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### Understanding [Embedded - FPGAs \(Field Programmable Gate Array\)](#)

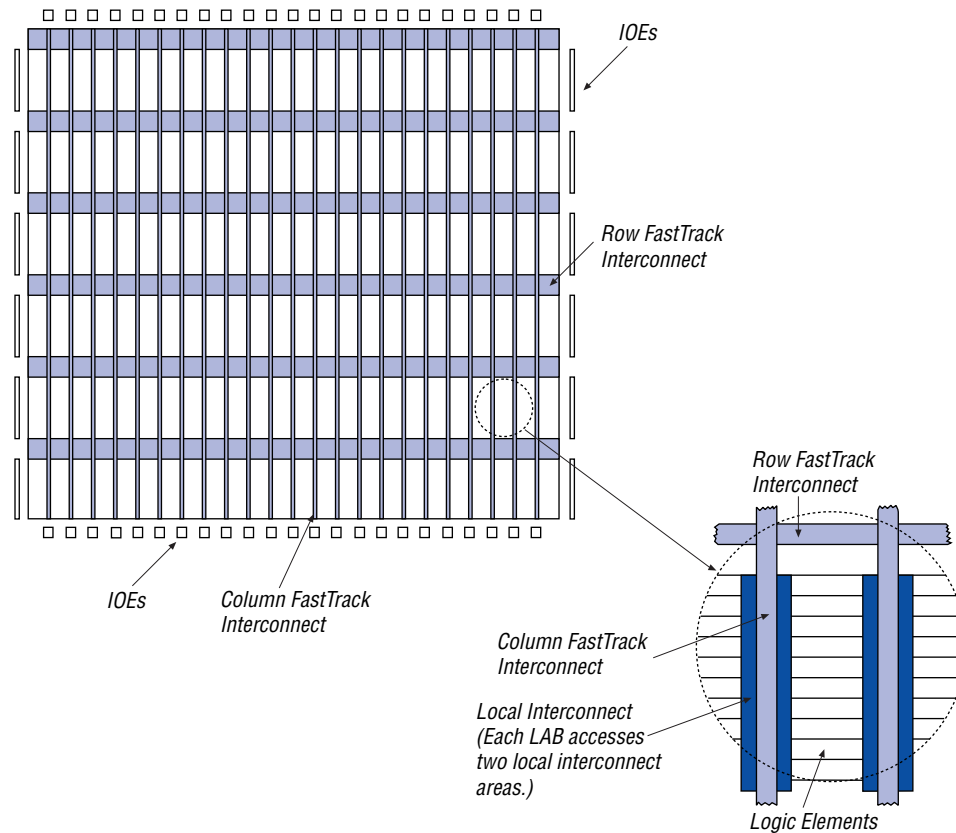
Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

### Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

#### Details

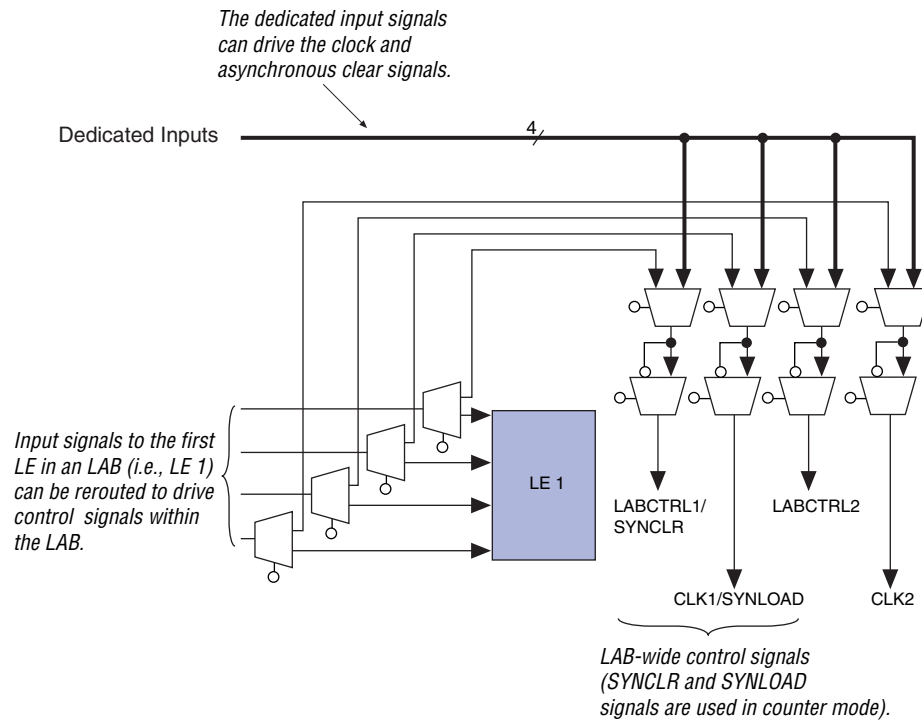
Product Status	Obsolete
Number of LABs/CLBs	196
Number of Logic Elements/Cells	1960
Total RAM Bits	-
Number of I/O	219
Number of Gates	24000
Voltage - Supply	3V ~ 3.6V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	256-BGA
Supplier Device Package	256-FBGA (17x17)
Purchase URL	<a href="https://www.e-xfl.com/product-detail/intel/epf6024afc256-1">https://www.e-xfl.com/product-detail/intel/epf6024afc256-1</a>

**Figure 1. OptiFLEX Architecture Block Diagram**

FLEX 6000 devices provide four dedicated, global inputs that drive the control inputs of the flipflops to ensure efficient distribution of high-speed, low-skew control signals. These inputs use dedicated routing channels that provide shorter delays and lower skews than the FastTrack Interconnect. These inputs can also be driven by internal logic, providing an ideal solution for a clock divider or an internally generated asynchronous clear signal that clears many registers in the device. The dedicated global routing structure is built into the device, eliminating the need to create a clock tree.

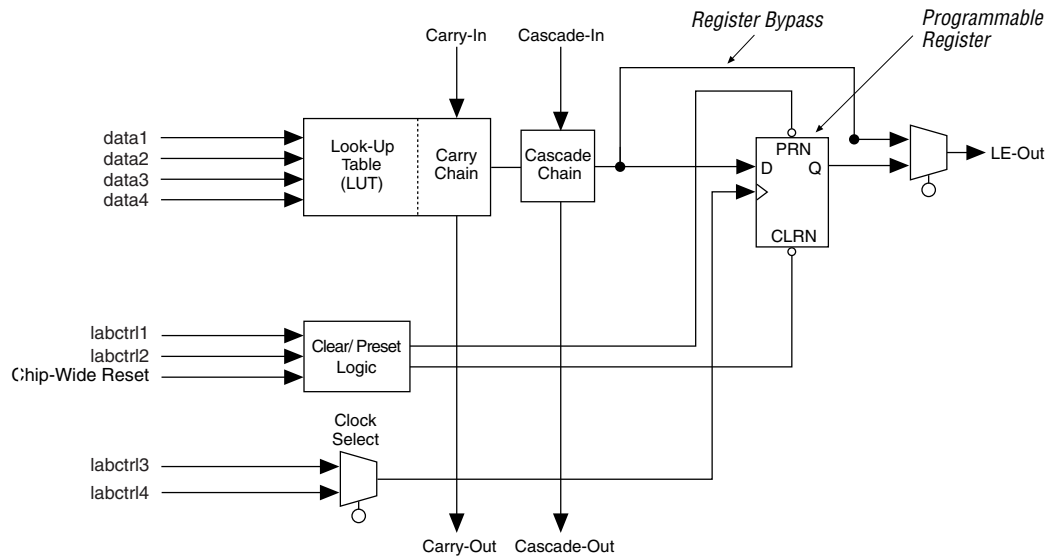
### Logic Array Block

An LAB consists of ten LEs, their associated carry and cascade chains, the LAB control signals, and the LAB local interconnect. The LAB provides the coarse-grained structure of the FLEX 6000 architecture, and facilitates efficient routing with optimum device utilization and high performance.

**Figure 3. LAB Control Signals**

## Logic Element

An LE, the smallest unit of logic in the FLEX 6000 architecture, has a compact size that provides efficient logic usage. Each LE contains a four-input LUT, which is a function generator that can quickly implement any function of four variables. An LE contains a programmable flipflop, carry and cascade chains. Additionally, each LE drives both the local and the FastTrack Interconnect. See [Figure 4](#).

**Figure 4. Logic Element**

The programmable flipflop in the LE can be configured for D, T, JK, or SR operation. The clock and clear control signals on the flipflop can be driven by global signals, general-purpose I/O pins, or any internal logic. For combinatorial functions, the flipflop is bypassed and the output of the LUT drives the outputs of the LE. The LE output can drive both the local interconnect and the FastTrack Interconnect.

The FLEX 6000 architecture provides two types of dedicated high-speed data paths that connect adjacent LEs without using local interconnect paths: carry chains and cascade chains. A carry chain supports high-speed arithmetic functions such as counters and adders, while a cascade chain implements wide-input functions such as equivalent comparators with minimum delay. Carry and cascade chains connect LEs 2 through 10 in an LAB and all LABs in the same half of the row. Because extensive use of carry and cascade chains can reduce routing flexibility, these chains should be limited to speed-critical portions of a design.

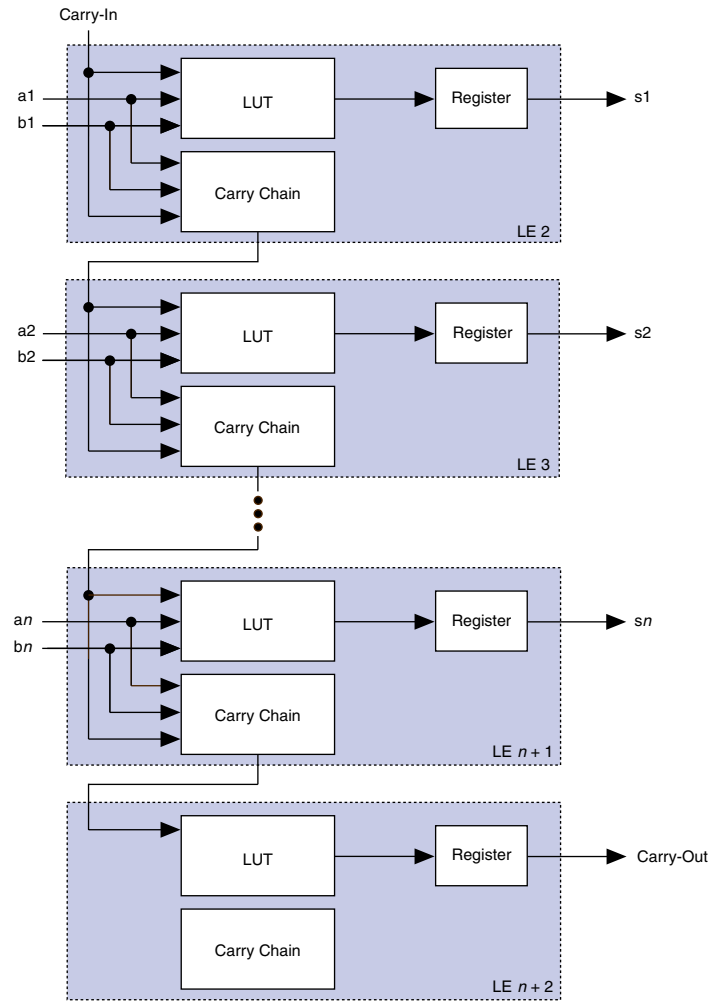
### *Carry Chain*

The carry chain provides a very fast (0.1 ns) carry-forward function between LEs. The carry-in signal from a lower-order bit drives forward into the higher-order bit via the carry chain, and feeds into both the LUT and the next portion of the carry chain. This feature allows the FLEX 6000 architecture to implement high-speed counters, adders, and comparators of arbitrary width. Carry chain logic can be created automatically by the Altera software during design processing, or manually by the designer during design entry. Parameterized functions such as LPM and DesignWare functions automatically take advantage of carry chains for the appropriate functions.

Because the first LE of each LAB can generate control signals for that LAB, the first LE in each LAB is not included in carry chains. In addition, the inputs of the first LE in each LAB may be used to generate synchronous clear and load enable signals for counters implemented with carry chains.

Carry chains longer than nine LEs are implemented automatically by linking LABs together. For enhanced fitting, a long carry chain skips alternate LABs in a row. A carry chain longer than one LAB skips either from an even-numbered LAB to another even-numbered LAB, or from an odd-numbered LAB to another odd-numbered LAB. For example, the last LE of the first LAB in a row carries to the second LE of the third LAB in the row. In addition, the carry chain does not cross the middle of the row. For instance, in the EPF6016 device, the carry chain stops at the 11th LAB in a row and a new carry chain begins at the 12th LAB.

Figure 5 shows how an  $n$ -bit full adder can be implemented in  $n + 1$  LEs with the carry chain. One portion of the LUT generates the sum of two bits using the input signals and the carry-in signal; the sum is routed to the output of the LE. Although the register can be bypassed for simple adders, it can be used for an accumulator function. Another portion of the LUT and the carry chain logic generates the carry-out signal, which is routed directly to the carry-in signal of the next-higher-order bit. The final carry-out signal is routed to an LE, where it is driven onto the FastTrack Interconnect.

**Figure 5. Carry Chain Operation**

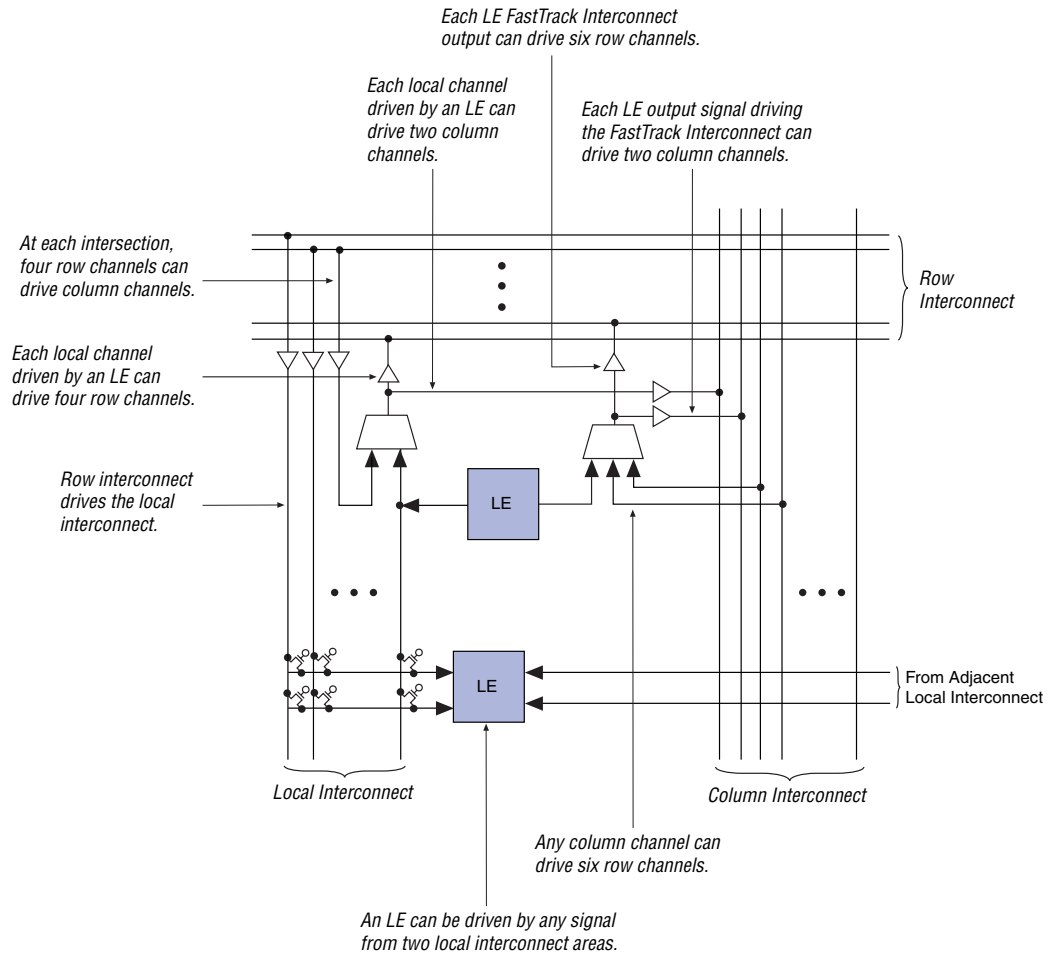
### *Cascade Chain*

The cascade chain enables the FLEX 6000 architecture to implement very wide fan-in functions. Adjacent LUTs can be used to implement portions of the function in parallel; the cascade chain serially connects the intermediate values. The cascade chain can use a logical AND or logical OR gate (via De Morgan's inversion) to connect the outputs of adjacent LEs. Each additional LE provides four more inputs to the effective width of a function, with a delay as low as 0.5 ns per LE. Cascade chain logic can be created automatically by the Altera software during design processing, or manually by the designer during design entry. Parameterized functions such as LPM and DesignWare functions automatically take advantage of cascade chains for the appropriate functions.

A cascade chain implementing an AND gate can use the register in the last LE; a cascade chain implementing an OR gate cannot use this register because of the inversion required to implement the OR gate.

Because the first LE of an LAB can generate control signals for that LAB, the first LE in each LAB is not included in cascade chains. Moreover, cascade chains longer than nine bits are automatically implemented by linking several LABs together. For easier routing, a long cascade chain skips every other LAB in a row. A cascade chain longer than one LAB skips either from an even-numbered LAB to another even-numbered LAB, or from an odd-numbered LAB to another odd-numbered LAB. For example, the last LE of the first LAB in a row cascades to the second LE of the third LAB. The cascade chain does not cross the center of the row. For example, in an EPF6016 device, the cascade chain stops at the 11th LAB in a row and a new cascade chain begins at the 12th LAB.

Figure 6 shows how the cascade function can connect adjacent LEs to form functions with a wide fan-in. In this example, functions of  $4n$  variables are implemented with  $n$  LEs. The cascade chain requires 3.4 ns to decode a 16-bit address.

**Figure 10. LAB Connections to Row & Column Interconnects**

For improved routability, the row interconnect consists of full-length and half-length channels. The full-length channels connect to all LABs in a row; the half-length channels connect to the LABs in half of the row. In addition to providing a predictable, row-wide interconnect, this architecture provides increased routing resources. Two neighboring LABs can be connected using a half-length channel, which saves the other half of the channel for the other half of the row. One-third of the row channels are half-length channels.

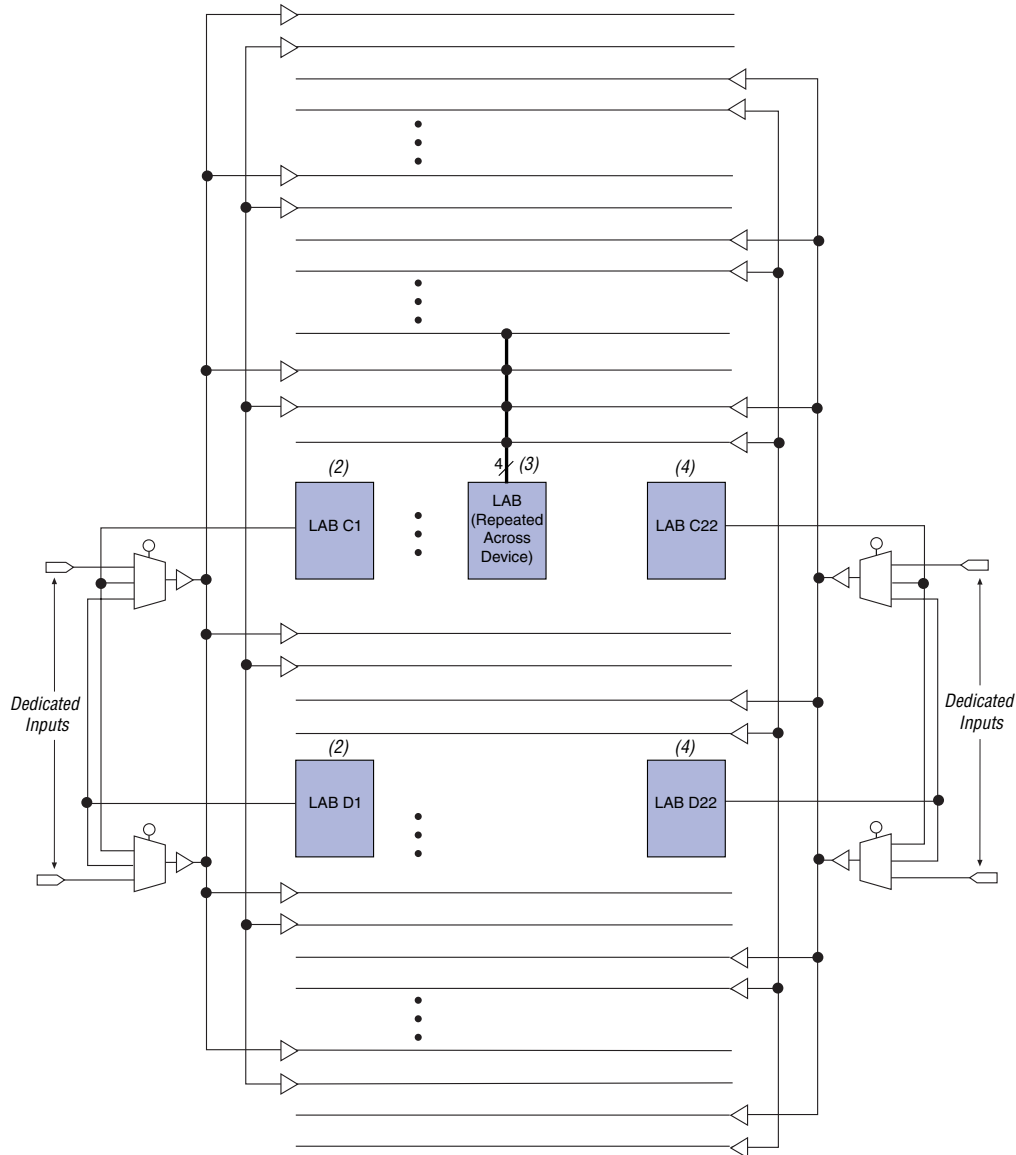


Table 5 summarizes the FastTrack Interconnect resources available in each FLEX 6000 device.

<b>Table 5. FLEX 6000 FastTrack Interconnect Resources</b>				
<b>Device</b>	<b>Rows</b>	<b>Channels per Row</b>	<b>Columns</b>	<b>Channels per Column</b>
EPF6010A	4	144	22	20
EPF6016 EPF6016A	6	144	22	20
EPF6024A	7	186	28	30

In addition to general-purpose I/O pins, FLEX 6000 devices have four dedicated input pins that provide low-skew signal distribution across the device. These four inputs can be used for global clock and asynchronous clear control signals. These signals are available as control signals for all LEs in the device. The dedicated inputs can also be used as general-purpose data inputs because they can feed the local interconnect of each LAB in the device. Using dedicated inputs to route data signals provides a fast path for high fan-out signals.

The local interconnect from LABs located at either end of two rows can drive a global control signal. For instance, in an EPF6016 device, LABs C1, D1, C22, and D22 can all drive global control signals. When an LE drives a global control signal, the dedicated input pin that drives that signal cannot be used. Any LE in the device can drive a global control signal by driving the FastTrack Interconnect into the appropriate LAB. To minimize delay, however, the Altera software places the driving LE in the appropriate LAB. The LE-driving-global signal feature is optimized for speed for control signals; regular data signals are better routed on the FastTrack Interconnect and do not receive any advantage from being routed on global signals. This LE-driving-global control signal feature is controlled by the designer and is not used automatically by the Altera software. See Figure 11.

**Figure 11. Global Clock & Clear Distribution** *Note (1)***Notes:**

- (1) The global clock and clear distribution signals are shown for EPF6016 and EPF6016A devices. In EPF6010A devices, LABs in rows B and C drive global signals. In EPF6024A devices, LABs in rows C and E drive global signals.
- (2) The local interconnect from LABs C1 and D1 can drive two global control signals on the left side.
- (3) Global signals drive into every LAB as clock, asynchronous clear, preset, and data signals.
- (4) The local interconnect from LABs C22 and D22 can drive two global control signals on the right side.

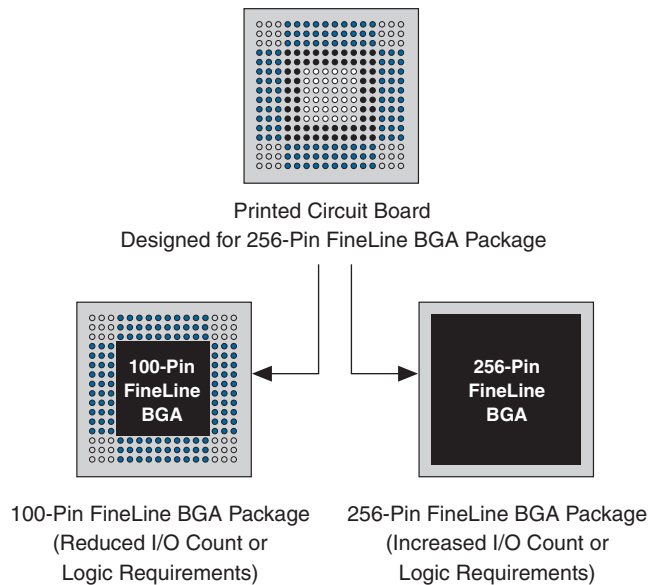
**Figure 15. SameFrame Pin-Out Example**

Table 6 lists the 3.3-V FLEX 6000 devices with the SameFrame pin-out feature.

**Table 6. 3.3-V FLEX 6000 Devices with SameFrame Pin-Outs**

Device	100-Pin FineLine BGA	256-Pin FineLine BGA
EPF6016A	V	V
EPF6024A		V

## Output Configuration

This section discusses slew-rate control, the MultiVolt I/O interface, power sequencing, and hot-socketing for FLEX 6000 devices.

### Slew-Rate Control

The output buffer in each IOE has an adjustable output slew-rate that can be configured for low-noise or high-speed performance. A slower slew-rate reduces system noise and adds a maximum delay of 6.8 ns. The fast slew-rate should be used for speed-critical outputs in systems that are adequately protected against noise. Designers can specify the slew-rate on a pin-by-pin basis during design entry or assign a default slew rate to all pins on a device-wide basis. The slew-rate setting affects only the falling edge of the output.

The instruction register length for FLEX 6000 devices is three bits. [Table 9](#) shows the boundary-scan register length for FLEX 6000 devices.

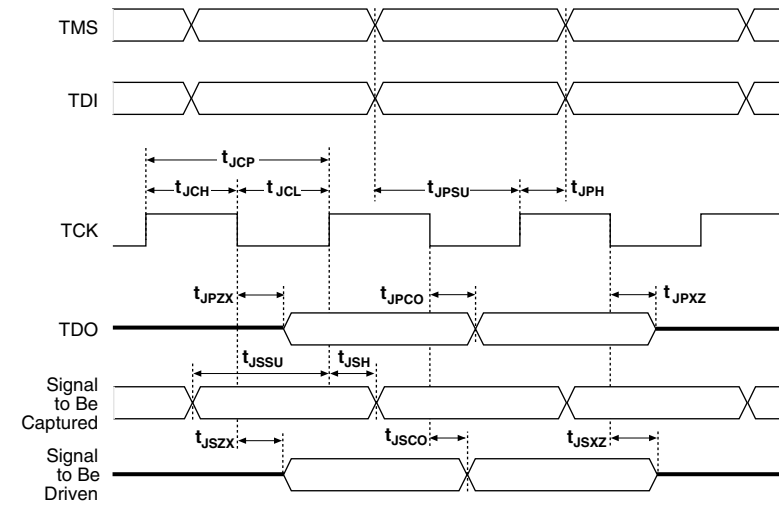
<b>Table 9. FLEX 6000 Device Boundary-Scan Register Length</b>	
<b>Device</b>	<b>Boundary-Scan Register Length</b>
EPF6010A	522
EPF6016	621
EPF6016A	522
EPF6024A	666

FLEX 6000 devices include a weak pull-up on JTAG pins.

f See [Application Note 39 \(IEEE 1149.1 \(JTAG\) Boundary-Scan Testing in Altera Devices\)](#) for more information.

[Figure 16](#) shows the timing requirements for the JTAG signals.

**Figure 16. JTAG Waveforms**



[Table 10](#) shows the JTAG timing parameters and values for FLEX 6000 devices.

**Table 10. JTAG Timing Parameters & Values**

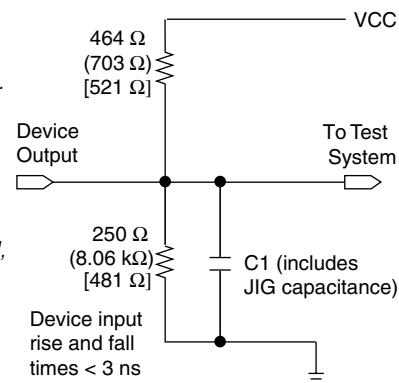
Symbol	Parameter	Min	Max	Unit
$t_{JCP}$	TCK clock period	100		ns
$t_{JCH}$	TCK clock high time	50		ns
$t_{JCL}$	TCK clock low time	50		ns
$t_{JPSU}$	JTAG port setup time	20		ns
$t_{JPH}$	JTAG port hold time	45		ns
$t_{JPCO}$	JTAG port clock-to-output		25	ns
$t_{JPZX}$	JTAG port high impedance to valid output		25	ns
$t_{JPXZ}$	JTAG port valid output to high impedance		25	ns
$t_{JSSU}$	Capture register setup time	20		ns
$t_{JSH}$	Capture register hold time	45		ns
$t_{JSCO}$	Update register clock-to-output		35	ns
$t_{JSZX}$	Update register high impedance to valid output		35	ns
$t_{JSXZ}$	Update register valid output to high impedance		35	ns

## Generic Testing

Each FLEX 6000 device is functionally tested. Complete testing of each configurable SRAM bit and all logic functionality ensures 100% configuration yield. AC test measurements for FLEX 6000 devices are made under conditions equivalent to those shown in [Figure 17](#). Multiple test patterns can be used to configure devices during all stages of the production flow.

**Figure 17. AC Test Conditions**

Power supply transients can affect AC measurements. Simultaneous transitions of multiple outputs should be avoided for accurate measurement. Threshold tests must not be performed under AC conditions. Large-amplitude, fast-ground-current transients normally occur as the device outputs discharge the load capacitances. When these transients flow through the parasitic inductance between the device ground pin and the test system ground, significant reductions in observable noise immunity can result. Numbers without parentheses are for 5.0-V devices or outputs. Numbers in parentheses are for 3.3-V devices or outputs. Numbers in brackets are for 2.5-V devices or outputs.



## Operating Conditions

Tables 11 through 18 provide information on absolute maximum ratings, recommended operating conditions, operating conditions, and capacitance for 5.0-V and 3.3-V FLEX 6000 devices.

**Table 11. FLEX 6000 5.0-V Device Absolute Maximum Ratings** *Note (1)*

Symbol	Parameter	Conditions	Min	Max	Unit
$V_{CC}$	Supply voltage	With respect to ground (2)	−2.0	7.0	V
$V_I$	DC input voltage		−2.0	7.0	V
$I_{OUT}$	DC output current, per pin		−25	25	mA
$T_{STG}$	Storage temperature	No bias	−65	150	°C
$T_{AMB}$	Ambient temperature	Under bias	−65	135	°C
$T_J$	Junction temperature	PQFP, TQFP, and BGA packages		135	°C

**Table 12. FLEX 6000 5.0-V Device Recommended Operating Conditions**

Symbol	Parameter	Conditions	Min	Max	Unit
$V_{CCINT}$	Supply voltage for internal logic and input buffers	(3), (4)	4.75 (4.50)	5.25 (5.50)	V
$V_{CCIO}$	Supply voltage for output buffers, 5.0-V operation	(3), (4)	4.75 (4.50)	5.25 (5.50)	V
	Supply voltage for output buffers, 3.3-V operation	(3), (4)	3.00 (3.00)	3.60 (3.60)	V
$V_I$	Input voltage		−0.5	$V_{CCINT} + 0.5$	V
$V_O$	Output voltage		0	$V_{CCIO}$	V
$T_J$	Operating temperature	For commercial use	0	85	°C
		For industrial use	−40	100	°C
$t_R$	Input rise time			40	ns
$t_F$	Input fall time			40	ns

**Table 15. FLEX 6000 3.3-V Device Absolute Maximum Ratings** *Note (1)*

Symbol	Parameter	Conditions	Min	Max	Unit
$V_{CC}$	Supply voltage	With respect to ground (2)	−0.5	4.6	V
$V_I$	DC input voltage		−2.0	5.75	V
$I_{OUT}$	DC output current, per pin		−25	25	mA
$T_{STG}$	Storage temperature	No bias	−65	150	°C
$T_{AMB}$	Ambient temperature	Under bias	−65	135	°C
$T_J$	Junction temperature	PQFP, PLCC, and BGA packages		135	°C

**Table 16. FLEX 6000 3.3-V Device Recommended Operating Conditions**

Symbol	Parameter	Conditions	Min	Max	Unit
$V_{CCINT}$	Supply voltage for internal logic and input buffers	(3), (4)	3.00 (3.00)	3.60 (3.60)	V
$V_{CCIO}$	Supply voltage for output buffers, 3.3-V operation	(3), (4)	3.00 (3.00)	3.60 (3.60)	V
	Supply voltage for output buffers, 2.5-V operation	(3), (4)	2.30 (2.30)	2.70 (2.70)	V
$V_I$	Input voltage		−0.5	5.75	V
$V_O$	Output voltage		0	$V_{CCIO}$	V
$T_J$	Operating temperature	For commercial use	0	85	°C
		For industrial use	−40	100	°C
$t_R$	Input rise time			40	ns
$t_F$	Input fall time			40	ns

<b>Table 17. FLEX 6000 3.3-V Device DC Operating Conditions</b> <i>Notes (5), (6)</i>						
Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{IH}$	High-level input voltage		1.7		5.75	V
$V_{IL}$	Low-level input voltage		-0.5		0.8	V
$V_{OH}$	3.3-V high-level TTL output voltage	$I_{OH} = -8$ mA DC, $V_{CCIO} = 3.00$ V (7)	2.4			V
	3.3-V high-level CMOS output voltage	$I_{OH} = -0.1$ mA DC, $V_{CCIO} = 3.00$ V (7)	$V_{CCIO} - 0.2$			V
	2.5-V high-level output voltage	$I_{OH} = -100$ $\mu$ A DC, $V_{CCIO} = 2.30$ V (7)	2.1			V
		$I_{OH} = -1$ mA DC, $V_{CCIO} = 2.30$ V (7)	2.0			V
		$I_{OH} = -2$ mA DC, $V_{CCIO} = 2.30$ V (7)	1.7			V
$V_{OL}$	3.3-V low-level TTL output voltage	$I_{OL} = 8$ mA DC, $V_{CCIO} = 3.00$ V (8)			0.45	V
	3.3-V low-level CMOS output voltage	$I_{OL} = 0.1$ mA DC, $V_{CCIO} = 3.00$ V (8)			0.2	V
	2.5-V low-level output voltage	$I_{OL} = 100$ $\mu$ A DC, $V_{CCIO} = 2.30$ V (8)			0.2	V
		$I_{OL} = 1$ mA DC, $V_{CCIO} = 2.30$ V (8)			0.4	V
		$I_{OL} = 2$ mA DC, $V_{CCIO} = 2.30$ V (8)			0.7	V
$I_I$	Input pin leakage current	$V_I = 5.3$ V to ground (8)	-10		10	$\mu$ A
$I_{OZ}$	Tri-stated I/O pin leakage current	$V_O = 5.3$ V to ground (8)	-10		10	$\mu$ A
$I_{CC0}$	$V_{CC}$ supply current (standby)	$V_I =$ ground, no load		0.5	5	mA

<b>Table 18. FLEX 6000 3.3-V Device Capacitance</b> <i>Note (9)</i>					
Symbol	Parameter	Conditions	Min	Max	Unit
$C_{IN}$	Input capacitance for I/O pin	$V_{IN} = 0$ V, $f = 1.0$ MHz		8	pF
$C_{INCLK}$	Input capacitance for dedicated input	$V_{IN} = 0$ V, $f = 1.0$ MHz		12	pF
$C_{OUT}$	Output capacitance	$V_{OUT} = 0$ V, $f = 1.0$ MHz		8	pF

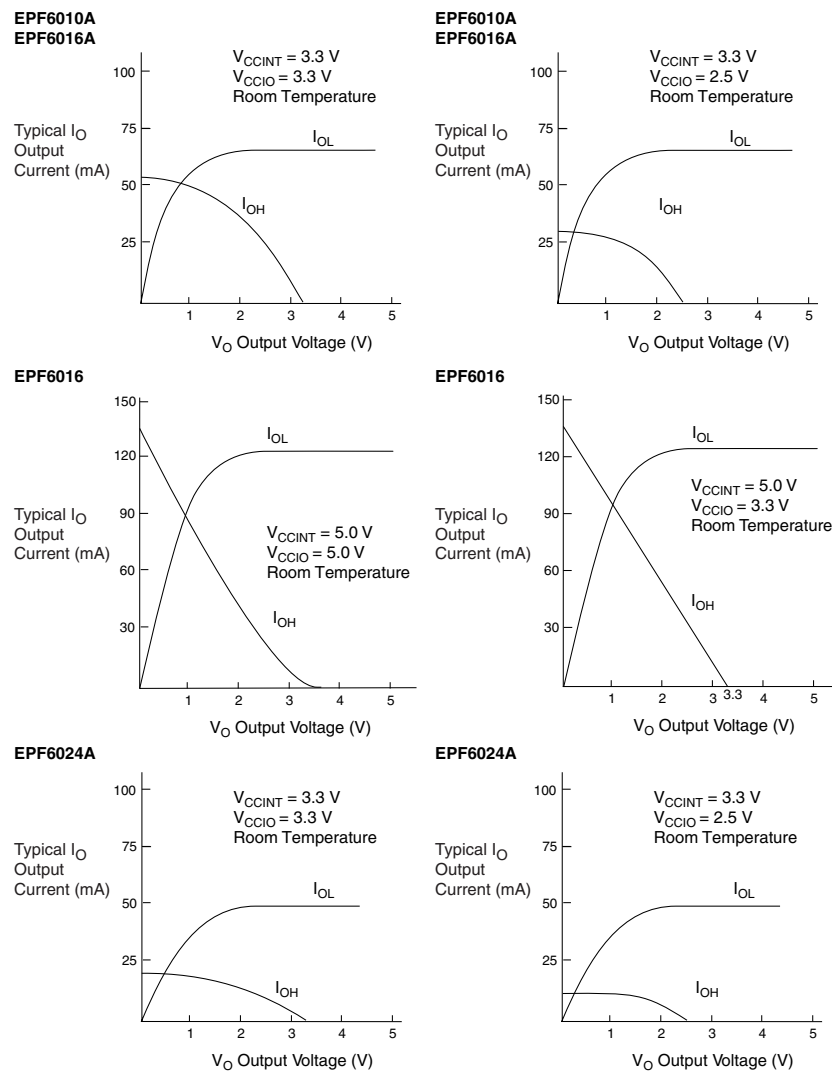
**Notes to tables:**

- (1) See the *Operating Requirements for Altera Devices Data Sheet*.
- (2) The minimum DC input voltage is -0.5 V. During transitions, the inputs may undershoot to -2.0 V or overshoot to 5.75 V for input currents less than 100 mA and periods shorter than 20 ns.
- (3) Numbers in parentheses are for industrial-temperature-range devices.
- (4) Maximum  $V_{CC}$  rise time is 100 ms.  $V_{CC}$  must rise monotonically.
- (5) Typical values are for  $T_A = 25^\circ$  C and  $V_{CC} = 3.3$  V.
- (6) These values are specified under [Table 16 on page 33](#).
- (7) The  $I_{OH}$  parameter refers to high-level TTL or CMOS output current.
- (8) The  $I_{OL}$  parameter refers to low-level TTL, PCI, or CMOS output current. This parameter applies to open-drain pins as well as output pins.
- (9) Capacitance is sample-tested only.



Figure 18 shows the typical output drive characteristics of 5.0-V and 3.3-V FLEX 6000 devices with 5.0-V, 3.3-V, and 2.5-V  $V_{CCIO}$ . When  $V_{CCIO} = 5.0$  V on EPF6016 devices, the output driver is compliant with the *PCI Local Bus Specification, Revision 2.2* for 5.0-V operation. When  $V_{CCIO} = 3.3$  V on the EPF6010A and EPF6016A devices, the output driver is compliant with the *PCI Local Bus Specification, Revision 2.2* for 3.3-V operation.

Figure 18. Output Drive Characteristics



Tables 19 through 21 describe the FLEX 6000 internal timing microparameters, which are expressed as worst-case values. Using hand calculations, these parameters can be used to estimate design performance. However, before committing designs to silicon, actual worst-case performance should be modeled using timing simulation and timing analysis. Tables 22 and 23 describe FLEX 6000 external timing parameters.

<b>Table 19. LE Timing Microparameters</b> <i>Note (1)</i>		
<b>Symbol</b>	<b>Parameter</b>	<b>Conditions</b>
$t_{REG\_TO\_REG}$	LUT delay for LE register feedback in carry chain	
$t_{CASC\_TO\_REG}$	Cascade-in to register delay	
$t_{CARRY\_TO\_REG}$	Carry-in to register delay	
$t_{DATA\_TO\_REG}$	LE input to register delay	
$t_{CASC\_TO\_OUT}$	Cascade-in to LE output delay	
$t_{CARRY\_TO\_OUT}$	Carry-in to LE output delay	
$t_{DATA\_TO\_OUT}$	LE input to LE output delay	
$t_{REG\_TO\_OUT}$	Register output to LE output delay	
$t_{SU}$	LE register setup time before clock; LE register recovery time after asynchronous clear	
$t_H$	LE register hold time after clock	
$t_{CO}$	LE register clock-to-output delay	
$t_{CLR}$	LE register clear delay	
$t_C$	LE register control signal delay	
$t_{LD\_CLR}$	Synchronous load or clear delay in counter mode	
$t_{CARRY\_TO\_CARRY}$	Carry-in to carry-out delay	
$t_{REG\_TO\_CARRY}$	Register output to carry-out delay	
$t_{DATA\_TO\_CARRY}$	LE input to carry-out delay	
$t_{CARRY\_TO\_CASC}$	Carry-in to cascade-out delay	
$t_{CASC\_TO\_CASC}$	Cascade-in to cascade-out delay	
$t_{REG\_TO\_CASC}$	Register-out to cascade-out delay	
$t_{DATA\_TO\_CASC}$	LE input to cascade-out delay	
$t_{CH}$	LE register clock high time	
$t_{CL}$	LE register clock low time	

<b>Table 23. External Timing Parameters</b>		
<b>Symbol</b>	<b>Parameter</b>	<b>Conditions</b>
$t_{\text{INSU}}$	Setup time with global clock at LE register	(8)
$t_{\text{INH}}$	Hold time with global clock at LE register	(8)
$t_{\text{OUTCO}}$	Clock-to-output delay with global clock with LE register using FastFLEX I/O pin	(8)

**Notes to tables:**

- (1) Microparameters are timing delays contributed by individual architectural elements and cannot be measured explicitly.
- (2) Operating conditions:  
 $V_{\text{CCIO}} = 5.0 \text{ V} \pm 5\%$  for commercial use in 5.0-V FLEX 6000 devices.  
 $V_{\text{CCIO}} = 5.0 \text{ V} \pm 10\%$  for industrial use in 5.0-V FLEX 6000 devices.  
 $V_{\text{CCIO}} = 3.3 \text{ V} \pm 10\%$  for commercial or industrial use in 3.3-V FLEX 6000 devices.
- (3) Operating conditions:  
 $V_{\text{CCIO}} = 3.3 \text{ V} \pm 10\%$  for commercial or industrial use in 5.0-V FLEX 6000 devices.  
 $V_{\text{CCIO}} = 2.5 \text{ V} \pm 0.2 \text{ V}$  for commercial or industrial use in 3.3-V FLEX 6000 devices.
- (4) Operating conditions:  
 $V_{\text{CCIO}} = 2.5 \text{ V}, 3.3 \text{ V}, \text{ or } 5.0 \text{ V}.$
- (5) These parameters are worst-case values for typical applications. Post-compilation timing simulation and timing analysis are required to determine actual worst-case performance.
- (6) This timing parameter shows the delay of a register-to-register test pattern and is used to determine speed grades. There are 12 LEs, including source and destination registers. The row and column interconnects between the registers vary in length.
- (7) This timing parameter is shown for reference and is specified by characterization.
- (8) This timing parameter is specified by characterization.

Tables 24 through 28 show the timing information for EPF6010A and EPF6016A devices.

Table 24. LE Timing Microparameters for EPF6010A & EPF6016A Devices (Part 1 of 2)							
Parameter	Speed Grade						Unit
	-1		-2		-3		
	Min	Max	Min	Max	Min	Max	
$t_{REG\_TO\_REG}$		1.2		1.3		1.7	ns
$t_{CASC\_TO\_REG}$		0.9		1.0		1.2	ns
$t_{CARRY\_TO\_REG}$		0.9		1.0		1.2	ns
$t_{DATA\_TO\_REG}$		1.1		1.2		1.5	ns
$t_{CASC\_TO\_OUT}$		1.3		1.4		1.8	ns
$t_{CARRY\_TO\_OUT}$		1.6		1.8		2.3	ns
$t_{DATA\_TO\_OUT}$		1.7		2.0		2.5	ns
$t_{REG\_TO\_OUT}$		0.4		0.4		0.5	ns
$t_{SU}$	0.9		1.0		1.3		ns
$t_H$	1.4		1.7		2.1		ns

Tables 29 through 33 show the timing information for EPF6016 devices.

Table 29. LE Timing Microparameters for EPF6016 Devices					
Parameter	Speed Grade				Unit
	-2		-3		
	Min	Max	Min	Max	
$t_{REG\_TO\_REG}$		2.2		2.8	ns
$t_{CASC\_TO\_REG}$		0.9		1.2	ns
$t_{CARRY\_TO\_REG}$		1.6		2.1	ns
$t_{DATA\_TO\_REG}$		2.4		3.0	ns
$t_{CASC\_TO\_OUT}$		1.3		1.7	ns
$t_{CARRY\_TO\_OUT}$		2.4		3.0	ns
$t_{DATA\_TO\_OUT}$		2.7		3.4	ns
$t_{REG\_TO\_OUT}$		0.3		0.5	ns
$t_{SU}$	1.1		1.6		ns
$t_H$	1.8		2.3		ns
$t_{CO}$		0.3		0.4	ns
$t_{CLR}$		0.5		0.6	ns
$t_C$		1.2		1.5	ns
$t_{LD\_CLR}$		1.2		1.5	ns
$t_{CARRY\_TO\_CARRY}$		0.2		0.4	ns
$t_{REG\_TO\_CARRY}$		0.8		1.1	ns
$t_{DATA\_TO\_CARRY}$		1.7		2.2	ns
$t_{CARRY\_TO\_CASC}$		1.7		2.2	ns
$t_{CASC\_TO\_CASC}$		0.9		1.2	ns
$t_{REG\_TO\_CASC}$		1.6		2.0	ns
$t_{DATA\_TO\_CASC}$		1.7		2.1	ns
$t_{CH}$	4.0		4.0		ns
$t_{CL}$	4.0		4.0		ns

Table 30. IOE Timing Microparameters for EPF6016 Devices					
Parameter	Speed Grade				Unit
	-2		-3		
	Min	Max	Min	Max	
$t_{OD1}$		2.3		2.8	ns
$t_{OD2}$		4.6		5.1	ns

**Table 30. IOE Timing Microparameters for EPF6016 Devices**

Table 30. IOE Timing Microparameters for EPF6016 Devices					
Parameter	Speed Grade				Unit
	-2		-3		
	Min	Max	Min	Max	
$t_{OD3}$		4.7		5.2	ns
$t_{XZ}$		2.3		2.8	ns
$t_{ZX1}$		2.3		2.8	ns
$t_{ZX2}$		4.6		5.1	ns
$t_{ZX3}$		4.7		5.2	ns
$t_{IOE}$		0.5		0.6	ns
$t_{IN}$		3.3		4.0	ns
$t_{IN\_DELAY}$		4.6		5.6	ns

**Table 31. Interconnect Timing Microparameters for EPF6016 Devices**

Table 31. Interconnect Timing Microparameters for EPF6016 Devices					
Parameter	Speed Grade				Unit
	-2		-3		
	Min	Max	Min	Max	
$t_{LOCAL}$		0.8		1.0	ns
$t_{ROW}$		2.9		3.3	ns
$t_{COL}$		2.3		2.5	ns
$t_{DIN\_D}$		4.9		6.0	ns
$t_{DIN\_C}$		4.8		6.0	ns
$t_{LEGLOBAL}$		3.1		3.9	ns
$t_{LABCARRY}$		0.4		0.5	ns
$t_{LABCASC}$		0.8		1.0	ns

**Table 32. External Reference Timing Parameters for EPF6016 Devices**

Table 32. External Reference Timing Parameters for EPF6016 Devices					
Parameter	Speed Grade				Unit
	-2		-3		
	Min	Max	Min	Max	
t <sub>1</sub>		53.0		65.0	ns
t <sub>DDR</sub>		16.0		20.0	ns