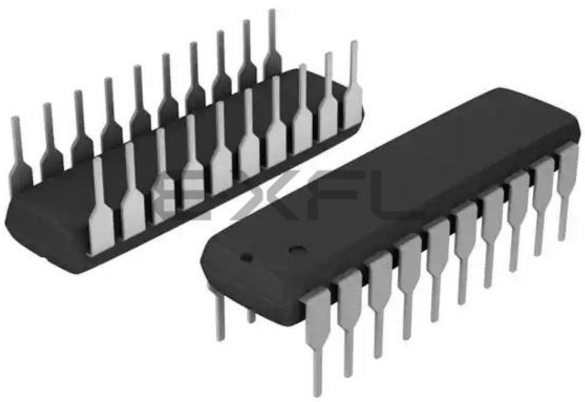
NXP USA Inc. - P87LPC764BN,112 Datasheet



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Applications of "Embedded - Microcontrollers"

Details	
Product Status	Obsolete
Core Processor	8051
Core Size	8-Bit
Speed	20MHz
Connectivity	I²C, UART/USART
Peripherals	Brown-out Detect/Reset, LED, POR, WDT
Number of I/O	18
Program Memory Size	4KB (4K x 8)
Program Memory Type	OTP
EEPROM Size	-
RAM Size	128 x 8
Voltage - Supply (Vcc/Vdd)	2.7V ~ 6V
Data Converters	-
Oscillator Type	Internal
Operating Temperature	0°C ~ 70°C (TA)
Mounting Type	Through Hole
Package / Case	20-DIP (0.300", 7.62mm)
Supplier Device Package	20-DIP
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/p87lpc764bn-112

Name	Description	SFR Address	Bit Functions and Addresses MSB							LSB	Reset Value
			D7	D6	D5	D4	D3	D2	D1	D0	
PSW*	Program status word	D0h	CY	AC	F0	RS1	RS0	OV	F1	P	00h
PT0AD#	Port 0 digital input disable	F6h									00h
			9F	9E	9D	9C	9B	9A	99	98	
SCON*	Serial port control	98h	SM0	SM1	SM2	REN	TB8	RB8	TI	RI	00h
SBUF	Serial port data buffer register	99h									xxh
SADDR#	Serial port address register	A9h									00h
SADEN#	Serial port address enable	B9h									00h
SP	Stack pointer	81h									07h
			8F	8E	8D	8C	8B	8A	89	88	
TCON*	Timer 0 and 1 control	88h	TF1	TR1	TF0	TR0	IE1	IT1	IE0	IT0	00h
TH0	Timer 0 high byte	8Ch									00h
TH1	Timer 1 high byte	8Dh									00h
TL0	Timer 0 low byte	8Ah									00h
TL1	Timer 1 low byte	8Bh									00h
TMOD	Timer 0 and 1 mode	89h	GATE	C/$\overline{T}$	M1	M0	GATE	C/$\overline{T}$	M1	M0	00h
WDCON#	Watchdog control register	A7h	–	–	WDOVF	WDRUN	WDCLK	WDS2	WDS1	WDS0	Note 4
WDRST#	Watchdog reset register	A6h									xxh

NOTES:
* SFRs are bit addressable.
# SFRs are modified from or added to the 80C51 SFRs.

1. Unimplemented bits in SFRs are X (unknown) at all times. Ones should not be written to these bits since they may be used for other purposes in future derivatives. The reset value shown in the table for these bits is 0.
2. I/O port values at reset are determined by the PRHI bit in the UCFG1 configuration byte.
3. The PCON reset value is x x BOF POF–0 0 0 0b. The BOF and POF flags are not affected by reset. The POF flag is set by hardware upon power up. The BOF flag is set by the occurrence of a brownout reset/interrupt and upon power up.
4. The WDCON reset value is xx11 0000b for a Watchdog reset, xx01 0000b for all other reset causes if the watchdog is enabled, and xx00 0000b for all other reset causes if the watchdog is disabled.

Internal Reference Voltage

An internal reference voltage generator may supply a default reference when a single comparator input pin is used. The value of the internal reference voltage, referred to as V_{ref}, is 1.28 V ±10%.

Comparator Interrupt

Each comparator has an interrupt flag CMFn contained in its configuration register. This flag is set whenever the comparator output changes state. The flag may be polled by software or may be used to generate an interrupt. The interrupt will be generated when the corresponding enable bit ECn in the IEN1 register is set and the interrupt system is enabled via the EA bit in the IEN0 register.

Comparators and Power Reduction Modes

Either or both comparators may remain enabled when Power Down or Idle mode is activated. The comparators will continue to function in the power reduction mode. If a comparator interrupt is enabled, a change of the comparator output state will generate an interrupt and wake up the processor. If the comparator output to a pin is enabled, the pin should be configured in the push-pull mode in order to obtain fast switching times while in power down mode. The reason is that with the oscillator stopped, the temporary strong pull-up that normally occurs during switching on a quasi-bidirectional port pin does not take place.

Comparators consume power in Power Down and Idle modes, as well as in the normal operating mode. This fact should be taken into account when system power consumption is an issue.

Comparator Configuration Example

The code shown in Figure 5 is an example of initializing one comparator. Comparator 1 is configured to use the CIN1A and CMPREF inputs, outputs the comparator result to the CMP1 pin, and generates an interrupt when the comparator output changes.

The interrupt routine used for the comparator must clear the interrupt flag (CMF1 in this case) before returning.

```
CmpInit:
    mov      PT0AD,#30h        ; Disable digital inputs on pins that are used
                               ;   for analog functions: CIN1A, CMPREF.
    anl      P0M2,#0cfh        ; Disable digital outputs on pins that are used
    orl      P0M1,#30h         ;   for analog functions: CIN1A, CMPREF.
    mov      CMP1,#24h         ; Turn on comparator 1 and set up for:
                               ;   – Positive input on CIN1A.
                               ;   – Negative input from CMPREF pin.
                               ;   – Output to CMP1 pin enabled.
    call     delay10us         ; The comparator has to start up for at
                               ;   least 10 microseconds before use.
    anl      CMP1,#0feh        ; Clear comparator 1 interrupt flag.
    setb     EC1               ; Enable the comparator 1 interrupt. The
                               ;   priority is left at the current value.
    setb     EA                ; Enable the interrupt system (if needed).
    ret                        ; Return to caller.
                                                                  SU01189
```

Figure 5.

I^2C Serial Interface

The I^2C bus uses two wires (SDA and SCL) to transfer information between devices connected to the bus. The main features of the bus are:

- Bidirectional data transfer between masters and slaves.
- Serial addressing of slaves (no added wiring).
- Acknowledgment after each transferred byte.
- Multimaster bus.
- Arbitration between simultaneously transmitting masters without corruption of serial data on bus.

The I^2C subsystem includes hardware to simplify the software required to drive the I^2C bus. The hardware is a single bit interface which in addition to including the necessary arbitration and framing error checks, includes clock stretching and a bus timeout timer. The interface is synchronized to software either through polled loops or interrupts.

Refer to the application note AN422, entitled “Using the 8XC751 Microcontroller as an I^2C Bus Master” for additional discussion of the 8xC76x I^2C interface and sample driver routines.

The P87LPC764 I^2C implementation duplicates that of the 87C751 and 87C752 except for the following details:

- The interrupt vector addresses for both the I^2C interrupt and the Timer I interrupt.
- The I^2C SFR addresses (I2CON, I2CFG, I2DAT).
- The location of the I^2C interrupt enable bit and the name of the SFR it is located within (EI2 is Bit 0 in IEN1).
- The location of the Timer I interrupt enable bit and the name of the SFR it is located within (ETI is Bit 7 in IEN1).
- The I^2C and Timer I interrupts have a settable priority.

Timer I is used to both control the timing of the I^2C bus and also to detect a “bus locked” condition, by causing an interrupt when nothing happens on the I^2C bus for an inordinately long period of time while a transmission is in progress. If this interrupt occurs, the program has the opportunity to attempt to correct the fault and resume I^2C operation.

Six time spans are important in I^2C operation and are insured by timer I:

- The MINIMUM HIGH time for SCL when this device is the master.
- The MINIMUM LOW time for SCL when this device is a master. This is not very important for a single-bit hardware interface like this one, because the SCL low time is stretched until the software responds to the I^2C flags. The software response time normally meets or exceeds the MIN LO time. In cases where the software responds within MIN HI + MIN LO) time, timer I will ensure that the minimum time is met.
- The MINIMUM SCL HIGH TO SDA HIGH time in a stop condition.
- The MINIMUM SDA HIGH TO SDA LOW time between I^2C stop and start conditions (4.7ms, see I^2C specification).
- The MINIMUM SDA LOW TO SCL LOW time in a start condition.
- The MAXIMUM SCL CHANGE time while an I^2C frame is in progress. A frame is in progress between a start condition and the following stop condition. This time span serves to detect a lack of software response on this device as well as external I^2C problems. SCL “stuck low” indicates a faulty master or slave. SCL “stuck high” may mean a faulty device, or that noise induced onto the I^2C bus caused all masters to withdraw from I^2C arbitration.

The first five of these times are 4.7 ms (see I^2C specification) and are covered by the low order three bits of timer I. Timer I is clocked by the P87LPC764 CPU clock. Timer I can be pre-loaded with one of four values to optimize timing for different oscillator frequencies. At lower frequencies, software response time is increased and will degrade maximum performance of the I^2C bus. See special function register I2CFG description for prescale values (CT0, CT1).

The MAXIMUM SCL CHANGE time is important, but its exact span is not critical. The complete 10 bits of timer I are used to count out the maximum time. When I^2C operation is enabled, this counter is cleared by transitions on the SCL pin. The timer does not run between I^2C frames (i.e., whenever reset or stop occurred more recently than the last start). When this counter is running, it will carry out after 1020 to 1023 machine cycles have elapsed since a change on SCL. A carry out causes a hardware reset of the I^2C interface and generates an interrupt if the Timer I interrupt is enabled. In cases where the bus hang-up is due to a lack of software response by this device, the reset releases SCL and allows I^2C operation among other devices to continue.

Timer I is enabled to run, and will reset the I^2C interface upon overflow, if the TIRUN bit in the I2CFG register is set. The Timer I interrupt may be enabled via the ETI bit in IEN1, and its priority set by the PTIH and PTI bits in the IP1H and IP1 registers respectively.

I^2C Interrupts

If I^2C interrupts are enabled (EA and EI2 are both set to 1), an I^2C interrupt will occur whenever the ATN flag is set by a start, stop, arbitration loss, or data ready condition (refer to the description of ATN following). In practice, it is not efficient to operate the I^2C interface in this fashion because the I^2C interrupt service routine would somehow have to distinguish between hundreds of possible conditions. Also, since I^2C can operate at a fairly high rate, the software may execute faster if the code simply waits for the I^2C interface.

Typically, the I^2C interrupt should only be used to indicate a start condition at an idle slave device, or a stop condition at an idle master device (if it is waiting to use the I^2C bus). This is accomplished by enabling the I^2C interrupt only during the aforementioned conditions.

Reading I2CON

RDAT — The data from SDA is captured into “Receive DATa” whenever a rising edge occurs on SCL. RDAT is also available (with seven low-order zeros) in the I2DAT register. The difference between reading it here and there is that reading I2DAT clears DRDY, allowing the I^2C to proceed on to another bit. Typically, the first seven bits of a received byte are read from I2DAT, while the 8th is read here. Then I2DAT can be written to send the Acknowledge bit and clear DRDY.

ATN — “ATteNtion” is 1 when one or more of DRDY, ARL, STR, or STP is 1. Thus, ATN comprises a single bit that can be tested to release the I^2C service routine from a “wait loop.”

DRDY — “Data ReaDY” (and thus ATN) is set when a rising edge occurs on SCL, except at idle slave. DRDY is cleared by writing CDR = 1, or by writing or reading the I2DAT register. The following low period on SCL is stretched until the program responds by clearing DRDY.

ARL "Arbitration Loss" is 1 when transmit Active was set, but this device lost arbitration to another transmitter. Transmit Active is cleared when ARL is 1. There are four separate cases in which ARL is set.

1. If the program sent a 1 or repeated start, but another device sent a 0, or a stop, so that SDA is 0 at the rising edge of SCL. (If the other device sent a stop, the setting of ARL will be followed shortly by STP being set.)

2. If the program sent a 1, but another device sent a repeated start, and it drove SDA low before SCL could be driven low. (This type of ARL is always accompanied by STR = 1.)

3. In master mode, if the program sent a repeated start, but another device sent a 1, and it drove SCL low before this device could drive SDA low.

4. In master mode, if the program sent stop, but it could not be sent because another device sent a 0.

STR "STaRt" is set to a 1 when an I^2C start condition is detected at a non-idle slave or at a master. (STR is not set when an idle slave becomes active due to a start bit; the slave has nothing useful to do until the rising edge of SCL sets DRDY.)

STP "SToP" is set to 1 when an I^2C stop condition is detected at a non-idle slave or at a master. (STP is not set for a stop condition at an idle slave.)

MASTER "MASTER" is 1 if this device is currently a master on the I^2C. MASTER is set when MASTRQ is 1 and the bus is not busy (i.e., if a start bit hasn't been received since reset or a "Timer I" time-out, or if a stop has been received since the last start). MASTER is cleared when ARL is set, or after the software writes MASTRQ = 0 and then XSTP = 1.

Writing I2CON

Typically, for each bit in an I^2C message, a service routine waits for ATN = 1. Based on DRDY, ARL, STR, and STP, and on the current bit position in the message, it may then write I2CON with one or more of the following bits, or it may read or write the I2DAT register.

CXA Writing a 1 to "Clear Xmit Active" clears the Transmit Active state. (Reading the I2DAT register also does this.)

Regarding Transmit Active

Transmit Active is set by writing the I2DAT register, or by writing I2CON with XSTR = 1 or XSTP = 1. The I^2C interface will only drive the SDA line low when Transmit Active is set, and the ARL bit will only be set to 1 when Transmit Active is set. Transmit Active is cleared by reading the I2DAT register, or by writing I2CON with CXA = 1. Transmit Active is automatically cleared when ARL is 1.

IDLE Writing 1 to "IDLE" causes a slave's I^2C hardware to ignore the I^2C until the next start condition (but if MASTRQ is 1, then a stop condition will cause this device to become a master).

CDR Writing a 1 to "Clear Data Ready" clears DRDY. (Reading or writing the I2DAT register also does this.)

CARL Writing a 1 to "Clear Arbitration Loss" clears the ARL bit.

CSTR Writing a 1 to "Clear STaRt" clears the STR bit.

CSTP Writing a 1 to "Clear SToP" clears the STP bit. Note that if one or more of DRDY, ARL, STR, or STP is 1, the low time of SCL is stretched until the service routine responds by clearing them.

XSTR Writing 1s to "Xmit repeated STaRt" and CDR tells the I^2C hardware to send a repeated start condition. This should only be at a master. Note that XSTR need not and should not be used to send an "initial" (non-repeated) start; it is sent automatically by the I^2C hardware. Writing XSTR = 1 includes the effect of writing I2DAT with XDAT = 1; it sets Transmit Active and releases SDA to high during the SCL low time. After SCL goes high, the I^2C hardware waits for the suitable minimum time and then drives SDA low to make the start condition.

XSTP Writing 1s to "Xmit SToP" and CDR tells the I^2C hardware to send a stop condition. This should only be done at a master. If there are no more messages to initiate, the service routine should clear the MASTRQ bit in I2CFG to 0 before writing XSTP with 1. Writing XSTP = 1 includes the effect of writing I2DAT with XDAT = 0; it sets Transmit Active and drives SDA low during the SCL low time. After SCL goes high, the I^2C hardware waits for the suitable minimum time and then releases SDA to high to make the stop condition.

External Interrupt Inputs

The P87LPC764 has two individual interrupt inputs as well as the Keyboard Interrupt function. The latter is described separately elsewhere in this section. The two interrupt inputs are identical to those present on the standard 80C51 microcontroller.

The external sources can be programmed to be level-activated or transition-activated by setting or clearing bit IT1 or IT0 in Register TCON. If ITn = 0, external interrupt n is triggered by a detected low at the $\overline{\text{INTn}}$ pin. If ITn = 1, external interrupt n is edge triggered. In this mode if successive samples of the $\overline{\text{INTn}}$ pin show a high in one cycle and a low in the next cycle, interrupt request flag IEn in TCON is set, causing an interrupt request.

Since the external interrupt pins are sampled once each machine cycle, an input high or low should hold for at least 6 CPU Clocks to ensure proper sampling. If the external interrupt is transition-activated, the external source has to hold the request pin high for at least one machine cycle, and then hold it low for at least one machine cycle. This is to ensure that the transition is seen and that interrupt request flag IEn is set. IEn is automatically cleared by the CPU when the service routine is called.

If the external interrupt is level-activated, the external source must hold the request active until the requested interrupt is actually generated. If the external interrupt is still asserted when the interrupt service routine is completed another interrupt will be generated. It is not necessary to clear the interrupt flag IEn when the interrupt is level sensitive, it simply tracks the input pin level.

If an external interrupt is enabled when the P87LPC764 is put into Power Down or Idle mode, the interrupt will cause the processor to wake up and resume operation. Refer to the section on Power Reduction Modes for details.

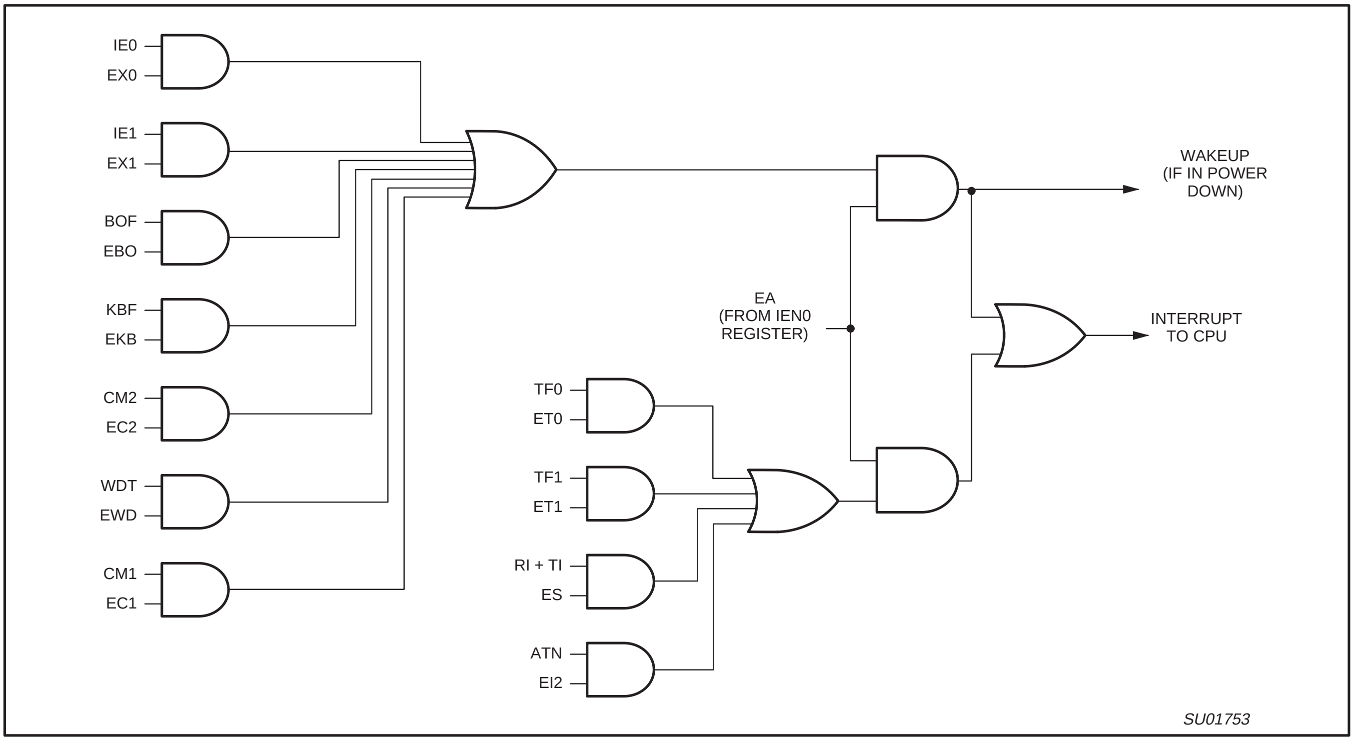

Figure 9. Interrupt Sources, Interrupt Enables, and Power Down Wakeup Sources

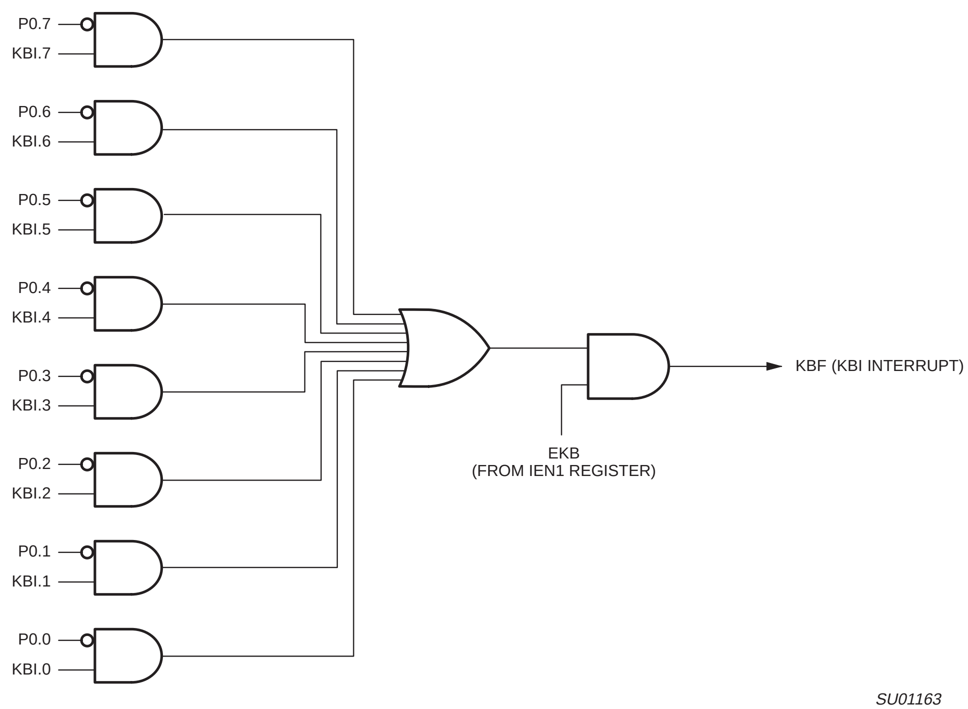

Figure 14. Keyboard Interrupt

KBI Address: 86h Reset Value: 00h

Not Bit Addressable

7	6	5	4	3	2	1	0
KBI.7	KBI.6	KBI.5	KBI.4	KBI.3	KBI.2	KBI.1	KBI.0

BIT	SYMBOL	FUNCTION
KBI.7	KBI.7	When set, enables P0.7 as a cause of a Keyboard Interrupt.
KBI.6	KBI.6	When set, enables P0.6 as a cause of a Keyboard Interrupt.
KBI.5	KBI.5	When set, enables P0.5 as a cause of a Keyboard Interrupt.
KBI.4	KBI.4	When set, enables P0.4 as a cause of a Keyboard Interrupt.
KBI.3	KBI.3	When set, enables P0.3 as a cause of a Keyboard Interrupt.
KBI.2	KBI.2	When set, enables P0.2 as a cause of a Keyboard Interrupt.
KBI.1	KBI.1	When set, enables P0.1 as a cause of a Keyboard Interrupt.
KBI.0	KBI.0	When set, enables P0.0 as a cause of a Keyboard Interrupt.

Note: the Keyboard Interrupt must be enabled in order for the settings of the KBI register to be effective. The interrupt flag (KBF) is located at bit 7 of AUXR1.

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Figure 15. Keyboard Interrupt Register (KBI)

For correct activation of Brownout Detect, the V_{DD} fall time must be no faster than 50 mV/µs. When V_{DD} is restored, is should not rise faster than 2 mV/µs in order to insure a proper reset.

The brownout voltage (2.5 V or 3.8 V) is selected via the BOV bit in the EPROM configuration register UCFG1. When unprogrammed (BOV = 1), the brownout detect voltage is 2.5 V. When programmed (BOV = 0), the brownout detect voltage is 3.8 V.

If the Brownout Detect function is not required in an application, it may be disabled, thus saving power. Brownout Detect is disabled by setting the control bit BOD in the AUXR1 register (AUXR1.6).

Power On Detection
The Power On Detect has a function similar to the Brownout Detect, but is designed to work as power comes up initially, before the power supply voltage reaches a level where Brownout Detect can work. When this feature is activated, the POF flag in the PCON register is set to indicate an initial power up condition. The POF flag will remain set until cleared by software.

Power Reduction Modes

The P87LPC764 supports Idle and Power Down modes of power reduction.

Idle Mode
The Idle mode leaves peripherals running in order to allow them to activate the processor when an interrupt is generated. Any enabled interrupt source or Reset may terminate Idle mode. Idle mode is entered by setting the IDL bit in the PCON register (see Figure 19).

Power Down Mode
The Power Down mode stops the oscillator in order to absolutely minimize power consumption. Power Down mode is entered by setting the PD bit in the PCON register (see Figure 19).

The processor can be made to exit Power Down mode via Reset or one of the interrupt sources shown in Table 5. This will occur if the interrupt is enabled and its priority is higher than any interrupt currently in progress.

In Power Down mode, the power supply voltage may be reduced to the RAM keep-alive voltage V_{RAM}. This retains the RAM contents at the point where Power Down mode was entered. SFR contents are not guaranteed after V_{DD} has been lowered to V_{RAM}, therefore it is recommended to wake up the processor via Reset in this case. V_{DD} must be raised to within the operating range before the Power Down mode is exited. Since the watchdog timer has a separate oscillator, it may reset the processor upon overflow if it is running during Power Down.

Note that if the Brownout Detect reset is enabled, the processor will be put into reset as soon as V_{DD} drops below the brownout voltage. If Brownout Detect is configured as an interrupt and is enabled, it will wake up the processor from Power Down mode when V_{DD} drops below the brownout voltage.

When the processor wakes up from Power Down mode, it will start the oscillator immediately and begin execution when the oscillator is stable. Oscillator stability is determined by counting 1024 CPU clocks after start-up when one of the crystal oscillator configurations is used, or 256 clocks after start-up for the internal RC or external clock input configurations.

Some chip functions continue to operate and draw power during Power Down mode, increasing the total power used during Power Down. These include the Brownout Detect, Watchdog Timer, and Comparators.

PCON Address: 87h

Not Bit Addressable

Reset Value:
- 30h for a Power On reset
- 20h for a Brownout reset
- 00h for other reset sources

7	6	5	4	3	2	1	0
SMOD1	SMOD0	BOF	POF	GF1	GF0	PD	IDL

BIT	SYMBOL	FUNCTION
PCON.7	SMOD1	When set, this bit doubles the UART baud rate for modes 1, 2, and 3.
PCON.6	SMOD0	This bit selects the function of bit 7 of the SCON SFR. When 0, SCON.7 is the SM0 bit. When 1, SCON.7 is the FE (Framing Error) flag. See Figure 28 for additional information.
PCON.5	BOF	Brown Out Flag. Set automatically when a brownout reset or interrupt has occurred. Also set at power on. Cleared by software. Refer to the Power Monitoring Functions section for additional information.
PCON.4	POF	Power On Flag. Set automatically when a power-on reset has occurred. Cleared by software. Refer to the Power Monitoring Functions section for additional information.
PCON.3	GF1	General purpose flag 1. May be read or written by user software, but has no effect on operation.
PCON.2	GF0	General purpose flag 0. May be read or written by user software, but has no effect on operation.
PCON.1	PD	Power Down control bit. Setting this bit activates Power Down mode operation. Cleared when the Power Down mode is terminated (see text).
PCON.0	IDL	Idle mode control bit. Setting this bit activates Idle mode operation. Cleared when the Idle mode is terminated (see text).

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Figure 19. Power Control Register (PCON)

Mode 0

Putting either Timer into Mode 0 makes it look like an 8048 Timer, which is an 8-bit Counter with a divide-by-32 prescaler. Figure 24 shows Mode 0 operation.

In this mode, the Timer register is configured as a 13-bit register. As the count rolls over from all 1s to all 0s, it sets the Timer interrupt flag TFn. The count input is enabled to the Timer when TRn = 1 and either GATE = 0 or $\overline{\text{INTn}}$ = 1. (Setting GATE = 1 allows the Timer to be controlled by external input $\overline{\text{INTn}}$, to facilitate pulse width measurements). TRn is a control bit in the Special Function Register TCON (Figure 23). The GATE bit is in the TMOD register.

The 13-bit register consists of all 8 bits of THn and the lower 5 bits of TLn. The upper 3 bits of TLn are indeterminate and should be ignored. Setting the run flag (TRn) does not clear the registers.

Mode 0 operation is the same for Timer 0 and Timer 1. See Figure 24. There are two different GATE bits, one for Timer 1 (TMOD.7) and one for Timer 0 (TMOD.3).

TCON Address: 88h Reset Value: 00h

Bit Addressable

7	6	5	4	3	2	1	0
TF1	TR1	TF0	TR0	IE1	IT1	IE0	IT0

BIT	SYMBOL	FUNCTION
TCON.7	TF1	Timer 1 overflow flag. Set by hardware on Timer/Counter overflow. Cleared by hardware when the interrupt is processed, or by software.
TCON.6	TR1	Timer 1 Run control bit. Set/cleared by software to turn Timer/Counter 1 on/off.
TCON.5	TF0	Timer 0 overflow flag. Set by hardware on Timer/Counter overflow. Cleared by hardware when the processor vectors to the interrupt routine, or by software.
TCON.4	TR0	Timer 0 Run control bit. Set/cleared by software to turn Timer/Counter 0 on/off.
TCON.3	IE1	Interrupt 1 Edge flag. Set by hardware when external interrupt 1 edge is detected. Cleared by hardware when the interrupt is processed, or by software.
TCON.2	IT1	Interrupt 1 Type control bit. Set/cleared by software to specify falling edge/low level triggered external interrupts.
TCON.1	IE0	Interrupt 0 Edge flag. Set by hardware when external interrupt 0 edge is detected. Cleared by hardware when the interrupt is processed, or by software.
TCON.0	IT0	Interrupt 0 Type control bit. Set/cleared by software to specify falling edge/low level triggered external interrupts.

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Figure 23. Timer/Counter Control Register (TCON)

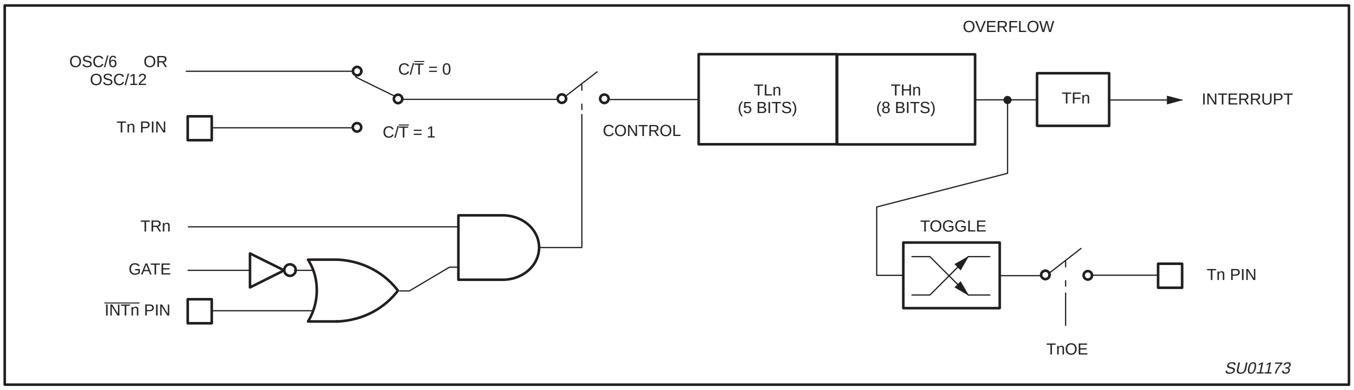


Figure 24. Timer/Counter 0 or 1 in Mode 0 (13-Bit Counter)

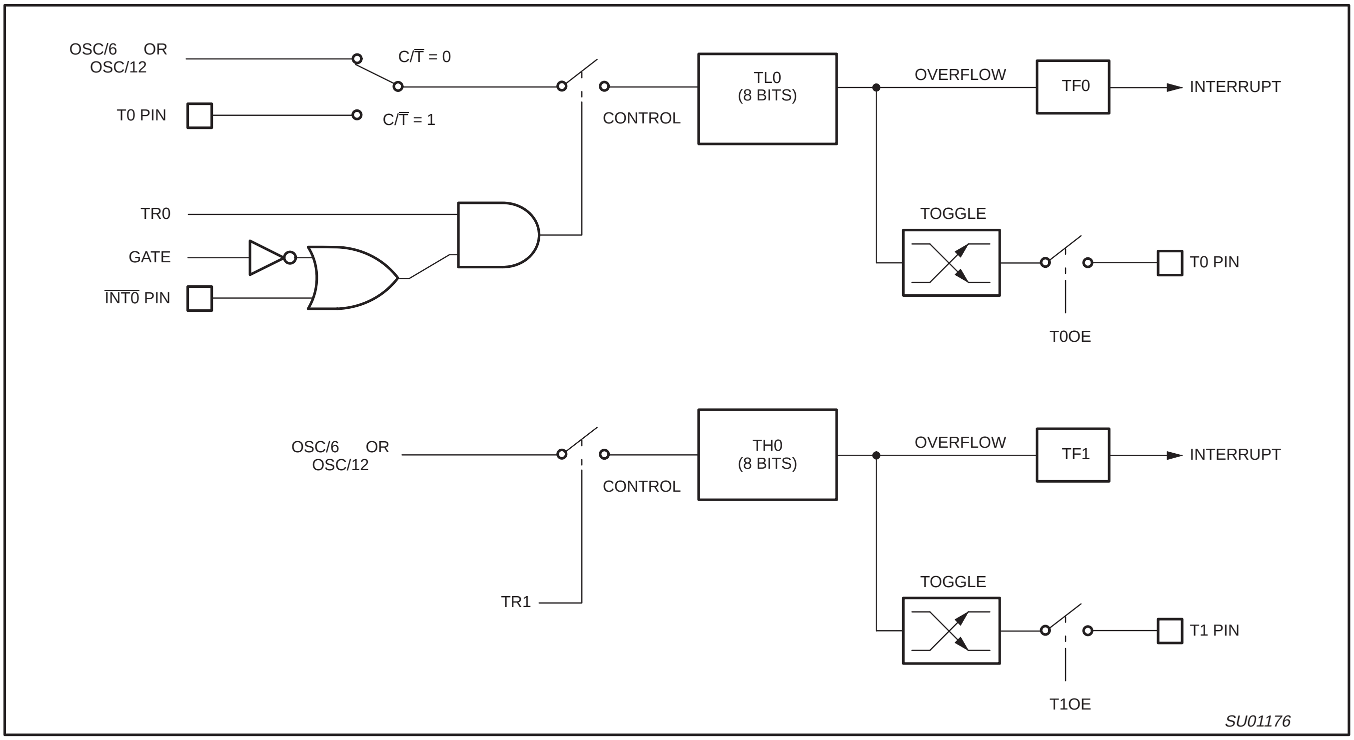


Figure 27. Timer/Counter 0 Mode 3 (Two 8-Bit Counters)

Timer Overflow Toggle Output

Timers 0 and 1 can be configured to automatically toggle a port output whenever a timer overflow occurs. The same device pins that are used for the T0 and T1 count inputs are also used for the timer toggle outputs. This function is enabled by control bits T0OE and T1OE in the P2M1 register, and apply to Timer 0 and Timer 1 respectively. The port outputs will be a logic 1 prior to the first timer overflow when this mode is turned on.

UART

The P87LPC764 includes an enhanced 80C51 UART. The baud rate source for the UART is timer 1 for modes 1 and 3, while the rate is fixed in modes 0 and 2. Because CPU clocking is different on the P87LPC764 than on the standard 80C51, baud rate calculation is somewhat different. Enhancements over the standard 80C51 UART include Framing Error detection and automatic address recognition.

The serial port is full duplex, meaning it can transmit and receive simultaneously. It is also receive-buffered, meaning it can commence reception of a second byte before a previously received byte has been read from the SBUF register. However, if the first byte still hasn't been read by the time reception of the second byte is complete, the first byte will be lost. The serial port receive and transmit registers are both accessed through Special Function Register SBUF. Writing to SBUF loads the transmit register, and reading SBUF accesses a physically separate receive register.

The serial port can be operated in 4 modes:

Mode 0

Serial data enters and exits through RxD. TxD outputs the shift clock. 8 bits are transmitted or received, LSB first. The baud rate is fixed at 1/6 of the CPU clock frequency.

Mode 1

10 bits are transmitted (through TxD) or received (through RxD): a start bit (logical 0), 8 data bits (LSB first), and a stop bit (logical 1). When data is received, the stop bit is stored in RB8 in Special Function Register SCON. The baud rate is variable and is determined by the Timer 1 overflow rate.

Mode 2

11 bits are transmitted (through TxD) or received (through RxD): start bit (logical 0), 8 data bits (LSB first), a programmable 9th data bit, and a stop bit (logical 1). When data is transmitted, the 9th data bit (TB8 in SCON) can be assigned the value of 0 or 1. Or, for example, the parity bit (P, in the PSW) could be moved into TB8. When data is received, the 9th data bit goes into RB8 in Special Function Register SCON, while the stop bit is ignored. The baud rate is programmable to either 1/16 or 1/32 of the CPU clock frequency, as determined by the SMOD1 bit in PCON.

Mode 3

11 bits are transmitted (through TxD) or received (through RxD): a start bit (logical 0), 8 data bits (LSB first), a programmable 9th data bit, and a stop bit (logical 1). In fact, Mode 3 is the same as Mode 2 in all respects except baud rate. The baud rate in Mode 3 is variable and is determined by the Timer 1 overflow rate.

In all four modes, transmission is initiated by any instruction that uses SBUF as a destination register. Reception is initiated in Mode 0 by the condition RI = 0 and REN = 1. Reception is initiated in the other modes by the incoming start bit if REN = 1.

More About UART Mode 0

Serial data enters and exits through RxD. TxD outputs the shift clock. 8 bits are transmitted/received: 8 data bits (LSB first). The baud rate is fixed at 1/6 the CPU clock frequency. Figure 29 shows a simplified functional diagram of the serial port in Mode 0, and associated timing.

Transmission is initiated by any instruction that uses SBUF as a destination register. The "write to SBUF" signal at S6P2 also loads a 1 into the 9th position of the transmit shift register and tells the TX Control block to commence a transmission. The internal timing is such that one full machine cycle will elapse between "write to SBUF" and activation of SEND.

SEND enables the output of the shift register to the alternate output function line of P1.1 and also enable SHIFT CLOCK to the alternate output function line of P1.0. SHIFT CLOCK is low during S3, S4, and S5 of every machine cycle, and high during S6, S1, and S2. At S6P2 of every machine cycle in which SEND is active, the contents of the transmit shift are shifted to the right one position.

As data bits shift out to the right, zeros come in from the left. When the MSB of the data byte is at the output position of the shift register, then the 1 that was initially loaded into the 9th position, is just to the left of the MSB, and all positions to the left of that contain zeros. This condition flags the TX Control block to do one last shift and then deactivate SEND and set T1. Both of these actions occur at S1P1 of the 10th machine cycle after "write to SBUF." Reception is initiated by the condition REN = 1 and R1 = 0. At S6P2 of the next machine cycle, the RX Control unit writes the bits 11111110 t o the receive shift register, and in the next clock phase activates RECEIVE.

RECEIVE enable SHIFT CLOCK to the alternate output function line of P1.0. SHIFT CLOCK makes transitions at S3P1 and S6P1 of every machine cycle. At S6P2 of every machine cycle in which RECEIVE is active, the contents of the receive shift register are shifted to the left one position. The value that comes in from the right is the value that was sampled at the P1.1 pin at S5P2 of the same machine cycle.

As data bits come in from the right, 1s shift out to the left. When the 0 that was initially loaded into the rightmost position arrives at the leftmost position in the shift register, it flags the RX Control block to do one last shift and load SBUF. At S1P1 of the 10th machine cycle after the write to SCON that cleared RI, RECEIVE is cleared as RI is set.

More About UART Mode 1

Ten bits are transmitted (through TxD), or received (through RxD): a start bit (0), 8 data bits (LSB first), and a stop bit (1). On receive, the stop bit goes into RB8 in SCON. In the P87LPC764 the baud rate is determined by the Timer 1 overflow rate. Figure 30 shows a simplified functional diagram of the serial port in Mode 1, and associated timings for transmit receive.

Transmission is initiated by any instruction that uses SBUF as a destination register. The "write to SBUF" signal also loads a 1 into the 9th bit position of the transmit shift register and flags the TX Control unit that a transmission is requested. Transmission actually commences at S1P1 of the machine cycle following the next rollover in the divide-by-16 counter. (Thus, the bit times are synchronized to the divide-by-16 counter, not to the "write to SBUF" signal.)

The transmission begins with activation of SEND which puts the start bit at TxD. One bit time later, DATA is activated, which enables the output bit of the transmit shift register to TxD. The first shift pulse occurs one bit time after that.

As data bits shift out to the right, zeros are clocked in from the left. When the MSB of the data byte is at the output position of the shift register, then the 1 that was initially loaded into the 9th position is just to the left of the MSB, and all positions to the left of that contain zeros. This condition flags the TX Control unit to do one last shift and then deactivate SEND and set TI. This occurs at the 10th divide-by-16 rollover after "write to SBUF."

Reception is initiated by a detected 1-to-0 transition at RxD. For this purpose RxD is sampled at a rate of 16 times whatever baud rate has been established. When a transition is detected, the divide-by-16 counter is immediately reset, and 1FFH is written into the input shift register. Resetting the divide-by-16 counter aligns its rollovers with the boundaries of the incoming bit times.

The 16 states of the counter divide each bit time into 16ths. At the 7th, 8th, and 9th counter states of each bit time, the bit detector samples the value of RxD. The value accepted is the value that was seen in at least 2 of the 3 samples. This is done for noise rejection. If the value accepted during the first bit time is not 0, the receive circuits are reset and the unit goes back to looking for another 1-to-0 transition. This is to provide rejection of false start bits. If the start bit proves valid, it is shifted into the input shift register, and reception of the rest of the frame will proceed.

As data bits come in from the right, 1s shift out to the left. When the start bit arrives at the leftmost position in the shift register (which in mode 1 is a 9-bit register), it flags the RX Control block to do one last shift, load SBUF and RB8, and set RI. The signal to load SBUF and RB8, and to set RI, will be generated if, and only if, the following conditions are met at the time the final shift pulse is generated.: 1. R1 = 0, and 2. Either SM2 = 0, or the received stop bit = 1.

If either of these two conditions is not met, the received frame is irretrievably lost. If both conditions are met, the stop bit goes into RB8, the 8 data bits go into SBUF, and RI is activated. At this time, whether the above conditions are met or not, the unit goes back to looking for a 1-to-0 transition in RxD.

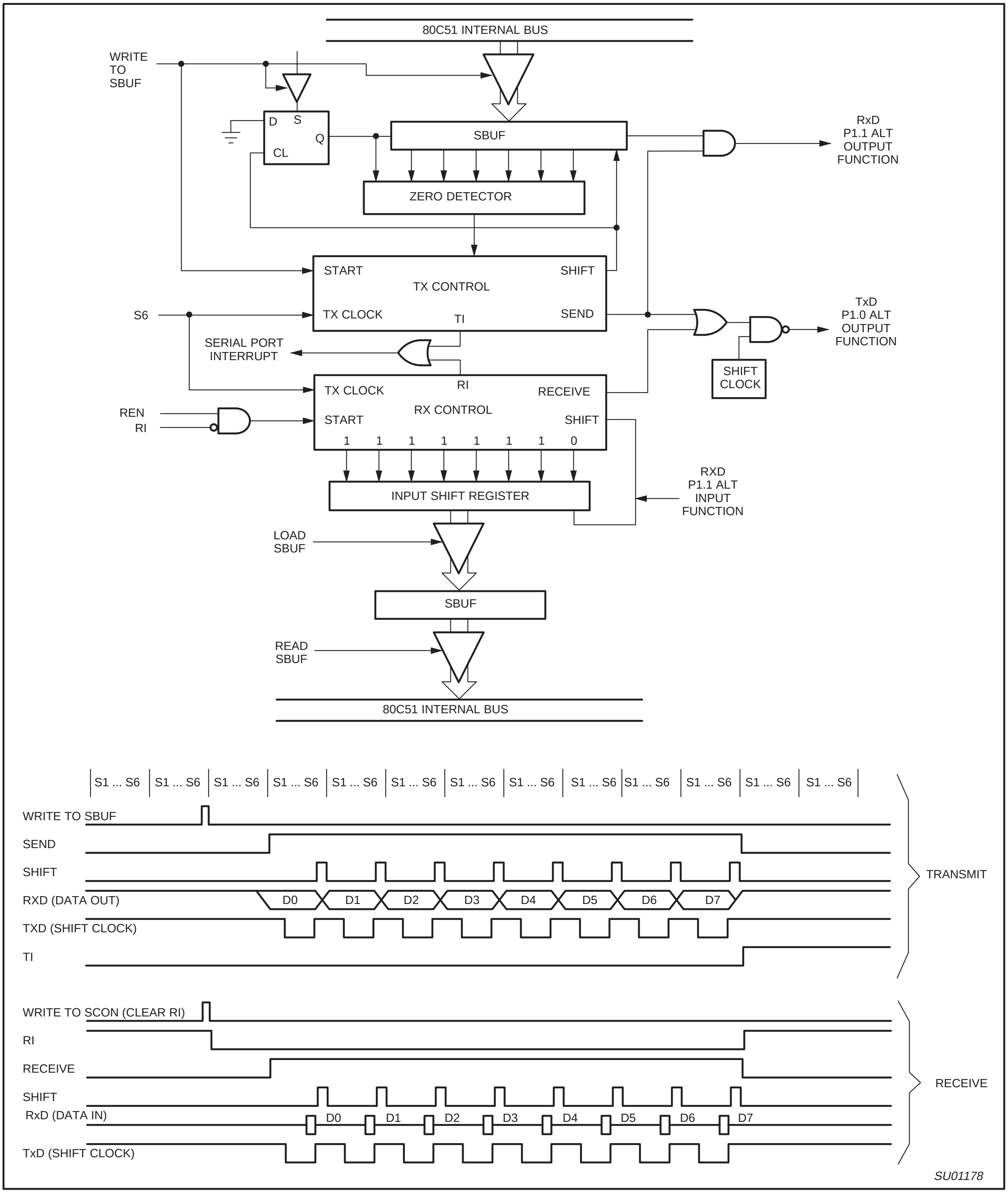


Figure 29. Serial Port Mode 0

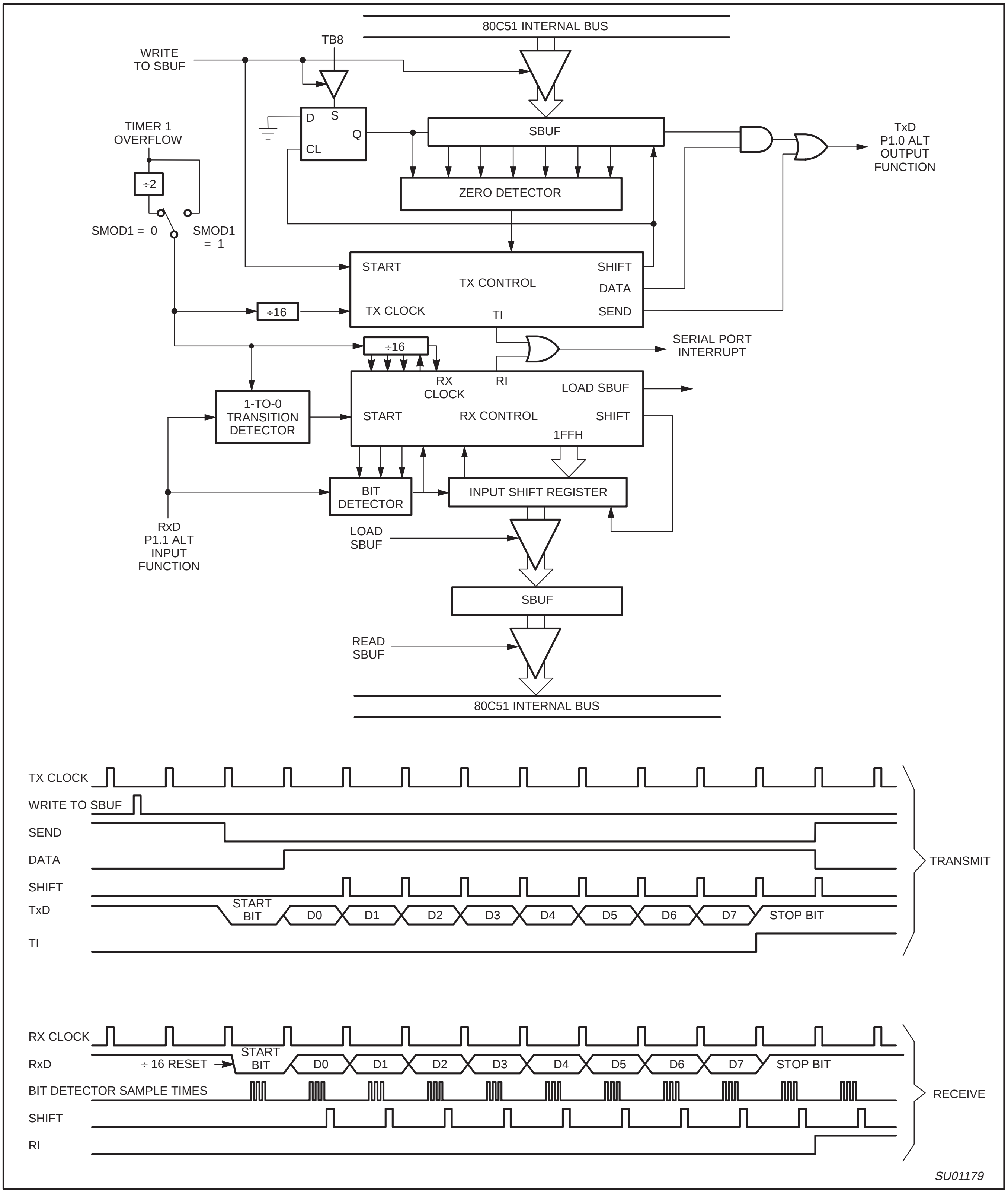


Figure 30. Serial Port Mode 1

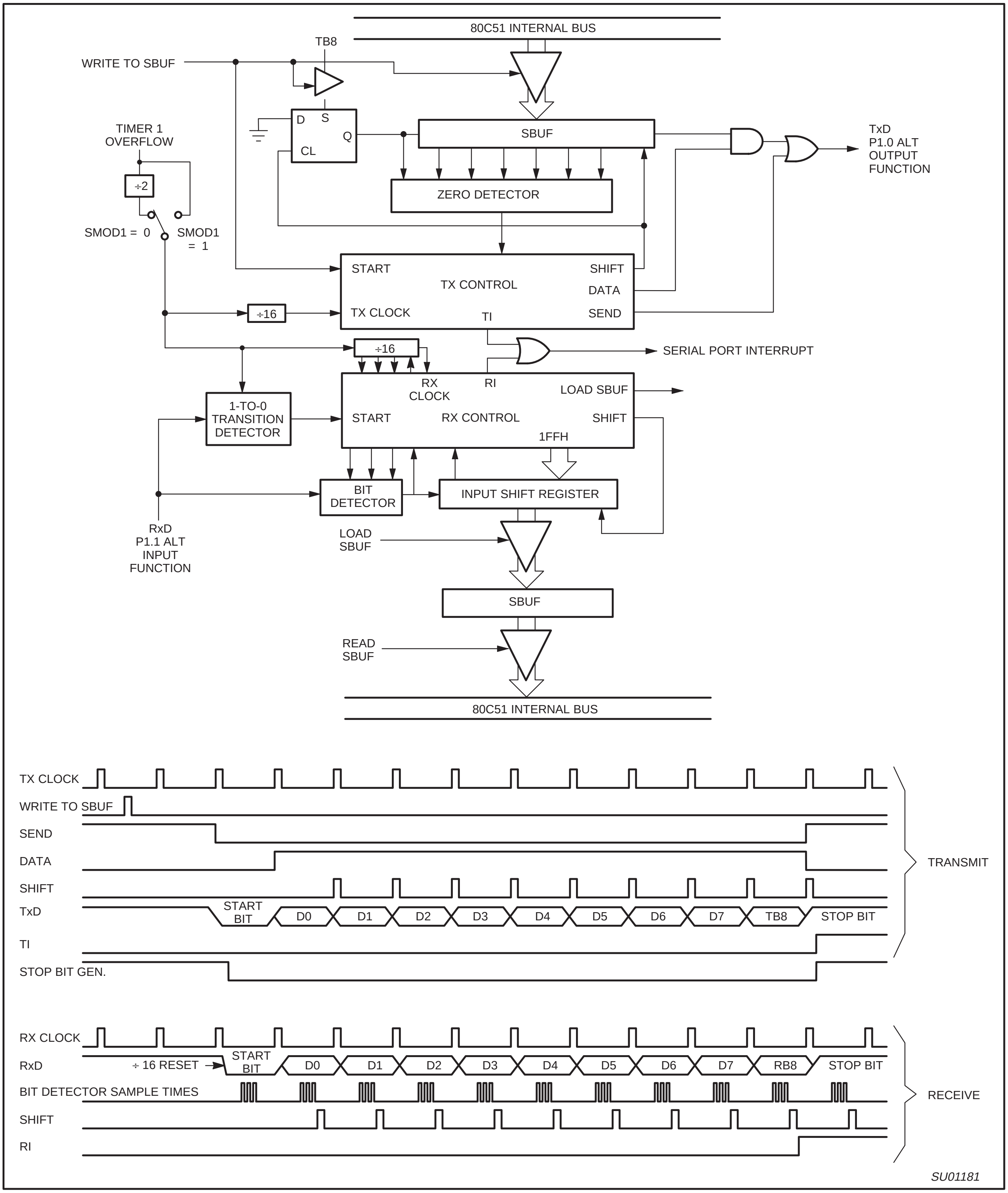


Figure 32. Serial Port Mode 3

EPROM Characteristics

Programming of the EPROM on the P87LPC764 is accomplished with a serial programming method. Commands, addresses, and data are transmitted to and from the device on two pins after programming mode is entered. Serial programming allows easy implementation of In-System Programming of the P87LPC764 in an application board. Details of In-System Programming can be found in application note AN466.

The P87LPC764 contains three signature bytes that can be read and used by an EPROM programming system to identify the device. The signature bytes designate the device as an P87LPC764 manufactured by Philips. The signature bytes may be read by the user program at addresses FC30h, FC31h and FC60h with the MOVC instruction, using the DPTR register for addressing.

A special user data area is also available for access via the MOVC instruction at addresses FCE0h through FCFFh. This “customer code” space is programmed in the same manner as the main code EPROM and may be used to store a serial number, manufacturing date, or other application information.

32-Byte Customer Code Space

A small supplemental EPROM space is reserved for use by the customer in order to identify code revisions, store checksums, add a serial number to each device, or any other desired use. This area exists in the code memory space from addresses FCE0h through FCFFh. Code execution from this space is not supported, but it may be read as data through the use of the MOVC instruction with the appropriate addresses. The memory may be programmed at the same time as the rest of the code memory and UCFG bytes are programmed.

System Configuration Bytes

A number of user configurable features of the P87LPC764 must be defined at power up and therefore cannot be set by the program after start of execution. Those features are configured through the use of two EPROM bytes that are programmed in the same manner as the EPROM program space. The contents of the two configuration bytes, UCFG1 and UCFG2, are shown in Figures 36 and 37. The values of these bytes may be read by the program through the use of the MOVX instruction at the addresses shown in the figure.

UCFG1 Address: FD00h Unprogrammed Value: FFh

7	6	5	4	3	2	1	0
WDTE	RPD	PRHI	BOV	CLKR	FOSC2	FOSC1	FOSC0

BIT	SYMBOL	FUNCTION
UCFG1.7	WDTE	Watchdog timer enable. When programmed (0), disables the watchdog timer. The timer may still be used to generate an interrupt.
UCFG1.6	RPD	Reset pin disable. When 1 disables the reset function of pin P1.5, allowing it to be used as an input only port pin.
UCFG1.5	PRHI	Port reset high. When 1, ports reset to a high state. When 0, ports reset to a low state.
UCFG1.4	BOV	Brownout voltage select. When 1, the brownout detect voltage is 2.5V. When 0, the brownout detect voltage is 3.8V. This is described in the Power Monitoring Functions section.
UCFG1.3	CLKR	Clock rate select. When 0, the CPU clock rate is divided by 2. This results in machine cycles taking 12 CPU clocks to complete as in the standard 80C51. For full backward compatibility, this division applies to peripheral timing as well.
UCFG1.2–0	FOSC2–FSOC0	CPU oscillator type select. See Oscillator section for additional information. Combinations other than those shown below should not be used. They are reserved for future use.

FOSC2–FOSC0	Oscillator Configuration
1 1 1	External clock input on X1 (default setting for an unprogrammed part).
0 1 1	Internal RC oscillator, 6 MHz. For tolerance, see AC Electrical Characteristics table.
0 1 0	Low frequency crystal, 20 kHz to 100 kHz.
0 0 1	Medium frequency crystal or resonator, 100 kHz to 4 MHz.
0 0 0	High frequency crystal or resonator, 4 MHz to 20 MHz.

SU01477

Figure 36. EPROM System Configuration Byte 1 (UCFG1)

UCFG2 Address: FD01h Unprogrammed Value: FFh

7	6	5	4	3	2	1	0
SB2	SB1	—	—	—	—	—	—

BIT	SYMBOL	FUNCTION
UCFG2.7, 6	SB2, SB1	EPROM security bits. See table entitled, “EPROM Security Bits” for details.
UCFG2.5–0	—	Reserved for future use.

SU01186

Figure 37. EPROM System Configuration Byte 2 (UCFG2)

Security Bits

When neither of the security bits are programmed, the code in the EPROM can be verified. When only security bit 1 is programmed, all further programming of the EPROM is disabled. At that point, only security bit 2 may still be programmed. When both security bits are programmed, EPROM verify is also disabled.

Table 11. EPROM Security Bits

SB2	SB1	Protection Description
1	1	Both security bits unprogrammed. No program security features enabled. EPROM is programmable and verifiable.
1	0	Only security bit 1 programmed. Further EPROM programming is disabled. Security bit 2 may still be programmed.
0	1	Only security bit 2 programmed. This combination is not supported.
0	0	Both security bits programmed. All EPROM verification and programming are disabled.

ABSOLUTE MAXIMUM RATINGS

PARAMETER	RATING	UNIT
Operating temperature under bias	–55 to +125	°C
Storage temperature range	–65 to +150	°C
Voltage on $\overline{RST}/V_{PP}$ pin to V_{SS}	0 to +11.0	V
Voltage on any other pin to V_{SS}	–0.5 to V_{DD}+0.5V	V
Maximum I_{OL} per I/O pin	20	mA
Power dissipation (based on package heat transfer, not device power consumption)	1.5	W

NOTES:

1. Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any conditions other than those described in the AC and DC Electrical Characteristics section of this specification are not implied.
2. This product includes circuitry specifically designed for the protection of its internal devices from the damaging effects of excessive static charge. Nonetheless, it is suggested that conventional precautions be taken to avoid applying greater than the rated maximum.
3. Parameters are valid over operating temperature range unless otherwise specified. All voltages are with respect to VSS unless otherwise noted.

DC ELECTRICAL CHARACTERISTICS (FOR P87LPC764BD, BN, BDH, FN, FD, FDH, BD/01, BDH/01)

V_{DD} = 2.7 V to 6.0 V unless otherwise specified; T_{amb} = 0°C to +70°C or –40°C to +85°C, unless otherwise specified.

SYMBOL	PARAMETER	TEST CONDITIONS	LIMITS			UNIT
			MIN	TYP[1]	MAX	
I_{DD}	Power supply current, operating	5.0 V, 20 MHz[11]		15	25	mA
		3.0 V, 10 MHz[11]		4	7	mA
I_{ID}	Power supply current, Idle mode	5.0 V, 20 MHz[11]		6	10	mA
		3.0 V, 10 MHz[11]		2	4	mA
I_{PD}	Power supply current, Power Down mode	5.0 V[11]		1	10	µA
		3.0 V[11]		1	5	µA
V_{RAM}	RAM keep-alive voltage		1.5			V
V_{IL}	Input low voltage (TTL input)	4.0 V < V_{DD} < 6.0 V	–0.5		0.2 V_{DD}–0.1	V
		2.7 V < V_{DD} < 4.0 V	–0.5		0.7	V
V_{IL1}	Negative going threshold (Schmitt input)		–0.5		0.3 V_{DD}	V
V_{IH}	Input high voltage (TTL input)		0.2 V_{DD}+0.9		V_{DD}+0.5	V
V_{IH1}	Positive going threshold (Schmitt input)		0.7 V_{DD}		V_{DD}+0.5	V
HYS	Hysteresis voltage			0.2 V_{DD}		V
V_{OL}	Output low voltage all ports[5, 9]	I_{OL} = 3.2 mA, V_{DD} = 2.7 V			0.4	V
V_{OL1}	Output low voltage all ports[5, 9]	I_{OL} = 20 mA, V_{DD} = 2.7 V			1.0	V
V_{OH}	Output high voltage, all ports[3]	I_{OH} = –20 µA, V_{DD} = 2.7 V	V_{DD}–0.7			V
		I_{OH} = –30 µA, V_{DD} = 4.5 V	V_{DD}–0.7			V
V_{OH1}	Output high voltage, all ports[4]	I_{OH} = –1.0 mA, V_{DD} = 2.7 V	V_{DD}–0.7			V
C_{IO}	Input/Output pin capacitance[10]				15	pF
I_{IL}	Logical 0 input current, all ports[8]	V_{IN} = 0.4 V			–50	µA
I_{LI}	Input leakage current, all ports[7]	V_{IN} = V_{IL} or V_{IH}			±2	µA
I_{TL}	Logical 1 to 0 transition current, all ports[3, 6]	V_{IN} = 1.5 V at V_{DD} = 3.0 V	–30		–250	µA
		V_{IN} = 2.0 V at V_{DD} = 5.5 V	–150		–650	µA
R_{RST}	Internal reset pull-up resistor[14]		40		225	kΩ
V_{BOLOW}	Brownout trip voltage with BOV = 1[12]		2.35		2.69	V
V_{BOHI}	Brownout trip voltage with BOV = 0		3.45		3.99	V
V_{REF}	Reference voltage		1.11	1.26	1.41	V
t_C (V_{REF})	Temperature coefficient			tbd		ppm/°C
SS	Supply sensitivity			tbd		%/V

NOTES:

1. Typical ratings are not guaranteed. The values listed are at room temperature, 5 V.
2. See other Figures for details.
   Active mode: $I_{CC(MAX)}$ = tbd
   Idle mode: $I_{CC(MAX)}$ = tbd
3. Ports in quasi-bidirectional mode with weak pull-up (applies to all port pins with pull-ups). Does not apply to open drain pins.
4. Ports in PUSH-PULL mode. Does not apply to open drain pins.
5. In all output modes except high impedance mode.
6. Port pins source a transition current when used in quasi-bidirectional mode and externally driven from 1 to 0. This current is highest when V_{IN} is approximately 2 V.
7. Measured with port in high impedance mode. Parameter is guaranteed but not tested at cold temperature.
8. Measured with port in quasi-bidirectional mode.
9. Under steady state (non-transient) conditions, I_{OL} must be externally limited as follows:
   Maximum I_{OL} per port pin: 20 mA
   Maximum total I_{OL} for all outputs: 80 mA
   Maximum total I_{OH} for all outputs: 5 mA
   If I_{OL} exceeds the test condition, V_{OL} may exceed the related specification. Pins are not guaranteed to sink current greater than the listed test conditions.
10. Pin capacitance is characterized but not tested.
11. The I_{DD}, I_{ID}, and I_{PD} specifications are measured using an external clock with the following functions disabled: comparators, brownout detect, and watchdog timer. For V_{DD} = 3 V, LPEP = 1. Refer to the appropriate figures on the following pages for additional current drawn by each of these functions and detailed graphs for other frequency and voltage combinations.
12. Devices initially operating at V_{DD} = 2.7V or above and at f_{OSC} = 10 MHz or less are guaranteed to continue to execute instructions correctly at the brownout trip point. Initial power-on operation below V_{DD} = 2.7 V is not guaranteed.
13. Devices initially operating at V_{DD} = 4.0 V or above and at f_{OSC} = 20 MHz or less are guaranteed to continue to execute instructions correctly at the brownout trip point. Initial power-on operation below V_{DD} = 4.0 V and F_{OSC} > 10 MHz is not guaranteed.
14. This internal resistor is disconnected if P1.5 is used as a general purpose input pin instead of the reset pin.

DC ELECTRICAL CHARACTERISTICS (FOR P87LPC764HDH)

V_{DD} = 4.5 V to 5.5 V; T_{amb} = –40°C to +125°C.

SYMBOL	PARAMETER	TEST CONDITIONS	LIMITS			UNIT
			MIN	TYP[1]	MAX	
I_{DD}	Power supply current, operating	5.0 V, 20 MHz[11]		15	25	mA
I_{ID}	Power supply current, Idle mode	5.0 V, 20 MHz[11]		6	10	mA
I_{PD}	Power supply current, Power Down mode	5.0 V[11]		1	10	µA
V_{RAM}	RAM keep-alive voltage		1.5			V
V_{IL}	Input low voltage (TTL input)	4.0 V < V_{DD} < 6.0 V	–0.5		0.2 V_{DD}–0.1	V
V_{IL1}	Negative going threshold (Schmitt input)		–0.5		0.3 V_{DD}	V
V_{IH}	Input high voltage (TTL input)		0.2 V_{DD}+0.9		V_{DD}+0.5	V
V_{IH1}	Positive going threshold (Schmitt input)		0.7 V_{DD}		V_{DD}+0.5	V
HYS	Hysteresis voltage			0.2 V_{DD}		V
V_{OL}	Output low voltage all ports[5, 9]	I_{OL} = 3.2 mA, V_{DD} = 2.7 V			0.4	V
V_{OL1}	Output low voltage all ports[5, 9]	I_{OL} = 20 mA, V_{DD} = 2.7 V			1.0	V
V_{OH}	Output high voltage, all ports[3]	I_{OH} = –30 µA, V_{DD} = 4.5 V	V_{DD}–0.7			V
V_{OH1}	Output high voltage, all ports[4]	I_{OH} = –1.0 mA, V_{DD} = 2.7 V	V_{DD}–0.7			V
C_{IO}	Input/Output pin capacitance[10]				15	pF
I_{IL}	Logical 0 input current, all ports[8]	V_{IN} = 0.4 V			–50	µA
I_{LI}	Input leakage current, all ports[7]	V_{IN} = V_{IL} or V_{IH}			±2	µA
I_{TL}	Logical 1 to 0 transition current, all ports[3, 6]	V_{IN} = 2.0 V at V_{DD} = 5.5 V	–150		–650	µA
R_{RST}	Internal reset pull-up resistor[14]		40		225	kΩ
V_{BOLOW}	Brownout trip voltage with BOV = 1[12]		2.35		2.69	V
V_{BOHI}	Brownout trip voltage with BOV = 0		3.45		3.99	V
V_{REF}	Reference voltage		1.11	1.26	1.41	V
t_C (V_{REF})	Temperature coefficient			tbd		ppm/°C
SS	Supply sensitivity			tbd		%/V

NOTES:
1. Typical ratings are not guaranteed. The values listed are at room temperature, 5 V.
2. See other Figures for details.
   Active mode: $I_{CC(MAX)}$ = tbd
   Idle mode: $I_{CC(MAX)}$ = tbd
3. Ports in quasi-bidirectional mode with weak pull-up (applies to all port pins with pull-ups). Does not apply to open drain pins.
4. Ports in PUSH-PULL mode. Does not apply to open drain pins.
5. In all output modes except high impedance mode.
6. Port pins source a transition current when used in quasi-bidirectional mode and externally driven from 1 to 0. This current is highest when V_{IN} is approximately 2 V.
7. Measured with port in high impedance mode. Parameter is guaranteed but not tested at cold temperature.
8. Measured with port in quasi-bidirectional mode.
9. Under steady state (non-transient) conditions, I_{OL} must be externally limited as follows:
   Maximum I_{OL} per port pin: 20 mA
   Maximum total I_{OL} for all outputs: 80 mA
   Maximum total I_{OH} for all outputs: 5 mA
   If I_{OL} exceeds the test condition, V_{OL} may exceed the related specification. Pins are not guaranteed to sink current greater than the listed test conditions.
10. Pin capacitance is characterized but not tested.
11. The I_{DD}, I_{ID}, and I_{PD} specifications are measured using an external clock with the following functions disabled: comparators, brownout detect, and watchdog timer. For V_{DD} = 3 V, LPEP = 1. Refer to the appropriate figures on the following pages for additional current drawn by each of these functions and detailed graphs for other frequency and voltage combinations.
12. Devices initially operating at V_{DD} = 2.7V or above and at f_{OSC} = 10 MHz or less are guaranteed to continue to execute instructions correctly at the brownout trip point. Initial power-on operation below V_{DD} = 2.7 V is not guaranteed.
13. Devices initially operating at V_{DD} = 4.0 V or above and at f_{OSC} = 20 MHz or less are guaranteed to continue to execute instructions correctly at the brownout trip point. Initial power-on operation below V_{DD} = 4.0 V and F_{OSC} > 10 MHz is not guaranteed.
14. This internal resistor is disconnected if P1.5 is used as a general purpose input pin instead of the reset pin.

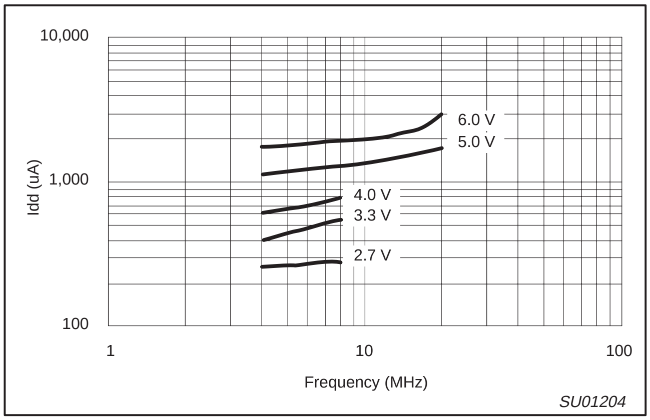


Figure 42. Typical Idd versus frequency (high frequency oscillator, 25°C)

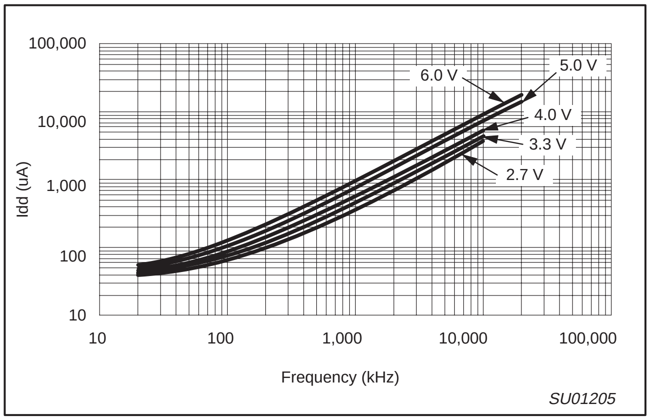


Figure 43. Typical Active Idd versus frequency (external clock, 25°C, LPEP=0)

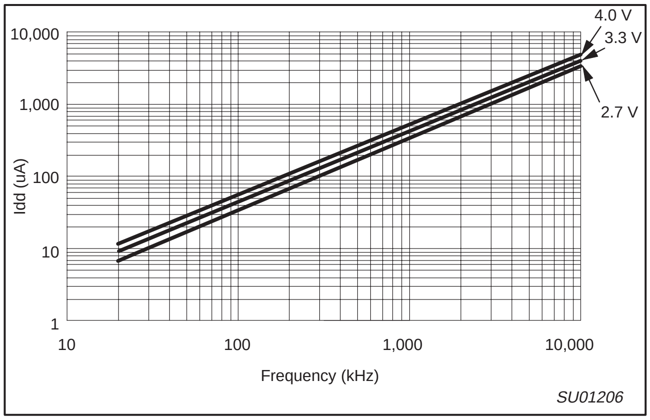


Figure 44. Typical Active Idd versus frequency (external clock, 25°C, LPEP=1)

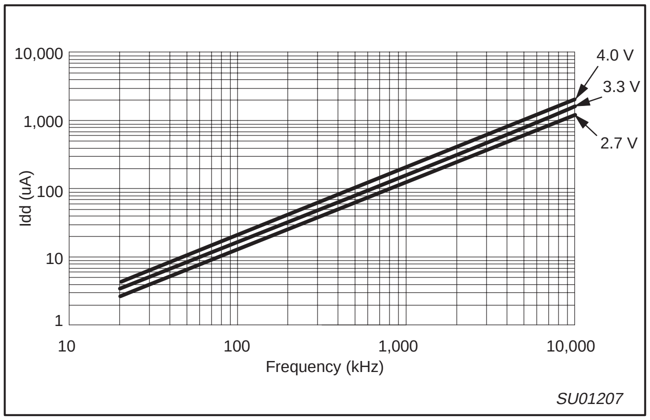


Figure 45. Typical Idle Idd versus frequency (external clock, 25°C, LPEP=1)

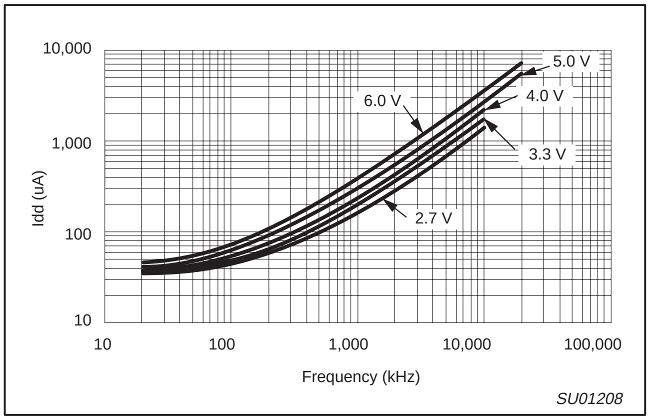
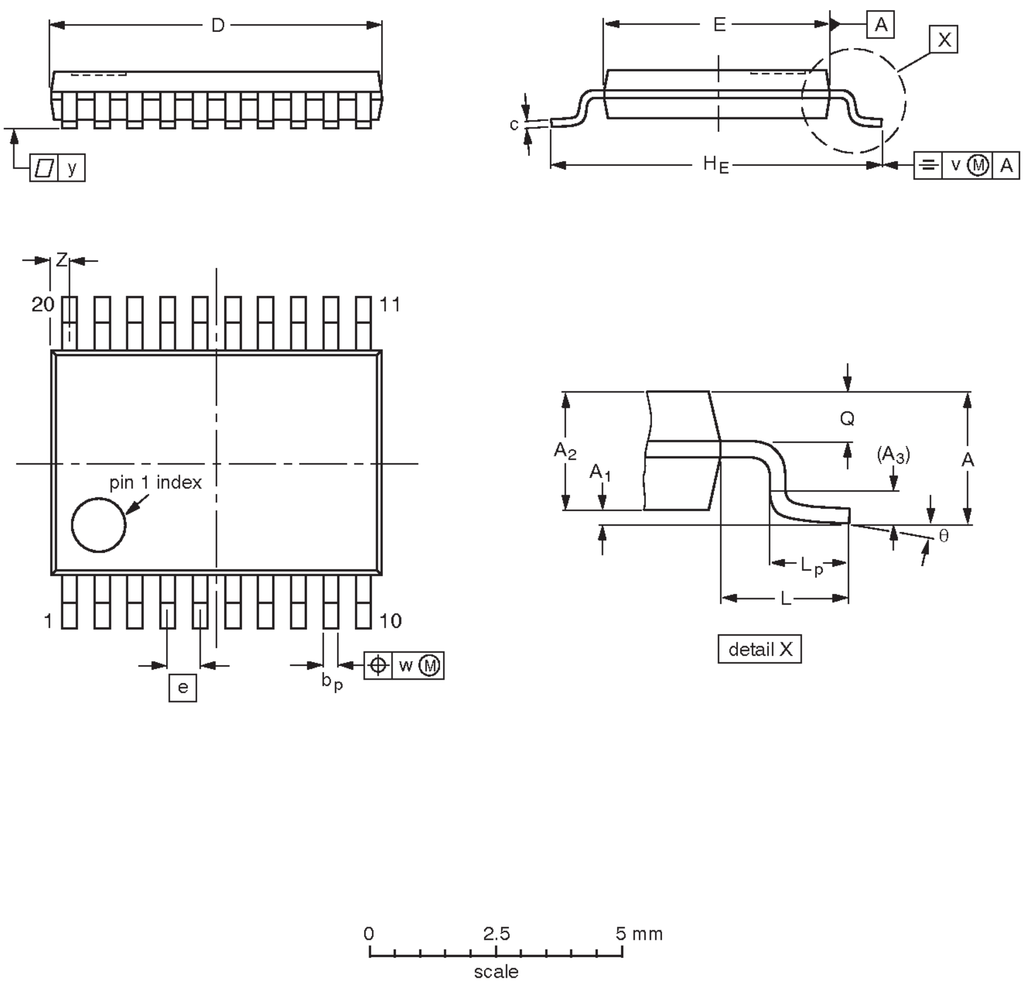


Figure 46. Typical Idle Idd versus frequency (external clock, 25°C, LPEP=0)

TSSOP20: plastic thin shrink small outline package; 20 leads; body width 4.4 mm **SOT360-1**



DIMENSIONS (mm are the original dimensions)

UNIT	A max.	A_1	A_2	A_3	b_p	c	D (1)	E (2)	e	H_E	L	L_p	Q	v	w	y	Z (1)	θ
mm	1.1	0.15 0.05	0.95 0.80	0.25	0.30 0.19	0.2 0.1	6.6 6.4	4.5 4.3	0.65	6.6 6.2	1	0.75 0.50	0.4 0.3	0.2	0.13	0.1	0.5 0.2	8^o 0^o

Notes

1. Plastic or metal protrusions of 0.15 mm maximum per side are not included.
2. Plastic interlead protrusions of 0.25 mm maximum per side are not included.

OUTLINE VERSION	REFERENCES				EUROPEAN PROJECTION	ISSUE DATE
	IEC	JEDEC	JEITA			
SOT360-1		MO-153				~~99-12-27~~ 03-02-19

REVISION HISTORY

Rev	Date	Description
_11	20030903	**Product data (9397 750 11121); ECN 853-2401 30269** Modifications: • Added BD/01, BDH/01 and HDH part types
_10	20011026	**Preliminary data (9397 750 09017); previous release**