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### Understanding [Embedded - Microprocessors](#)

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

### Applications of [Embedded - Microprocessors](#)

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

#### Details

Product Status	Active
Core Processor	-
Number of Cores/Bus Width	-
Speed	-
Co-Processors/DSP	-
RAM Controllers	-
Graphics Acceleration	-
Display & Interface Controllers	-
Ethernet	-
SATA	-
USB	-
Voltage - I/O	-
Operating Temperature	-
Security Features	-
Package / Case	-
Supplier Device Package	-
Purchase URL	<a href="https://www.e-xfl.com/pro/item?MUrl=&amp;PartUrl=mpc862tvr50b">https://www.e-xfl.com/pro/item?MUrl=&amp;PartUrl=mpc862tvr50b</a>

## Features

- Sleep—All units disabled except RTC, PIT, time base, and decremter with PLL active for fast wake up
- Deep sleep—All units disabled including PLL except RTC, PIT, time base, and decremter.
- Power down mode— All units powered down except PLL, RTC, PIT, time base and decremter
- Debug interface
  - Eight comparators: four operate on instruction address, two operate on data address, and two operate on data
  - Supports conditions: = ≠ < >
  - Each watchpoint can generate a break point internally
- 3.3 V operation with 5-V TTL compatibility except EXTAL and EXTCLK
- 357-pin plastic ball grid array (PBGA) package
- Operation up to 100MHz

The MPC862/857T/857DSL is comprised of three modules that each use the 32-bit internal bus: the MPC8xx core, the system integration unit (SIU), and the communication processor module (CPM). The MPC862P/862T block diagram is shown in [Figure 1](#). The MPC857T/857DSL block diagram is shown in [Figure 2](#).

**Table 4. Power Dissipation ( $P_D$ ) (continued)**

Die Revision	Frequency	Typical <sup>1</sup>	Maximum <sup>2</sup>	Unit
A.1, B.0 (2:1 Mode)	66 MHz	910	1060	mW
	80 MHz	1.06	1.20	W
B.0 (2:1 Mode)	100 MHz	1.35	1.54	W

<sup>1</sup> Typical power dissipation is measured at 3.3 V.

<sup>2</sup> Maximum power dissipation is measured at 3.5 V.

#### NOTE

Values in [Table 4](#) represent VDDL based power dissipation and do not include I/O power dissipation over VDDH. I/O power dissipation varies widely by application due to buffer current, depending on external circuitry.

## 6 DC Characteristics

[Table 5](#) provides the DC electrical characteristics for the MPC862/857T/857DSL.

**Table 5. DC Electrical Specifications**

Characteristic	Symbol	Min	Max	Unit
Operating voltage	VDDH, VDDL, KAPWR, VDDSYN	3.135	3.465	V
	KAPWR (power-down mode)	2.0	3.6	V
	KAPWR (all other operating modes)	VDDH – 0.4	VDDH	V
Input High Voltage (all inputs except EXTAL and EXTCLK)	VIH	2.0	5.5	V
Input Low Voltage <sup>1</sup>	VIL	GND	0.8	V
EXTAL, EXTCLK Input High Voltage	VIHC	0.7*(VCC)	VCC+0.3	V
Input Leakage Current, Vin = 5.5 V (Except TMS, $\overline{\text{TRST}}$ , DSCK and DSDI pins)	I <sub>in</sub>	—	100	μA
Input Leakage Current, Vin = 3.6 V (Except TMS, $\overline{\text{TRST}}$ , DSCK, and DSDI)	I <sub>in</sub>	—	10	μA
Input Leakage Current, Vin = 0 V (Except TMS, $\overline{\text{TRST}}$ , DSCK, and DSDI pins)	I <sub>in</sub>	—	10	μA
Input Capacitance <sup>2</sup>	C <sub>in</sub>	—	20	pF
Output High Voltage, IOH = -2.0 mA, VDDH = 3.0 V (Except XTAL, XFC, and Open drain pins)	VOH	2.4	—	V

**Table 5. DC Electrical Specifications (continued)**

Characteristic	Symbol	Min	Max	Unit
Output Low Voltage IOL = 2.0 mA (CLKOUT) IOL = 3.2 mA <sup>3</sup> IOL = 5.3 mA <sup>4</sup> IOL = 7.0 mA (TXD1/PA14, TXD2/PA12) IOL = 8.9 mA ( $\overline{TS}$ , $\overline{TA}$ , $\overline{TEA}$ , $\overline{BI}$ , $\overline{BB}$ , $\overline{HRESET}$ , $\overline{SRESET}$ )	VOL	—	0.5	V

<sup>1</sup>  $V_{IL}(\max)$  for the I<sup>2</sup>C interface is 0.8 V rather than the 1.5 V as specified in the I<sup>2</sup>C standard.

<sup>2</sup> Input capacitance is periodically sampled.

<sup>3</sup> A(0:31),  $\overline{TSIZ0}/\overline{REG}$ ,  $\overline{TSIZ1}$ , D(0:31), DP(0:3)/ $\overline{IRQ}$ (3:6),  $\overline{RD}/\overline{WR}$ ,  $\overline{BURST}$ ,  $\overline{RSV}/\overline{IRQ2}$ ,  $\overline{IP\_B}(0:1)/\overline{IWP}(0:1)/\overline{VFLS}(0:1)$ ,  $\overline{IP\_B2}/\overline{IOIS16\_B}/\overline{AT2}$ ,  $\overline{IP\_B3}/\overline{IWP2}/\overline{VF2}$ ,  $\overline{IP\_B4}/\overline{LWP0}/\overline{VF0}$ ,  $\overline{IP\_B5}/\overline{LWP1}/\overline{VF1}$ ,  $\overline{IP\_B6}/\overline{DSDI}/\overline{AT0}$ ,  $\overline{IP\_B7}/\overline{PTR}/\overline{AT3}$ ,  $\overline{RXD1}/\overline{PA15}$ ,  $\overline{RXD2}/\overline{PA13}$ ,  $\overline{L1TXDB}/\overline{PA11}$ ,  $\overline{L1RXDB}/\overline{PA10}$ ,  $\overline{L1TXDA}/\overline{PA9}$ ,  $\overline{L1RXDA}/\overline{PA8}$ ,  $\overline{TIN1}/\overline{L1RCLKA}/\overline{BRGO1}/\overline{CLK1}/\overline{PA7}$ ,  $\overline{BRGCLK1}/\overline{TOUT1}/\overline{CLK2}/\overline{PA6}$ ,  $\overline{TIN2}/\overline{L1TCLKA}/\overline{BRGO2}/\overline{CLK3}/\overline{PA5}$ ,  $\overline{TOUT2}/\overline{CLK4}/\overline{PA4}$ ,  $\overline{TIN3}/\overline{BRGO3}/\overline{CLK5}/\overline{PA3}$ ,  $\overline{BRGCLK2}/\overline{L1RCLKB}/\overline{TOUT3}/\overline{CLK6}/\overline{PA2}$ ,  $\overline{TIN4}/\overline{BRGO4}/\overline{CLK7}/\overline{PA1}$ ,  $\overline{L1TCLKB}/\overline{TOUT4}/\overline{CLK8}/\overline{PA0}$ ,  $\overline{REJECT1}/\overline{SPISEL}/\overline{PB31}$ ,  $\overline{SPICLK}/\overline{PB30}$ ,  $\overline{SPIMOSI}/\overline{PB29}$ ,  $\overline{BRGO4}/\overline{SPIMISO}/\overline{PB28}$ ,  $\overline{BRGO1}/\overline{I2CSDA}/\overline{PB27}$ ,  $\overline{BRGO2}/\overline{I2CSCL}/\overline{PB26}$ ,  $\overline{SMTXD1}/\overline{PB25}$ ,  $\overline{SMRXD1}/\overline{PB24}$ ,  $\overline{SMSYN1}/\overline{SDACK1}/\overline{PB23}$ ,  $\overline{SMSYN2}/\overline{SDACK2}/\overline{PB22}$ ,  $\overline{SMTXD2}/\overline{L1CLKOB}/\overline{PB21}$ ,  $\overline{SMRXD2}/\overline{L1CLKOA}/\overline{PB20}$ ,  $\overline{L1ST1}/\overline{RTS1}/\overline{PB19}$ ,  $\overline{L1ST2}/\overline{RTS2}/\overline{PB18}$ ,  $\overline{L1ST3}/\overline{L1RQB}/\overline{PB17}$ ,  $\overline{L1ST4}/\overline{L1RQA}/\overline{PB16}$ ,  $\overline{BRGO3}/\overline{PB15}$ ,  $\overline{RSTRT1}/\overline{PB14}$ ,  $\overline{L1ST1}/\overline{RTS1}/\overline{DREQ0}/\overline{PC15}$ ,  $\overline{L1ST2}/\overline{RTS2}/\overline{DREQ1}/\overline{PC14}$ ,  $\overline{L1ST3}/\overline{L1RQB}/\overline{PC13}$ ,  $\overline{L1ST4}/\overline{L1RQA}/\overline{PC12}$ ,  $\overline{CTS1}/\overline{PC11}$ ,  $\overline{TGATE1}/\overline{CD1}/\overline{PC10}$ ,  $\overline{CTS2}/\overline{PC9}$ ,  $\overline{TGATE2}/\overline{CD2}/\overline{PC8}$ ,  $\overline{CTS3}/\overline{SDACK2}/\overline{L1TSYNCB}/\overline{PC7}$ ,  $\overline{CD3}/\overline{L1RSYNCB}/\overline{PC6}$ ,  $\overline{CTS4}/\overline{SDACK1}/\overline{L1TSYNCA}/\overline{PC5}$ ,  $\overline{CD4}/\overline{L1RSYNCA}/\overline{PC4}$ ,  $\overline{PD15}/\overline{L1TSYNCA}$ ,  $\overline{PD14}/\overline{L1RSYNCA}$ ,  $\overline{PD13}/\overline{L1TSYNCB}$ ,  $\overline{PD12}/\overline{L1RSYNCB}$ ,  $\overline{PD11}/\overline{RXD3}$ ,  $\overline{PD10}/\overline{TXD3}$ ,  $\overline{PD9}/\overline{RXD4}$ ,  $\overline{PD8}/\overline{TXD4}$ ,  $\overline{PD5}/\overline{REJECT2}$ ,  $\overline{PD6}/\overline{RTS4}$ ,  $\overline{PD7}/\overline{RTS3}$ ,  $\overline{PD4}/\overline{REJECT3}$ ,  $\overline{PD3}$ ,  $\overline{MII\_MDC}$ ,  $\overline{MII\_TX\_ER}$ ,  $\overline{MII\_EN}$ ,  $\overline{MII\_MDIO}$ ,  $\overline{MII\_TXD}[0:3]$ .

<sup>4</sup>  $\overline{BDIP}/\overline{GPL\_B}(5)$ ,  $\overline{BR}$ ,  $\overline{BG}$ ,  $\overline{FRZ}/\overline{IRQ6}$ ,  $\overline{CS}(0:5)$ ,  $\overline{CS}(6)/\overline{CE}(1)\_B$ ,  $\overline{CS}(7)/\overline{CE}(2)\_B$ ,  $\overline{WE0}/\overline{BS\_B0}/\overline{IORD}$ ,  $\overline{WE1}/\overline{BS\_B1}/\overline{IOWR}$ ,  $\overline{WE2}/\overline{BS\_B2}/\overline{PCOE}$ ,  $\overline{WE3}/\overline{BS\_B3}/\overline{PCWE}$ ,  $\overline{BS\_A}(0:3)$ ,  $\overline{GPL\_A0}/\overline{GPL\_B0}$ ,  $\overline{OE}/\overline{GPL\_A1}/\overline{GPL\_B1}$ ,  $\overline{GPL\_A}(2:3)/\overline{GPL\_B}(2:3)/\overline{CS}(2:3)$ ,  $\overline{UPWAITA}/\overline{GPL\_A4}$ ,  $\overline{UPWAITB}/\overline{GPL\_B4}$ ,  $\overline{GPL\_A5}$ ,  $\overline{ALE\_A}$ ,  $\overline{CE1\_A}$ ,  $\overline{CE2\_A}$ ,  $\overline{ALE\_B}/\overline{DSCK}/\overline{AT1}$ ,  $\overline{OP}(0:1)$ ,  $\overline{OP2}/\overline{MODCK1}/\overline{STS}$ ,  $\overline{OP3}/\overline{MODCK2}/\overline{DSDO}$ ,  $\overline{BADDR}(28:30)$ .

## 7 Thermal Calculation and Measurement

For the following discussions,  $P_D = (V_{DD} \times I_{DD}) + P_{I/O}$ , where  $P_{I/O}$  is the power dissipation of the I/O drivers.

### 7.1 Estimation with Junction-to-Ambient Thermal Resistance

An estimation of the chip junction temperature,  $T_J$ , in °C can be obtained from the equation:

$$T_J = T_A + (R_{\theta JA} \times P_D)$$

where:

$T_A$  = ambient temperature (°C)

$R_{\theta JA}$  = package junction-to-ambient thermal resistance (°C/W)

$P_D$  = power dissipation in package

The junction-to-ambient thermal resistance is an industry standard value which provides a quick and easy estimation of thermal performance. However, the answer is only an estimate; test cases have demonstrated that errors of a factor of two (in the quantity  $T_J - T_A$ ) are possible.

## 7.6 References

Semiconductor Equipment and Materials International  
 805 East Middlefield Rd.  
 Mountain View, CA 94043

(415) 964-5111

MIL-SPEC and EIA/JESD (JEDEC) Specifications  
 (Available from Global Engineering Documents)

800-854-7179 or  
 303-397-7956

JEDEC Specifications

<http://www.jedec.org>

1. C.E. Triplett and B. Joiner, "An Experimental Characterization of a 272 PBGA Within an Automotive Engine Controller Module," Proceedings of SemiTherm, San Diego, 1998, pp. 47-54.
2. B. Joiner and V. Adams, "Measurement and Simulation of Junction to Board Thermal Resistance and Its Application in Thermal Modeling," Proceedings of SemiTherm, San Diego, 1999, pp. 212-220.

## 8 Layout Practices

Each  $V_{CC}$  pin on the MPC862/857T/857DSL should be provided with a low-impedance path to the board's supply. Each GND pin should likewise be provided with a low-impedance path to ground. The power supply pins drive distinct groups of logic on chip. The  $V_{CC}$  power supply should be bypassed to ground using at least four 0.1  $\mu$ F by-pass capacitors located as close as possible to the four sides of the package. The capacitor leads and associated printed circuit traces connecting to chip  $V_{CC}$  and GND should be kept to less than half an inch per capacitor lead. A four-layer board is recommended, employing two inner layers as  $V_{CC}$  and GND planes.

All output pins on the MPC862/857T/857DSL have fast rise and fall times. Printed circuit (PC) trace interconnection length should be minimized in order to minimize undershoot and reflections caused by these fast output switching times. This recommendation particularly applies to the address and data busses. Maximum PC trace lengths of six inches are recommended. Capacitance calculations should consider all device loads as well as parasitic capacitances due to the PC traces. Attention to proper PCB layout and bypassing becomes especially critical in systems with higher capacitive loads because these loads create higher transient currents in the  $V_{CC}$  and GND circuits. Pull up all unused inputs or signals that will be inputs during reset. Special care should be taken to minimize the noise levels on the PLL supply pins.

## 9 Bus Signal Timing

The maximum bus speed supported by the MPC862/857T/857DSL is 66 MHz. Higher-speed parts must be operated in half-speed bus mode (for example, an MPC862/857T/857DSL used at 80MHz must be configured for a 40 MHz bus). [Table 6](#) shows the period ranges for standard part frequencies.

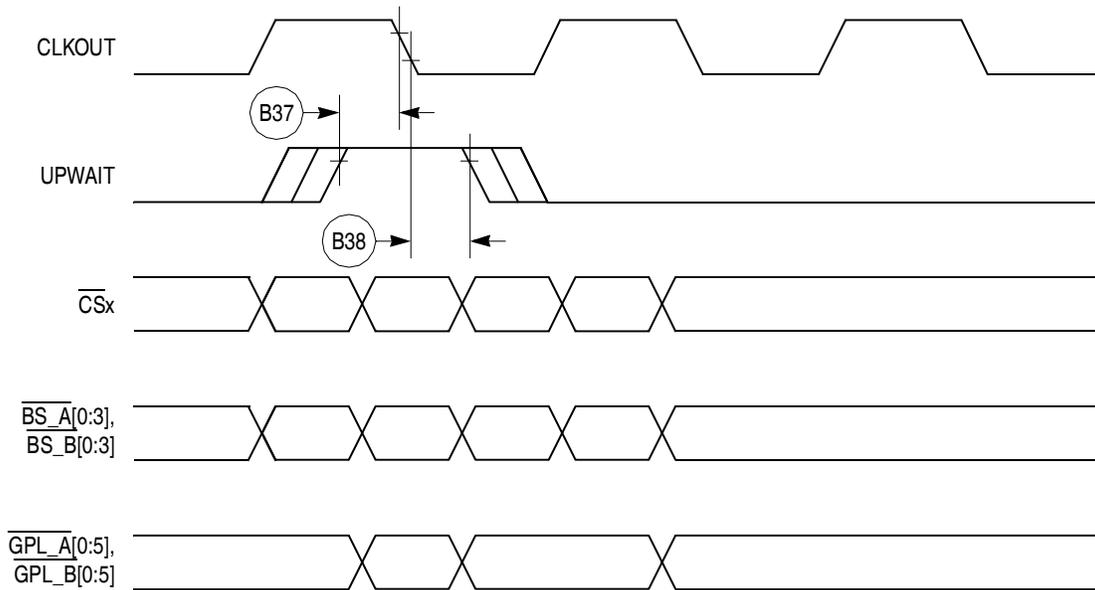
**Table 6. Period Range for Standard Part Frequencies**

Freq	50 MHz		66 MHz		80 MHz		100 MHz	
	Min	Max	Min	Max	Min	Max	Min	Max
Period	20.00	30.30	15.15	30.30	25.00	30.30	20.00	30.30

Table 7. Bus Operation Timings (continued)

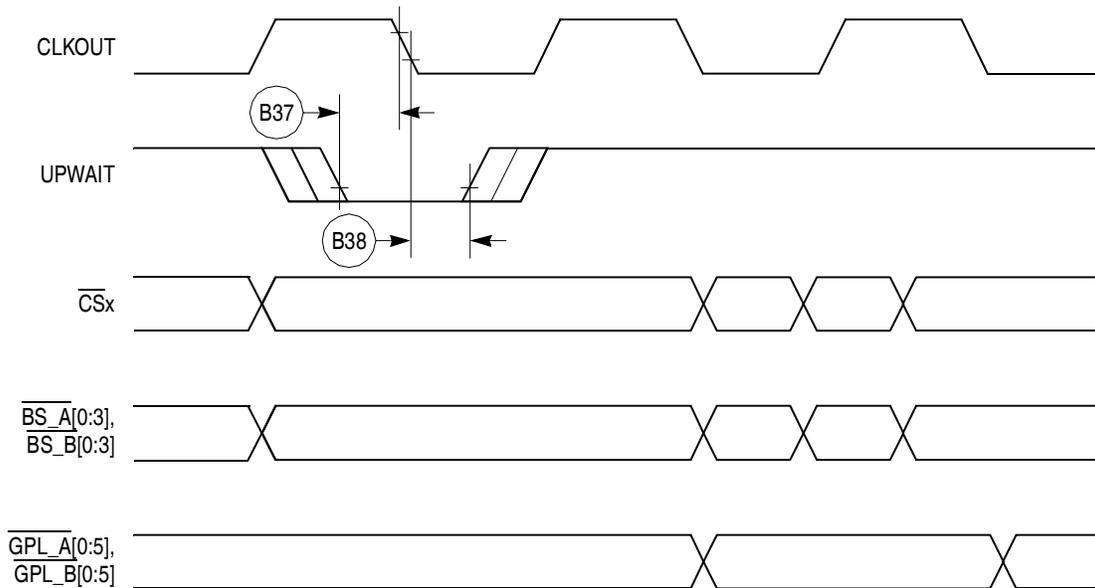
Num	Characteristic	33 MHz		40 MHz		50 MHz		66 MHz		Unit
		Min	Max	Min	Max	Min	Max	Min	Max	
B27	A(0:31) and BADDR(28:30) to $\overline{CS}$ asserted GPCM ACS = 10, TRLX = 1 (MIN = 1.25 x B1 - 2.00)	35.90	—	29.30	—	23.00	—	16.90	—	ns
B27a	A(0:31) and BADDR(28:30) to $\overline{CS}$ asserted GPCM ACS = 11, TRLX = 1 (MIN = 1.50 x B1 - 2.00)	43.50	—	35.50	—	28.00	—	20.70	—	ns
B28	CLKOUT rising edge to $\overline{WE}$ (0:3) negated GPCM write access CSNT = 0 (MAX = 0.00 x B1 + 9.00)	—	9.00	—	9.00	—	9.00	—	9.00	ns
B28a	CLKOUT falling edge to $\overline{WE}$ (0:3) negated GPCM write access TRLX = 0, 1, CSNT = 1, EBDF = 0 (MAX = 0.25 x B1 + 6.80)	7.60	14.30	6.30	13.00	5.00	11.80	3.80	10.50	ns
B28b	CLKOUT falling edge to $\overline{CS}$ negated GPCM write access TRLX = 0,1, CSNT = 1 ACS = 10 or ACS = 11, EBDF = 0 (MAX = 0.25 x B1 + 6.80)	—	14.30	—	13.00	—	11.80	—	10.50	ns
B28c	CLKOUT falling edge to $\overline{WE}$ (0:3) negated GPCM write access TRLX = 0, CSNT = 1 write access TRLX = 0,1, CSNT = 1, EBDF = 1 (MAX = 0.375 x B1 + 6.6)	10.90	18.00	10.90	18.00	7.00	14.30	5.20	12.30	ns
B28d	CLKOUT falling edge to $\overline{CS}$ negated GPCM write access TRLX = 0,1, CSNT = 1, ACS = 10, or ACS = 11, EBDF = 1 (MAX = 0.375 x B1 + 6.6)	—	18.00	—	18.00	—	14.30	—	12.30	ns
B29	$\overline{WE}$ (0:3) negated to D(0:31), DP(0:3) High-Z GPCM write access, CSNT = 0, EBDF = 0 (MIN = 0.25 x B1 - 2.00)	5.60	—	4.30	—	3.00	—	1.80	—	ns
B29a	$\overline{WE}$ (0:3) negated to D(0:31), DP(0:3) High-Z GPCM write access, TRLX = 0, CSNT = 1, EBDF = 0 (MIN = 0.50 x B1 - 2.00)	13.20	—	10.50	—	8.00	—	5.60	—	ns
B29b	$\overline{CS}$ negated to D(0:31), DP(0:3), High Z GPCM write access, ACS = 00, TRLX = 0,1 & CSNT = 0 (MIN = 0.25 x B1 - 2.00)	5.60	—	4.30	—	3.00	—	1.80	—	ns
B29c	$\overline{CS}$ negated to D(0:31), DP(0:3) High-Z GPCM write access, TRLX = 0, CSNT = 1, ACS = 10, or ACS = 11 EBDF = 0 (MIN = 0.50 x B1 - 2.00)	13.20	—	10.50	—	8.00	—	5.60	—	ns

Figure 19 provides the timing for the asynchronous asserted UPWAIT signal controlled by the UPM.



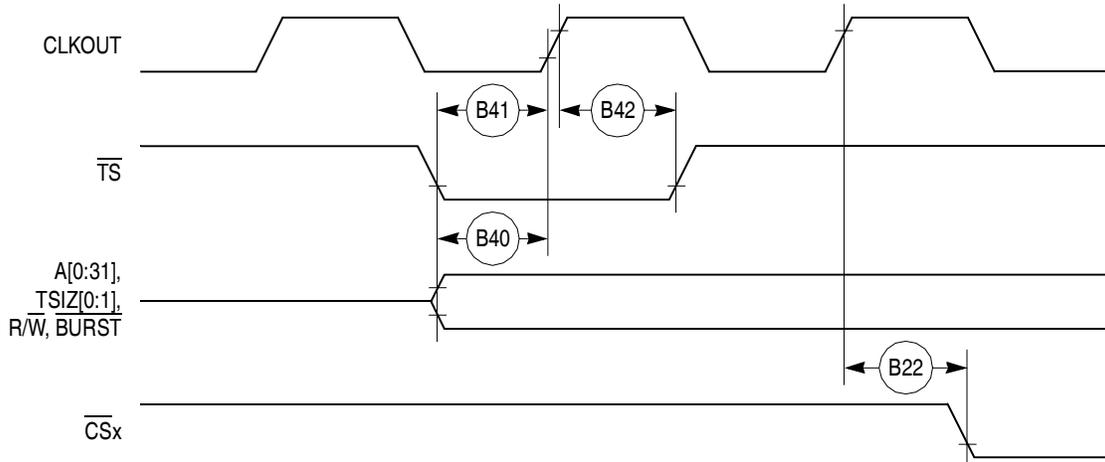
**Figure 19. Asynchronous UPWAIT Asserted Detection in UPM Handled Cycles Timing**

Figure 20 provides the timing for the asynchronous negated UPWAIT signal controlled by the UPM.



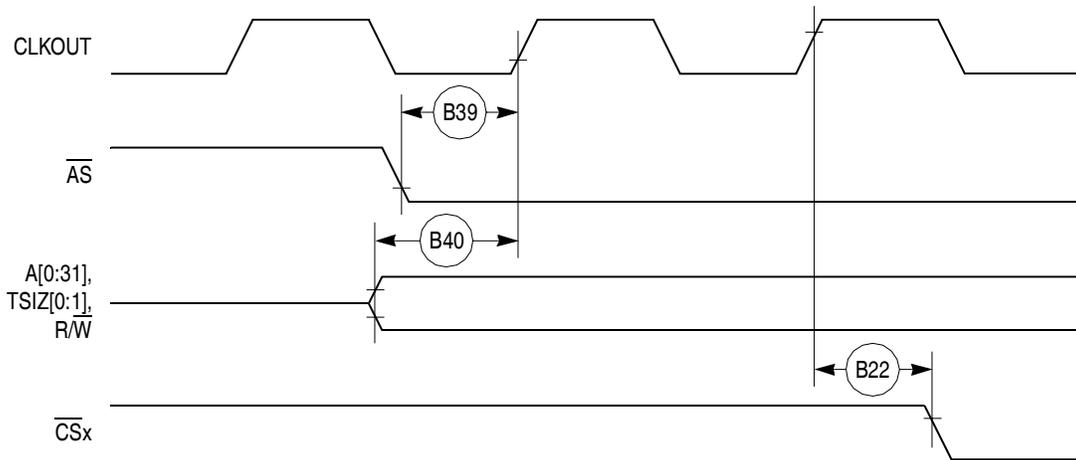
**Figure 20. Asynchronous UPWAIT Negated Detection in UPM Handled Cycles Timing**

Figure 21 provides the timing for the synchronous external master access controlled by the GPCM.



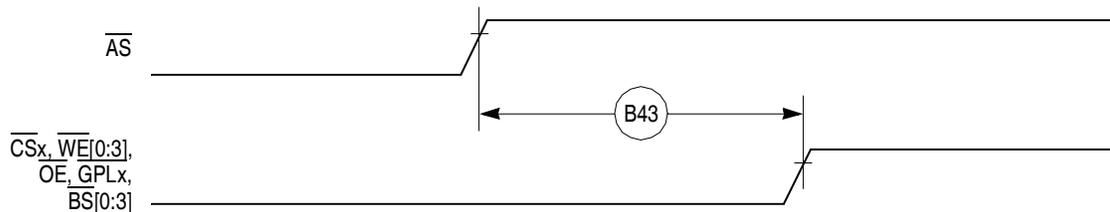
**Figure 21. Synchronous External Master Access Timing (GPCM Handled ACS = 00)**

Figure 22 provides the timing for the asynchronous external master memory access controlled by the GPCM.



**Figure 22. Asynchronous External Master Memory Access Timing (GPCM Controlled—ACS = 00)**

Figure 23 provides the timing for the asynchronous external master control signals negation.



**Figure 23. Asynchronous External Master—Control Signals Negation Timing**

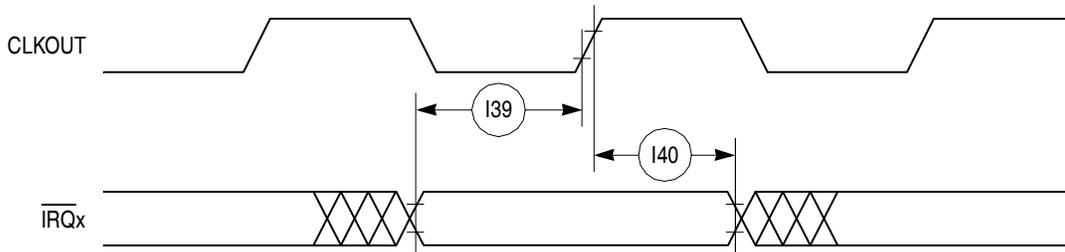
Table 8 provides interrupt timing for the MPC862/857T/857DSL.

**Table 8. Interrupt Timing**

Num	Characteristic <sup>1</sup>	All Frequencies		Unit
		Min	Max	
I39	$\overline{\text{IRQ}}_x$ valid to CLKOUT rising edge (set up time)	6.00		ns
I40	$\overline{\text{IRQ}}_x$ hold time after CLKOUT	2.00		ns
I41	$\overline{\text{IRQ}}_x$ pulse width low	3.00		ns
I42	$\overline{\text{IRQ}}_x$ pulse width high	3.00		ns
I43	$\overline{\text{IRQ}}_x$ edge-to-edge time	$4 \times T_{\text{CLKOUT}}$		—

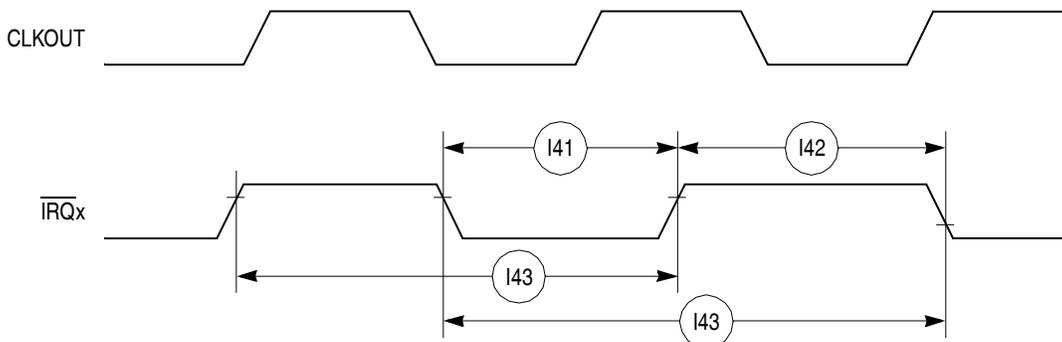
<sup>1</sup> The timings I39 and I40 describe the testing conditions under which the  $\overline{\text{IRQ}}$  lines are tested when being defined as level sensitive. The  $\overline{\text{IRQ}}$  lines are synchronized internally and do not have to be asserted or negated with reference to the CLKOUT. The timings I41, I42, and I43 are specified to allow the correct function of the  $\overline{\text{IRQ}}$  lines detection circuitry, and has no direct relation with the total system interrupt latency that the MPC862/857T/857DSL is able to support.

Figure 24 provides the interrupt detection timing for the external level-sensitive lines.



**Figure 24. Interrupt Detection Timing for External Level Sensitive Lines**

Figure 25 provides the interrupt detection timing for the external edge-sensitive lines.



**Figure 25. Interrupt Detection Timing for External Edge Sensitive Lines**

Figure 27 provides the PCMCIA access cycle timing for the external bus write.

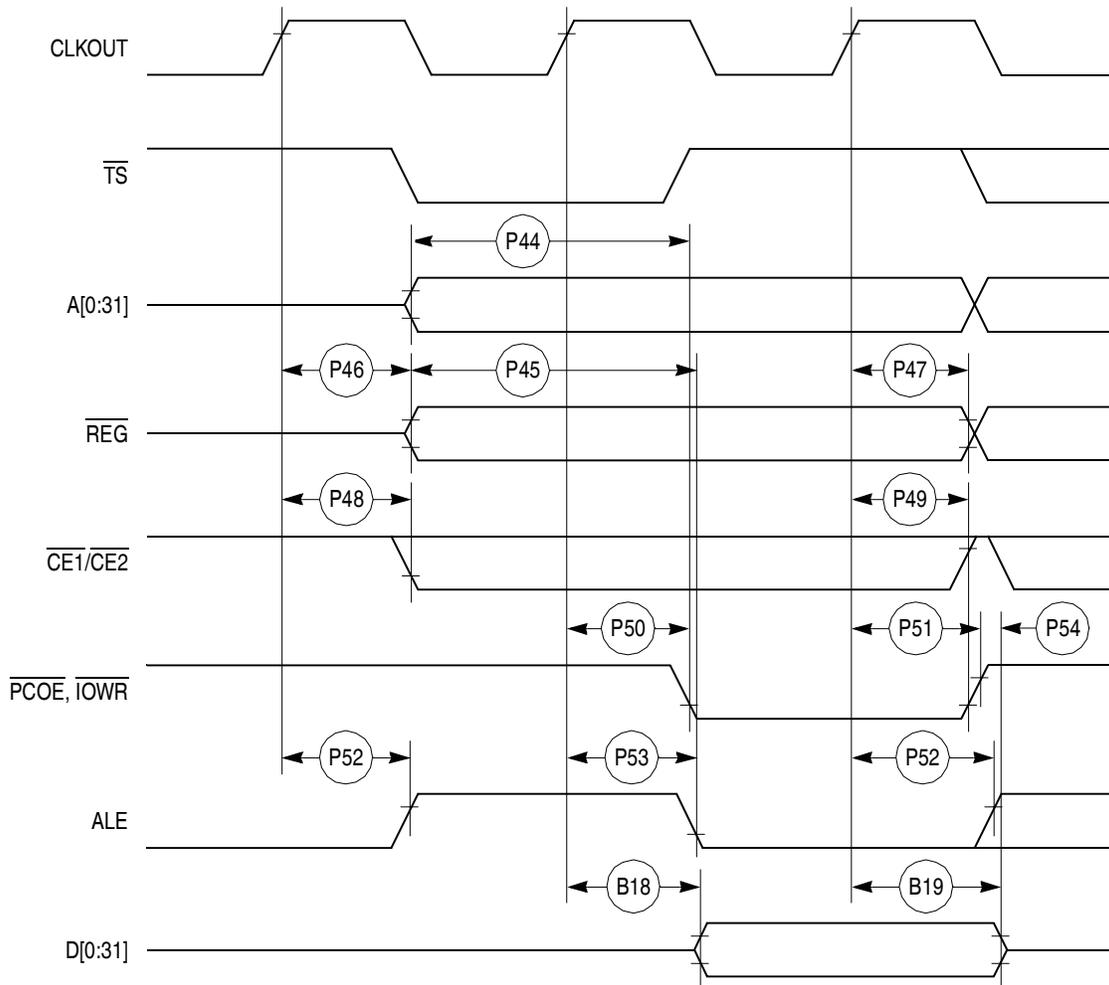


Figure 27. PCMCIA Access Cycles Timing External Bus Write

Figure 28 provides the PCMCIA  $\overline{WAIT}$  signals detection timing.

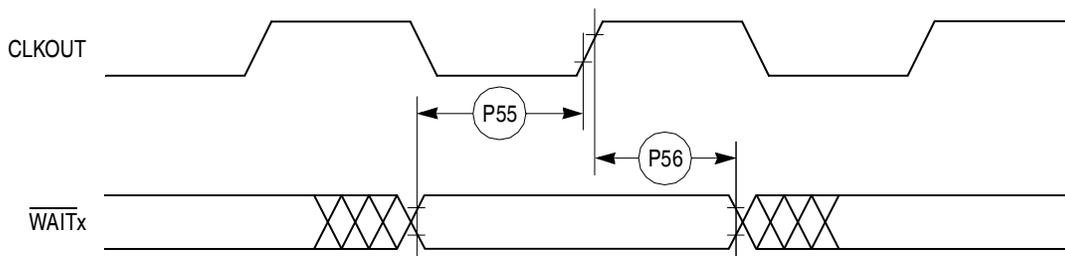


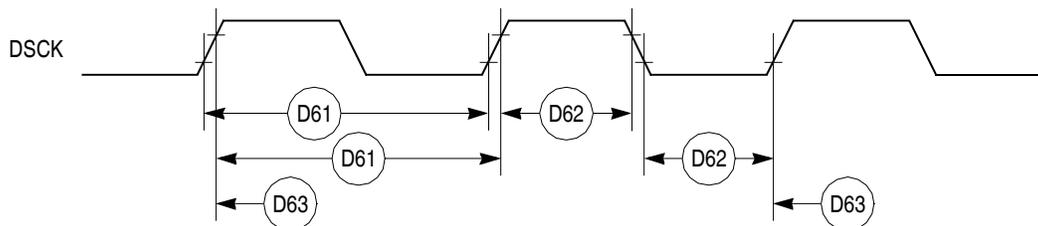
Figure 28. PCMCIA  $\overline{WAIT}$  Signals Detection Timing

Table 11 shows the debug port timing for the MPC862/857T/857DSL.

**Table 11. Debug Port Timing**

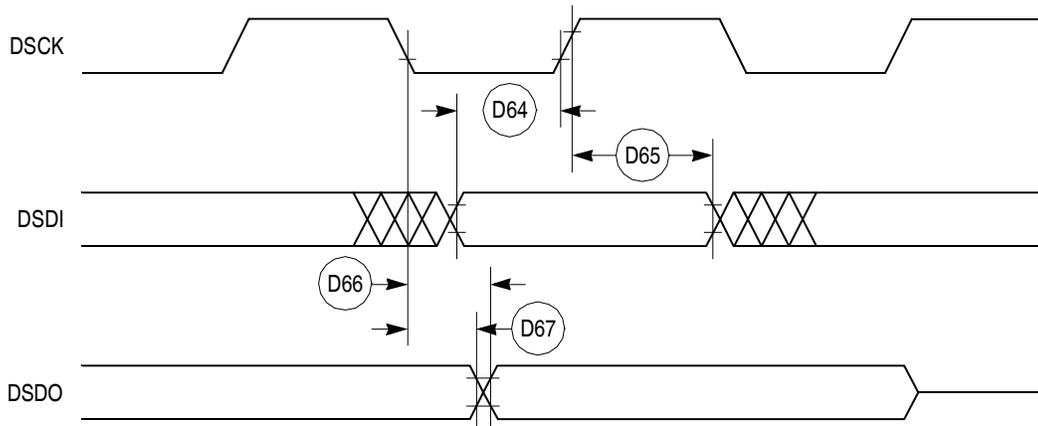
Num	Characteristic	All Frequencies		Unit
		Min	Max	
D61	DSCK cycle time	$3 \times T_{\text{CLOCKOUT}}$		-
D62	DSCK clock pulse width	$1.25 \times T_{\text{CLOCKOUT}}$		-
D63	DSCK rise and fall times	0.00	3.00	ns
D64	DSDI input data setup time	8.00		ns
D65	DSDI data hold time	5.00		ns
D66	DSCK low to DSDO data valid	0.00	15.00	ns
D67	DSCK low to DSDO invalid	0.00	2.00	ns

Figure 31 provides the input timing for the debug port clock.

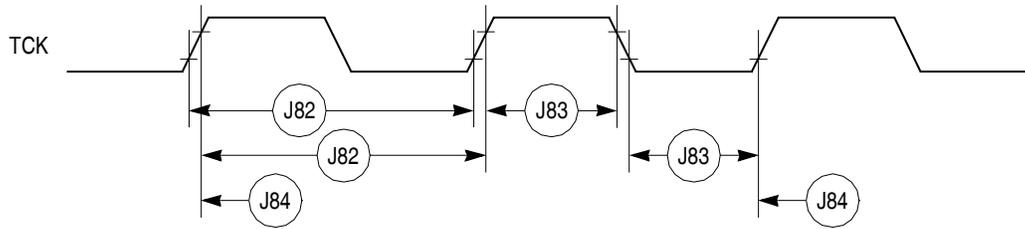
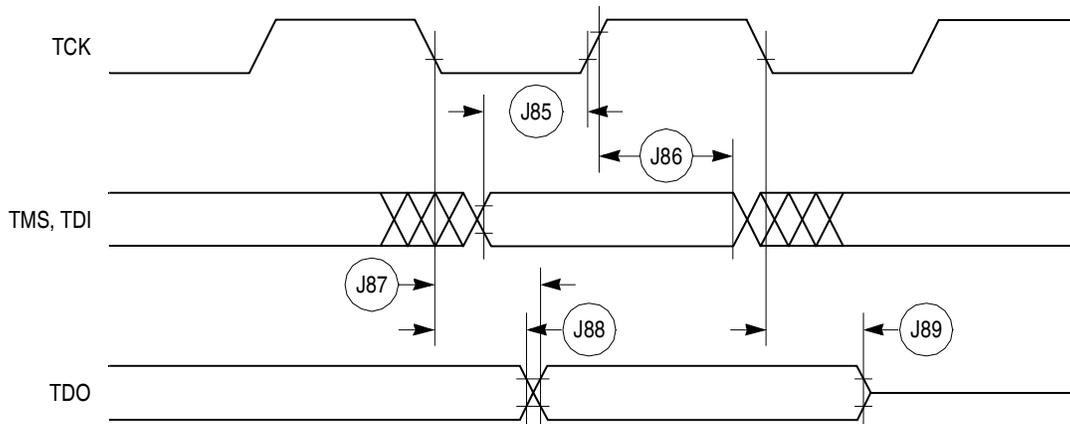
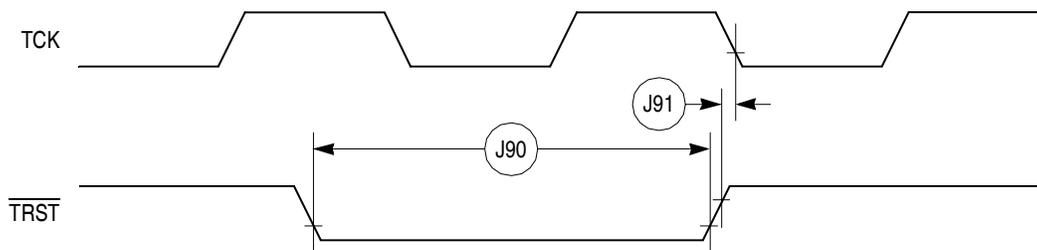
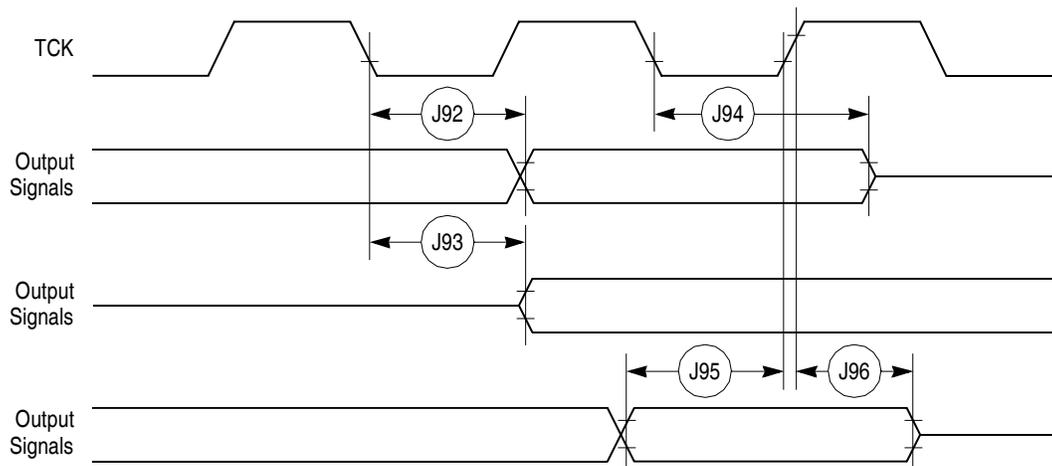


**Figure 31. Debug Port Clock Input Timing**

Figure 32 provides the timing for the debug port.

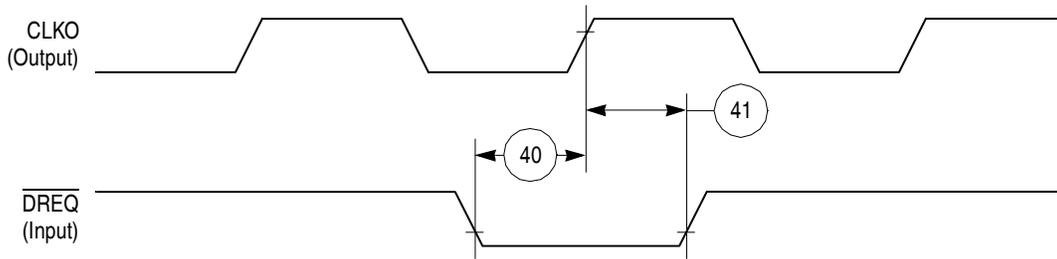


**Figure 32. Debug Port Timings**

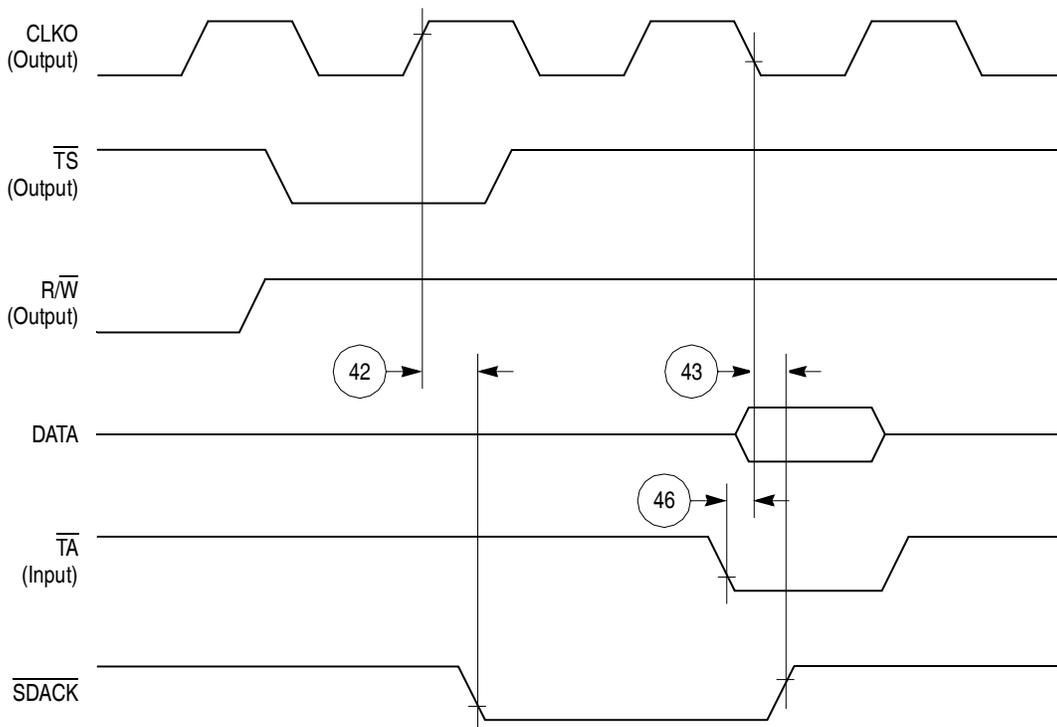

**Figure 36. JTAG Test Clock Input Timing**

**Figure 37. JTAG Test Access Port Timing Diagram**

**Figure 38. JTAG  $\overline{\text{TRST}}$  Timing Diagram**

**Figure 39. Boundary Scan (JTAG) Timing Diagram**

**Table 16. IDMA Controller Timing (continued)**

Num	Characteristic	All Frequencies		Unit
		Min	Max	
43	$\overline{\text{SDACK}}$ negation delay from clock low	—	12	ns
44	$\overline{\text{SDACK}}$ negation delay from $\overline{\text{TA}}$ low	—	20	ns
45	$\overline{\text{SDACK}}$ negation delay from clock high	—	15	ns
46	$\overline{\text{TA}}$ assertion to falling edge of the clock setup time (applies to external $\overline{\text{TA}}$ )	7	—	ns



**Figure 46. IDMA External Requests Timing Diagram**



**Figure 47.  $\overline{\text{SDACK}}$  Timing Diagram—Peripheral Write, Externally-Generated  $\overline{\text{TA}}$**

## 11.7 SCC in NMSI Mode Electrical Specifications

Table 20 provides the NMSI external clock timing.

**Table 20. NMSI External Clock Timing**

Num	Characteristic	All Frequencies		Unit
		Min	Max	
100	RCLK1 and TCLK1 width high <sup>1</sup>	1/SYNCCLK	—	ns
101	RCLK1 and TCLK1 width low	1/SYNCCLK +5	—	ns
102	RCLK1 and TCLK1 rise/fall time	—	15.00	ns
103	TXD1 active delay (from TCLK1 falling edge)	0.00	50.00	ns
104	$\overline{\text{RTS1}}$ active/inactive delay (from TCLK1 falling edge)	0.00	50.00	ns
105	$\overline{\text{CTS1}}$ setup time to TCLK1 rising edge	5.00	—	ns
106	RXD1 setup time to RCLK1 rising edge	5.00	—	ns
107	RXD1 hold time from RCLK1 rising edge <sup>2</sup>	5.00	—	ns
108	$\overline{\text{CD1}}$ setup Time to RCLK1 rising edge	5.00	—	ns

<sup>1</sup> The ratios SyncCLK/RCLK1 and SyncCLK/TCLK1 must be greater than or equal to 2.25/1.

<sup>2</sup> Also applies to  $\overline{\text{CD}}$  and  $\overline{\text{CTS}}$  hold time when they are used as an external sync signal.

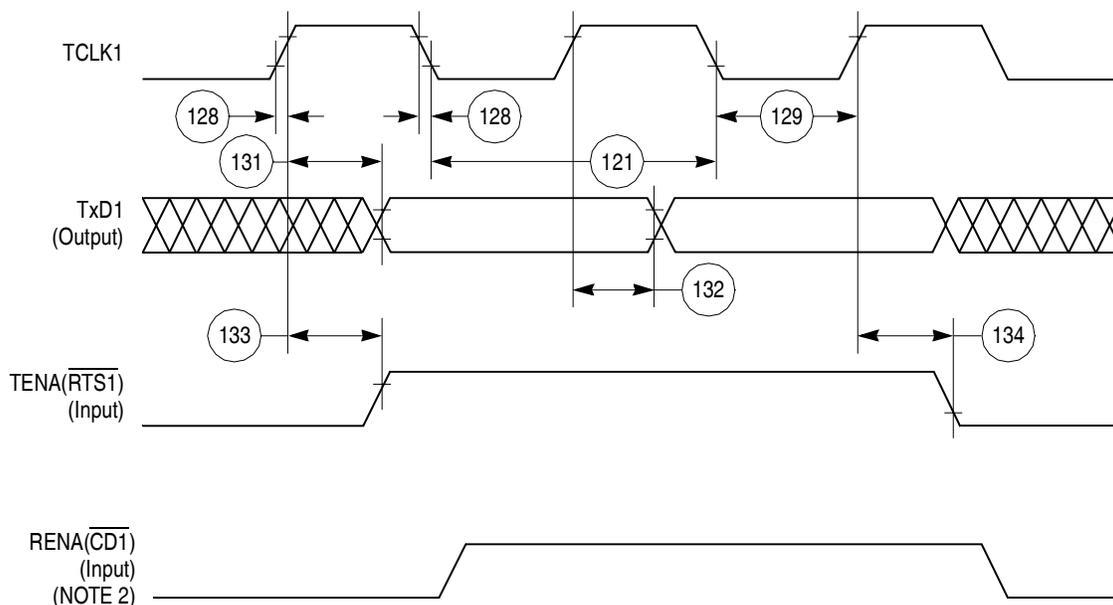
Table 21 provides the NMSI internal clock timing.

**Table 21. NMSI Internal Clock Timing**

Num	Characteristic	All Frequencies		Unit
		Min	Max	
100	RCLK1 and TCLK1 frequency <sup>1</sup>	0.00	SYNCCLK/3	MHz
102	RCLK1 and TCLK1 rise/fall time	—	—	ns
103	TXD1 active delay (from TCLK1 falling edge)	0.00	30.00	ns
104	$\overline{\text{RTS1}}$ active/inactive delay (from TCLK1 falling edge)	0.00	30.00	ns
105	$\overline{\text{CTS1}}$ setup time to TCLK1 rising edge	40.00	—	ns
106	RXD1 setup time to RCLK1 rising edge	40.00	—	ns
107	RXD1 hold time from RCLK1 rising edge <sup>2</sup>	0.00	—	ns
108	$\overline{\text{CD1}}$ setup time to RCLK1 rising edge	40.00	—	ns

<sup>1</sup> The ratios SyncCLK/RCLK1 and SyncCLK/TCLK1 must be greater or equal to 3/1.

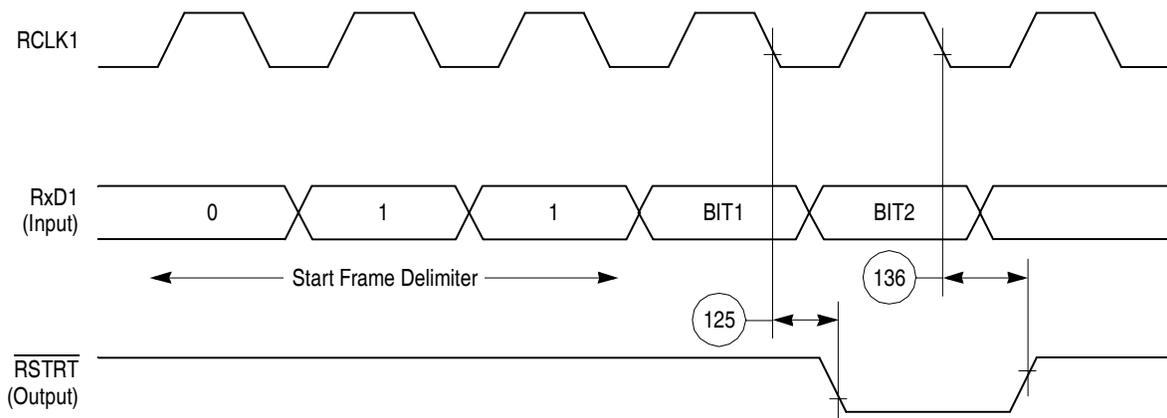
<sup>2</sup> Also applies to  $\overline{\text{CD}}$  and  $\overline{\text{CTS}}$  hold time when they are used as an external sync signals.



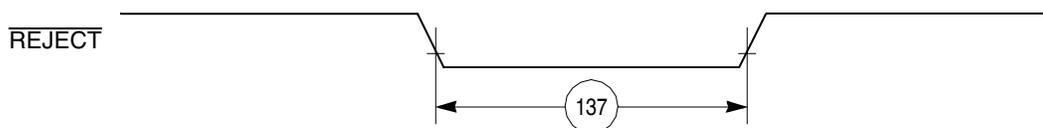
NOTES:

1. Transmit clock invert (TCI) bit in GSMR is set.
2. If RENA is deasserted before TENA, or RENA is not asserted at all during transmit, then the CSL bit is set in the buffer descriptor at the end of the frame transmission.

**Figure 62. Ethernet Transmit Timing Diagram**



**Figure 63. CAM Interface Receive Start Timing Diagram**



**Figure 64. CAM Interface REJECT Timing Diagram**

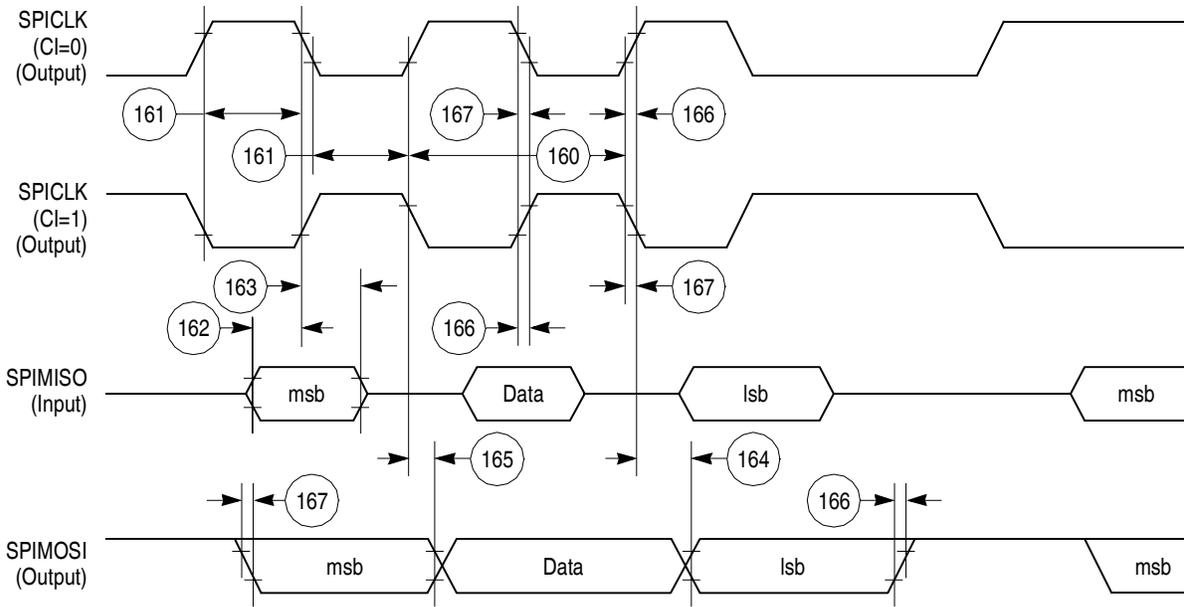


Figure 67. SPI Master (CP = 1) Timing Diagram

## 11.11 SPI Slave AC Electrical Specifications

Table 25 provides the SPI slave timings as shown in Figure 68 though Figure 69.

Table 25. SPI Slave Timing

Num	Characteristic	All Frequencies		Unit
		Min	Max	
170	Slave cycle time	2	—	$t_{cyc}$
171	Slave enable lead time	15	—	ns
172	Slave enable lag time	15	—	ns
173	Slave clock (SPICLK) high or low time	1	—	$t_{cyc}$
174	Slave sequential transfer delay (does not require deselect)	1	—	$t_{cyc}$
175	Slave data setup time (inputs)	20	—	ns
176	Slave data hold time (inputs)	20	—	ns
177	Slave access time	—	50	ns

Figure 71 shows signal timings during UTOPIA receive operations.

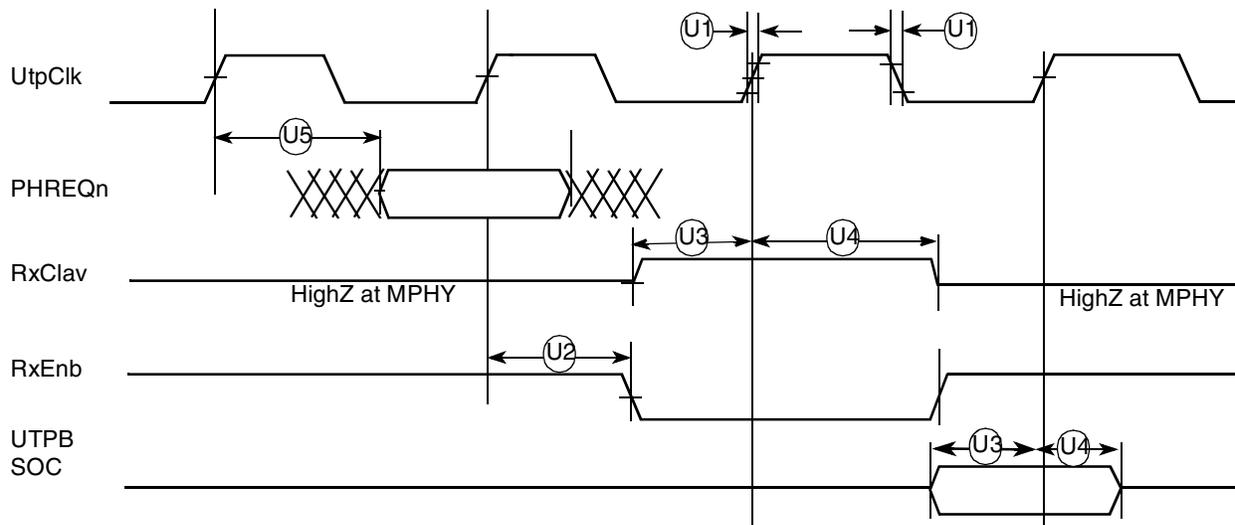


Figure 71. UTOPIA Receive Timing

Figure 72 shows signal timings during UTOPIA transmit operations.

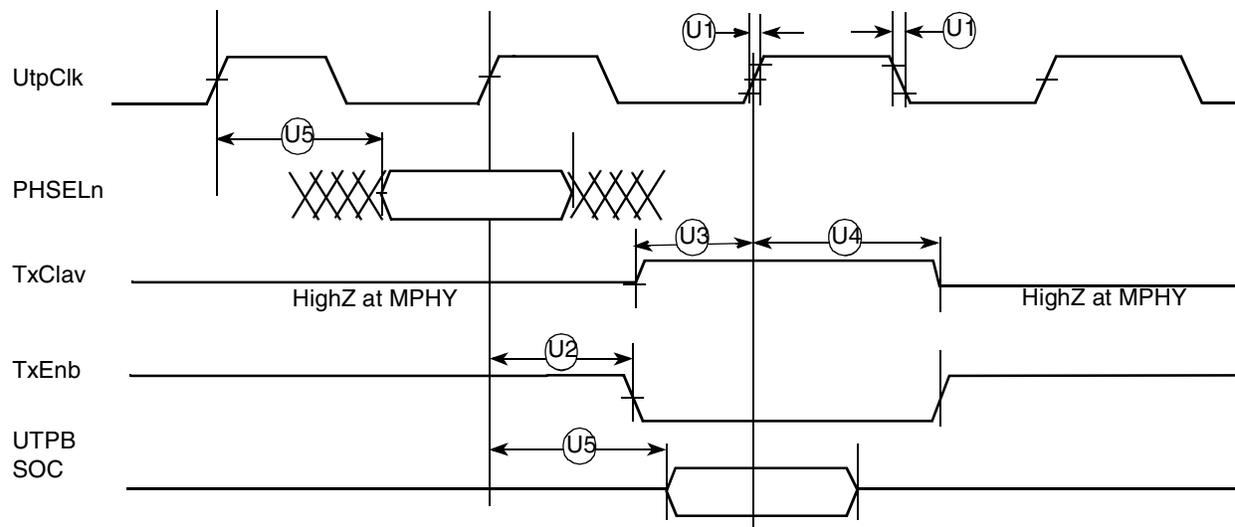


Figure 72. UTOPIA Transmit Timing

## 13 FEC Electrical Characteristics

This section provides the AC electrical specifications for the Fast Ethernet controller (FEC). Note that the timing specifications for the MII signals are independent of system clock frequency (part speed designation). Furthermore, MII signals use TTL signal levels compatible with devices operating at either 5.0 or 3.3 V.

### 13.1 MII Receive Signal Timing (MII\_RXD[3:0], MII\_RX\_DV, MII\_RX\_ER, MII\_RX\_CLK)

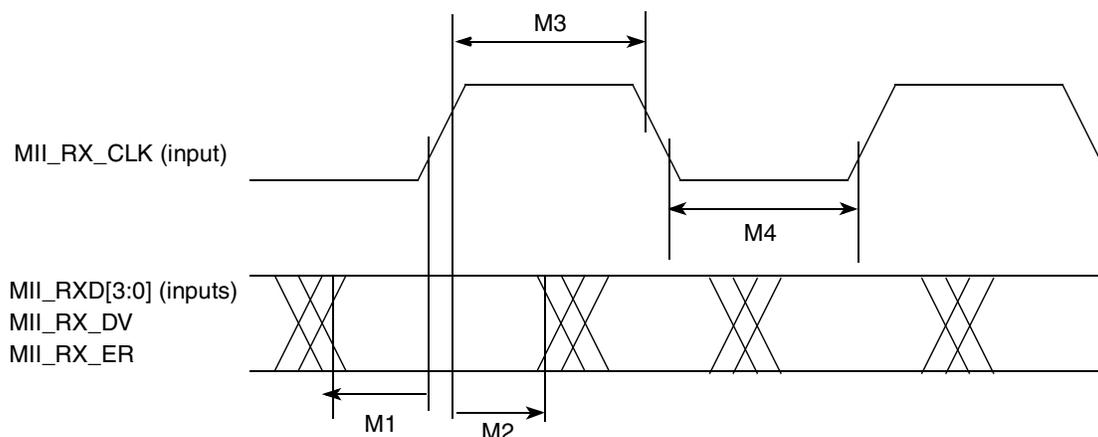
The receiver functions correctly up to a MII\_RX\_CLK maximum frequency of 25MHz +1%. There is no minimum frequency requirement. In addition, the processor clock frequency must exceed the MII\_RX\_CLK frequency - 1%.

Table 29 provides information on the MII receive signal timing.

**Table 29. MII Receive Signal Timing**

Num	Characteristic	Min	Max	Unit
M1	MII_RXD[3:0], MII_RX_DV, MII_RX_ER to MII_RX_CLK setup	5	—	ns
M2	MII_RX_CLK to MII_RXD[3:0], MII_RX_DV, MII_RX_ER hold	5	—	ns
M3	MII_RX_CLK pulse width high	35%	65%	MII_RX_CLK period
M4	MII_RX_CLK pulse width low	35%	65%	MII_RX_CLK period

Figure 73 shows MII receive signal timing.



**Figure 73. MII Receive Signal Timing Diagram**

### 13.2 MII Transmit Signal Timing (MII\_TXD[3:0], MII\_TX\_EN, MII\_TX\_ER, MII\_TX\_CLK)

The transmitter functions correctly up to a MII\_TX\_CLK maximum frequency of 25 MHz +1%. There is no minimum frequency requirement. In addition, the processor clock frequency must exceed the MII\_TX\_CLK frequency - 1%.

Table 30 provides information on the MII transmit signal timing.

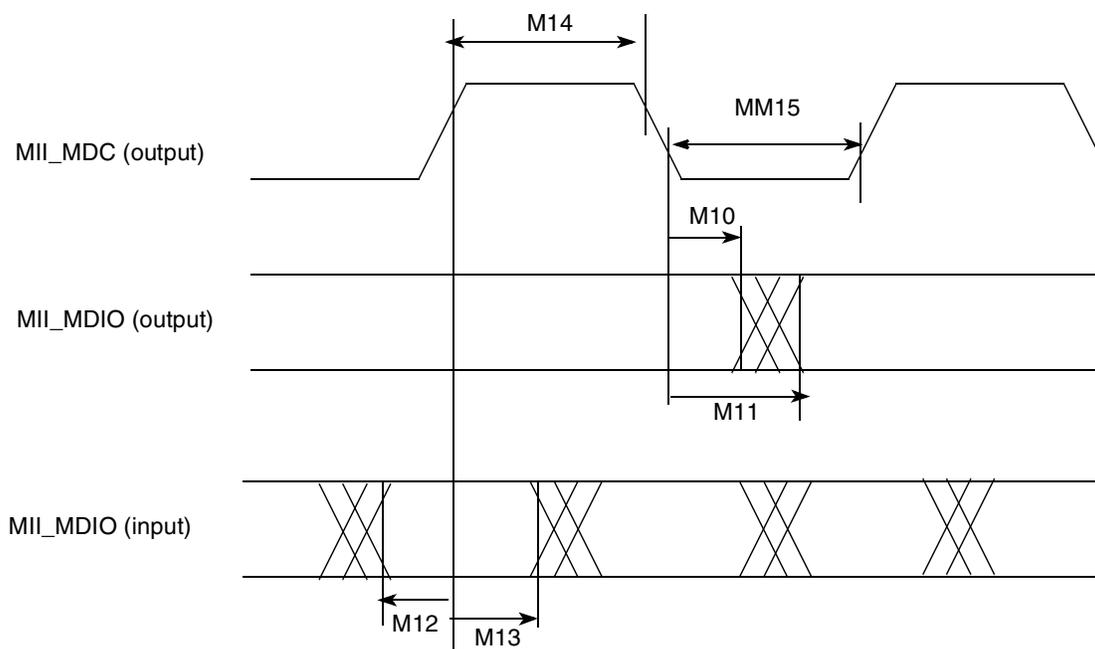
**Table 30. MII Transmit Signal Timing**

Num	Characteristic	Min	Max	Unit
M5	MII_TX_CLK to MII_TXD[3:0], MII_TX_EN, MII_TX_ER invalid	5	—	ns
M6	MII_TX_CLK to MII_TXD[3:0], MII_TX_EN, MII_TX_ER valid	—	25	

**Table 32. MII Serial Management Channel Timing**

Num	Characteristic	Min	Max	Unit
M10	MII_MDC falling edge to MII_MDIO output invalid (minimum propagation delay)	0	—	ns
M11	MII_MDC falling edge to MII_MDIO output valid (max prop delay)	—	25	ns
M12	MII_MDIO (input) to MII_MDC rising edge setup	10	—	ns
M13	MII_MDIO (input) to MII_MDC rising edge hold	0	—	ns
M14	MII_MDC pulse width high	40%	60%	MII_MDC period
M15	MII_MDC pulse width low	40%	60%	MII_MDC period

Figure 76 shows the MII serial management channel timing diagram.


**Figure 76. MII Serial Management Channel Timing Diagram**

## 14 Mechanical Data and Ordering Information

Table 33 provides information on the MPC862/857T/857DSL derivative devices.

**Table 33. MPC862/857T/857DSL Derivatives**

Device	Number of SCCs <sup>1</sup>	Ethernet Support	Multi-Channel HDLC Support	ATM Support	Cache Size	
					Instruction	Data
MPC862T	Four	10/100 Mbps	Yes	Yes	4 Kbytes	4 Kbytes
MPC862P	Four	10/100 Mbps	Yes	Yes	16 Kbytes	8 Kbytes

Table 35 contains a list of the MPC862 input and output signals and shows multiplexing and pin assignments.

**Table 35. Pin Assignments**

Name	Pin Number	Type
A[0:31]	B19, B18, A18, C16, B17, A17, B16, A16, D15, C15, B15, A15, C14, B14, A14, D12, C13, B13, D9, D11, C12, B12, B10, B11, C11, D10, C10, A13, A10, A12, A11, A9	Bidirectional Three-state
TSIZ0 $\overline{\text{REG}}$	B9	Bidirectional Three-state
TSIZ1	C9	Bidirectional Three-state
$\overline{\text{RD}}/\overline{\text{WR}}$	B2	Bidirectional Three-state
$\overline{\text{BURST}}$	F1	Bidirectional Three-state
$\overline{\text{BDIP}}$ $\overline{\text{GPL\_B5}}$	D2	Output
$\overline{\text{TS}}$	F3	Bidirectional Active Pull-up
$\overline{\text{TA}}$	C2	Bidirectional Active Pull-up
$\overline{\text{TEA}}$	D1	Open-drain
$\overline{\text{BI}}$	E3	Bidirectional Active Pull-up
$\overline{\text{IRQ2}}$ RSV	H3	Bidirectional Three-state
$\overline{\text{IRQ4}}$ $\overline{\text{KR}}$ $\overline{\text{RETRY}}$ SPKROUT	K1	Bidirectional Three-state
$\overline{\text{CR}}$ $\overline{\text{IRQ3}}$	F2	Input
D[0:31]	W14, W12, W11, W10, W13, W9, W7, W6, U13, T11, V11, U11, T13, V13, V10, T10, U10, T12, V9, U9, V8, U8, T9, U12, V7, T8, U7, V12, V6, W5, U6, T7	Bidirectional Three-state
DP0 $\overline{\text{IRQ3}}$	V3	Bidirectional Three-state
DP1 $\overline{\text{IRQ4}}$	V5	Bidirectional Three-state
DP2 $\overline{\text{IRQ5}}$	W4	Bidirectional Three-state
DP3 $\overline{\text{IRQ6}}$	V4	Bidirectional Three-state

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