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Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Obsolete
Core Processor	MPC8xx
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	50MHz
Co-Processors/DSP	Communications; CPM
RAM Controllers	DRAM
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10Mbps (4), 10/100Mbps (1)
SATA	-
USB	-
Voltage - I/O	3.3V
Operating Temperature	-40°C ~ 115°C (TA)
Security Features	-
Package / Case	357-BBGA
Supplier Device Package	357-PBGA (25x25)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mpc862tcvr50b

Table 4. Power Dissipation (P_D) (continued)

Die Revision	Frequency	Typical [1]	Maximum [2]	Unit
A.1, B.0 (2:1 Mode)	66 MHz	910	1060	mW
	80 MHz	1.06	1.20	W
B.0 (2:1 Mode)	100 MHz	1.35	1.54	W

1 Typical power dissipation is measured at 3.3 V.

2 Maximum power dissipation is measured at 3.5 V.

NOTE

Values in Table 4 represent VDDL based power dissipation and do not include I/O power dissipation over VDDH. I/O power dissipation varies widely by application due to buffer current, depending on external circuitry.

6 DC Characteristics

Table 5 provides the DC electrical characteristics for the MPC862/857T/857DSL.

Table 5. DC Electrical Specifications

Characteristic	Symbol	Min	Max	Unit
Operating voltage	VDDH, VDDL, KAPWR, VDDSYN	3.135	3.465	V
	KAPWR (power-down mode)	2.0	3.6	V
	KAPWR (all other operating modes)	VDDH – 0.4	VDDH	V
Input High Voltage (all inputs except EXTAL and EXTCLK)	VIH	2.0	5.5	V
Input Low Voltage [1]	VIL	GND	0.8	V
EXTAL, EXTCLK Input High Voltage	VIHC	0.7*(VCC)	VCC+0.3	V
Input Leakage Current, Vin = 5.5 V (Except TMS, $\overline{\text{TRST}}$, DSCK and DSDI pins)	I_{in}	—	100	µA
Input Leakage Current, Vin = 3.6 V (Except TMS, $\overline{\text{TRST}}$, DSCK, and DSDI)	I_{In}	—	10	µA
Input Leakage Current, Vin = 0 V (Except TMS, $\overline{\text{TRST}}$, DSCK, and DSDI pins)	I_{In}	—	10	µA
Input Capacitance [2]	C_{in}	—	20	pF
Output High Voltage, IOH = -2.0 mA, VDDH = 3.0 V (Except XTAL, XFC, and Open drain pins)	VOH	2.4	—	V

Table 5. DC Electrical Specifications (continued)

Characteristic	Symbol	Min	Max	Unit
Output Low Voltage IOL = 2.0 mA (CLKOUT) IOL = 3.2 mA [3] IOL = 5.3 mA [4] IOL = 7.0 mA (TXD1/PA14, TXD2/PA12) IOL = 8.9 mA (TS, TA, TEA, BI, BB, HRESET, SRESET)	VOL	—	0.5	V

1 V_{IL}(max) for the I^2C interface is 0.8 V rather than the 1.5 V as specified in the I^2C standard.

2 Input capacitance is periodically sampled.

3 A(0:31), TSIZ0/REG, TSIZ1, D(0:31), DP(0:3)/IRQ(3:6), RD/WR, BURST, RSV/IRQ2, IP_B(0:1)/IWP(0:1)/VFLS(0:1), IP_B2/IOIS16_B/AT2, IP_B3/IWP2/VF2, IP_B4/LWP0/VF0, IP_B5/LWP1/VF1, IP_B6/DSDI/AT0, IP_B7/PTR/AT3, RXD1 /PA15, RXD2/PA13, L1TXDB/PA11, L1RXDB/PA10, L1TXDA/PA9, L1RXDA/PA8, TIN1/L1RCLKA/BRGO1/CLK1/PA7, BRGCLK1/TOUT1/CLK2/PA6, TIN2/L1TCLKA/BRGO2/CLK3/PA5, TOUT2/CLK4/PA4, TIN3/BRGO3/CLK5/PA3, BRGCLK2/L1RCLKB/TOUT3/CLK6/PA2, TIN4/BRGO4/CLK7/PA1, L1TCLKB/TOUT4/CLK8/PA0, REJCT1/SPISEL/PB31, SPICLK/PB30, SPIMOSI/PB29, BRGO4/SPIMISO/PB28, BRGO1/I2CSDA/PB27, BRGO2/I2CSCL/PB26, SMTXD1/PB25, SMRXD1/PB24, SMSYN1/SDACK1/PB23, SMSYN2/SDACK2/PB22, SMTXD2/L1CLKOB/PB21, SMRXD2/L1CLKOA/PB20, L1ST1/RTS1/PB19, L1ST2/RTS2/PB18, L1ST3/L1RQB/PB17, L1ST4/L1RQA/PB16, BRGO3/PB15, RSTRT1/PB14, L1ST1/RTS1/DREQ0/PC15, L1ST2/RTS2/DREQ1/PC14, L1ST3/L1RQB/PC13, L1ST4/L1RQA/PC12, CTS1/PC11, TGATE1/CD1/PC10, CTS2/PC9, TGATE2/CD2/PC8, CTS3/SDACK2/L1TSYNCB/PC7, CD3/L1RSYNCB/PC6, CTS4/SDACK1/L1TSYNCA/PC5, CD4/L1RSYNCA/PC4, PD15/L1TSYNCA, PD14/L1RSYNCA, PD13/L1TSYNCB, PD12/L1RSYNCB, PD11/RXD3, PD10/TXD3, PD9/RXD4, PD8/TXD4, PD5/REJECT2, PD6/RTS4, PD7/RTS3, PD4/REJECT3, PD3, MII_MDC, MII_TX_ER, MII_EN, MII_MDIO, MII_TXD[0:3].

4 BDIP/GPL_B(5), BR, BG, FRZ/IRQ6, CS(0:5), CS(6)/CE(1)_B, CS(7)/CE(2)_B, WE0/BS_B0/IORD, WE1/BS_B1/IOWR, WE2/BS_B2/PCOE, WE3/BS_B3/PCWE, BS_A(0:3), GPL_A0/GPL_B0, OE/GPL_A1/GPL_B1, GPL_A(2:3)/GPL_B(2:3)/CS(2:3), UPWAITA/GPL_A4, UPWAITB/GPL_B4, GPL_A5, ALE_A, CE1_A, CE2_A, ALE_B/DSCK/AT1, OP(0:1), OP2/MODCK1/STS, OP3/MODCK2/DSDO, BADDR(28:30).

7 Thermal Calculation and Measurement

For the following discussions, P_D= (VDD x IDD) + PI/O, where PI/O is the power dissipation of the I/O drivers.

7.1 Estimation with Junction-to-Ambient Thermal Resistance

An estimation of the chip junction temperature, T_J, in °C can be obtained from the equation:

$$T_J = T_A + (R_{\theta JA} \times P_D)$$

where:

T_A = ambient temperature (ºC)

$R_{\theta JA}$ = package junction-to-ambient thermal resistance (ºC/W)

P_D = power dissipation in package

The junction-to-ambient thermal resistance is an industry standard value which provides a quick and easy estimation of thermal performance. However, the answer is only an estimate; test cases have demonstrated that errors of a factor of two (in the quantity T_J-T_A) are possible.

If the board temperature is known, an estimate of the junction temperature in the environment can be made using the following equation:

$$T_J = T_B + (R_{\theta JB} \times P_D)$$

where:

$R_{\theta JB}$ = junction-to-board thermal resistance (°C/W)

T_B = board temperature (°C)

P_D = power dissipation in package

If the board temperature is known and the heat loss from the package case to the air can be ignored, acceptable predictions of junction temperature can be made. For this method to work, the board and board mounting must be similar to the test board used to determine the junction-to-board thermal resistance, namely a 2s2p (board with a power and a ground plane) and vias attaching the thermal balls to the ground plane.

7.4 Estimation Using Simulation

When the board temperature is not known, a thermal simulation of the application is needed. The simple two resistor model can be used with the thermal simulation of the application [2], or a more accurate and complex model of the package can be used in the thermal simulation.

7.5 Experimental Determination

To determine the junction temperature of the device in the application after prototypes are available, the thermal characterization parameter (Ψ_{JT}) can be used to determine the junction temperature with a measurement of the temperature at the top center of the package case using the following equation:

$$T_J = T_T + (\Psi_{JT} \times P_D)$$

where:

Ψ_{JT} = thermal characterization parameter

T_T = thermocouple temperature on top of package

P_D = power dissipation in package

The thermal characterization parameter is measured per JESD51-2 specification published by JEDEC using a 40-gauge type T thermocouple epoxied to the top center of the package case. The thermocouple should be positioned so that the thermocouple junction rests on the package. A small amount of epoxy is placed over the thermocouple junction and over about 1 mm of wire extending from the junction. The thermocouple wire is placed flat against the package case to avoid measurement errors caused by cooling effects of the thermocouple wire.

7.6 References

Semiconductor Equipment and Materials International 805 East Middlefield Rd. Mountain View, CA 94043	(415) 964-5111
MIL-SPEC and EIA/JESD (JEDEC) Specifications (Available from Global Engineering Documents)	800-854-7179 or 303-397-7956
JEDEC Specifications	http://www.jedec.org

1. C.E. Triplett and B. Joiner, “An Experimental Characterization of a 272 PBGA Within an Automotive Engine Controller Module,” Proceedings of SemiTherm, San Diego, 1998, pp. 47-54.

2. B. Joiner and V. Adams, “Measurement and Simulation of Junction to Board Thermal Resistance and Its Application in Thermal Modeling,” Proceedings of SemiTherm, San Diego, 1999, pp. 212-220.

8 Layout Practices

Each V_{CC} pin on the MPC862/857T/857DSL should be provided with a low-impedance path to the board’s supply. Each GND pin should likewise be provided with a low-impedance path to ground. The power supply pins drive distinct groups of logic on chip. The V_{CC} power supply should be bypassed to ground using at least four 0.1 µF by-pass capacitors located as close as possible to the four sides of the package. The capacitor leads and associated printed circuit traces connecting to chip V_{CC} and GND should be kept to less than half an inch per capacitor lead. A four-layer board is recommended, employing two inner layers as V_{CC} and GND planes.

All output pins on the MPC862/857T/857DSL have fast rise and fall times. Printed circuit (PC) trace interconnection length should be minimized in order to minimize undershoot and reflections caused by these fast output switching times. This recommendation particularly applies to the address and data busses. Maximum PC trace lengths of six inches are recommended. Capacitance calculations should consider all device loads as well as parasitic capacitances due to the PC traces. Attention to proper PCB layout and bypassing becomes especially critical in systems with higher capacitive loads because these loads create higher transient currents in the V_{CC} and GND circuits. Pull up all unused inputs or signals that will be inputs during reset. Special care should be taken to minimize the noise levels on the PLL supply pins.

9 Bus Signal Timing

The maximum bus speed supported by the MPC862/857T/857DSL is 66 MHz. Higher-speed parts must be operated in half-speed bus mode (for example, an MPC862/857T/857DSL used at 80MHz must be configured for a 40 MHz bus). Table 6 shows the period ranges for standard part frequencies.

Table 6. Period Range for Standard Part Frequencies

Freq	50 MHz		66 MHz		80 MHz		100 MHz	
	Min	Max	Min	Max	Min	Max	Min	Max
Period	20.00	30.30	15.15	30.30	25.00	30.30	20.00	30.30

Table 7. Bus Operation Timings (continued)

Num	Characteristic	33 MHz		40 MHz		50 MHz		66 MHz		Unit
		Min	Max	Min	Max	Min	Max	Min	Max	
B8a	CLKOUT to TSIZ(0:1), REG, RSV, AT(0:3) BDIP, PTR valid (MAX = 0.25 x B1 + 6.3)	7.60	13.80	6.30	12.50	5.00	11.30	3.80	10.00	ns
B8b	CLKOUT to BR, BG, VFLS(0:1), VF(0:2), IWP(0:2), FRZ, LWP(0:1), STS Valid [4] (MAX = 0.25 x B1 + 6.3)	7.60	13.80	6.30	12.50	5.00	11.30	3.80	10.00	ns
B9	CLKOUT to A(0:31), BADDR(28:30), RD/WR, BURST, D(0:31), DP(0:3), TSIZ(0:1), REG, RSV, AT(0:3), PTR High-Z (MAX = 0.25 x B1 + 6.3)	7.60	13.80	6.30	12.50	5.00	11.30	3.80	10.00	ns
B11	CLKOUT to TS, BB assertion (MAX = 0.25 x B1 + 6.0)	7.60	13.60	6.30	12.30	5.00	11.00	3.80	11.30	ns
B11a	CLKOUT to TA, BI assertion (when driven by the memory controller or PCMCIA interface) (MAX = 0.00 x B1 + 9.30 [5])	2.50	9.30	2.50	9.30	2.50	9.30	2.50	9.80	ns
B12	CLKOUT to TS, BB negation (MAX = 0.25 x B1 + 4.8)	7.60	12.30	6.30	11.00	5.00	9.80	3.80	8.50	ns
B12a	CLKOUT to TA, BI negation (when driven by the memory controller or PCMCIA interface) (MAX = 0.00 x B1 + 9.00)	2.50	9.00	2.50	9.00	2.50	9.00	2.50	9.00	ns
B13	CLKOUT to TS, BB High-Z (MIN = 0.25 x B1)	7.60	21.60	6.30	20.30	5.00	19.00	3.80	14.00	ns
B13a	CLKOUT to TA, BI High-Z (when driven by the memory controller or PCMCIA interface) (MIN = 0.00 x B1 + 2.5)	2.50	15.00	2.50	15.00	2.50	15.00	2.50	15.00	ns
B14	CLKOUT to TEA assertion (MAX = 0.00 x B1 + 9.00)	2.50	9.00	2.50	9.00	2.50	9.00	2.50	9.00	ns
B15	CLKOUT to TEA High-Z (MIN = 0.00 x B1 + 2.50)	2.50	15.00	2.50	15.00	2.50	15.00	2.50	15.00	ns
B16	TA, BI valid to CLKOUT (setup time) (MIN = 0.00 x B1 + 6.00)	6.00	—	6.00	—	6.00	—	6.00	—	ns
B16a	TEA, KR, RETRY, CR valid to CLKOUT (setup time) (MIN = 0.00 x B1 + 4.5)	4.50	—	4.50	—	4.50	—	4.50	—	ns
B16b	BB, BG, BR, valid to CLKOUT (setup time) [6] (4MIN = 0.00 x B1 + 0.00)	4.00	—	4.00	—	4.00	—	4.00	—	ns
B17	CLKOUT to TA, TEA, BI, BB, BG, BR valid (hold time) (MIN = 0.00 x B1 + 1.00 [7])	1.00	—	1.00	—	1.00	—	2.00	—	ns

Figure 6 provides the timing for the synchronous output signals.

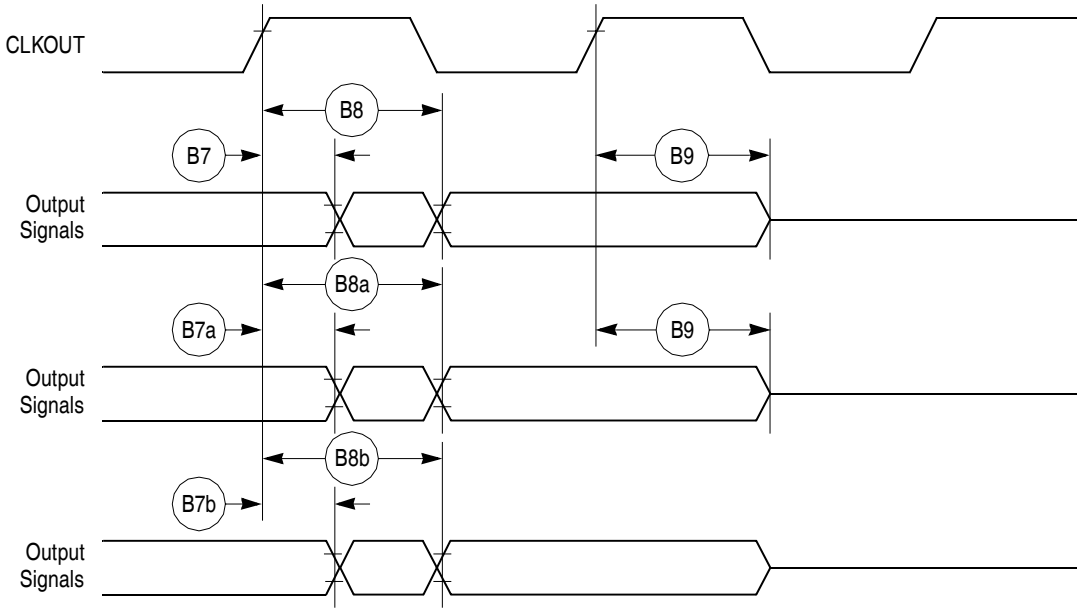


Figure 6. Synchronous Output Signals Timing

Figure 7 provides the timing for the synchronous active pull-up and open-drain output signals.

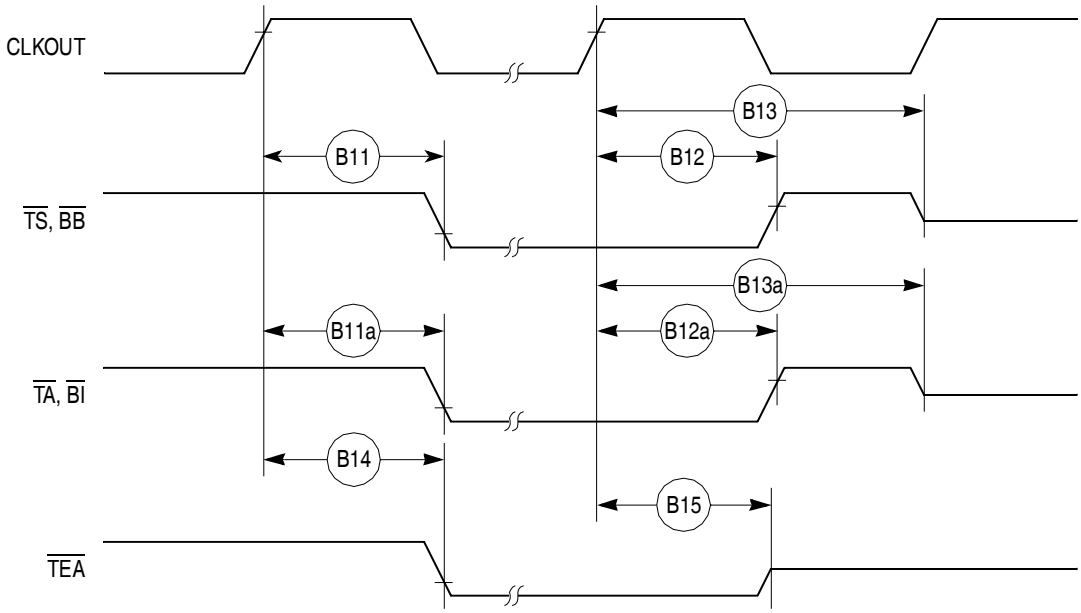


Figure 7. Synchronous Active Pull-Up Resistor and Open-Drain Outputs Signals Timing

Figure 15 through Figure 17 provide the timing for the external bus write controlled by various GPCM factors.

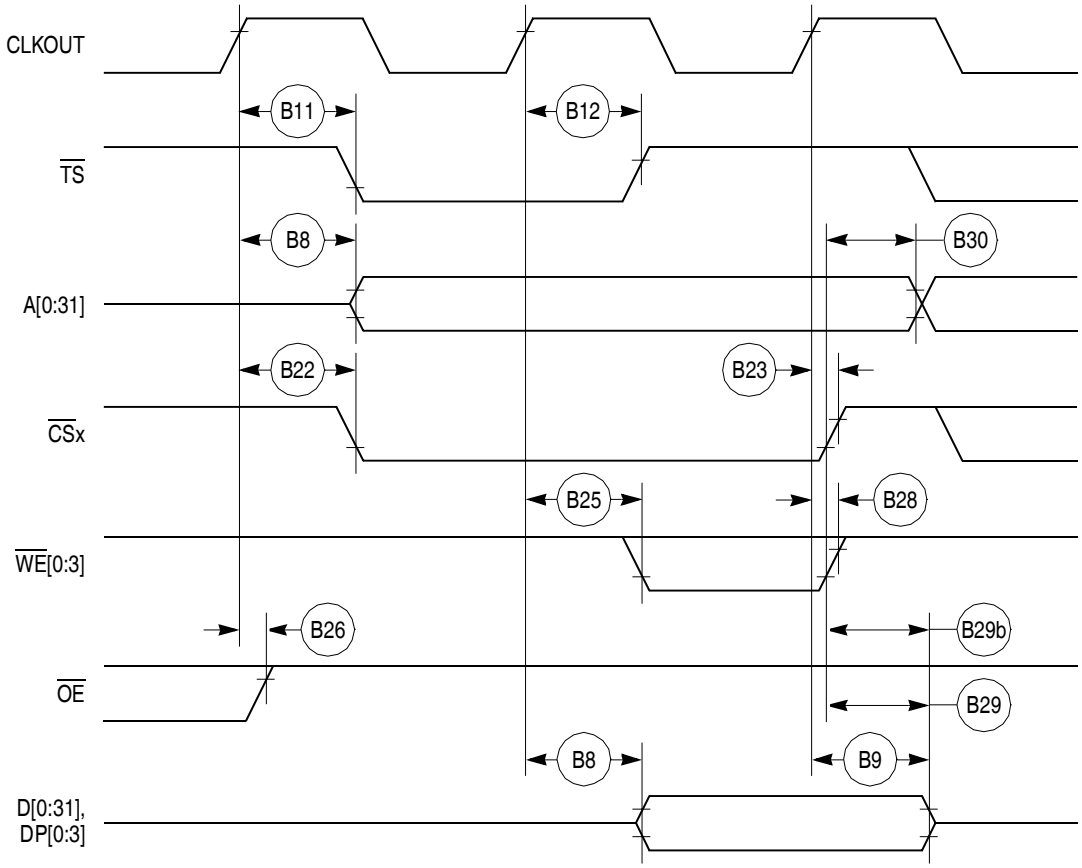


Figure 15. External Bus Write Timing (GPCM Controlled—TRLX = 0,1 CSNT = 0)

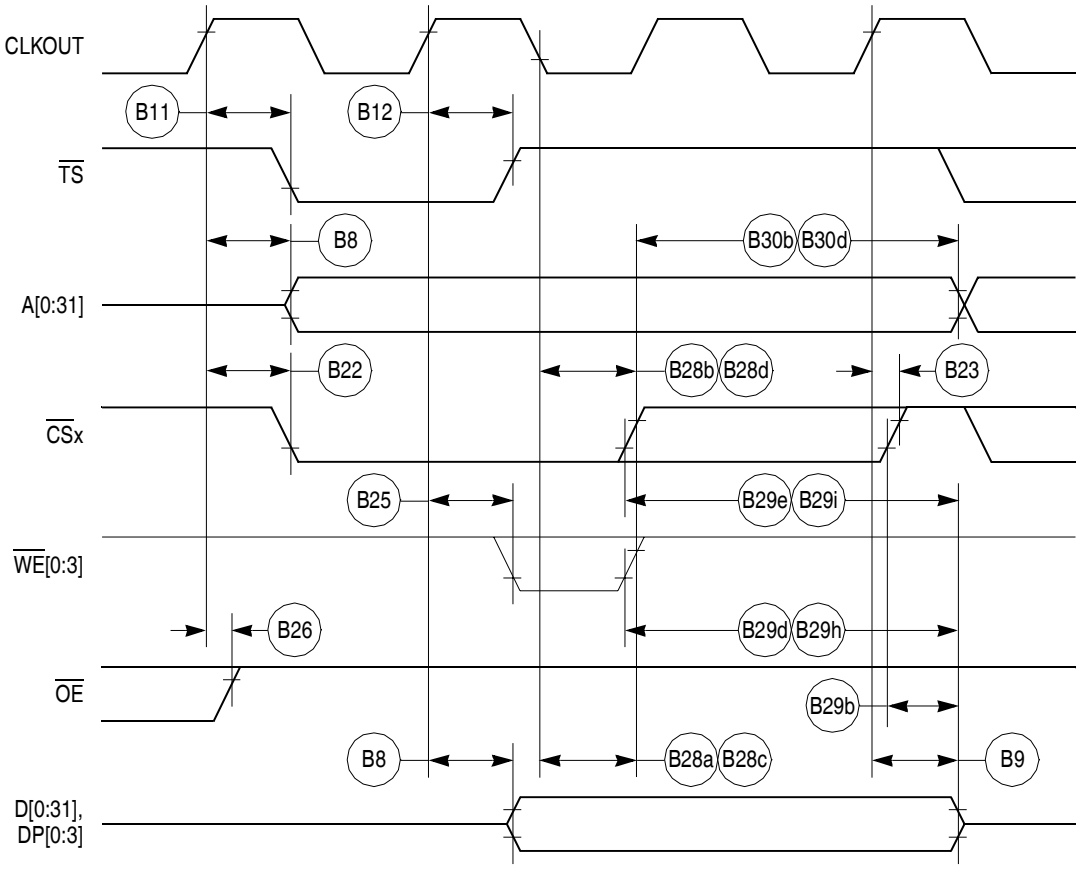


Figure 17. External Bus Write Timing (GPCM Controlled—TRLX = 0,1, CSNT = 1)

Figure 21 provides the timing for the synchronous external master access controlled by the GPCM.

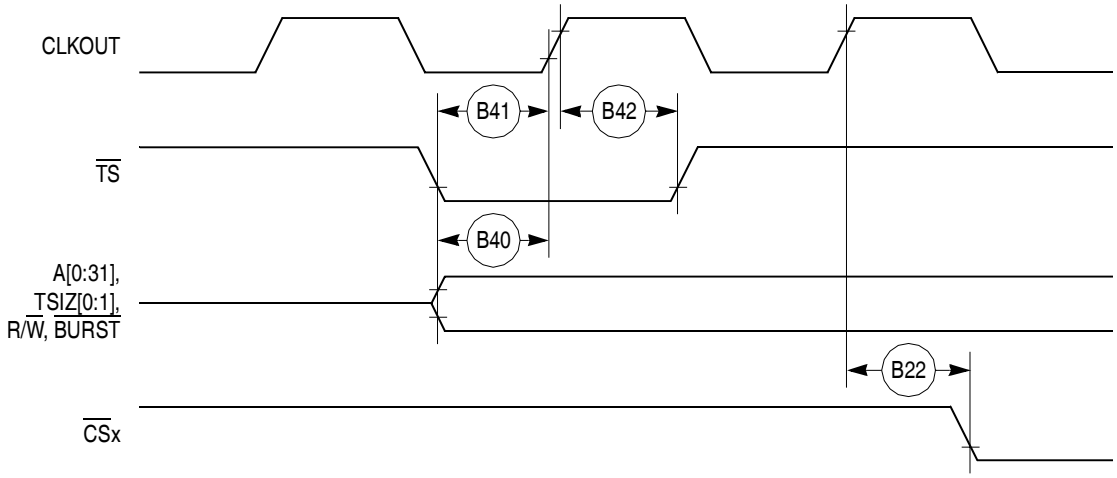


Figure 21. Synchronous External Master Access Timing (GPCM Handled ACS = 00)

Figure 22 provides the timing for the asynchronous external master memory access controlled by the GPCM.

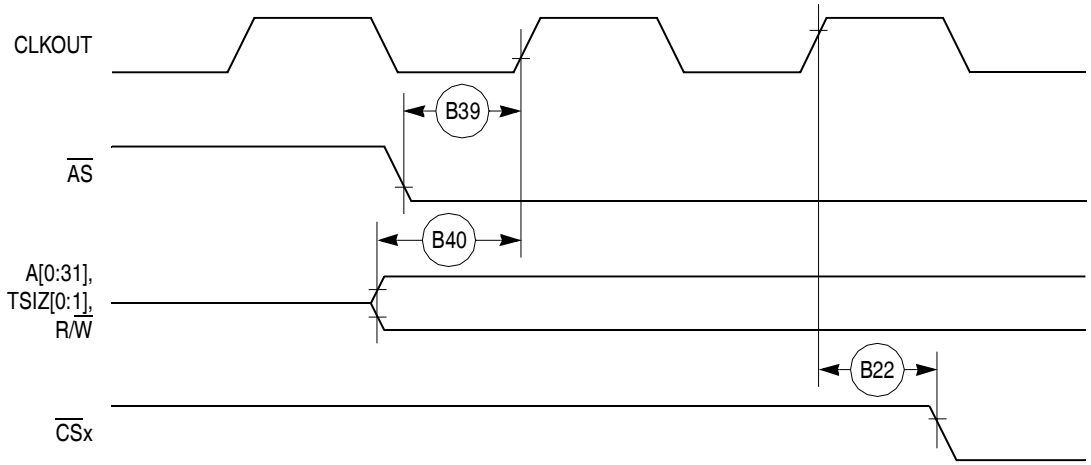


Figure 22. Asynchronous External Master Memory Access Timing (GPCM Controlled—ACS = 00)

Figure 23 provides the timing for the asynchronous external master control signals negation.

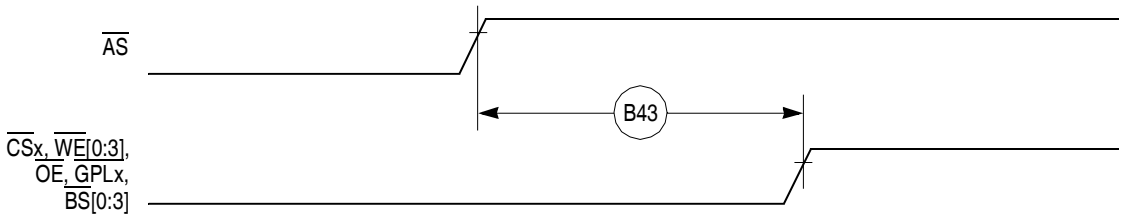


Figure 23. Asynchronous External Master—Control Signals Negation Timing

Table 10 shows the PCMCIA port timing for the MPC862/857T/857DSL.

Table 10. PCMCIA Port Timing

Num	Characteristic	33 MHz		40 MHz		50 MHz		66 MHz		Unit
		Min	Max	Min	Max	Min	Max	Min	Max	
P57	CLKOUT to OPx Valid (MAX = 0.00 x B1 + 19.00)	—	19.00	—	19.00	—	19.00	—	19.00	ns
P58	$\overline{\text{HRESET}}$ negated to OPx drive [1] (MIN = 0.75 x B1 + 3.00)	25.70	—	21.70	—	18.00	—	14.40	—	ns
P59	IP_Xx valid to CLKOUT rising edge (MIN = 0.00 x B1 + 5.00)	5.00	—	5.00	—	5.00	—	5.00	—	ns
P60	CLKOUT rising edge to IP_Xx invalid (MIN = 0.00 x B1 + 1.00)	1.00	—	1.00	—	1.00	—	1.00	—	ns

1 OP2 and OP3 only.

Figure 29 provides the PCMCIA output port timing for the MPC862/857T/857DSL.

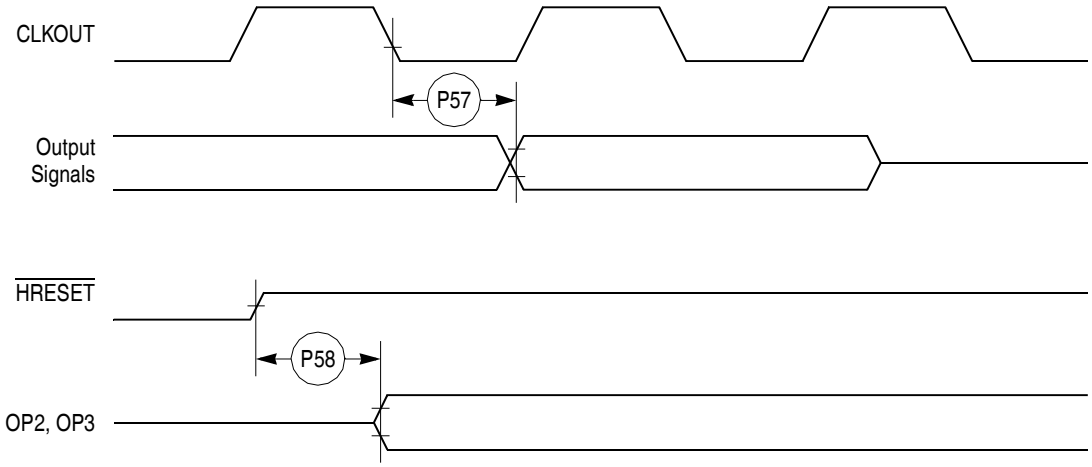


Figure 29. PCMCIA Output Port Timing

Figure 30 provides the PCMCIA output port timing for the MPC862/857T/857DSL.

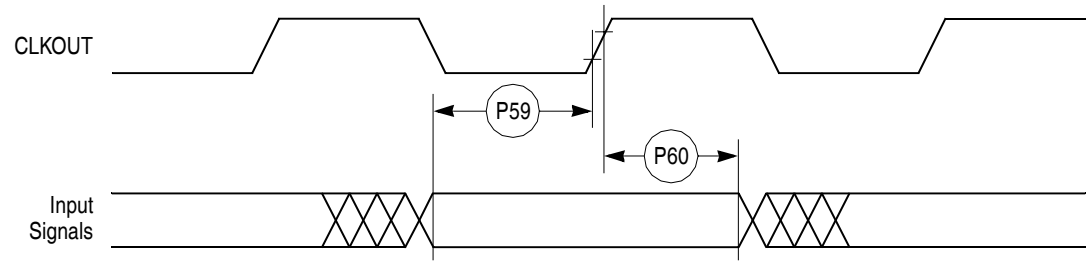


Figure 30. PCMCIA Input Port Timing

Table 12 shows the reset timing for the MPC862/857T/857DSL.

Table 12. Reset Timing

Num	Characteristic	33 MHz		40 MHz		50 MHz		66 MHz		Unit
		Min	Max	Min	Max	Min	Max	Min	Max	
R69	CLKOUT to $\overline{\text{HRESET}}$ high impedance (MAX = 0.00 x B1 + 20.00)	—	20.00	—	20.00	—	20.00	—	20.00	ns
R70	CLKOUT to $\overline{\text{SRESET}}$ high impedance (MAX = 0.00 x B1 + 20.00)	—	20.00	—	20.00	—	20.00	—	20.00	ns
R71	$\overline{\text{RSTCONF}}$ pulse width (MIN = 17.00 x B1)	515.20	—	425.00	—	340.00	—	257.60	—	ns
R72	—	—	—	—	—	—	—	—	—	—
R73	Configuration data to HRESET rising edge set up time (MIN = 15.00 x B1 + 50.00)	504.50	—	425.00	—	350.00	—	277.30	—	ns
R74	Configuration data to $\overline{\text{RSTCONF}}$ rising edge set up time (MIN = 0.00 x B1 + 350.00)	350.00	—	350.00	—	350.00	—	350.00	—	ns
R75	Configuration data hold time after $\overline{\text{RSTCONF}}$ negation (MIN = 0.00 x B1 + 0.00)	0.00	—	0.00	—	0.00	—	0.00	—	ns
R76	Configuration data hold time after $\overline{\text{HRESET}}$ negation (MIN = 0.00 x B1 + 0.00)	0.00	—	0.00	—	0.00	—	0.00	—	ns
R77	$\overline{\text{HRESET}}$ and $\overline{\text{RSTCONF}}$ asserted to data out drive (MAX = 0.00 x B1 + 25.00)	—	25.00	—	25.00	—	25.00	—	25.00	ns
R78	$\overline{\text{RSTCONF}}$ negated to data out high impedance. (MAX = 0.00 x B1 + 25.00)	—	25.00	—	25.00	—	25.00	—	25.00	ns
R79	CLKOUT of last rising edge before chip three-states $\overline{\text{HRESET}}$ to data out high impedance. (MAX = 0.00 x B1 + 25.00)	—	25.00	—	25.00	—	25.00	—	25.00	ns
R80	DSDI, DSCK set up (MIN = 3.00 x B1)	90.90	—	75.00	—	60.00	—	45.50	—	ns
R81	DSDI, DSCK hold time (MIN = 0.00 x B1 + 0.00)	0.00	—	0.00	—	0.00	—	0.00	—	ns
R82	$\overline{\text{SRESET}}$ negated to CLKOUT rising edge for DSDI and DSCK sample (MIN = 8.00 x B1)	242.40	—	200.00	—	160.00	—	121.20	—	ns

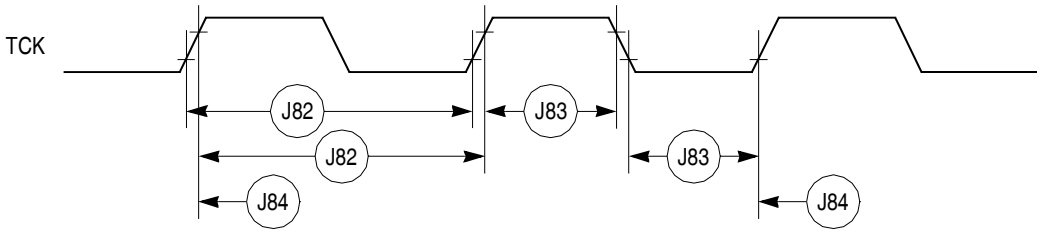


Figure 36. JTAG Test Clock Input Timing

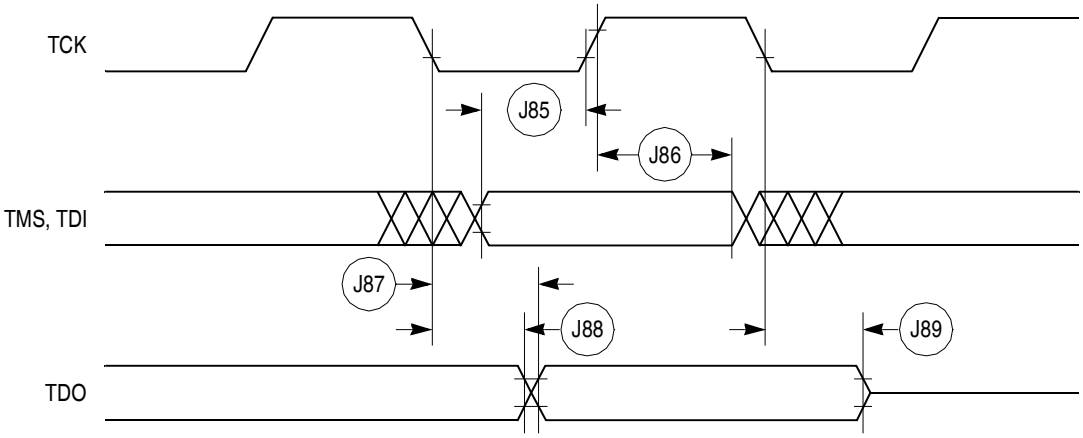


Figure 37. JTAG Test Access Port Timing Diagram

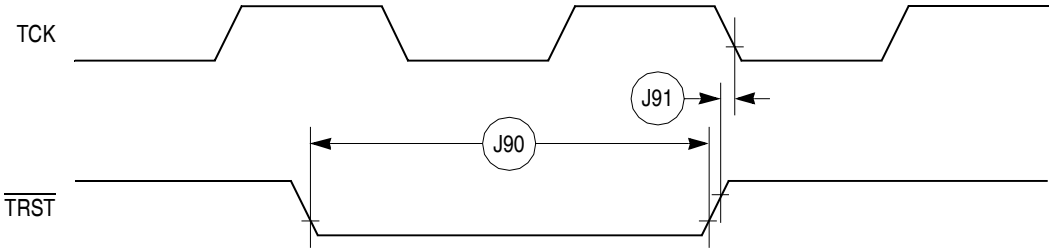


Figure 38. JTAG $\overline{\text{TRST}}$ Timing Diagram

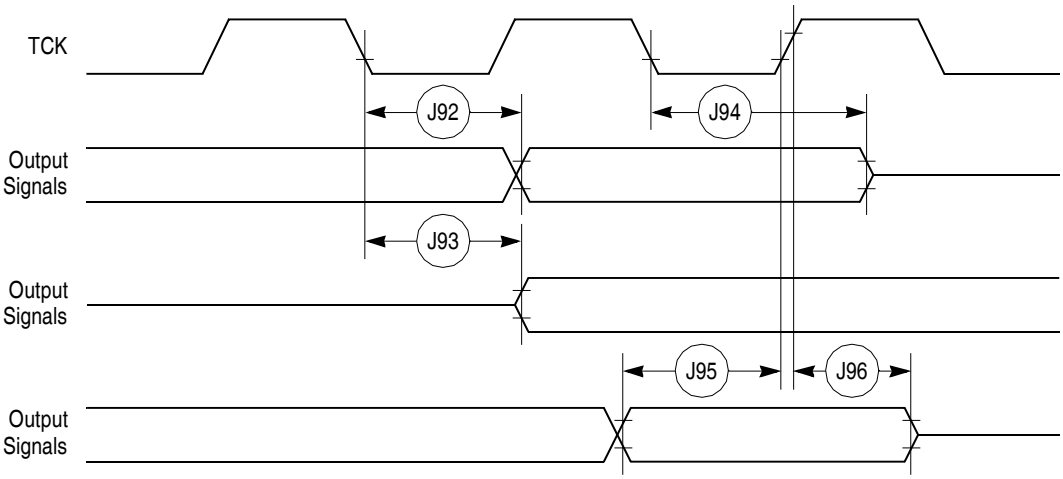


Figure 39. Boundary Scan (JTAG) Timing Diagram

11 CPM Electrical Characteristics

This section provides the AC and DC electrical specifications for the communications processor module (CPM) of the MPC862/857T/857DSL.

11.1 PIP/PIO AC Electrical Specifications

Table 14 provides the PIP/PIO AC timings as shown in Figure 40 though Figure 44.

Table 14. PIP/PIO Timing

Num	Characteristic	All Frequencies		Unit
		Min	Max	
21	Data-in setup time to STBI low	0	—	ns
22	Data-in hold time to STBI high	2.5 – t3 [1]	—	clk
23	STBI pulse width	1.5	—	clk
24	STBO pulse width	1 clk – 5 ns	—	ns
25	Data-out setup time to STBO low	2	—	clk
26	Data-out hold time from STBO high	5	—	clk
27	STBI low to STBO low (Rx interlock)	—	2	clk
28	STBI low to STBO high (Tx interlock)	2	—	clk
29	Data-in setup time to clock high	15	—	ns
30	Data-in hold time from clock high	7.5	—	ns
31	Clock low to data-out valid (CPU writes data, control, or direction)	—	25	ns

1 t3 = Specification 23

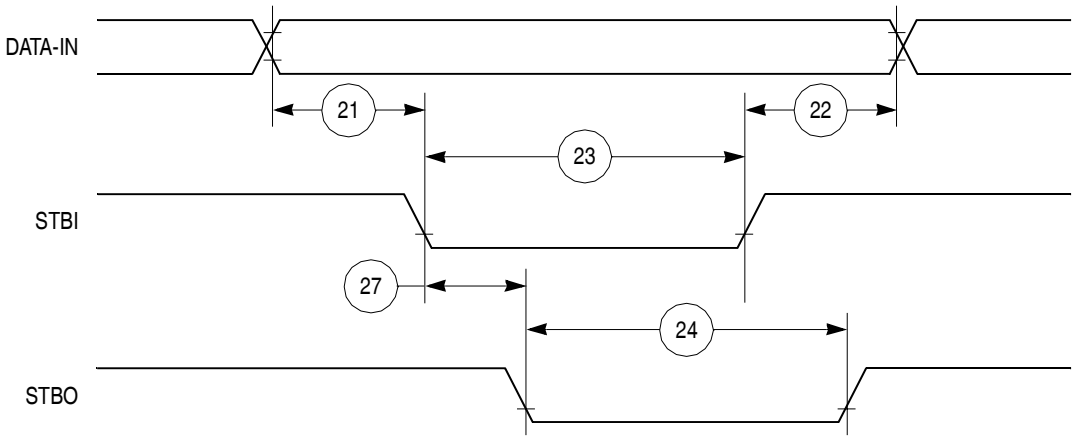


Figure 40. PIP Rx (Interlock Mode) Timing Diagram

Table 19. SI Timing (continued)

Num	Characteristic	All Frequencies		Unit
		Min	Max	
83a	L1RCLK, L1TCLK width high (DSC = 1)[3]	P + 10	—	ns
84	L1CLK edge to L1CLKO valid (DSC = 1)	—	30.00	ns
85	$\overline{\text{L1RQ}}$ valid before falling edge of L1TSYNC[4]	1.00	—	L1TCLK
86	L1GR setup time[2]	42.00	—	ns
87	L1GR hold time	42.00	—	ns
88	L1CLK edge to L1SYNC valid (FSD = 00) CNT = 0000, BYT = 0, DSC = 0)	—	0.00	ns

1 The ratio SyncCLK/L1RCLK must be greater than 2.5/1.

2 These specs are valid for IDL mode only.

3 Where P = 1/CLKOUT. Thus for a 25-MHz CLKO1 rate, P = 40 ns.

4 These strobes and TxD on the first bit of the frame become valid after L1CLK edge or L1SYNC, whichever is later.

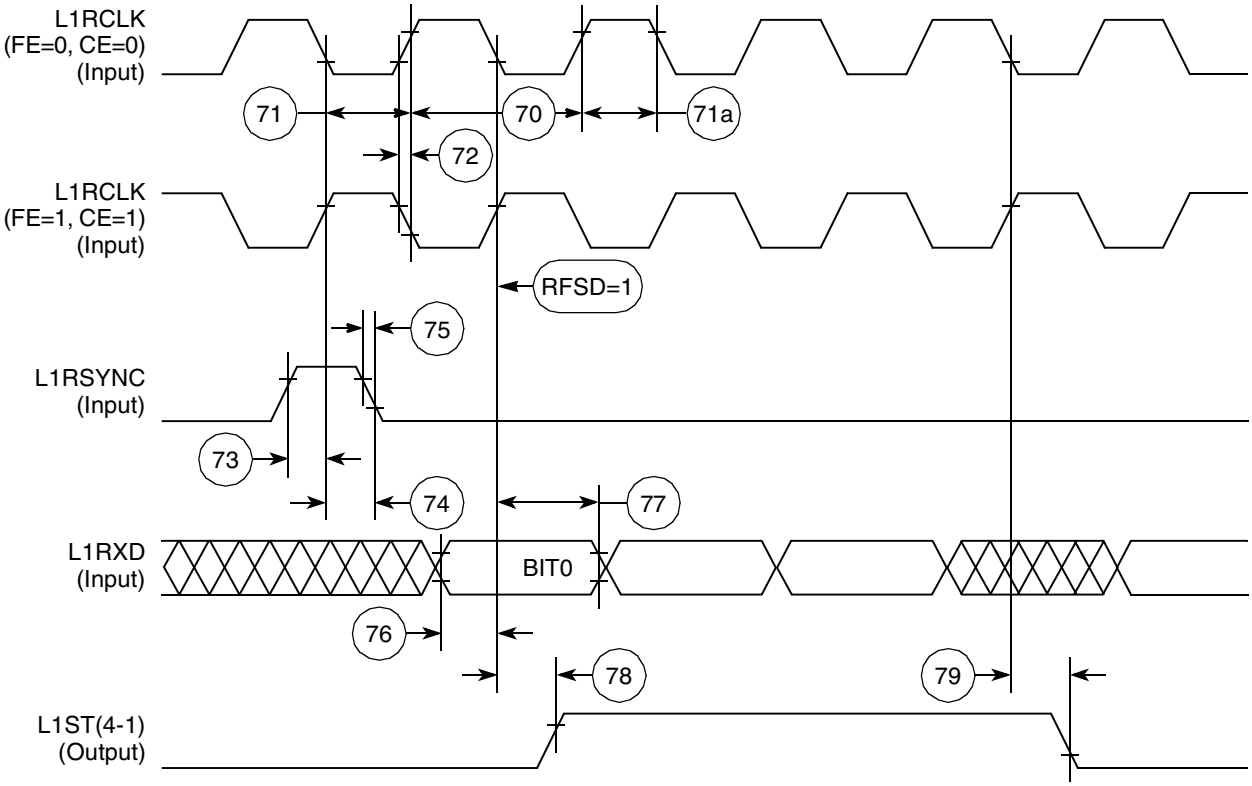
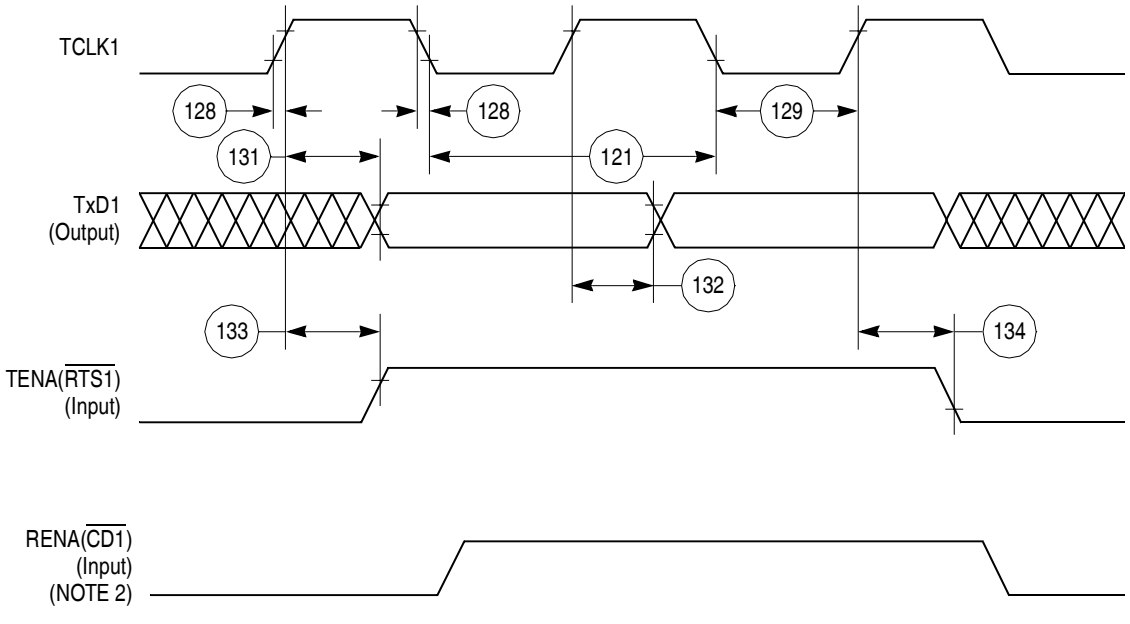


Figure 52. SI Receive Timing Diagram with Normal Clocking (DSC = 0)



NOTES:
1. Transmit clock invert (TCI) bit in GSMR is set.
2. If RENA is deasserted before TENA, or RENA is not asserted at all during transmit, then the CSL bit is set in the buffer descriptor at the end of the frame transmission.

Figure 62. Ethernet Transmit Timing Diagram

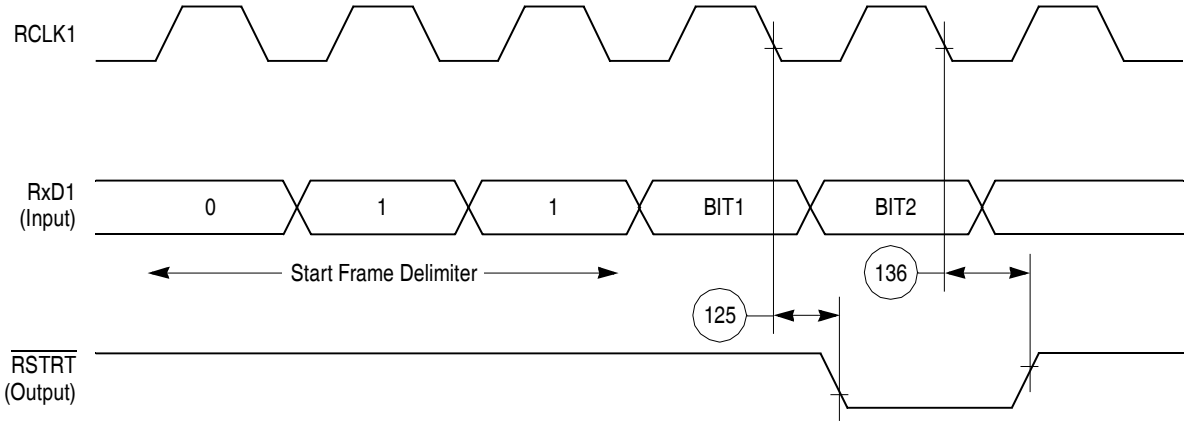


Figure 63. CAM Interface Receive Start Timing Diagram

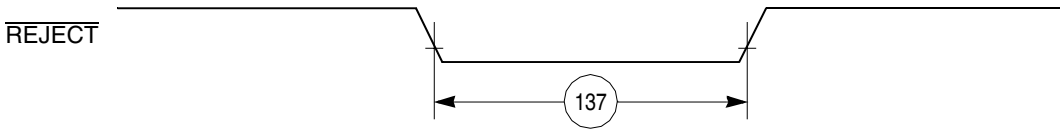


Figure 64. CAM Interface REJECT Timing Diagram

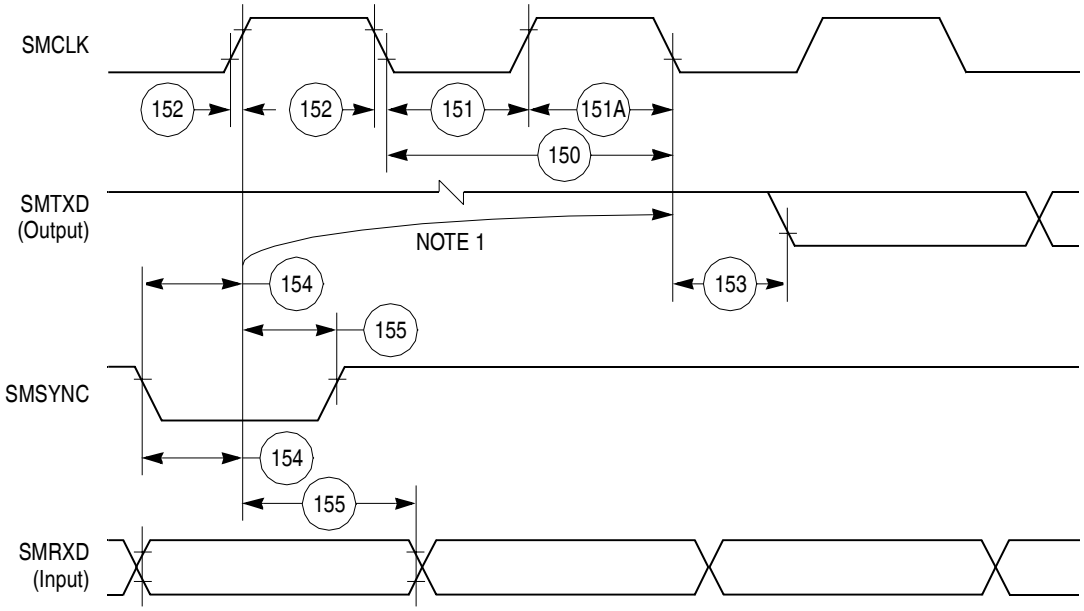
11.9 SMC Transparent AC Electrical Specifications

Table 23 provides the SMC transparent timings as shown in Figure 65.

Table 23. SMC Transparent Timing

Num	Characteristic	All Frequencies		Unit
		Min	Max	
150	SMCLK clock period [1]	100	—	ns
151	SMCLK width low	50	—	ns
151A	SMCLK width high	50	—	ns
152	SMCLK rise/fall time	—	15	ns
153	SMTXD active delay (from SMCLK falling edge)	10	50	ns
154	SMRXD/SMSYNC setup time	20	—	ns
155	RXD1/SMSYNC hold time	5	—	ns

1 SyncCLK must be at least twice as fast as SMCLK.



NOTE:
1. This delay is equal to an integer number of character-length clocks.

Figure 65. SMC Transparent Timing Diagram

Figure 70 shows the I^{2}C bus timing.

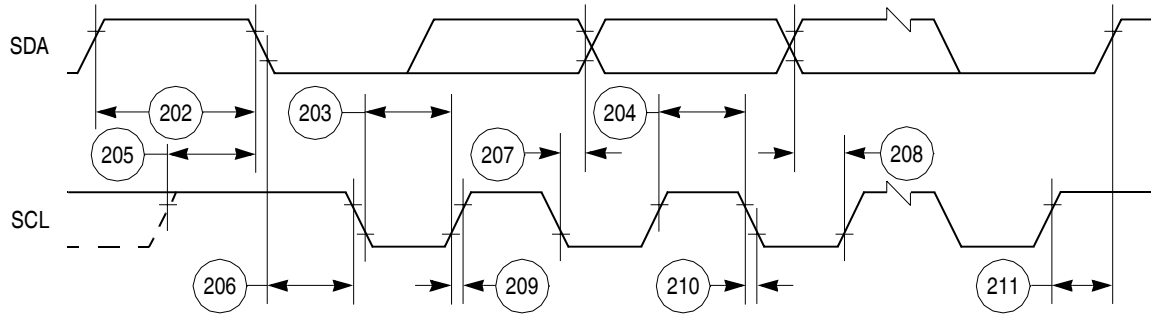


Figure 70. I^{2}C Bus Timing Diagram

12 UTOPIA AC Electrical Specifications

Table 28 shows the AC electrical specifications for the UTOPIA interface.

Table 28. UTOPIA AC Electrical Specifications

Num	Signal Characteristic	Direction	Min	Max	Unit
U1	UtpClk rise/fall time (Internal clock option)	Output		4 ns	ns
	Duty cycle		50	50	%
	Frequency			33	MHz
U1a	UtpClk rise/fall time (external clock option)	Input		4ns	ns
	Duty cycle		40	60	%
	Frequency			33	MHz
U2	$\overline{\text{RxEnb}}$ and $\overline{\text{TxEnb}}$ active delay	Output	2 ns	16 ns	ns
U3	UTPB, SOC, Rxclav and Txclav setup time	Input	4 ns		ns
U4	UTPB, SOC, Rxclav and Txclav hold time	Input	1 ns		ns
U5	UTPB, SOC active delay (and PHREQ and PHSEL active delay in MPHY mode)	Output	2 ns	16 ns	ns

Table 35. Pin Assignments (continued)

Name	Pin Number	Type
$\overline{\text{GPL\_A5}}$	D3	Output
$\overline{\text{PORESET}}$	R2	Input
$\overline{\text{RSTCONF}}$	P3	Input
$\overline{\text{HRESET}}$	N4	Open-drain
$\overline{\text{SRESET}}$	P2	Open-drain
XTAL	P1	Analog Output
EXTAL	N1	Analog Input (3.3 V only)
XFC	T2	Analog Input
CLKOUT	W3	Output
EXTCLK	N2	Input (3.3 V only)
TEXP	N3	Output
ALE_A MII-TXD1	K2	Output
$\overline{\text{CE1\_A}}$ MII-TXD2	B3	Output
$\overline{\text{CE2\_A}}$ MII-TXD3	A3	Output
$\overline{\text{WAIT\_A}}$ SOC_Split[2]	R3	Input
$\overline{\text{WAIT\_B}}$	R4	Input
IP_A0 UTPB_Split0[2] MII-RXD3	T5	Input
IP_A1 UTPB_Split1[2] MII-RXD2	T4	Input
IP_A2 $\overline{\text{IOIS16\_A}}$ UTPB_Split2[2] MII-RXD1	U3	Input
IP_A3 UTPB_Split3[2] MII-RXD0	W2	Input
IP_A4 UTPB_Split4[2] MII-RXCLK	U4	Input
IP_A5 UTPB_Split5[2] MII-RXERR	U5	Input

Table 35. Pin Assignments (continued)

Name	Pin Number	Type
PC13 L1RQb L1ST3 RTS3	E18	Bidirectional
PC12 L1RQa L1ST4 RTS4	F18	Bidirectional
PC11 CTS1	J19	Bidirectional
PC10 CD1 TGATE1	K19	Bidirectional
PC9 CTS2	L18	Bidirectional
PC8 CD2 TGATE2	M18	Bidirectional
PC7 CTS3 L1TSYNCB SDACK2	M16	Bidirectional
PC6 CD3 L1RSYNCB	R19	Bidirectional
PC5 CTS4 L1TSYNCA SDACK1	T18	Bidirectional
PC4 CD4 L1RSYNCA	T17	Bidirectional
PD15 L1TSYNCA MII-RXD3 UTPB0	U17	Bidirectional
PD14 L1RSYNCA MII-RXD2 UTPB1	V19	Bidirectional
PD13 L1TSYNCB MII-RXD1 UTPB2	V18	Bidirectional

15 Document Revision History

Table 36 lists significant changes between revisions of this document.

Table 36. Document Revision History

Rev. No.	Date	Substantive Changes
0	2001	Initial revision
0.1	9/2001	Change extended temperature from 95 to 105
0.2	11/2001	Revised for new template, changed Table 7 B23 max value @ 66 MHz from 2 ns to 8 ns.
0.3	4/2002	• Timing modified and equations added, for Rev. A and B devices. • Modified power numbers and temperature ranges. Added ESAR UTOPIA timing.
1.0	9/2002	• Specification changed to include the MPC857T and MPC857DSL. • Changed maximum operating frequency from 80 MHz to 100 MHz. • Removed MPC862DP, DT, and SR derivatives and part numbers. • Corrected power dissipation numbers. • Changed UTOPIA maximum frequency from 50 MHz to 33 MHz. • Changed part number ordering information to Rev. B devices only. • To maximum ratings for temperature, added frequency ranges.
1.1	5/2003	Changed SPI Master Timing Specs. 162 and 164
1.2	8/2003	• Changed B28a through B28d and B29b to show that TRLX can be 0 or 1. • Non-technical reformatting
2.0	11/2004	• Added a table footnote to Table 5 DC Electrical Specifications about meeting the VIL Max of the I2C Standard. • Updated document template.
3.0	2/2006	• Changed Tj from 95C to 105C in table 34