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Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of [Embedded - Microprocessors](#)

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details

Product Status	Obsolete
Core Processor	MPC8xx
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	66MHz
Co-Processors/DSP	Communications; CPM
RAM Controllers	DRAM
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10Mbps (1), 10/100Mbps (1)
SATA	-
USB	-
Voltage - I/O	3.3V
Operating Temperature	0°C ~ 105°C (TA)
Security Features	-
Package / Case	357-BBGA
Supplier Device Package	357-PBGA (25x25)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mpc857tzq66b

Features

- Sleep—All units disabled except RTC, PIT, time base, and decremter with PLL active for fast wake up
- Deep sleep—All units disabled including PLL except RTC, PIT, time base, and decremter.
- Power down mode— All units powered down except PLL, RTC, PIT, time base and decremter
- Debug interface
 - Eight comparators: four operate on instruction address, two operate on data address, and two operate on data
 - Supports conditions: = $\neq$ < >
 - Each watchpoint can generate a break point internally
- 3.3 V operation with 5-V TTL compatibility except EXTAL and EXTCLK
- 357-pin plastic ball grid array (PBGA) package
- Operation up to 100MHz

The MPC862/857T/857DSL is comprised of three modules that each use the 32-bit internal bus: the MPC8xx core, the system integration unit (SIU), and the communication processor module (CPM). The MPC862P/862T block diagram is shown in [Figure 1](#). The MPC857T/857DSL block diagram is shown in [Figure 2](#).

4 Thermal Characteristics

Table 3 shows the thermal characteristics for the MPC862/857T/857DSL.

Table 3. MPC862/857T/857DSL Thermal Resistance Data

Rating	Environment		Symbol	Value	Unit
Junction to ambient ¹	Natural Convection	Single layer board (1s)	$R_{\theta JA}$ ²	37	°C/W
		Four layer board (2s2p)	$R_{\theta JMA}$ ³	23	
	Air flow (200 ft/min)	Single layer board (1s)	$R_{\theta JMA}$ ³	30	
		Four layer board (2s2p)	$R_{\theta JMA}$ ³	19	
Junction to board ⁴			$R_{\theta JB}$	13	
Junction to case ⁵			$R_{\theta JC}$	6	
Junction to package top ⁶	Natural Convection		Ψ_{JT}	2	
	Air flow (200 ft/min)		Ψ_{JT}	2	

¹ Junction temperature is a function of on-chip power dissipation, package thermal resistance, mounting site (board) temperature, ambient temperature, air flow, power dissipation of other components on the board, and board thermal resistance.

² Per SEMI G38-87 and JEDEC JESD51-2 with the single layer board horizontal.

³ Per JEDEC JESD51-6 with the board horizontal.

⁴ Thermal resistance between the die and the printed circuit board per JEDEC JESD51-8. Board temperature is measured on the top surface of the board near the package.

⁵ Indicates the average thermal resistance between the die and the case top surface as measured by the cold plate method (MIL SPEC-883 Method 1012.1) with the cold plate temperature used for the case temperature. For exposed pad packages where the pad would be expected to be soldered, junction to case thermal resistance is a simulated value from the junction to the exposed pad without contact resistance.

⁶ Thermal characterization parameter indicating the temperature difference between package top and the junction temperature per JEDEC JESD51-2.

5 Power Dissipation

Table 4 provides power dissipation information. The modes are 1:1, where CPU and bus speeds are equal, and 2:1 mode, where CPU frequency is twice bus speed.

Table 4. Power Dissipation (P_D)

Die Revision	Frequency	Typical ¹	Maximum ²	Unit
0 (1:1 Mode)	50 MHz	656	735	mW
	66 MHz	TBD	TBD	mW
A.1, B.0 (1:1 Mode)	50 MHz	630	760	mW
	66 MHz	890	1000	mW

Table 5. DC Electrical Specifications (continued)

Characteristic	Symbol	Min	Max	Unit
Output Low Voltage IOL = 2.0 mA (CLKOUT) IOL = 3.2 mA ³ IOL = 5.3 mA ⁴ IOL = 7.0 mA (TXD1/PA14, TXD2/PA12) IOL = 8.9 mA ($\overline{TS}$, $\overline{TA}$, $\overline{TEA}$, $\overline{BI}$, $\overline{BB}$, $\overline{HRESET}$, $\overline{SRESET}$)	VOL	—	0.5	V

¹ $V_{IL}(\text{max})$ for the I²C interface is 0.8 V rather than the 1.5 V as specified in the I²C standard.

² Input capacitance is periodically sampled.

³ A(0:31), $\overline{TSI0}/\overline{REG}$, $\overline{TSI1}$, D(0:31), DP(0:3)/ $\overline{IRQ}$ (3:6), $\overline{RD}/\overline{WR}$, $\overline{BURST}$, $\overline{RSV}/\overline{IRQ2}$, $\overline{IP_B}(0:1)/\overline{IWP}(0:1)/\overline{VFLS}(0:1)$, $\overline{IP_B2}/\overline{IOIS16_B}/\overline{AT2}$, $\overline{IP_B3}/\overline{IWP2}/\overline{VF2}$, $\overline{IP_B4}/\overline{LWP0}/\overline{VF0}$, $\overline{IP_B5}/\overline{LWP1}/\overline{VF1}$, $\overline{IP_B6}/\overline{DSDI}/\overline{AT0}$, $\overline{IP_B7}/\overline{PTR}/\overline{AT3}$, $\overline{RXD1}/\overline{PA15}$, $\overline{RXD2}/\overline{PA13}$, $\overline{L1TXDB}/\overline{PA11}$, $\overline{L1RXDB}/\overline{PA10}$, $\overline{L1TXDA}/\overline{PA9}$, $\overline{L1RXDA}/\overline{PA8}$, $\overline{TIN1}/\overline{L1RCLKA}/\overline{BRGO1}/\overline{CLK1}/\overline{PA7}$, $\overline{BRGCLK1}/\overline{TOUT1}/\overline{CLK2}/\overline{PA6}$, $\overline{TIN2}/\overline{L1TCLKA}/\overline{BRGO2}/\overline{CLK3}/\overline{PA5}$, $\overline{TOUT2}/\overline{CLK4}/\overline{PA4}$, $\overline{TIN3}/\overline{BRGO3}/\overline{CLK5}/\overline{PA3}$, $\overline{BRGCLK2}/\overline{L1RCLKB}/\overline{TOUT3}/\overline{CLK6}/\overline{PA2}$, $\overline{TIN4}/\overline{BRGO4}/\overline{CLK7}/\overline{PA1}$, $\overline{L1TCLKB}/\overline{TOUT4}/\overline{CLK8}/\overline{PA0}$, $\overline{REJECT1}/\overline{SPISEL}/\overline{PB31}$, $\overline{SPICLK}/\overline{PB30}$, $\overline{SPIMOSI}/\overline{PB29}$, $\overline{BRGO4}/\overline{SPIMISO}/\overline{PB28}$, $\overline{BRGO1}/\overline{I2CSDA}/\overline{PB27}$, $\overline{BRGO2}/\overline{I2CSCL}/\overline{PB26}$, $\overline{SMTXD1}/\overline{PB25}$, $\overline{SMRXD1}/\overline{PB24}$, $\overline{SMSYN1}/\overline{SDACK1}/\overline{PB23}$, $\overline{SMSYN2}/\overline{SDACK2}/\overline{PB22}$, $\overline{SMTXD2}/\overline{L1CLKOB}/\overline{PB21}$, $\overline{SMRXD2}/\overline{L1CLKOA}/\overline{PB20}$, $\overline{L1ST1}/\overline{RTS1}/\overline{PB19}$, $\overline{L1ST2}/\overline{RTS2}/\overline{PB18}$, $\overline{L1ST3}/\overline{L1RQB}/\overline{PB17}$, $\overline{L1ST4}/\overline{L1RQA}/\overline{PB16}$, $\overline{BRGO3}/\overline{PB15}$, $\overline{RSTR1}/\overline{PB14}$, $\overline{L1ST1}/\overline{RTS1}/\overline{DREQ0}/\overline{PC15}$, $\overline{L1ST2}/\overline{RTS2}/\overline{DREQ1}/\overline{PC14}$, $\overline{L1ST3}/\overline{L1RQB}/\overline{PC13}$, $\overline{L1ST4}/\overline{L1RQA}/\overline{PC12}$, $\overline{CTS1}/\overline{PC11}$, $\overline{TGATE1}/\overline{CD1}/\overline{PC10}$, $\overline{CTS2}/\overline{PC9}$, $\overline{TGATE2}/\overline{CD2}/\overline{PC8}$, $\overline{CTS3}/\overline{SDACK2}/\overline{L1TSYNCB}/\overline{PC7}$, $\overline{CD3}/\overline{L1RSYNCB}/\overline{PC6}$, $\overline{CTS4}/\overline{SDACK1}/\overline{L1TSYNCA}/\overline{PC5}$, $\overline{CD4}/\overline{L1RSYNCA}/\overline{PC4}$, $\overline{PD15}/\overline{L1TSYNCA}$, $\overline{PD14}/\overline{L1RSYNCA}$, $\overline{PD13}/\overline{L1TSYNCB}$, $\overline{PD12}/\overline{L1RSYNCB}$, $\overline{PD11}/\overline{RXD3}$, $\overline{PD10}/\overline{TXD3}$, $\overline{PD9}/\overline{RXD4}$, $\overline{PD8}/\overline{TXD4}$, $\overline{PD5}/\overline{REJECT2}$, $\overline{PD6}/\overline{RTS4}$, $\overline{PD7}/\overline{RTS3}$, $\overline{PD4}/\overline{REJECT3}$, $\overline{PD3}$, $\overline{MII_MDC}$, $\overline{MII_TX_ER}$, $\overline{MII_EN}$, $\overline{MII_MDIO}$, $\overline{MII_TXD}[0:3]$.

⁴ $\overline{BDIP}/\overline{GPL_B}(5)$, $\overline{BR}$, $\overline{BG}$, $\overline{FRZ}/\overline{IRQ6}$, $\overline{CS}(0:5)$, $\overline{CS}(6)/\overline{CE}(1)_B$, $\overline{CS}(7)/\overline{CE}(2)_B$, $\overline{WE0}/\overline{BS_B0}/\overline{IORD}$, $\overline{WE1}/\overline{BS_B1}/\overline{IOWR}$, $\overline{WE2}/\overline{BS_B2}/\overline{PCOE}$, $\overline{WE3}/\overline{BS_B3}/\overline{PCWE}$, $\overline{BS_A}(0:3)$, $\overline{GPL_A0}/\overline{GPL_B0}$, $\overline{OE}/\overline{GPL_A1}/\overline{GPL_B1}$, $\overline{GPL_A}(2:3)/\overline{GPL_B}(2:3)/\overline{CS}(2:3)$, $\overline{UPWAITA}/\overline{GPL_A4}$, $\overline{UPWAITB}/\overline{GPL_B4}$, $\overline{GPL_A5}$, $\overline{ALE_A}$, $\overline{CE1_A}$, $\overline{CE2_A}$, $\overline{ALE_B}/\overline{DSCK}/\overline{AT1}$, $\overline{OP}(0:1)$, $\overline{OP2}/\overline{MODCK1}/\overline{STS}$, $\overline{OP3}/\overline{MODCK2}/\overline{DSDO}$, $\overline{BADDR}(28:30)$.

7 Thermal Calculation and Measurement

For the following discussions, $P_D = (V_{DD} \times I_{DD}) + P_{I/O}$, where $P_{I/O}$ is the power dissipation of the I/O drivers.

7.1 Estimation with Junction-to-Ambient Thermal Resistance

An estimation of the chip junction temperature, T_J , in °C can be obtained from the equation:

$$T_J = T_A + (R_{\theta JA} \times P_D)$$

where:

T_A = ambient temperature (°C)

$R_{\theta JA}$ = package junction-to-ambient thermal resistance (°C/W)

P_D = power dissipation in package

The junction-to-ambient thermal resistance is an industry standard value which provides a quick and easy estimation of thermal performance. However, the answer is only an estimate; test cases have demonstrated that errors of a factor of two (in the quantity $T_J - T_A$) are possible.

Table 7. Bus Operation Timings (continued)

Num	Characteristic	33 MHz		40 MHz		50 MHz		66 MHz		Unit
		Min	Max	Min	Max	Min	Max	Min	Max	
B17a	CLKOUT to $\overline{KR}$, $\overline{RETRY}$, $\overline{CR}$ valid (hold time) (MIN = $0.00 \times B1 + 2.00$)	2.00	—	2.00	—	2.00	—	2.00	—	ns
B18	D(0:31), DP(0:3) valid to CLKOUT rising edge (setup time) ⁸ (MIN = $0.00 \times B1 + 6.00$)	6.00	—	6.00	—	6.00	—	6.00	—	ns
B19	CLKOUT rising edge to D(0:31), DP(0:3) valid (hold time) ⁸ (MIN = $0.00 \times B1 + 1.00$ ⁹)	1.00	—	1.00	—	1.00	—	2.00	—	ns
B20	D(0:31), DP(0:3) valid to CLKOUT falling edge (setup time) ¹⁰ (MIN = $0.00 \times B1 + 4.00$)	4.00	—	4.00	—	4.00	—	4.00	—	ns
B21	CLKOUT falling edge to D(0:31), DP(0:3) valid (hold Time) ¹⁰ (MIN = $0.00 \times B1 + 2.00$)	2.00	—	2.00	—	2.00	—	2.00	—	ns
B22	CLKOUT rising edge to $\overline{CS}$ asserted GPCM ACS = 00 (MAX = $0.25 \times B1 + 6.3$)	7.60	13.80	6.30	12.50	5.00	11.30	3.80	10.00	ns
B22a	CLKOUT falling edge to $\overline{CS}$ asserted GPCM ACS = 10, TRLX = 0 (MAX = $0.00 \times B1 + 8.00$)	—	8.00	—	8.00	—	8.00	—	8.00	ns
B22b	CLKOUT falling edge to $\overline{CS}$ asserted GPCM ACS = 11, TRLX = 0, EBDF = 0 (MAX = $0.25 \times B1 + 6.3$)	7.60	13.80	6.30	12.50	5.00	11.30	3.80	10.00	ns
B22c	CLKOUT falling edge to $\overline{CS}$ asserted GPCM ACS = 11, TRLX = 0, EBDF = 1 (MAX = $0.375 \times B1 + 6.6$)	10.90	18.00	10.90	18.00	7.00	14.30	5.20	12.30	ns
B23	CLKOUT rising edge to $\overline{CS}$ negated GPCM read access, GPCM write access ACS = 00, TRLX = 0 & CSNT = 0 (MAX = $0.00 \times B1 + 8.00$)	2.00	8.00	2.00	8.00	2.00	8.00	2.00	8.00	ns
B24	A(0:31) and BADDR(28:30) to $\overline{CS}$ asserted GPCM ACS = 10, TRLX = 0 (MIN = $0.25 \times B1 - 2.00$)	5.60	—	4.30	—	3.00	—	1.80	—	ns
B24a	A(0:31) and BADDR(28:30) to $\overline{CS}$ asserted GPCM ACS = 11 TRLX = 0 (MIN = $0.50 \times B1 - 2.00$)	13.20	—	10.50	—	8.00	—	5.60	—	ns
B25	CLKOUT rising edge to $\overline{OE}$, $\overline{WE}$ (0:3) asserted (MAX = $0.00 \times B1 + 9.00$)	—	9.00		9.00		9.00		9.00	ns
B26	CLKOUT rising edge to $\overline{OE}$ negated (MAX = $0.00 \times B1 + 9.00$)	2.00	9.00	2.00	9.00	2.00	9.00	2.00	9.00	ns

Table 7. Bus Operation Timings (continued)

Num	Characteristic	33 MHz		40 MHz		50 MHz		66 MHz		Unit
		Min	Max	Min	Max	Min	Max	Min	Max	
B37	UPWAIT valid to CLKOUT falling edge ¹² (MIN = 0.00 x B1 + 6.00)	6.00	—	6.00	—	6.00	—	6.00	—	ns
B38	CLKOUT falling edge to UPGATE valid ¹² (MIN = 0.00 x B1 + 1.00)	1.00	—	1.00	—	1.00	—	1.00	—	ns
B39	$\overline{AS}$ valid to CLKOUT rising edge ¹³ (MIN = 0.00 x B1 + 7.00)	7.00	—	7.00	—	7.00	—	7.00	—	ns
B40	A(0:31), TSIZ(0:1), RD/WR, BURST, valid to CLKOUT rising edge (MIN = 0.00 x B1 + 7.00)	7.00	—	7.00	—	7.00	—	7.00	—	ns
B41	$\overline{TS}$ valid to CLKOUT rising edge (setup time) (MIN = 0.00 x B1 + 7.00)	7.00	—	7.00	—	7.00	—	7.00	—	ns
B42	CLKOUT rising edge to $\overline{TS}$ valid (hold time) (MIN = 0.00 x B1 + 2.00)	2.00	—	2.00	—	2.00	—	2.00	—	ns
B43	$\overline{AS}$ negation to memory controller signals negation (MAX = TBD)	—	TBD	—	TBD	—	TBD	—	TBD	ns

¹ Phase and frequency jitter performance results are only valid if the input jitter is less than the prescribed value.

² If the rate of change of the frequency of EXTAL is slow (i.e. it does not jump between the minimum and maximum values in one cycle) or the frequency of the jitter is fast (i.e., it does not stay at an extreme value for a long time) then the maximum allowed jitter on EXTAL can be up to 2%.

³ The timings specified in B4 and B5 are based on full strength clock.

⁴ The timing for $\overline{BR}$ output is relevant when the MPC862/857T/857DSL is selected to work with external bus arbiter. The timing for $\overline{BG}$ output is relevant when the MPC862/857T/857DSL is selected to work with internal bus arbiter.

⁵ For part speeds above 50MHz, use 9.80ns for B11a.

⁶ The timing required for $\overline{BR}$ input is relevant when the MPC862/857T/857DSL is selected to work with internal bus arbiter. The timing for $\overline{BG}$ input is relevant when the MPC862/857T/857DSL is selected to work with external bus arbiter.

⁷ For part speeds above 50MHz, use 2ns for B17.

⁸ The D(0:31) and DP(0:3) input timings B18 and B19 refer to the rising edge of the CLKOUT in which the $\overline{TA}$ input signal is asserted.

⁹ For part speeds above 50MHz, use 2ns for B19.

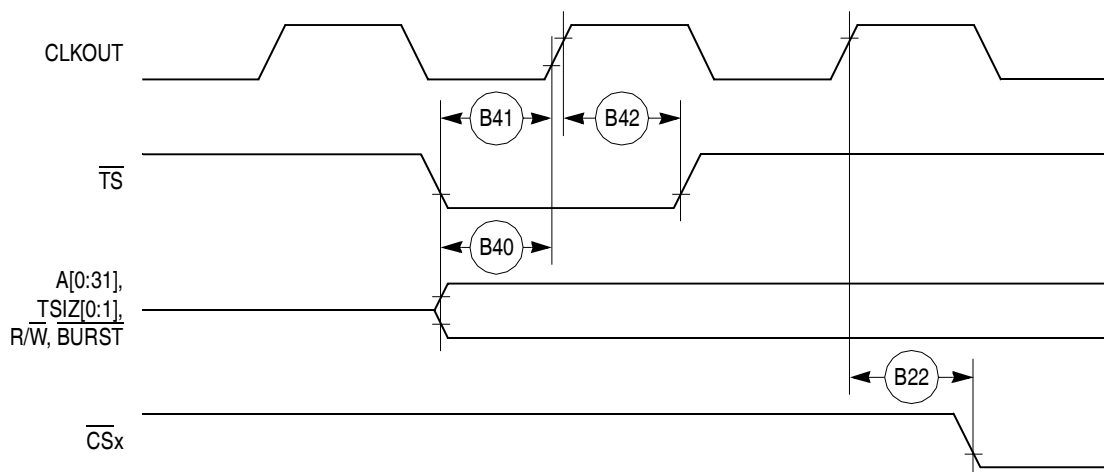
¹⁰ The D(0:31) and DP(0:3) input timings B20 and B21 refer to the falling edge of the CLKOUT. This timing is valid only for read accesses controlled by chip-selects under control of the UPM in the memory controller, for data beats where DLT3 = 1 in the UPM RAM words. (This is only the case where data is latched on the falling edge of CLKOUT.)

¹¹ The timing B30 refers to $\overline{CS}$ when ACS = 00 and to $\overline{WE}(0:3)$ when CSNT = 0.

¹² The signal UPGATE is considered asynchronous to the CLKOUT and synchronized internally. The timings specified in B37 and B38 are specified to enable the freeze of the UPM output signals as described in Figure 19.

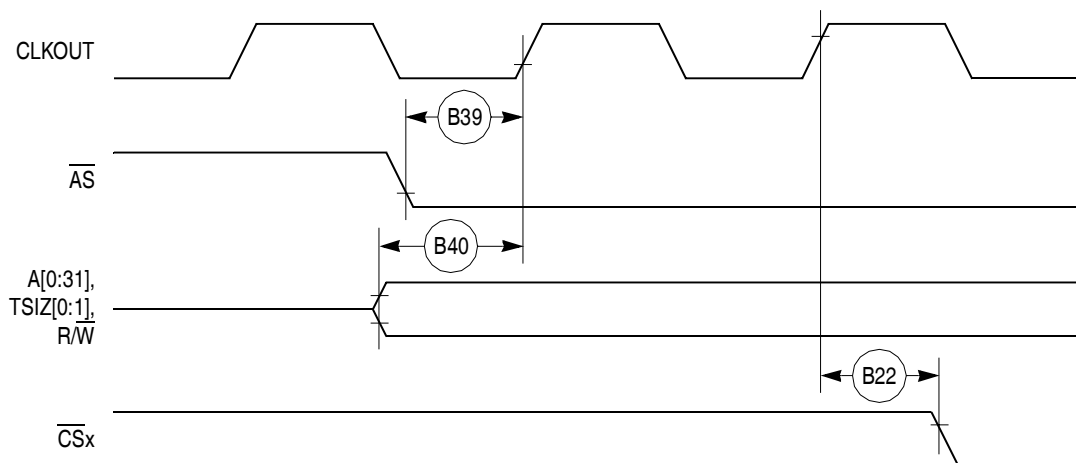
¹³ The $\overline{AS}$ signal is considered asynchronous to the CLKOUT. The timing B39 is specified in order to allow the behavior specified in Figure 22.

Figure 21 provides the timing for the synchronous external master access controlled by the GPCM.



**Figure 21. Synchronous External Master Access Timing
(GPCM Handled ACS = 00)**

Figure 22 provides the timing for the asynchronous external master memory access controlled by the GPCM.



**Figure 22. Asynchronous External Master Memory Access Timing
(GPCM Controlled—ACS = 00)**

Figure 23 provides the timing for the asynchronous external master control signals negation.

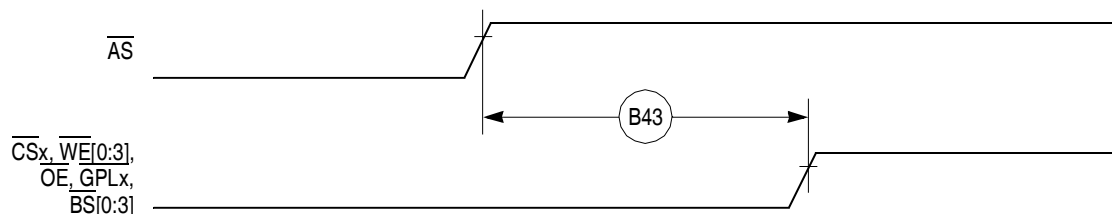


Figure 23. Asynchronous External Master—Control Signals Negation Timing

Table 8 provides interrupt timing for the MPC862/857T/857DSL.

Table 8. Interrupt Timing

Num	Characteristic ¹	All Frequencies		Unit
		Min	Max	
I39	$\overline{\text{IRQ}}_x$ valid to CLKOUT rising edge (set up time)	6.00		ns
I40	$\overline{\text{IRQ}}_x$ hold time after CLKOUT	2.00		ns
I41	$\overline{\text{IRQ}}_x$ pulse width low	3.00		ns
I42	$\overline{\text{IRQ}}_x$ pulse width high	3.00		ns
I43	$\overline{\text{IRQ}}_x$ edge-to-edge time	$4 \times T_{\text{CLKOUT}}$		—

¹ The timings I39 and I40 describe the testing conditions under which the $\overline{\text{IRQ}}$ lines are tested when being defined as level sensitive. The $\overline{\text{IRQ}}$ lines are synchronized internally and do not have to be asserted or negated with reference to the CLKOUT.

The timings I41, I42, and I43 are specified to allow the correct function of the $\overline{\text{IRQ}}$ lines detection circuitry, and has no direct relation with the total system interrupt latency that the MPC862/857T/857DSL is able to support.

Figure 24 provides the interrupt detection timing for the external level-sensitive lines.

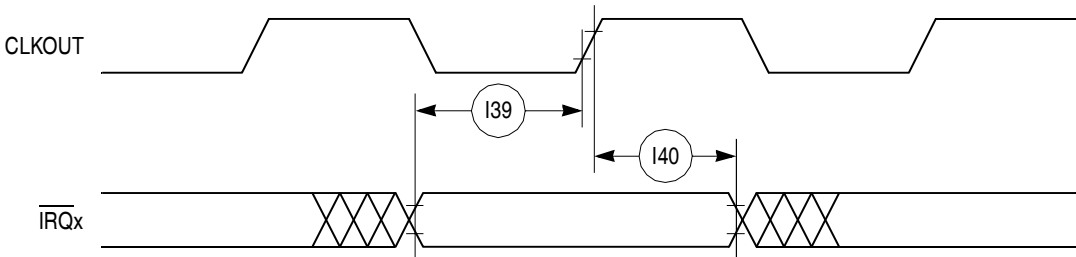


Figure 24. Interrupt Detection Timing for External Level Sensitive Lines

Figure 25 provides the interrupt detection timing for the external edge-sensitive lines.

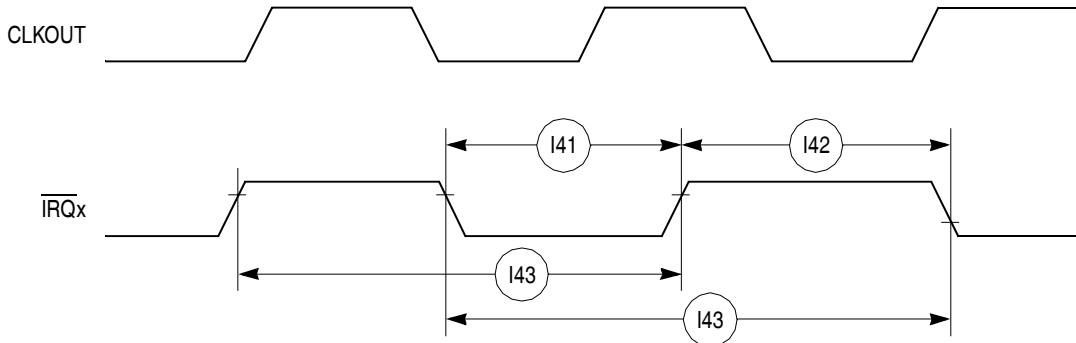


Figure 25. Interrupt Detection Timing for External Edge Sensitive Lines

Table 12 shows the reset timing for the MPC862/857T/857DSL.

Table 12. Reset Timing

Num	Characteristic	33 MHz		40 MHz		50 MHz		66 MHz		Unit
		Min	Max	Min	Max	Min	Max	Min	Max	
R69	CLKOUT to $\overline{\text{HRESET}}$ high impedance (MAX = 0.00 x B1 + 20.00)	—	20.00	—	20.00	—	20.00	—	20.00	ns
R70	CLKOUT to $\overline{\text{SRESET}}$ high impedance (MAX = 0.00 x B1 + 20.00)	—	20.00	—	20.00	—	20.00	—	20.00	ns
R71	$\overline{\text{RSTCONF}}$ pulse width (MIN = 17.00 x B1)	515.20	—	425.00	—	340.00	—	257.60	—	ns
R72	—	—	—	—	—	—	—	—	—	—
R73	Configuration data to $\overline{\text{HRESET}}$ rising edge set up time (MIN = 15.00 x B1 + 50.00)	504.50	—	425.00	—	350.00	—	277.30	—	ns
R74	Configuration data to $\overline{\text{RSTCONF}}$ rising edge set up time (MIN = 0.00 x B1 + 350.00)	350.00	—	350.00	—	350.00	—	350.00	—	ns
R75	Configuration data hold time after $\overline{\text{RSTCONF}}$ negation (MIN = 0.00 x B1 + 0.00)	0.00	—	0.00	—	0.00	—	0.00	—	ns
R76	Configuration data hold time after $\overline{\text{HRESET}}$ negation (MIN = 0.00 x B1 + 0.00)	0.00	—	0.00	—	0.00	—	0.00	—	ns
R77	$\overline{\text{HRESET}}$ and $\overline{\text{RSTCONF}}$ asserted to data out drive (MAX = 0.00 x B1 + 25.00)	—	25.00	—	25.00	—	25.00	—	25.00	ns
R78	$\overline{\text{RSTCONF}}$ negated to data out high impedance. (MAX = 0.00 x B1 + 25.00)	—	25.00	—	25.00	—	25.00	—	25.00	ns
R79	CLKOUT of last rising edge before chip three-states $\overline{\text{HRESET}}$ to data out high impedance. (MAX = 0.00 x B1 + 25.00)	—	25.00	—	25.00	—	25.00	—	25.00	ns
R80	DSDI, DSCK set up (MIN = 3.00 x B1)	90.90	—	75.00	—	60.00	—	45.50	—	ns
R81	DSDI, DSCK hold time (MIN = 0.00 x B1 + 0.00)	0.00	—	0.00	—	0.00	—	0.00	—	ns
R82	$\overline{\text{SRESET}}$ negated to CLKOUT rising edge for DSDI and DSCK sample (MIN = 8.00 x B1)	242.40	—	200.00	—	160.00	—	121.20	—	ns

Figure 35 provides the reset timing for the debug port configuration.

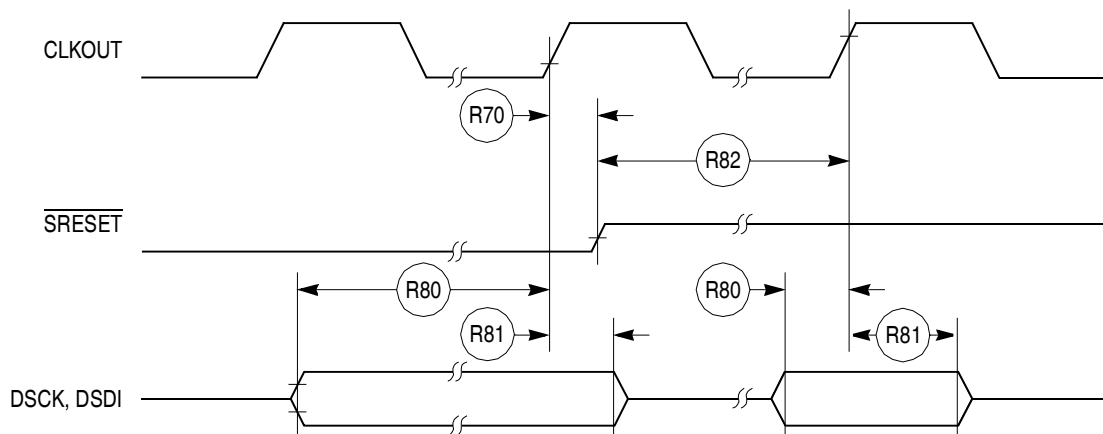


Figure 35. Reset Timing—Debug Port Configuration

10 IEEE 1149.1 Electrical Specifications

Table 13 provides the JTAG timings for the MPC862/857T/857DSL shown in Figure 36 though Figure 39.

Table 13. JTAG Timing

Num	Characteristic	All Frequencies		Unit
		Min	Max	
J82	TCK cycle time	100.00	—	ns
J83	TCK clock pulse width measured at 1.5 V	40.00	—	ns
J84	TCK rise and fall times	0.00	10.00	ns
J85	TMS, TDI data setup time	5.00	—	ns
J86	TMS, TDI data hold time	25.00	—	ns
J87	TCK low to TDO data valid	—	27.00	ns
J88	TCK low to TDO data invalid	0.00	—	ns
J89	TCK low to TDO high impedance	—	20.00	ns
J90	$\overline{\text{TRST}}$ assert time	100.00	—	ns
J91	$\overline{\text{TRST}}$ setup time to TCK low	40.00	—	ns
J92	TCK falling edge to output valid	—	50.00	ns
J93	TCK falling edge to output valid out of high impedance	—	50.00	ns
J94	TCK falling edge to output high impedance	—	50.00	ns
J95	Boundary scan input valid to TCK rising edge	50.00	—	ns
J96	TCK rising edge to boundary scan input invalid	50.00	—	ns

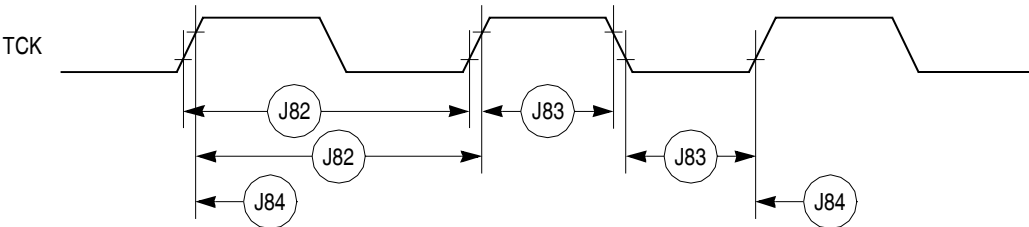


Figure 36. JTAG Test Clock Input Timing

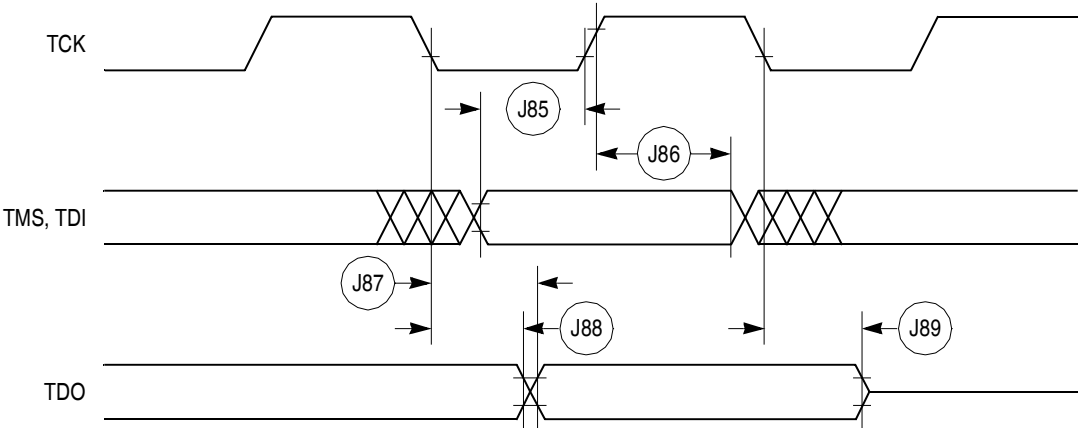


Figure 37. JTAG Test Access Port Timing Diagram

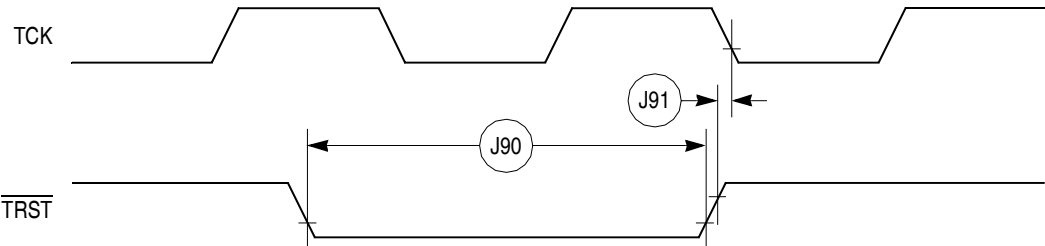


Figure 38. JTAG $\overline{\text{TRST}}$ Timing Diagram

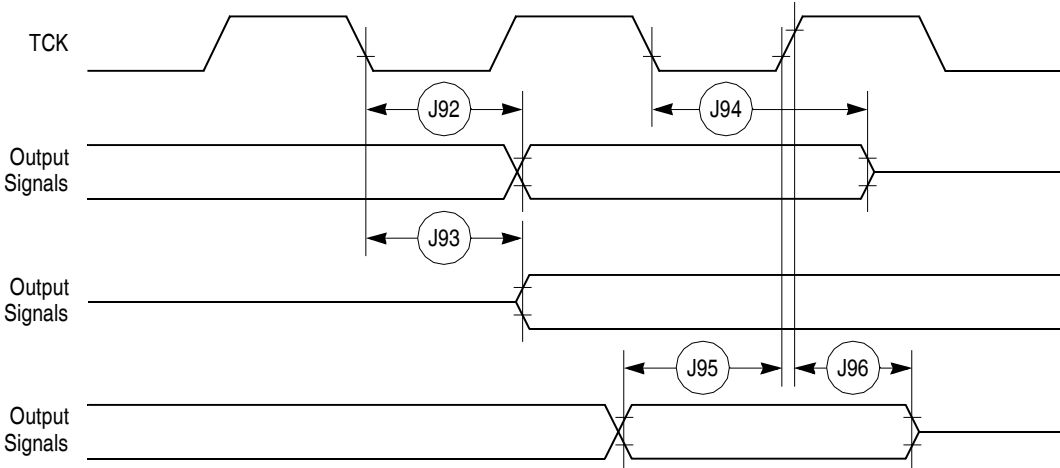


Figure 39. Boundary Scan (JTAG) Timing Diagram

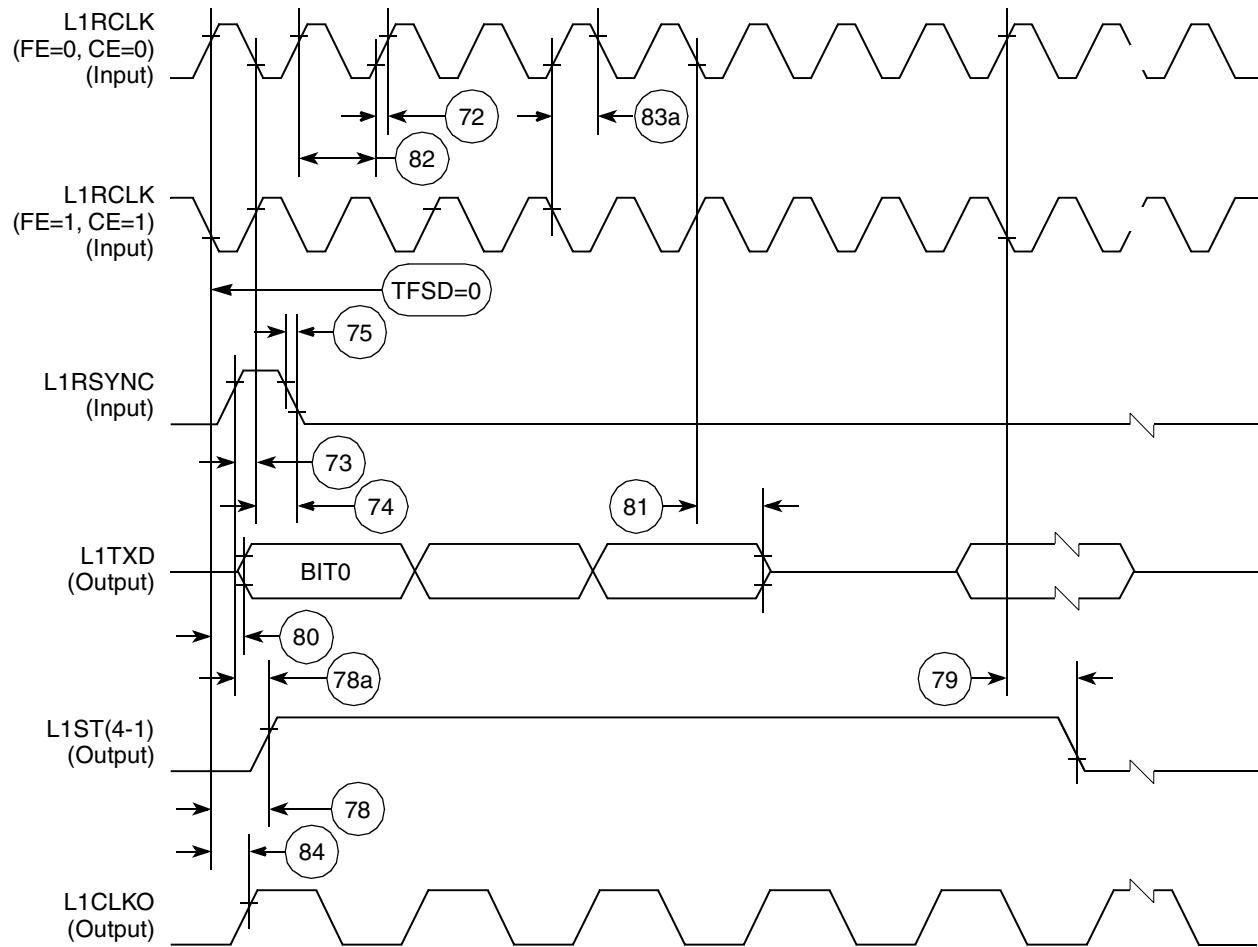


Figure 55. SI Transmit Timing with Double Speed Clocking (DSC = 1)

Figure 57 through Figure 59 show the NMSI timings.

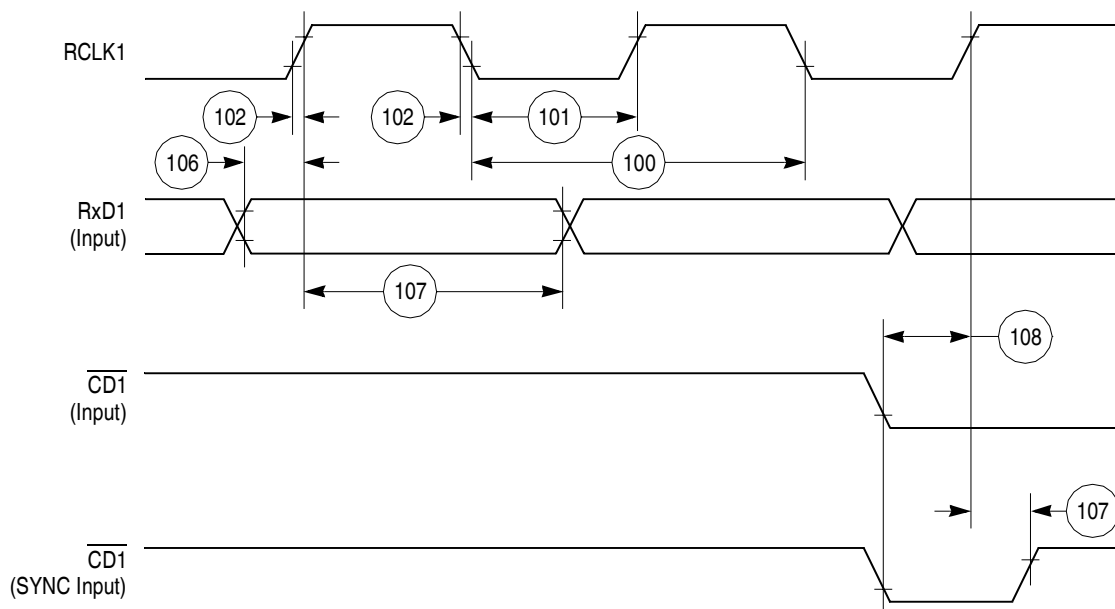


Figure 57. SCC NMSI Receive Timing Diagram

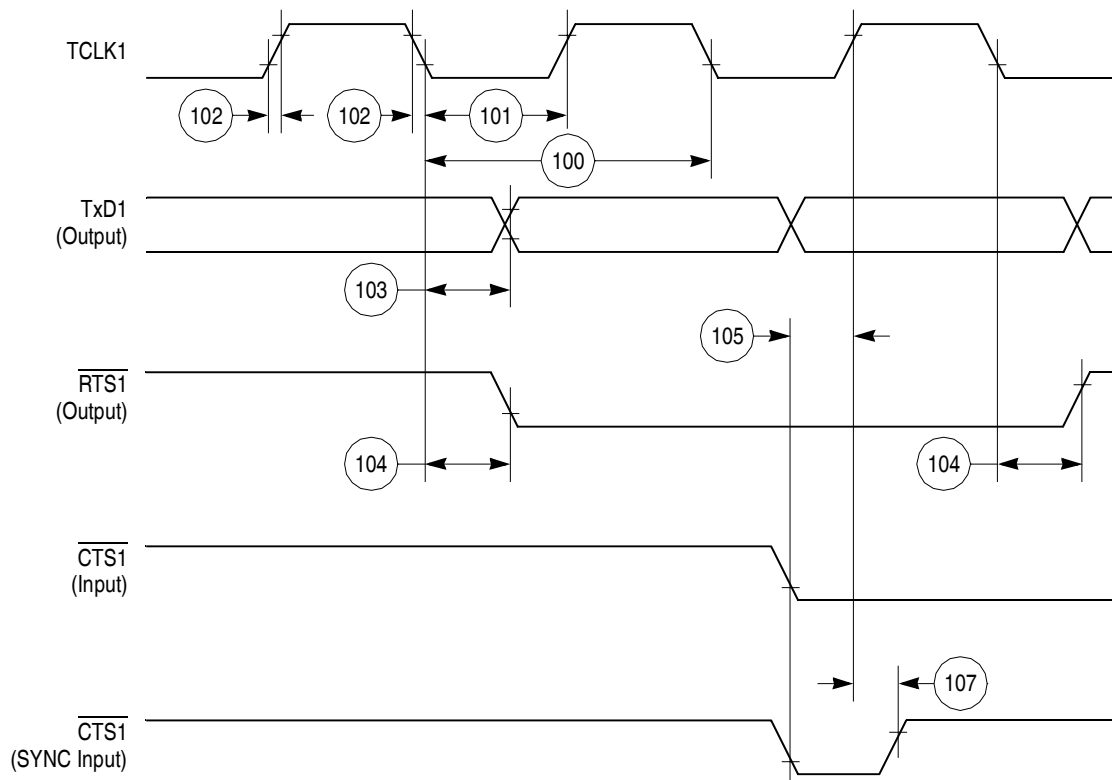


Figure 58. SCC NMSI Transmit Timing Diagram

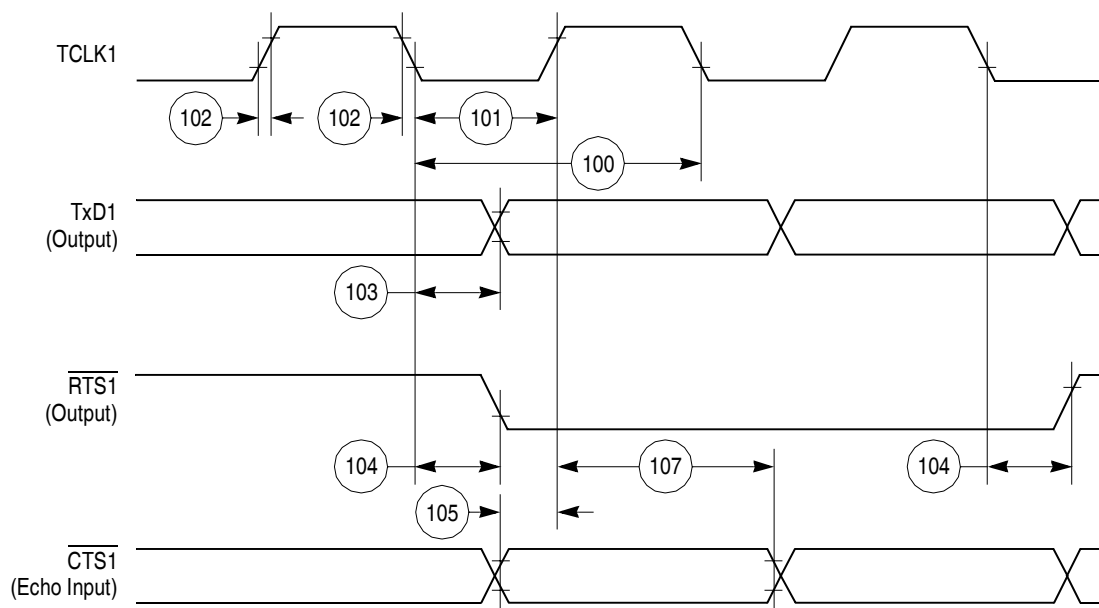


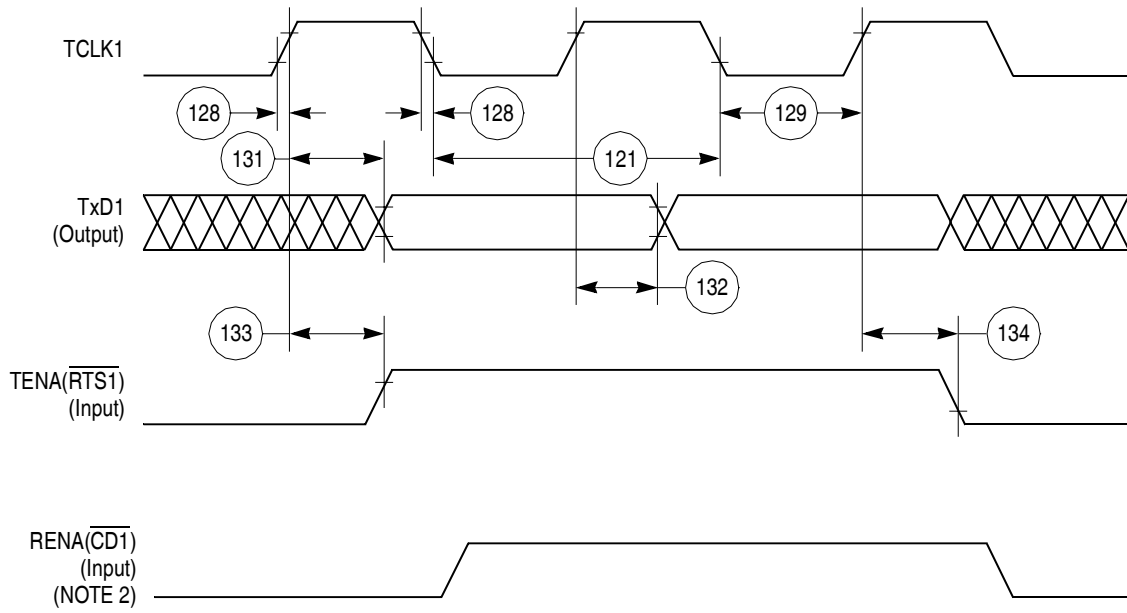
Figure 59. HDLC Bus Timing Diagram

11.8 Ethernet Electrical Specifications

Table 22 provides the Ethernet timings as shown in Figure 60 through Figure 64.

Table 22. Ethernet Timing

Num	Characteristic	All Frequencies		Unit
		Min	Max	
120	CLSN width high	40	—	ns
121	RCLK1 rise/fall time	—	15	ns
122	RCLK1 width low	40	—	ns
123	RCLK1 clock period ¹	80	120	ns
124	RXD1 setup time	20	—	ns
125	RXD1 hold time	5	—	ns
126	RENA active delay (from RCLK1 rising edge of the last data bit)	10	—	ns
127	RENA width low	100	—	ns
128	TCLK1 rise/fall time	—	15	ns
129	TCLK1 width low	40	—	ns
130	TCLK1 clock period ¹	99	101	ns
131	TXD1 active delay (from TCLK1 rising edge)	10	50	ns
132	TXD1 inactive delay (from TCLK1 rising edge)	10	50	ns
133	TENA active delay (from TCLK1 rising edge)	10	50	ns



- NOTES:
1. Transmit clock invert (TCI) bit in GSMR is set.
 2. If RENA is deasserted before TENA, or RENA is not asserted at all during transmit, then the CSL bit is set in the buffer descriptor at the end of the frame transmission.

Figure 62. Ethernet Transmit Timing Diagram

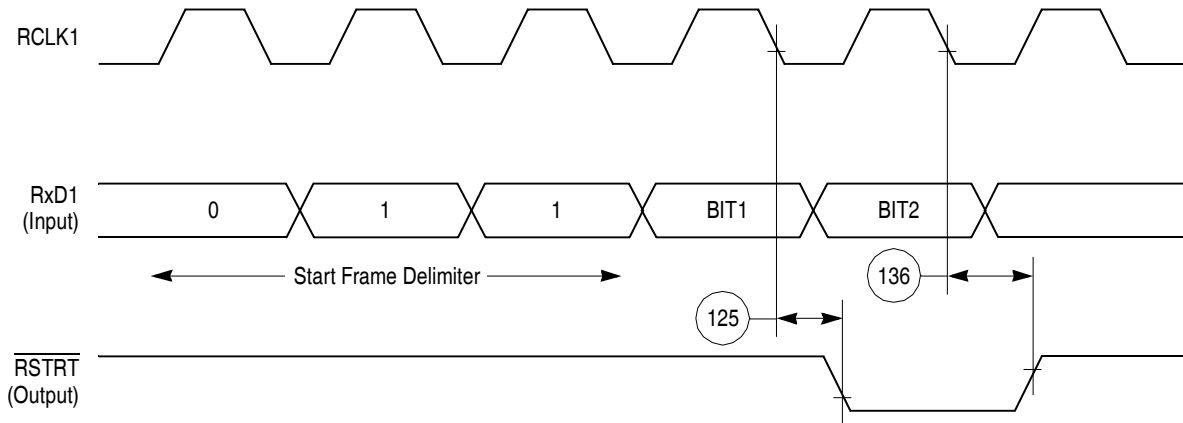


Figure 63. CAM Interface Receive Start Timing Diagram

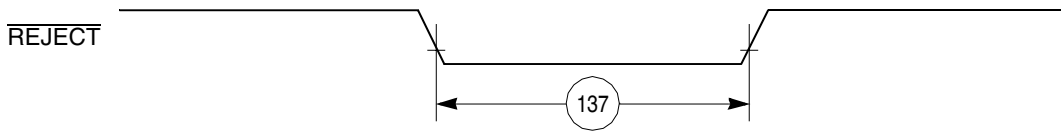


Figure 64. CAM Interface $\overline{\text{REJECT}}$ Timing Diagram

11.10 SPI Master AC Electrical Specifications

Table 24 provides the SPI master timings as shown in Figure 66 though Figure 67.

Table 24. SPI Master Timing

Num	Characteristic	All Frequencies		Unit
		Min	Max	
160	MASTER cycle time	4	1024	t_{cyc}
161	MASTER clock (SCK) high or low time	2	512	t_{cyc}
162	MASTER data setup time (inputs)	15	—	ns
163	Master data hold time (inputs)	0	—	ns
164	Master data valid (after SCK edge)	—	10	ns
165	Master data hold time (outputs)	0	—	ns
166	Rise time output	—	15	ns
167	Fall time output	—	15	ns

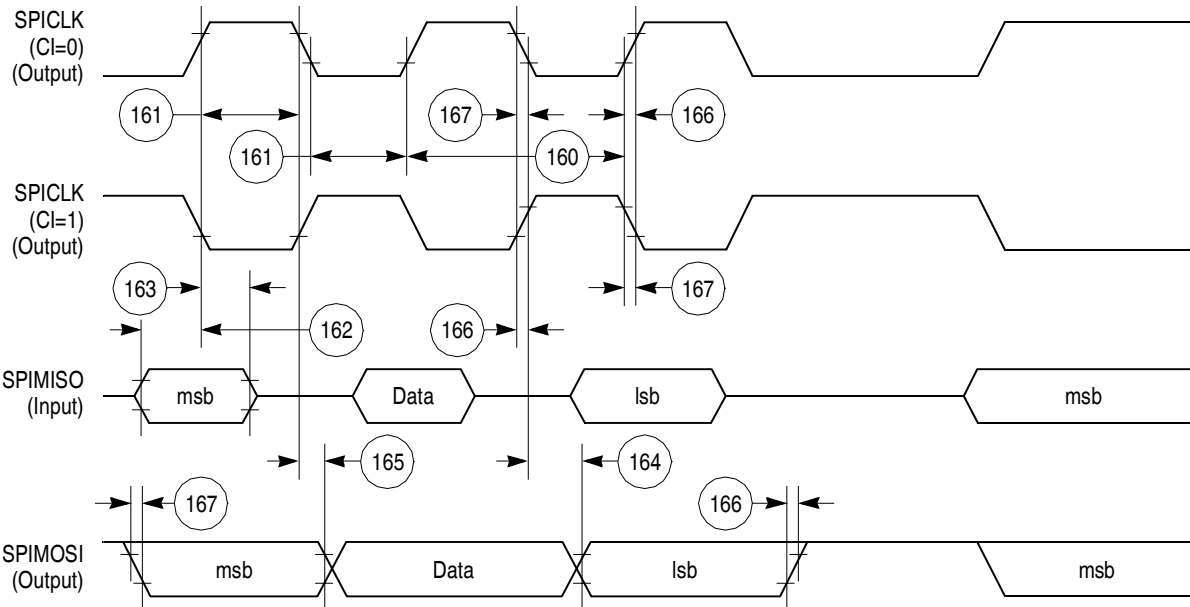


Figure 66. SPI Master (CP = 0) Timing Diagram

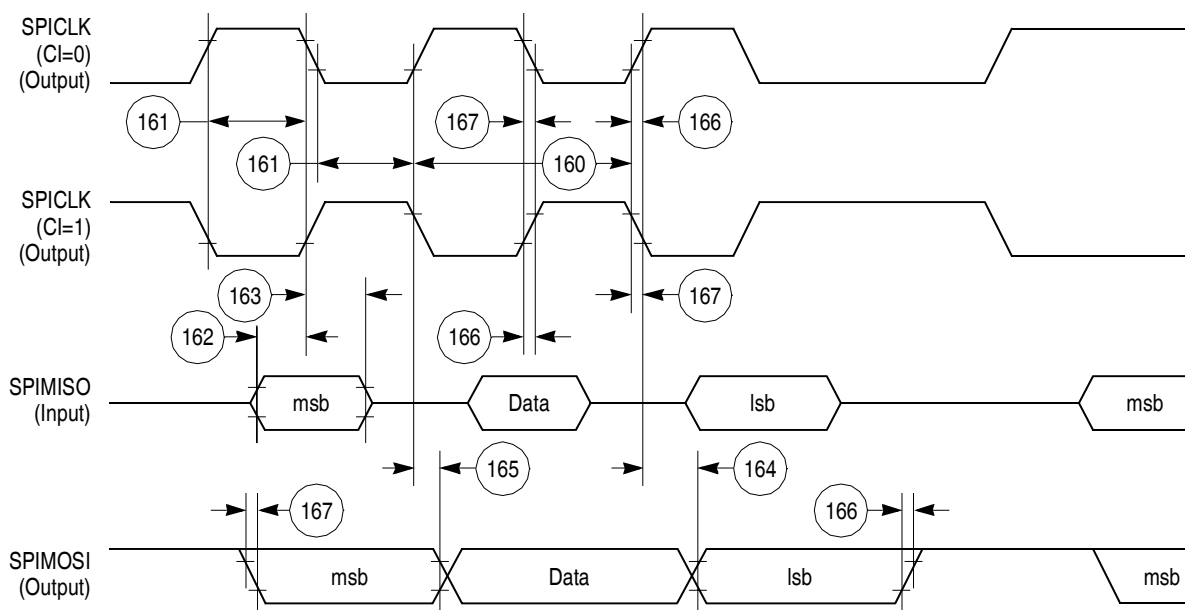


Figure 67. SPI Master (CP = 1) Timing Diagram

11.11 SPI Slave AC Electrical Specifications

Table 25 provides the SPI slave timings as shown in Figure 68 through Figure 69.

Table 25. SPI Slave Timing

Num	Characteristic	All Frequencies		Unit
		Min	Max	
170	Slave cycle time	2	—	t_{cyc}
171	Slave enable lead time	15	—	ns
172	Slave enable lag time	15	—	ns
173	Slave clock (SPICLK) high or low time	1	—	t_{cyc}
174	Slave sequential transfer delay (does not require deselect)	1	—	t_{cyc}
175	Slave data setup time (inputs)	20	—	ns
176	Slave data hold time (inputs)	20	—	ns
177	Slave access time	—	50	ns

11.12 I²C AC Electrical Specifications

Table 26 provides the I²C (SCL < 100 KHz) timings.

Table 26. I²C Timing (SCL < 100 KHz)

Num	Characteristic	All Frequencies		Unit
		Min	Max	
200	SCL clock frequency (slave)	0	100	kHz
200	SCL clock frequency (master) ¹	1.5	100	kHz
202	Bus free time between transmissions	4.7	—	μs
203	Low period of SCL	4.7	—	μs
204	High period of SCL	4.0	—	μs
205	Start condition setup time	4.7	—	μs
206	Start condition hold time	4.0	—	μs
207	Data hold time	0	—	μs
208	Data setup time	250	—	ns
209	SDL/SCL rise time	—	1	μs
210	SDL/SCL fall time	—	300	ns
211	Stop condition setup time	4.7	—	μs

¹ SCL frequency is given by $SCL = BRGCLK_frequency / ((BRG\ register + 3) * pre_scaler * 2)$.
The ratio $SyncClk/(BRGCLK/pre_scaler)$ must be greater or equal to 4/1.

Table 27 provides the I²C (SCL > 100 kHz) timings.

Table 27. I²C Timing (SCL > 100 kHz)

Num	Characteristic	Expression	All Frequencies		Unit
			Min	Max	
200	SCL clock frequency (slave)	fSCL	0	BRGCLK/48	Hz
200	SCL clock frequency (master) ¹	fSCL	BRGCLK/16512	BRGCLK/48	Hz
202	Bus free time between transmissions	—	$1/(2.2 * fSCL)$	—	s
203	Low period of SCL	—	$1/(2.2 * fSCL)$	—	s
204	High period of SCL	—	$1/(2.2 * fSCL)$	—	s
205	Start condition setup time	—	$1/(2.2 * fSCL)$	—	s
206	Start condition hold time	—	$1/(2.2 * fSCL)$	—	s
207	Data hold time	—	0	—	s
208	Data setup time	—	$1/(40 * fSCL)$	—	s
209	SDL/SCL rise time	—	—	$1/(10 * fSCL)$	s
210	SDL/SCL fall time	—	—	$1/(33 * fSCL)$	s
211	Stop condition setup time	—	$1/2(2.2 * fSCL)$	—	s

¹ SCL frequency is given by $SCL = BrgClk_frequency / ((BRG\ register + 3) * pre_scaler * 2)$.
The ratio $SyncClk/(Brg_Clk/pre_scaler)$ must be greater or equal to 4/1.

Table 35. Pin Assignments (continued)

Name	Pin Number	Type
IP_A6 UTPB_Split6 ² MII-TXERR	T6	Input
IP_A7 UTPB_Split7 ² MII-RXDV	T3	Input
ALE_B DSCK/AT1	J1	Bidirectional Three-state
IP_B[0:1] IWP[0:1] VFLS[0:1]	H2, J3	Bidirectional
IP_B2 IOIS16_B AT2	J2	Bidirectional Three-state
IP_B3 IWP2 VF2	G1	Bidirectional
IP_B4 LWP0 VF0	G2	Bidirectional
IP_B5 LWP1 VF1	J4	Bidirectional
IP_B6 DSDI AT0	K3	Bidirectional Three-state
IP_B7 PTR AT3	H1	Bidirectional Three-state
OP0 MII-TXD0 UtpClk_Split ²	L4	Bidirectional
OP1	L2	Output
OP2 MODCK1 STS	L1	Bidirectional
OP3 MODCK2 DSDO	M4	Bidirectional
BADDR30 REG	K4	Output
BADDR[28:29]	M3, M2	Output
AS	L3	Input

Table 35. Pin Assignments (continued)

Name	Pin Number	Type
PD12 L1RSYNCB MII-MDC UTPB3	R16	Bidirectional
PD11 RXD3 MII-TXERR RXENB	T16	Bidirectional
PD10 TXD3 MII-RXD0 TXENB	W18	Bidirectional
PD9 RXD4 MII-TXD0 UTPCLK	V17	Bidirectional
PD8 TXD4 MII-MDC MII-RXCLK	W17	Bidirectional
PD7 RTS3 MII-RXERR UTPB4	T15	Bidirectional
PD6 RTS4 MII-RXDV UTPB5	V16	Bidirectional
PD5 REJECT2 MII-TXD3 UTPB6	U15	Bidirectional
PD4 REJECT3 MII-TXD2 UTPB7	U16	Bidirectional
PD3 REJECT4 MII-TXD1 SOC	W16	Bidirectional
TMS	G18	Input
TDI DSDI	H17	Input
TCK DSCK	H16	Input



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