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"[Embedded - Microcontrollers](#)" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

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Details

Product Status	Active
Core Processor	Coldfire V2
Core Size	32-Bit Single-Core
Speed	60MHz
Connectivity	Ethernet, I ² C, SPI, UART/USART
Peripherals	DMA, LVD, POR, PWM, WDT
Number of I/O	73
Program Memory Size	256KB (256K x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	32K x 8
Voltage - Supply (Vcc/Vdd)	3V ~ 3.6V
Data Converters	A/D 8x12b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	121-LBGA
Supplier Device Package	121-MAPBGA (12x12)
Purchase URL	https://www.e-xfl.com/pro/item?MUrl=&PartUrl=mcf52233cvm60

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1 MCF52235 Family Configurations

Table 1. MCF52235 Family Configurations

Module	52230	52231	52232	52233	52234	52235	52236
Version 2 ColdFire Core with EMAC (Enhanced Multiply-Accumulate Unit)	•	•	•	•	•	•	•
System Clock (MHz)	60	60	50	60	60	60	50
Performance (Dhrystone 2.1 MIPS)	56	56	46	56	56	56	46
Flash / Static RAM (SRAM)	128/32 Kbytes	128/32 Kbytes	128/32 Kbytes	256/32 Kbytes	256/32 Kbytes	256/32 Kbytes	256/32 Kbytes
Interrupt Controllers (INTC0/INTC1)	•	•	•	•	•	•	•
Fast Analog-to-Digital Converter (ADC)	•	•	•	•	•	•	•
Random Number Generator and Crypto Acceleration Unit (CAU)	—	—	—	—	—	•	—
FlexCAN 2.0B Module	—	•	—	—	•	•	—
Fast Ethernet Controller (FEC) with on-chip interface (EPHY)	•	•	•	•	•	•	•
Four-channel Direct-Memory Access (DMA)	•	•	•	•	•	•	•
Software Watchdog Timer (WDT)	•	•	•	•	•	•	•
Programmable Interrupt Timer	2	2	2	2	2	2	2
Four-Channel General Purpose Timer	•	•	•	•	•	•	•
32-bit DMA Timers	4	4	4	4	4	4	4
QSPI	•	•	•	•	•	•	•
UART(s)	3	3	3	3	3	3	3
I ² C	•	•	•	•	•	•	•
Eight/Four-channel 8/16-bit PWM Timer	•	•	•	•	•	•	•
General Purpose I/O Module (GPIO)	•	•	•	•	•	•	•
Chip Configuration and Reset Controller Module	•	•	•	•	•	•	•
Background Debug Mode (BDM)	•	•	•	•	•	•	•
JTAG - IEEE 1149.1 Test Access Port ¹	•	•	•	•	•	•	•
Package	80 LQFP 112 LQFP	80 LQFP 112 LQFP	80 LQFP	80 LQFP 112 LQFP	112 LQFP 121 MAPBGA	112 LQFP 121 MAPBGA	80 LQFP

¹ The full debug/trace interface is available only on the 112- and 121-pin packages. A reduced debug interface is bonded on the 80-pin package.

1.2.1 Feature Overview

The MCF52235 family includes the following features:

- Version 2 ColdFire variable-length RISC processor core
 - Static operation
 - 32-bit address and data paths on-chip
 - Up to 60 MHz processor core frequency
 - Sixteen general-purpose, 32-bit data and address registers
 - Implements ColdFire ISA_A with extensions to support the user stack pointer register and four new instructions for improved bit processing (ISA_A+)
 - Enhanced Multiply-Accumulate (EMAC) unit with 32-bit accumulator to support $16 \times 16 \rightarrow 32$ or $32 \times 32 \rightarrow 32$ operations
 - Cryptography Acceleration Unit (CAU)
 - Tightly-coupled coprocessor to accelerate software-based encryption and message digest functions
 - FIPS-140 compliant random number generator
 - Support for DES, 3DES, AES, MD5, and SHA-1 algorithms
 - Illegal instruction decode that allows for 68K emulation support
- System debug support
 - Real time trace for determining dynamic execution path
 - Background debug mode (BDM) for in-circuit debugging (DEBUG_B+)
 - Real time debug support, with six hardware breakpoints (4 PC, 1 address and 1 data) that can be configured into a 1- or 2-level trigger
- On-chip memories
 - Up to 32 Kbytes of dual-ported SRAM on CPU internal bus, supporting core and DMA access with standby power supply support
 - Up to 256 Kbytes of interleaved Flash memory supporting 2-1-1-1 accesses
- Power management
 - Fully static operation with processor sleep and whole chip stop modes
 - Rapid response to interrupts from the low-power sleep mode (wake-up feature)
 - Clock enable/disable for each peripheral when not used
- Fast Ethernet Controller (FEC)
 - 10/100 BaseT/TX capability, half duplex or full duplex
 - On-chip transmit and receive FIFOs
 - Built-in dedicated DMA controller
 - Memory-based flexible descriptor rings
- On-chip Ethernet Transceiver (EPHY)
 - Digital adaptive equalization
 - Supports auto-negotiation
 - Baseline wander correction
 - Full-/Half-duplex support in all modes
 - Loopback modes
 - Supports MDIO preamble suppression
 - Jumbo packet
- FlexCAN 2.0B module

- Unique vector number for each interrupt source
- Ability to mask any individual interrupt source or all interrupt sources (global mask-all)
- Support for hardware and software interrupt acknowledge (IACK) cycles
- Combinatorial path to provide wake-up from low power modes
- DMA controller
 - Four fully programmable channels
 - Dual-address transfer support with 8-, 16-, and 32-bit data capability, along with support for 16-byte (4 x 32-bit) burst transfers
 - Source/destination address pointers that can increment or remain constant
 - 24-bit byte transfer counter per channel
 - Auto-alignment transfers supported for efficient block movement
 - Bursting and cycle steal support
 - Software-programmable DMA requesters for the UARTs (3) and 32-bit timers (4)
- Reset
 - Separate reset in and reset out signals
 - Seven sources of reset:
 - Power-on reset (POR)
 - External
 - Software
 - Watchdog
 - Loss of clock
 - Loss of lock
 - Low-voltage detection (LVD)
 - Status flag indication of source of last reset
- Chip integration module (CIM)
 - System configuration during reset
 - Selects one of three clock modes
 - Configures output pad drive strength
 - Unique part identification number and part revision number
- General purpose I/O interface
 - Up to 56 bits of general purpose I/O
 - Bit manipulation supported via set/clear functions
 - Programmable drive strengths
 - Unused peripheral pins may be used as extra GPIO
- JTAG support for system level board testing

1.2.2 V2 Core Overview

The version 2 ColdFire processor core is comprised of two separate pipelines decoupled by an instruction buffer. The two-stage instruction fetch pipeline (IFP) is responsible for instruction-address generation and instruction fetch. The instruction buffer is a first-in-first-out (FIFO) buffer that holds prefetched instructions awaiting execution in the operand execution pipeline (OEP). The OEP includes two pipeline stages. The first stage decodes instructions and selects operands (DSOC); the second stage (AGEX) performs instruction execution and calculates operand effective addresses, if needed.

The V2 core implements the ColdFire instruction set architecture revision A+ with added support for a separate user stack pointer register and four new instructions to assist in bit processing. Additionally, the MCF52235 core includes the enhanced multiply-accumulate (EMAC) unit for improved signal processing capabilities. The EMAC implements a three-stage arithmetic

	1	2	3	4	5	6	7	8	9	10	11
A	TCLK	SDA	SCL	$\overline{\text{IRQ15}}$	$\overline{\text{IRQ14}}$	$\overline{\text{IRQ13}}$	VSSA	VDDA	AN1	AN7	AN5
B	TMS	$\overline{\text{RCON}}$	GPT0	GPT3	PWM5	PWM1	VRL	VRH	AN2	AN6	AN4
C	$\overline{\text{TRST}}$	TDO	TDI	GPT2	PWM7	PWM3	$\overline{\text{IRQ12}}$	AN0	AN3	LNKLED	ACTLED
D	DTIN1	DTIN0	ALLPST	GPT1	VDDX	VDDX	VDD	VDDR	PST2	PST3	SPDLED
E	DDATA3	$\overline{\text{IRQ9}}$	$\overline{\text{IRQ8}}$	VSS	VSS	VDDX	VSS	VDD	PST0	PST1	PHY_RXN
F	DDATA0	DDATA1	DDATA2	VSS	VSS	VSS	VSS	VSS	PHY_VSSRX	PHY_VDDR	PHY_RXP
G	DTIN2	$\overline{\text{IRQ5}}$	$\overline{\text{IRQ6}}$	JTAG_EN	VDDX	VDDX	VDDX	PHY_VSSA	PHY_VSSTX	PHY_VDDTX	PHY_TXP
H	DTIN3	$\overline{\text{URTS0}}$	$\overline{\text{URTS1}}$	QSPI_DIN	QSPI_CS1	VDDX	TEST	TXLED	RXLED	PHY_VDDA	PHY_TXN
J	SYNCB	$\overline{\text{UCTS0}}$	$\overline{\text{UCTS1}}$	QSPI_DOUT	QSPI_CS2	$\overline{\text{RSTI}}$	XTAL	$\overline{\text{IRQ1}}$	COLLED	DUPLED	PHY_RBIAS
K	SYNCA	URXD0	URXD1	QSPI_CLK	QSPI_CS3	VDDPLL	VSSPLL	$\overline{\text{IRQ2}}$	$\overline{\text{IRQ11}}$	$\overline{\text{URTS2}}$	URXD2
L	$\overline{\text{IRQ10}}$	UTXD0	UTXD1	QSPI_CS0	$\overline{\text{IRQ4}}$	$\overline{\text{RSTO}}$	EXTAL	$\overline{\text{IRQ3}}$	$\overline{\text{IRQ7}}$	$\overline{\text{UCTS2}}$	UTXD2

Figure 4. 121 MAPBGA Pin Assignments

Table 3. Pin Functions by Primary and Alternate Purpose (continued)

Pin Group	Primary Function	Secondary Function	Tertiary Function	Quaternary Function	Drive Strength/Control ¹	Wired OR Control	Pull-up/Pull-down ²	Pin on 121 MAPBGA	Pin on 112 LQFP	Pin on 80 LQFP
Continued Interrupts ³	IRQ11	—	—	PGP[3]	PSDR[43]	—	Pull-Up ⁶	K9	57	41
	IRQ10	—	—	PGP[2]	PSDR[42]	—	Pull-Up ⁶	L1	29	—
	IRQ9	—	—	PGP[1]	PSDR[41]	—	Pull-Up ⁶	E2	11	—
	IRQ8	—	—	PGP[0]	PSDR[40]	—	Pull-Up	E3	10	—
	IRQ7	—	—	PNQ[7]	Low	—	Pull-Up ⁶	L9	56	40
	IRQ6	—	FEC_RXER	PNQ[6]	Low	—	Pull-Up ⁶	G3	19	—
	IRQ5	—	FEC_RXD[1]	PNQ[5]	Low	—	Pull-Up ⁶	G2	20	—
	IRQ4	—	—	PNQ[4]	Low	—	Pull-Up ⁶	L5	41	29
	IRQ3	—	FEC_RXD[2]	PNQ[3]	Low	—	Pull-Up ⁶	L8	53	—
	IRQ2	—	FEC_RXD[3]	PNQ[2]	Low	—	Pull-Up ⁶	K8	54	—
	IRQ1	SYNCA	PWM1	PNQ[1]	High	—	Pull-Up ⁶	J8	55	39
JTAG/BDM	JTAG_EN	—	—	—	N/A	N/A	Pull-Down	G4	18	12
	TCLK/PSTCLK	CLKOUT	—	—	High	—	Pull-Up ⁷	A1	1	1
	TDI/DSI	—	—	—	N/A	N/A	Pull-Up ⁷	C3	4	4
	TDO/DSO	—	—	—	High	N/A	—	C2	5	5
	TMS/BKPT	—	—	—	N/A	N/A	Pull-Up ⁷	B1	2	2
	TRST/DSCLK	—	—	—	N/A	N/A	Pull-Up	C1	6	6
Mode Selection	RCON/EZPCS	—	—	—	N/A	N/A	Pull-Up	B2	3	3
PWM	PWM7	—	—	PTD[3]	PDSR[31]	—	—	C5	104	—
	PWM5	—	—	PTD[2]	PDSR[30]	—	—	B5	103	—
	PWM3	—	—	PTD[1]	PDSR[29]	—	—	C6	100	—
	PWM1	—	—	PTD[0]	PDSR[28]	—	—	B6	99	—

1.3 Reset Signals

Table 4 describes signals that are used to either reset the chip or as a reset indication.

Table 4. Reset Signals

Signal Name	Abbreviation	Function	I/O
Reset In	$\overline{\text{RSTI}}$	Primary reset input to the device. Asserting $\overline{\text{RSTI}}$ immediately resets the CPU and peripherals.	I
Reset Out	$\overline{\text{RSTO}}$	Driven low for 512 CPU clocks after the reset source has deasserted.	O

1.4 PLL and Clock Signals

Table 5 describes signals that are used to support the on-chip clock generation circuitry.

Table 5. PLL and Clock Signals

Signal Name	Abbreviation	Function	I/O
External Clock In	EXTAL	Crystal oscillator or external clock input.	I
Crystal	XTAL	Crystal oscillator output.	O
Clock Out	CLKOUT	This output signal reflects the internal system clock.	O

1.5 Mode Selection

Table 6 describes signals used in mode selection, Table 6 describes particular clocking modes.

Table 6. Mode Selection Signals

Signal Name	Abbreviation	Function	I/O
Reset Configuration	RCON	The Serial Flash Programming mode is entered by asserting the RCON pin (with the TEST pin negated) as the chip comes out of reset. During this mode, the EzPort has access to the Flash memory which can be programmed from an external device.	—
Test	TEST	Reserved for factory testing only and in normal modes of operation should be connected to VSS to prevent unintentional activation of test functions.	I

1.6 External Interrupt Signals

Table 7 describes the external interrupt signals.

Table 7. External Interrupt Signals

Signal Name	Abbreviation	Function	I/O
External Interrupts	$\overline{\text{IRQ}}[15:1]$	External interrupt sources.	I

1.7 Queued Serial Peripheral Interface (QSPI)

Table 8 describes QSPI signals.

Table 8. Queued Serial Peripheral Interface (QSPI) Signals

Signal Name	Abbreviation	Function	I/O
QSPI Synchronous Serial Output	QSPI_DOUT	Provides the serial data from the QSPI and can be programmed to be driven on the rising or falling edge of QSPI_CLK.	O
QSPI Synchronous Serial Data Input	QSPI_DIN	Provides the serial data to the QSPI and can be programmed to be sampled on the rising or falling edge of QSPI_CLK.	I
QSPI Serial Clock	QSPI_CLK	Provides the serial clock from the QSPI. The polarity and phase of QSPI_CLK are programmable.	O
Synchronous Peripheral Chip Selects	QSPI_CS[3:0]	QSPI peripheral chip selects that can be programmed to be active high or low.	O

1.8 Fast Ethernet Controller EPHY Signals

Table 9 describes the Fast Ethernet Controller (FEC) signals.

Table 9. Fast Ethernet Controller (FEC) Signals

Signal Name	Abbreviation	Function	I/O
Twisted Pair Input +	RXP	Differential Ethernet twisted-pair input pin. This pin is high-impedance out of reset.	I
Twisted Pair Input -	RXN	Differential Ethernet twisted-pair input pin. This pin is high-impedance out of reset.	I
Twisted Pair Output +	TXN	Differential Ethernet twisted-pair output pin. This pin is high-impedance out of reset.	O
Twisted Pair Output -	TXP	Differential Ethernet twisted-pair output pin. This pin is high-impedance out of reset.	O
Bias Control Resistor	RBIAS	Connect a 12.4 k Ω (1.0%) external resistor, RBIAS, between the PHY_RBIAS pin and analog ground. Place this resistor as near to the chip pin as possible. Stray capacitance must be kept to less than 10 pF (>50 pF causes instability). No high-speed signals can be permitted in the region of RBIAS.	I
Activity LED	ACT_LED	Indicates when the EPHY is transmitting or receiving	O
Link LED	LINK_LED	Indicates when the EPHY has a valid link	O
Speed LED	SPD_LED	Indicates the speed of the EPHY connection	O
Duplex LED	DUPLED	Indicates the duplex (full or half) of the EPHY connection	O
Collision LED	COLLED	Indicates if the EPHY detects a collision	O
Transmit LED	TXLED	Indicates if the EPHY is transmitting	O
Receive LED	RXLED	Indicates if the EPHY is receiving	O

1.12 ADC Signals

Table 13 describes the signals of the Analog-to-Digital Converter.

Table 13. ADC Signals

Signal Name	Abbreviation	Function	I/O
Analog Inputs	AN[7:0]	Inputs to the A-to-D converter.	I
Analog Reference	V _{RH}	Reference voltage high and low inputs.	I
	V _{RL}		I
Analog Supply	V _{DDA}	Isolate the ADC circuitry from power supply noise	—
	V _{SSA}		—

1.13 General Purpose Timer Signals

Table 14 describes the General Purpose Timer Signals.

Table 14. GPT Signals

Signal Name	Abbreviation	Function	I/O
General Purpose Timer Input/Output	GPT[3:0]	Inputs to or outputs from the general purpose timer module	I/O

1.14 Pulse Width Modulator Signals

Table 15 describes the PWM signals.

Table 15. PWM Signals

Signal Name	Abbreviation	Function	I/O
PWM Output Channels	PWM[7:0]	Pulse width modulated output for PWM channels	O

1.15 Debug Support Signals

These signals are used as the interface to the on-chip JTAG controller and also to interface to the BDM logic.

Table 16. Debug Support Signals

Signal Name	Abbreviation	Function	I/O
JTAG Enable	JTAG_EN	Select between debug module and JTAG signals at reset	I
Test Reset	$\overline{\text{TRST}}$	This active-low signal is used to initialize the JTAG logic asynchronously.	I
Test Clock	TCLK	Used to synchronize the JTAG logic.	I
Test Mode Select	TMS	Used to sequence the JTAG state machine. TMS is sampled on the rising edge of TCLK.	I
Test Data Input	TDI	Serial input for test instructions and data. TDI is sampled on the rising edge of TCLK.	I

2.1 Maximum Ratings

Table 19. Absolute Maximum Ratings^{1, 2}

Rating	Symbol	Value	Unit
Supply voltage	V_{DD}	−0.3 to +4.0	V
Clock synthesizer supply voltage	V_{DDPLL}	−0.3 to +4.0	V
Digital input voltage ³	V_{IN}	−0.3 to + 4.0	V
EXTAL pin voltage	V_{EXTAL}	0 to 3.3	V
XTAL pin voltage	V_{XTAL}	0 to 3.3	V
Instantaneous maximum current Single pin limit (applies to all pins) ^{4, 5}	I_{DD}	25	mA
Operating temperature range (packaged)	T_A ($T_L - T_H$)	−40 to 85	°C
Storage temperature range	T_{stg}	−65 to 150	°C

¹ Functional operating conditions are given in DC Electrical Specifications. Absolute Maximum Ratings are stress ratings only, and functional operation at the maxima is not guaranteed. Stress beyond those listed may affect device reliability or cause permanent damage to the device.

² This device contains circuitry protecting against damage due to high static voltage or electrical fields; however, it is advised that normal precautions be taken to avoid application of any voltages higher than maximum-rated voltages to this high-impedance circuit. Reliability of operation is enhanced if unused inputs are tied to an appropriate logic voltage level (e.g., either V_{SS} or V_{DD}).

³ Input must be current limited to the I_{DD} value specified. To determine the value of the required current-limiting resistor, calculate resistance values for positive and negative clamp voltages, then use the larger of the two values.

⁴ All functional non-supply pins are internally clamped to V_{SS} and V_{DD} .

⁵ The power supply must maintain regulation within operating V_{DD} range during instantaneous and operating maximum current conditions. If positive injection current ($V_{in} > V_{DD}$) is greater than I_{DD} , the injection current may flow out of V_{DD} and could result in external power supply going out of regulation. Ensure external V_{DD} load shunts current greater than maximum injection current. This is the greatest risk when the MCU is not consuming power (ex; no clock). The power supply must maintain regulation within operating V_{DD} range during instantaneous and operating maximum current conditions.

Table 20 lists thermal resistance values.

NOTE

The use of this device in one- or two-layer board designs is not recommended due to the limited thermal conductance provided by those boards.

Table 20. Thermal Characteristics

Characteristic	Symbol	Package ¹	Value	Unit
Junction to ambient, natural convection	θ_{JA}	80-pin LQFP, four-layer board	36.0 ^{2,3}	°C/W
		112-pin LQFP, four-layer board	35.0	
		121 MAPBGA, four-layer board	32	
		80-pin LQFP, one-layer board ¹	49.0 ¹	
		121 MAPBGA, one-layer board ¹	56 ¹	
		112-pin LQFP, one-layer board ¹	44.0 ¹	
Junction to ambient (@200 ft/min)	θ_{JMA}	80-pin LQFP, four-layer board	30.0	°C/W
		112-pin LQFP, four-layer board	29.0	
		121 MAPBGA, four-layer board	28	
		80-pin LQFP, one-layer board ¹	39.0 ¹	
		112-pin LQFP, one-layer board ¹	35.0 ¹	
		121 MAPBGA, one-layer board ¹	46 ¹	
Junction to board	θ_{JB}	80-pin LQFP	22.0 ⁴	°C/W
		112-pin LQFP	23.0	
		121 MAPBGA, four-layer board	18	
Junction to case	θ_{JC}	80-pin LQFP	6.0 ⁵	°C/W
		112-pin LQFP	6.0	
		121 MAPBGA	10	
Junction to top of package, natural convection	Ψ_{jt}	80-pin LQFP	2.0 ⁶	°C/W
		112-pin LQFP	2.0 ⁶	
		121 MAPBGA	2.0 ⁶	
Maximum operating junction temperature	T_j	All	130	°C

¹ The use of this device in one- or two-layer board designs is not recommended due to the limited thermal conductance provided by those boards.

² θ_{JMA} and Ψ_{jt} parameters are simulated in conformance with EIA/JESD Standard 51-2 for natural convection. Freescale recommends the use of θ_{JMA} and power dissipation specifications in the system design to prevent device junction temperatures from exceeding the rated specification. System designers should be aware that device junction temperatures can be significantly influenced by board layout and surrounding devices. Conformance to the device junction temperature specification can be verified by physical measurement in the customer's system using the Ψ_{jt} parameter, the device power dissipation, and the method described in EIA/JESD Standard 51-2.

³ Per JEDEC JESD51-6 with the board horizontal.

⁴ Thermal resistance between the die and the printed circuit board in conformance with JEDEC JESD51-8. Board temperature is measured on the top surface of the board near the package.

⁵ Thermal resistance between the die and the case top surface as measured by the cold plate method (MIL SPEC-883 Method 1012.1).

⁶ Thermal characterization parameter indicating the temperature difference between package top and the junction temperature per JEDEC JESD51-2. When Greek letters are not available, the thermal characterization parameter is written in conformance with Psi-JT.

Table 23. Active Current Consumption Specifications

Characteristic	Symbol	Typical				Peak	Unit
		Running from SRAM, EPHY Off	Running from Flash, EPHY Off	Running from Flash, EPHY 10BaseT	Running from Flash, EPHY 100BaseT		
Active current, core and I/O PLL @25 MHz PLL @60 MHz	$I_{DDR} + I_{DDX} + I_{DDA}$	75 130	82 138	150 220	260 310	290 340	mA
Analog supply current Normal operation Low-power STOP	I_{DDA}	20 15	20 15	20 15	20 15	30 50	mA μ A

Table 24. Current Consumption Specifications in Low-Power Modes¹

Mode ²	PLL @25 MHz (typical) ³	PLL @60 MHz (typical) ³	PLL @60 MHz (peak) ⁴	Unit
STOP mode 3 (STPMD[1:0]=11)	0.2		1.0	mA
STOP mode 2 (STPMD[1:0]=10)	7		—	
STOP mode 1 (STPMD[1:0]=01)	10	12	—	
STOP mode 0 (STPMD[1:0]=00)	10	12	—	
WAIT	16	27	—	
DOZE	16	27	—	
RUN	25	45	—	

¹ All values are measured with a 3.30 V power supply.

² Refer to the “Power Management” chapter in the *MCF52235 ColdFire® Integrated Microcontroller Reference Manual* for more information on low-power modes.

³ These values were obtained with CLKOUT and all peripheral clocks except for the CFM clock disabled prior to entering low-power mode. The tests were performed at room temperature. All code was executed from flash memory; running code from SRAM further reduces power consumption.

⁴ These values were obtained with CLKOUT and all peripheral clocks enabled. All code was executed from flash memory.

2.4 Phase Lock Loop Electrical Specifications

Table 25. Oscillator and PLL Electrical Specifications
 $(V_{DD} \text{ and } V_{DDPLL} = 2.7 \text{ to } 3.6 \text{ V}, V_{SS} = V_{SSPLL} = 0 \text{ V})$

Characteristic	Symbol	Min	Max	Unit
Clock Source Frequency Range of EXTAL Frequency Range • Crystal • External ¹	$f_{crystal}$ f_{ext}	0.5 0	25.0 60.0	MHz
PLL reference frequency range	f_{ref_pll}	2	10.0	MHz
System frequency ² External clock mode On-Chip PLL Frequency	f_{sys}	0 $f_{ref} / 32$	60 60	MHz
Loss of reference frequency ^{3, 5}	f_{LOR}	100	1000	kHz
Self clocked mode frequency ^{4, 5}	f_{SCM}	1	5	MHz
Crystal start-up time ^{5, 6}	t_{cst}	—	10	ms
EXTAL input high voltage Crystal reference External reference	V_{IHEXT}	$V_{DD} - 1.0$ 2.0	V_{DD} 3.0 ⁷	V
EXTAL input low voltage Crystal reference External reference	V_{ILEXT}	V_{SS} V_{SS}	1.0 0.8	V
XTAL output high voltage $I_{OH} = 1.0 \text{ mA}$	V_{OL}	$V_{DD} - 1.0$	—	V
XTAL output low voltage $I_{OL} = 1.0 \text{ mA}$	V_{OL}	—	0.5	V
XTAL load capacitance ⁸		—	—	pF
PLL lock time ^{5,9}	t_{ipll}	—	500	μs
Power-up to lock time ^{5, 7,9} With crystal reference Without crystal reference	t_{iplk}	— —	10.5 500	ms μs
Duty cycle of reference ⁵	t_{dc}	40	60	% f_{sys}
Frequency un-LOCK range	f_{UL}	-1.5	1.5	% f_{sys}
Frequency LOCK range	f_{LCK}	-0.75	0.75	% f_{sys}
CLKOUT period Jitter ^{5, 6, 8, 10,11} , measured at f_{sys} Max Peak-to-peak jitter (clock edge to clock edge) Long term jitter (averaged over 2 ms interval)	C_{jitter}	— —	10 0.01	% f_{sys}

¹ In external clock mode, it is possible to run the chip directly from an external clock source without enabling the PLL.

² All internal registers retain data at 0 Hz.

³ Loss of reference frequency is the reference frequency detected internally that transitions the PLL into self-clocked mode.

⁴ Self-clocked mode frequency is the frequency that the PLL operates at when the reference frequency falls below f_{LOR} with default MFD/RFD settings.

⁵ This parameter is characterized before qualification rather than 100% tested.

⁶ Proper PC board layout procedures must be followed to achieve specifications.

Table 29. I²C Output Timing Specifications between I2C_SCL and I2C_SDA

Num	Characteristic	Min	Max	Units
I1 ¹	Start condition hold time	$6 \times t_{CYC}$	—	ns
I2 ¹	Clock low period	$10 \times t_{CYC}$	—	ns
I3 ²	I2C_SCL/I2C_SDA rise time ($V_{IL} = 0.5 \text{ V}$ to $V_{IH} = 2.4 \text{ V}$)	—	—	μs
I4 ¹	Data hold time	$7 \times t_{CYC}$	—	ns
I5 ³	I2C_SCL/I2C_SDA fall time ($V_{IH} = 2.4 \text{ V}$ to $V_{IL} = 0.5 \text{ V}$)	—	3	ns
I6 ¹	Clock high time	$10 \times t_{CYC}$	—	ns
I7 ¹	Data setup time	$2 \times t_{CYC}$	—	ns
I8 ¹	Start condition setup time (for repeated start condition only)	$20 \times t_{CYC}$	—	ns
I9 ¹	Stop condition setup time	$10 \times t_{CYC}$	—	ns

¹ Output numbers depend on the value programmed into the IFDR; an IFDR programmed with the maximum frequency (IFDR = 0x20) results in minimum output timings as shown in Table 29. The I²C interface is designed to scale the actual data transition time to move it to the middle of the SCL low period. The actual position is affected by the prescale and division values programmed into the IFDR; however, the numbers given in Table 29 are minimum values.

² Because SCL and SDA are open-collector-type outputs, which the processor can only actively drive low, the time SCL or SDA take to reach a high level depends on external signal capacitance and pull-up resistor values.

³ Specified at a nominal 50-pF load.

Figure 8 shows timing for the values in Table 28 and Table 29.

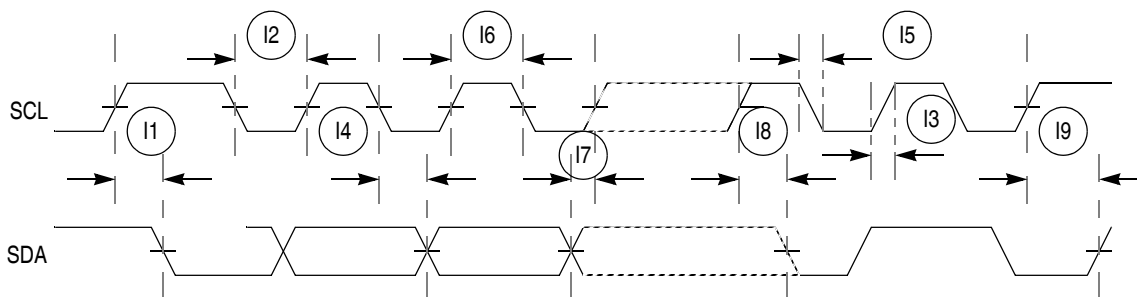

Figure 8. I²C Input/Output Timings

Table 35. ADC Parameters¹ (continued)

Name	Characteristic	Min	Typical	Max	Unit
f_{ADIC}	ADC internal clock	0.1	—	5.0	MHz
R_{AD}	Conversion range	V_{REFL}	—	V_{REFH}	V
t_{ADPU}	ADC power-up time ⁵	—	6	13	t_{AIC} cycles ⁶
t_{REC}	Recovery from auto standby	—	0	1	t_{AIC} cycles
t_{ADC}	Conversion time	—	6	—	t_{AIC} cycles
t_{ADS}	Sample time	—	1	—	t_{AIC} cycles
C_{ADI}	Input capacitance	—	See Figure 12	—	pF
X_{IN}	Input impedance	—	See Figure 12	—	W
I_{ADI}	Input injection current ⁷ , per pin	—	—	3	mA
I_{VREFH}	V_{REFH} current	—	0	—	mA
V_{OFFSET}	Offset voltage internal reference	—	± 11	± 15	mV
E_{GAIN}	Gain error (transfer path)	.99	1	1.01	—
V_{OFFSET}	Offset voltage external reference	—	± 3	—	mV
SNR	Signal-to-noise ratio	—	62 to 66	—	dB
THD	Total harmonic distortion	—	–75	—	dB
SFDR	Spurious free dynamic range	—	75	—	dB
SINAD	Signal-to-noise plus distortion	—	65	—	dB
ENOB	Effective number OF bits	9.1	10.6	—	Bits

¹ All measurements were made at $V_{\text{DD}} = 3.3\text{V}$, $V_{\text{REFH}} = 3.3\text{V}$, and $V_{\text{REFL}} = \text{ground}$

² INL measured from $V_{\text{IN}} = V_{\text{REFL}}$ to $V_{\text{IN}} = V_{\text{REFH}}$

³ LSB = Least Significant Bit

⁴ INL measured from $V_{\text{IN}} = 0.1V_{\text{REFH}}$ to $V_{\text{IN}} = 0.9V_{\text{REFH}}$

⁵ Includes power-up of ADC and V_{REF}

⁶ ADC clock cycles

⁷ The current that can be injected or sourced from an unselected ADC signal input without impacting the performance of the ADC

2.9.1 Equivalent Circuit for ADC Inputs

Figure 10-17 shows the ADC input circuit during sample and hold. S1 and S2 are always open/closed at the same time that S3 is closed/open. When S1/S2 are closed and S3 is open, one input of the sample and hold circuit moves to $(V_{\text{REFH}} - V_{\text{REFL}})/2$, while the other charges to the analog input voltage. When the switches are flipped, the charge on C1 and C2 are averaged via S3, with the result that a single-ended analog input is switched to a differential voltage centered about $(V_{\text{REFH}} - V_{\text{REFL}})/2$. The switches switch on every cycle of the ADC clock (open one-half ADC clock, closed one-half ADC clock). There are additional capacitances associated with the analog input pad, routing, etc., but these do not filter into the S/H output voltage, as S1 provides isolation during the charge-sharing phase. One aspect of this circuit is that there is an ongoing input current, which is a function of the analog input voltage, V_{REF} , and the ADC clock frequency.

2.13 JTAG and Boundary Scan Timing

Table 39. JTAG and Boundary Scan Timing

Num	Characteristics ¹	Symbol	Min	Max	Unit
J1	TCLK frequency of operation	f_{JCYC}	DC	1/4	$f_{sys}/2$
J2	TCLK cycle period	t_{JCYC}	$4 \times t_{CYC}$	—	ns
J3	TCLK clock pulse width	t_{JCW}	26	—	ns
J4	TCLK rise and fall times	t_{JCRF}	0	3	ns
J5	Boundary scan input data setup time to TCLK rise	t_{BSDST}	4	—	ns
J6	Boundary scan input data hold time after TCLK rise	t_{BSDHT}	26	—	ns
J7	TCLK low to boundary scan output data valid	t_{BSDV}	0	33	ns
J8	TCLK low to boundary scan output high Z	t_{BSDZ}	0	33	ns
J9	TMS, TDI input data setup time to TCLK rise	t_{TAPBST}	4	—	ns
J10	TMS, TDI input data hold time after TCLK rise	t_{TAPBHT}	10	—	ns
J11	TCLK low to TDO data valid	t_{TDODV}	0	26	ns
J12	TCLK low to TDO high Z	t_{TDODZ}	0	8	ns
J13	\overline{TRST} assert time	t_{TRSTAT}	100	—	ns
J14	\overline{TRST} setup time (negation) to TCLK high	t_{TRSTST}	10	—	ns

¹ JTAG_EN is expected to be a static signal. Hence, it is not associated with any timing.

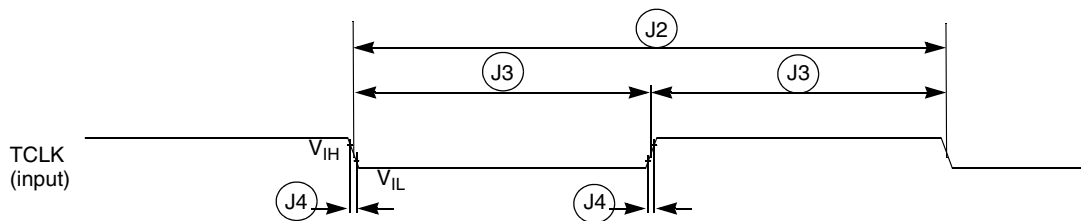


Figure 14. Test Clock Input Timing

2.14 Debug AC Timing Specifications

Table 40 lists specifications for the debug AC timing parameters shown in Figure 19.

Table 40. Debug AC Timing Specification

Num	Characteristic	60 MHz		Units
		Min	Max	
D1	PST, DDATA to CLKOUT setup	4	—	ns
D2	CLKOUT to PST, DDATA hold	1.5	—	ns
D3	DSI-to-DSCLK setup	$1 \times t_{CYC}$	—	ns
D4 ¹	DSCLK-to-DSO hold	$4 \times t_{CYC}$	—	ns
D5	DSCLK cycle time	$5 \times t_{CYC}$	—	ns
D6	\overline{BKPT} input data setup time to CLKOUT Rise	4	—	ns
D7	\overline{BKPT} input data hold time to CLKOUT Rise	1.5	—	ns
D8	CLKOUT high to \overline{BKPT} high Z	0.0	10.0	ns

¹ DSCLK and DSI are synchronized internally. D4 is measured from the synchronized DSCLK input relative to the rising edge of CLKOUT.

Figure 18 shows real-time trace timing for the values in Table 40.

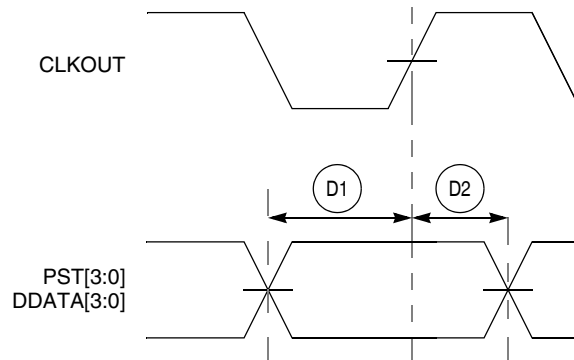


Figure 18. Real-Time Trace AC Timing

Figure 19 shows BDM serial port AC timing for the values in Table 40.

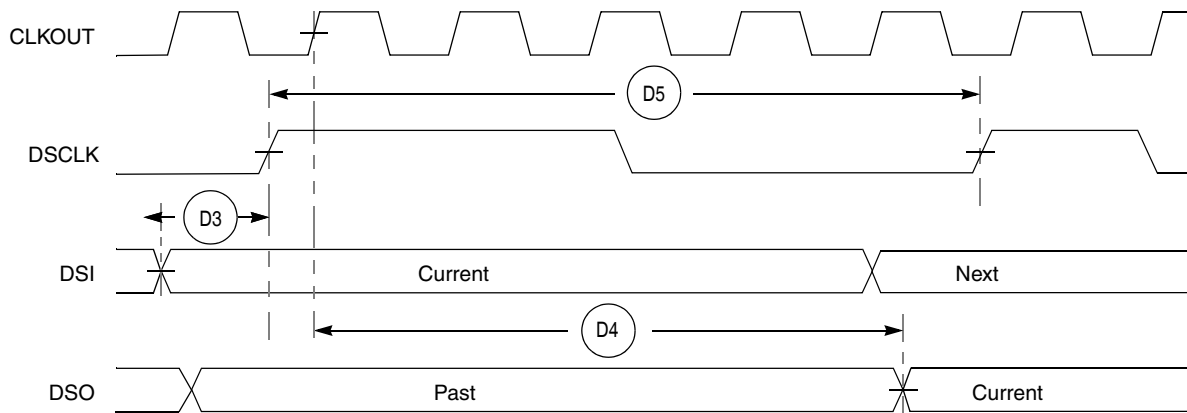
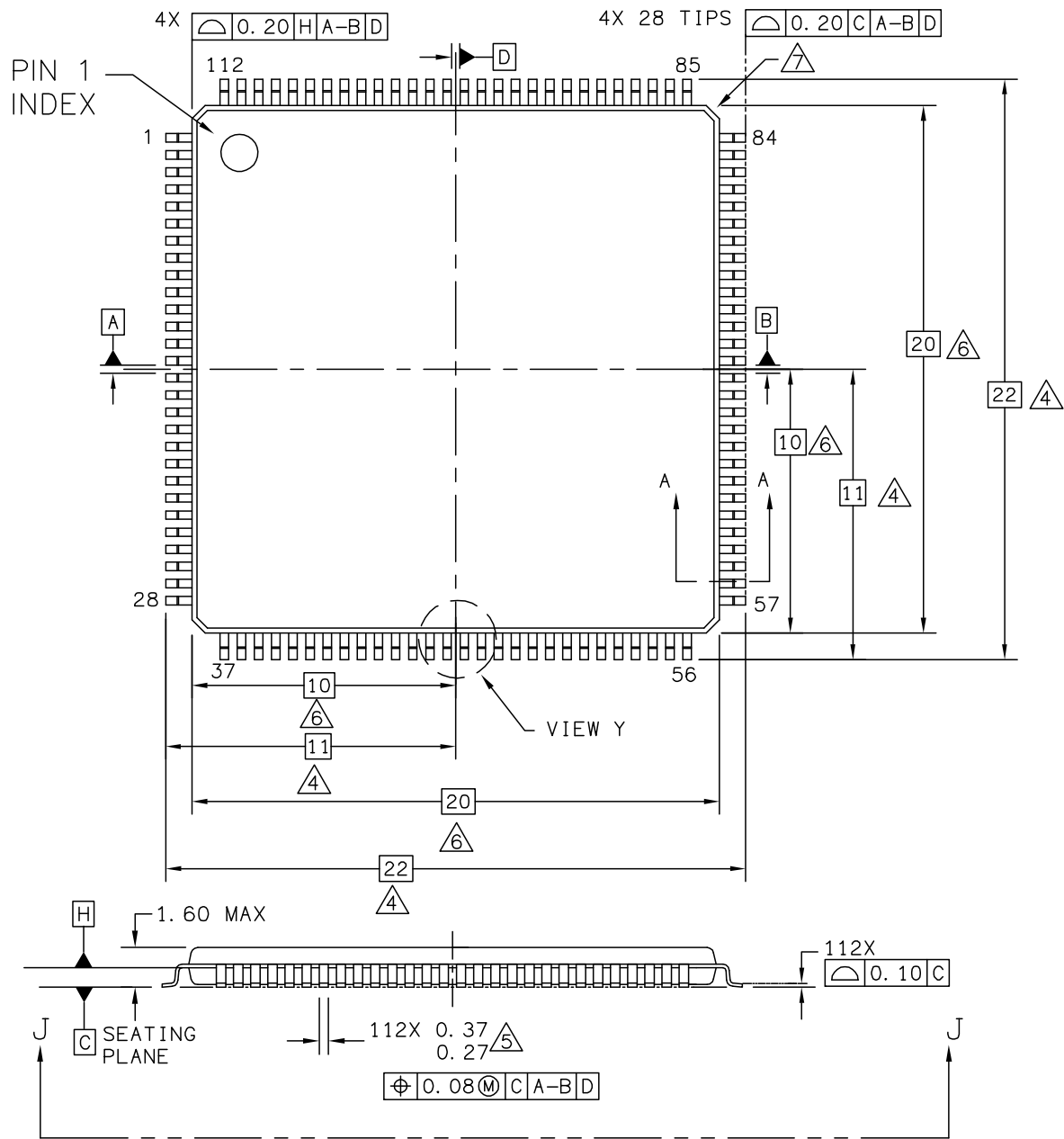


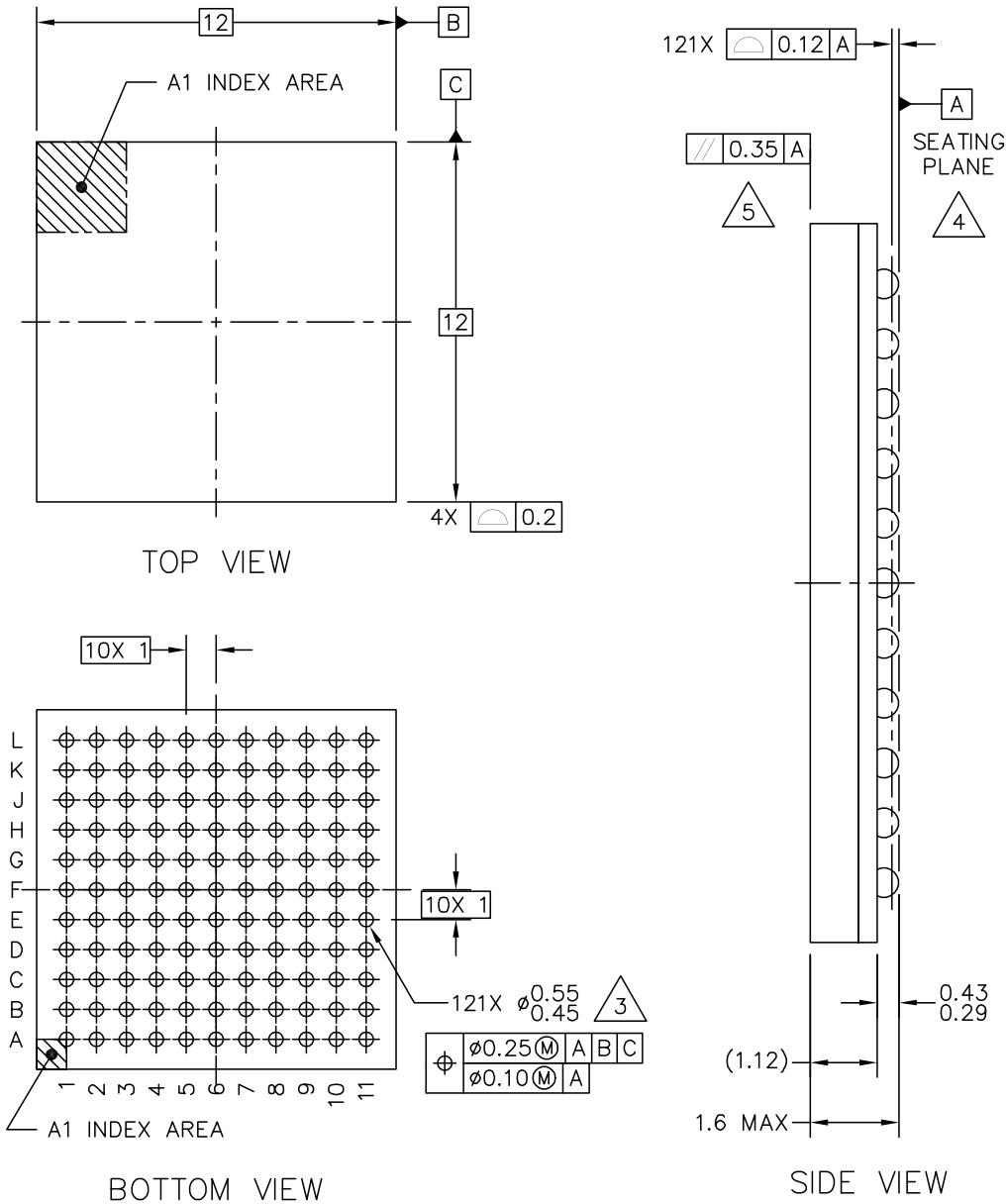
Figure 19. BDM Serial Port AC Timing

3.2 112-pin LQFP Package



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		CASE NUMBER: 987-02		25 MAY 2005
		STANDARD: JEDEC MS-026 BFA		

3.3 121 MAPBGA Package



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	CASE NUMBER: 1817-01		15 NOV 2005
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