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### What is "[Embedded - Microcontrollers](#)"?

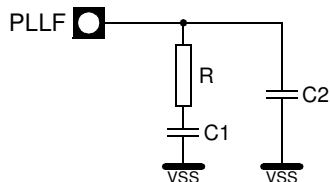
"[Embedded - Microcontrollers](#)" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

### Applications of "[Embedded - Microcontrollers](#)"

#### Details

Product Status	Active
Core Processor	80C51
Core Size	8-Bit
Speed	48MHz
Connectivity	I <sup>2</sup> C, SPI, UART/USART, USB
Peripherals	LED, POR, PWM, WDT
Number of I/O	18
Program Memory Size	16KB (16K x 8)
Program Memory Type	FLASH
EEPROM Size	4K x 8
RAM Size	1.25K x 8
Voltage - Supply (Vcc/Vdd)	2.7V ~ 5.5V
Data Converters	-
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	32-VQFN Exposed Pad
Supplier Device Package	32-QFN (7x7)
Purchase URL	<a href="https://www.e-xfl.com/product-detail/microchip-technology/at89c5130a-putum">https://www.e-xfl.com/product-detail/microchip-technology/at89c5130a-putum</a>

**Figure 5-4.** PLL Filter Connection

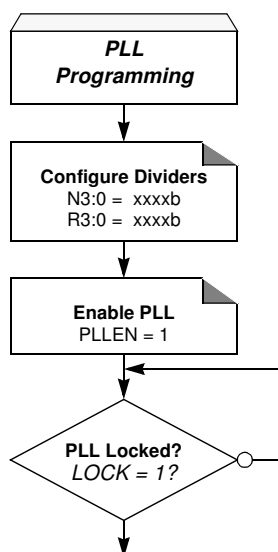


The typical values are:  $R = 100\ \Omega$ ,  $C1 = 10\ \text{nf}$ ,  $C2 = 2.2\ \text{nF}$ .

### 5.3.2 PLL Programming

The PLL is programmed using the flow shown in Figure 5-5. As soon as clock generation is enabled user must wait until the lock indicator is set to ensure the clock output is stable.

**Figure 5-5.** PLL Programming Flow



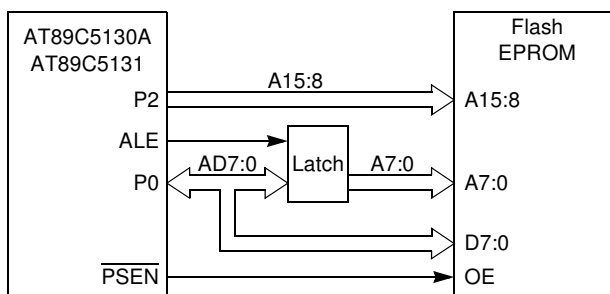
### 5.3.3 Divider Values

To generate a 48 MHz clock using the PLL, the divider values have to be configured following the oscillator frequency. The typical divider values are shown in Table 5-1.

**Table 5-1.** Typical Divider Values

Oscillator Frequency	R+1	N+1	PLLDIV
3 MHz	16	1	F0h
6 MHz	8	1	70h
8 MHz	6	1	50h
12 MHz	4	1	30h
16 MHz	3	1	20h
18 MHz	8	3	72h
20 MHz	12	5	B4h
24 MHz	2	1	10h

**Figure 8-2.** External Code Memory Interface Structure



**Table 8-1.** External Data Memory Interface Signals

Signal Name	Type	Description	Alternate Function
A15:8	O	<b>Address Lines</b> Upper address lines for the external bus.	P2.7:0
AD7:0	I/O	<b>Address/Data Lines</b> Multiplexed lower address lines and data for the external memory.	P0.7:0
ALE	O	<b>Address Latch Enable</b> ALE signals indicates that valid address information are available on lines AD7:0.	-
$\overline{\text{PSEN}}$	O	<b>Program Store Enable Output</b> This signal is active low during external code fetch or external code read (MOVC instruction).	-

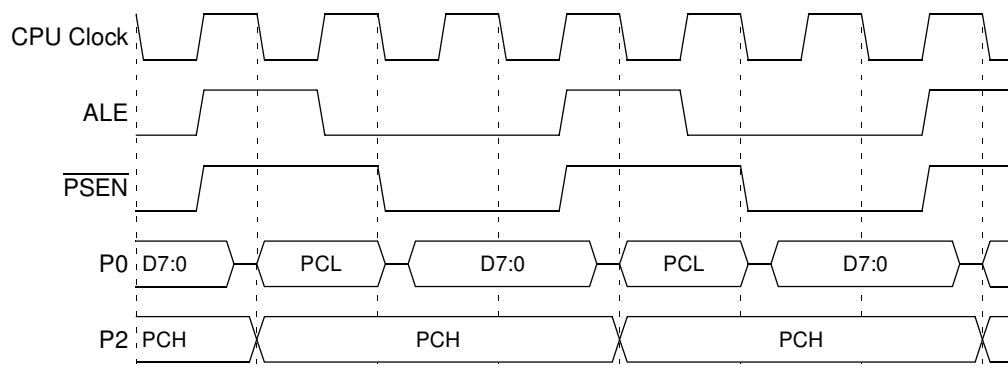
### 8.1.2 External Bus Cycles

This section describes the bus cycles the AT89C5130A/31A-M executes to fetch code (see Figure 8-3) in the external program/code memory.

External memory cycle takes 6 CPU clock periods. This is equivalent to 12 oscillator clock periods in standard mode or 6 oscillator clock periods in X2 mode. For further information on X2 mode (see the clock Section).

For simplicity, the accompanying figure depicts the bus cycle waveforms in idealized form and do not provide precise timing information.

**Figure 8-3.** External Code Fetch Waveforms



## 11. In-System Programming (ISP)

With the implementation of the User Space (FM0) and the Boot Space (FM1) in Flash technology the AT89C5130A/31A-M allows the system engineer the development of applications with a very high level of flexibility. This flexibility is based on the possibility to alter the customer program at any stages of a product's life:

- Before mounting the chip on the PCB, FM0 flash can be programmed with the application code. FM1 is always preprogrammed by Atmel with a USB bootloader.<sup>(1)</sup>
- Once the chip is mounted on the PCB, it can be programmed by serial mode via the USB bus.

Note: 1. The user can also program his own bootloader in FM1.

This ISP allows code modification over the total lifetime of the product.

Besides the default Bootloaders Atmel provide customers all the needed Application-Programming-Interfaces (API) which are needed for the ISP. The API are located in the Boot memory.

This allow the customer to have a full use of the 32-Kbyte user memory.

### 11.1 Flash Programming and Erasure

There are three methods for programming the Flash memory:

- The Atmel bootloader located in FM1 is activated by the application. Low level API routines (located in FM1) will be used to program FM0. The interface used for serial downloading to FM0 is the USB. API can be called also by user's bootloader located in FM0 at [SBV]00h.
- A further method exist in activating the Atmel boot loader by hardware activation. See the Section "Hardware Registers".
- The FM0 can be programmed also by the parallel mode using a programmer.

Bit Number	Bit Mnemonic	Description															
5	M0	Pulse length Cleared to stretch MOVX control: the $\overline{RD}$ and the $\overline{WR}$ pulse length is 6 clock periods (default). Set to stretch MOVX control: the $\overline{RD}$ and the $\overline{WR}$ pulse length is 30 clock periods.															
4	-	Reserved The value read from this bit is indeterminate. Do not set this bit															
3	XRS1	ERAM Size															
2	XRS0	<table border="1"> <thead> <tr> <th>XRS1</th> <th>XRS0</th> <th>ERAM size</th> </tr> </thead> <tbody> <tr> <td>0</td> <td>0</td> <td>256 bytes</td> </tr> <tr> <td>0</td> <td>1</td> <td>512 bytes</td> </tr> <tr> <td>1</td> <td>0</td> <td>768 bytes</td> </tr> <tr> <td>1</td> <td>1</td> <td>1024 bytes (default)</td> </tr> </tbody> </table>	XRS1	XRS0	ERAM size	0	0	256 bytes	0	1	512 bytes	1	0	768 bytes	1	1	1024 bytes (default)
XRS1	XRS0	ERAM size															
0	0	256 bytes															
0	1	512 bytes															
1	0	768 bytes															
1	1	1024 bytes (default)															
1	EXTRAM	<b>EXTRAM bit</b> Cleared to access internal ERAM using MOVX at $\overline{Ri}$ at DPTR. Set to access external memory.															
0	AO	<b>ALE Output bit</b> Cleared, ALE is emitted at a constant rate of 1/6 the oscillator frequency (or 1/3 if X2 mode is used) (default). Set, ALE is active only when a MOVX or MOVC instruction is used.															

Reset Value = 0X0X 1100b

Not bit addressable

**Table 13-1.** T2CON Register  
T2CON - Timer 2 Control Register (C8h)

7	6	5	4	3	2	1	0
TF2	EXF2	RCLK	TCLK	EXEN2	TR2	C/T2#	CP/RL2#
Bit Number	Bit Mnemonic	Description					
7	TF2	<b>Timer 2 overflow Flag</b> Must be cleared by software. Set by hardware on Timer 2 overflow, if RCLK = 0 and TCLK = 0.					
6	EXF2	<b>Timer 2 External Flag</b> Set when a capture or a reload is caused by a negative transition on T2EX pin if EXEN2 = 1. When set, causes the CPU to vector to Timer 2 interrupt routine when Timer 2 interrupt is enabled. Must be cleared by software. EXF2 doesn't cause an interrupt in Up/down counter mode (DCEN = 1).					
5	RCLK	<b>Receive Clock bit</b> Cleared to use Timer 1 overflow as receive clock for serial port in mode 1 or 3. Set to use Timer 2 overflow as receive clock for serial port in mode 1 or 3.					
4	TCLK	<b>Transmit Clock bit</b> Cleared to use Timer 1 overflow as transmit clock for serial port in mode 1 or 3. Set to use Timer 2 overflow as transmit clock for serial port in mode 1 or 3.					
3	EXEN2	<b>Timer 2 External Enable bit</b> Cleared to ignore events on T2EX pin for Timer 2 operation. Set to cause a capture or reload when a negative transition on T2EX pin is detected, if Timer 2 is not used to clock the serial port.					
2	TR2	<b>Timer 2 Run control bit</b> Cleared to turn off Timer 2. Set to turn on Timer 2.					
1	C/T2#	<b>Timer/Counter 2 select bit</b> Cleared for timer operation (input from internal clock system: $F_{CLK\ PERIPH}$ ). Set for counter operation (input from T2 input pin, falling edge trigger). Must be 0 for clock out mode.					
0	CP/RL2#	<b>Timer 2 Capture/Reload bit</b> If RCLK = 1 or TCLK = 1, CP/RL2# is ignored and timer is forced to Auto-reload on Timer 2 overflow. Cleared to Auto-reload on Timer 2 overflows or negative transitions on T2EX pin if EXEN2 = 1. Set to capture on negative transitions on T2EX pin if EXEN2 = 1.					

Reset Value = 0000 0000b

Bit addressable

**Table 13-2.** T2MOD Register  
T2MOD - Timer 2 Mode Control Register (C9h)

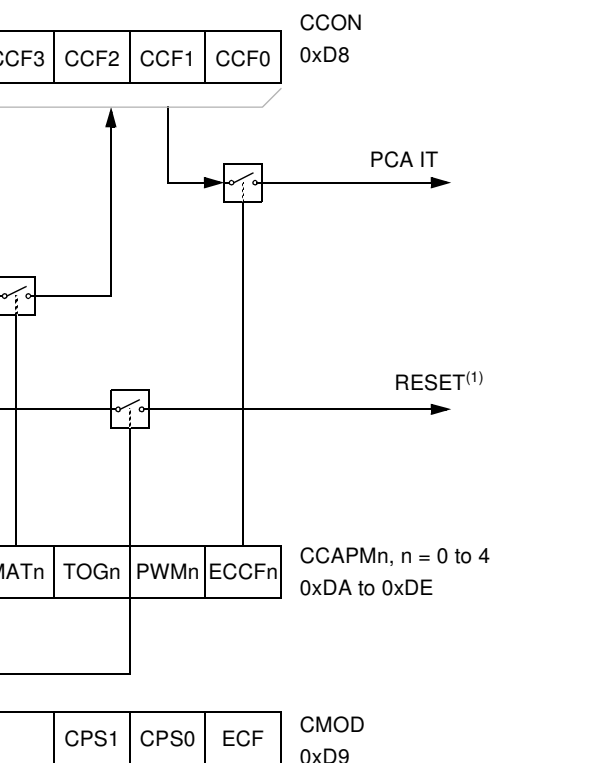
7	6	5	4	3	2	1	0
-	-	-	-	-	-	T2OE	DCEN

Bit Number	Bit Mnemonic	Description
7	-	<b>Reserved</b> The value read from this bit is indeterminate. Do not set this bit.
6	-	<b>Reserved</b> The value read from this bit is indeterminate. Do not set this bit.
5	-	<b>Reserved</b> The value read from this bit is indeterminate. Do not set this bit.
4	-	<b>Reserved</b> The value read from this bit is indeterminate. Do not set this bit.
3	-	<b>Reserved</b> The value read from this bit is indeterminate. Do not set this bit.
2	-	<b>Reserved</b> The value read from this bit is indeterminate. Do not set this bit.
1	T2OE	<b>Timer 2 Output Enable bit</b> Cleared to program P1.0/T2 as clock input or I/O port. Set to program P1.0/T2 as clock output.
0	DCEN	<b>Down Counter Enable bit</b> Cleared to disable Timer 2 as up/down counter. Set to enable Timer 2 as up/down counter.

Reset Value = XXXX XX00b

Not bit addressable



APnH should be set with a non zero value, otherwise CCAPnH will set the ECOM bit.

ECOM so that an unwanted match doesn't occur to CCAPnH will set ECOM. For this reason, user n CCAPnH. Of course, the ECOM bit can still be

associated with the PCA module will toggle each time a read occurs to the module's capture registers. To activate this feature, the module's CCAPMn SFR must be set (see

CAPnH before writing the ECOMn bit.



## 15. Serial I/O Port

The serial I/O port in the AT89C5130A/31A-M is compatible with the serial I/O port in the 80C52. It provides both synchronous and asynchronous communication modes. It operates as an Universal Asynchronous Receiver and Transmitter (UART) in three full-duplex modes (modes 1, 2 and 3). Asynchronous transmission and reception can occur simultaneously and at different baud rates.

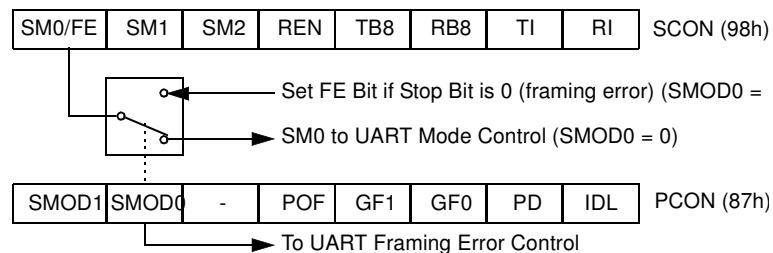
Serial I/O port includes the following enhancements:

- Framing error detection
- Automatic address recognition

### 15.1 Framing Error Detection

Framing bit error detection is provided for the three asynchronous modes (modes 1, 2 and 3). To enable the framing bit error detection feature, set SMOD0 bit in PCON register (see Figure 15-1).

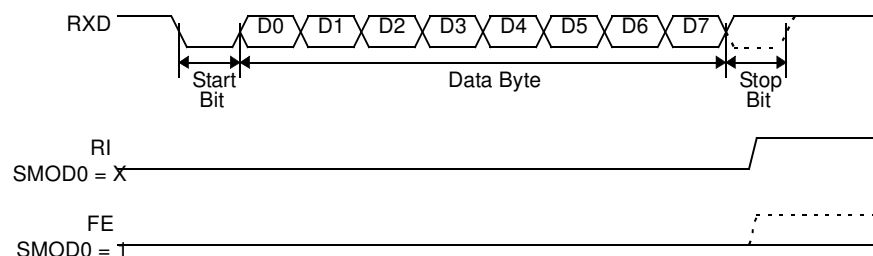
**Figure 15-1.** Framing Error Block Diagram



When this feature is enabled, the receiver checks each incoming data frame for a valid stop bit. An invalid stop bit may result from noise on the serial lines or from simultaneous transmission by two CPUs. If a valid stop bit is not found, the Framing Error bit (FE) in SCON register (See [Table 15-1](#)) bit is set.

Software may examine FE bit after each reception to check for data errors. Once set, only software or a reset can clear FE bit. Subsequently received frames with valid stop bits cannot clear FE bit. When FE feature is enabled, RI rises on stop bit instead of the last data bit (See [Figure 15-2](#) and [Figure 15-3](#)).

**Figure 15-2.** UART Timings in Mode 1

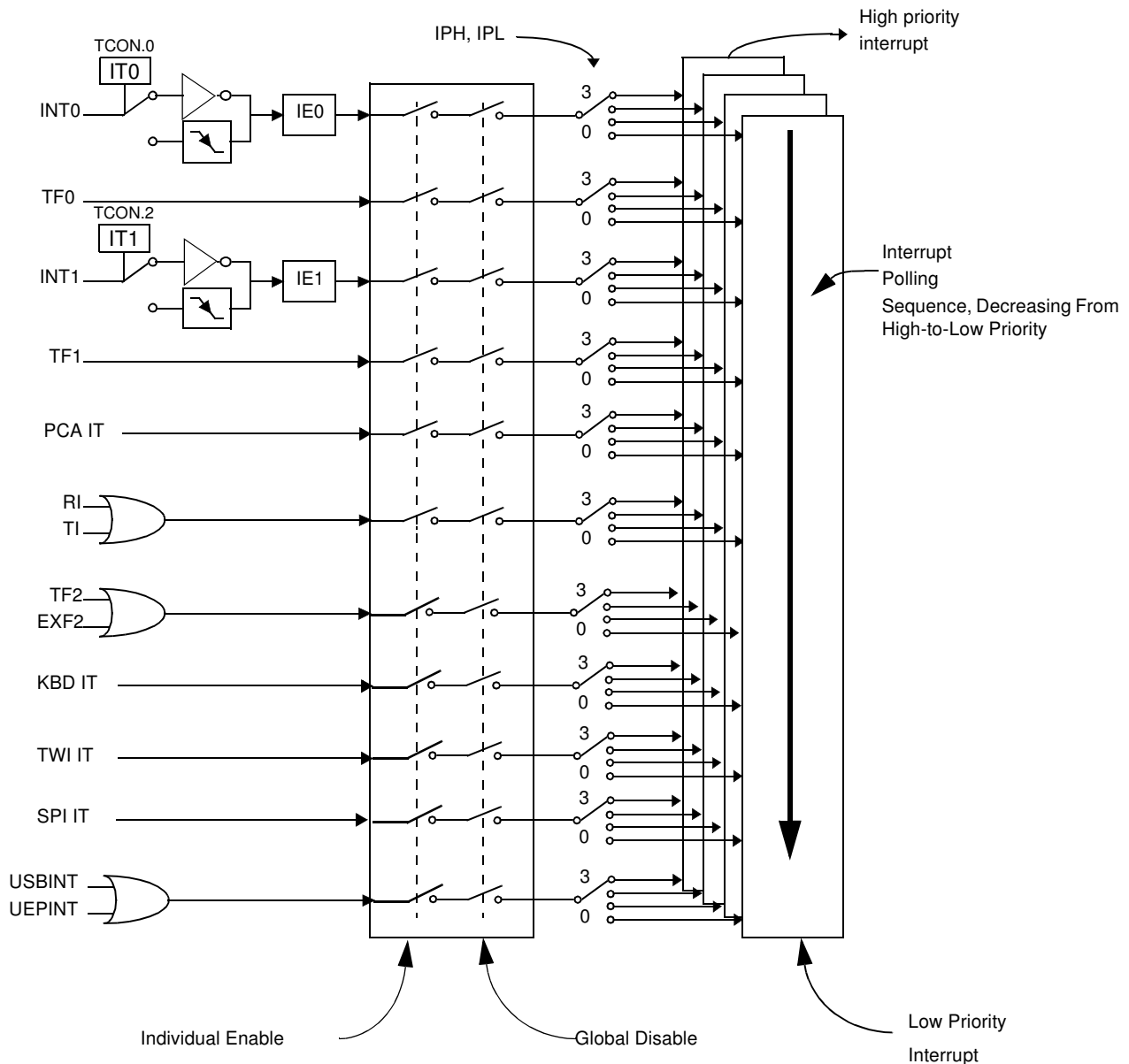


## 16. Interrupt System

### 16.1 Overview

The AT89C5130A/31A-M has a total of 11 interrupt vectors: two external interrupts ( $\overline{\text{INT0}}$  and  $\overline{\text{INT1}}$ ), three timer interrupts (timers 0, 1 and 2), the serial port interrupt, SPI interrupt, Keyboard interrupt, USB interrupt and the PCA global interrupt. These interrupts are shown in Figure 16-1.

**Figure 16-1.** Interrupt Control System



Each of the interrupt sources can be individually enabled or disabled by setting or clearing a bit in the Interrupt Enable register (Table 16-2). This register also contains a global disable bit, which must be cleared to disable all interrupts at once.

Bit Number	Bit Mnemonic	Description
7	EA	<b>Enable All interrupt bit</b> Cleared to disable all interrupts. Set to enable all interrupts.
6	EC	<b>PCA interrupt enable bit</b> Cleared to disable. Set to enable.
5	ET2	<b>Timer 2 overflow interrupt Enable bit</b> Cleared to disable Timer 2 overflow interrupt. Set to enable Timer 2 overflow interrupt.
4	ES	<b>Serial port Enable bit</b> Cleared to disable serial port interrupt. Set to enable serial port interrupt.
3	ET1	<b>Timer 1 overflow interrupt Enable bit</b> Cleared to disable Timer 1 overflow interrupt. Set to enable Timer 1 overflow interrupt.
2	EX1	<b>External interrupt 1 Enable bit</b> Cleared to disable external interrupt 1. Set to enable external interrupt 1.
1	ET0	<b>Timer 0 overflow interrupt Enable bit</b> Cleared to disable timer 0 overflow interrupt. Set to enable timer 0 overflow interrupt.
0	EX0	<b>External interrupt 0 Enable bit</b> Cleared to disable external interrupt 0. Set to enable external interrupt 0.

Reset Value = 0000 0000b

Bit addressable

**Table 16-3.** IPL0 Register  
IPL0 - Interrupt Priority Register (B8h)

7	6	5	4	3	2	1	0
-	PPCL	PT2L	PSL	PT1L	PX1L	PT0L	PX0L

### 19.2.3 SPI Serial Clock (SCK)

This signal is used to synchronize the data movement both in and out the devices through their MOSI and MISO lines. It is driven by the Master for eight clock cycles which allows to exchange one byte on the serial lines.

### 19.2.4 Slave Select ( $\overline{SS}$ )

Each Slave peripheral is selected by one Slave Select pin ( $\overline{SS}$ ). This signal must stay low for any message for a Slave. It is obvious that only one Master ( $\overline{SS}$  high level) can drive the network. The Master may select each Slave device by software through port pins (Figure 19-1). To prevent bus conflicts on the MISO line, only one slave should be selected at a time by the Master for a transmission.

In a Master configuration, the  $\overline{SS}$  line can be used in conjunction with the MODF flag in the SPI Status register (SPSTA) to prevent multiple masters from driving MOSI and SCK (see Section "Error Conditions", page 98).

A high level on the  $\overline{SS}$  pin puts the MISO line of a Slave SPI in a high-impedance state.

The  $\overline{SS}$  pin could be used as a general-purpose if the following conditions are met:

- The device is configured as a Master and the SSDIS control bit in SPCON is set. This kind of configuration can be found when only one Master is driving the network and there is no way that the  $\overline{SS}$  pin could be pulled low. Therefore, the MODF flag in the SPSTA will never be set<sup>(1)</sup>.
- The Device is configured as a Slave with CPHA and SSDIS control bits set<sup>(2)</sup> This kind of configuration can happen when the system comprises one Master and one Slave only. Therefore, the device should always be selected and there is no reason that the Master uses the  $\overline{SS}$  pin to select the communicating Slave device.

Notes: 1. Clearing SSDIS control bit does not clear MODF.

2. Special care should be taken not to set SSDIS control bit when CPHA = '0' because in this mode, the  $\overline{SS}$  is used to start the transmission.

### 19.2.5 Baud Rate

In Master mode, the baud rate can be selected from a baud rate generator which is controlled by three bits in the SPCON register: SPR2, SPR1 and SPR0. The Master clock is chosen from one of seven clock rates resulting from the division of the internal clock by 4, 8, 16, 32, 64 or 128.

Table 19-1 gives the different clock rates selected by SPR2:SPR1:SPR0:

**Table 19-1.** SPI Master Baud Rate Selection

SPR2	SPR1	SPR0	Clock Rate	Baud Rate Divisor (BD)
0	0	0	Don't Use	No BRG
0	0	1	$F_{CLK\ PERIPH}/4$	4
0	1	0	$F_{CLK\ PERIPH}/8$	8
0	1	1	$F_{CLK\ PERIPH}/16$	16
1	0	0	$F_{CLK\ PERIPH}/32$	32
1	0	1	$F_{CLK\ PERIPH}/64$	64
1	1	0	$F_{CLK\ PERIPH}/128$	128

## 20. Two Wire Interface (TWI)

This section describes the 2-wire interface. The 2-wire bus is a bi-directional 2-wire serial communication standard. It is designed primarily for simple but efficient integrated circuit (IC) control. The system is comprised of two lines, SCL (Serial Clock) and SDA (Serial Data) that carry information between the ICs connected to them. The serial data transfer is limited to 400 Kbit/s in standard mode. Various communication configuration can be designed using this bus. [Figure 20-1](#) shows a typical 2-wire bus configuration. All the devices connected to the bus can be master and slave.

**Figure 20-1.** 2-wire Bus Configuration

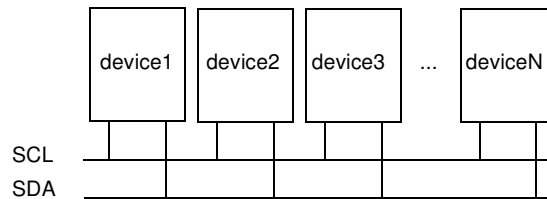
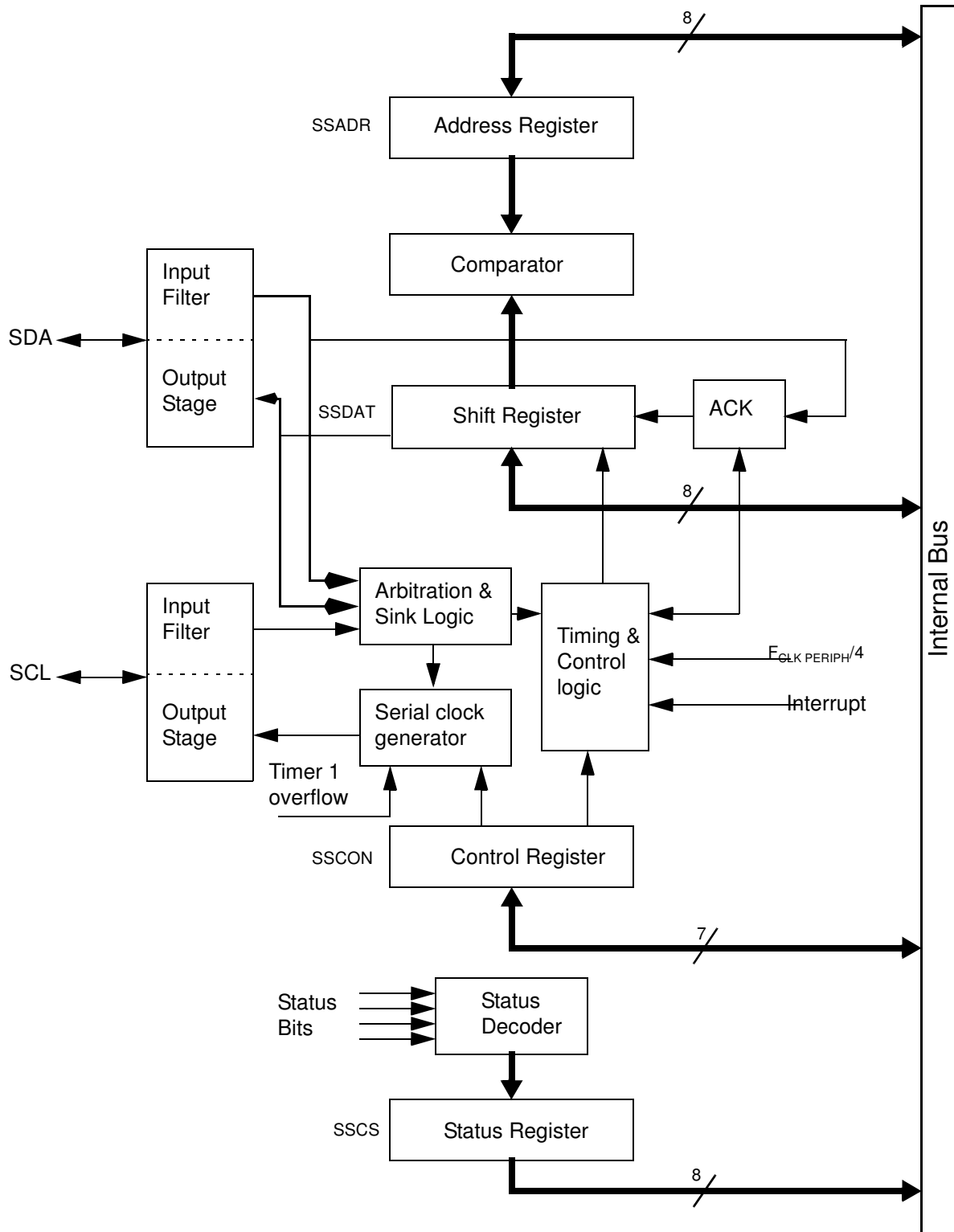


Figure 20-2. Block Diagram



The Endpoint 0 is the Default Control Endpoint and will always be configured in Control type.

- Endpoint direction configuration

For Bulk, Interrupt and Isochronous endpoints, the direction is defined with the EPDIR bit of the UEPCONX register with the following values:

- IN:EPDIR = 1b
- OUT:EPDIR = 0b

For Control endpoints, the EPDIR bit has no effect.

- Summary of Endpoint Configuration:

Do not forget to select the correct endpoint number in the UEPNUM register before accessing to endpoint specific registers.

**Table 21-1.** Summary of Endpoint Configuration

Endpoint Configuration	EPEN	EPDIR	EPTYPE	UEPCONX
Disabled	0b	Xb	XXb	0XXX XXXb
Control	1b	Xb	00b	80h
Bulk-in	1b	1b	10b	86h
Bulk-out	1b	0b	10b	82h
Interrupt-In	1b	1b	11b	87h
Interrupt-Out	1b	0b	11b	83h
Isochronous-In	1b	1b	01b	85h
Isochronous-Out	1b	0b	01b	81h

- Endpoint FIFO reset

Before using an endpoint, its FIFO will be reset. This action resets the FIFO pointer to its original value, resets the byte counter of the endpoint (UBYCTLX and UBYCTHX registers), and resets the data toggle bit (DTGL bit in UEPCONX).

The reset of an endpoint FIFO is performed by setting to 1 and resetting to 0 the corresponding bit in the UEPRST register.

For example, in order to reset the Endpoint number 2 FIFO, write 0000 0100b then 0000 0000b in the UEPRST register.

Note that the endpoint reset doesn't reset the bank number for ping-pong endpoints.

## 21.3 Read/Write Data FIFO

### 21.3.1 FIFO Mapping

Depending on the selected endpoint through the UEPNUM register, the UEPDATX register allows to access the corresponding endpoint data fifo.

The firmware has to clear one of these two bits after having read all the data FIFO to allow a new packet to be stored in the corresponding bank.

If the Host sends more bytes than supported by the endpoint FIFO, the overflow data won't be stored, but the USB controller will consider that the packet is valid if the CRC is correct.

### 21.6.3 Isochronous IN Transactions in Standard Mode

An endpoint will be first enabled and configured before being able to send Isochronous packets.

The firmware will fill the FIFO with the data to be sent and set the TXRDY bit in the UEPSTAX register to allow the USB controller to send the data stored in FIFO at the next IN request concerning this endpoint.

If the TXRDY bit is not set when the IN request occurs, nothing will be sent by the USB controller.

When the IN packet has been sent, the TXCMPL bit in the UEPSTAX register is set by the USB controller. This triggers a USB interrupt if enabled. The firmware will clear the TXCMPL bit before filling the endpoint FIFO with new data.

The firmware will never write more bytes than supported by the endpoint FIFO

### 21.6.4 Isochronous IN Transactions in Ping-pong Mode

An endpoint will be first enabled and configured before being able to send Isochronous packets.

The firmware will fill the FIFO bank 0 with the data to be sent and set the TXRDY bit in the UEPSTAX register to allow the USB controller to send the data stored in FIFO at the next IN request concerning the endpoint. The FIFO banks are automatically switched, and the firmware can immediately write into the endpoint FIFO bank 1.

If the TXRDY bit is not set when the IN request occurs, nothing will be sent by the USB controller.

When the IN packet concerning the bank 0 has been sent, the TXCMPL bit is set by the USB controller. This triggers a USB interrupt if enabled. The firmware will clear the TXCMPL bit before filling the endpoint FIFO bank 0 with new data. The FIFO banks are then automatically switched.

When the IN packet concerning the bank 1 has been sent, the TXCMPL bit is set by the USB controller. This triggers a USB interrupt if enabled. The firmware will clear the TXCMPL bit before filling the endpoint FIFO bank 1 with new data.

The bank switch is performed by the USB controller each time the TXRDY bit is set by the firmware. Until the TXRDY bit has been set by the firmware for an endpoint bank, the USB controller won't send anything at each IN requests concerning this bank.

The firmware will never write more bytes than supported by the endpoint FIFO.

## 21.7 Miscellaneous

### 21.7.1 USB Reset

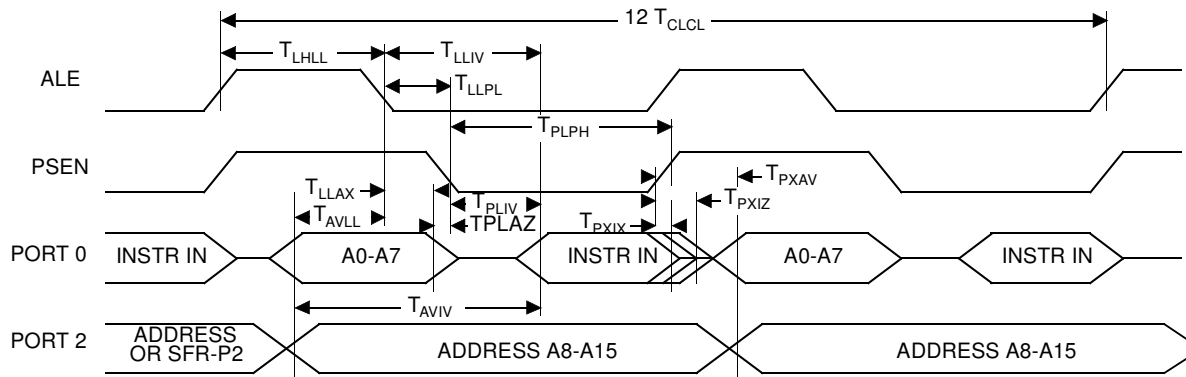
The EORINT bit in the USBINT register is set by hardware when a End Of Reset has been detected on the USB bus. This triggers a USB interrupt if enabled. The USB controller is still enabled, but all the USB registers are reset by hardware. The firmware will clear the EORINT bit to allow the next USB reset detection.



**Table 27-4.** AC Parameters for a Variable Clock

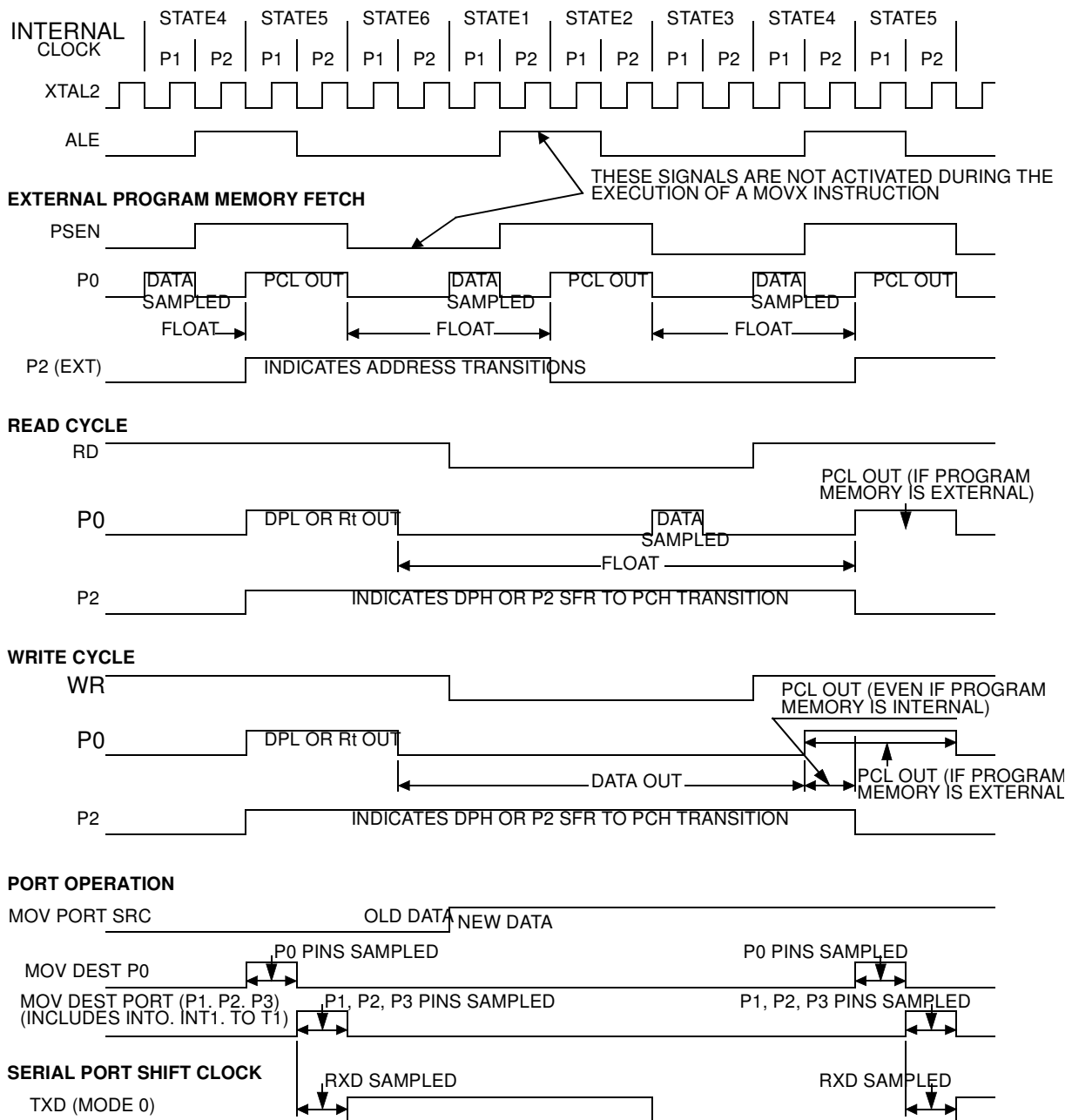
Symbol	Type	Standard Clock	X2 Clock	X Parameter	Units
$T_{LHLL}$	Min	$2 T - x$	$T - x$	10	ns
$T_{AVLL}$	Min	$T - x$	$0.5 T - x$	15	ns
$T_{LLAX}$	Min	$T - x$	$0.5 T - x$	15	ns
$T_{LLIV}$	Max	$4 T - x$	$2 T - x$	30	ns
$T_{LLPL}$	Min	$T - x$	$0.5 T - x$	10	ns
$T_{PLPH}$	Min	$3 T - x$	$1.5 T - x$	20	ns
$T_{PLIV}$	Max	$3 T - x$	$1.5 T - x$	40	ns
$T_{PXIX}$	Min	x	x	0	ns
$T_{PXIZ}$	Max	$T - x$	$0.5 T - x$	7	ns
$T_{AVIV}$	Max	$5 T - x$	$2.5 T - x$	40	ns
$T_{PLAZ}$	Max	x	x	10	ns

### 27.4.3 External Program Memory Read Cycle



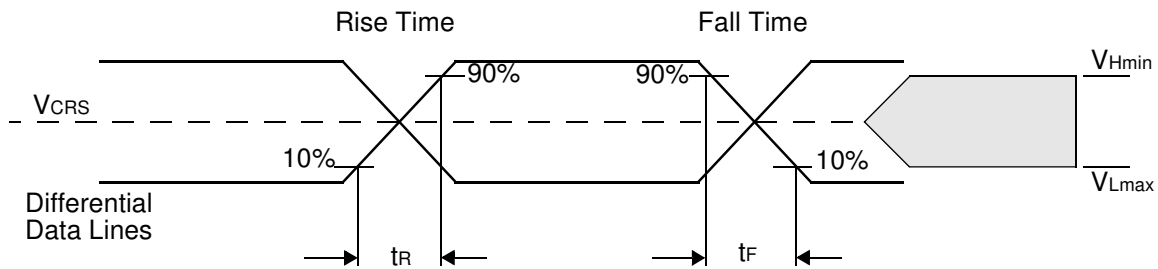
## 27.4.13 Clock Waveforms

Valid in normal clock mode. In X2 mode XTAL2 must be changed to XTAL2/2.



This diagram indicates when signals are clocked internally. The time it takes the signals to propagate to the pins, however, ranges from 25 to 125 ns. This propagation delay is dependent on variables such as temperature and pin loading. Propagation also varies from output to output and component. Typically though ( $T_A = 25^\circ\text{C}$  fully loaded)  $\overline{\text{RD}}$  and  $\overline{\text{WR}}$  propagation delays are approximately 50 ns. The other signals are typically 85 ns. Propagation delays are incorporated in the AC specifications.

## 27.5 USB AC Parameters



**Table 27-14.** USB AC Parameters

Symbol	Parameter	Min	Typ	Max	Unit	Test Conditions
$t_R$	Rise Time	4		20	ns	
$t_F$	Fall Time	4		20	ns	
$t_{FDRATE}$	Full-speed Data Rate	11.9700		12.0300	Mb/s	
$V_{CRS}$	Crossover Voltage	1.3		2.0	V	
$t_{DJ1}$	Source Jitter Total to Next Transaction	-3.5		3.5	ns	
$t_{DJ2}$	Source Jitter Total for Paired Transactions	-4		4	ns	
$t_{JR1}$	Receiver Jitter to Next Transaction	-18.5		18.5	ns	
$t_{JR2}$	Receiver Jitter for Paired Transactions	-9		9	ns	

## 27.6 SPI Interface AC Parameters

### 27.6.0.1 Definition of Symbols

**Table 27-15.** SPI Interface Timing Symbol Definitions

Signals	
C	Clock
I	Data In
O	Data Out

Conditions	
H	High
L	Low
V	Valid
X	No Longer Valid
Z	Floating

### 27.6.0.2 Timings

Test conditions: capacitive load on all pins= 50 pF.

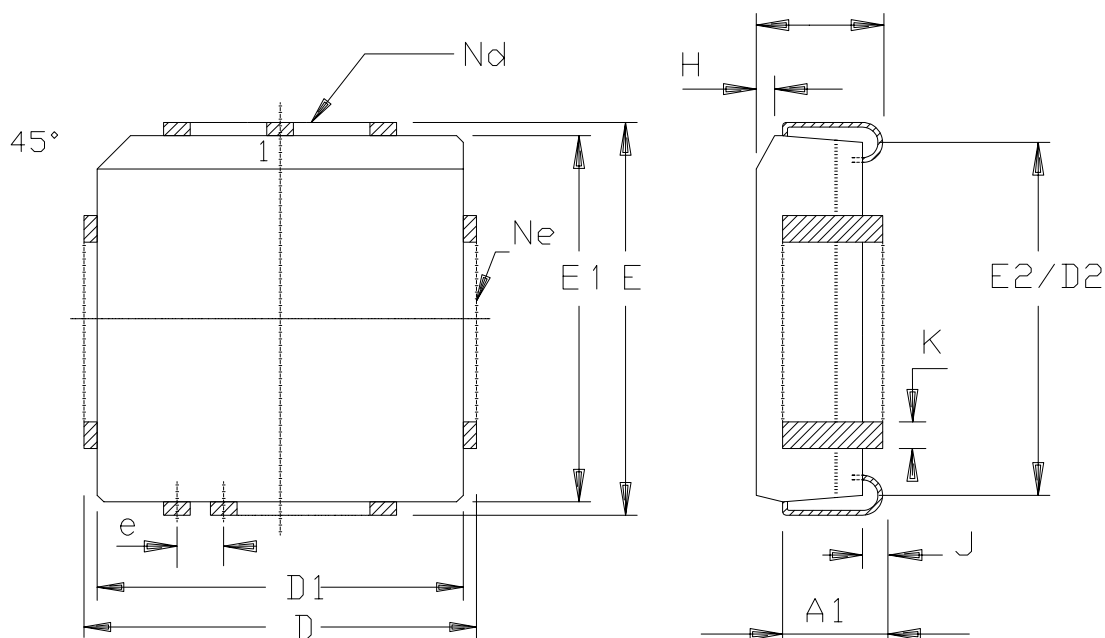
**Table 27-16.** SPI Interface Master AC Timing

$V_{DD} = 2.7 \text{ to } 5.5 \text{ V}$ ,  $T_A = -40 \text{ to } +85^\circ\text{C}$

Symbol	Parameter	Min	Max	Unit
<b>Slave Mode</b>				
$T_{CHCH}$	Clock Period	2		$T_{PER}$
$T_{CHCX}$	Clock High Time	0.8		$T_{PER}$
$T_{CLCX}$	Clock Low Time	0.8		$T_{PER}$
$T_{SLCH}, T_{SLCL}$	$\overline{SS}$ Low to Clock edge	100		ns
$T_{IVCL}, T_{IVCH}$	Input Data Valid to Clock Edge	50		ns
$T_{CLIX}, T_{CHIX}$	Input Data Hold after Clock Edge	50		ns
$T_{CLOV}, T_{CHOV}$	Output Data Valid after Clock Edge		50	ns
$T_{CLOX}, T_{CHOX}$	Output Data Hold Time after Clock Edge	0		ns
$T_{CLSH}, T_{CHSH}$	$\overline{SS}$ High after Clock Edge	0		ns
$T_{SLOV}$	$\overline{SS}$ Low to Output Data Valid		$4T_{PER}+20$	ns
$T_{SHOX}$	Output Data Hold after $\overline{SS}$ High		$2T_{PER}+100$	ns
$T_{SHSL}$	$\overline{SS}$ High to $\overline{SS}$ Low	$2T_{PER}+120$		
$T_{ILIH}$	Input Rise Time		2	$\mu\text{s}$
$T_{IHIL}$	Input Fall Time		2	$\mu\text{s}$
$T_{OLOH}$	Output Rise time		100	ns
$T_{OHOL}$	Output Fall Time		100	ns
<b>Master Mode</b>				
$T_{CHCH}$	Clock Period	4		$T_{PER}$
$T_{CHCX}$	Clock High Time	$2T_{PER}-20$		ns
$T_{CLCX}$	Clock Low Time	$2T_{PER}-20$		ns
$T_{IVCL}, T_{IVCH}$	Input Data Valid to Clock Edge	50		ns
$T_{CLIX}, T_{CHIX}$	Input Data Hold after Clock Edge	50		ns
$T_{CLOV}, T_{CHOV}$	Output Data Valid after Clock Edge		20	ns
$T_{CLOX}, T_{CHOX}$	Output Data Hold Time after Clock Edge	0		ns

Note:  $T_{PER}$  is XTAL period when SPI interface operates in X2 mode or twice XTAL period when SPI interface operates in X1 mode.

## 29.2 52-lead PLCC



	MM		INCH	
A	4. 20	4. 57	. 165	. 180
A1	2. 29	3. 30	. 090	. 130
D	19. 94	20. 19	. 785	. 795
D1	19. 05	19. 25	. 750	. 758
D2	17. 53	18. 54	. 690	. 730
E	19. 94	20. 19	. 785	. 795
E1	19. 05	19. 25	. 750	. 758
E2	17. 53	18. 54	. 690	. 730
e	1. 27	BSC	. 050	BSC
H	1. 07	1. 42	. 042	. 056
J	0. 51	—	. 020	—
K	0. 33	0. 53	. 013	. 021
Nd	13		13	
Ne	13		13	
PKG STD		00		

### STANDARD NOTES FOR PLCC

1/ CONTROLLING DIMENSIONS : INCHES

2/ DIMENSIONING AND TOLERANCING PER ANSI Y 14.5M - 1982.

3/ "D" AND "E1" DIMENSIONS DO NOT INCLUDE MOLD FLASH OR PROTUSIONS.

MOLD FLASH OR PROTUSIONS SHALL NOT EXCEED 0.20 mm (.008 INCH) PER SIDE.