

Welcome to [E-XFL.COM](https://www.e-xfl.com)

### What is "[Embedded - Microcontrollers](#)"?

"[Embedded - Microcontrollers](#)" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

### Applications of "[Embedded - Microcontrollers](#)"

#### Details

Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	20MHz
Connectivity	I <sup>2</sup> C, LINbus, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	17
Program Memory Size	14KB (8K x 14)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	512 x 8
Voltage - Supply (Vcc/Vdd)	2.3V ~ 5.5V
Data Converters	A/D 12x10b; D/A 1x5b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	20-SOIC (0.295", 7.50mm Width)
Supplier Device Package	20-SOIC
Purchase URL	<a href="https://www.e-xfl.com/product-detail/microchip-technology/pic16f1509-e-so">https://www.e-xfl.com/product-detail/microchip-technology/pic16f1509-e-so</a>

# PIC16(L)F1508/9

---

## 2.1 Automatic Interrupt Context Saving

During interrupts, certain registers are automatically saved in shadow registers and restored when returning from the interrupt. This saves stack space and user code. See **Section 7.5 “Automatic Context Saving”**, for more information.

## 2.2 16-Level Stack with Overflow and Underflow

These devices have a hardware stack memory 15 bits wide and 16 words deep. A Stack Overflow or Underflow will set the appropriate bit (STKOVF or STKUNF) in the PCON register, and if enabled, will cause a software Reset. See **Section 3.5 “Stack”** for more details.

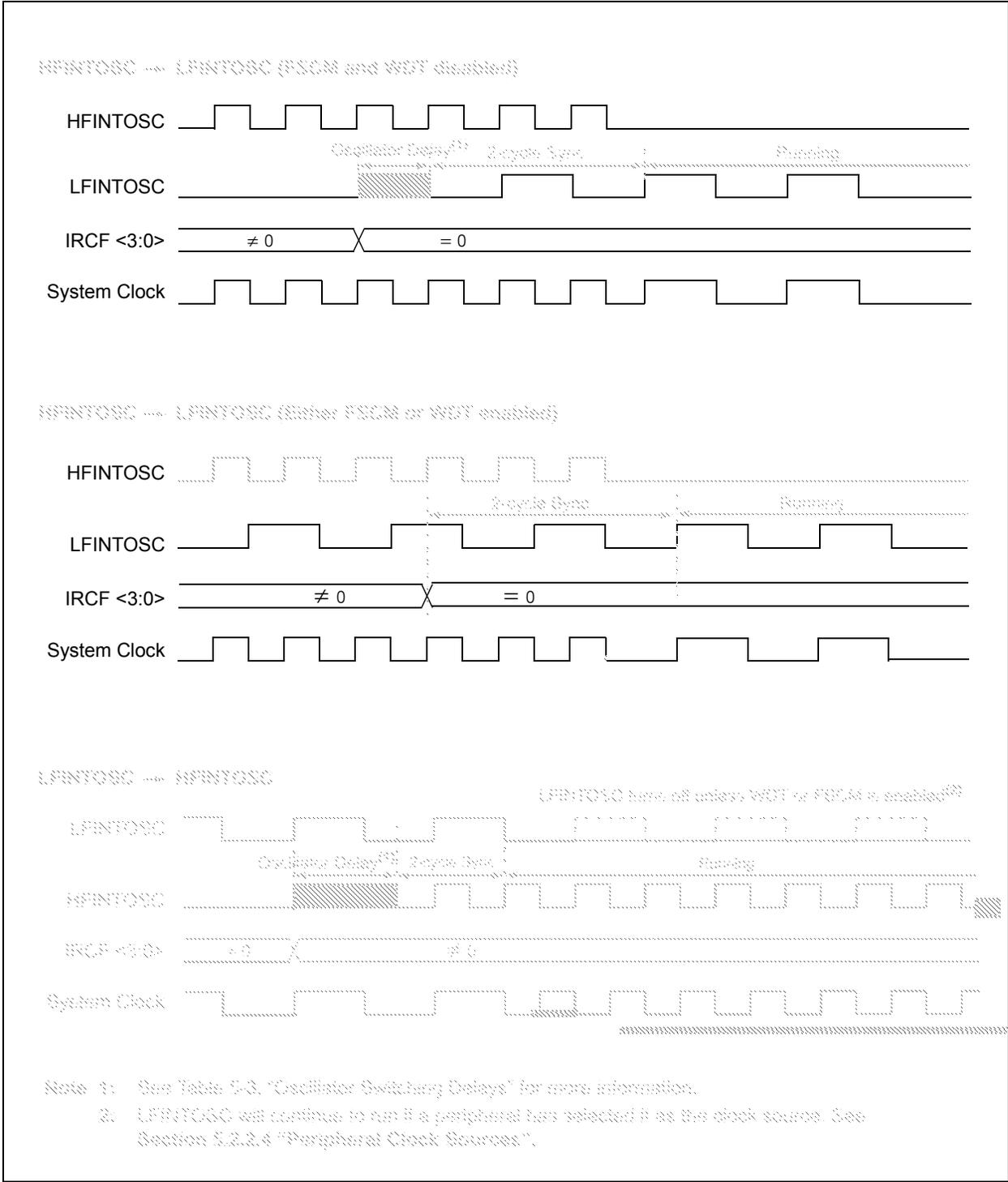
## 2.3 File Select Registers

There are two 16-bit File Select Registers (FSR). FSRs can access all file registers and program memory, which allows one Data Pointer for all memory. When an FSR points to program memory, there is one additional instruction cycle in instructions using INDF to allow the data to be fetched. General purpose memory can now also be addressed linearly, providing the ability to access contiguous data larger than 80 bytes. There are also new instructions to support the FSRs. See **Section 3.6 “Indirect Addressing”** for more details.

## 2.4 Instruction Set

There are 49 instructions for the enhanced mid-range CPU to support the features of the CPU. See **Section 28.0 “Instruction Set Summary”** for more details.

FIGURE 5-7: INTERNAL OSCILLATOR SWITCH TIMING



## 7.3 Interrupts During Sleep

Some interrupts can be used to wake from Sleep. To wake from Sleep, the peripheral must be able to operate without the system clock. The interrupt source must have the appropriate Interrupt Enable bit(s) set prior to entering Sleep.

On waking from Sleep, if the GIE bit is also set, the processor will branch to the interrupt vector. Otherwise, the processor will continue executing instructions after the `SLEEP` instruction. The instruction directly after the `SLEEP` instruction will always be executed before branching to the ISR. Refer to **Section 8.0 “Power-Down Mode (Sleep)”** for more details.

## 7.4 INT Pin

The INT pin can be used to generate an asynchronous edge-triggered interrupt. This interrupt is enabled by setting the INTE bit of the INTCON register. The INTEDG bit of the OPTION\_REG register determines on which edge the interrupt will occur. When the INTEDG bit is set, the rising edge will cause the interrupt. When the INTEDG bit is clear, the falling edge will cause the interrupt. The INTF bit of the INTCON register will be set when a valid edge appears on the INT pin. If the GIE and INTE bits are also set, the processor will redirect program execution to the interrupt vector.

## 7.5 Automatic Context Saving

Upon entering an interrupt, the return PC address is saved on the stack. Additionally, the following registers are automatically saved in the shadow registers:

- W register
- STATUS register (except for  $\overline{TO}$  and  $\overline{PD}$ )
- BSR register
- FSR registers
- PCLATH register

Upon exiting the Interrupt Service Routine, these registers are automatically restored. Any modifications to these registers during the ISR will be lost. If modifications to any of these registers are desired, the corresponding shadow register should be modified and the value will be restored when exiting the ISR. The shadow registers are available in Bank 31 and are readable and writable. Depending on the user's application, other registers may also need to be saved.

**TABLE 15-3: SUMMARY OF REGISTERS ASSOCIATED WITH ADC**

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
ADCON0	—	CHS<4:0>					GO/DONE	ADON	134
ADCON1	ADFM	ADCS<2:0>			—	—	ADPREF<1:0>		135
ADCON2	TRIGSEL<3:0>				—	—	—	—	136
ADRESH	ADC Result Register High								137, 138
ADRESL	ADC Result Register Low								137, 138
ANSELA	—	—	—	ANSA4	—	ANSA2	ANSA1	ANSA0	110
ANSELB	—	—	ANSB5	ANSB4	—	—	—	—	114
ANSELC	ANSC7	ANSC6	—	—	ANSC3	ANSC2	ANSC1	ANSC0	118
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	75
PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSP1IE	—	TMR2IE	TMR1IE	76
PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSP1IF	—	TMR2IF	TMR1IF	79
TRISA	—	—	TRISA5	TRISA4	— <sup>(1)</sup>	TRISA2	TRISA1	TRISA0	109
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	—	—	—	—	113
TRISC	TRISC7	TRISC6	TRISC5	TRISC4	TRISC3	TRISC2	TRISC1	TRISC0	117
FVRCON	FVREN	FVRRDY	TSEN	TSRNG	CDAFVR<1:0>		ADFVR<1:0>		125

**Legend:** x = unknown, u = unchanged, — = unimplemented read as '0', α = value depends on condition. Shaded cells are not used for ADC module.

**Note 1:** Unimplemented, read as '1'.

## 19.9 Register Definitions: Timer1 Control

**REGISTER 19-1: T1CON: TIMER1 CONTROL REGISTER**

R/W-0/u	R/W-0/u	R/W-0/u	R/W-0/u	R/W-0/u	R/W-0/u	U-0	R/W-0/u
TMR1CS<1:0>		T1CKPS<1:0>		T1OSCEN	$\overline{T1SYNC}$	—	TMR1ON
bit 7							bit 0

**Legend:**

R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

- bit 7-6     **TMR1CS<1:0>**: Timer1 Clock Source Select bits  
 11 = Timer1 clock source is LFINTOSC  
 10 = Timer1 clock source is pin or oscillator:  
     If **T1OSCEN** = 0:  
     External clock from T1CKI pin (on the rising edge)  
     If **T1OSCEN** = 1:  
     Crystal oscillator on SOSC1/SOSCO pins  
 01 = Timer1 clock source is system clock (Fosc)  
 00 = Timer1 clock source is instruction clock (Fosc/4)
- bit 5-4     **T1CKPS<1:0>**: Timer1 Input Clock Prescale Select bits  
 11 = 1:8 Prescale value  
 10 = 1:4 Prescale value  
 01 = 1:2 Prescale value  
 00 = 1:1 Prescale value
- bit 3       **T1OSCEN**: LP Oscillator Enable Control bit  
 1 = Secondary oscillator circuit enabled for Timer1  
 0 = Secondary oscillator circuit disabled for Timer1
- bit 2        **$\overline{T1SYNC}$** : Timer1 Synchronization Control bit  
 1 = Do not synchronize asynchronous clock input  
 0 = Synchronize asynchronous clock input with system clock (Fosc)
- bit 1       **Unimplemented**: Read as '0'
- bit 0       **TMR1ON**: Timer1 On bit  
 1 = Enables Timer1  
 0 = Stops Timer1 and clears Timer1 gate flip-flop

**TABLE 19-5: SUMMARY OF REGISTERS ASSOCIATED WITH TIMER1**

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
ANSELA	—	—	—	ANSA4	—	ANSA2	ANSA1	ANSA0	110
APFCON	—	—	—	SSEL	T1GSEL	—	CLC1SEL	NCO1SEL	107
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	75
OSCSTAT	SOSCR	—	OSTS	HFIOFR	—	—	LFIOFR	HFIOFS	60
PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSP1IE	—	TMR2IE	TMR1IE	76
PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSP1IF	—	TMR2IF	TMR1IF	79
TMR1H	Holding Register for the Most Significant Byte of the 16-bit TMR1 Count								159*
TMR1L	Holding Register for the Least Significant Byte of the 16-bit TMR1 Count								159*
TRISA	—	—	TRISA5	TRISA4	— <sup>(1)</sup>	TRISA2	TRISA1	TRISA0	109
T1CON	TMR1CS<1:0>		T1CKPS<1:0>		T1OSCEN	T1SYNC	—	TMR1ON	163
T1GCON	TMR1GE	T1GPOL	T1GTM	T1GSPM	T1GGO/ DONE	T1GVAL	T1GSS<1:0>		164

**Legend:** — = unimplemented location, read as '0'. Shaded cells are not used by the Timer1 module.

\* Page provides register information.

**Note 1:** Unimplemented, read as '1'.

# PIC16(L)F1508/9

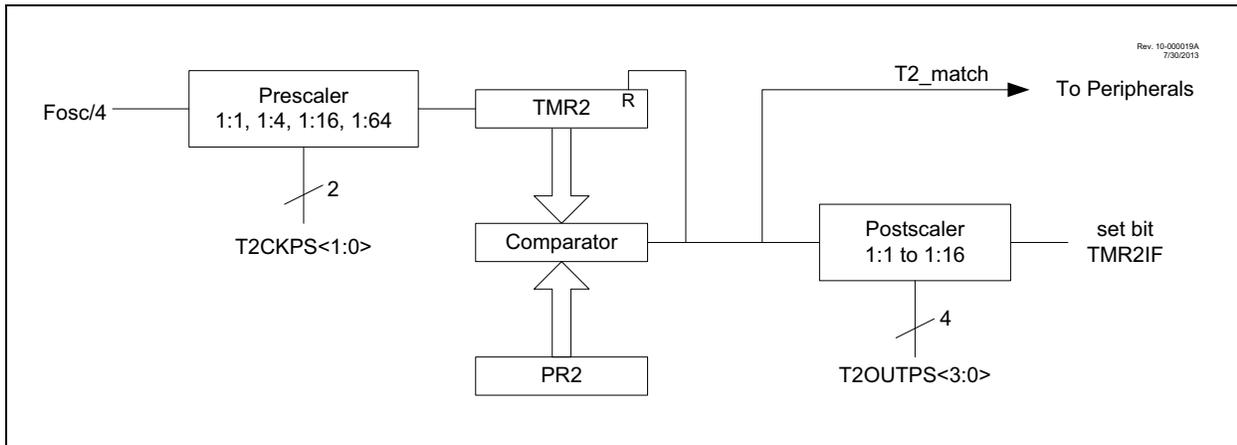
## 20.0 TIMER2 MODULE

The Timer2 module incorporates the following features:

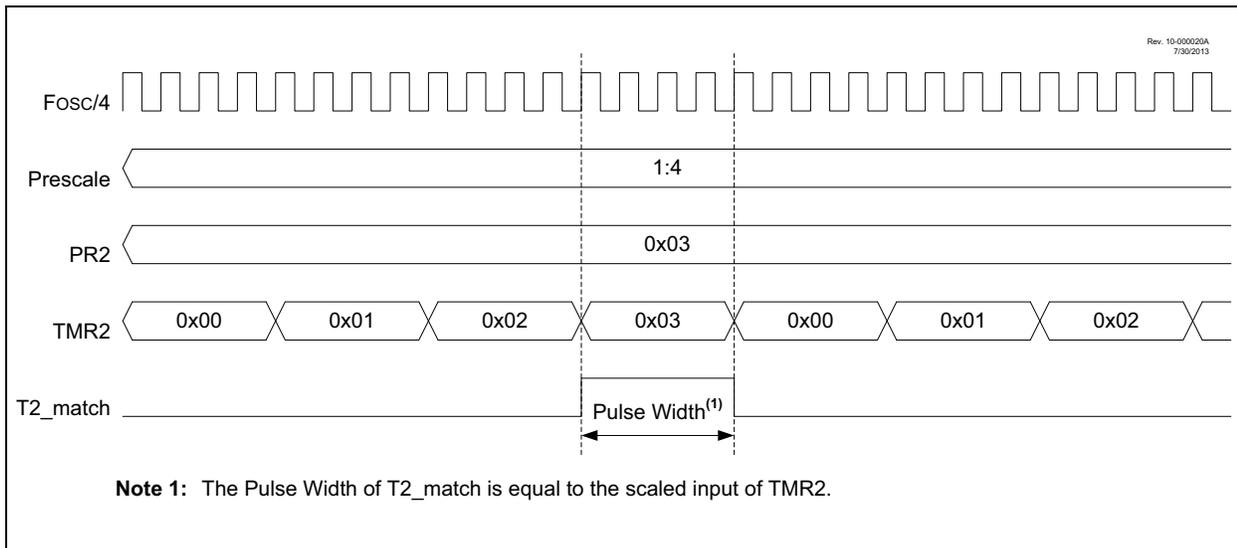
- 8-bit Timer and Period registers (TMR2 and PR2, respectively)
- Readable and writable (both registers)
- Software programmable prescaler (1:1, 1:4, 1:16, and 1:64)
- Software programmable postscaler (1:1 to 1:16)
- Interrupt on TMR2 match with PR2

See Figure 20-1 for a block diagram of Timer2.

**FIGURE 20-1: TIMER2 BLOCK DIAGRAM**



**FIGURE 20-2: TIMER2 TIMING DIAGRAM**



# PIC16(L)F1508/9

## 21.2.3 SPI MASTER MODE

The master can initiate the data transfer at any time because it controls the SCKx line. The master determines when the slave (Processor 2, Figure 21-5) is to broadcast data by the software protocol.

In Master mode, the data is transmitted/received as soon as the SSPxBUF register is written to. If the SPI is only going to receive, the SDOx output could be disabled (programmed as an input). The SSPxSR register will continue to shift in the signal present on the SDIx pin at the programmed clock rate. As each byte is received, it will be loaded into the SSPxBUF register as if a normal received byte (interrupts and Status bits appropriately set).

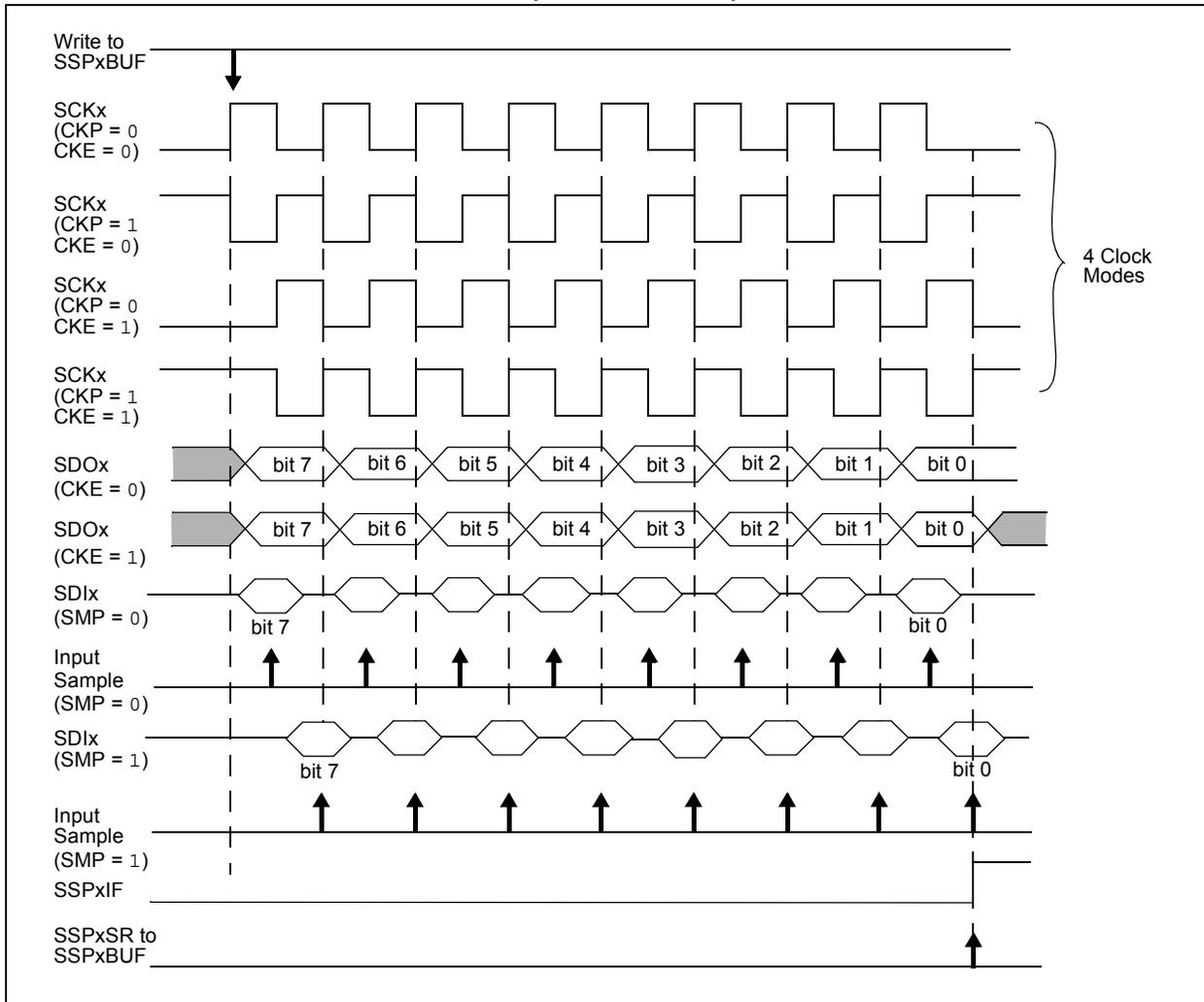
The clock polarity is selected by appropriately programming the CKP bit of the SSPxCON1 register and the CKE bit of the SSPxSTAT register. This then, would give waveforms for SPI communication as shown in Figure 21-6, Figure 21-8, Figure 21-9 and Figure 21-10, where the MSb is transmitted first. In Master mode, the SPI clock rate (bit rate) is user programmable to be one of the following:

- $F_{osc}/4$  (or  $T_{CY}$ )
- $F_{osc}/16$  (or  $4 * T_{CY}$ )
- $F_{osc}/64$  (or  $16 * T_{CY}$ )
- Timer2 output/2
- $F_{osc}/(4 * (SSPxADD + 1))$

Figure 21-6 shows the waveforms for Master mode.

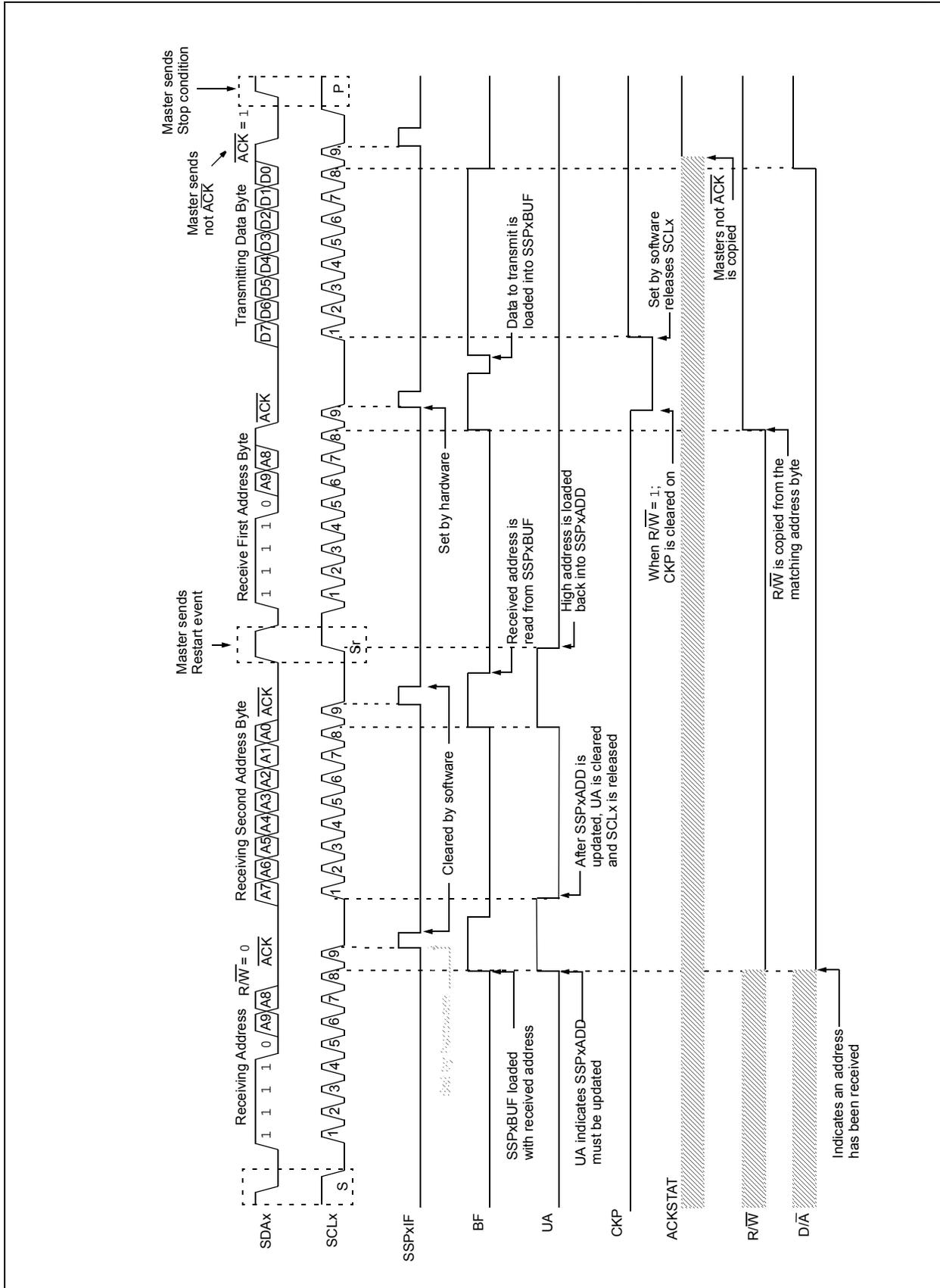
When the CKE bit is set, the SDOx data is valid before there is a clock edge on SCKx. The change of the input sample is shown based on the state of the SMP bit. The time when the SSPxBUF is loaded with the received data is shown.

**FIGURE 21-6: SPI MODE WAVEFORM (MASTER MODE)**



# PIC16(L)F1508/9

FIGURE 21-22: I<sup>2</sup>C SLAVE, 10-BIT ADDRESS, TRANSMISSION (SEN = 0, AHEN = 0, DHEN = 0)



## 21.5.6 CLOCK STRETCHING

Clock stretching occurs when a device on the bus holds the SCLx line low, effectively pausing communication. The slave may stretch the clock to allow more time to handle data or prepare a response for the master device. A master device is not concerned with stretching as anytime it is active on the bus and not transferring data it is stretching. Any stretching done by a slave is invisible to the master software and handled by the hardware that generates SCLx.

The CKP bit of the SSPxCON1 register is used to control stretching in software. Any time the CKP bit is cleared, the module will wait for the SCLx line to go low and then hold it. Setting CKP will release SCLx and allow more communication.

### 21.5.6.1 Normal Clock Stretching

Following an  $\overline{\text{ACK}}$  if the R/W bit of SSPxSTAT is set, a read request, the slave hardware will clear CKP. This allows the slave time to update SSPxBUF with data to transfer to the master. If the SEN bit of SSPxCON2 is set, the slave hardware will always stretch the clock after the  $\overline{\text{ACK}}$  sequence. Once the slave is ready, CKP is set by software and communication resumes.

**Note 1:** The BF bit has no effect on if the clock will be stretched or not. This is different than previous versions of the module that would not stretch the clock, clear CKP, if SSPxBUF was read before the ninth falling edge of SCLx.

**2:** Previous versions of the module did not stretch the clock for a transmission if SSPxBUF was loaded before the ninth falling edge of SCLx. It is now always cleared for read requests.

### 21.5.6.2 10-bit Addressing Mode

In 10-bit Addressing mode, when the UA bit is set, the clock is always stretched. This is the only time the SCLx is stretched without CKP being cleared. SCLx is released immediately after a write to SSPxADD.

**Note:** Previous versions of the module did not stretch the clock if the second address byte did not match.

### 21.5.6.3 Byte NACKing

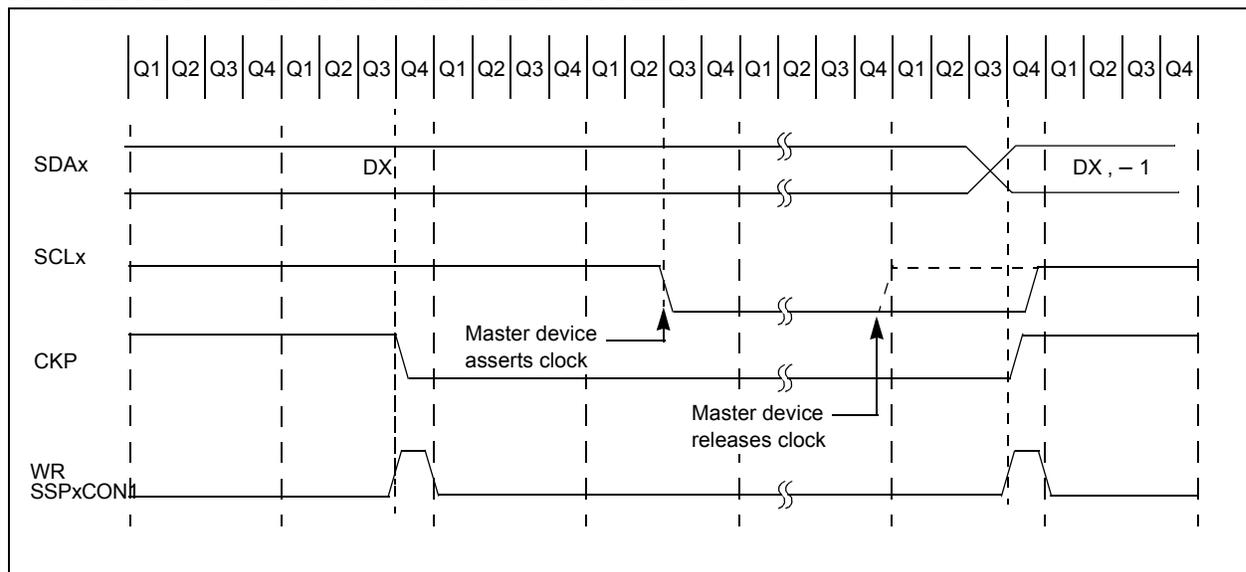
When the AHEN bit of SSPxCON3 is set; CKP is cleared by hardware after the eighth falling edge of SCLx for a received matching address byte. When the DHEN bit of SSPxCON3 is set, CKP is cleared after the eighth falling edge of SCLx for received data.

Stretching after the eighth falling edge of SCLx allows the slave to look at the received address or data and decide if it wants to ACK the received data.

## 21.5.7 CLOCK SYNCHRONIZATION AND THE CKP BIT

Any time the CKP bit is cleared, the module will wait for the SCLx line to go low and then hold it. However, clearing the CKP bit will not assert the SCLx output low until the SCLx output is already sampled low. Therefore, the CKP bit will not assert the SCLx line until an external I<sup>2</sup>C master device has already asserted the SCLx line. The SCLx output will remain low until the CKP bit is set and all other devices on the I<sup>2</sup>C bus have released SCLx. This ensures that a write to the CKP bit will not violate the minimum high time requirement for SCLx (see Figure 21-23).

**FIGURE 21-23: CLOCK SYNCHRONIZATION TIMING**



# PIC16(L)F1508/9

## 21.6.5 I<sup>2</sup>C MASTER MODE REPEATED START CONDITION TIMING

A Repeated Start condition (Figure 21-27) occurs when the RSEN bit of the SSPxCON2 register is programmed high and the master state machine is no longer active. When the RSEN bit is set, the SCLx pin is asserted low. When the SCLx pin is sampled low, the Baud Rate Generator is loaded and begins counting. The SDAx pin is released (brought high) for one Baud Rate Generator count (TBRG). When the Baud Rate Generator times out, if SDAx is sampled high, the SCLx pin will be deasserted (brought high). When SCLx is sampled high, the Baud Rate Generator is reloaded and begins counting. SDAx and SCLx must be sampled high for one TBRG. This action is then followed by assertion of the SDAx pin (SDAx = 0) for one TBRG while SCLx is high. SCLx is asserted low. Following this, the RSEN bit of the SSPxCON2 register will be

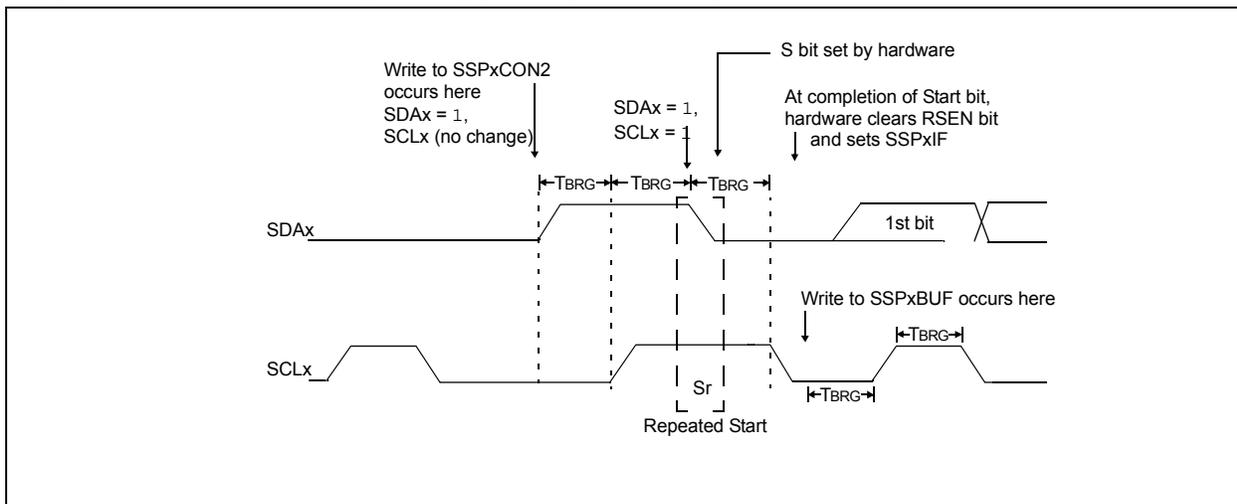
automatically cleared and the Baud Rate Generator will not be reloaded, leaving the SDAx pin held low. As soon as a Start condition is detected on the SDAx and SCLx pins, the S bit of the SSPxSTAT register will be set. The SSPxIF bit will not be set until the Baud Rate Generator has timed out.

**Note 1:** If RSEN is programmed while any other event is in progress, it will not take effect.

**2:** A bus collision during the Repeated Start condition occurs if:

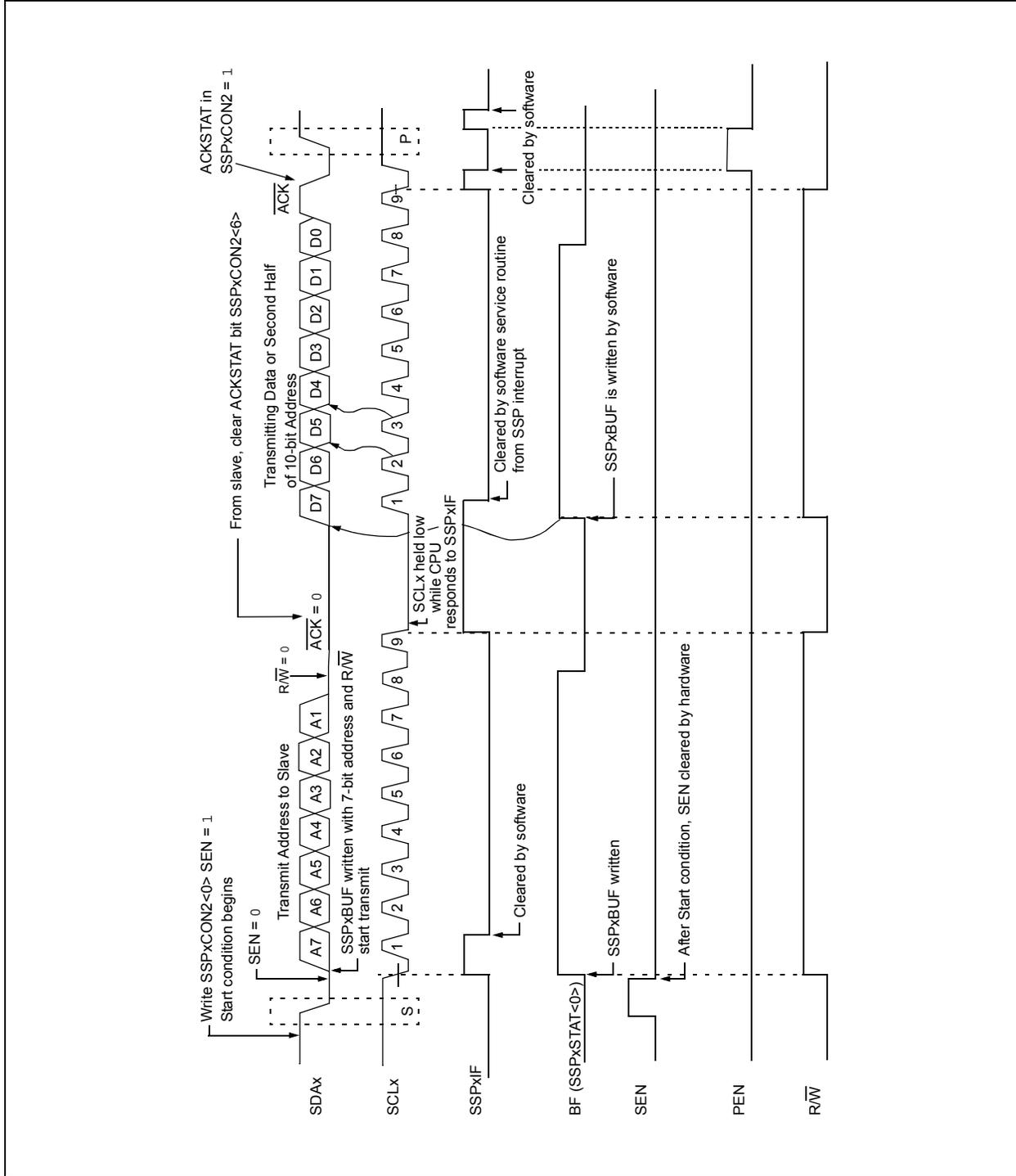
- SDAx is sampled low when SCLx goes from low-to-high.
- SCLx goes low before SDAx is asserted low. This may indicate that another master is attempting to transmit a data '1'.

**FIGURE 21-27: REPEAT START CONDITION WAVEFORM**



# PIC16(L)F1508/9

FIGURE 21-28: I<sup>2</sup>C MASTER MODE WAVEFORM (TRANSMISSION, 7 OR 10-BIT ADDRESS)



## 21.7 BAUD RATE GENERATOR

The MSSP module has a Baud Rate Generator available for clock generation in both I<sup>2</sup>C and SPI Master modes. The Baud Rate Generator (BRG) reload value is placed in the SSPxADD register (Register 21-6). When a write occurs to SSPxBUF, the Baud Rate Generator will automatically begin counting down.

Once the given operation is complete, the internal clock will automatically stop counting and the clock pin will remain in its last state.

An internal signal “Reload” in Figure 21-40 triggers the value from SSPxADD to be loaded into the BRG counter. This occurs twice for each oscillation of the

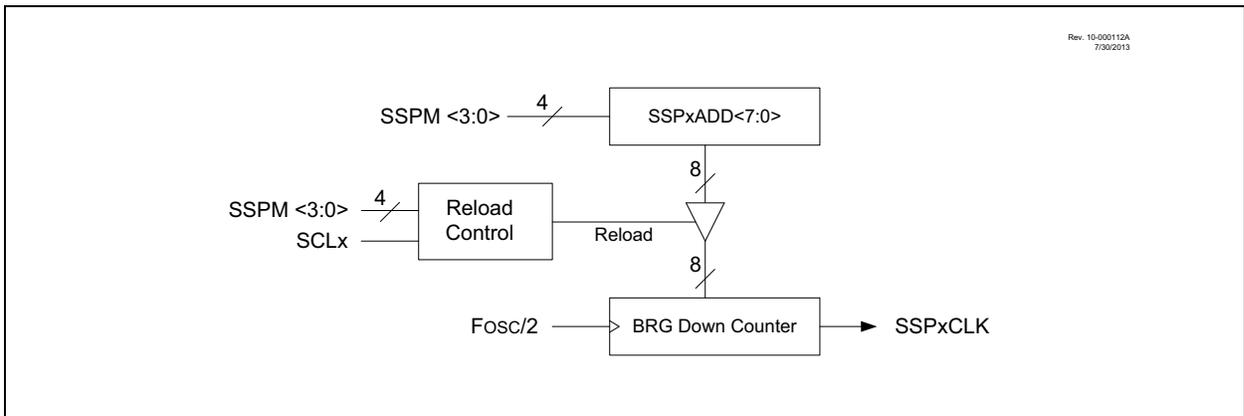
module clock line. The logic dictating when the reload signal is asserted depends on the mode the MSSP is being operated in.

Table 21-4 demonstrates clock rates based on instruction cycles and the BRG value loaded into SSPxADD.

**EQUATION 21-1:**

$$F_{CLOCK} = \frac{F_{OSC}}{(SSPxADD + 1)(4)}$$

**FIGURE 21-40: BAUD RATE GENERATOR BLOCK DIAGRAM**



**Note:** Values of 0x00, 0x01 and 0x02 are not valid for SSPxADD when used as a Baud Rate Generator for I<sup>2</sup>C. This is an implementation limitation.

**TABLE 21-4: MSSP CLOCK RATE W/BRG**

Fosc	Fcy	BRG Value	F <sub>CLOCK</sub> (Two Rollovers of BRG)
16 MHz	4 MHz	09h	400 kHz
16 MHz	4 MHz	0Ch	308 kHz
16 MHz	4 MHz	27h	100 kHz
4 MHz	1 MHz	09h	100 kHz

**Note:** Refer to the I/O port electrical and timing specifications in Table 29-9 and Figure 29-7 to ensure the system is designed to support the I/O timing requirements.

**TABLE 22-1: SUMMARY OF REGISTERS ASSOCIATED WITH ASYNCHRONOUS TRANSMISSION**

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on Page
BAUDCON	ABDOVF	RCIDL	—	SCKP	BRG16	—	WUE	ABDEN	235
INTCON	GIE	PEIE	TMR0IE	INTE	IOCIE	TMR0IF	INTF	IOCIF	75
PIE1	TMR1GIE	ADIE	RCIE	TXIE	SSP1IE	—	TMR2IE	TMR1IE	76
PIR1	TMR1GIF	ADIF	RCIF	TXIF	SSP1IF	—	TMR2IF	TMR1IF	79
RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	234*
SPBRGL	BRG<7:0>								236*
SPBRGH	BRG<15:8>								236*
TRISB	TRISB7	TRISB6	TRISB5	TRISB4	TRISB3	TRISB2	TRISB1	TRISB0	113
TXREG	EUSART Transmit Data Register								225
TXSTA	CSRC	TX9	TXEN	SYNC	SENDER	BRGH	TRMT	TX9D	233

**Legend:** — = unimplemented location, read as '0'. Shaded cells are not used for asynchronous transmission.

\* Page provides register information.

# PIC16(L)F1508/9

**TABLE 22-5: BAUD RATES FOR ASYNCHRONOUS MODES**

BAUD RATE	SYNC = 0, BRGH = 0, BRG16 = 0											
	Fosc = 20.000 MHz			Fosc = 18.432 MHz			Fosc = 16.000 MHz			Fosc = 11.0592 MHz		
	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)
300	—	—	—	—	—	—	—	—	—	—	—	—
1200	1221	1.73	255	1200	0.00	239	1202	0.16	207	1200	0.00	143
2400	2404	0.16	129	2400	0.00	119	2404	0.16	103	2400	0.00	71
9600	9470	-1.36	32	9600	0.00	29	9615	0.16	25	9600	0.00	17
10417	10417	0.00	29	10286	-1.26	27	10417	0.00	23	10165	-2.42	16
19.2k	19.53k	1.73	15	19.20k	0.00	14	19.23k	0.16	12	19.20k	0.00	8
57.6k	—	—	—	57.60k	0.00	7	—	—	—	57.60k	0.00	2
115.2k	—	—	—	—	—	—	—	—	—	—	—	—

BAUD RATE	SYNC = 0, BRGH = 0, BRG16 = 0											
	Fosc = 8.000 MHz			Fosc = 4.000 MHz			Fosc = 3.6864 MHz			Fosc = 1.000 MHz		
	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)
300	—	—	—	300	0.16	207	300	0.00	191	300	0.16	51
1200	1202	0.16	103	1202	0.16	51	1200	0.00	47	1202	0.16	12
2400	2404	0.16	51	2404	0.16	25	2400	0.00	23	—	—	—
9600	9615	0.16	12	—	—	—	9600	0.00	5	—	—	—
10417	10417	0.00	11	10417	0.00	5	—	—	—	—	—	—
19.2k	—	—	—	—	—	—	19.20k	0.00	2	—	—	—
57.6k	—	—	—	—	—	—	57.60k	0.00	0	—	—	—
115.2k	—	—	—	—	—	—	—	—	—	—	—	—

BAUD RATE	SYNC = 0, BRGH = 1, BRG16 = 0											
	Fosc = 20.000 MHz			Fosc = 18.432 MHz			Fosc = 16.000 MHz			Fosc = 11.0592 MHz		
	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)	Actual Rate	% Error	SPBRG value (decimal)
300	—	—	—	—	—	—	—	—	—	—	—	—
1200	—	—	—	—	—	—	—	—	—	—	—	—
2400	—	—	—	—	—	—	—	—	—	—	—	—
9600	9615	0.16	129	9600	0.00	119	9615	0.16	103	9600	0.00	71
10417	10417	0.00	119	10378	-0.37	110	10417	0.00	95	10473	0.53	65
19.2k	19.23k	0.16	64	19.20k	0.00	59	19.23k	0.16	51	19.20k	0.00	35
57.6k	56.82k	-1.36	21	57.60k	0.00	19	58.82k	2.12	16	57.60k	0.00	11
115.2k	113.64k	-1.36	10	115.2k	0.00	9	111.1k	-3.55	8	115.2k	0.00	5

## REGISTER 24-5: CLCxGLS0: GATE 1 LOGIC SELECT REGISTER

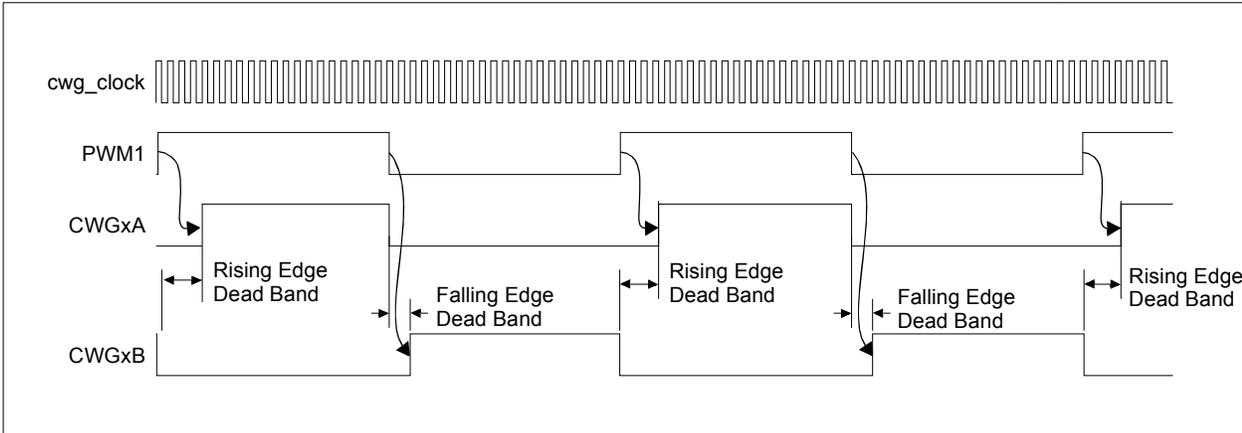
| R/W-x/u  |
|----------|----------|----------|----------|----------|----------|----------|----------|
| LCxG1D4T | LCxG1D4N | LCxG1D3T | LCxG1D3N | LCxG1D2T | LCxG1D2N | LCxG1D1T | LCxG1D1N |
| bit 7    |          |          |          |          |          |          | bit 0    |

### Legend:

R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

- bit 7      **LCxG1D4T:** Gate 1 Data 4 True (non-inverted) bit  
1 = lcx4T is gated into lcxg1  
0 = lcx4T is not gated into lcxg1
- bit 6      **LCxG1D4N:** Gate 1 Data 4 Negated (inverted) bit  
1 = lcx4N is gated into lcxg1  
0 = lcx4N is not gated into lcxg1
- bit 5      **LCxG1D3T:** Gate 1 Data 3 True (non-inverted) bit  
1 = lcx3T is gated into lcxg1  
0 = lcx3T is not gated into lcxg1
- bit 4      **LCxG1D3N:** Gate 1 Data 3 Negated (inverted) bit  
1 = lcx3N is gated into lcxg1  
0 = lcx3N is not gated into lcxg1
- bit 3      **LCxG1D2T:** Gate 1 Data 2 True (non-inverted) bit  
1 = lcx2T is gated into lcxg1  
0 = lcx2T is not gated into lcxg1
- bit 2      **LCxG1D2N:** Gate 1 Data 2 Negated (inverted) bit  
1 = lcx2N is gated into lcxg1  
0 = lcx2N is not gated into lcxg1
- bit 1      **LCxG1D1T:** Gate 1 Data 1 True (non-inverted) bit  
1 = lcx1T is gated into lcxg1  
0 = lcx1T is not gated into lcxg1
- bit 0      **LCxG1D1N:** Gate 1 Data 1 Negated (inverted) bit  
1 = lcx1N is gated into lcxg1  
0 = lcx1N is not gated into lcxg1

**FIGURE 26-2: TYPICAL CWG OPERATION WITH PWM1 (NO AUTO-SHUTDOWN)**



## 26.5 Dead-Band Control

Dead-band control provides for non-overlapping output signals to prevent shoot-through current in power switches. The CWG contains two 6-bit dead-band counters. One dead-band counter is used for the rising edge of the input source control. The other is used for the falling edge of the input source control.

Dead band is timed by counting CWG clock periods from zero up to the value in the rising or falling dead-band counter registers. See CWGxDBR and CWGxDBF registers (Register 26-4 and Register 26-5, respectively).

## 26.6 Rising Edge Dead Band

The rising edge dead-band delays the turn-on of the CWGxA output from when the CWGxB output is turned off. The rising edge dead-band time starts when the rising edge of the input source signal goes true. When this happens, the CWGxB output is immediately turned off and the rising edge dead-band delay time starts. When the rising edge dead-band delay time is reached, the CWGxA output is turned on.

The CWGxDBR register sets the duration of the dead-band interval on the rising edge of the input source signal. This duration is from 0 to 64 counts of dead band.

Dead band is always counted off the edge on the input source signal. A count of 0 (zero), indicates that no dead band is present.

If the input source signal is not present for enough time for the count to be completed, no output will be seen on the respective output.

## 26.7 Falling Edge Dead Band

The falling edge dead band delays the turn-on of the CWGxB output from when the CWGxA output is turned off. The falling edge dead-band time starts when the falling edge of the input source goes true. When this happens, the CWGxA output is immediately turned off and the falling edge dead-band delay time starts. When the falling edge dead-band delay time is reached, the CWGxB output is turned on.

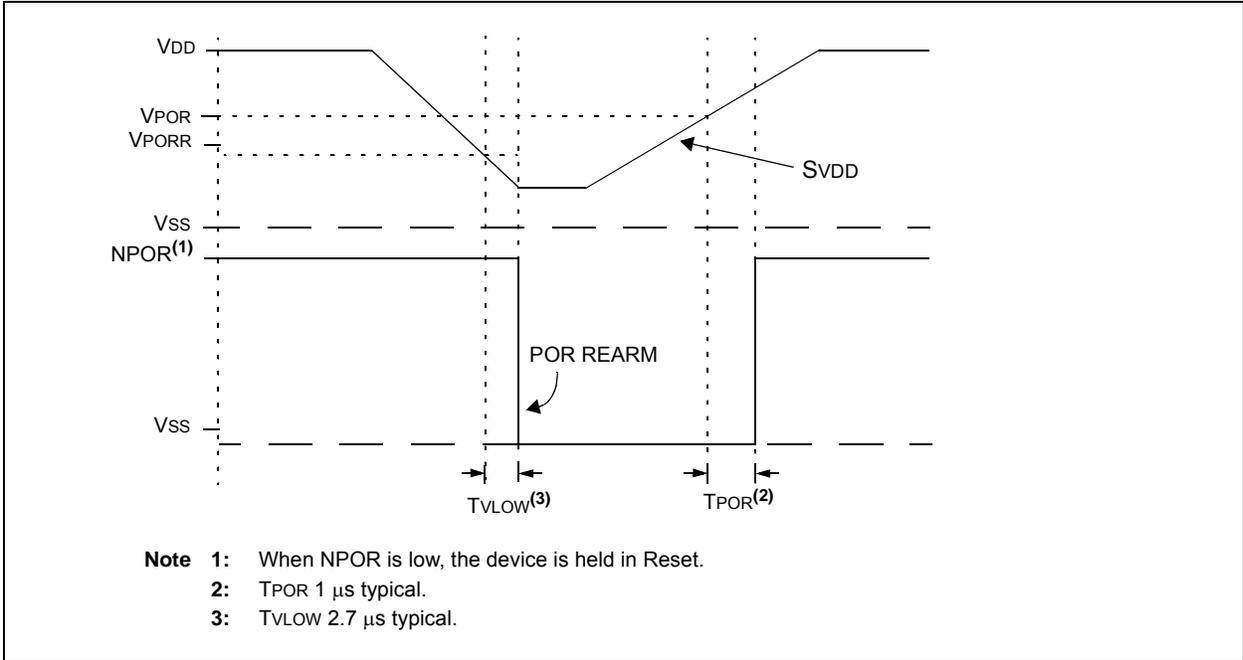
The CWGxDBF register sets the duration of the dead-band interval on the falling edge of the input source signal. This duration is from 0 to 64 counts of dead band.

Dead band is always counted off the edge on the input source signal. A count of 0 (zero), indicates that no dead band is present.

If the input source signal is not present for enough time for the count to be completed, no output will be seen on the respective output.

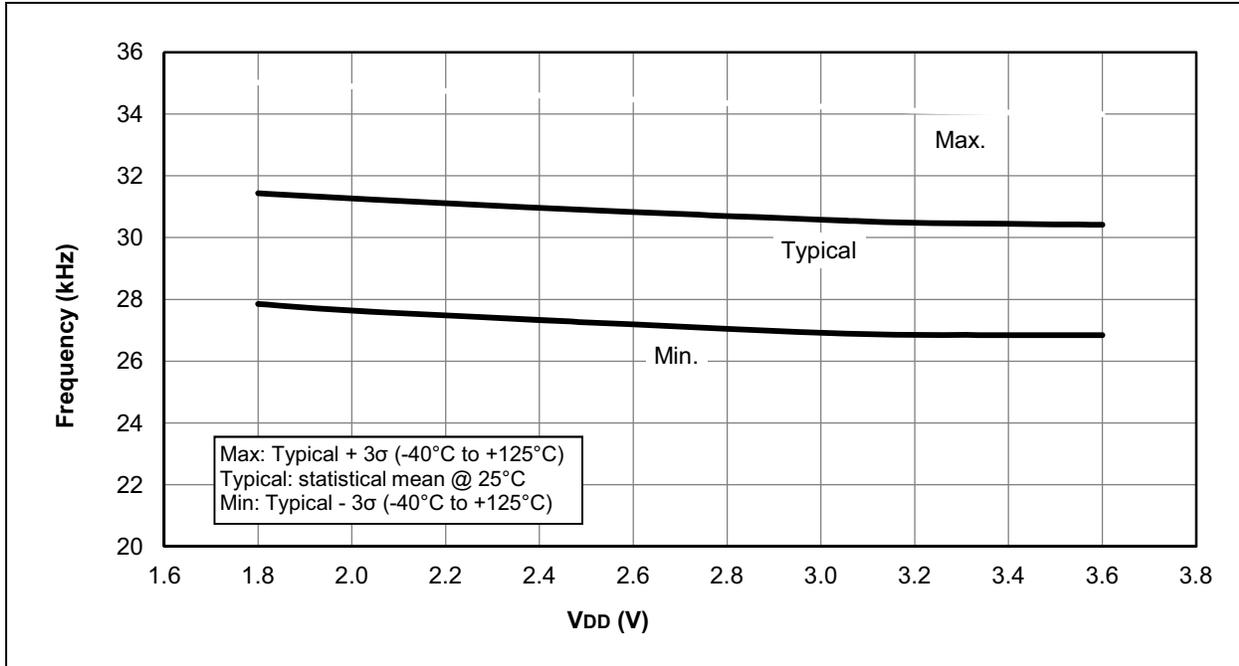
Refer to Figure 26-3 and Figure 26-4 for examples.

FIGURE 29-3: POR AND POR REARM WITH SLOW RISING V<sub>DD</sub>

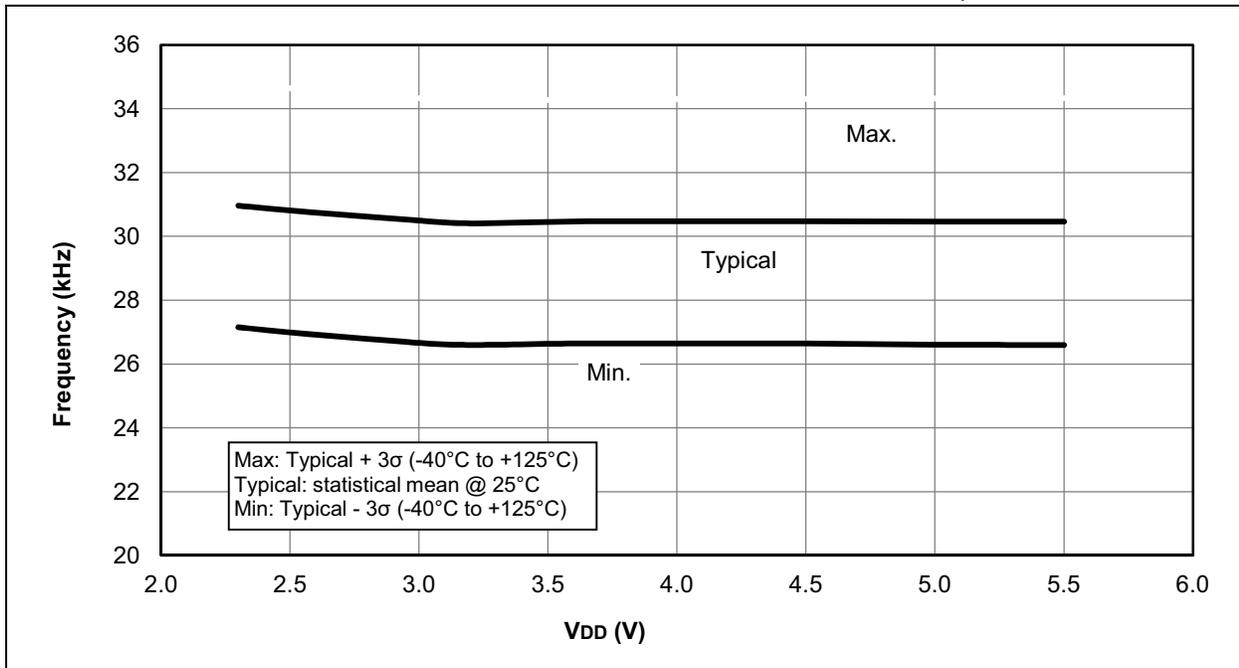


# PIC16(L)F1508/9

**FIGURE 30-70: LFINTOSC FREQUENCY OVER V<sub>DD</sub> AND TEMPERATURE, PIC16LF1508/9 ONLY**



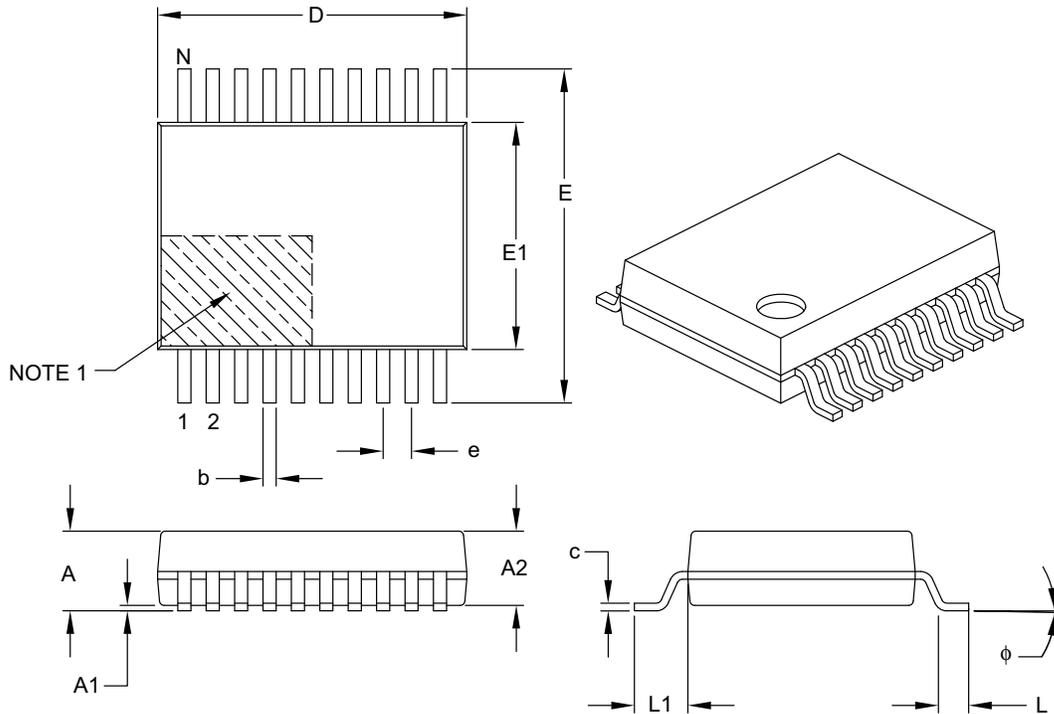
**FIGURE 30-71: LFINTOSC FREQUENCY OVER V<sub>DD</sub> AND TEMPERATURE, PIC16F1508/9 ONLY**



# PIC16(L)F1508/9

## 20-Lead Plastic Shrink Small Outline (SS) – 5.30 mm Body [SSOP]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at <http://www.microchip.com/packaging>



Dimension Limits	Units	MILLIMETERS		
		MIN	NOM	MAX
Number of Pins	N	20		
Pitch	e	0.65 BSC		
Overall Height	A	–	–	2.00
Molded Package Thickness	A2	1.65	1.75	1.85
Standoff	A1	0.05	–	–
Overall Width	E	7.40	7.80	8.20
Molded Package Width	E1	5.00	5.30	5.60
Overall Length	D	6.90	7.20	7.50
Foot Length	L	0.55	0.75	0.95
Footprint	L1	1.25 REF		
Lead Thickness	c	0.09	–	0.25
Foot Angle	$\phi$	0°	4°	8°
Lead Width	b	0.22	–	0.38

**Notes:**

- Pin 1 visual index feature may vary, but must be located within the hatched area.
- Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.20 mm per side.
- Dimensioning and tolerancing per ASME Y14.5M.

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-072B