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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

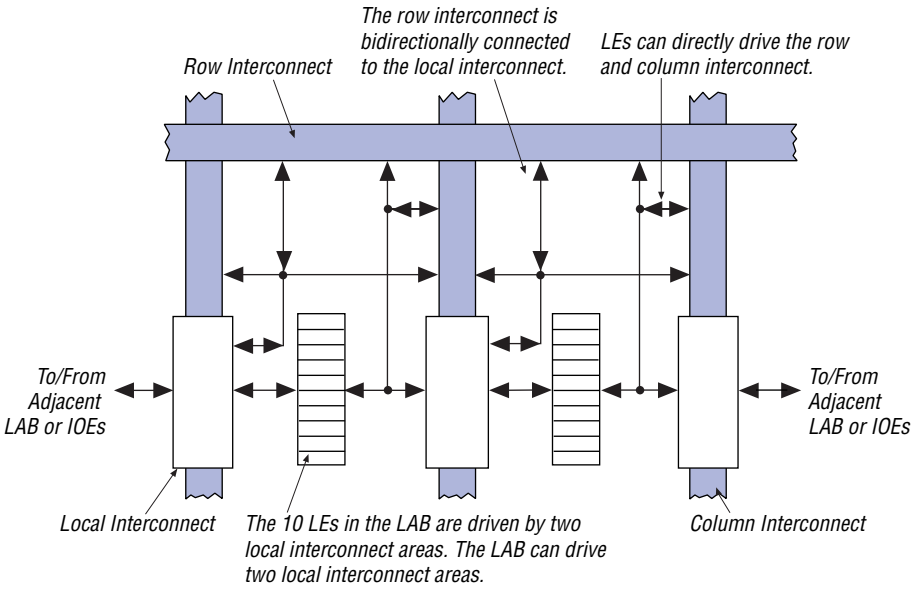
Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	196
Number of Logic Elements/Cells	1960
Total RAM Bits	-
Number of I/O	117
Number of Gates	24000
Voltage - Supply	3V ~ 3.6V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	144-LQFP
Supplier Device Package	144-TQFP (20x20)
Purchase URL	https://www.e-xfl.com/product-detail/intel/epf6024atc144-2n

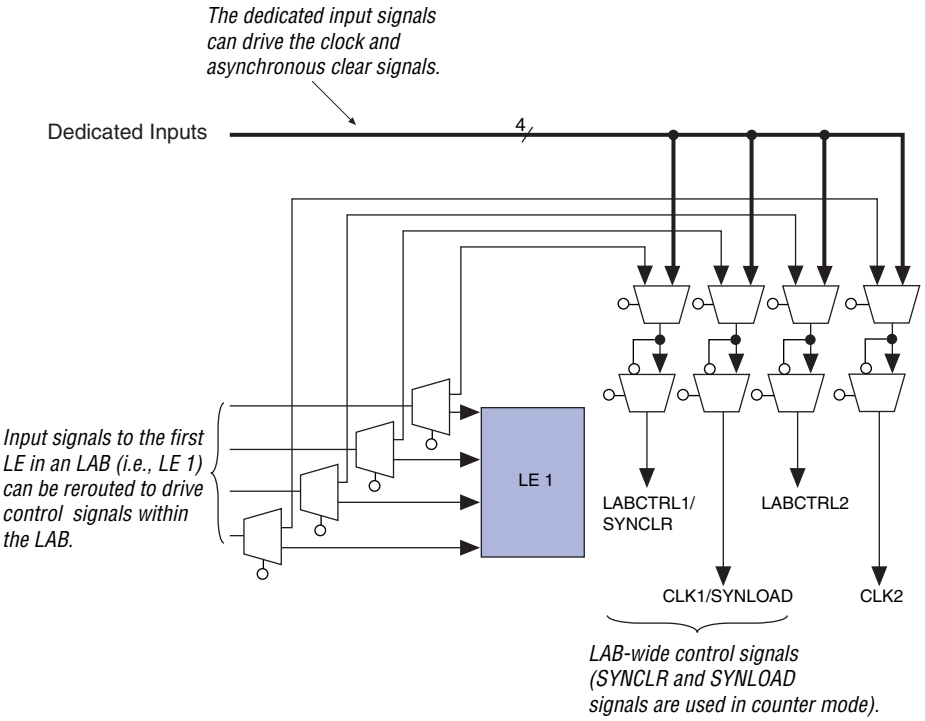
The interleaved LAB structure—an innovative feature of the FLEX 6000 architecture—allows each LAB to drive two local interconnects. This feature minimizes the use of the FastTrack Interconnect, providing higher performance. An LAB can drive 20 LEs in adjacent LABs via the local interconnect, which maximizes fitting flexibility while minimizing die size. See Figure 2.

Figure 2. Logic Array Block



In most designs, the registers only use global clock and clear signals. However, in some cases, other clock or asynchronous clear signals are needed. In addition, counters may also have synchronous clear or load signals. In a design that uses non-global clock and clear signals, inputs from the first LE in an LAB are re-routed to drive the control signals for that LAB. See Figure 3.

Figure 3. LAB Control Signals



Logic Element

An LE, the smallest unit of logic in the FLEX 6000 architecture, has a compact size that provides efficient logic usage. Each LE contains a four-input LUT, which is a function generator that can quickly implement any function of four variables. An LE contains a programmable flipflop, carry and cascade chains. Additionally, each LE drives both the local and the FastTrack Interconnect. See Figure 4.

Carry Chain

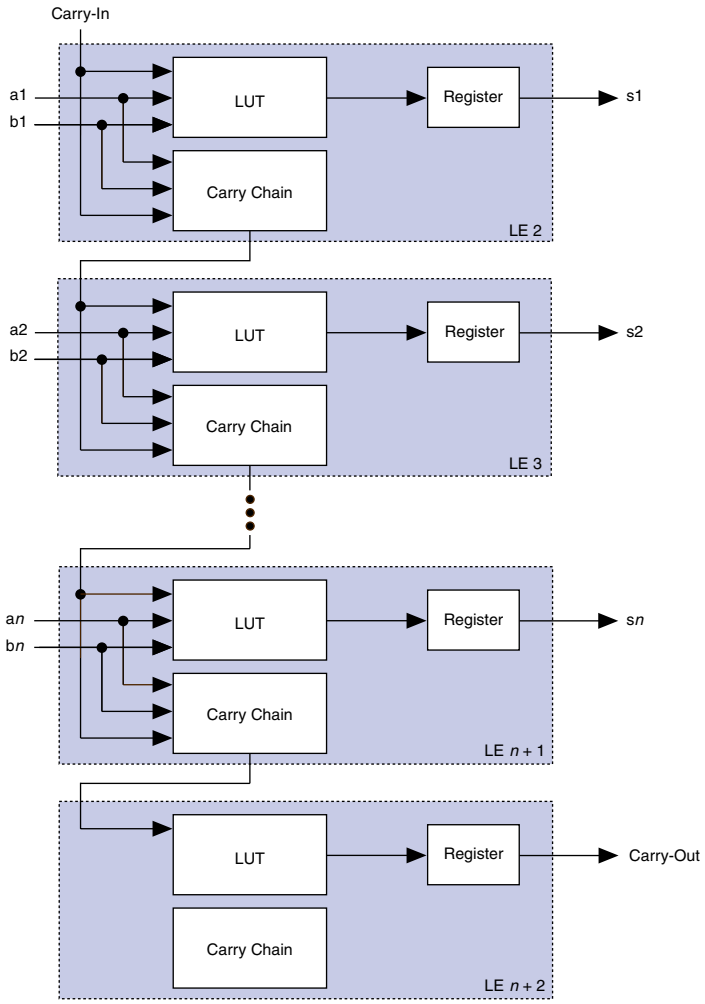
The carry chain provides a very fast (0.1 ns) carry-forward function between LEs. The carry-in signal from a lower-order bit drives forward into the higher-order bit via the carry chain, and feeds into both the LUT and the next portion of the carry chain. This feature allows the FLEX 6000 architecture to implement high-speed counters, adders, and comparators of arbitrary width. Carry chain logic can be created automatically by the Altera software during design processing, or manually by the designer during design entry. Parameterized functions such as LPM and DesignWare functions automatically take advantage of carry chains for the appropriate functions.

Because the first LE of each LAB can generate control signals for that LAB, the first LE in each LAB is not included in carry chains. In addition, the inputs of the first LE in each LAB may be used to generate synchronous clear and load enable signals for counters implemented with carry chains.

Carry chains longer than nine LEs are implemented automatically by linking LABs together. For enhanced fitting, a long carry chain skips alternate LABs in a row. A carry chain longer than one LAB skips either from an even-numbered LAB to another even-numbered LAB, or from an odd-numbered LAB to another odd-numbered LAB. For example, the last LE of the first LAB in a row carries to the second LE of the third LAB in the row. In addition, the carry chain does not cross the middle of the row. For instance, in the EPF6016 device, the carry chain stops at the 11th LAB in a row and a new carry chain begins at the 12th LAB.

Figure 5 shows how an *n*-bit full adder can be implemented in *n* + 1 LEs with the carry chain. One portion of the LUT generates the sum of two bits using the input signals and the carry-in signal; the sum is routed to the output of the LE. Although the register can be bypassed for simple adders, it can be used for an accumulator function. Another portion of the LUT and the carry chain logic generates the carry-out signal, which is routed directly to the carry-in signal of the next-higher-order bit. The final carry-out signal is routed to an LE, where it is driven onto the FastTrack Interconnect.

Figure 5. Carry Chain Operation



Normal Mode

The normal mode is suitable for general logic applications, combinatorial functions, or wide decoding functions that can take advantage of a cascade chain. In normal mode, four data inputs from the LAB local interconnect and the carry-in are inputs to a 4-input LUT. The Altera software automatically selects the carry-in or the DATA3 signal as one of the inputs to the LUT. The LUT output can be combined with the cascade-in signal to form a cascade chain through the cascade-out signal.

Arithmetic Mode

The arithmetic mode is ideal for implementing adders, accumulators, and comparators. An LE in arithmetic mode uses two 3-input LUTs. One LUT computes a 3-input function; the other generates a carry output. As shown in Figure 7, the first LUT uses the carry-in signal and two data inputs from the LAB local interconnect to generate a combinatorial or registered output. For example, when implementing an adder, this output is the sum of three signals: DATA1, DATA2, and carry-in. The second LUT uses the same three signals to generate a carry-out signal, thereby creating a carry chain. The arithmetic mode also supports simultaneous use of the cascade chain.

The Altera software implements logic functions to use the arithmetic mode automatically where appropriate; the designer does not have to decide how the carry chain will be used.

Counter Mode

The counter mode offers counter enable, synchronous up/down control, synchronous clear, and synchronous load options. The counter enable and synchronous up/down control signals are generated from the data inputs of the LAB local interconnect. The synchronous clear and synchronous load options are LAB-wide signals that affect all registers in the LAB. Consequently, if any of the LEs in a LAB use counter mode, other LEs in that LAB must be used as part of the same counter or be used for a combinatorial function. In addition, the Altera software automatically places registers that are not in the counter into other LABs.

The counter mode uses two 3-input LUTs: one generates the counter data and the other generates the fast carry bit. A 2-to-1 multiplexer provides synchronous loading, and another AND gate provides synchronous clearing. If the cascade function is used by an LE in counter mode, the synchronous clear or load will override any signal carried on the cascade chain. The synchronous clear overrides the synchronous load.

Either the counter enable or the up/down control may be used for a given counter. Moreover, the synchronous load can be used as a count enable by routing the register output into the data input automatically when requested by the designer.

The second LE of each LAB has a special function for counter mode; the carry-in of the LE can be driven by a fast feedback path from the register. This function gives a faster counter speed for counter carry chains starting in the second LE of an LAB.

The Altera software implements functions to use the counter mode automatically where appropriate. The designer does not have to decide how the carry chain will be used.

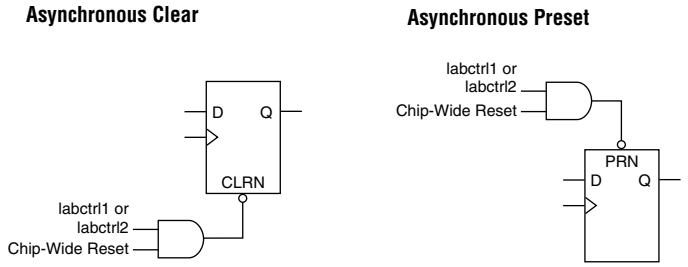
Internal Tri-State Emulation

Internal tri-state emulation provides internal tri-states without the limitations of a physical tri-state bus. In a physical tri-state bus, the tri-state buffers' output enable (OE) signals select which signal drives the bus. However, if multiple OE signals are active, contending signals can be driven onto the bus. Conversely, if no OE signals are active, the bus will float. Internal tri-state emulation resolves contending tri-state buffers to a low value and floating buses to a high value, thereby eliminating these problems. The Altera software automatically implements tri-state bus functionality with a multiplexer.

Clear & Preset Logic Control

Logic for the programmable register's clear and preset functions is controlled by the LAB-wide signals `LABCTRL1` and `LABCTRL2`. The LE register has an asynchronous clear that can implement an asynchronous preset. Either `LABCTRL1` or `LABCTRL2` can control the asynchronous clear or preset. Because the clear and preset functions are active-low, the Altera software automatically assigns a logic high to an unused clear or preset signal. The clear and preset logic is implemented in either the asynchronous clear or asynchronous preset mode, which is chosen during design entry (see Figure 8).

Figure 8. LE Clear & Preset Modes



Asynchronous Clear

The flipflop can be cleared by either LABCTRL1 or LABCTRL2.

Asynchronous Preset

An asynchronous preset is implemented with an asynchronous clear. The Altera software provides preset control by using the clear and inverting the input and output of the register. Inversion control is available for the inputs to both LEs and IOEs. Therefore, this technique can be used when a register drives logic or drives a pin.

In addition to the two clear and preset modes, FLEX 6000 devices provide a chip-wide reset pin (DEV_CLRn) that can reset all registers in the device. The option to use this pin is set in the Altera software before compilation. The chip-wide reset overrides all other control signals. Any register with an asynchronous preset will be preset when the chip-wide reset is asserted because of the inversion technique used to implement the asynchronous preset.

The Altera software can use a programmable NOT-gate push-back technique to emulate simultaneous preset and clear or asynchronous load. However, this technique uses an additional three LEs per register.

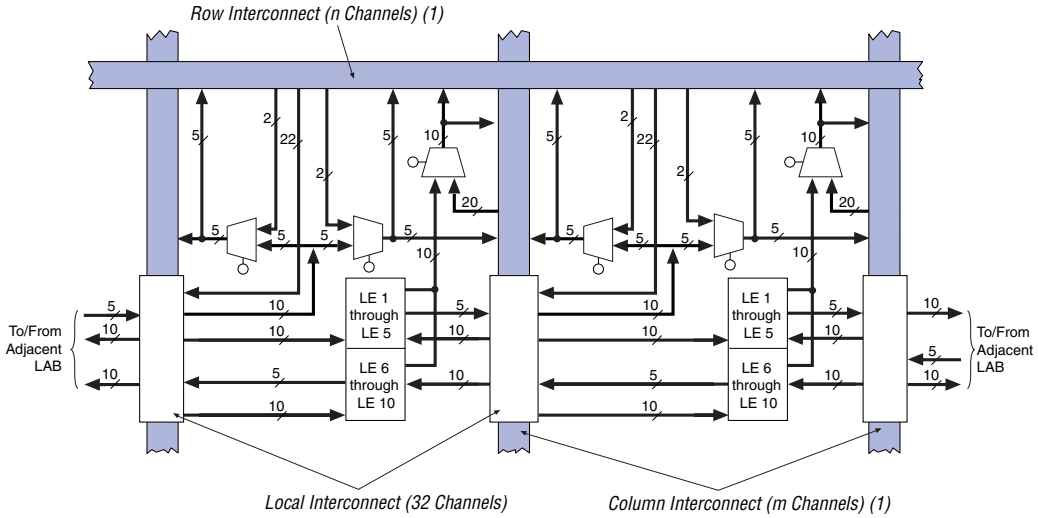
FastTrack Interconnect

In the FLEX 6000 OptiFLEX architecture, connections between LEs and device I/O pins are provided by the FastTrack Interconnect, a series of continuous horizontal and vertical routing channels that traverse the device. This global routing structure provides predictable performance, even for complex designs. In contrast, the segmented routing in FPGAs requires switch matrices to connect a variable number of routing paths, increasing the delays between logic resources and reducing performance.

The FastTrack Interconnect consists of column and row interconnect channels that span the entire device. Each row of LABs is served by a dedicated row interconnect, which routes signals between LABs in the same row, and also routes signals from I/O pins to LABs. Additionally, the local interconnect routes signals between LEs in the same LAB and in adjacent LABs. The column interconnect routes signals between rows and routes signals from I/O pins to rows.

LEs 1 through 5 of an LAB drive the local interconnect to the right, while LEs 6 through 10 drive the local interconnect to the left. The `DATA1` and `DATA3` inputs of each LE are driven by the local interconnect to the left; `DATA2` and `DATA4` are driven by the local interconnect to the right. The local interconnect also routes signals from LEs to I/O pins. Figure 9 shows an overview of the FLEX 6000 interconnect architecture. LEs in the first and last columns have drivers on both sides so that all LEs in the LAB can drive I/O pins via the local interconnect.

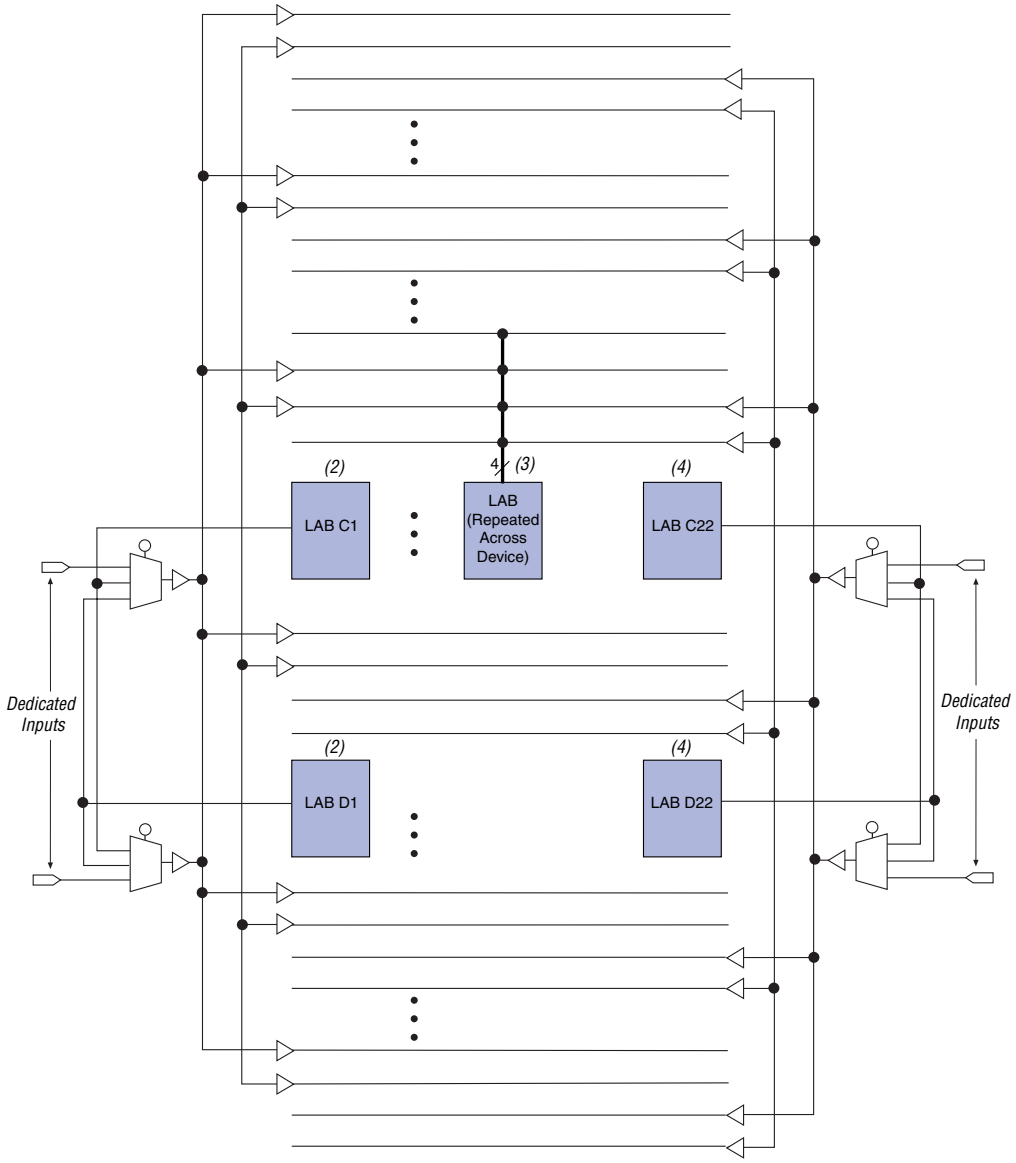
Figure 9. FastTrack Interconnect Architecture



Note:

(1) For EPF6010A, EPF6016, and EPF6016A devices, n = 144 channels and m = 20 channels; for EPF6024A devices, n = 186 channels and m = 30 channels.

Figure 11. Global Clock & Clear Distribution *Note (1)*



Notes:

(1) The global clock and clear distribution signals are shown for EPF6016 and EPF6016A devices. In EPF6010A devices, LABs in rows B and C drive global signals. In EPF6024A devices, LABs in rows C and E drive global signals.
(2) The local interconnect from LABs C1 and D1 can drive two global control signals on the left side.
(3) Global signals drive into every LAB as clock, asynchronous clear, preset, and data signals.
(4) The local interconnect from LABs C22 and D22 can drive two global control signals on the right side.

Figure 15. SameFrame Pin-Out Example

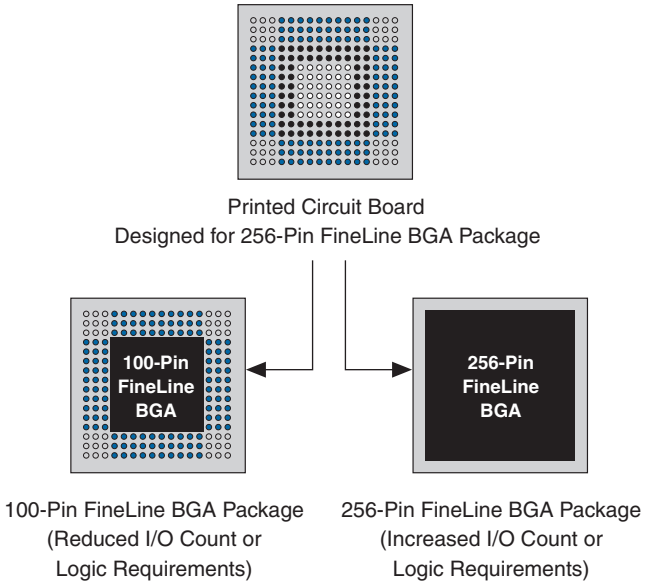


Table 6 lists the 3.3-V FLEX 6000 devices with the SameFrame pin-out feature.

Table 6. 3.3-V FLEX 6000 Devices with SameFrame Pin-Outs

Device	100-Pin FineLine BGA	256-Pin FineLine BGA
EPF6016A	v	v
EPF6024A		v

Output Configuration

This section discusses slew-rate control, the MultiVolt I/O interface, power sequencing, and hot-socketing for FLEX 6000 devices.

Slew-Rate Control

The output buffer in each IOE has an adjustable output slew-rate that can be configured for low-noise or high-speed performance. A slower slew-rate reduces system noise and adds a maximum delay of 6.8 ns. The fast slew-rate should be used for speed-critical outputs in systems that are adequately protected against noise. Designers can specify the slew-rate on a pin-by-pin basis during design entry or assign a default slew rate to all pins on a device-wide basis. The slew-rate setting affects only the falling edge of the output.

Open-drain output pins on 5.0-V or 3.3-V FLEX 6000 devices (with a pull-up resistor to the 5.0-V supply) can drive 5.0-V CMOS input pins that require a V_{IH} of 3.5 V. When the open-drain pin is active, it will drive low. When the pin is inactive, the trace will be pulled up to 5.0 V by the resistor. The open-drain pin will only drive low or tri-state; it will never drive high. The rise time is dependent on the value of the pull-up resistor and load impedance. The I_{OL} current specification should be considered when selecting a pull-up resistor.

Output pins on 5.0-V FLEX 6000 devices with V_{CCIO} = 3.3 V or 5.0 V (with a pull-up resistor to the 5.0-V supply) can also drive 5.0-V CMOS input pins. In this case, the pull-up transistor will turn off when the pin voltage exceeds 3.3 V. Therefore, the pin does not have to be open-drain.

Power Sequencing & Hot-Socketing

Because FLEX 6000 family devices can be used in a mixed-voltage environment, they have been designed specifically to tolerate any possible power-up sequence. The V_{CCIO} and V_{CCINT} power planes can be powered in any order.

Signals can be driven into 3.3-V FLEX 6000 devices before and during power up without damaging the device. Additionally, FLEX 6000 devices do not drive out during power up. Once operating conditions are reached, FLEX 6000 devices operate as specified by the user.

IEEE Std. 1149.1 (JTAG) Boundary-Scan Support

All FLEX 6000 devices provide JTAG BST circuitry that comply with the IEEE Std. 1149.1-1990 specification. Table 8 shows JTAG instructions for FLEX 6000 devices. JTAG BST can be performed before or after configuration, but not during configuration (except when you disable JTAG support in user mode).

1 See *Application Note 39 (IEEE 1149.1 (JTAG) Boundary-Scan Testing in Altera Devices)* for more information on JTAG BST circuitry.

Table 8. FLEX 6000 JTAG Instructions

JTAG Instruction	Description
SAMPLE/PRELOAD	Allows a snapshot of the signals at the device pins to be captured and examined during normal device operation, and permits an initial data pattern to be output at the device pins.
EXTEST	Allows the external circuitry and board-level interconnections to be tested by forcing a test pattern at the output pins and capturing test result at the input pins.
BYPASS	Places the 1-bit bypass register between the `TDI` and `TDO` pins, which allows the BST data to pass synchronously through the selected device to adjacent devices during normal device operation.

Operating Conditions

Tables 11 through 18 provide information on absolute maximum ratings, recommended operating conditions, operating conditions, and capacitance for 5.0-V and 3.3-V FLEX 6000 devices.

Table 11. FLEX 6000 5.0-V Device Absolute Maximum Ratings *Note (1)*

Symbol	Parameter	Conditions	Min	Max	Unit
V_{CC}	Supply voltage	With respect to ground *(2)*	–2.0	7.0	V
V_I	DC input voltage		–2.0	7.0	V
I_{OUT}	DC output current, per pin		–25	25	mA
T_{STG}	Storage temperature	No bias	–65	150	° C
T_{AMB}	Ambient temperature	Under bias	–65	135	° C
T_J	Junction temperature	PQFP, TQFP, and BGA packages		135	° C

Table 12. FLEX 6000 5.0-V Device Recommended Operating Conditions

Symbol	Parameter	Conditions	Min	Max	Unit
V_{CCINT}	Supply voltage for internal logic and input buffers	*(3)*, *(4)*	4.75 (4.50)	5.25 (5.50)	V
V_{CCIO}	Supply voltage for output buffers, 5.0-V operation	*(3)*, *(4)*	4.75 (4.50)	5.25 (5.50)	V
	Supply voltage for output buffers, 3.3-V operation	*(3)*, *(4)*	3.00 (3.00)	3.60 (3.60)	V
V_I	Input voltage		–0.5	V_{CCINT} + 0.5	V
V_O	Output voltage		0	V_{CCIO}	V
T_J	Operating temperature	For commercial use	0	85	° C
		For industrial use	–40	100	° C
t_R	Input rise time			40	ns
t_F	Input fall time			40	ns

Table 17. FLEX 6000 3.3-V Device DC Operating Conditions *Notes (5), (6)*

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V_{IH}	High-level input voltage		1.7		5.75	V
V_{IL}	Low-level input voltage		–0.5		0.8	V
V_{OH}	3.3-V high-level TTL output voltage	I_{OH} = –8 mA DC, V_{CCIO} = 3.00 V *(7)*	2.4			V
	3.3-V high-level CMOS output voltage	I_{OH} = –0.1 mA DC, V_{CCIO} = 3.00 V *(7)*	V_{CCIO} – 0.2			V
	2.5-V high-level output voltage	I_{OH} = –100 µA DC, V_{CCIO} = 2.30 V *(7)*	2.1			V
		I_{OH} = –1 mA DC, V_{CCIO} = 2.30 V *(7)*	2.0			V
		I_{OH} = –2 mA DC, V_{CCIO} = 2.30 V *(7)*	1.7			V
V_{OL}	3.3-V low-level TTL output voltage	I_{OL} = 8 mA DC, V_{CCIO} = 3.00 V *(8)*			0.45	V
	3.3-V low-level CMOS output voltage	I_{OL} = 0.1 mA DC, V_{CCIO} = 3.00 V *(8)*			0.2	V
	2.5-V low-level output voltage	I_{OL} = 100 µA DC, V_{CCIO} = 2.30 V *(8)*			0.2	V
		I_{OL} = 1 mA DC, V_{CCIO} = 2.30 V *(8)*			0.4	V
		I_{OL} = 2 mA DC, V_{CCIO} = 2.30 V *(8)*			0.7	V
I_I	Input pin leakage current	V_I = 5.3 V to ground *(8)*	–10		10	µA
I_{OZ}	Tri-stated I/O pin leakage current	V_O = 5.3 V to ground *(8)*	–10		10	µA
I_{CC0}	V_{CC} supply current (standby)	V_I = ground, no load		0.5	5	mA

Table 18. FLEX 6000 3.3-V Device Capacitance *Note (9)*

Symbol	Parameter	Conditions	Min	Max	Unit
C_{IN}	Input capacitance for I/O pin	V_{IN} = 0 V, f = 1.0 MHz		8	pF
C_{INCLK}	Input capacitance for dedicated input	V_{IN} = 0 V, f = 1.0 MHz		12	pF
C_{OUT}	Output capacitance	V_{OUT} = 0 V, f = 1.0 MHz		8	pF

Notes to tables:

(1) See the *Operating Requirements for Altera Devices Data Sheet*.
(2) The minimum DC input voltage is –0.5 V. During transitions, the inputs may undershoot to –2.0 V or overshoot to 5.75 V for input currents less than 100 mA and periods shorter than 20 ns.
(3) Numbers in parentheses are for industrial-temperature-range devices.
(4) Maximum V_{CC} rise time is 100 ms. V_{CC} must rise monotonically.
(5) Typical values are for T_A = 25° C and V_{CC} = 3.3 V.
(6) These values are specified under Table 16 on page 33.
(7) The I_{OH} parameter refers to high-level TTL or CMOS output current.
(8) The I_{OL} parameter refers to low-level TTL, PCI, or CMOS output current. This parameter applies to open-drain pins as well as output pins.
(9) Capacitance is sample-tested only.

Figure 18 shows the typical output drive characteristics of 5.0-V and 3.3-V FLEX 6000 devices with 5.0-V, 3.3-V, and 2.5-V V_{CCIO}. When V_{CCIO} = 5.0 V on EPF6016 devices, the output driver is compliant with the ***PCI Local Bus Specification, Revision 2.2*** for 5.0-V operation. When V_{CCIO} = 3.3 V on the EPF6010A and EPF6016A devices, the output driver is compliant with the ***PCI Local Bus Specification, Revision 2.2*** for 3.3-V operation.

Figure 18. Output Drive Characteristics

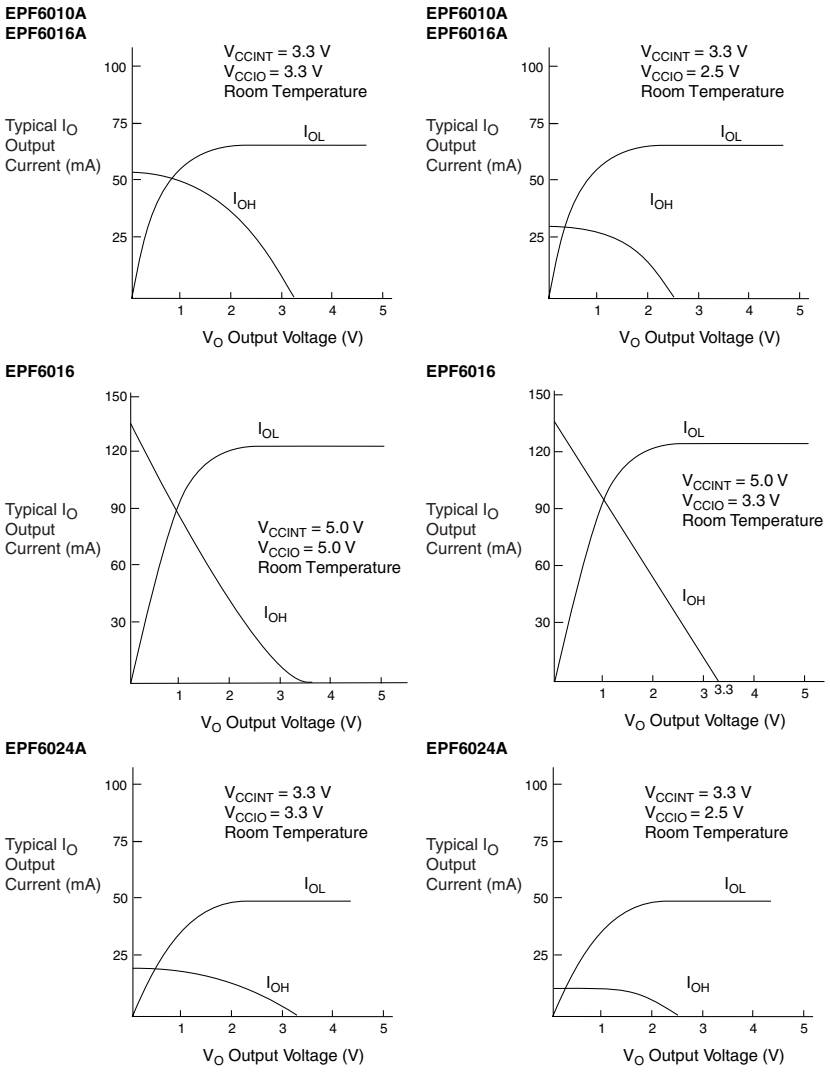


Figure 19. FLEX 6000 Timing Model

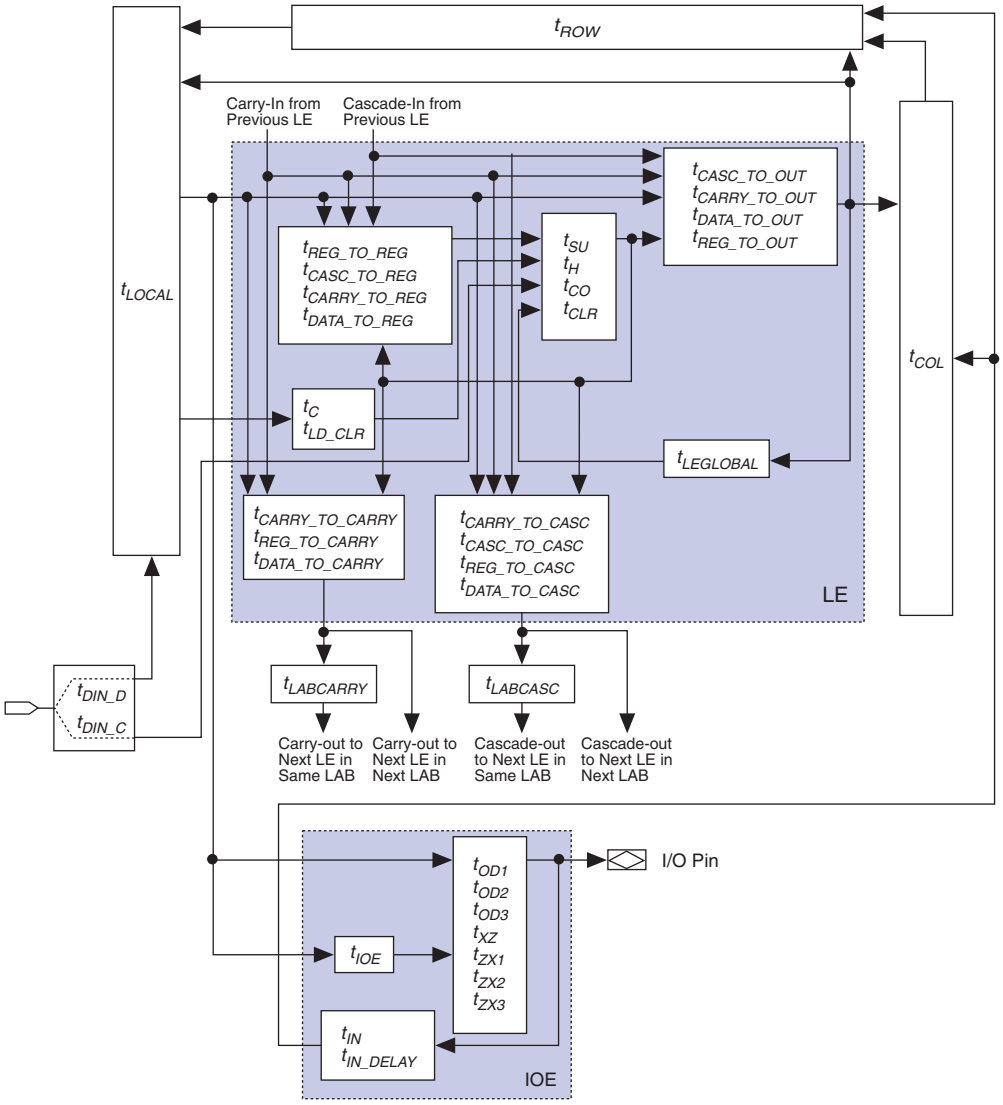


Table 23. External Timing Parameters		
Symbol	**Parameter**	**Conditions**
t_{INSU}	Setup time with global clock at LE register	*(8)*
t_{INH}	Hold time with global clock at LE register	*(8)*
t_{OUTCO}	Clock-to-output delay with global clock with LE register using FastFLEX I/O pin	*(8)*

Notes to tables:

(1) Microparameters are timing delays contributed by individual architectural elements and cannot be measured explicitly.
(2) Operating conditions:
V_{CCIO} = 5.0 V ±5% for commercial use in 5.0-V FLEX 6000 devices.
V_{CCIO} = 5.0 V ±10% for industrial use in 5.0-V FLEX 6000 devices.
V_{CCIO} = 3.3 V ±10% for commercial or industrial use in 3.3-V FLEX 6000 devices.
(3) Operating conditions:
V_{CCIO} = 3.3 V ±10% for commercial or industrial use in 5.0-V FLEX 6000 devices.
V_{CCIO} = 2.5 V ±0.2 V for commercial or industrial use in 3.3-V FLEX 6000 devices.
(4) Operating conditions:
V_{CCIO} = 2.5 V, 3.3 V, or 5.0 V.
(5) These parameters are worst-case values for typical applications. Post-compilation timing simulation and timing analysis are required to determine actual worst-case performance.
(6) This timing parameter shows the delay of a register-to-register test pattern and is used to determine speed grades. There are 12 LEs, including source and destination registers. The row and column interconnects between the registers vary in length.
(7) This timing parameter is shown for reference and is specified by characterization.
(8) This timing parameter is specified by characterization.

Tables 24 through 28 show the timing information for EPF6010A and EPF6016A devices.

Table 24. LE Timing Microparameters for EPF6010A & EPF6016A Devices (Part 1 of 2)							
Parameter	**Speed Grade**						**Unit**
	-1		**-2**		**-3**		
	Min	**Max**	**Min**	**Max**	**Min**	**Max**	
$t_{REG\_TO\_REG}$		1.2		1.3		1.7	ns
$t_{CASC\_TO\_REG}$		0.9		1.0		1.2	ns
$t_{CARRY\_TO\_REG}$		0.9		1.0		1.2	ns
$t_{DATA\_TO\_REG}$		1.1		1.2		1.5	ns
$t_{CASC\_TO\_OUT}$		1.3		1.4		1.8	ns
$t_{CARRY\_TO\_OUT}$		1.6		1.8		2.3	ns
$t_{DATA\_TO\_OUT}$		1.7		2.0		2.5	ns
$t_{REG\_TO\_OUT}$		0.4		0.4		0.5	ns
t_{SU}	0.9		1.0		1.3		ns
t_{H}	1.4		1.7		2.1		ns

Table 26. Interconnect Timing Microparameters for EPF6010A & EPF6016A Devices

Parameter	Speed Grade						Unit
	-1		-2		-3		
	Min	Max	Min	Max	Min	Max	
t_{LOCAL}		0.7		0.7		1.0	ns
t_{ROW}		2.9		3.2		3.2	ns
t_{COL}		1.2		1.3		1.4	ns
t_{DIN_D}		5.4		5.7		6.4	ns
t_{DIN_C}		4.3		5.0		6.1	ns
$t_{LEGLOBAL}$		2.6		3.0		3.7	ns
$t_{LABCARRY}$		0.7		0.8		0.9	ns
$t_{LABCASC}$		1.3		1.4		1.8	ns

Table 27. External Reference Timing Parameters for EPF6010A & EPF6016A Devices

Parameter	Device	Speed Grade						Unit
		-1		-2		-3		
		Min	Max	Min	Max	Min	Max	
t_1	EPF6010A		37.6		43.6		53.7	ns
	EPF6016A		38.0		44.0		54.1	ns

Table 28. External Timing Parameters for EPF6010A & EPF6016A Devices

Parameter	Speed Grade						Unit
	-1		-2		-3		
	Min	Max	Min	Max	Min	Max	
t_{INSU}	2.1 *(1)*		2.4 *(1)*		3.3 *(1)*		ns
t_{INH}	0.2 *(2)*		0.3 *(2)*		0.1 *(2)*		ns
t_{OUTCO}	2.0	7.1	2.0	8.2	2.0	10.1	ns

Notes:
(1) Setup times are longer when the *Increase Input Delay* option is turned on. The setup time values are shown with the *Increase Input Delay* option turned off.
(2) Hold time is zero when the *Increase Input Delay* option is turned on.

Table 30. IOE Timing Microparameters for EPF6016 Devices

Parameter	Speed Grade				Unit
	-2		-3		
	Min	Max	Min	Max	
t_{OD3}		4.7		5.2	ns
t_{XZ}		2.3		2.8	ns
t_{ZX1}		2.3		2.8	ns
t_{ZX2}		4.6		5.1	ns
t_{ZX3}		4.7		5.2	ns
t_{IOE}		0.5		0.6	ns
t_{IN}		3.3		4.0	ns
t_{IN_DELAY}		4.6		5.6	ns

Table 31. Interconnect Timing Microparameters for EPF6016 Devices

Parameter	Speed Grade				Unit
	-2		-3		
	Min	Max	Min	Max	
t_{LOCAL}		0.8		1.0	ns
t_{ROW}		2.9		3.3	ns
t_{COL}		2.3		2.5	ns
t_{DIN_D}		4.9		6.0	ns
t_{DIN_C}		4.8		6.0	ns
$t_{LEGLOBAL}$		3.1		3.9	ns
$t_{LABCARRY}$		0.4		0.5	ns
$t_{LABCASC}$		0.8		1.0	ns

Table 32. External Reference Timing Parameters for EPF6016 Devices

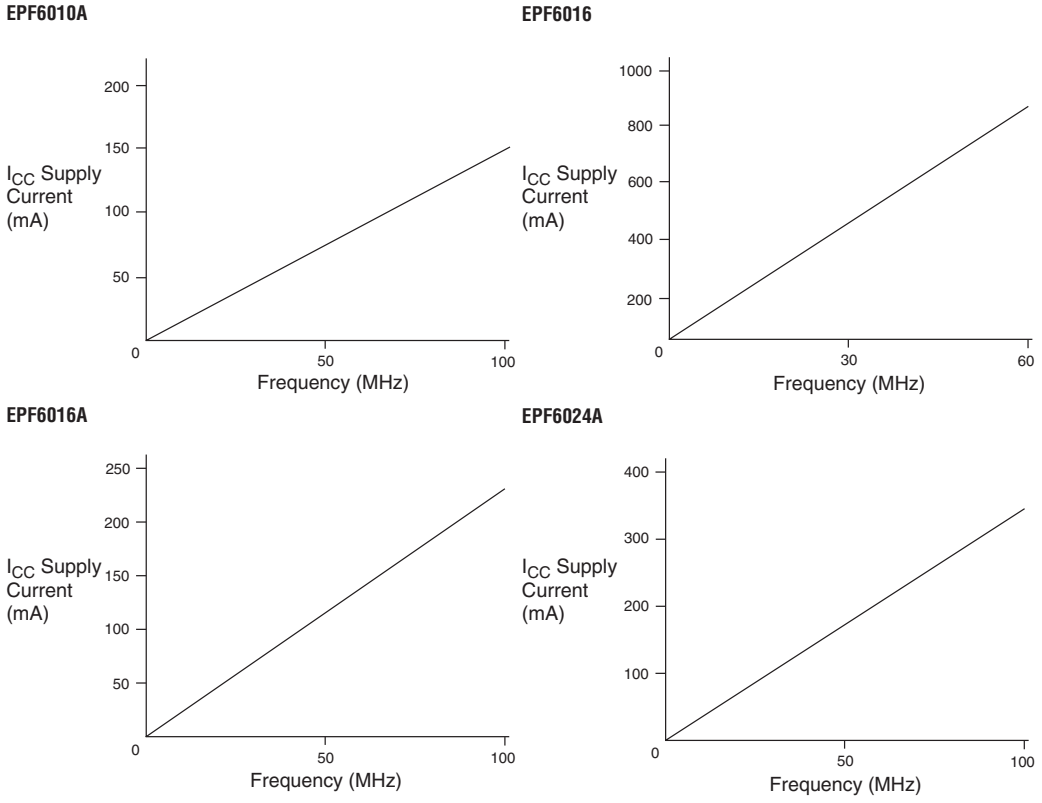
Parameter	Speed Grade				Unit
	-2		-3		
	Min	Max	Min	Max	
t_1		53.0		65.0	ns
t_{DRR}		16.0		20.0	ns

This calculation provides an I_{CC} estimate based on typical conditions with no output load. The actual I_{CC} should be verified during operation because this measurement is sensitive to the actual pattern in the device and the environmental operating conditions.

To better reflect actual designs, the power model (and the constant K in the power calculation equations shown above) for continuous interconnect FLEX devices assumes that LEs drive FastTrack Interconnect channels. In contrast, the power model of segmented FPGAs assumes that all LEs drive only one short interconnect segment. This assumption may lead to inaccurate results, compared to measured power consumption for an actual design in a segmented interconnect FPGA.

Figure 20 shows the relationship between the current and operating frequency for EPF6010A, EPF6016, EPF6016A, and EPF6024A devices.

Figure 20. $I_{CCACTIVE}$ vs. Operating Frequency



Device Configuration & Operation

The FLEX 6000 architecture supports several configuration schemes to load a design into the device(s) on the circuit board. This section summarizes the device operating modes and available device configuration schemes.

f See *Application Note 116 (Configuring APEX 20K, FLEX 10K & FLEX 6000 Devices)* for detailed information on configuring FLEX 6000 devices, including sample schematics, timing diagrams, configuration options, pins names, and timing parameters.