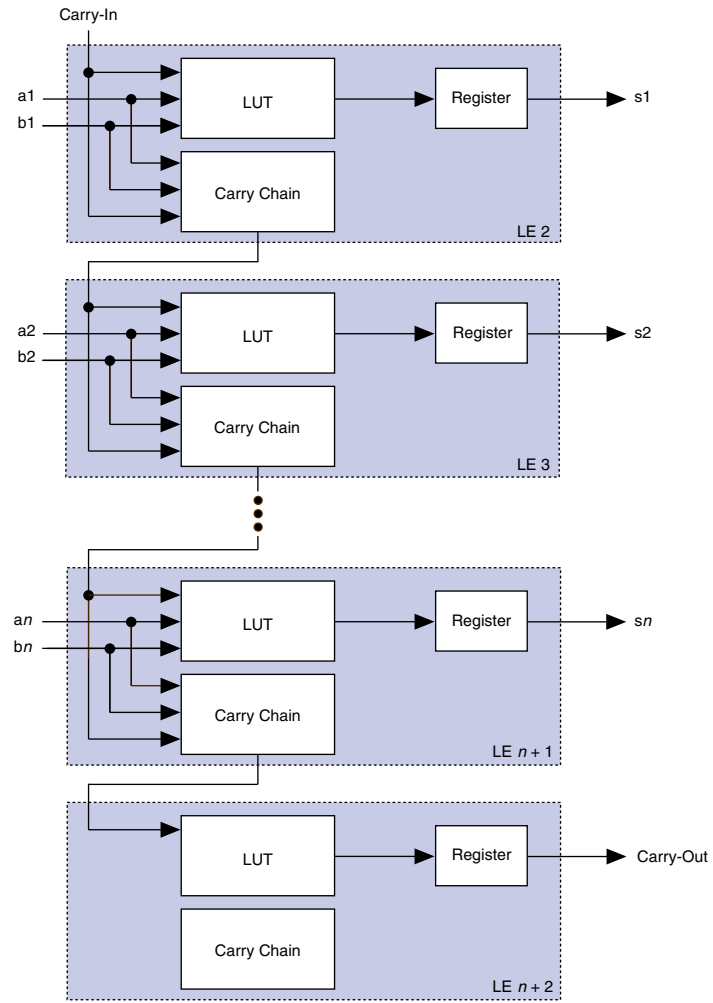
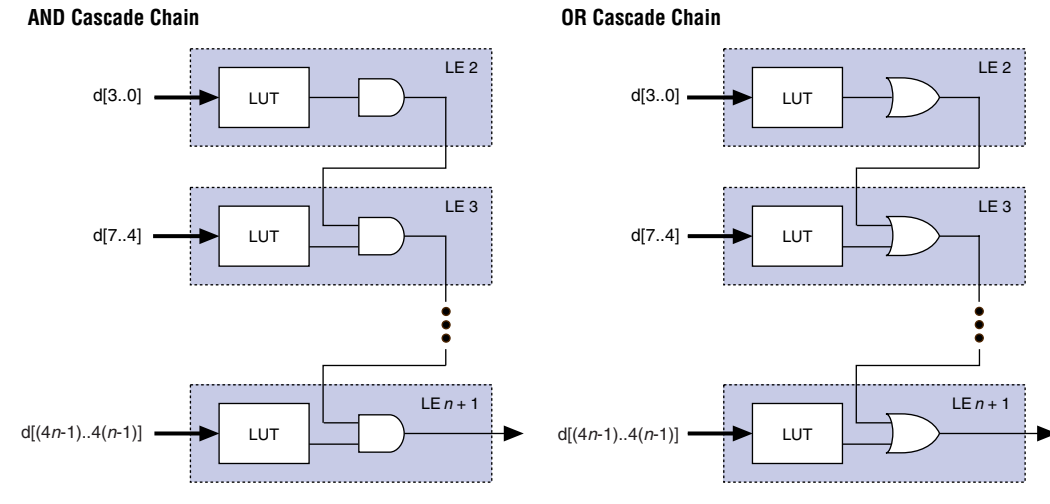


Figure 6. Cascade Chain Operation



### LE Operating Modes

The FLEX 6000 LE can operate in one of the following three modes:

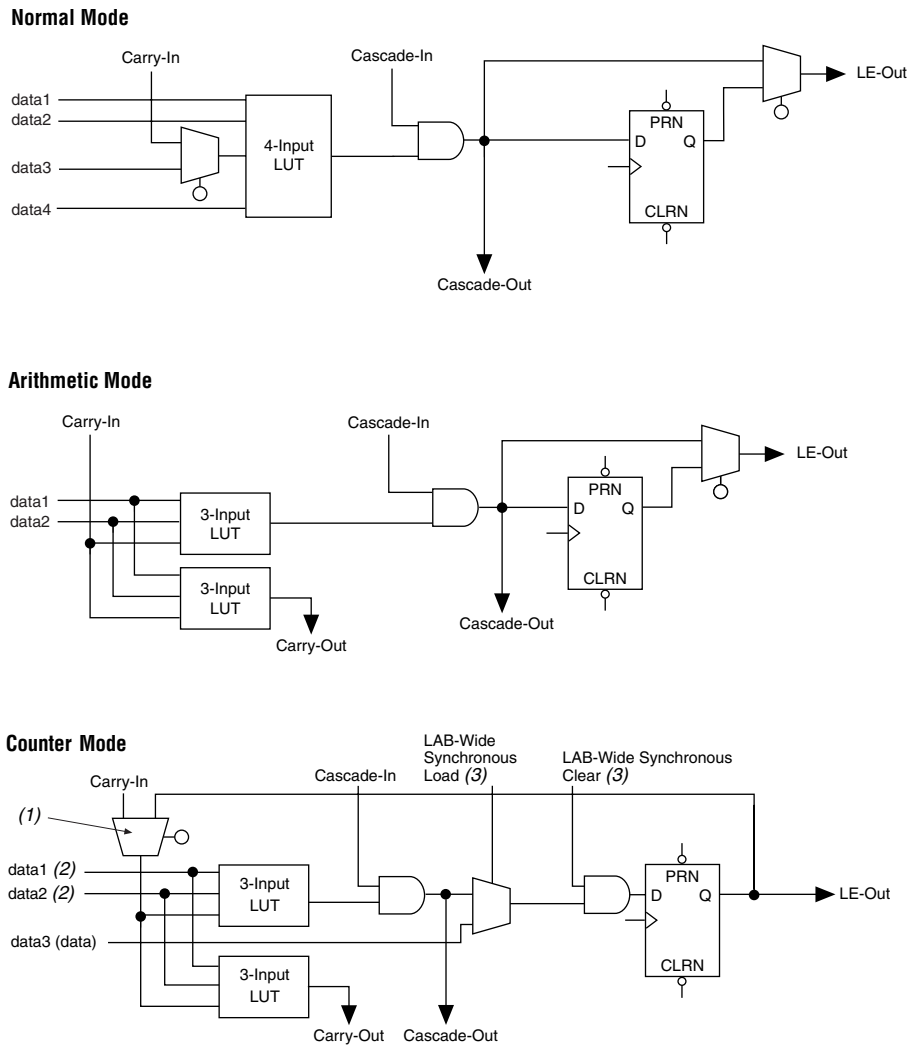
- Normal mode
- Arithmetic mode
- Counter mode

Each of these modes uses LE resources differently. In each mode, seven available inputs to the LE—the four data inputs from the LAB local interconnect, the feedback from the programmable register, and the carry-in and cascade-in from the previous LE—are directed to different destinations to implement the desired logic function. LAB-wide signals provide clock, asynchronous clear, synchronous clear, and synchronous load control for the register. The Altera software, in conjunction with parameterized functions such as LPM and DesignWare functions, automatically chooses the appropriate mode for common functions such as counters, adders, and multipliers. If required, the designer can also create special-purpose functions to use an LE operating mode for optimal performance.

Figure 7 shows the LE operating modes.



Figure 7. LE Operating Modes



**Notes:**

- (1) The register feedback multiplexer is available on LE 2 of each LAB.
- (2) The data1 and data2 input signals can supply a clock enable, up or down control, or register feedback signals for all LEs other than the second LE in an LAB.
- (3) The LAB-wide synchronous clear and LAB-wide synchronous load affect all registers in an LAB.

### Normal Mode

The normal mode is suitable for general logic applications, combinatorial functions, or wide decoding functions that can take advantage of a cascade chain. In normal mode, four data inputs from the LAB local interconnect and the carry-in are inputs to a 4-input LUT. The Altera software automatically selects the carry-in or the DATA3 signal as one of the inputs to the LUT. The LUT output can be combined with the cascade-in signal to form a cascade chain through the cascade-out signal.

### Arithmetic Mode

The arithmetic mode is ideal for implementing adders, accumulators, and comparators. An LE in arithmetic mode uses two 3-input LUTs. One LUT computes a 3-input function; the other generates a carry output. As shown in [Figure 7](#), the first LUT uses the carry-in signal and two data inputs from the LAB local interconnect to generate a combinatorial or registered output. For example, when implementing an adder, this output is the sum of three signals: DATA1, DATA2, and carry-in. The second LUT uses the same three signals to generate a carry-out signal, thereby creating a carry chain. The arithmetic mode also supports simultaneous use of the cascade chain.

The Altera software implements logic functions to use the arithmetic mode automatically where appropriate; the designer does not have to decide how the carry chain will be used.

### Counter Mode

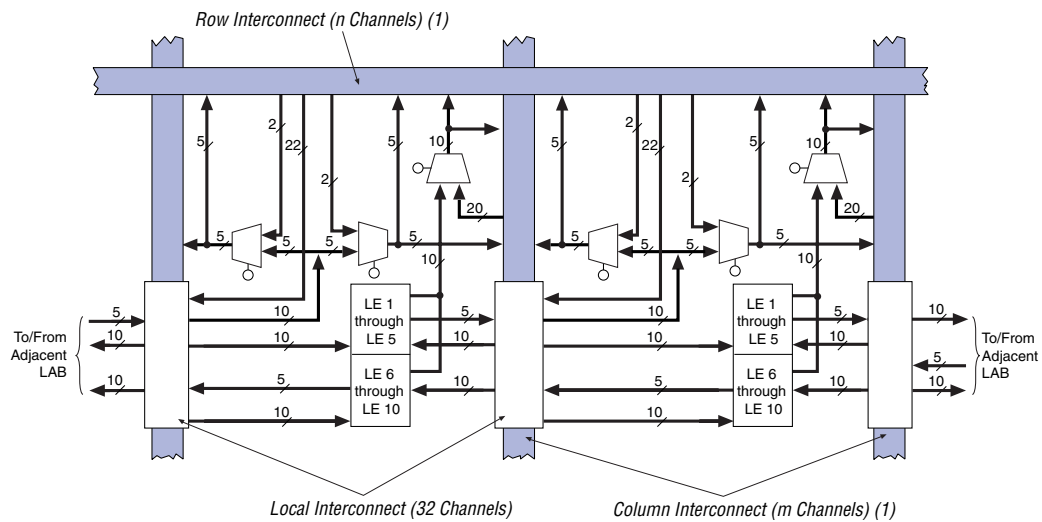
The counter mode offers counter enable, synchronous up/down control, synchronous clear, and synchronous load options. The counter enable and synchronous up/down control signals are generated from the data inputs of the LAB local interconnect. The synchronous clear and synchronous load options are LAB-wide signals that affect all registers in the LAB. Consequently, if any of the LEs in a LAB use counter mode, other LEs in that LAB must be used as part of the same counter or be used for a combinatorial function. In addition, the Altera software automatically places registers that are not in the counter into other LABs.

The counter mode uses two 3-input LUTs: one generates the counter data and the other generates the fast carry bit. A 2-to-1 multiplexer provides synchronous loading, and another AND gate provides synchronous clearing. If the cascade function is used by an LE in counter mode, the synchronous clear or load will override any signal carried on the cascade chain. The synchronous clear overrides the synchronous load.

The FastTrack Interconnect consists of column and row interconnect channels that span the entire device. Each row of LABs is served by a dedicated row interconnect, which routes signals between LABs in the same row, and also routes signals from I/O pins to LABs. Additionally, the local interconnect routes signals between LEs in the same LAB and in adjacent LABs. The column interconnect routes signals between rows and routes signals from I/O pins to rows.

LEs 1 through 5 of an LAB drive the local interconnect to the right, while LEs 6 through 10 drive the local interconnect to the left. The DATA1 and DATA3 inputs of each LE are driven by the local interconnect to the left; DATA2 and DATA4 are driven by the local interconnect to the right. The local interconnect also routes signals from LEs to I/O pins. Figure 9 shows an overview of the FLEX 6000 interconnect architecture. LEs in the first and last columns have drivers on both sides so that all LEs in the LAB can drive I/O pins via the local interconnect.

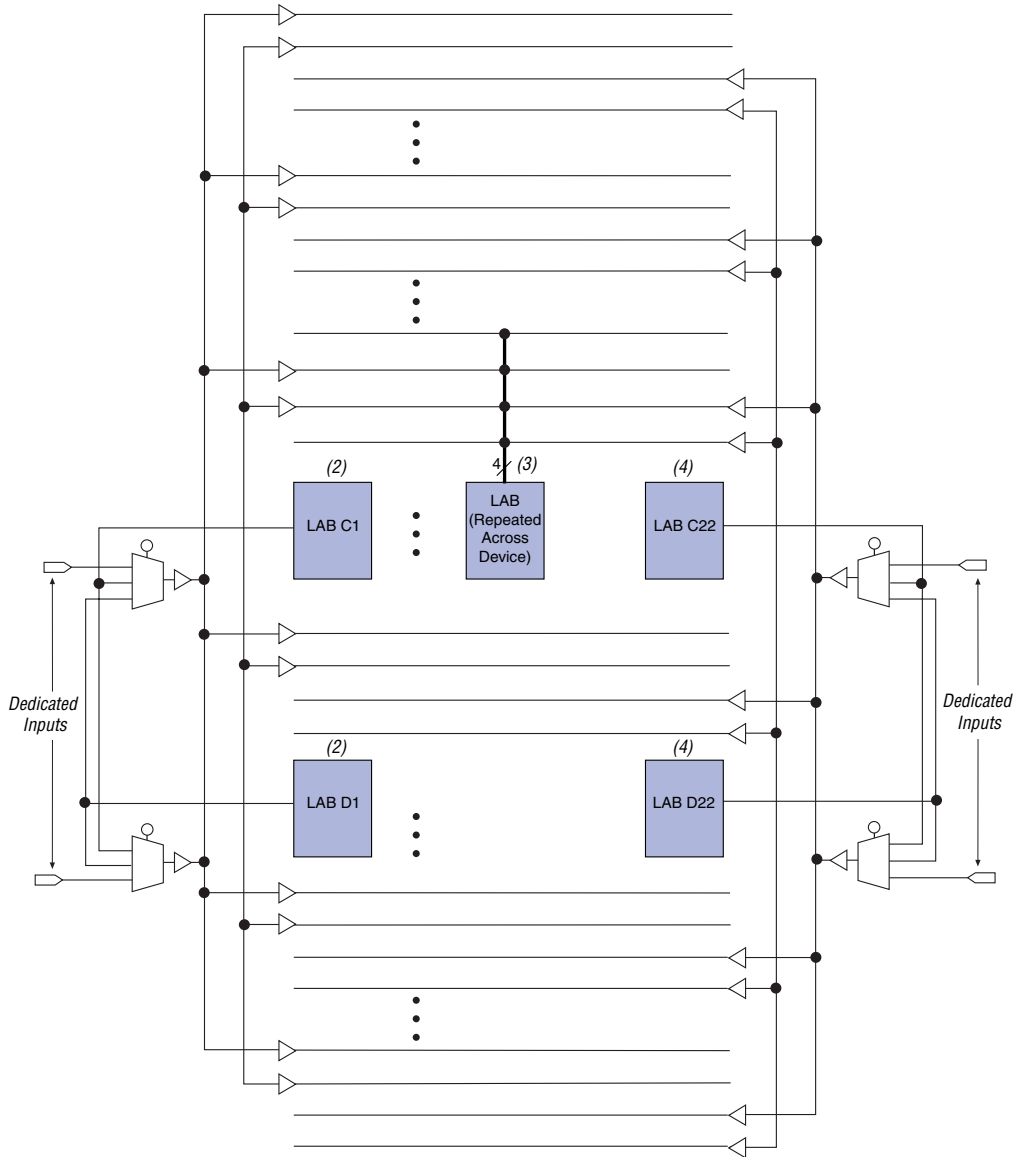
Figure 9. FastTrack Interconnect Architecture



Note:

- (1) For EPF6010A, EPF6016, and EPF6016A devices,  $n = 144$  channels and  $m = 20$  channels; for EPF6024A devices,  $n = 186$  channels and  $m = 30$  channels.

Figure 11. Global Clock & Clear Distribution *Note (1)*



**Notes:**

- (1) The global clock and clear distribution signals are shown for EPF6016 and EPF6016A devices. In EPF6010A devices, LABs in rows B and C drive global signals. In EPF6024A devices, LABs in rows C and E drive global signals.
- (2) The local interconnect from LABs C1 and D1 can drive two global control signals on the left side.
- (3) Global signals drive into every LAB as clock, asynchronous clear, preset, and data signals.
- (4) The local interconnect from LABs C22 and D22 can drive two global control signals on the right side.

Tables 19 through 21 describe the FLEX 6000 internal timing microparameters, which are expressed as worst-case values. Using hand calculations, these parameters can be used to estimate design performance. However, before committing designs to silicon, actual worst-case performance should be modeled using timing simulation and timing analysis. Tables 22 and 23 describe FLEX 6000 external timing parameters.

<b>Table 19. LE Timing Microparameters</b> <i>Note (1)</i>		
<b>Symbol</b>	<b>Parameter</b>	<b>Conditions</b>
$t_{REG\_TO\_REG}$	LUT delay for LE register feedback in carry chain	
$t_{CASC\_TO\_REG}$	Cascade-in to register delay	
$t_{CARRY\_TO\_REG}$	Carry-in to register delay	
$t_{DATA\_TO\_REG}$	LE input to register delay	
$t_{CASC\_TO\_OUT}$	Cascade-in to LE output delay	
$t_{CARRY\_TO\_OUT}$	Carry-in to LE output delay	
$t_{DATA\_TO\_OUT}$	LE input to LE output delay	
$t_{REG\_TO\_OUT}$	Register output to LE output delay	
$t_{SU}$	LE register setup time before clock; LE register recovery time after asynchronous clear	
$t_H$	LE register hold time after clock	
$t_{CO}$	LE register clock-to-output delay	
$t_{CLR}$	LE register clear delay	
$t_C$	LE register control signal delay	
$t_{LD\_CLR}$	Synchronous load or clear delay in counter mode	
$t_{CARRY\_TO\_CARRY}$	Carry-in to carry-out delay	
$t_{REG\_TO\_CARRY}$	Register output to carry-out delay	
$t_{DATA\_TO\_CARRY}$	LE input to carry-out delay	
$t_{CARRY\_TO\_CASC}$	Carry-in to cascade-out delay	
$t_{CASC\_TO\_CASC}$	Cascade-in to cascade-out delay	
$t_{REG\_TO\_CASC}$	Register-out to cascade-out delay	
$t_{DATA\_TO\_CASC}$	LE input to cascade-out delay	
$t_{CH}$	LE register clock high time	
$t_{CL}$	LE register clock low time	

Symbol	Parameter	Conditions
$t_{INSU}$	Setup time with global clock at LE register	(8)
$t_{INH}$	Hold time with global clock at LE register	(8)
$t_{OUTCO}$	Clock-to-output delay with global clock with LE register using FastFLEX I/O pin	(8)

**Notes to tables:**

- (1) Microparameters are timing delays contributed by individual architectural elements and cannot be measured explicitly.
- (2) Operating conditions:  
 $V_{CCIO} = 5.0\text{ V} \pm 5\%$  for commercial use in 5.0-V FLEX 6000 devices.  
 $V_{CCIO} = 5.0\text{ V} \pm 10\%$  for industrial use in 5.0-V FLEX 6000 devices.  
 $V_{CCIO} = 3.3\text{ V} \pm 10\%$  for commercial or industrial use in 3.3-V FLEX 6000 devices.
- (3) Operating conditions:  
 $V_{CCIO} = 3.3\text{ V} \pm 10\%$  for commercial or industrial use in 5.0-V FLEX 6000 devices.  
 $V_{CCIO} = 2.5\text{ V} \pm 0.2\text{ V}$  for commercial or industrial use in 3.3-V FLEX 6000 devices.
- (4) Operating conditions:  
 $V_{CCIO} = 2.5\text{ V}, 3.3\text{ V}, \text{ or } 5.0\text{ V}.$
- (5) These parameters are worst-case values for typical applications. Post-compilation timing simulation and timing analysis are required to determine actual worst-case performance.
- (6) This timing parameter shows the delay of a register-to-register test pattern and is used to determine speed grades. There are 12 LEs, including source and destination registers. The row and column interconnects between the registers vary in length.
- (7) This timing parameter is shown for reference and is specified by characterization.
- (8) This timing parameter is specified by characterization.

Tables 24 through 28 show the timing information for EPF6010A and EPF6016A devices.

Parameter	Speed Grade						Unit
	-1		-2		-3		
	Min	Max	Min	Max	Min	Max	
$t_{REG\_TO\_REG}$		1.2		1.3		1.7	ns
$t_{CASC\_TO\_REG}$		0.9		1.0		1.2	ns
$t_{CARRY\_TO\_REG}$		0.9		1.0		1.2	ns
$t_{DATA\_TO\_REG}$		1.1		1.2		1.5	ns
$t_{CASC\_TO\_OUT}$		1.3		1.4		1.8	ns
$t_{CARRY\_TO\_OUT}$		1.6		1.8		2.3	ns
$t_{DATA\_TO\_OUT}$		1.7		2.0		2.5	ns
$t_{REG\_TO\_OUT}$		0.4		0.4		0.5	ns
$t_{SU}$	0.9		1.0		1.3		ns
$t_{H}$	1.4		1.7		2.1		ns

Parameter	Speed Grade						Unit
	-1		-2		-3		
	Min	Max	Min	Max	Min	Max	
$t_{CO}$		0.3		0.4		0.4	ns
$t_{CLR}$		0.4		0.4		0.5	ns
$t_C$		1.8		2.1		2.6	ns
$t_{LD\_CLR}$		1.8		2.1		2.6	ns
$t_{CARRY\_TO\_CARRY}$		0.1		0.1		0.1	ns
$t_{REG\_TO\_CARRY}$		1.6		1.9		2.3	ns
$t_{DATA\_TO\_CARRY}$		2.1		2.5		3.0	ns
$t_{CARRY\_TO\_CASC}$		1.0		1.1		1.4	ns
$t_{CASC\_TO\_CASC}$		0.5		0.6		0.7	ns
$t_{REG\_TO\_CASC}$		1.4		1.7		2.1	ns
$t_{DATA\_TO\_CASC}$		1.1		1.2		1.5	ns
$t_{CH}$	2.5		3.0		3.5		ns
$t_{CL}$	2.5		3.0		3.5		ns

Parameter	Speed Grade						Unit
	-1		-2		-3		
	Min	Max	Min	Max	Min	Max	
$t_{OD1}$		1.9		2.2		2.7	ns
$t_{OD2}$		4.1		4.8		5.8	ns
$t_{OD3}$		5.8		6.8		8.3	ns
$t_{XZ}$		1.4		1.7		2.1	ns
$t_{XZ1}$		1.4		1.7		2.1	ns
$t_{XZ2}$		3.6		4.3		5.2	ns
$t_{XZ3}$		5.3		6.3		7.7	ns
$t_{IOE}$		0.5		0.6		0.7	ns
$t_{IN}$		3.6		4.1		5.1	ns
$t_{IN\_DELAY}$		4.8		5.4		6.7	ns

**Table 30. IOE Timing Microparameters for EPF6016 Devices**

Parameter	Speed Grade				Unit
	-2		-3		
	Min	Max	Min	Max	
$t_{OD3}$		4.7		5.2	ns
$t_{XZ}$		2.3		2.8	ns
$t_{ZX1}$		2.3		2.8	ns
$t_{ZX2}$		4.6		5.1	ns
$t_{ZX3}$		4.7		5.2	ns
$t_{IOE}$		0.5		0.6	ns
$t_{IN}$		3.3		4.0	ns
$t_{IN\_DELAY}$		4.6		5.6	ns

**Table 31. Interconnect Timing Microparameters for EPF6016 Devices**

Parameter	Speed Grade				Unit
	-2		-3		
	Min	Max	Min	Max	
$t_{LOCAL}$		0.8		1.0	ns
$t_{ROW}$		2.9		3.3	ns
$t_{COL}$		2.3		2.5	ns
$t_{DIN\_D}$		4.9		6.0	ns
$t_{DIN\_C}$		4.8		6.0	ns
$t_{LEGLOBAL}$		3.1		3.9	ns
$t_{LABCARRY}$		0.4		0.5	ns
$t_{LABCASC}$		0.8		1.0	ns

**Table 32. External Reference Timing Parameters for EPF6016 Devices**

Parameter	Speed Grade				Unit
	-2		-3		
	Min	Max	Min	Max	
$t_1$		53.0		65.0	ns
$t_{DRR}$		16.0		20.0	ns



**Table 35. IOE Timing Microparameters for EPF6024A Devices**

Parameter	Speed Grade						Unit
	-1		-2		-3		
	Min	Max	Min	Max	Min	Max	
$t_{OD1}$		1.9		2.1		2.5	ns
$t_{OD2}$		4.0		4.4		5.3	ns
$t_{OD3}$		7.0		7.8		9.3	ns
$t_{XZ}$		4.3		4.8		5.8	ns
$t_{XZ1}$		4.3		4.8		5.8	ns
$t_{XZ2}$		6.4		7.1		8.6	ns
$t_{XZ3}$		9.4		10.5		12.6	ns
$t_{IOE}$		0.5		0.6		0.7	ns
$t_{IN}$		3.3		3.7		4.4	ns
$t_{IN\_DELAY}$		5.3		5.9		7.0	ns

**Table 36. Interconnect Timing Microparameters for EPF6024A Devices**

Parameter	Speed Grade						Unit
	-1		-2		-3		
	Min	Max	Min	Max	Min	Max	
$t_{LOCAL}$		0.8		0.8		1.1	ns
$t_{ROW}$		3.0		3.1		3.3	ns
$t_{COL}$		3.0		3.2		3.4	ns
$t_{DIN\_D}$		5.4		5.6		6.2	ns
$t_{DIN\_C}$		4.6		5.1		6.1	ns
$t_{LEGLOBAL}$		3.1		3.5		4.3	ns
$t_{LABCARRY}$		0.6		0.7		0.8	ns
$t_{LABCASC}$		0.3		0.3		0.4	ns

**Table 37. External Reference Timing Parameters for EPF6024A Devices**

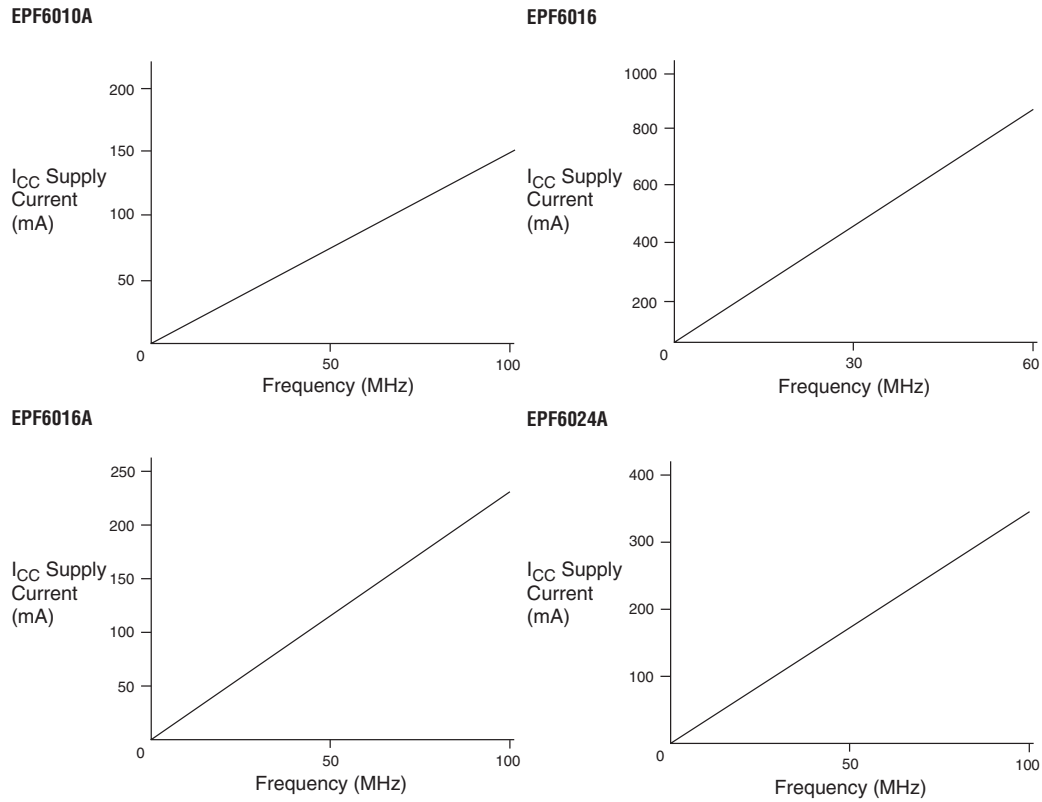
Parameter	Speed Grade						Unit
	-1		-2		-3		
	Min	Max	Min	Max	Min	Max	
$t_1$		45.0		50.0		60.0	ns

This calculation provides an  $I_{CC}$  estimate based on typical conditions with no output load. The actual  $I_{CC}$  should be verified during operation because this measurement is sensitive to the actual pattern in the device and the environmental operating conditions.

To better reflect actual designs, the power model (and the constant K in the power calculation equations shown above) for continuous interconnect FLEX devices assumes that LEs drive FastTrack Interconnect channels. In contrast, the power model of segmented FPGAs assumes that all LEs drive only one short interconnect segment. This assumption may lead to inaccurate results, compared to measured power consumption for an actual design in a segmented interconnect FPGA.

Figure 20 shows the relationship between the current and operating frequency for EPF6010A, EPF6016, EPF6016A, and EPF6024A devices.

Figure 20.  $I_{CCACTIVE}$  vs. Operating Frequency



## Device Configuration & Operation

The FLEX 6000 architecture supports several configuration schemes to load a design into the device(s) on the circuit board. This section summarizes the device operating modes and available device configuration schemes.

- f See [Application Note 116 \(Configuring APEX 20K, FLEX 10K & FLEX 6000 Devices\)](#) for detailed information on configuring FLEX 6000 devices, including sample schematics, timing diagrams, configuration options, pins names, and timing parameters.

## Operating Modes

The FLEX 6000 architecture uses SRAM configuration elements that require configuration data to be loaded every time the circuit powers up. This process of physically loading the SRAM data into a FLEX 6000 device is known as configuration. During initialization—a process that occurs immediately after configuration—the device resets registers, enables I/O pins, and begins to operate as a logic device. The I/O pins are tri-stated during power-up, and before and during configuration. The configuration and initialization processes of a device are referred to as *command mode*; normal device operation is called *user mode*.

SRAM configuration elements allow FLEX 6000 devices to be reconfigured in-circuit by loading new configuration data into the device. Real-time reconfiguration is performed by forcing the device into command mode with a device pin, loading different configuration data, reinitializing the device, and resuming user-mode operation. The entire reconfiguration process requires less than 100 ms and is used to dynamically reconfigure an entire system. Also, in-field system upgrades can be performed by distributing new configuration files.

## Configuration Schemes

The configuration data for a FLEX 6000 device can be loaded with one of three configuration schemes, which is chosen on the basis of the target application. An EPC1 or EPC1441 configuration device or intelligent controller can be used to control the configuration of a FLEX 6000 device, allowing automatic configuration on system power-up.

Multiple FLEX 6000 devices can be configured in any of the three configuration schemes by connecting the configuration enable input (nCE) and configuration enable output (nCEO) pins on each device.

Table 40 shows the data sources for each configuration scheme.

<b>Configuration Scheme</b>	<b>Data Source</b>
Configuration device	EPC1 or EPC1441 configuration device
Passive serial (PS)	BitBlaster™, ByteBlasterMV™, or MasterBlaster™ download cables, or serial data source
Passive serial asynchronous (PSA)	BitBlaster, ByteBlasterMV, or MasterBlaster download cables, or serial data source



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