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Understanding <u>Embedded - FPGAs (Field Programmable Gate Array)</u>

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	196
Number of Logic Elements/Cells	1960
Total RAM Bits	-
Number of I/O	219
Number of Gates	24000
Voltage - Supply	3V ~ 3.6V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	256-BGA
Supplier Device Package	256-FBGA (17x17)
Purchase URL	https://www.e-xfl.com/product-detail/intel/epf6024afc256-3

Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

Functional Description

The FLEX 6000 OptiFLEX architecture consists of logic elements (LEs). Each LE includes a 4-input look-up table (LUT), which can implement any 4-input function, a register, and dedicated paths for carry and cascade chain functions. Because each LE contains a register, a design can be easily pipelined without consuming more LEs. The specified gate count for FLEX 6000 devices includes all LUTs and registers.

LEs are combined into groups called logic array blocks (LABs); each LAB contains 10 LEs. The Altera software automatically places related LEs into the same LAB, minimizing the number of required interconnects. Each LAB can implement a medium-sized block of logic, such as a counter or multiplexer.

Signal interconnections within FLEX 6000 devices—and to and from device pins—are provided via the routing structure of the FastTrack Interconnect. The routing structure is a series of fast, continuous row and column channels that run the entire length and width of the device. Any LE or pin can feed or be fed by any other LE or pin via the FastTrack Interconnect. See "FastTrack Interconnect" on page 17 of this data sheet for more information.

Each I/O pin is fed by an I/O element (IOE) located at the end of each row and column of the FastTrack Interconnect. Each IOE contains a bidirectional I/O buffer. Each IOE is placed next to an LAB, where it can be driven by the local interconnect of that LAB. This feature allows fast clock-to-output times of less than 8 ns when a pin is driven by any of the 10 LEs in the adjacent LAB. Also, any LE can drive any pin via the row and column interconnect. I/O pins can drive the LE registers via the row and column interconnect, providing setup times as low as 2 ns and hold times of 0 ns. IOEs provide a variety of features, such as JTAG BST support, slew-rate control, and tri-state buffers.

Figure 1 shows a block diagram of the FLEX 6000 OptiFLEX architecture. Each group of ten LEs is combined into an LAB, and the LABs are arranged into rows and columns. The LABs are interconnected by the FastTrack Interconnect. IOEs are located at the end of each FastTrack Interconnect row and column.

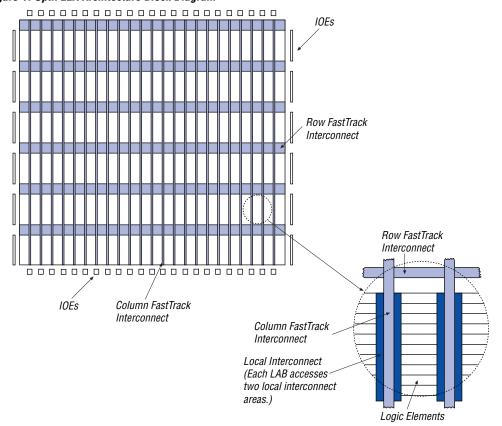


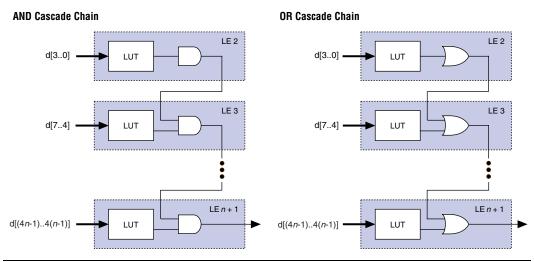
Figure 1. OptiFLEX Architecture Block Diagram

FLEX 6000 devices provide four dedicated, global inputs that drive the control inputs of the flipflops to ensure efficient distribution of high-speed, low-skew control signals. These inputs use dedicated routing channels that provide shorter delays and lower skews than the FastTrack Interconnect. These inputs can also be driven by internal logic, providing an ideal solution for a clock divider or an internally generated asynchronous clear signal that clears many registers in the device. The dedicated global routing structure is built into the device, eliminating the need to create a clock tree.

Logic Array Block

An LAB consists of ten LEs, their associated carry and cascade chains, the LAB control signals, and the LAB local interconnect. The LAB provides the coarse-grained structure of the FLEX 6000 architecture, and facilitates efficient routing with optimum device utilization and high performance.

Figure 6. Cascade Chain Operation



LE Operating Modes

The FLEX 6000 LE can operate in one of the following three modes:

- Normal mode
- Arithmetic mode
- Counter mode

Each of these modes uses LE resources differently. In each mode, seven available inputs to the LE—the four data inputs from the LAB local interconnect, the feedback from the programmable register, and the carry-in and cascade-in from the previous LE—are directed to different destinations to implement the desired logic function. LAB-wide signals provide clock, asynchronous clear, synchronous clear, and synchronous load control for the register. The Altera software, in conjunction with parameterized functions such as LPM and DesignWare functions, automatically chooses the appropriate mode for common functions such as counters, adders, and multipliers. If required, the designer can also create special-purpose functions to use an LE operating mode for optimal performance.

Figure 7 shows the LE operating modes.

Either the counter enable or the up/down control may be used for a given counter. Moreover, the synchronous load can be used as a count enable by routing the register output into the data input automatically when requested by the designer.

The second LE of each LAB has a special function for counter mode; the carry-in of the LE can be driven by a fast feedback path from the register. This function gives a faster counter speed for counter carry chains starting in the second LE of an LAB.

The Altera software implements functions to use the counter mode automatically where appropriate. The designer does not have to decide how the carry chain will be used.

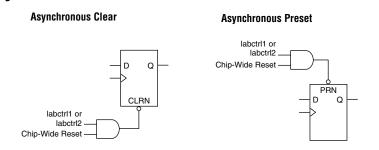
Internal Tri-State Emulation

Internal tri-state emulation provides internal tri-states without the limitations of a physical tri-state bus. In a physical tri-state bus, the tri-state buffers' output enable (OE) signals select which signal drives the bus. However, if multiple OE signals are active, contending signals can be driven onto the bus. Conversely, if no OE signals are active, the bus will float. Internal tri-state emulation resolves contending tri-state buffers to a low value and floating buses to a high value, thereby eliminating these problems. The Altera software automatically implements tri-state bus functionality with a multiplexer.

Clear & Preset Logic Control

Logic for the programmable register's clear and preset functions is controlled by the LAB-wide signals LABCTRL1 and LABCTRL2. The LE register has an asynchronous clear that can implement an asynchronous preset. Either LABCTRL1 or LABCTRL2 can control the asynchronous clear or preset. Because the clear and preset functions are active-low, the Altera software automatically assigns a logic high to an unused clear or preset signal. The clear and preset logic is implemented in either the asynchronous clear or asynchronous preset mode, which is chosen during design entry (see Figure 8).

Figure 8. LE Clear & Preset Modes



Asynchronous Clear

The flipflop can be cleared by either LABCTRL1 or LABCTRL2.

Asynchronous Preset

An asynchronous preset is implemented with an asynchronous clear. The Altera software provides preset control by using the clear and inverting the input and output of the register. Inversion control is available for the inputs to both LEs and IOEs. Therefore, this technique can be used when a register drives logic or drives a pin.

In addition to the two clear and preset modes, FLEX 6000 devices provide a chip-wide reset pin (DEV_CLRn) that can reset all registers in the device. The option to use this pin is set in the Altera software before compilation. The chip-wide reset overrides all other control signals. Any register with an asynchronous preset will be preset when the chip-wide reset is asserted because of the inversion technique used to implement the asynchronous preset.

The Altera software can use a programmable NOT-gate push-back technique to emulate simultaneous preset and clear or asynchronous load. However, this technique uses an additional three LEs per register.

FastTrack Interconnect

In the FLEX 6000 OptiFLEX architecture, connections between LEs and device I/O pins are provided by the FastTrack Interconnect, a series of continuous horizontal and vertical routing channels that traverse the device. This global routing structure provides predictable performance, even for complex designs. In contrast, the segmented routing in FPGAs requires switch matrices to connect a variable number of routing paths, increasing the delays between logic resources and reducing performance.

Table 5 summarizes the FastTrack Interconnect resources available in each FLEX 6000 device.

Table 5. FLEX 600	Table 5. FLEX 6000 FastTrack Interconnect Resources					
Device	Rows	Channels per Row	Columns	Channels per Column		
EPF6010A	4	144	22	20		
EPF6016 EPF6016A	6	144	22	20		
EPF6024A	7	186	28	30		

In addition to general-purpose I/O pins, FLEX 6000 devices have four dedicated input pins that provide low-skew signal distribution across the device. These four inputs can be used for global clock and asynchronous clear control signals. These signals are available as control signals for all LEs in the device. The dedicated inputs can also be used as general-purpose data inputs because they can feed the local interconnect of each LAB in the device. Using dedicated inputs to route data signals provides a fast path for high fan-out signals.

The local interconnect from LABs located at either end of two rows can drive a global control signal. For instance, in an EPF6016 device, LABs C1, D1, C22, and D22 can all drive global control signals. When an LE drives a global control signal, the dedicated input pin that drives that signal cannot be used. Any LE in the device can drive a global control signal by driving the FastTrack Interconnect into the appropriate LAB. To minimize delay, however, the Altera software places the driving LE in the appropriate LAB. The LE-driving-global signal feature is optimized for speed for control signals; regular data signals are better routed on the FastTrack Interconnect and do not receive any advantage from being routed on global signals. This LE-driving-global control signal feature is controlled by the designer and is not used automatically by the Altera software. See Figure 11.

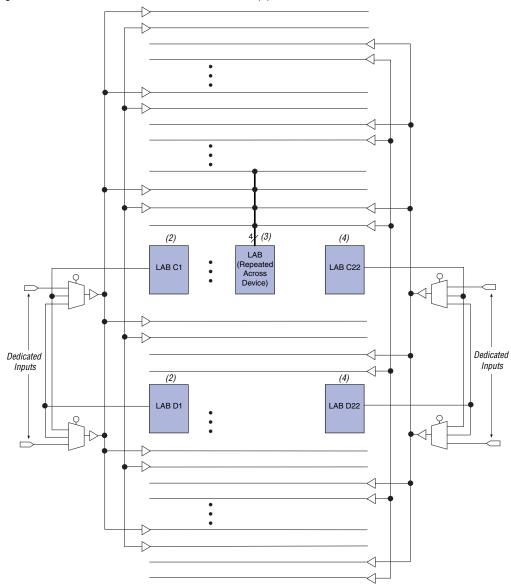


Figure 11. Global Clock & Clear Distribution Note (1)

Notes:

- The global clock and clear distribution signals are shown for EPF6016 and EPF6016A devices. In EPF6010A devices, LABs in rows B and C drive global signals. In EPF6024A devices, LABs in rows C and E drive global signals. The local interconnect from LABs C1 and D1 can drive two global control signals on the left side.
- (2)
- Global signals drive into every LAB as clock, asynchronous clear, preset, and data signals. (3)
- The local interconnect from LABs C22 and D22 can drive two global control signals on the right side.

I/O Elements

An IOE contains a bidirectional I/O buffer and a tri-state buffer. IOEs can be used as input, output, or bidirectional pins. An IOE receives its data signals from the adjacent local interconnect, which can be driven by a row or column interconnect (allowing any LE in the device to drive the IOE) or by an adjacent LE (allowing fast clock-to-output delays). A FastFLEXTM I/O pin is a row or column output pin that receives its data signals from the adjacent local interconnect driven by an adjacent LE. The IOE receives its output enable signal through the same path, allowing individual output enables for every pin and permitting emulation of open-drain buffers. The Altera Compiler uses programmable inversion to invert the data or output enable signals automatically where appropriate. Open-drain emulation is provided by driving the data input low and toggling the OE of each IOE. This emulation is possible because there is one OE per pin.

A chip-wide output enable feature allows the designer to disable all pins of the device by asserting one pin (DEV_OE). This feature is useful during board debugging or testing.

Figure 12 shows the IOE block diagram.

To Row or Column Interconnect

Chip-Wide Output Enable

From LAB Local Interconnect

Slew-Rate
Control

Figure 12. IOE Block Diagram

MultiVolt I/O Interface

The FLEX 6000 device architecture supports the MultiVolt I/O interface feature, which allows FLEX 6000 devices to interface with systems of differing supply voltages. The EPF6016 device can be set for 3.3-V or 5.0-V I/O pin operation. This device has one set of $V_{\rm CC}$ pins for internal operation and input buffers (VCCINT), and another set for output drivers (VCCIO).

The VCCINT pins on 5.0-V FLEX 6000 devices must always be connected to a 5.0-V power supply. With a 5.0-V V_{CCINT} level, input voltages are at TTL levels and are therefore compatible with 3.3-V and 5.0-V inputs.

The VCCIO pins on 5.0-V FLEX 6000 devices can be connected to either a 3.3-V or 5.0-V power supply, depending on the output requirements. When the VCCIO pins are connected to a 5.0-V power supply, the output levels are compatible with 5.0-V systems. When the VCCIO pins are connected to a 3.3-V power supply, the output high is 3.3 V and is therefore compatible with 3.3-V or 5.0-V systems. Devices operating with VCCIO levels lower than 4.75 V incur a nominally greater timing delay of t_{OD2} instead of t_{OD1} .

On 3.3-V FLEX 6000 devices, the VCCINT pins must be connected to a 3.3-V power supply. Additionally, 3.3-V FLEX 6000A devices can interface with 2.5-V, 3.3-V, or 5.0-V systems when the VCCIO pins are tied to 2.5 V. The output can drive 2.5-V systems, and the inputs can be driven by 2.5-V, 3.3-V, or 5.0-V systems. When the VCCIO pins are tied to 3.3 V, the output can drive 3.3-V or 5.0-V systems. MultiVolt I/Os are not supported on 100-pin TQFP or 100-pin FineLine BGA packages.

Table 7	describes	FLFX 6000	MultiVolt I	/O support.
Table /	describes	TLLA UUUU	munu v On i	/ O subboit.

Table 7.	FLEX 600	00 MultiVo	It I/O Sup	port			
V _{CCINT}	V _{CCIO}	Input Signal (V) Output Signal (V)					
(V)	(V)	2.5	3.3	5.0	2.5	3.3	5.0
3.3	2.5	v	V	v	V		
3.3	3.3	v	v	v	v (1)	v	v
5.0	3.3		v	v		v	v
5.0	5.0		V	v			V

Note:

(1) When $V_{\rm CCIO} = 3.3~{\rm V}$, a FLEX 6000 device can drive a 2.5-V device that has 3.3-V tolerant inputs.

Open-drain output pins on 5.0-V or 3.3-V FLEX 6000 devices (with a pull-up resistor to the 5.0-V supply) can drive 5.0-V CMOS input pins that require a $V_{\rm IH}$ of 3.5 V. When the open-drain pin is active, it will drive low. When the pin is inactive, the trace will be pulled up to 5.0 V by the resistor. The open-drain pin will only drive low or tri-state; it will never drive high. The rise time is dependent on the value of the pull-up resistor and load impedance. The $I_{\rm OL}$ current specification should be considered when selecting a pull-up resistor.

Output pins on 5.0-V FLEX 6000 devices with V_{CCIO} = 3.3 V or 5.0 V (with a pull-up resistor to the 5.0-V supply) can also drive 5.0-V CMOS input pins. In this case, the pull-up transistor will turn off when the pin voltage exceeds 3.3 V. Therefore, the pin does not have to be open-drain.

Power Sequencing & Hot-Socketing

Because FLEX 6000 family devices can be used in a mixed-voltage environment, they have been designed specifically to tolerate any possible power-up sequence. The $\rm V_{CCIO}$ and $\rm V_{CCINT}$ power planes can be powered in any order.

Signals can be driven into 3.3-V FLEX 6000 devices before and during power up without damaging the device. Additionally, FLEX 6000 devices do not drive out during power up. Once operating conditions are reached, FLEX 6000 devices operate as specified by the user.

IEEE Std. 1149.1 (JTAG) Boundary-Scan Support

All FLEX 6000 devices provide JTAG BST circuitry that comply with the IEEE Std. 1149.1-1990 specification. Table 8 shows JTAG instructions for FLEX 6000 devices. JTAG BST can be performed before or after configuration, but not during configuration (except when you disable JTAG support in user mode).

See Application Note 39 (IEEE 1149.1 (JTAG) Boundary-Scan Testing in Altera Devices) for more information on JTAG BST circuitry.

Table 8. FLEX 6000	Table 8. FLEX 6000 JTAG Instructions			
JTAG Instruction	Description			
SAMPLE/PRELOAD	Allows a snapshot of the signals at the device pins to be captured and examined during normal device operation, and permits an initial data pattern to be output at the device pins.			
EXTEST	Allows the external circuitry and board-level interconnections to be tested by forcing a test pattern at the output pins and capturing test result at the input pins.			
BYPASS	Places the 1-bit bypass register between the TDI and TDO pins, which allows the BST data to pass synchronously through the selected device to adjacent devices during normal device operation.			

Operating Conditions

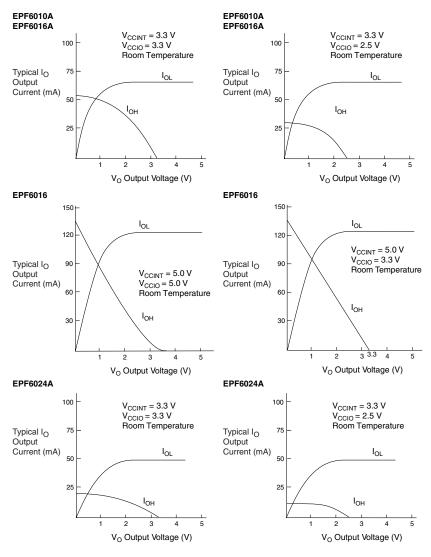
Tables 11 through 18 provide information on absolute maximum ratings, recommended operating conditions, operating conditions, and capacitance for 5.0-V and 3.3-V FLEX 6000 devices.

Table 1	Table 11. FLEX 6000 5.0-V Device Absolute Maximum Ratings Note (1)				
Symbol	Parameter	Conditions	Min	Max	Unit
V _{CC}	Supply voltage	With respect to ground (2)	-2.0	7.0	٧
VI	DC input voltage		-2.0	7.0	V
I _{OUT}	DC output current, per pin		-25	25	mA
T _{STG}	Storage temperature	No bias	-65	150	° C
T _{AMB}	Ambient temperature	Under bias	-65	135	° C
TJ	Junction temperature	PQFP, TQFP, and BGA packages		135	° C

Symbol	Parameter	Conditions	Min	Max	Unit
V _{CCINT}	Supply voltage for internal logic and input buffers	(3), (4)	4.75 (4.50)	5.25 (5.50)	V
V _{CCIO}	Supply voltage for output buffers, 5.0-V operation	(3), (4)	4.75 (4.50)	5.25 (5.50)	V
	Supply voltage for output buffers, 3.3-V operation	(3), (4)	3.00 (3.00)	3.60 (3.60)	V
V _I	Input voltage		-0.5	V _{CCINT} + 0.5	V
Vo	Output voltage		0	V _{CCIO}	V
TJ	Operating temperature	For commercial use	0	85	° C
		For industrial use	-40	100	° C
t _R	Input rise time			40	ns
t _F	Input fall time			40	ns

Figure 18 shows the typical output drive characteristics of 5.0-V and 3.3-V FLEX 6000 devices with 5.0-V, 3.3-V, and 2.5-V $V_{\rm CCIO}$. When $V_{\rm CCIO}=5.0$ V on EPF6016 devices, the output driver is compliant with the *PCI Local Bus Specification, Revision* 2.2 for 5.0-V operation. When $V_{\rm CCIO}=3.3$ V on the EPF6010A and EPF6016A devices, the output driver is compliant with the *PCI Local Bus Specification, Revision* 2.2 for 3.3-V operation.

Figure 18. Output Drive Characteristics



Timing Model

The continuous, high-performance FastTrack Interconnect routing resources ensure predictable performance and accurate simulation and timing analysis. This predictable performance contrasts with that of FPGAs, which use a segmented connection scheme and therefore have unpredictable performance.

Device performance can be estimated by following the signal path from a source, through the interconnect, to the destination. For example, the registered performance between two LEs on the same row can be calculated by adding the following parameters:

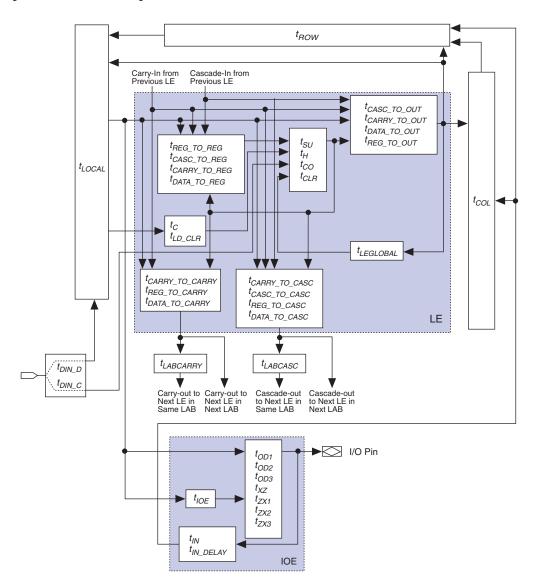
- LE register clock-to-output delay ($t_{CO} + t_{REG_TO_OUT}$)
- Routing delay $(t_{ROW} + t_{LOCAL})$
- LE LUT delay ($t_{DATA_TO_REG}$)
- LE register setup time (t_{SU})

The routing delay depends on the placement of the source and destination LEs. A more complex registered path may involve multiple combinatorial LEs between the source and destination LEs.

Timing simulation and delay prediction are available with the Simulator and Timing Analyzer, or with industry-standard EDA tools. The Simulator offers both pre-synthesis functional simulation to evaluate logic design accuracy and post-synthesis timing simulation with 0.1-ns resolution. The Timing Analyzer provides point-to-point timing delay information, setup and hold time analysis, and device-wide performance analysis.

Figure 19 shows the overall timing model, which maps the possible routing paths to and from the various elements of the FLEX 6000 device.

Figure 19. FLEX 6000 Timing Model



Symbol	Parameter	Conditions
t _{OD1}	Output buffer and pad delay, slow slew rate = off, V _{CCIO} = V _{CCINT}	C1 = 35 pF (2)
t _{OD2}	Output buffer and pad delay, slow slew rate = off, V _{CCIO} = low voltage	C1 = 35 pF (3)
t _{OD3}	Output buffer and pad delay, slow slew rate = on	C1 = 35 pF (4)
t_{XZ}	Output buffer disable delay	C1 = 5 pF
t _{ZX1}	Output buffer enable delay, slow slew rate = off, V _{CCIO} = V _{CCINT}	C1 = 35 pF (2)
t_{ZX2}	Output buffer enable delay, slow slew rate = off, V _{CCIO} = low voltage	C1 = 35 pF (3)
t _{ZX3}	IOE output buffer enable delay, slow slew rate = on	C1 = 35 pF (4)
t _{IOE}	Output enable control delay	
t _{IN}	Input pad and buffer to FastTrack Interconnect delay	
t _{IN_DELAY}	Input pad and buffer to FastTrack Interconnect delay with additional delay turned on	

Table 21. Int	Table 21. Interconnect Timing Microparameters Note (1)				
Symbol	Parameter	Conditions			
t _{LOCAL}	LAB local interconnect delay				
t _{ROW}	Row interconnect routing delay	(5)			
t _{COL}	Column interconnect routing delay	(5)			
t _{DIN_D}	Dedicated input to LE data delay	(5)			
t _{DIN_C}	Dedicated input to LE control delay				
t _{LEGLOBAL}	LE output to LE control via internally-generated global signal delay	(5)			
t _{LABCARRY}	Routing delay for the carry-out of an LE driving the carry-in signal of a different LE in a different LAB				
t _{LABCASC}	Routing delay for the cascade-out signal of an LE driving the cascade-in signal of a different LE in a different LAB				

Table 22. External Reference Timing Parameters		
Symbol	Parameter	Conditions
t ₁	Register-to-register test pattern	(6)
t _{DRR}	Register-to-register delay via 4 LEs, 3 row interconnects, and 4 local interconnects	(7)

Table 23. External Timing Parameters			
Symbol	Parameter	Conditions	
t _{INSU}	Setup time with global clock at LE register	(8)	
t _{INH}	Hold time with global clock at LE register	(8)	
t _{оитсо}	Clock-to-output delay with global clock with LE register using FastFLEX I/O pin	(8)	

Notes to tables:

- Microparameters are timing delays contributed by individual architectural elements and cannot be measured explicitly.
- (2) Operating conditions:
 - $\hat{V_{CCIO}} = \widecheck{5}.0~V \pm 5\%$ for commercial use in 5.0-V FLEX 6000 devices.
 - $V_{CCIO} = 5.0 \text{ V} \pm 10\%$ for industrial use in 5.0-V FLEX 6000 devices.
 - $V_{CCIO} = 3.3 \text{ V} \pm 10\%$ for commercial or industrial use in 3.3-V FLEX 6000 devices.
- (3) Operating conditions:
 - $V_{CCIO} = 3.3 \text{ V} \pm 10\%$ for commercial or industrial use in 5.0-V FLEX 6000 devices.
 - V_{CCIO} = 2.5 V ±0.2 V for commercial or industrial use in 3.3-V FLEX 6000 devices.
- (4) Operating conditions:
 - $V_{\text{CCIO}} = 2.5 \text{ V}, 3.3 \text{ V}, \text{ or } 5.0 \text{ V}.$
- (5) These parameters are worst-case values for typical applications. Post-compilation timing simulation and timing analysis are required to determine actual worst-case performance.
- (6) This timing parameter shows the delay of a register-to-register test pattern and is used to determine speed grades. There are 12 LEs, including source and destination registers. The row and column interconnects between the registers vary in length.
- 7) This timing parameter is shown for reference and is specified by characterization.
- (8) This timing parameter is specified by characterization.

Tables 24 through 28 show the timing information for EPF6010A and EPF6016A devices.

Parameter			Speed	Grade			Unit
	_	1	-	2	-	3	
	Min	Max	Min	Max	Min	Max	
treg_to_reg		1.2		1.3		1.7	ns
[†] CASC_TO_REG		0.9		1.0		1.2	ns
t _{CARRY_TO_REG}		0.9		1.0		1.2	ns
^t DATA_TO_REG		1.1		1.2		1.5	ns
tcasc_to_out		1.3		1.4		1.8	ns
t _{CARRY_TO_OUT}		1.6		1.8		2.3	ns
t _{DATA_TO_OUT}		1.7		2.0		2.5	ns
t _{REG_TO_OUT}		0.4		0.4		0.5	ns
t _{su}	0.9		1.0		1.3		ns
t _H	1.4		1.7		2.1		ns

Parameter	Speed Grade							
	-1		-2		-3		1	
	Min	Max	Min	Max	Min	Max		
t _{LOCAL}		0.7		0.7		1.0	ns	
t _{ROW}		2.9		3.2		3.2	ns	
t _{COL}		1.2		1.3		1.4	ns	
t _{DIN_D}		5.4		5.7		6.4	ns	
t _{DIN_C}		4.3		5.0		6.1	ns	
[†] LEGLOBAL		2.6		3.0		3.7	ns	
t _{LABCARRY}		0.7		0.8		0.9	ns	
t _{LABCASC}		1.3		1.4		1.8	ns	

Table 27. External Reference Timing Parameters for EPF6010A & EPF6016A Devices								
Parameter	Device	rice Speed Grade						Unit
		-	-1		-2		-3	
		Min	Max	Min	Max	Min	Max	
t ₁	EPF6010A		37.6		43.6		53.7	ns
	EPF6016A		38.0		44.0		54.1	ns

Table 28. External Timing Parameters for EPF6010A & EPF6016A Devices								
Parameter	Speed Grade							
	-1		-2		-3			
	Min	Max	Min	Max	Min	Max		
t _{INSU}	2.1 (1)		2.4 (1)		3.3 (1)		ns	
t _{INH}	0.2 (2)		0.3 (2)		0.1 (2)		ns	
t _{оитсо}	2.0	7.1	2.0	8.2	2.0	10.1	ns	

Notes:

Setup times are longer when the *Increase Input Delay* option is turned on. The setup time values are shown with the *Increase Input Delay* option turned off.
 Hold time is zero when the *Increase Input Delay* option is turned on.

Table 38. External Timing Parameters for EPF6024A Devices								
Parameter	Speed Grade							
	-1		-2		-3		1	
	Min	Max	Min	Max	Min	Max		
t _{INSU}	2.0 (1)		2.2 (1)		2.6 (1)		ns	
t _{INH}	0.2 (2)		0.2 (2)		0.3 (2)		ns	
t _{outco}	2.0	7.4	2.0	8.2	2.0	9.9	ns	

Notes:

- (1) Setup times are longer when the *Increase Input Delay* option is turned on. The setup time values are shown with the *Increase Input Delay* option turned off.
- (2) Hold time is zero when the *Increase Input Delay* option is turned on.

Power Consumption

The supply power (P) for FLEX 6000 devices can be calculated with the following equations:

$$\begin{array}{ll} P &=& P_{INT} + P_{IO} \\ P &=& (I_{CCSTANDBY} + I_{CCACTIVE}) \times V_{CC} + P_{IO} \end{array}$$

Typical $I_{CCSTANDBY}$ values are shown as I_{CC0} in the "FLEX 6000 Device DC Operating Conditions" table on pages 31 and 33 of this data sheet. The $I_{CCACTIVE}$ value depends on the switching frequency and the application logic. This value is based on the amount of current that each LE typically consumes. The P_{IO} value, which depends on the device output load characteristics and switching frequency, can be calculated using the guidelines given in *Application Note 74 (Evaluating Power for Altera Devices)*.

The I_{CCACTIVE} value can be calculated with the following equation:

$$I_{CCACTIVE} = K \times f_{MAX} \times N \times tog_{LC} \times \frac{\mu A}{MHz \times LE}$$

Where:

 f_{MAX} = Maximum operating frequency in MHz

N = Total number of LEs used in a FLEX 6000 device tog_{LC} = Average percentage of LEs toggling at each clock

(typically 12.5%)

K = Constant, shown in Table 39

Table 39. K Constant Values					
Device	K Value				
EPF6010A	14				
EPF6016	88				
EPF6016A	14				
EPF6024A	14				

This calculation provides an I_{CC} estimate based on typical conditions with no output load. The actual I_{CC} should be verified during operation because this measurement is sensitive to the actual pattern in the device and the environmental operating conditions.

To better reflect actual designs, the power model (and the constant K in the power calculation equations shown above) for continuous interconnect FLEX devices assumes that LEs drive FastTrack Interconnect channels. In contrast, the power model of segmented FPGAs assumes that all LEs drive only one short interconnect segment. This assumption may lead to inaccurate results, compared to measured power consumption for an actual design in a segmented interconnect FPGA.

Figure 20 shows the relationship between the current and operating frequency for EPF6010A, EPF6016, EPF6016A, and EPF6024A devices.

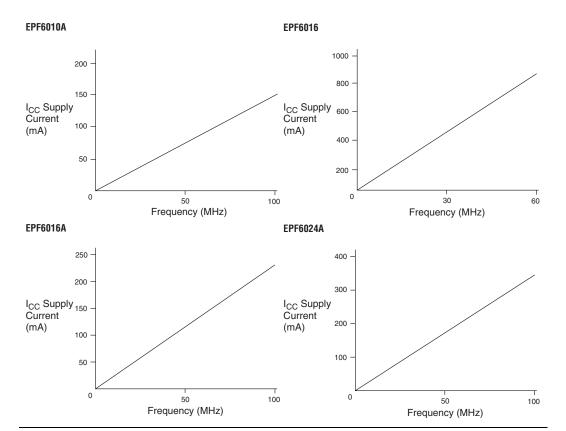


Figure 20. I_{CCACTIVE} vs. Operating Frequency

Device Configuration & Operation

The FLEX 6000 architecture supports several configuration schemes to load a design into the device(s) on the circuit board. This section summarizes the device operating modes and available device configuration schemes.

See Application Note 116 (Configuring APEX 20K, FLEX 10K & FLEX 6000 Devices) for detailed information on configuring FLEX 6000 devices, including sample schematics, timing diagrams, configuration options, pins names, and timing parameters.