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Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Obsolete
Core Processor	MPC8xx
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	66MHz
Co-Processors/DSP	Communications; CPM
RAM Controllers	DRAM
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10Mbps (1), 10/100Mbps (1)
SATA	-
USB	-
Voltage - I/O	3.3V
Operating Temperature	0°C ~ 95°C (TA)
Security Features	-
Package / Case	357-BBGA
Supplier Device Package	357-PBGA (25x25)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/kmpc859dslzp66a

Table 1 shows the functionality supported by the members of the MPC866/859 family.

2 Features

Table 1. MPC866 Family Functionality

Part	Cache		Ethernet		SCC	SMC
	Instruction	Data	10T	10/100		
MPC866P	16 Kbytes	8 Kbytes	Up to 4	1	4	2
MPC866T	4 Kbytes	4 Kbytes	Up to 4	1	4	2
MPC859P	16 Kbytes	8 Kbytes	1	1	1	2
MPC859T	4 Kbytes	4 Kbytes	1	1	1	2
MPC859DSL	4 Kbytes	4 Kbytes	1	1	1 [1]	1 [2]
MPC852T [3]	4 KBytes	4 Kbytes	2	1	2	1

1 On the MPC859DSL, the SCC (SCC1) is for ethernet only. Also, the MPC859DSL does not support the Time Slot Assigner (TSA).

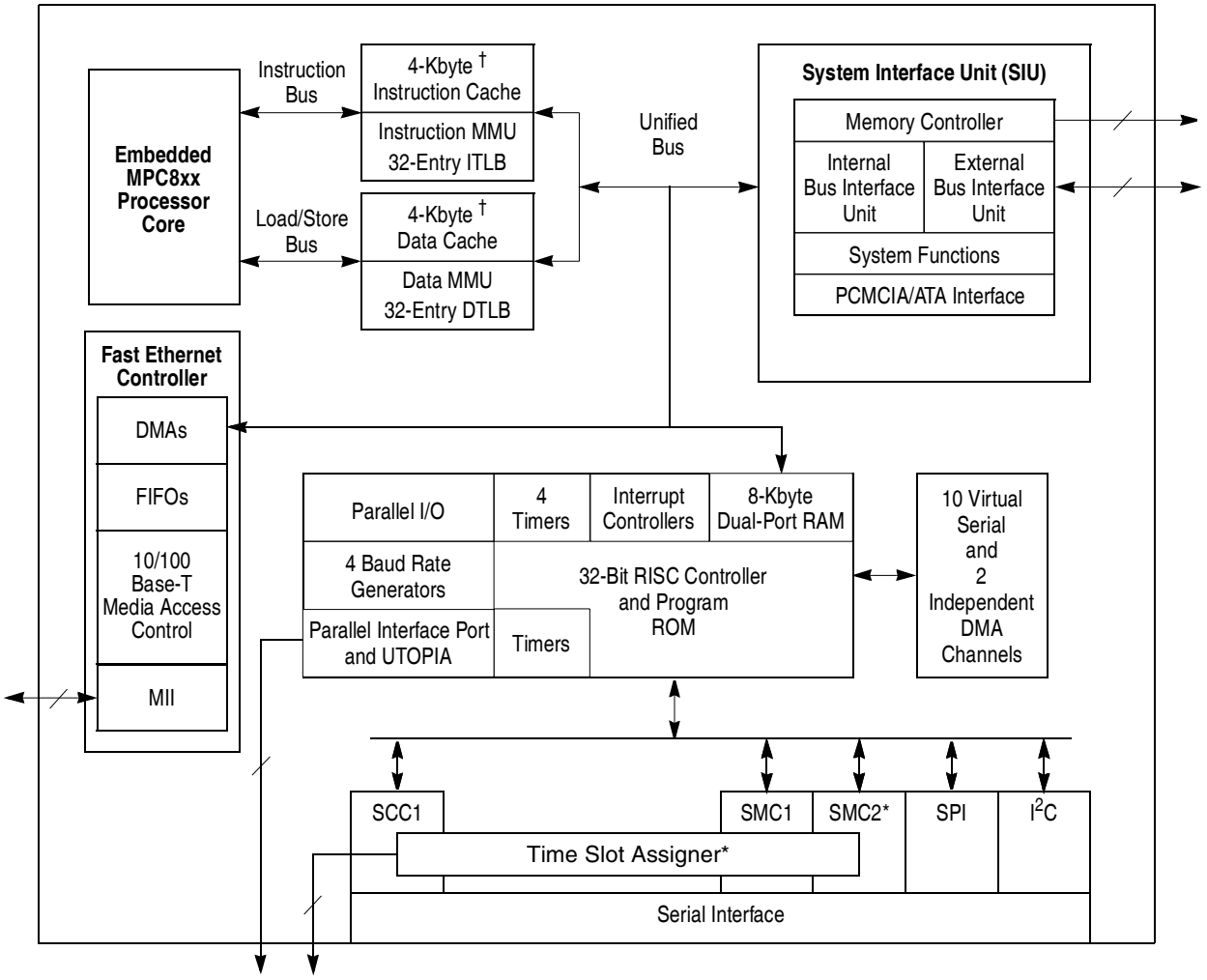
2 On the MPC859DSL, the SMC (SMC1) is for UART only.

3 For more details on the MPC852T, please refer to the *MPC852T Hardware Specifications.*

The following list summarizes the key MPC866/859 features:

- Embedded single-issue, 32-bit PowerPC™ core (implementing the PowerPC architecture) with thirty-two 32-bit general-purpose registers (GPRs)
  - The core performs branch prediction with conditional prefetch, without conditional execution
  - 4- or 8-Kbyte data cache and 4- or 16-Kbyte instruction cache (see Table 1)
    - 16-Kbyte instruction cache (MPC866P and MPC859P) is four-way, set-associative with 256 sets; 4-Kbyte instruction cache (MPC866T, MPC859T, and MPC859DSL) is two-way, set-associative with 128 sets.
    - 8-Kbyte data cache (MPC866P and MPC859P) is two-way, set-associative with 256 sets; 4-Kbyte data cache(MPC866T, MPC859T, and MPC859DSL) is two-way, set-associative with 128 sets.
    - Cache coherency for both instruction and data caches is maintained on 128-bit (4-word) cache blocks
    - Caches are physically addressed, implement a least recently used (LRU) replacement algorithm, and are lockable on a cache block basis.
  - MMUs with 32-entry TLB, fully associative instruction and data TLBs
  - MMUs support multiple page sizes of 4, 16, and 512 Kbytes, and 8 Mbytes; 16 virtual address spaces and 16 protection groups.
  - Advanced on-chip-emulation debug mode
- The MPC866/859 provides enhanced ATM functionality over that of the MPC860SAR. The MPC866/859 adds major new features available in 'enhanced SAR' (ESAR) mode, including the following:
  - Improved operation, administration, and maintenance (OAM) support
  - OAM performance monitoring (PM) support
  - Multiple APC priority levels available to support a range of traffic pace requirements

  - ATM port-to-port switching capability without the need for RAM-based microcode
  - Simultaneous MII (10/100Base-T) and UTOPIA (half-duplex) capability
  - Optional statistical cell counters per PHY
  - UTOPIA level 2 compliant interface with added FIFO buffering to reduce the total cell transmission time. (The earlier UTOPIA level 1 specification is also supported.)
    - Multi-PHY support on the MPC866, MPC859P, and MPC859T
    - Four PHY support on the MPC866/859
  - Parameter RAM for both SPI and I^2C can be relocated without RAM-based microcode
  - Supports full-duplex UTOPIA both master (ATM side) and slave (PHY side) operation using a 'split' bus
  - AAL2/VBR functionality is ROM-resident.
- Up to 32-bit data bus (dynamic bus sizing for 8, 16, and 32 bits)
- Thirty-two address lines
- Memory controller (eight banks)
  - Contains complete dynamic RAM (DRAM) controller
  - Each bank can be a chip select or $\overline{\text{RAS}}$ to support a DRAM bank
  - Up to 30 wait states programmable per memory bank
  - Glueless interface to page mode/EDO/SDRAM, SRAM, EPROMs, flash EPROMs, and other memory devices.
  - DRAM controller programmable to support most size and speed memory interfaces
  - Four $\overline{\text{CAS}}$ lines, four $\overline{\text{WE}}$ lines, and one $\overline{\text{OE}}$ line
  - Boot chip-select available at reset (options for 8-, 16-, or 32-bit memory)
  - Variable block sizes (32 Kbytes–256 Mbytes)
  - Selectable write protection
  - On-chip bus arbitration logic
- General-purpose timers
  - Four 16-bit timers cascadable to be two 32-bit timers
  - Gate mode can enable/disable counting
  - Interrupt can be masked on reference match and event capture
- Fast Ethernet controller (FEC)
  - Simultaneous MII (10/100Base-T) and UTOPIA operation when using the UTOPIA multiplexed bus
- System integration unit (SIU)
  - Bus monitor
  - Software watchdog
  - Periodic interrupt timer (PIT)
  - Low-power stop mode
  - Clock synthesizer
  - Decrementer and time base from the PowerPC architecture
  - Reset controller
  - IEEE 1149.1 test access port (JTAG)



† The MPC859P has a 16-Kbyte instruction cache and a 8-Kbyte data cache.

* The MPC859DSL does not contain SMC2 nor the time slot assigner, and provides eight SDMA controllers.

Figure 2. MPC859P/859T/MPC859DSL Block Diagram

5 Power Dissipation

Table 5 shows power dissipation information. The modes are 1:1, where CPU and bus speeds are equal, and 2:1 mode, where CPU frequency is twice the bus speed.

Table 5. Power Dissipation (P_D)

Die Revision	Bus Mode	CPU Frequency	Typical [1]	Maximum [2]	Unit
0	1:1	50 MHz	110	140	mW
		66 MHz	150	180	mW
	2:1	66 MHz	140	160	mW
		80 MHz	170	200	mW
		100 MHz	210	250	mW
		133 MHz	260	320	mW

1 Typical power dissipation at VDDL and VDDSYN is at 1.8 V. and VDDH is at 3.3 V.

2 Maximum power dissipation at VDDL and VDDSYN is at 1.9 V, and VDDH is at 3.465 V.

NOTE

Values in Table 5 represent VDDL based power dissipation and do not include I/O power dissipation over VDDH. I/O power dissipation varies widely by application due to buffer current, depending on external circuitry. The VDDSYN power dissipation is negligible.

6 DC Characteristics

Table 6 shows the DC electrical characteristics for the MPC866/859.

Table 6. DC Electrical Specifications

Characteristic	Symbol	Min	Max	Unit
Operating voltage	VDDL (core)	1.7	1.9	V
	VDDH (I/O)	3.135	3.465	V
	VDDSYN [1]	1.7	1.9	V
	Difference between VDDL to VDDSYN	—	100	mV
Input high voltage (all inputs except EXTAL and EXTCLK) [2]	VIH	2.0	3.465	V

Table 6. DC Electrical Specifications (continued)

Characteristic	Symbol	Min	Max	Unit
Input low voltage	VIL	GND	0.8	V
EXTAL, EXTCLK input high voltage	VIHC	0.7*(VDDH)	VDDH	V
Input leakage current, Vin = 5.5V (except TMS, TRST, DSCK and DSDI pins) for 5 Volts Tolerant Pins [2]	I_{in}	—	100	µA
Input leakage current, Vin = VDDH (except TMS, TRST, DSCK, and DSDI)	I_{In}	—	10	µA
Input leakage current, Vin = 0 V (except TMS, TRST, DSCK and DSDI pins)	I_{In}	—	10	µA
Input capacitance [3]	C_{in}	—	20	pF
Output high voltage, IOH = – 2.0 mA, except XTAL, and Open drain pins	VOH	2.4	—	V
Output low voltage • IOL = 2.0 mA (CLKOUT) • IOL = 3.2 mA [4] • IOL = 5.3 mA [5] • IOL = 7.0 mA (TXD1/PA14, TXD2/PA12) • IOL = 8.9 mA (TS, TA, TEA, BI, BB, HRESET, SRESET)	VOL	—	0.5	V

1 The difference between VDDL and VDDSYN can not be more than 100 m V.

2 The signals PA[0:15], PB[14:31], PC[4:15], PD[3:15], TDI, TDO, TCK, TRST_B, TMS, MII_TXEN, MII_MDIO are 5 V tolerant.

3 Input capacitance is periodically sampled.

4 A(0:31), TSIZ0/REG, TSIZ1, D(0:31), DP(0:3)/IRQ(3:6), RD/WR, BURST, RSV/IRQ2, IP_B(0:1)/IWP(0:1)/VFLS(0:1), IP_B2/IOIS16_B/AT2, IP_B3/IWP2/VF2, IP_B4/LWP0/VF0, IP_B5/LWP1/VF1, IP_B6/DSDI/AT0, IP_B7/PTR/AT3, RXD1 /PA15, RXD2/PA13, L1TXDB/PA11, L1RXDB/PA10, L1TXDA/PA9, L1RXDA/PA8, TIN1/L1RCLKA/BRGO1/CLK1/PA7, BRGCLK1/TOUT1/CLK2/PA6, TIN2/L1TCLKA/BRGO2/CLK3/PA5, TOUT2/CLK4/PA4, TIN3/BRGO3/CLK5/PA3, BRGCLK2/L1RCLKB/TOUT3/CLK6/PA2, TIN4/BRGO4/CLK7/PA1, L1TCLKB/TOUT4/CLK8/PA0, REJCT1/SPISEL/PB31, SPICLK/PB30, SPIMOSI/PB29, BRGO4/SPIMISO/PB28, BRGO1/I2CSDA/PB27, BRGO2/I2CSCL/PB26, SMTXD1/PB25, SMRXD1/PB24, SMSYN1/SDACK1/PB23, SMSYN2/SDACK2/PB22, SMTXD2/L1CLKOB/PB21, SMRXD2/L1CLKOA/PB20, L1ST1/RTS1/PB19, L1ST2/RTS2/PB18, L1ST3/L1RQB/PB17, L1ST4/L1RQA/PB16, BRGO3/PB15, RSTRT1/PB14, L1ST1/RTS1/DREQ0/PC15, L1ST2/RTS2/DREQ1/PC14, L1ST3/L1RQB/PC13, L1ST4/L1RQA/PC12, CTS1/PC11, TGATE1/CD1/PC10, CTS2/PC9, TGATE2/CD2/PC8, CTS3/SDACK2/L1TSYNCB/PC7, CD3/L1RSYNCB/PC6, CTS4/SDACK1/L1TSYNCA/PC5, CD4/L1RSYNCA/PC4, PD15/L1TSYNCA, PD14/L1RSYNCA, PD13/L1TSYNCB, PD12/L1RSYNCB, PD11/RXD3, PD10/TXD3, PD9/RXD4, PD8/TXD4, PD5/REJECT2, PD6/RTS4, PD7/RTS3, PD4/REJECT3, PD3, MII_MDC, MII_TX_ER, MII_EN, MII_MDIO, MII_TXD[0:3].

5 BDIP/GPL_B(5), BR, BG, FRZ/IRQ6, CS(0:5), CS(6)/CE(1)_B, CS(7)/CE(2)_B, WE0/BS_B0/IORD, WE1/BS_B1/IOWR, WE2/BS_B2/PCOE, WE3/BS_B3/PCWE, BS_A(0:3), GPL_A0/GPL_B0, OE/GPL_A1/GPL_B1, GPL_A(2:3)/GPL_B(2:3)/CS(2:3), UPWAITA/GPL_A4, UPWAITB/GPL_B4, GPL_A5, ALE_A, CE1_A, CE2_A, ALE_B/DSCK/AT1, OP(0:1), OP2/MODCK1/STS, OP3/MODCK2/DSDO, BADDR(28:30).

7 Thermal Calculation and Measurement

For the following discussions, P_D = (VDDL x IDDL) + PI/O, where PI/O is the power dissipation of the I/O drivers. The VDDSYN power dissipation is negligible.

7.1 Estimation with Junction-to-Ambient Thermal Resistance

An estimation of the chip junction temperature, T_J, in °C can be obtained from the equation:

$$T_J = T_A + (R_{\theta JA} \times P_D)$$

where:

T_A = ambient temperature (ºC)

$R_{\theta JA}$ = package junction-to-ambient thermal resistance (ºC/W)

P_D = power dissipation in package

The junction-to-ambient thermal resistance is an industry standard value that provides a quick and easy estimation of thermal performance. However, the answer is only an estimate; test cases have demonstrated that errors of a factor of two (in the quantity T_J-T_A) are possible.

7.2 Estimation with Junction-to-Case Thermal Resistance

Historically, the thermal resistance has frequently been expressed as the sum of a junction-to-case thermal resistance and a case-to-ambient thermal resistance:

$$R_{\theta JA} = R_{\theta JC} + R_{\theta CA}$$

where:

$R_{\theta JA}$ = junction-to-ambient thermal resistance (ºC/W)

$R_{\theta JC}$ = junction-to-case thermal resistance (ºC/W)

$R_{\theta CA}$ = case-to-ambient thermal resistance (ºC/W)

$R_{\theta JC}$ is device related and cannot be influenced by the user. The user adjusts the thermal environment to affect the case-to-ambient thermal resistance, $R_{\theta CA}$. For instance, the user can change the airflow around the device, add a heat sink, change the mounting arrangement on the printed-circuit board, or change the thermal dissipation on the printed-circuit board surrounding the device. This thermal model is most useful for ceramic packages with heat sinks where some 90% of the heat flows through the case and the heat sink to the ambient environment. For most packages, a better model is required.

7.3 Estimation with Junction-to-Board Thermal Resistance

A simple package thermal model that has demonstrated reasonable accuracy (about 20%) is a two-resistor model consisting of a junction-to-board and a junction-to-case thermal resistance. The junction-to-case covers the situation where a heat sink is used or where a substantial amount of heat is dissipated from the top of the package. The junction-to-board thermal resistance describes the thermal performance when most of the heat is conducted to the printed-circuit board. It has been observed that the thermal performance of most plastic packages and especially PBGA packages is strongly dependent on the board temperature; see Figure 3.

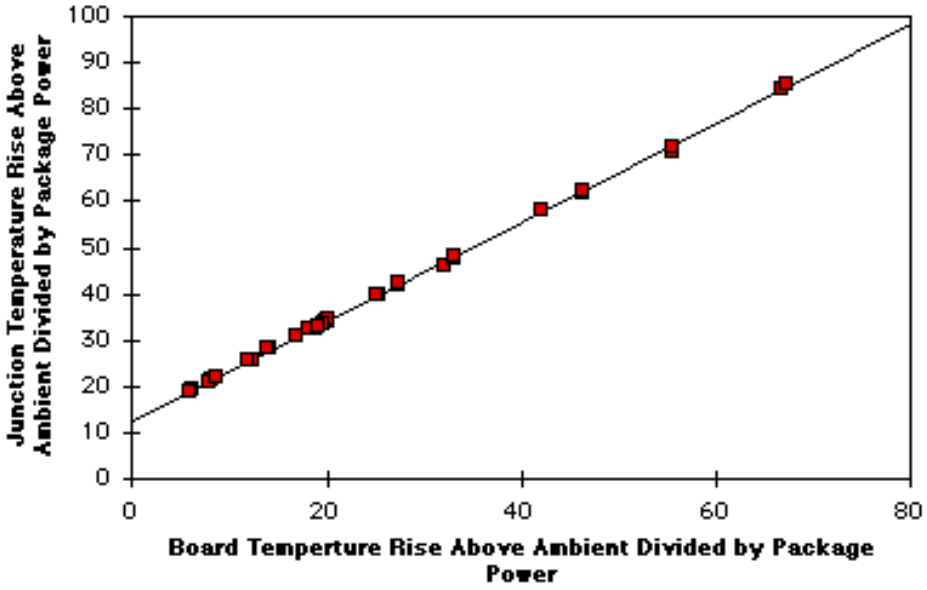


Figure 3. Effect of Board Temperature Rise on Thermal Behavior

If the board temperature is known, an estimate of the junction temperature in the environment can be made using the following equation:

$T_J = T_B + (R_{\theta JB} \times P_D)$

where:

$R_{\theta JB}$ = junction-to-board thermal resistance (ºC/W)

T_B = board temperature ºC

P_D = power dissipation in package

If the board temperature is known and the heat loss from the package case to the air can be ignored, acceptable predictions of junction temperature can be made. For this method to work, the board and board mounting must be similar to the test board used to determine the junction-to-board thermal resistance, namely a 2s2p (board with a power and a ground plane) and vias attaching the thermal balls to the ground plane.

7.4 Estimation Using Simulation

When the board temperature is not known, a thermal simulation of the application is needed. The simple two-resistor model can be used with the thermal simulation of the application [2], or a more accurate and complex model of the package can be used in the thermal simulation.

This recommendation particularly applies to the address and data buses. Maximum PC trace lengths of 6” are recommended. Capacitance calculations should consider all device loads as well as parasitic capacitances due to the PC traces. Attention to proper PCB layout and bypassing becomes especially critical in systems with higher capacitive loads because these loads create higher transient currents in the V_{DD} and GND circuits. Pull up all unused inputs or signals that will be inputs during reset. Special care should be taken to minimize the noise levels on the PLL supply pins. For more information, please refer to Section 14.4.3, Clock Synthesizer Power (VDDSYN, VSSSYN, VSSSYN1), in the *MPC866 User’s Manual*.

10 Bus Signal Timing

The maximum bus speed supported by the MPC866/859 is 66 MHz. Higher-speed parts must be operated in half-speed bus mode (for example, an MPC866/859 used at 100 MHz must be configured for a 50-MHz bus). Table 7 and Table 8 show the frequency ranges for standard part frequencies.

Table 7. Frequency Ranges for Standard Part Frequencies (1:1 Bus Mode)

Part Freq	50 MHz		66 MHz	
	Min	Max	Min	Max
Core	40	50	40	66.67
Bus	40	50	40	66.67

Table 8. Frequency Ranges for Standard Part Frequencies (2:1 Bus Mode)

Part Freq	50 MHz		66 MHz		100 MHz		133 MHz	
	Min	Max	Min	Max	Min	Max	Min	Max
Core	40	50	40	66.67	40	100	40	133.34
Bus	20	25	20	33.33	20	50	20	66.67

Table 9 shows the timings for the MPC866/859 at 33, 40, 50, and 66 MHz bus operation. The timing for the MPC866/859 bus shown in this table assumes a 50-pF load for maximum delays and a 0-pF load for minimum delays. CLKOUT assumes a 100-pF load maximum delay.

Table 9. Bus Operation Timings

Num	Characteristic	33 MHz		40 MHz		50 MHz		66 MHz		Unit
		Min	Max	Min	Max	Min	Max	Min	Max	
B1	Bus Period (CLKOUT) See Table 7	—	—	—	—	—	—	—	—	ns
B1a	EXTCLK to CLKOUT phase skew	– 2	+2	– 2	+2	– 2	+2	– 2	+2	ns
B1b	CLKOUT frequency jitter peak-to-peak	—	1	—	1	—	1	—	1	ns
B1c	Frequency jitter on EXTCLK	—	0.50	—	0.50	—	0.50	—	0.50	%

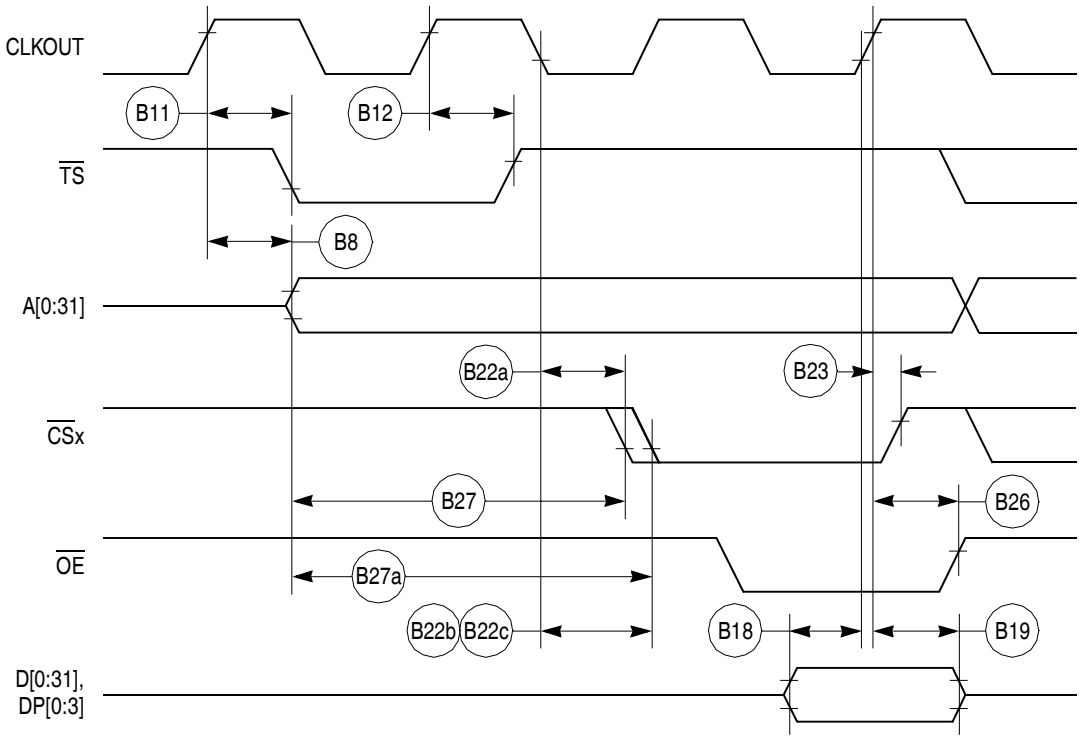


Figure 15. External Bus Read Timing (GPCM Controlled—TRLX = 0 or 1, ACS = 10, ACS = 11)

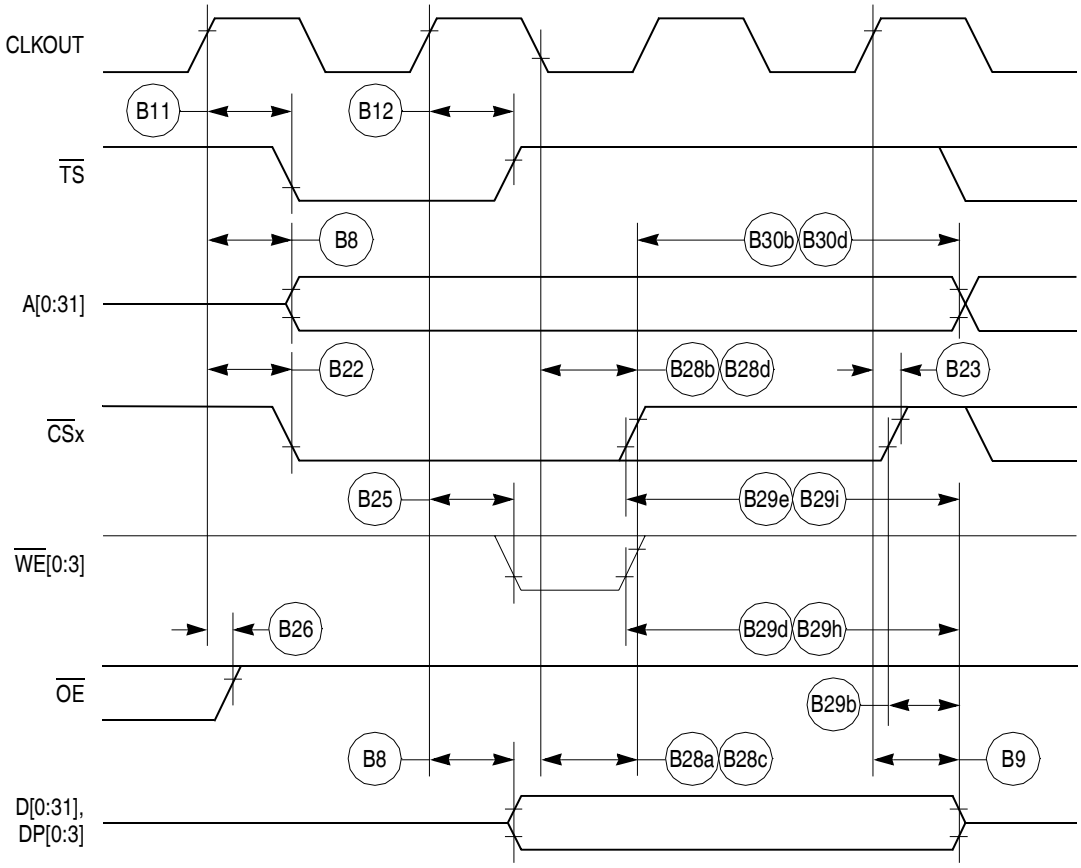


Figure 18. External Bus Write Timing (GPCM Controlled—TRLX = 1, CSNT = 1)

Figure 27 shows the PCMCIA access cycle timing for the external bus read.

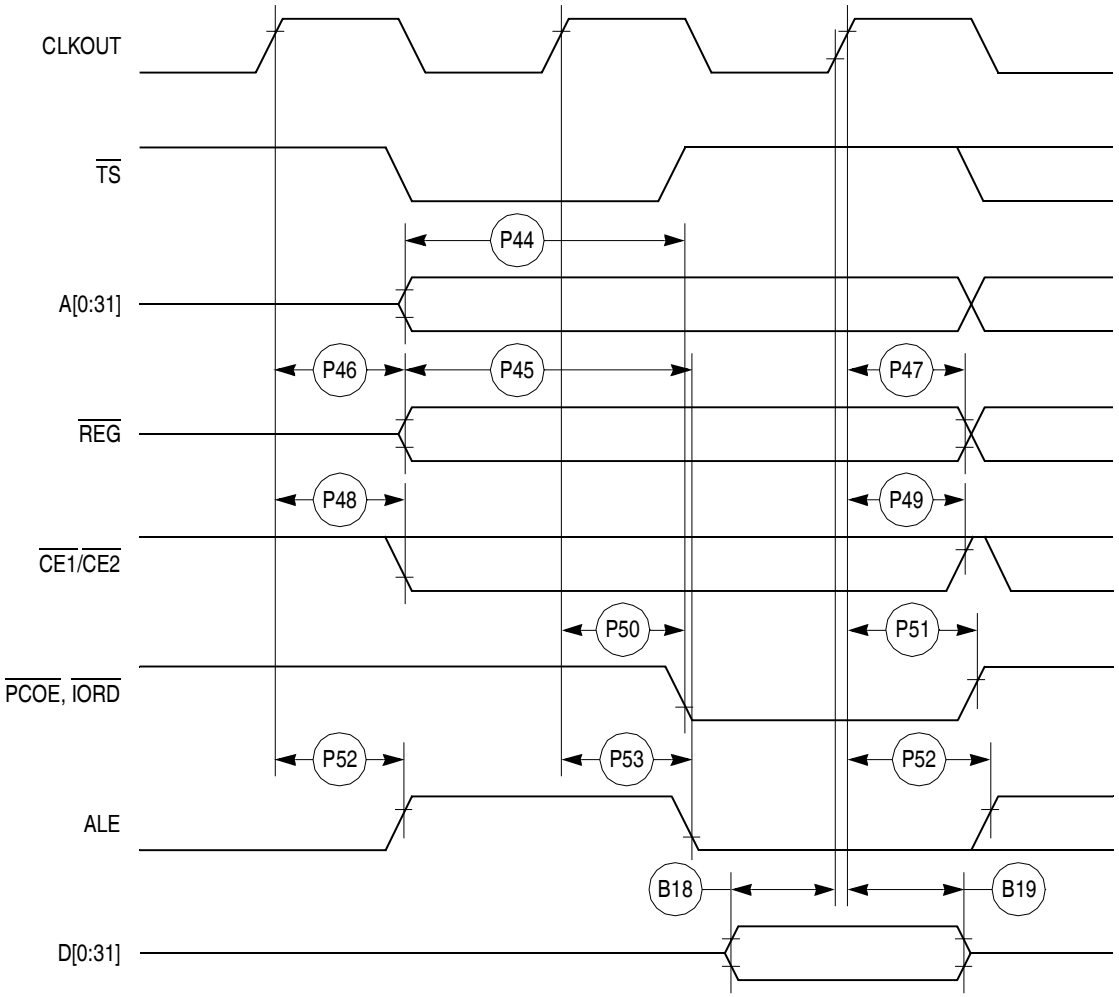


Figure 27. PCMCIA Access Cycles Timing External Bus Read

Table 14 shows the reset timing for the MPC866/859.

Table 14. Reset Timing

Num	Characteristic	33 MHz		40 MHz		50 MHz		66 MHz		Unit
		Min	Max	Min	Max	Min	Max	Min	Max	
R69	CLKOUT to $\overline{\text{HRESET}}$ high impedance (MAX = 0.00 x B1 + 20.00)	—	20.00	—	20.00	—	20.00	—	20.00	ns
R70	CLKOUT to $\overline{\text{SRESET}}$ high impedance (MAX = 0.00 x B1 + 20.00)	—	20.00	—	20.00	—	20.00	—	20.00	ns
R71	$\overline{\text{RSTCONF}}$ pulse width (MIN = 17.00 x B1)	515.20	—	425.00	—	340.00	—	257.60	—	ns
R72	—	—	—	—	—	—	—	—	—	—
R73	Configuration data to HRESET rising edge setup time (MIN = 15.00 x B1 + 50.00)	504.50	—	425.00	—	350.00	—	277.30	—	ns
R74	Configuration data to $\overline{\text{RSTCONF}}$ rising edge setup time (MIN = 0.00 x B1 + 350.00)	350.00	—	350.00	—	350.00	—	350.00	—	ns
R75	Configuration data hold time after $\overline{\text{RSTCONF}}$ negation (MIN = 0.00 x B1 + 0.00)	0.00	—	0.00	—	0.00	—	0.00	—	ns
R76	Configuration data hold time after $\overline{\text{HRESET}}$ negation (MIN = 0.00 x B1 + 0.00)	0.00	—	0.00	—	0.00	—	0.00	—	ns
R77	$\overline{\text{HRESET}}$ and $\overline{\text{RSTCONF}}$ asserted to data out drive (MAX = 0.00 x B1 + 25.00)	—	25.00	—	25.00	—	25.00	—	25.00	ns
R78	$\overline{\text{RSTCONF}}$ negated to data out high impedance (MAX = 0.00 x B1 + 25.00)	—	25.00	—	25.00	—	25.00	—	25.00	ns
R79	CLKOUT of last rising edge before chip three-states $\overline{\text{HRESET}}$ to data out high impedance (MAX = 0.00 x B1 + 25.00)	—	25.00	—	25.00	—	25.00	—	25.00	ns
R80	DSDI, DSCK setup (MIN = 3.00 x B1)	90.90	—	75.00	—	60.00	—	45.50	—	ns
R81	DSDI, DSCK hold time (MIN = 0.00 x B1 + 0.00)	0.00	—	0.00	—	0.00	—	0.00	—	ns
R82	$\overline{\text{SRESET}}$ negated to CLKOUT rising edge for DSDI and DSCK sample (MIN = 8.00 x B1)	242.40	—	200.00	—	160.00	—	121.20	—	ns

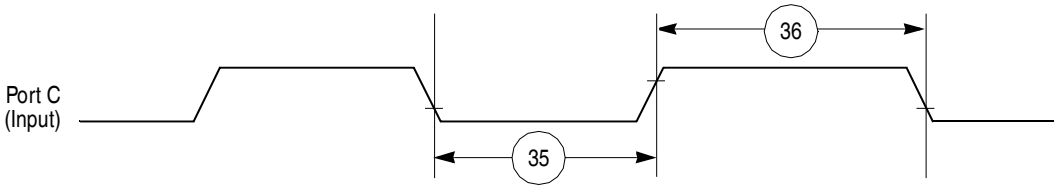


Figure 46. Port C Interrupt Detection Timing

12.3 IDMA Controller AC Electrical Specifications

Table 18 shows the IDMA controller timings as shown in Figure 47 through Figure 50.

Table 18. IDMA Controller Timing

Num	Characteristic	All Frequencies		Unit
		Min	Max	
40	$\overline{\text{DREQ}}$ setup time to clock high	7	—	ns
41	$\overline{\text{DREQ}}$ hold time from clock high	3	—	ns
42	$\overline{\text{SDACK}}$ assertion delay from clock high	—	12	ns
43	$\overline{\text{SDACK}}$ negation delay from clock low	—	12	ns
44	$\overline{\text{SDACK}}$ negation delay from $\overline{\text{TA}}$ low	—	20	ns
45	$\overline{\text{SDACK}}$ negation delay from clock high	—	15	ns
46	$\overline{\text{TA}}$ assertion to falling edge of the clock setup time (applies to external $\overline{\text{TA}}$)	7	—	ns

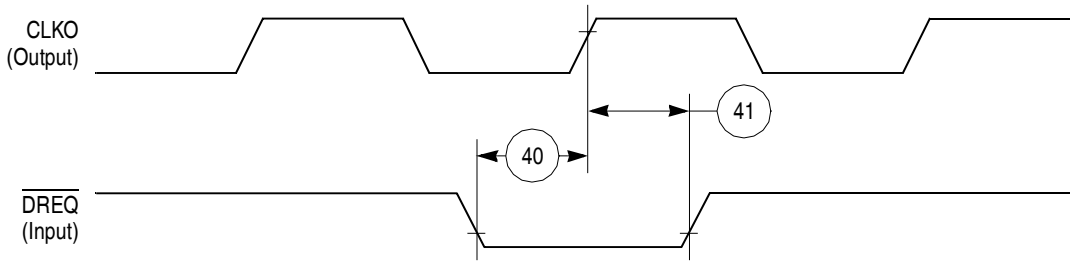


Figure 47. IDMA External Requests Timing Diagram

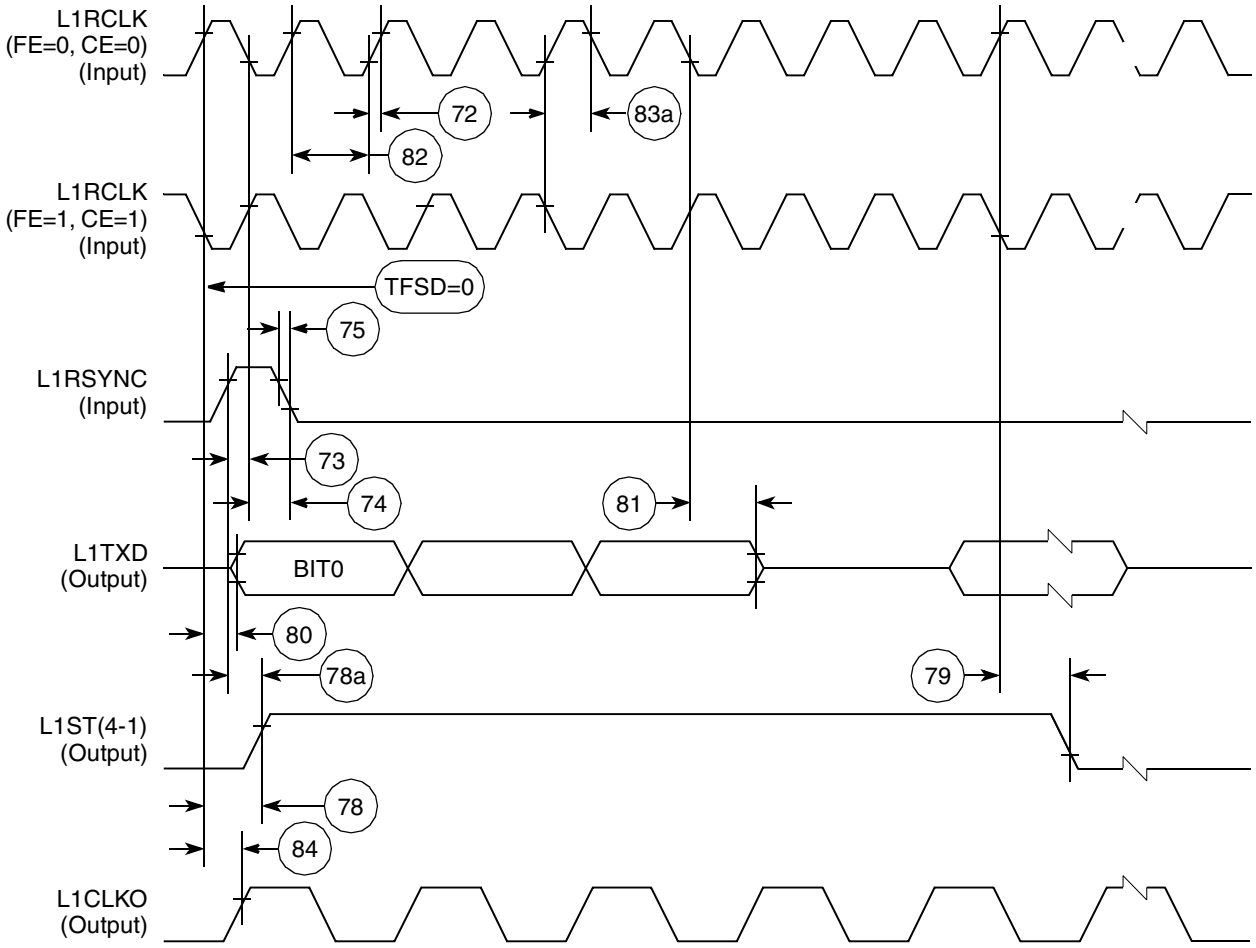


Figure 56. SI Transmit Timing with Double Speed Clocking (DSC = 1)

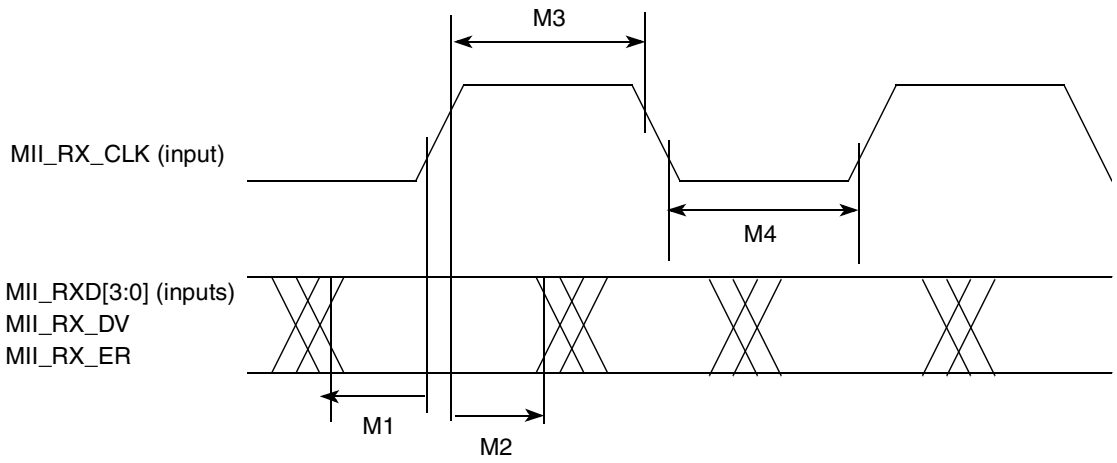


Figure 74. MII Receive Signal Timing Diagram

14.2 MII Transmit Signal Timing (MII_TXD[3:0], MII_TX_EN, MII_TX_ER, MII_TX_CLK)

The transmitter functions correctly up to a MII_TX_CLK maximum frequency of 25 MHz +1%. There is no minimum frequency requirement. In addition, the processor clock frequency must exceed the MII_TX_CLK frequency - 1%.

Table 34 shows information on the MII transmit signal timing.

Table 34. MII Transmit Signal Timing

Num	Characteristic	Min	Max	Unit
M5	MII_TX_CLK to MII_TXD[3:0], MII_TX_EN, MII_TX_ER invalid	5	—	ns
M6	MII_TX_CLK to MII_TXD[3:0], MII_TX_EN, MII_TX_ER valid	—	25	—
M7	MII_TX_CLK pulse width high	35%	65%	MII_TX_CLK period
M8	MII_TX_CLK pulse width low	35%	65%	MII_TX_CLK period

Figure 75 shows the MII transmit signal timing diagram.

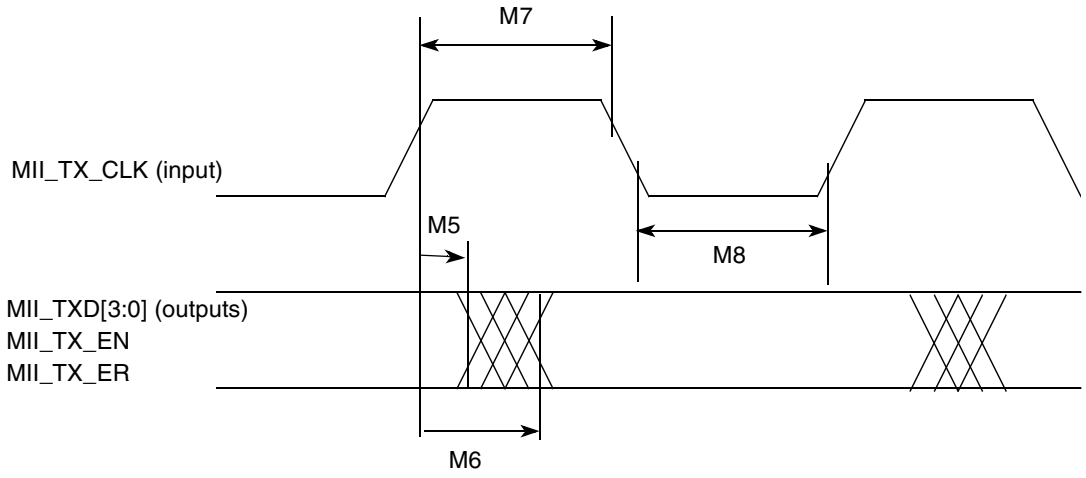


Figure 75. MII Transmit Signal Timing Diagram

14.3 MII Async Inputs Signal Timing (MII_CRS, MII_COL)

Table 35 shows the timing for on the MII async inputs signal.

Table 35. MII Async Inputs Signal Timing

Num	Characteristic	Min	Max	Unit
M9	MII_CRS, MII_COL minimum pulse width	1.5	—	MII_TX_CLK period

Figure 76 shows the MII asynchronous inputs signal timing diagram.

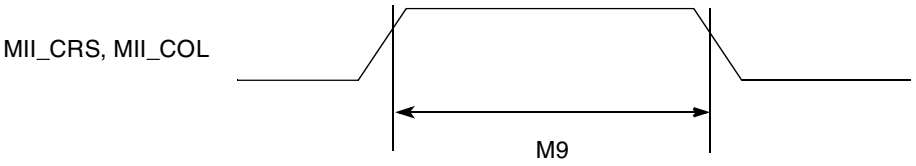


Figure 76. MII Async Inputs Timing Diagram

14.4 MII Serial Management Channel Timing (MII_MDIO, MII_MDC)

Table 36 shows the timing for the MII serial management channel signal. The FEC functions correctly with a maximum MDC frequency in excess of 2.5 MHz. The exact upper bound is under investigation.

Table 36. MII Serial Management Channel Timing

Num	Characteristic	Min	Max	Unit
M10	MII_MDC falling edge to MII_MDIO output invalid (minimum propagation delay)	0	—	ns
M11	MII_MDC falling edge to MII_MDIO output valid (maximum propagation delay)	—	25	ns
M12	MII_MDIO (input) to MII_MDC rising edge setup	10	—	ns

15.1 Pin Assignments

Figure 78 shows the top view pinout of the PBGA package. For additional information, see the *MPC866 PowerQUICC Family User's Manual*.

NOTE: This is the top view of the device.

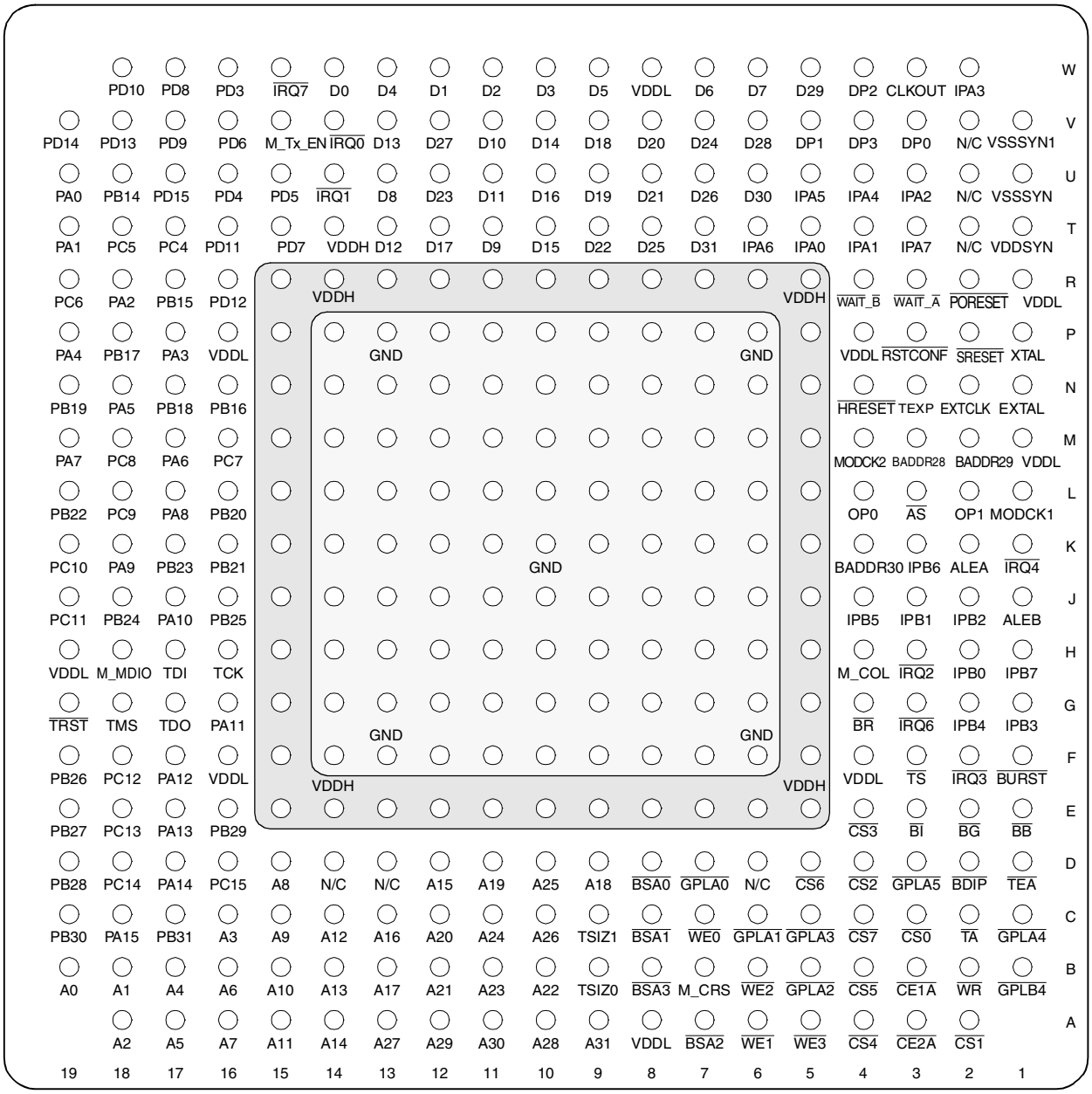


Figure 78. Pinout of the PBGA Package

Table 39. Pin Assignments (continued)

Name	Pin Number	Type
PA5 CLK3 L1TCLKA BRGO2 TIN2	N18	Bidirectional
PA4 CLK4 $\overline{\text{TOUT2}}$	P19	Bidirectional
PA3 CLK5 BRGO3 TIN3	P17	Bidirectional
PA2 CLK6 $\overline{\text{TOUT3}}$ L1RCLKB	R18	Bidirectional
PA1 CLK7 BRGO4 TIN4	T19	Bidirectional
PA0 CLK8 $\overline{\text{TOUT4}}$ L1TCLKB	U19	Bidirectional
PB31 $\overline{\text{SPISEL}}$ $\overline{\text{REJECT1}}$	C17	Bidirectional (Optional: Open-drain)
PB30 SPICLK $\overline{\text{RSTRT2}}$	C19	Bidirectional (Optional: Open-drain)
PB29 SPIMOSI	E16	Bidirectional (Optional: Open-drain)
PB28 SPIMISO BRGO4	D19	Bidirectional (Optional: Open-drain)
PB27 I2CSDA BRGO1	E19	Bidirectional (Optional: Open-drain)
PB26 I2CSCL BRGO2	F19	Bidirectional (Optional: Open-drain)

Table 39. Pin Assignments (continued)

Name	Pin Number	Type
PB25 RXADDR3[2] SMTXD1	J16	Bidirectional (Optional: Open-drain)
PB24 TXADDR3[2] SMRXD1	J18	Bidirectional (Optional: Open-drain)
PB23 TXADDR2[2] $\overline{\text{SDACK1}}$ $\overline{\text{SMSYN1}}$	K17	Bidirectional (Optional: Open-drain)
PB22 TXADDR4[2] $\overline{\text{SDACK2}}$ $\overline{\text{SMSYN2}}$	L19	Bidirectional (Optional: Open-drain)
PB21 SMTXD2 L1CLKOB PHSEL1[1] TXADDR1[2]	K16	Bidirectional (Optional: Open-drain)
PB20 SMRXD2 L1CLKOA PHSEL0[1] TXADDR0[2]	L16	Bidirectional (Optional: Open-drain)
PB19 $\overline{\text{RTS1}}$ L1ST1	N19	Bidirectional (Optional: Open-drain)
PB18 RXADDR4[2] $\overline{\text{RTS2}}$ L1ST2	N17	Bidirectional (Optional: Open-drain)
PB17 $\overline{\text{L1RQb}}$ L1ST3 $\overline{\text{RTS3}}$ PHREQ1[1] RXADDR1[2]	P18	Bidirectional (Optional: Open-drain)

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