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NXP USA Inc. - P87LPC764BD,512 Datasheet



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Details

Product Status	Obsolete
Core Processor	8051
Core Size	8-Bit
Speed	20MHz
Connectivity	I ² C, UART/USART
Peripherals	Brown-out Detect/Reset, LED, POR, WDT
Number of I/O	18
Program Memory Size	4KB (4K x 8)
Program Memory Type	OTP
EEPROM Size	-
RAM Size	128 x 8
Voltage - Supply (Vcc/Vdd)	2.7V ~ 6V
Data Converters	-
Oscillator Type	Internal
Operating Temperature	0°C ~ 70°C (TA)
Mounting Type	Surface Mount
Package / Case	20-SOIC (0.295", 7.50mm Width)
Supplier Device Package	20-50
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/p87lpc764bd-512

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BLOCK DIAGRAM



P87LPC764

PIN DESCRIPTIONS

MNEMONIC	PIN NO.	TYPE		NAME AND FUNCTION					
P0.0–P0.7	1, 13, 14, 16–20	I/O	Port 0 : Port 0 is an 8-bit I/O port with a user-configurable output type. Port 0 latches are configured in the quasi-bidirectional mode and have either ones or zeros written to them during reset, as determined by the PRHI bit in the UCFG1 configuration byte. The operation of port 0 pins as inputs and outputs depends upon the port configuration selected. Each port pin is configured independently. Refer to the section on I/O port configuration and the DC Electrical Characteristics for details.						
			Port 0 also p	rovides vario	us special functions as described below.				
	1	0	P0.0	CMP2	Comparator 2 output.				
	20	I	P0.1	CIN2B	Comparator 2 positive input B.				
	19	I	P0.2	CIN2A	Comparator 2 positive input A.				
	18	I	P0.3	CIN1B	Comparator 1 positive input B.				
	17	I	P0.4	CIN1A	Comparator 1 positive input A.				
	16	I	P0.5	CMPREF	Comparator reference (negative) input.				
	14	0	P0.6	CMP1	Comparator 1 output.				
	13	I/O	P0.7	T1	Timer/counter 1 external count input or overflow output.				
P1.0–P1.7	2–4, 8–12	I/O	Port 1 : Port 1 is an 8-bit I/O port with a user-configurable output type, except for three pins as not below. Port 1 latches are configured in the quasi-bidirectional mode and have either ones or zeros written to them during reset, as determined by the PRHI bit in the UCFG1 configuration byte. The operation of the configurable port 1 pins as inputs and outputs depends upon the port configuration selected. Each of the configurable port pins are programmed independently. Refer to the section of port configuration and the DC Electrical Characteristics for details.						
			Port 1 also p	Port 1 also provides various special functions as described below.					
	12	0	P1.0	TxD	Transmitter output for the serial port.				
	11	I	P1.1	RxD	Receiver input for the serial port.				
	10	I/O	P1.2	T0	Timer/counter 0 external count input or overflow output.				
		1/0		SCL	drain, in order to conform to I ² C specifications.				
	9	I	P1.3	INT0	External interrupt 0 input.				
		I/O		SDA	I ² C serial data input/output. When configured as an output, P1.3 is open drain, in order to conform to I ² C specifications.				
	8	I	P1.4	INT1	External interrupt 1 input.				
	4	I	P1.5	RST	External Reset input (if selected via EPROM configuration). A low on this pin resets the microcontroller, causing I/O ports and peripherals to take on their default states, and the processor begins execution at address 0. When used as a port pin, P1.5 is a Schmitt trigger input only.				
P2.0-P2.1	6, 7	I/O	Port 2 : Port 2 is a 2-bit I/O port with a user-configurable output type. Port 2 latches are configured in th quasi-bidirectional mode and have either ones or zeros written to them during reset, as determined by the PRHI bit in the UCFG1 configuration byte. The operation of port 2 pins as inputs and outputs depends upon the port configuration selected. Each port pin is configured independently. Refer to the section on I/O port configuration and the DC Electrical Characteristics for details.						
			Port 2 also p	t 2 also provides various special functions as described below.					
	7	0	P2.0	X2	Output from the oscillator amplifier (when a crystal oscillator option is selected via the EPROM configuration).				
				CLKOUT CPU clock divided by 6 clock output when enabled via SFR bit a conjunction with internal RC oscillator or external clock input.					
	6		P2.1	X1	Input to the oscillator circuit and internal clock generator circuits (when selected via the EPROM configuration).				
V _{SS}	5	I	Ground: 0V	reference.					
V _{DD}	15	I	Power Supp Power Down	ly: This is the modes.	e power supply voltage for normal operation as well as Idle and				

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Internal Reference Voltage

An internal reference voltage generator may supply a default reference when a single comparator input pin is used. The value of the internal reference voltage, referred to as V_{ref} , is 1.28 V ±10%.

Comparator Interrupt

Each comparator has an interrupt flag CMFn contained in its configuration register. This flag is set whenever the comparator output changes state. The flag may be polled by software or may be used to generate an interrupt. The interrupt will be generated when the corresponding enable bit ECn in the IEN1 register is set and the interrupt system is enabled via the EA bit in the IEN0 register.

Comparators and Power Reduction Modes

Either or both comparators may remain enabled when Power Down or Idle mode is activated. The comparators will continue to function in the power reduction mode. If a comparator interrupt is enabled, a change of the comparator output state will generate an interrupt and wake up the processor. If the comparator output to a pin is enabled, the pin should be configured in the push-pull mode in order to obtain fast switching times while in power down mode. The reason is that with the oscillator stopped, the temporary strong pull-up that normally occurs during switching on a quasi-bidirectional port pin does not take place.

Comparators consume power in Power Down and Idle modes, as well as in the normal operating mode. This fact should be taken into account when system power consumption is an issue.

Comparator Configuration Example

The code shown in Figure 5 is an example of initializing one comparator. Comparator 1 is configured to use the CIN1A and CMPREF inputs, outputs the comparator result to the CMP1 pin, and generates an interrupt when the comparator output changes.

The interrupt routine used for the comparator must clear the interrupt flag (CMF1 in this case) before returning.

CmpInit:		
mov	PT0AD,#30h	; Disable digital inputs on pins that are used
		; for analog functions: CIN1A, CMPREF.
anl	POM2,#0cfh	; Disable digital outputs on pins that are used
orl	P0M1,#30h	; for analog functions: CIN1A, CMPREF.
mov	CMP1,#24h	; Turn on comparator 1 and set up for:
		; - Positive input on CIN1A.
		; - Negative input from CMPREF pin.
		; - Output to CMP1 pin enabled.
call	delay10us	; The comparator has to start up for at
		; least 10 microseconds before use.
anl	CMP1,#0feh	; Clear comparator 1 interrupt flag.
setb	EC1	; Enable the comparator 1 interrupt. The
		; priority is left at the current value.
setb	EA	; Enable the interrupt system (if needed).
ret		; Return to caller.
		SU01189

Figure 5.

I²C Serial Interface

The I^2C bus uses two wires (SDA and SCL) to transfer information between devices connected to the bus. The main features of the bus are:

- Bidirectional data transfer between masters and slaves.
- Serial addressing of slaves (no added wiring).
- Acknowledgment after each transferred byte.
- Multimaster bus.
- Arbitration between simultaneously transmitting masters without corruption of serial data on bus.

The l^2C subsystem includes hardware to simplify the software required to drive the l^2C bus. The hardware is a single bit interface which in addition to including the necessary arbitration and framing error checks, includes clock stretching and a bus timeout timer. The interface is synchronized to software either through polled loops or interrupts.

Refer to the application note AN422, entitled "Using the 8XC751 Microcontroller as an I²C Bus Master" for additional discussion of the 8xC76x I²C interface and sample driver routines.

The P87LPC764 $I^{2}C$ implementation duplicates that of the 87C751 and 87C752 except for the following details:

- The interrupt vector addresses for both the I²C interrupt and the Timer I interrupt.
- The I²C SFR addresses (I2CON, I2CFG, I2DAT).
- The location of the I²C interrupt enable bit and the name of the SFR it is located within (EI2 is Bit 0 in IEN1).
- The location of the Timer I interrupt enable bit and the name of the SFR it is located within (ETI is Bit 7 in IEN1).
- The I²C and Timer I interrupts have a settable priority.

Timer I is used to both control the timing of the I^2C bus and also to detect a "bus locked" condition, by causing an interrupt when nothing happens on the I^2C bus for an inordinately long period of time while a transmission is in progress. If this interrupt occurs, the program has the opportunity to attempt to correct the fault and resume I^2C operation.

Six time spans are important in I²C operation and are insured by timer I:

- The MINIMUM HIGH time for SCL when this device is the master.
- The MINIMUM LOW time for SCL when this device is a master. This is not very important for a single-bit hardware interface like this one, because the SCL low time is stretched until the software responds to the I²C flags. The software response time normally meets or exceeds the MIN LO time. In cases where the software responds within MIN HI + MIN LO) time, timer I will ensure that the minimum time is met.
- The MINIMUM SCL HIGH TO SDA HIGH time in a stop condition.
- The MINIMUM SDA HIGH TO SDA LOW time between I²C stop and start conditions (4.7ms, see I²C specification).
- The MINIMUM SDA LOW TO SCL LOW time in a start condition.
- The MAXIMUM SCL CHANGE time while an I²C frame is in progress. A frame is in progress between a start condition and the following stop condition. This time span serves to detect a lack of software response on this device as well as external I²C

problems. SCL "stuck low" indicates a faulty master or slave. SCL "stuck high" may mean a faulty device, or that noise induced onto the I²C bus caused all masters to withdraw from I²C arbitration.

The first five of these times are 4.7 ms (see I^2C specification) and are covered by the low order three bits of timer I. Timer I is clocked by the P87LPC764 CPU clock. Timer I can be pre-loaded with one of four values to optimize timing for different oscillator frequencies. At lower frequencies, software response time is increased and will degrade maximum performance of the I^2C bus. See special function register I2CFG description for prescale values (CT0, CT1).

The MAXIMUM SCL CHANGE time is important, but its exact span is not critical. The complete 10 bits of timer I are used to count out the maximum time. When I²C operation is enabled, this counter is cleared by transitions on the SCL pin. The timer does not run between I²C frames (i.e., whenever reset or stop occurred more recently than the last start). When this counter is running, it will carry out after 1020 to 1023 machine cycles have elapsed since a change on SCL. A carry out causes a hardware reset of the I²C interface and generates an interrupt if the Timer I interrupt is enabled. In cases where the bus hang-up is due to a lack of software response by this device, the reset releases SCL and allows I²C operation among other devices to continue.

Timer I is enabled to run, and will reset the I²C interface upon overflow, if the TIRUN bit in the I2CFG register is set. The Timer I interrupt may be enabled via the ETI bit in IEN1, and its priority set by the PTIH and PTI bits in the IP1H and IP1 registers respectively.

I²C Interrupts

If I²C interrupts are enabled (EA and EI2 are both set to 1), an I²C interrupt will occur whenever the ATN flag is set by a start, stop, arbitration loss, or data ready condition (refer to the description of ATN following). In practice, it is not efficient to operate the I²C interface in this fashion because the I²C interrupt service routine would somehow have to distinguish between hundreds of possible conditions. Also, since I²C can operate at a fairly high rate, the software may execute faster if the code simply waits for the I²C interface.

Typically, the l^2C interrupt should only be used to indicate a start condition at an idle slave device, or a stop condition at an idle master device (if it is waiting to use the l^2C bus). This is accomplished by enabling the l^2C interrupt only during the aforementioned conditions.

Reading I2CON

- RDAT The data from SDA is captured into "Receive DATa" whenever a rising edge occurs on SCL. RDAT is also available (with seven low-order zeros) in the I2DAT register. The difference between reading it here and there is that reading I2DAT clears DRDY, allowing the I²C to proceed on to another bit. Typically, the first seven bits of a received byte are read from I2DAT, while the 8th is read here. Then I2DAT can be written to send the Acknowledge bit and clear DRDY.
- ATN "ATteNtion" is 1 when one or more of DRDY, ARL, STR, or STP is 1. Thus, ATN comprises a single bit that can be tested to release the I²C service routine from a "wait loop."
- DRDY "Data ReaDY" (and thus ATN) is set when a rising edge occurs on SCL, except at idle slave. DRDY is cleared by writing CDR = 1, or by writing or reading the I2DAT register. The following low period on SCL is stretched until the program responds by clearing DRDY.

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I2CON	Addres	s: D8h									Reset Value: 81h
	Bit Add	Iressable ¹	7	6	5	4	3	2	1	0	
		READ	RDAT	ATN	DRDY	ARL	STR	STP	MASTER		
		WRITE	CXA	IDLE	CDR	CARL	CSTR	CSTP	XSTR	XSTP	
Bľ	т	SYMBOL	FUN	CTION							
120	CON.7	RDAT	Read	: the mos	t recently	received of	data bit.				
	"	CXA	Write	: clears th	ne transmi	t active fla	ıg.				
120	CON.6	ATN	Read	: ATN = 1	if any of t	he flags D	DRDY, ARI	L, STR, o	r STP = 1.		
	"	IDLE	Write	Write: in the I ² C slave mode, writing a 1 to this bit causes the I ² C hardware to ignore the bus until it is needed again.							
120	CON.5	DRDY	Read	: Data Re	ady flag,	set when t	there is a i	rising edg	e on SCL.		
	"	CDR	Write	: writing a	1 to this I	oit clears t	he DRDY	flag.			
120	CON.4	ARL	Read	: Arbitrati	on Loss fla	ag, set wh	en arbitra	tion is los	t while in the	e transmit	mode.
	"	CARL	Write	: writing a	1 to this I	oit clears t	he CARL	flag.			
120	CON.3	STR	Read	: Start fla	g, set whe	en a start o	condition is	s detected	l at a maste	r or non-ie	dle slave.
	"	CSTR	Write	: writing a	1 to this I	oit clears t	he STR fla	ag.			
120	CON.2	STP	Read	: Stop flag	g, set whe	n a stop c	ondition is	s detected	l at a maste	r or non-io	dle slave.
	"	CSTP	Write	: writing a	1 to this I	oit clears t	he STP fla	ag.			
120	CON.1	MASTER	Read	: indicate	s whether	this devic	e is curre	ntly as bu	s master.		
	"	XSTR	Write	: writing a	1 to this I	oit causes	a repeate	ed start co	ndition to be	e generate	ed.
120	CON.0	_	Read	: undefine	ed.						
	"	XSTP	Write	: writing a	1 to this I	oit causes	a stop co	ndition to	be generate	ed.	SU01155

Figure 6. I²C Control Register (I2CON)

I2DAT	Addres Not Bit	s: D9h Addressab	ole							
			7	6	5	4	3	2	1	0
		READ	RDAT	_	_	_	_	—	_	_
		WRITE	XDAT	_		_		_	_	_
F	BIT	SYMBOL	. FUN	CTION						
Ľ	2DAT.7	RDAT	Read I2DA	Read: the most recently received data bit, captured from SDA at every rising edge of SCL. Reading 2DAT also clears DRDY and the Transmit Active state.						
	"	XDAT	Write Trans	Vrite: sets the data for the next transmitted bit. Writing I2DAT also clears DRDY and sets the ransmit Active state.						
ľ	2DAT.6-0	-	Unus	ed.						

Figure 7. I²C Data Register (I2DAT)

Checking ATN and DRDY

When a program detects ATN = 1, it should next check DRDY. If DRDY = 1, then if it receives the last bit, it should capture the data from RDAT (in I2DAT or I2CON). Next, if the next bit is to be sent, it should be written to I2DAT. One way or another, it should clear DRDY and then return to monitoring ATN. Note that if any of ARL,

STR, or STP is set, clearing DRDY will not release SCL to high, so that the I^2C will not go on to the next bit. If a program detects ATN = 1, and DRDY = 0, it should go on to examine ARL, STR, and STP.

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Table 1. Interaction of TIRUN with SLAVEN, MASTRQ, and MASTER

SLAVEN, MASTRQ, MASTER	TIRUN	OPERATING MODE
All 0	0	The I ² C interface is disabled. Timer I is cleared and does not run. This is the state assumed after a reset. If an I ² C application wants to ignore the I ² C at certain times, it should write SLAVEN, MASTRQ, and TIRUN all to zero.
All 0	1	The I ² C interface is disabled.
Any or all 1	0	The I ² C interface is enabled. The 3 low-order bits of Timer I run for min-time generation, but the hi-order bits do not, so that there is no checking for I ² C being "hung." This configuration can be used for very slow I ² C operation.
Any or all 1	1	The I ² C interface is enabled. Timer I runs during frames on the I ² C, and is cleared by transitions on SCL, and by Start and Stop conditions. This is the normal state for I ² C operation.

Table 2. CT1, CT0 Values

СТ1, СТ0	Min Time Count (Machine Cycles)	CPU Clock Max (for 100 kHz I ² C)	Timeout Period (Machine Cycles)		
1 0	7	8.4 MHz	1023		
0 1	6	7.2 MHz	1022		
0 0	5	6.0 MHz	1021		
1 1	4	4.8 MHz	1020		

Interrupts

The P87LPC764 uses a four priority level interrupt structure. This allows great flexibility in controlling the handling of the P87LPC764's many interrupt sources. The P87LPC764 supports up to 12 interrupt sources.

Each interrupt source can be individually enabled or disabled by setting or clearing a bit in registers IEN0 or IEN1. The IEN0 register also contains a global disable bit, EA, which disables all interrupts at once.

Each interrupt source can be individually programmed to one of four priority levels by setting or clearing bits in the IP0, IP0H, IP1, and IP1H registers. An interrupt service routine in progress can be

Table 3. Summary of Interrupts

interrupted by a higher priority interrupt, but not by another interrupt of the same or lower priority. The highest priority interrupt service cannot be interrupted by any other interrupt source. So, if two requests of different priority levels are received simultaneously, the request of higher priority level is serviced.

If requests of the same priority level are received simultaneously, an internal polling sequence determines which request is serviced. This is called the arbitration ranking. Note that the arbitration ranking is only used to resolve simultaneous requests of the same priority level.

Table 3 summarizes the interrupt sources, flag bits, vector addresses, enable bits, priority bits, arbitration ranking, and whether each interrupt may wake up the CPU from Power Down mode.

Description	Interrupt Flag Bit(s)	Vector Address	Interrupt Enable Bit(s)	Interrupt Priority	Arbitration Ranking	Power Down Wakeup
External Interrupt 0	IE0	0003h	EX0 (IEN0.0)	IP0H.0, IP0.0	1 (highest)	Yes
Timer 0 Interrupt	TF0	000Bh	ET0 (IEN0.1)	IP0H.1, IP0.1	4	No
External Interrupt 1	IE1	0013h	EX1 (IEN0.2)	IP0H.2, IP0.2	6	Yes
Timer 1 Interrupt	TF1	001Bh	ET1 (IEN0.3)	IP0H.3, IP0.3	9	No
Serial Port Tx and Rx	TI & RI	0023h	ES (IEN0.4)	IP0H.4, IP0.4	11	No
Brownout Detect	BOF	002Bh	EBO (IEN0.5)	IP0H.5, IP0.5	2	Yes
I ² C Interrupt	ATN	0033h	EI2 (IEN1.0)	IP1H.0, IP1.0	5	No
KBI Interrupt	KBF	003Bh	EKB (IEN1.1)	IP1H.1, IP1.1	7	Yes
Comparator 2 interrupt	CMF2	0043h	EC2 (IEN1.2)	IP1H.2, IP1.2	10	Yes
Watchdog Timer	WDOVF	0053h	EWD (IEN0.6)	IP0H.6, IP0.6	3	Yes
Comparator 1 interrupt	CMF1	0063h	EC1 (IEN1.5)	IP1H.5, IP1.5	8	Yes
Timer I interrupt	_	0073h	ETI (IEN1.7)	IP1H.7, IP1.7	12 (lowest)	No

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I/O Ports

The P87LPC764 has 3 I/O ports, port 0, port 1, and port 2. The exact number of I/O pins available depend upon the oscillator and reset options chosen. At least 15 pins of the P87LPC764 may be used as I/Os when a two-pin external oscillator and an external reset circuit are used. Up to 18 pins may be available if fully on-chip oscillator and reset configurations are chosen.

All but three I/O port pins on the P87LPC764 may be software configured to one of four types on a bit-by-bit basis, as shown in Table 4. These are: quasi-bidirectional (standard 80C51 port outputs), push-pull, open drain, and input only. Two configuration registers for each port choose the output type for each port pin.

Table 4. Port Output Configuration Settings

PxM1.y	PxM2.y	Port Output Mode					
0	0	Quasi-bidirectional					
0	1	Push-Pull					
1	0	Input Only (High Impedance)					
1	1	Open Drain					

Quasi-Bidirectional Output Configuration

The default port output configuration for standard P87LPC764 I/O ports is the quasi-bidirectional output that is common on the 80C51 and most of its derivatives. This output type can be used as both an

input and output without the need to reconfigure the port. This is possible because when the port outputs a logic high, it is weakly driven, allowing an external device to pull the pin low. When the pin is pulled low, it is driven strongly and able to sink a fairly large current. These features are somewhat similar to an open drain output except that there are three pull-up transistors in the quasi-bidirectional output that serve different purposes.

One of these pull-ups, called the "very weak" pull-up, is turned on whenever the port latch for the pin contains a logic 1. The very weak pull-up sources a very small current that will pull the pin high if it is left floating.

A second pull-up, called the "weak" pull-up, is turned on when the port latch for the pin contains a logic 1 and the pin itself is also at a logic 1 level. This pull-up provides the primary source current for a quasi-bidirectional pin that is outputting a 1. If a pin that has a logic 1 on it is pulled low by an external device, the weak pull-up turns off, and only the very weak pull-up remains on. In order to pull the pin low under these conditions, the external device has to sink enough current to overpower the weak pull-up and take the voltage on the port pin below its input threshold.

The third pull-up is referred to as the "strong" pull-up. This pull-up is used to speed up low-to-high transitions on a quasi-bidirectional port pin when the port latch changes from a logic 0 to a logic 1. When this occurs, the strong pull-up turns on for a brief time, two CPU clocks, in order to pull the port pin high quickly. Then it turns off again.

The quasi-bidirectional port configuration is shown in Figure 10.



Figure 10. Quasi-Bidirectional Output

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Oscillator

The P87LPC764 provides several user selectable oscillator options, allowing optimization for a range of needs from high precision to lowest possible cost. These are configured when the EPROM is

programmed. Basic oscillator types that are supported include: low, medium, and high speed crystals, covering a range from 20 kHz to 20 MHz; ceramic resonators; and on-chip RC oscillator.

Low Frequency Oscillator Option

This option supports an external crystal in the range of 20 kHz to 100 kHz.

Table 5 shows capacitor values that may be used with a quartz crystal in this mode.

Table 5. Recommended oscillator capacitors for use with the low frequency oscillator option

Oscillator		V_{DD} = 2.7 to 4.5 V		V _{DD} = 4.5 to 6.0 V			
Frequency	Lower Limit	Optimal Value	Upper Limit	Lower Limit	Optimal Value	Upper Limit	
20 kHz	15 pF	15 pF	33 pF	33 pF	33 pF	47 pF	
32 kHz	15 pF	15 pF	33 pF	33 pF	33 pF	47 pF	
100 kHz	15 pF	15 pF	33 pF	15 pF	15 pF	33 pF	

Medium Frequency Oscillator Option

This option supports an external crystal in the range of 100 kHz to 4 MHz. Ceramic resonators are also supported in this configuration.

Table 6 shows capacitor values that may be used with a quartz crystal in this mode.

Table 6. Recommended oscillator capacitors for use with the medium frequency oscillator option

Oscillator Fre- quency		V_{DD} = 2.7 to 4.5 V		V _{DD} = 4.5 to 6.0 V			
	Lower Limit	Optimal Value	Upper Limit	Lower Limit	Optimal Value	Upper Limit	
100 kHz	33 pF	33 pF	47 pF	33 pF	33 pF	47 pF	
1 MHz	15 pF	15 pF	33 pF	15 pF	22 pF	47 pF	
4 MHz	15 pF	15 pF	33 pF	15 pF	15 pF	33 pF	

High Frequency Oscillator Option

This option supports an external crystal in the range of 4 to 20 MHz. Ceramic resonators are also supported in this configuration.

Table 7 shows capacitor values that may be used with a quartz crystal in this mode.

Table 7. Recommended oscillator capacitors for use with the high frequency oscillator option

Oscillator Frequency		V_{DD} = 2.7 to 4.5 V		V _{DD} = 4.5 to 6.0 V			
	Lower Limit	Optimal Value	Upper Limit	Lower Limit	Optimal Value	Upper Limit	
4 MHz	15 pF	33 pF	47 pF	15 pF	33 pF	68 pF	
8 MHz	15 pF	15 pF	33 pF	15 pF	33 pF	47 pF	
16 MHz	-	-	-	15 pF	15 pF	33 pF	
20 MHz	-	-	-	15 pF	15 pF	33 pF	

On-Chip RC Oscillator Option

The on-chip RC oscillator option has a typical frequency of 6 MHz and can be divided down for slower operation through the use of the DIVM register. For on-chip oscillator tolerance see AC Electrical Characteristics table. A clock output on the X2/P2.0 pin may be enabled when the on-chip RC oscillator is used.

External Clock Input Option

In this configuration, the processor clock is input from an external source driving the X1/P2.1 pin. The rate may be from 0 Hz up to 20 MHz when V_{DD} is above 4.5 V and up to 10 MHz when V_{DD} is below 4.5 V. When the external clock input mode is used, the X2/P2.0 pin may be used as a standard port pin. A clock output on the X2/P2.0 pin may be enabled when the external clock input is used.

Clock Output

The P87LPC764 supports a clock output function when either the on-chip RC oscillator or external clock input options are selected. This allows external devices to synchronize to the P87LPC764. When enabled, via the ENCLK bit in the P2M1 register, the clock output appears on the X2/CLKOUT pin whenever the on-chip oscillator is running, including in Idle mode. The frequency of the clock output is 1/6 of the CPU clock rate. If the clock output is not needed in Idle mode, it may be turned off prior to entering Idle, saving additional power. The clock output may also be enabled when the external clock input option is selected.



Figure 16. Using the Crystal Oscillator



Figure 17. Using an External Clock Input



Figure 18. Block Diagram of Oscillator Control

CPU Clock Modification: CLKR and DIVM

For backward compatibility, the CLKR configuration bit allows setting the P87LPC764 instruction and peripheral timing to match standard 80C51 timing by dividing the CPU clock by two. Default timing for the P87LPC764 is 6 CPU clocks per machine cycle while standard 80C51 timing is 12 clocks per machine cycle. This division also applies to peripheral timing, allowing 80C51 code that is oscillator frequency and/or timer rate dependent. The CLKR bit is located in the EPROM configuration register UCFG1, described under EPROM Characteristics

In addition to this, the CPU clock may be divided down from the oscillator rate by a programmable divider, under program control. This function is controlled by the DIVM register. If the DIVM register is set to zero (the default value), the CPU will be clocked by either the unmodified oscillator rate, or that rate divided by two, as determined by the previously described CLKR function.

When the DIVM register is set to some value N (between 1 and 255), the CPU clock is divided by 2 * (N + 1). Clock division values from 4 through 512 are thus possible. This feature makes it possible to temporarily run the CPU at a lower rate, reducing power consumption, in a manner similar to Idle mode. By dividing the clock, the CPU can retain the ability to respond to events other than those that can cause interrupts (i.e. events that allow exiting the Idle mode) by executing its normal program at a lower rate. This can allow bypassing the oscillator startup time in cases where Power Down mode would otherwise be used. The value of DIVM may be changed by the program at any time without interrupting code execution.

Power Monitoring Functions

The P87LPC764 incorporates power monitoring functions designed to prevent incorrect operation during initial power up and power loss or reduction during operation. This is accomplished with two hardware functions: Power-On Detect and Brownout Detect.

Brownout Detection

The Brownout Detect function allows preventing the processor from failing in an unpredictable manner if the power supply voltage drops below a certain level. The default operation is for a brownout detection to cause a processor reset, however it may alternatively be configured to generate an interrupt by setting the BOI bit in the AUXR1 register (AUXR1.5).

The P87LPC764 allows selection of two Brownout levels: 2.5 V or 3.8 V. When V_{DD} drops below the selected voltage, the brownout detector triggers and remains active until V_{DD} is returns to a level above the Brownout Detect voltage. When Brownout Detect causes a processor reset, that reset remains active as long as V_{DD} remains below the Brownout Detect voltage. When Brownout Detect generates an interrupt, that interrupt occurs once as V_{DD} crosses from above to below the Brownout Detect voltage. For the interrupt to be processed, the interrupt system and the BOI interrupt must both be enabled (via the EA and EBO bits in IEN0).

When Brownout Detect is activated, the BOF flag in the PCON register is set so that the cause of processor reset may be determined by software. This flag will remain set until cleared by software.

P87LPC764

Timer/Counters

The P87LPC764 has two general purpose counter/timers which are upward compatible with the standard 80C51 Timer 0 and Timer 1. Both can be configured to operate either as timers or event counters (see Figure 22). An option to automatically toggle the T0 and/or T1 pins upon timer overflow has been added.

In the "Timer" function, the register is incremented every machine cycle. Thus, one can think of it as counting machine cycles. Since a machine cycle consists of 6 CPU clock periods, the count rate is 1/6 of the CPU clock frequency. Refer to the section Enhanced CPU for a description of the CPU clock.

In the "Counter" function, the register is incremented in response to a 1-to-0 transition at its corresponding external input pin, T0 or T1. In this function, the external input is sampled once during every machine cycle. When the samples of the pin state show a high in one cycle and a low in the next cycle, the count is incremented. The new count value appears in the register during the cycle following the one in which the transition was detected. Since it takes 2 machine cycles (12 CPU clocks) to recognize a 1-to-0 transition, the maximum count rate is 1/6 of the CPU clock frequency. There are no restrictions on the duty cycle of the external input signal, but to ensure that a given level is sampled at least once before it changes, it should be held for at least one full machine cycle.

The "Timer" or "Counter" function is selected by control bits C/\overline{T} in the Special Function Register TMOD. In addition to the "Timer" or "Counter" selection, Timer 0 and Timer 1 have four operating modes, which are selected by bit-pairs (M1, M0) in TMOD. Modes 0, 1, and 2 are the same for both Timers/Counters. Mode 3 is different. The four operating modes are described in the following text.



Figure 22. Timer/Counter Mode Control Register (TMOD)



Figure 27. Timer/Counter 0 Mode 3 (Two 8-Bit Counters)

Timer Overflow Toggle Output

Timers 0 and 1 can be configured to automatically toggle a port output whenever a timer overflow occurs. The same device pins that are used for the T0 and T1 count inputs are also used for the timer toggle outputs. This function is enabled by control bits TOOE and T1OE in the P2M1 register, and apply to Timer 0 and Timer 1 respectively. The port outputs will be a logic 1 prior to the first timer overflow when this mode is turned on.

UART

The P87LPC764 includes an enhanced 80C51 UART. The baud rate source for the UART is timer 1 for modes 1 and 3, while the rate is fixed in modes 0 and 2. Because CPU clocking is different on the P87LPC764 than on the standard 80C51, baud rate calculation is somewhat different. Enhancements over the standard 80C51 UART include Framing Error detection and automatic address recognition.

The serial port is full duplex, meaning it can transmit and receive simultaneously. It is also receive-buffered, meaning it can commence reception of a second byte before a previously received byte has been read from the SBUF register. However, if the first byte still hasn't been read by the time reception of the second byte is complete, the first byte will be lost. The serial port receive and transmit registers are both accessed through Special Function Register SBUF. Writing to SBUF loads the transmit register, and reading SBUF accesses a physically separate receive register.

The serial port can be operated in 4 modes:

Mode 0

Serial data enters and exits through RxD. TxD outputs the shift clock. 8 bits are transmitted or received, LSB first. The baud rate is fixed at 1/6 of the CPU clock frequency.

Mode 1

10 bits are transmitted (through TxD) or received (through RxD): a start bit (logical 0), 8 data bits (LSB first), and a stop bit (logical 1). When data is received, the stop bit is stored in RB8 in Special Function Register SCON. The baud rate is variable and is determined by the Timer 1 overflow rate.

Mode 2

11 bits are transmitted (through TxD) or received (through RxD): start bit (logical 0), 8 data bits (LSB first), a programmable 9th data bit, and a stop bit (logical 1). When data is transmitted, the 9th data bit (TB8 in SCON) can be assigned the value of 0 or 1. Or, for example, the parity bit (P, in the PSW) could be moved into TB8. When data is received, the 9th data bit goes into RB8 in Special Function Register SCON, while the stop bit is ignored. The baud rate is programmable to either 1/16 or 1/32 of the CPU clock frequency, as determined by the SMOD1 bit in PCON.

Mode 3

11 bits are transmitted (through TxD) or received (through RxD): a start bit (logical 0), 8 data bits (LSB first), a programmable 9th data bit, and a stop bit (logical 1). In fact, Mode 3 is the same as Mode 2 in all respects except baud rate. The baud rate in Mode 3 is variable and is determined by the Timer 1 overflow rate.

In all four modes, transmission is initiated by any instruction that uses SBUF as a destination register. Reception is initiated in Mode 0 by the condition RI = 0 and REN = 1. Reception is initiated in the other modes by the incoming start bit if REN = 1.

Serial Port Control Register (SCON)

The serial port control and status register is the Special Function Register SCON, shown in Figure 28. This register contains not only the mode selection bits, but also the 9th data bit for transmit and receive (TB8 and RB8), and the serial port interrupt bits (TI and RI).

The Framing Error bit (FE) allows detection of missing stop bits in the received data stream. The FE bit shares the bit position SCON.7

with the SM0 bit. Which bit appears in SCON at any particular time is determined by the SMOD0 bit in the PCON register. If SMOD0 = 0, SCON.7 is the SM0 bit. If SMOD0 = 1, SCON.7 is the FE bit. Once set, the FE bit remains set until it is cleared by software. This allows detection of framing errors for a group of characters without the need for monitoring it for every character individually.

CON Addre	ss: 98h								Reset Value: 00h
Bit Addressable									
		7 6	5	4	3	2	1	0	
	SM	D/FE SM1	SM2	REN	TB8	RB8	TI	RI	
BIT	SYMBOL	FUNCTION							
SCON.7	FE	Framing Error cleared by so See SM0 bit	Framing Error. This bit is set by the UART receiver when an invalid stop bit is detected. Must be cleared by software. The SMOD0 bit in the PCON register must be 1 for this bit to be accessible.						
SCON.7	SM0	With SM1, de to be access	efines the s ible. See Fl	erial port r E bit above	node. The e.	e SMOD0 I	bit in the P	CON regis	ster must be 0 for this bit
SCON. 6	SM1	With SM0, de	efines the s	erial port r	node (see	table belo	ow).		
	<u>SM0, SM1</u>	UART Mode		Baud	Rate				
	0 0	0: shift regist	er	CPU	clock/6				
	0 1	1: 8-bit UAR	Г	Varia	ble (see t	ext)			
	10	2: 9-bit UAR	Г	CPU	clock/32	or CPU clo	ock/16		
	11	3: 9-bit UAR	Г	Varia	ble (see t	ext)			
SCON.5	SM2	Enables the to 1, then RI will not be ac	Enables the multiprocessor communication feature in Modes 2 and 3. In Mode 2 or 3, if SM2 is set to 1, then RI will not be activated if the received 9th data bit (RB8) is 0. In Mode 1, if SM2=1 then RI will not be activated if a valid stop bit was not received. In Mode 0, SM2 should be 0.						
SCON.4	REN	Enables seri	al reception	. Set by so	oftware to	enable re	ception. C	lear by sof	tware to disable reception.
SCON.3	TB8	The 9th data	bit that will	be transm	itted in M	odes 2 an	d 3. Set or	clear by s	oftware as desired.
SCON.2	RB8	In Modes 2 a was received	In Modes 2 and 3, is the 9th data bit that was received. In Mode 1, it SM2=0, RB8 is the stop bit that was received. In Mode 0, RB8 is not used.						
SCON.1	TI	Transmit inte of the stop b	Transmit interrupt flag. Set by hardware at the end of the 8th bit time in Mode 0, or at the beginning of the stop bit in the other modes, in any serial transmission. Must be cleared by software.						
SCON.0	RI	Receive inte the stop bit ti	rrupt flag. S me in the o	et by hard ther mode	ware at th s, in any s	ne end of t serial rece	he 8th bit t ption (exce	ime in Mo ept see SM	de 0, or halfway through /2). Must be cleared by
software.			SU01157						

Figure 28. Serial Port Control Register (SCON)

P87LPC764

Baud Rates

The baud rate in Mode 0 is fixed: Mode 0 Baud Rate = CPU clock/6. The baud rate in Mode 2 depends on the value of bit SMOD1 in Special Function Register PCON. If SMOD1 = 0 (which is the value on reset), the baud rate is 1/32 of the CPU clock frequency. If SMOD1 = 1, the baud rate is 1/16 of the CPU clock frequency.

Mode 2 Baud Rate =
$$\frac{1 + \text{SMOD1}}{32} \times \text{CPU}$$
 clock frequency

Using Timer 1 to Generate Baud Rates

When Timer 1 is used as the baud rate generator, the baud rates in Modes 1 and 3 are determined by the Timer 1 overflow rate and the value of SMOD1. The Timer 1 interrupt should be disabled in this

application. The Timer itself can be configured for either "timer" or "counter" operation, and in any of its 3 running modes. In the most typical applications, it is configured for "timer" operation, in the auto-reload mode (high nibble of TMOD = 0010b). In that case the baud rate is given by the formula:

Mode 1, 3 Baud Rate =
$$\frac{\frac{\text{CPU clock frequency}}{192 \text{ (or 96 if SMOD1 = 1)}}}{256 - (\text{TH1})}$$

Tables 6 and 7 list various commonly used baud rates and how they can be obtained using Timer 1 as the baud rate generator.

Table 9. Baud Rates, Timer Values, and CPU Clock Frequencies for SMOD1 = 0

Timer Count	Baud Rate								
Timer Count	2400	4800	9600	19.2k	38.4k	57.6k			
-1	0.4608	0.9216	* 1.8432	* 3.6864	* 7.3728	* 11.0592			
-2	0.9216	1.8432	* 3.6864	* 7.3728	* 14.7456				
-3	1.3824	2.7648	5.5296	* 11.0592	-	-			
-4	* 1.8432	* 3.6864	* 7.3728	* 14.7456	-	-			
-5	2.3040	4.6080	9.2160	* 18.4320	-	-			
-6	2.7648	5.5296	* 11.0592	-	-	-			
-7	3.2256	6.4512	12.9024	-	-	-			
-8	* 3.6864	* 7.3728	* 14.7456	-	-	-			
-9	4.1472	8.2944	16.5888	-	-	-			
-10	4.6080	9.2160	* 18.4320	-	-	-			



Figure 31. Serial Port Mode 2

AC ELECTRICAL CHARACTERISTICS (FOR P87LPC764BD, BDH, BN, FN, FD, FDH)

 $T_{amb} = 0$ °C to +70 °C or -40°C to +85°C, $V_{DD} = 2.7$ V to 6.0 V unless otherwise specified; $V_{SS} = 0$ V^{1,2,3}

	FIGURE	DADAMETED	LIMITS			
SYMBOL	FIGURE	PARAMETER	MIN	MAX	UNIT	
External Cl	ock	•			•	
f _{osc}	39	Oscillator frequency ($V_{DD} = 4.0 \text{ V}$ to 6.0 V)	0	20	MHz	
f _{osc}	39	Oscillator frequency (V _{DD} = 2.7 V to 6.0 V)		0	10	MHz
t _C	39	Clock period and CPU timing cycle		1/f _{osc}	_	ns
f _{osc(tol)}		On-chip RC oscillator tolerance. Applies to P87LPC7	10	10	%	
f _{osc(tol)}		On-chip RC oscillator tolerance, all other devices	25	25	%	
t _{CLCX}	39	Clock low-time ⁴	f _{OSC} = 20 MHz	20	_	ns
t _{CLCX}	39] [$f_{OSC} = 10 \text{ MHz}$	40	_	ns
t _{CHCX}	39	Clock high-time ⁴	f _{OSC} = 20 MHz	20	-	ns
t _{CHCX}	39	f _{OSC} = 10 M		40	-	ns
Shift Regis	ter					
t _{XLXL}	38	Serial port clock cycle time	6t _C	-	ns	
t _{QVXH}	38	Output data setup to clock rising edge	5t _C - 133	-	ns	
t _{XHQX}	38	Output data hold after clock rising edge	1t _C - 80	-	ns	
t _{XHDV}	38	Input data setup to clock rising edge	-	5t _C – 133	ns	
t _{XHDX}	38	Input data hold after clock rising edge	0	_	ns	

NOTES:

1. Parameters are valid over operating temperature range unless otherwise specified.

Load capacitance for all outputs = 80 pF.
Parts are guaranteed to operate down to 0 Hz.

4. Applies only to an external clock source, not when a crystal is connected to the X1 and X2 pins.

5. For availability of other devices with this specification, please contact Philips sales office.

AC ELECTRICAL CHARACTERISTICS (FOR P87LPC764BD/01, BDH/01)

 T_{amb} = 0 °C to +70 °C, V_{DD} = 2.7 V to 6.0 V unless otherwise specified; V_{SS} = 0 V^{1,2,3}

	FIGURE	DADAMETED	LIMITS				
SYMBOL	FIGURE PARAMETER				MAX	UNIT	
External Cl	ock	•		-			
f _{osc}	39	Oscillator frequency ($V_{DD} = 4.0 \text{ V}$ to 6.0 V)		0	20	MHz	
f _{osc}	39	Oscillator frequency ($V_{DD} = 2.7 \text{ V to } 6.0 \text{ V}$)		0	10	MHz	
t _C	39	Clock period and CPU timing cycle		1/f _{osc}	-	ns	
t _{CLCX}	39	Clock low-time ¹	f _{osc} = 20 MHz	20	-	ns	
t _{CLCX}	39	1	f _{osc} = 10 MHz	40	-	ns	
t _{CHCX}	39	Clock high-time ¹	f _{osc} = 20 MHz	20	-	ns	
t _{CHCX}	39	1	f _{osc} = 10 MHz	40	-	ns	
Internal RC	Internal RC Oscillator						
f _{osc(cal)}		On-chip RC oscillator calibration ²	$f_{osc(RC)} = 6 MHz$	-1	+1	%	
f _{osc(tol)}		On-chip RC oscillator, 0 °C to +50 °C ^{3,4} tol.	f _{osc(RC)} = 6 MHz	-2.5	+2.5	%	
f _{osc(tol)}		On-chip RC oscillator, 0 °C to +70 °C ³ tol.	f _{osc(RC)} = 6 MHz	-5 ⁵	+2.5	%	
Shift Regis	ter		-	-			
t _{XLXL}	38	Serial port clock cycle time		6t _C	-	ns	
t _{QVXH}	38	Output data setup to clock rising edge	5t _C – 133	-	ns		
t _{XHQX}	38	Output data hold after clock rising edge	1t _C - 80	-	ns		
t _{XHDV}	38	Input data setup to clock rising edge	-	5t _C – 133	ns		
t _{XHDX}	38	Input data hold after clock rising edge	0	-	ns		

NOTES:

1. Applies only to an external clock source, not when a crystal is connected to the X1 and X2 pins. 2. Tested at $V_{DD} = 5.0$ V and room temperature. 3. These parameters are characterized but not tested.

4. +/- 2.5% accuracy enables serial communication over the UART with the internal Oscillator.

5. Min frequency at hot temperature.

COMPARATOR ELECTRICAL CHARACTERISTICS (FOR P87LPC764HDH)

 V_{DD} = 4.5 V to 5.5 V; T_{amb} = –40°C to +125°C

SYMPOL	DADAMETED	TEST CONDITIONS				
STMBOL	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	
V _{IO}	Offset voltage comparator inputs ¹				±20	mV
V _{CR}	Common mode range comparator inputs		0		V _{DD} -0.3	V
CMRR	Common mode rejection ratio ¹				-50	dB
	Response time			250	500	ns
	Comparator enable to output valid				10	μs
IIL	Input leakage current, comparator	$0 < V_{IN} < V_{DD}$			±10	μA

NOTE:

1. This parameter is guaranteed by characterization, but not tested in production.



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