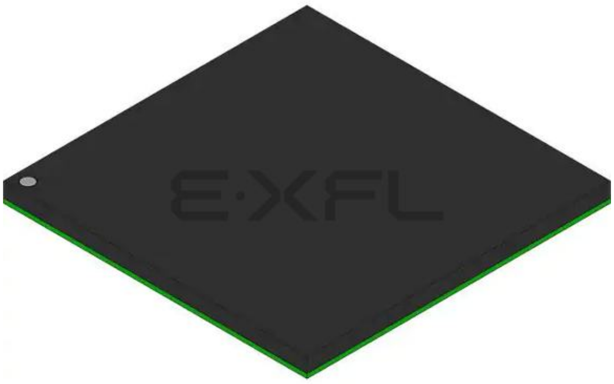

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Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Obsolete
Core Processor	MPC8xx
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	133MHz
Co-Processors/DSP	Communications; CPM
RAM Controllers	DRAM
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10Mbps (1), 10/100Mbps (1)
SATA	-
USB	-
Voltage - I/O	3.3V
Operating Temperature	0°C ~ 95°C (TA)
Security Features	-
Package / Case	357-BBGA
Supplier Device Package	357-PBGA (25x25)
Purchase URL	https://www.e-xfl.com/pro/item?MUrl=&PartUrl=mpc859pzp133a

Table 1 shows the functionality supported by the members of the MPC866/859 family.

2 Features

Table 1. MPC866 Family Functionality

Part	Cache		Ethernet		SCC	SMC
	Instruction	Data	10T	10/100		
MPC866P	16 Kbytes	8 Kbytes	Up to 4	1	4	2
MPC866T	4 Kbytes	4 Kbytes	Up to 4	1	4	2
MPC859P	16 Kbytes	8 Kbytes	1	1	1	2
MPC859T	4 Kbytes	4 Kbytes	1	1	1	2
MPC859DSL	4 Kbytes	4 Kbytes	1	1	1 [1]	1 [2]
MPC852T [3]	4 KBytes	4 Kbytes	2	1	2	1

[1] On the MPC859DSL, the SCC (SCC1) is for ethernet only. Also, the MPC859DSL does not support the Time Slot Assigner (TSA).

[2] On the MPC859DSL, the SMC (SMC1) is for UART only.

[3] For more details on the MPC852T, please refer to the *MPC852T Hardware Specifications.*

The following list summarizes the key MPC866/859 features:

- Embedded single-issue, 32-bit PowerPC™ core (implementing the PowerPC architecture) with thirty-two 32-bit general-purpose registers (GPRs)
  - The core performs branch prediction with conditional prefetch, without conditional execution
  - 4- or 8-Kbyte data cache and 4- or 16-Kbyte instruction cache (see Table 1)
    - 16-Kbyte instruction cache (MPC866P and MPC859P) is four-way, set-associative with 256 sets; 4-Kbyte instruction cache (MPC866T, MPC859T, and MPC859DSL) is two-way, set-associative with 128 sets.
    - 8-Kbyte data cache (MPC866P and MPC859P) is two-way, set-associative with 256 sets; 4-Kbyte data cache(MPC866T, MPC859T, and MPC859DSL) is two-way, set-associative with 128 sets.
    - Cache coherency for both instruction and data caches is maintained on 128-bit (4-word) cache blocks
    - Caches are physically addressed, implement a least recently used (LRU) replacement algorithm, and are lockable on a cache block basis.
  - MMUs with 32-entry TLB, fully associative instruction and data TLBs
  - MMUs support multiple page sizes of 4, 16, and 512 Kbytes, and 8 Mbytes; 16 virtual address spaces and 16 protection groups.
  - Advanced on-chip-emulation debug mode
- The MPC866/859 provides enhanced ATM functionality over that of the MPC860SAR. The MPC866/859 adds major new features available in 'enhanced SAR' (ESAR) mode, including the following:
  - Improved operation, administration, and maintenance (OAM) support
  - OAM performance monitoring (PM) support
  - Multiple APC priority levels available to support a range of traffic pace requirements

The MPC866/859 is comprised of three modules that each use a 32-bit internal bus: MPC8xx core, system integration unit (SIU), and communication processor module (CPM). The MPC866P block diagram is shown in Figure 1. The MPC859P/859T/859DSL block diagram is shown in Figure 2.

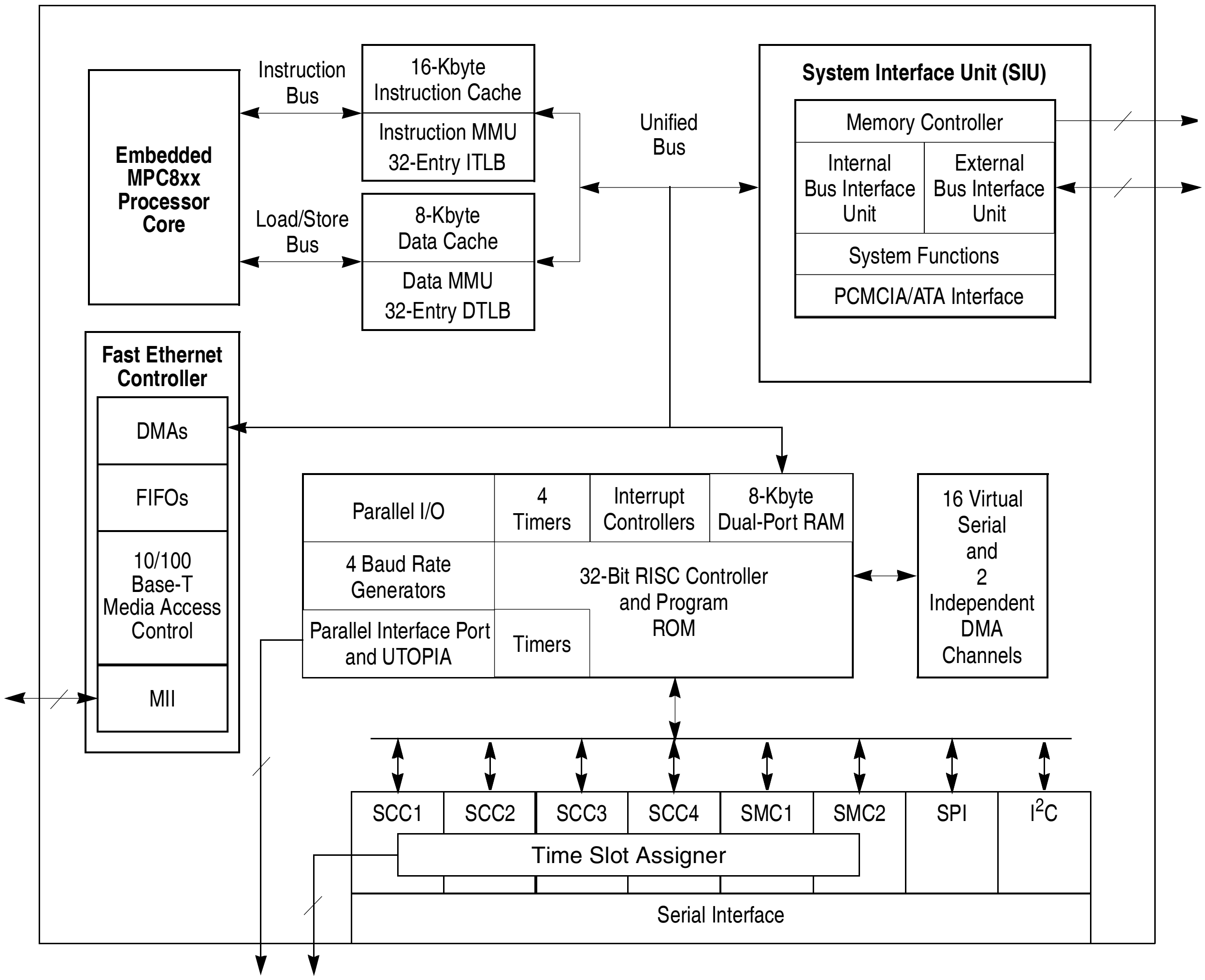

Figure 1. MPC866P Block Diagram

Table 6. DC Electrical Specifications (continued)

Characteristic	Symbol	Min	Max	Unit
Input low voltage	VIL	GND	0.8	V
EXTAL, EXTCLK input high voltage	VIHC	0.7*(VDDH)	VDDH	V
Input leakage current, Vin = 5.5V (except TMS, TRST, DSCK and DSDI pins) for 5 Volts Tolerant Pins [2]	I_{in}	—	100	µA
Input leakage current, Vin = VDDH (except TMS, TRST, DSCK, and DSDI)	I_{In}	—	10	µA
Input leakage current, Vin = 0 V (except TMS, TRST, DSCK and DSDI pins)	I_{In}	—	10	µA
Input capacitance [3]	C_{in}	—	20	pF
Output high voltage, IOH = – 2.0 mA, except XTAL, and Open drain pins	VOH	2.4	—	V
Output low voltage • IOL = 2.0 mA (CLKOUT) • IOL = 3.2 mA [4] • IOL = 5.3 mA [5] • IOL = 7.0 mA (TXD1/PA14, TXD2/PA12) • IOL = 8.9 mA (TS, TA, TEA, BI, BB, HRESET, SRESET)	VOL	—	0.5	V

1 The difference between VDDL and VDDSYN can not be more than 100 m V.

2 The signals PA[0:15], PB[14:31], PC[4:15], PD[3:15], TDI, TDO, TCK, TRST_B, TMS, MII_TXEN, MII_MDIO are 5 V tolerant.

3 Input capacitance is periodically sampled.

4 A(0:31), TSIZ0/REG, TSIZ1, D(0:31), DP(0:3)/IRQ(3:6), RD/WR, BURST, RSV/IRQ2, IP_B(0:1)/IWP(0:1)/VFLS(0:1), IP_B2/IOIS16_B/AT2, IP_B3/IWP2/VF2, IP_B4/LWP0/VF0, IP_B5/LWP1/VF1, IP_B6/DSDI/AT0, IP_B7/PTR/AT3, RXD1 /PA15, RXD2/PA13, L1TXDB/PA11, L1RXDB/PA10, L1TXDA/PA9, L1RXDA/PA8, TIN1/L1RCLKA/BRGO1/CLK1/PA7, BRGCLK1/TOUT1/CLK2/PA6, TIN2/L1TCLKA/BRGO2/CLK3/PA5, TOUT2/CLK4/PA4, TIN3/BRGO3/CLK5/PA3, BRGCLK2/L1RCLKB/TOUT3/CLK6/PA2, TIN4/BRGO4/CLK7/PA1, L1TCLKB/TOUT4/CLK8/PA0, REJCT1/SPISEL/PB31, SPICLK/PB30, SPIMOSI/PB29, BRGO4/SPIMISO/PB28, BRGO1/I2CSDA/PB27, BRGO2/I2CSCL/PB26, SMTXD1/PB25, SMRXD1/PB24, SMSYN1/SDACK1/PB23, SMSYN2/SDACK2/PB22, SMTXD2/L1CLKOB/PB21, SMRXD2/L1CLKOA/PB20, L1ST1/RTS1/PB19, L1ST2/RTS2/PB18, L1ST3/L1RQB/PB17, L1ST4/L1RQA/PB16, BRGO3/PB15, RSTRT1/PB14, L1ST1/RTS1/DREQ0/PC15, L1ST2/RTS2/DREQ1/PC14, L1ST3/L1RQB/PC13, L1ST4/L1RQA/PC12, CTS1/PC11, TGATE1/CD1/PC10, CTS2/PC9, TGATE2/CD2/PC8, CTS3/SDACK2/L1TSYNCB/PC7, CD3/L1RSYNCB/PC6, CTS4/SDACK1/L1TSYNCA/PC5, CD4/L1RSYNCA/PC4, PD15/L1TSYNCA, PD14/L1RSYNCA, PD13/L1TSYNCB, PD12/L1RSYNCB, PD11/RXD3, PD10/TXD3, PD9/RXD4, PD8/TXD4, PD5/REJECT2, PD6/RTS4, PD7/RTS3, PD4/REJECT3, PD3, MII_MDC, MII_TX_ER, MII_EN, MII_MDIO, MII_TXD[0:3].

5 BDIP/GPL_B(5), BR, BG, FRZ/IRQ6, CS(0:5), CS(6)/CE(1)_B, CS(7)/CE(2)_B, WE0/BS_B0/IORD, WE1/BS_B1/IOWR, WE2/BS_B2/PCOE, WE3/BS_B3/PCWE, BS_A(0:3), GPL_A0/GPL_B0, OE/GPL_A1/GPL_B1, GPL_A(2:3)/GPL_B(2:3)/CS(2:3), UPWAITA/GPL_A4, UPWAITB/GPL_B4, GPL_A5, ALE_A, CE1_A, CE2_A, ALE_B/DSCK/AT1, OP(0:1), OP2/MODCK1/STS, OP3/MODCK2/DSDO, BADDR(28:30).

7 Thermal Calculation and Measurement

For the following discussions, P_D = (VDDL x IDDL) + PI/O, where PI/O is the power dissipation of the I/O drivers. The VDDSYN power dissipation is negligible.

7.1 Estimation with Junction-to-Ambient Thermal Resistance

An estimation of the chip junction temperature, T_J, in °C can be obtained from the equation:

$$T_J = T_A + (R_{\theta JA} \times P_D)$$

where:

T_A = ambient temperature (ºC)

$R_{\theta JA}$ = package junction-to-ambient thermal resistance (ºC/W)

P_D = power dissipation in package

The junction-to-ambient thermal resistance is an industry standard value that provides a quick and easy estimation of thermal performance. However, the answer is only an estimate; test cases have demonstrated that errors of a factor of two (in the quantity T_J-T_A) are possible.

7.2 Estimation with Junction-to-Case Thermal Resistance

Historically, the thermal resistance has frequently been expressed as the sum of a junction-to-case thermal resistance and a case-to-ambient thermal resistance:

$$R_{\theta JA} = R_{\theta JC} + R_{\theta CA}$$

where:

$R_{\theta JA}$ = junction-to-ambient thermal resistance (ºC/W)

$R_{\theta JC}$ = junction-to-case thermal resistance (ºC/W)

$R_{\theta CA}$ = case-to-ambient thermal resistance (ºC/W)

$R_{\theta JC}$ is device related and cannot be influenced by the user. The user adjusts the thermal environment to affect the case-to-ambient thermal resistance, $R_{\theta CA}$. For instance, the user can change the airflow around the device, add a heat sink, change the mounting arrangement on the printed-circuit board, or change the thermal dissipation on the printed-circuit board surrounding the device. This thermal model is most useful for ceramic packages with heat sinks where some 90% of the heat flows through the case and the heat sink to the ambient environment. For most packages, a better model is required.

7.3 Estimation with Junction-to-Board Thermal Resistance

A simple package thermal model that has demonstrated reasonable accuracy (about 20%) is a two-resistor model consisting of a junction-to-board and a junction-to-case thermal resistance. The junction-to-case covers the situation where a heat sink is used or where a substantial amount of heat is dissipated from the top of the package. The junction-to-board thermal resistance describes the thermal performance when most of the heat is conducted to the printed-circuit board. It has been observed that the thermal performance of most plastic packages and especially PBGA packages is strongly dependent on the board temperature; see Figure 3.

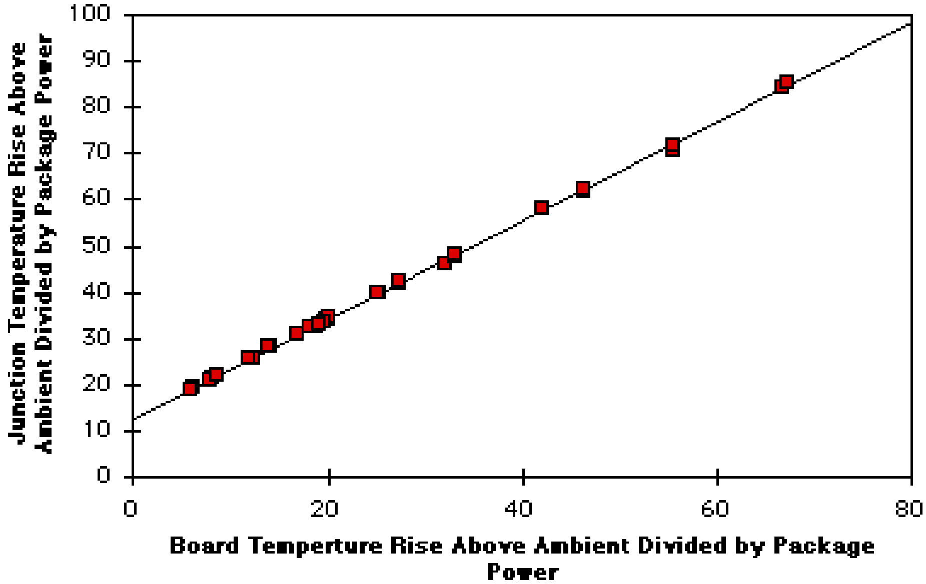

Figure 3. Effect of Board Temperature Rise on Thermal Behavior

If the board temperature is known, an estimate of the junction temperature in the environment can be made using the following equation:

$T_J = T_B + (R_{\theta JB} \times P_D)$

where:

$R_{\theta JB}$ = junction-to-board thermal resistance (ºC/W)

T_B = board temperature ºC

P_D = power dissipation in package

If the board temperature is known and the heat loss from the package case to the air can be ignored, acceptable predictions of junction temperature can be made. For this method to work, the board and board mounting must be similar to the test board used to determine the junction-to-board thermal resistance, namely a 2s2p (board with a power and a ground plane) and vias attaching the thermal balls to the ground plane.

7.4 Estimation Using Simulation

When the board temperature is not known, a thermal simulation of the application is needed. The simple two-resistor model can be used with the thermal simulation of the application [2], or a more accurate and complex model of the package can be used in the thermal simulation.

This recommendation particularly applies to the address and data buses. Maximum PC trace lengths of 6” are recommended. Capacitance calculations should consider all device loads as well as parasitic capacitances due to the PC traces. Attention to proper PCB layout and bypassing becomes especially critical in systems with higher capacitive loads because these loads create higher transient currents in the V_{DD} and GND circuits. Pull up all unused inputs or signals that will be inputs during reset. Special care should be taken to minimize the noise levels on the PLL supply pins. For more information, please refer to Section 14.4.3, Clock Synthesizer Power (VDDSYN, VSSSYN, VSSSYN1), in the *MPC866 User’s Manual*.

10 Bus Signal Timing

The maximum bus speed supported by the MPC866/859 is 66 MHz. Higher-speed parts must be operated in half-speed bus mode (for example, an MPC866/859 used at 100 MHz must be configured for a 50-MHz bus). Table 7 and Table 8 show the frequency ranges for standard part frequencies.

Table 7. Frequency Ranges for Standard Part Frequencies (1:1 Bus Mode)

Part Freq	50 MHz		66 MHz	
	Min	Max	Min	Max
Core	40	50	40	66.67
Bus	40	50	40	66.67

Table 8. Frequency Ranges for Standard Part Frequencies (2:1 Bus Mode)

Part Freq	50 MHz		66 MHz		100 MHz		133 MHz	
	Min	Max	Min	Max	Min	Max	Min	Max
Core	40	50	40	66.67	40	100	40	133.34
Bus	20	25	20	33.33	20	50	20	66.67

Table 9 shows the timings for the MPC866/859 at 33, 40, 50, and 66 MHz bus operation. The timing for the MPC866/859 bus shown in this table assumes a 50-pF load for maximum delays and a 0-pF load for minimum delays. CLKOUT assumes a 100-pF load maximum delay.

Table 9. Bus Operation Timings

Num	Characteristic	33 MHz		40 MHz		50 MHz		66 MHz		Unit
		Min	Max	Min	Max	Min	Max	Min	Max	
B1	Bus Period (CLKOUT) See Table 7	—	—	—	—	—	—	—	—	ns
B1a	EXTCLK to CLKOUT phase skew	– 2	+2	– 2	+2	– 2	+2	– 2	+2	ns
B1b	CLKOUT frequency jitter peak-to-peak	—	1	—	1	—	1	—	1	ns
B1c	Frequency jitter on EXTCLK	—	0.50	—	0.50	—	0.50	—	0.50	%

Table 9. Bus Operation Timings (continued)

Num	Characteristic	33 MHz		40 MHz		50 MHz		66 MHz		Unit
		Min	Max	Min	Max	Min	Max	Min	Max	
B1d	CLKOUT phase jitter peak-to-peak for OSCLK ≥ 15 MHz	—	4	—	4	—	4	—	4	ns
	CLKOUT phase jitter peak-to-peak for OSCLK < 15 MHz	—	5	—	5	—	5	—	5	ns
B2	CLKOUT pulse width low (MIN = 0.4 x B1, MAX = 0.6 x B1)	12.1	18.2	10.0	15.0	8.0	12.0	6.1	9.1	ns
B3	CLKOUT pulse width high (MIN = 0.4 x B1, MAX = 0.6 x B1)	12.1	18.2	10.0	15.0	8.0	12.0	6.1	9.1	ns
B4	CLKOUT rise time	—	4.00	—	4.00	—	4.00	—	4.00	ns
B5	CLKOUT fall time	—	4.00	—	4.00	—	4.00	—	4.00	ns
B7	CLKOUT to A(0:31), BADDR(28:30), RD/$\overline{WR}$, $\overline{BURST}$, D(0:31), DP(0:3) output hold (MIN = 0.25 x B1)	7.60	—	6.30	—	5.00	—	3.80	—	ns
B7a	CLKOUT to TSIZ(0:1), $\overline{REG}$, $\overline{RSV}$, AT(0:3), $\overline{BDIP}$, PTR output hold (MIN = 0.25 x B1)	7.60	—	6.30	—	5.00	—	3.80	—	ns
B7b	CLKOUT to $\overline{BR}$, $\overline{BG}$, FRZ, VFLS(0:1), VF(0:2), IWP(0:2), LWP(0:1), $\overline{STS}$ output hold (MIN = 0.25 x B1)	7.60	—	6.30	—	5.00	—	3.80	—	ns
B8	CLKOUT to A(0:31), BADDR(28:30) RD/$\overline{WR}$, $\overline{BURST}$, D(0:31), DP(0:3), valid (MAX = 0.25 x B1 + 6.3)	—	13.80	—	12.50	—	11.30	—	10.00	ns
B8a	CLKOUT to TSIZ(0:1), $\overline{REG}$, $\overline{RSV}$, AT(0:3), $\overline{BDIP}$, PTR valid (MAX = 0.25 x B1 + 6.3)	—	13.80	—	12.50	—	11.30	—	10.00	ns
B8b	CLKOUT to $\overline{BR}$, $\overline{BG}$, VFLS(0:1), VF(0:2), IWP(0:2), FRZ, LWP(0:1), $\overline{STS}$ valid [4] (MAX = 0.25 x B1 + 6.3)	—	13.80	—	12.50	—	11.30	—	10.00	ns
B9	CLKOUT to A(0:31), BADDR(28:30), RD/$\overline{WR}$, $\overline{BURST}$, D(0:31), DP(0:3), TSIZ(0:1), $\overline{REG}$, $\overline{RSV}$, AT(0:3), PTR High-Z (MAX = 0.25 x B1 + 6.3)	7.60	13.80	6.30	12.50	5.00	11.30	3.80	10.00	ns
B11	CLKOUT to $\overline{TS}$, $\overline{BB}$ assertion (MAX = 0.25 x B1 + 6.0)	7.60	13.60	6.30	12.30	5.00	11.00	3.80	9.80	ns
B11a	CLKOUT to $\overline{TA}$, $\overline{BI}$ assertion (when driven by the memory controller or PCMCIA interface) (MAX = 0.00 x B1 + 9.30 [1])	2.50	9.30	2.50	9.30	2.50	9.30	2.50	9.80	ns
B12	CLKOUT to $\overline{TS}$, $\overline{BB}$ negation (MAX = 0.25 x B1 + 4.8)	7.60	12.30	6.30	11.00	5.00	9.80	3.80	8.50	ns

Table 9. Bus Operation Timings (continued)

Num	Characteristic	33 MHz		40 MHz		50 MHz		66 MHz		Unit
		Min	Max	Min	Max	Min	Max	Min	Max	
B12a	CLKOUT to TA, BI negation (when driven by the memory controller or PCMCIA interface) (MAX = 0.00 x B1 + 9.00)	2.50	9.00	2.50	9.00	2.50	9.00	2.50	9.00	ns
B13	CLKOUT to TS, BB High-Z (MIN = 0.25 x B1)	7.60	21.60	6.30	20.30	5.00	19.00	3.80	14.00	ns
B13a	CLKOUT to TA, BI High-Z (when driven by the memory controller or PCMCIA interface) (MIN = 0.00 x B1 + 2.5)	2.50	15.00	2.50	15.00	2.50	15.00	2.50	15.00	ns
B14	CLKOUT to TEA assertion (MAX = 0.00 x B1 + 9.00)	2.50	9.00	2.50	9.00	2.50	9.00	2.50	9.00	ns
B15	CLKOUT to TEA High-Z (MIN = 0.00 x B1 + 2.50)	2.50	15.00	2.50	15.00	2.50	15.00	2.50	15.00	ns
B16	TA, BI valid to CLKOUT (setup time) (MIN = 0.00 x B1 + 6.00)	6.00	—	6.00	—	6.00	—	6.00	—	ns
B16a	TEA, KR, RETRY, CR valid to CLKOUT (setup time) (MIN = 0.00 x B1 + 4.5)	4.50	—	4.50	—	4.50	—	4.50	—	ns
B16b	BB, BG, BR, valid to CLKOUT (setup time) [2] (4 MIN = 0.00 x B1 + 0.00)	4.00	—	4.00	—	4.00	—	4.00	—	ns
B17	CLKOUT to TA, TEA, BI, BB, BG, BR valid (hold time) (MIN = 0.00 x B1 + 1.00 [3])	1.00	—	1.00	—	1.00	—	2.00	—	ns
B17a	CLKOUT to KR, RETRY, CR valid (hold time) (MIN = 0.00 x B1 + 2.00)	2.00	—	2.00	—	2.00	—	2.00	—	ns
B18	D(0:31), DP(0:3) valid to CLKOUT rising edge (setup time) [4] (MIN = 0.00 x B1 + 6.00)	6.00	—	6.00	—	6.00	—	6.00	—	ns
B19	CLKOUT rising edge to D(0:31), DP(0:3) valid (hold time) [4] (MIN = 0.00 x B1 + 1.00 [5])	1.00	—	1.00	—	1.00	—	2.00	—	ns
B20	D(0:31), DP(0:3) valid to CLKOUT falling edge (setup time) [6](MIN = 0.00 x B1 + 4.00)	4.00	—	4.00	—	4.00	—	4.00	—	ns
B21	CLKOUT falling edge to D(0:31), DP(0:3) valid (hold Time) [6] (MIN = 0.00 x B1 + 2.00)	2.00	—	2.00	—	2.00	—	2.00	—	ns
B22	CLKOUT rising edge to CS asserted GPCM ACS = 00 (MAX = 0.25 x B1 + 6.3)	7.60	13.80	6.30	12.50	5.00	11.30	3.80	10.00	ns
B22a	CLKOUT falling edge to CS asserted GPCM ACS = 10, TRLX = 0 (MAX = 0.00 x B1 + 8.00)	—	8.00	—	8.00	—	8.00	—	8.00	ns

Figure 19 shows the timing for the external bus controlled by the UPM.

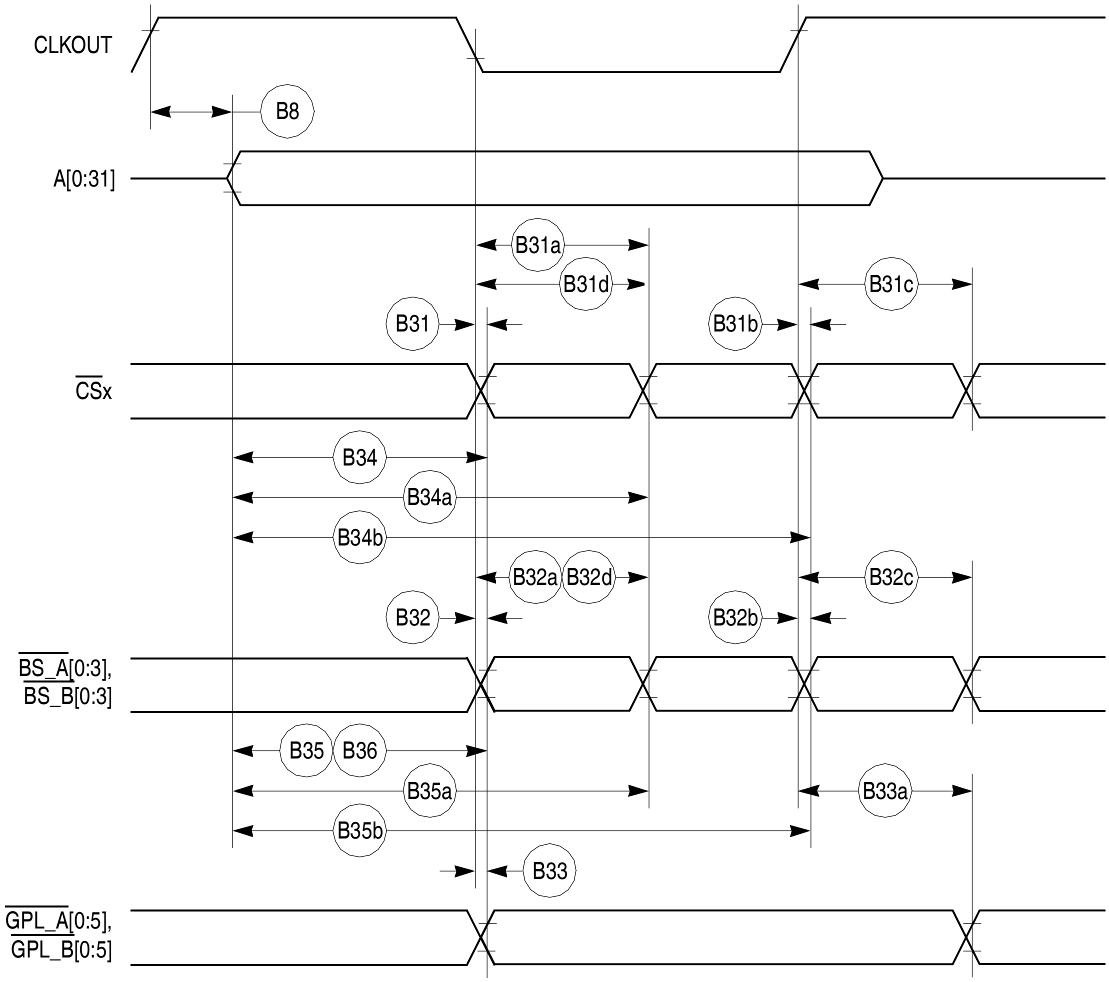

Figure 19. External Bus Timing (UPM Controlled Signals)

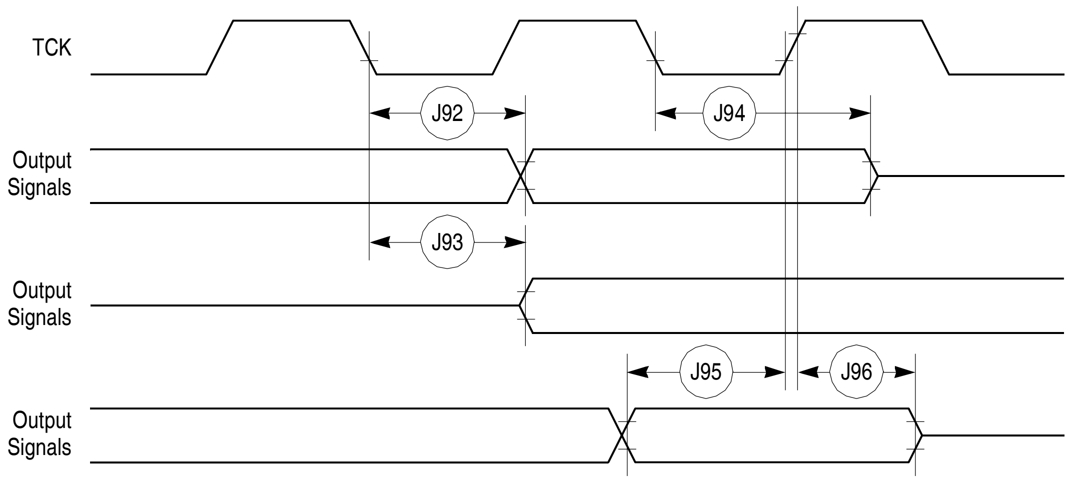

Figure 40. Boundary Scan (JTAG) Timing Diagram

12 CPM Electrical Characteristics

This section provides the AC and DC electrical specifications for the communications processor module (CPM) of the MPC866/859.

12.1 PIP/PIO AC Electrical Specifications

Table 16 shows the PIP/PIO AC timings as shown in Figure 41 through Figure 45.

Table 16. PIP/PIO Timing

Num	Characteristic	All Frequencies		Unit
		Min	Max	
21	Data-in setup time to STBI low	0	—	ns
22	Data-In hold time to STBI high	2.5 – t3 [1]	—	clk
23	STBI pulse width	1.5	—	clk
24	STBO pulse width	1 clk – 5ns	—	ns
25	Data-out setup time to STBO low	2	—	clk
26	Data-out hold time from STBO high	5	—	clk
27	STBI low to STBO low (Rx interlock)	—	2	clk
28	STBI low to STBO high (Tx interlock)	2	—	clk
29	Data-in setup time to clock high	15	—	ns
30	Data-in hold time from clock high	7.5	—	ns
31	Clock low to data-out valid (CPU writes data, control, or direction)	—	25	ns

[1] t3 = Specification 23

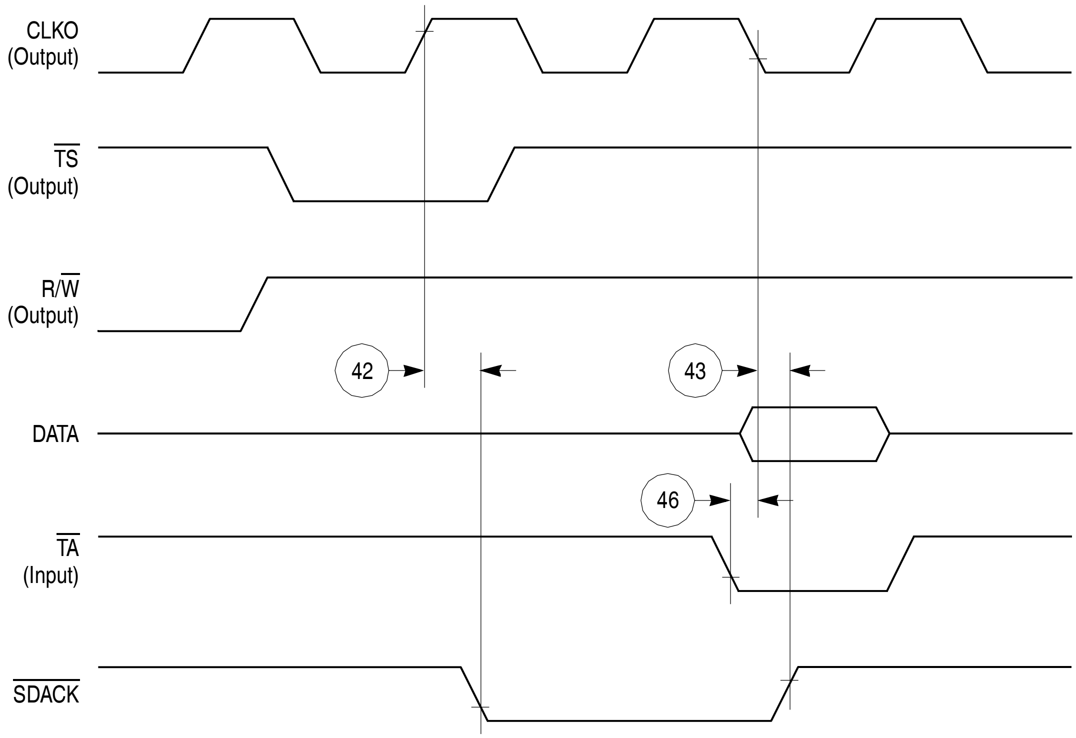

Figure 48. SDACK Timing Diagram—Peripheral Write, Externally-Generated TA

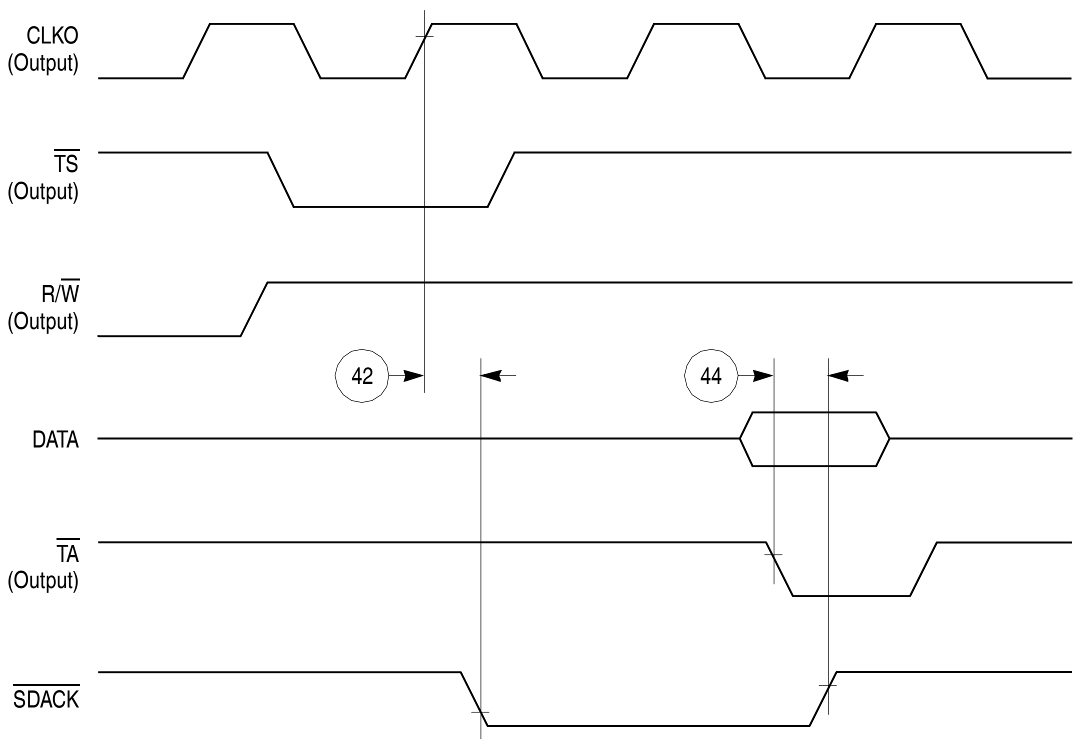

Figure 49. SDACK Timing Diagram—Peripheral Write, Internally-Generated TA

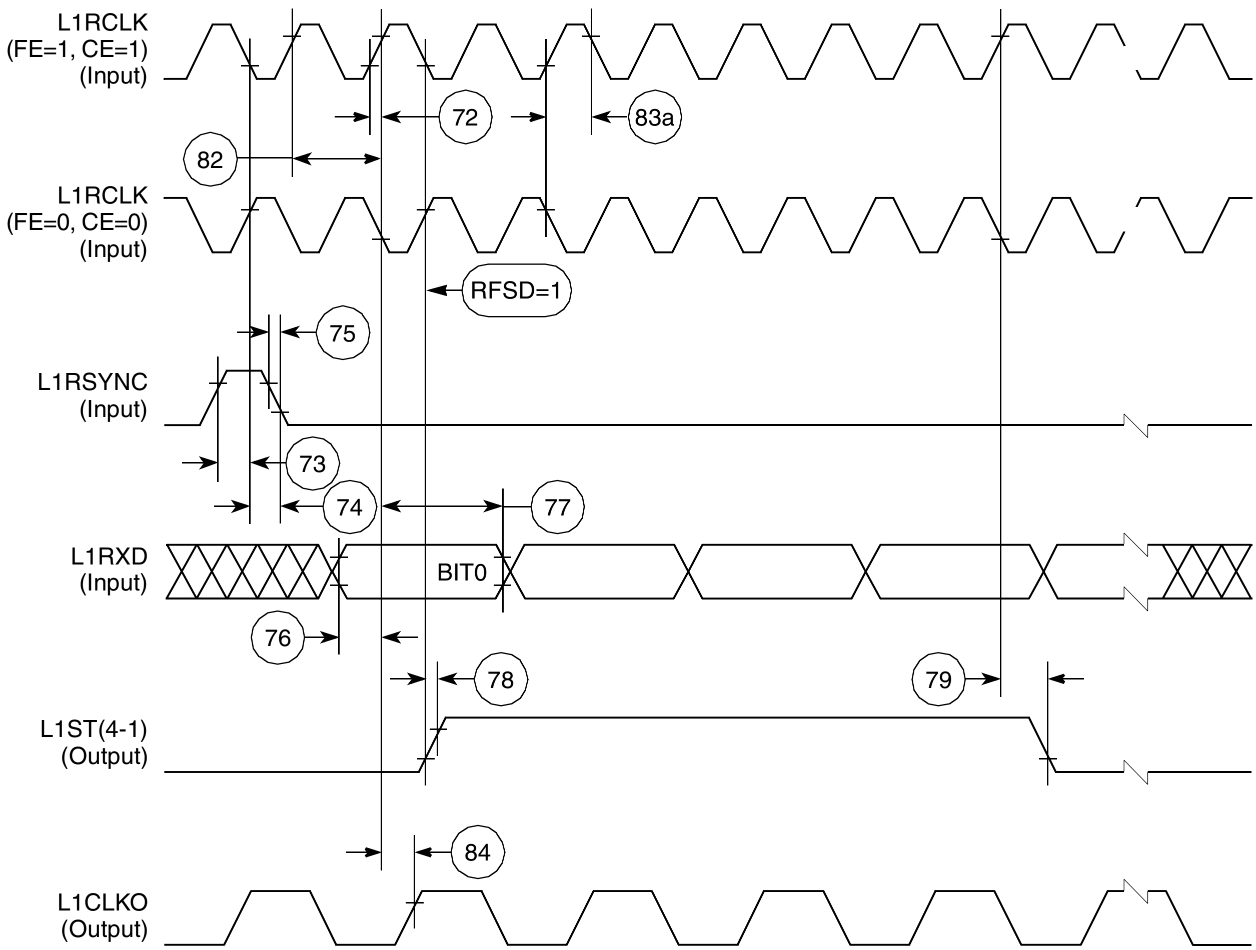

Figure 54. SI Receive Timing with Double-Speed Clocking (DSC = 1)

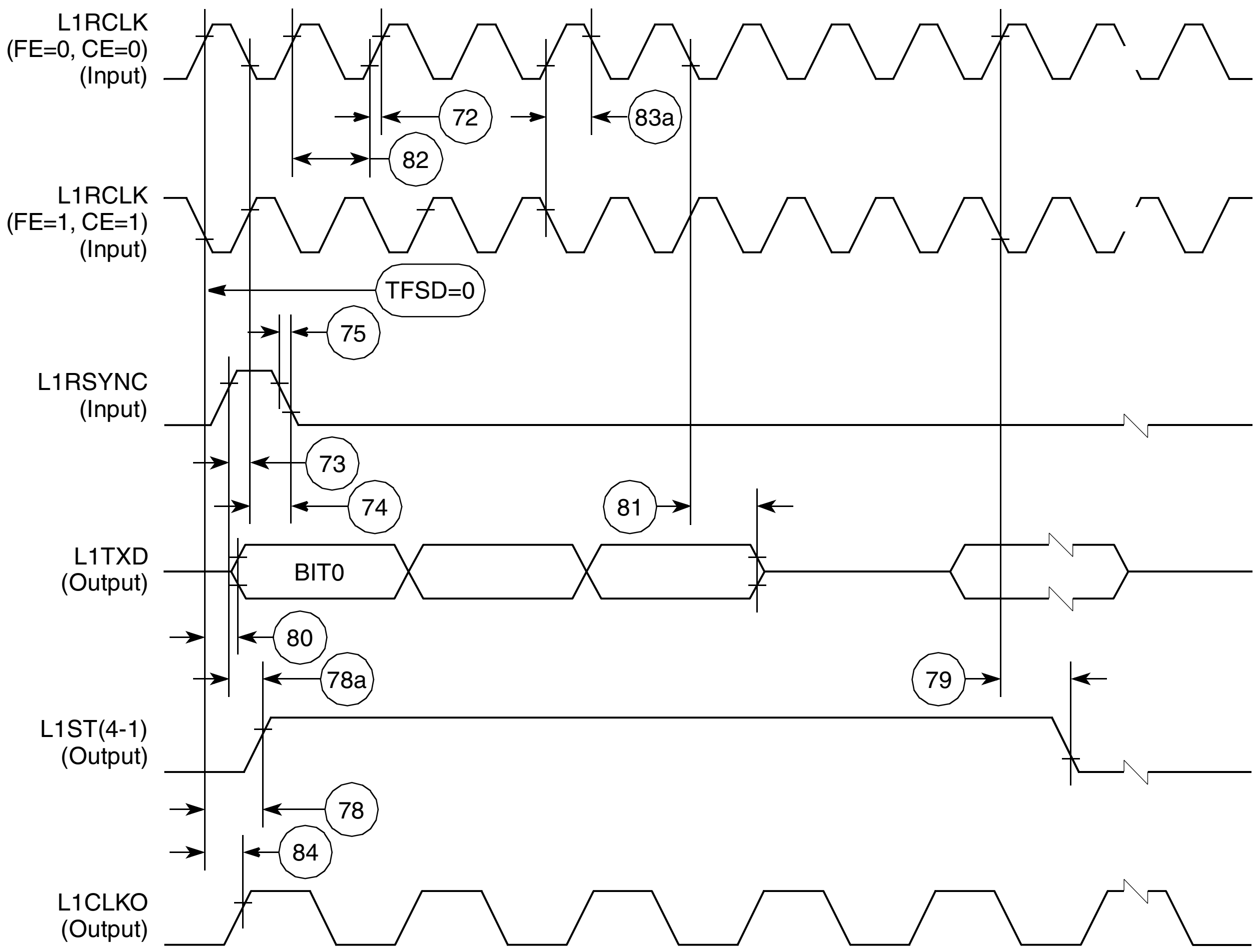

Figure 56. SI Transmit Timing with Double Speed Clocking (DSC = 1)

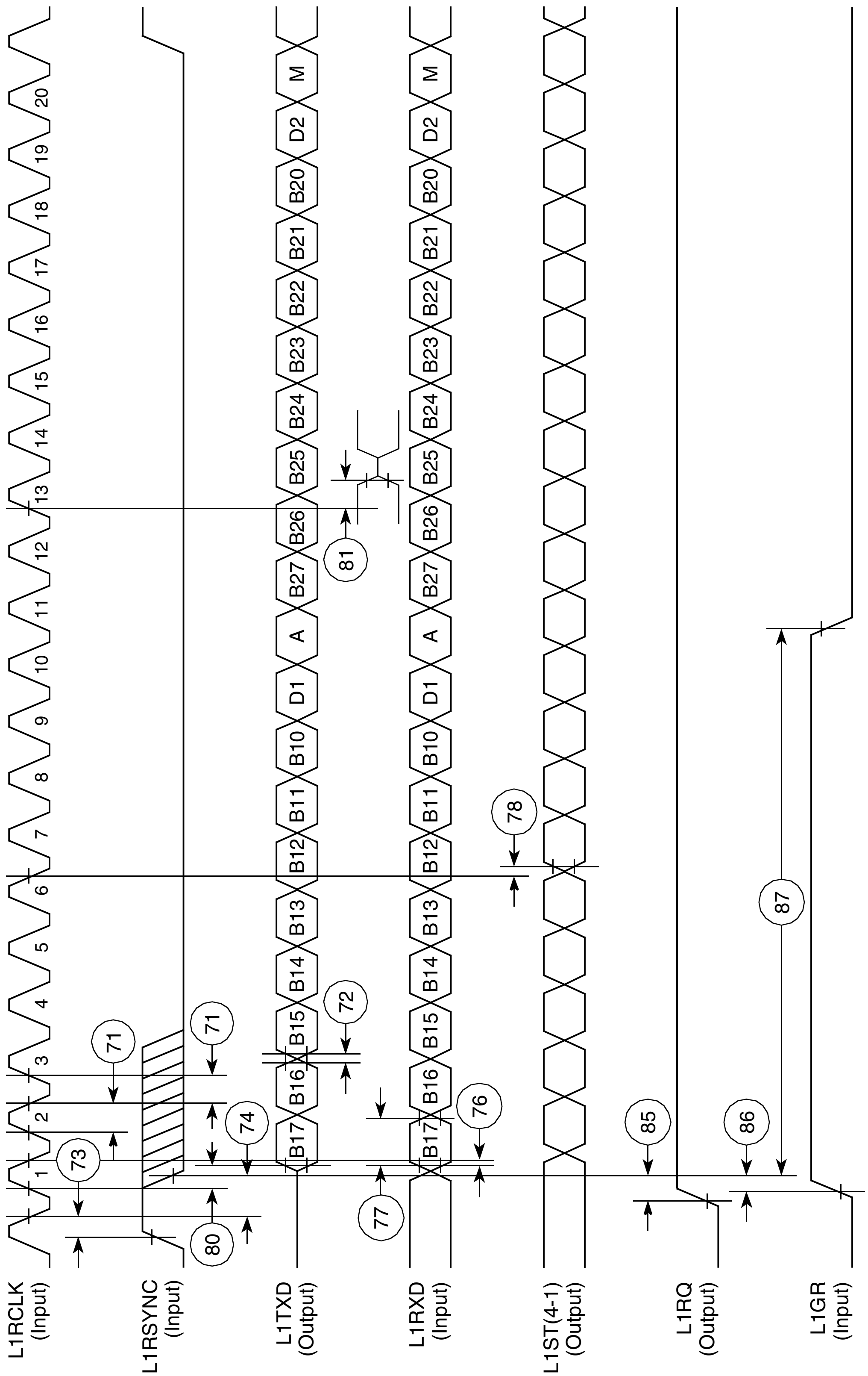

Figure 57. IDL Timing

Table 24. Ethernet Timing (continued)

Num	Characteristic	All Frequencies		Unit
		Min	Max	
134	TENA inactive delay (from TCLK1 rising edge)	10	50	ns
135	$\overline{\text{RSTRT}}$ active delay (from TCLK1 falling edge)	10	50	ns
136	$\overline{\text{RSTRT}}$ inactive delay (from TCLK1 falling edge)	10	50	ns
137	$\overline{\text{REJECT}}$ width low	1	—	CLK
138	CLKO1 low to $\overline{\text{SDACK}}$ asserted [2]	—	20	ns
139	CLKO1 low to $\overline{\text{SDACK}}$ negated [2]	—	20	ns

[1] The ratios SyncCLK/RCLK1 and SyncCLK/TCLK1 must be greater or equal to 2/1.

[2] $\overline{\text{SDACK}}$ is asserted whenever the SDMA writes the incoming frame DA into memory.

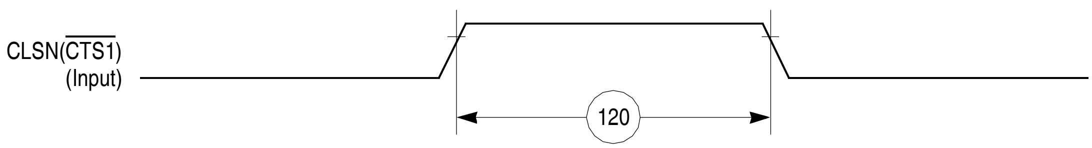

Figure 61. Ethernet Collision Timing Diagram

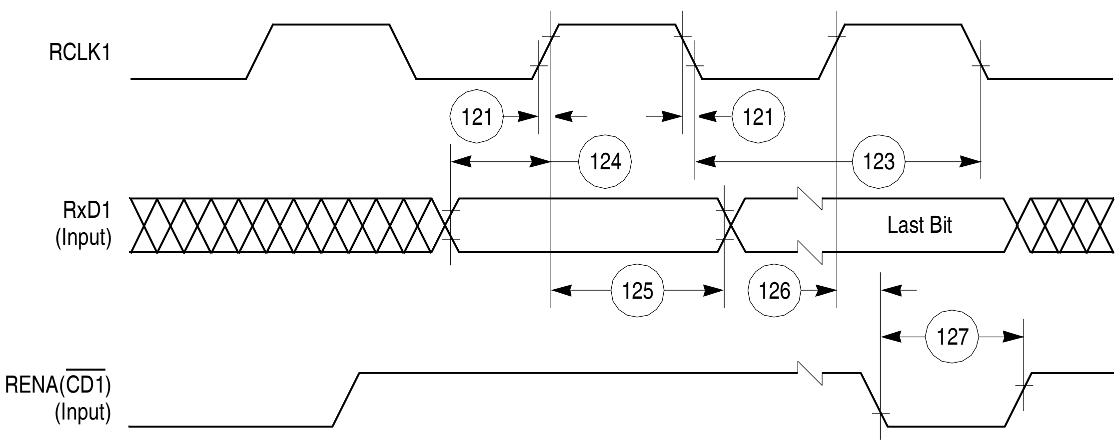

Figure 62. Ethernet Receive Timing Diagram

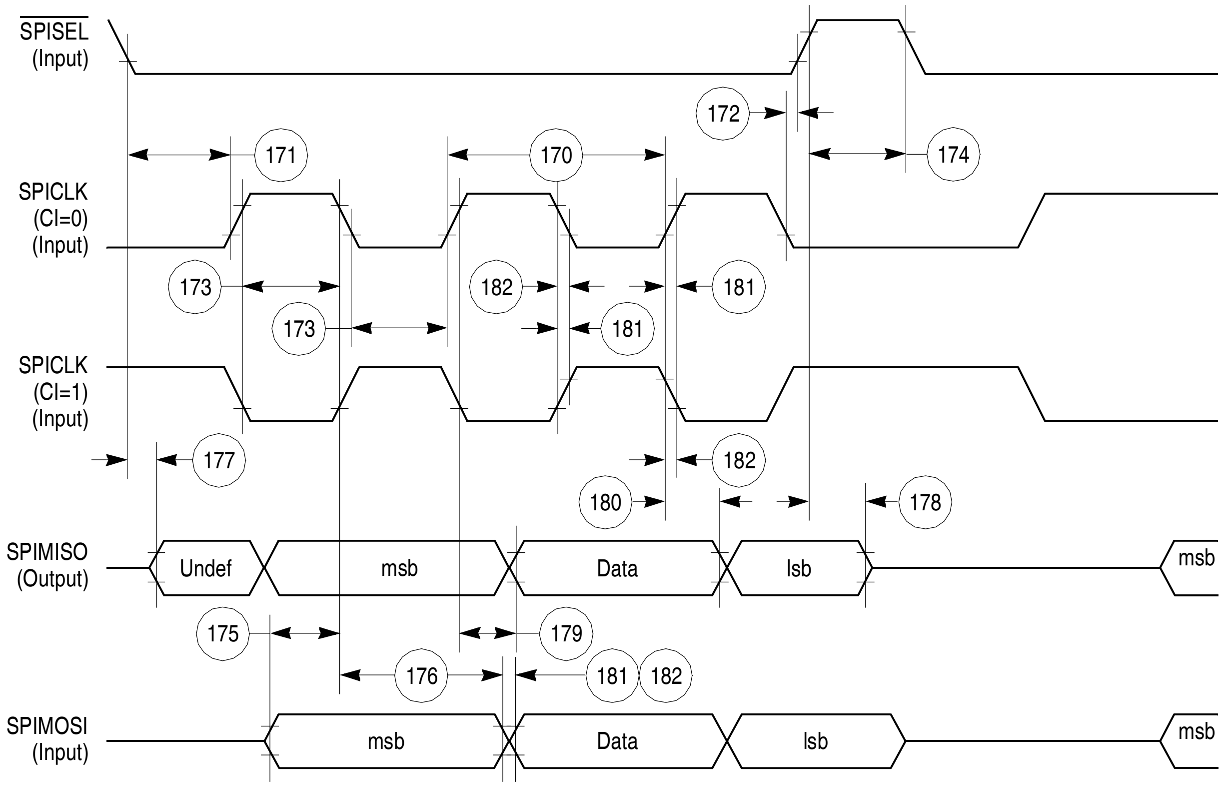

Figure 70. SPI Slave (CP = 1) Timing Diagram

12.12 I^2C AC Electrical Specifications

Table 28 shows the I^2C (SCL < 100 kHz) timings.

Table 28. I^2C Timing (SCL < 100 kHz)

Num	Characteristic	All Frequencies		Unit
		Min	Max	
200	SCL clock frequency (slave)	0	100	kHz
200	SCL clock frequency (master) [1]	1.5	100	kHz
202	Bus free time between transmissions	4.7	—	µs
203	Low period of SCL	4.7	—	µs
204	High period of SCL	4.0	—	µs
205	Start condition setup time	4.7	—	µs
206	Start condition hold time	4.0	—	µs
207	Data hold time	0	—	µs
208	Data setup time	250	—	ns
209	SDL/SCL rise time	—	1	µs
210	SDL/SCL fall time	—	300	ns
211	Stop condition setup time	4.7	—	µs

1 SCL frequency is given by SCL = BRGCLK_frequency / ((BRG register + 3) * pre_scaler * 2).
The ratio SyncClk/(BRGCLK/pre_scaler) must be greater or equal to 4/1.

Table 29 shows the I^2C (SCL > 100 kHz) timings.

Table 29. I^2C Timing (SCL > 100 kHz)

Num	Characteristic	Expression	All Frequencies		Unit
			Min	Max	
200	SCL clock frequency (slave)	fSCL	0	BRGCLK/48	Hz
200	SCL clock frequency (master) [1]	fSCL	BRGCLK/16512	BRGCLK/48	Hz
202	Bus free time between transmissions	—	1/(2.2 * fSCL)	—	s
203	Low period of SCL	—	1/(2.2 * fSCL)	—	s
204	High period of SCL	—	1/(2.2 * fSCL)	—	s
205	Start condition setup time	—	1/(2.2 * fSCL)	—	s
206	Start condition hold time	—	1/(2.2 * fSCL)	—	s
207	Data hold time	—	0	—	s
208	Data setup time	—	1/(40 * fSCL)	—	s
209	SDL/SCL rise time	—	—	1/(10 * fSCL)	s
210	SDL/SCL fall time	—	—	1/(33 * fSCL)	s
211	Stop condition setup time	—	1/2(2.2 * fSCL)	—	s

1 SCL frequency is given by SCL = BrgClk_frequency / ((BRG register + 3) * pre_scaler * 2).
The ratio SyncClk/(Brg_Clk/pre_scaler) must be greater or equal to 4/1.

Table 38. MPC866/859 Package/Frequency Orderable (continued)

Plastic ball grid array (VR suffix) Lead free	0° to 95°C	50	MPC859DSLVR50A
		66	MPC859DSLVR66A
		100	MPC859PVR100A MPC859TVR100A MPC866PVR100A MPC866TVR100A
		133	MPC859PVR133A MPC859TVR133A MPC866PVR133A MPC866TVR133A
Plastic ball grid array (CVR suffix) Lead free	–40° to 100°C	50	MPC859DSLCVR50A
		66	MPC859DSLCVR66A
		100	MPC859PCVR100A MPC859TCVR100A MPC866PCVR100A MPC866TCVR100A

Table 39. Pin Assignments (continued)

Name	Pin Number	Type
OP3 MODCK2 DSDO	M4	Bidirectional
BADDR30 $\overline{\text{REG}}$	K4	Output
BADDR[28:29]	M3, M2	Output
$\overline{\text{AS}}$	L3	Input
PA15 RXD1 RXD4	C18	Bidirectional
PA14 TXD1 TXD4	D17	Bidirectional (Optional: Open-drain)
PA13 RXD2	E17	Bidirectional
PA12 TXD2	F17	Bidirectional (Optional: Open-drain)
PA11 L1TXDB RXD3	G16	Bidirectional (Optional: Open-drain)
PA10 L1RXDB TXD3	J17	Bidirectional (Optional: Open-drain)
PA9 L1TXDA RXD4	K18	Bidirectional (Optional: Open-drain)
PA8 L1RXDA TXD4	L17	Bidirectional (Optional: Open-drain)
PA7 CLK1 L1RCLKA BRGO1 TIN1	M19	Bidirectional
PA6 CLK2 $\overline{\text{TOUT1}}$	M17	Bidirectional

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