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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

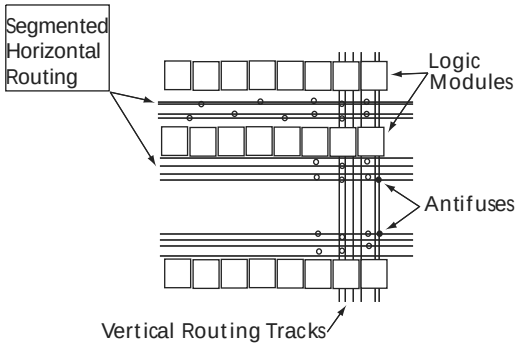
Details	
Product Status	Active
Number of LABs/CLBs	-
Number of Logic Elements/Cells	-
Total RAM Bits	-
Number of I/O	57
Number of Gates	3000
Voltage - Supply	3V ~ 3.6V, 4.75V ~ 5.25V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 70°C (TA)
Package / Case	100-BQFP
Supplier Device Package	100-PQFP (20x14)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/a40mx02-fpqg100

Contents

3.2.3.3 Antifuse Structures

An antifuse is a “normally open” structure. The use of antifuses to implement a programmable logic device results in highly testable structures as well as efficient programming algorithms. There are no pre-existing connections; temporary connections can be made using pass transistors. These temporary connections can isolate individual antifuses to be programmed and individual circuit structures to be tested, which can be done before and after programming. For instance, all metal tracks can be tested for continuity and shorts between adjacent tracks, and the functionality of all logic modules can be verified.

Figure 7 • **MX Routing Structure**



3.2.4 Clock Networks

The 40MX devices have one global clock distribution network (CLK). A signal can be put on the CLK network by being routed through the CLKBUF buffer.

In 42MX devices, there are two low-skew, high-fanout clock distribution networks, referred to as CLKA and CLKB. Each network has a clock module (CLKMOD) that can select the source of the clock signal from any of the following (Figure 8, page 11):

- Externally from the CLKA pad, using CLKBUF buffer
- Externally from the CLKB pad, using CLKBUF buffer
- Internally from the CLKINTA input, using CLKINT buffer
- Internally from the CLKINTB input, using CLKINT buffer

The clock modules are located in the top row of I/O modules. Clock drivers and a dedicated horizontal clock track are located in each horizontal routing channel.

Clock input pads in both 40MX and 42MX devices can also be used as normal I/Os, bypassing the clock networks.

The A42MX36 device has four additional register control resources, called quadrant clock networks (Figure 9, page 11). Each quadrant clock provides a local, high-fanout resource to the contiguous logic modules within its quadrant of the device. Quadrant clock signals can originate from specific I/O pins or from the internal array and can be used as a secondary register clock, register clear, or output enable.

Figure 13 • **Silicon Explorer II Setup with 42MX**

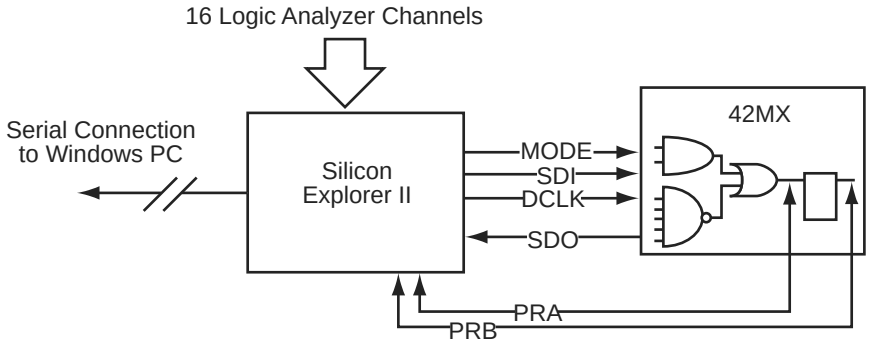


Table 8 • **Device Configuration Options for Probe Capability**

Security Fuse(s) Programmed	Mode	PRA, PRB[1]	SDI, SDO, DCLK[1]
No	LOW	User I/Os[2]	User I/Os[2]
No	HIGH	Probe Circuit Outputs	Probe Circuit Inputs
Yes	–	Probe Circuit Secured	Probe Circuit Secured

1. *Avoid using SDI, SDO, DCLK, PRA and PRB pins as input or bidirectional ports. Since these pins are active during probing, input signals will not pass through these pins and may cause contention.*
2. *If no user signal is assigned to these pins, they will behave as unused I/Os in this mode. See the Pin Descriptions,* page 83 *for information on unused I/O pins*

3.4.7 Design Consideration

It is recommended to use a series 70Ω termination resistor on every probe connector (SDI, SDO, MODE, DCLK, PRA and PRB). The 70 Ω series termination is used to prevent data transmission corruption during probing and reading back the checksum.

3.4.8 IEEE Standard 1149.1 Boundary Scan Test (BST) Circuitry

42MX24 and 42MX36 devices are compatible with IEEE Standard 1149.1 (informally known as Joint Testing Action Group Standard or JTAG), which defines a set of hardware architecture and mechanisms for cost-effective board-level testing. The basic MX boundary-scan logic circuit is composed of the TAP (test access port), TAP controller, test data registers and instruction register (Figure 14, page 18). This circuit supports all mandatory IEEE 1149.1 instructions (EXTEST, SAMPLE/PRELOAD and BYPASS) and some optional instructions. Table 9, page 18 describes the ports that control JTAG testing, while Table 10, page 18 describes the test instructions supported by these MX devices.

Each test section is accessed through the TAP, which has four associated pins: TCK (test clock input), TDI and TDO (test data input and output), and TMS (test mode selector).

The TAP controller is a four-bit state machine. The '1's and '0's represent the values that must be present at TMS at a rising edge of TCK for the given state transition to occur. IR and DR indicate that the instruction register or the data register is operating in that state.

The TAP controller receives two control inputs (TMS and TCK) and generates control and clock signals for the rest of the test logic architecture. On power-up, the TAP controller enters the Test-Logic-Reset state. To guarantee a reset of the controller from any of the possible states, TMS must remain high for five TCK cycles.

42MX24 and 42MX36 devices support three types of test data registers: bypass, device identification, and boundary scan. The bypass register is selected when no other register needs to be accessed in a device. This speeds up test data transfer to other devices in a test data path. The 32-bit device identification register is a shift register with four fields (lowest significant byte (LSB), ID number, part number and version). The boundary-scan register observes and controls the state of each I/O pin.

A sample calculation of the absolute maximum power dissipation allowed for a TQ176 package at commercial temperature and still air is given in the following equation

$$\text{MaximumPowerAllowed} = \frac{\text{Max} \cdot \text{junction temp} \cdot (°\text{C}) - \text{Max} \cdot \text{ambient temp} \cdot (°\text{C})}{\theta_{ja}(°(\text{C}/\text{W}))} = \frac{150°\text{C} - 70°\text{C}}{(28°\text{C})/\text{W}} = 2.86\text{W}$$

EQ 5

The maximum power dissipation for military-grade devices is a function of θ_{jc}. A sample calculation of the absolute maximum power dissipation allowed for CQFP 208-pin package at military temperature and still air is given in the following equation

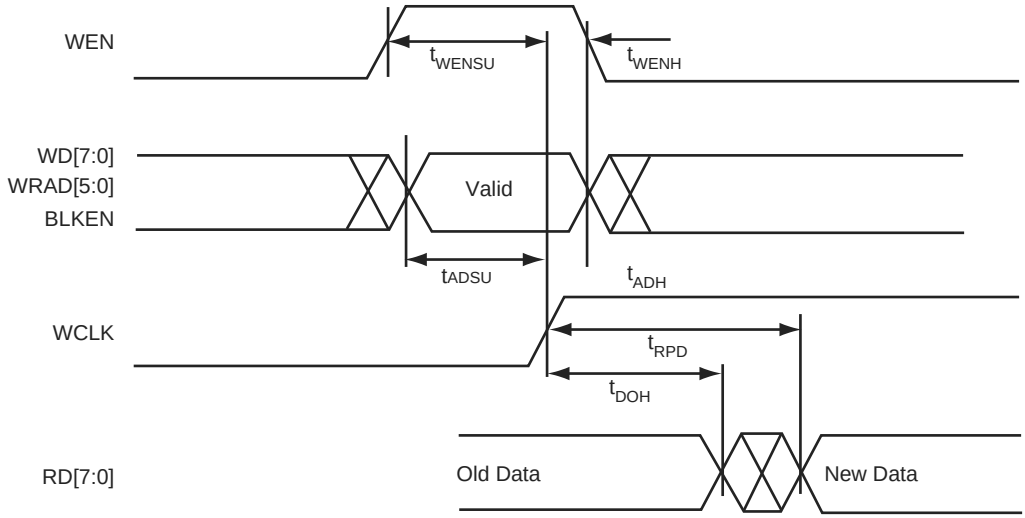
$$\text{MaximumPowerAllowed} = \frac{\text{Max} \cdot \text{junction temp} \cdot (°\text{C}) - \text{Max} \cdot \text{ambient temp} \cdot (°\text{C})}{\theta_{jc}(°(\text{C}/\text{W}))} = \frac{150°\text{C} - 125°\text{C}}{(6.3°\text{C})/\text{W}} = 3.97\text{W}$$

EQ 6

Table 27 • **Package Thermal Characteristics**

			θ_{ja}			
Plastic Packages	**Pin Count**	θ_{jc}	**Still Air**	**1.0 m/s 200 ft/min.**	**2.5 m/s 500 ft/min.**	**Units**
Plastic Quad Flat Pack	100	12.0	27.8	23.4	21.2	°C/W
Plastic Quad Flat Pack	144	10.0	26.2	22.8	21.1	°C/W
Plastic Quad Flat Pack	160	10.0	26.2	22.8	21.1	°C/W
Plastic Quad Flat Pack	208	8.0	26.1	22.5	20.8	°C/W
Plastic Quad Flat Pack	240	8.5	25.6	22.3	20.8	°C/W
Plastic Leaded Chip Carrier	44	16.0	20.0	24.5	22.0	°C/W
Plastic Leaded Chip Carrier	68	13.0	25.0	21.0	19.4	°C/W
Plastic Leaded Chip Carrier	84	12.0	22.5	18.9	17.6	°C/W
Thin Plastic Quad Flat Pack	176	11.0	24.7	19.9	18.0	°C/W
Very Thin Plastic Quad Flat Pack	80	12.0	38.2	31.9	29.4	°C/W
Very Thin Plastic Quad Flat Pack	100	10.0	35.3	29.4	27.1	°C/W
Plastic Ball Grid Array	272	3.0	18.3	14.9	13.9	°C/W
Ceramic Packages						
Ceramic Pin Grid Array	132	4.8	25.0	20.6	18.7	°C/W
Ceramic Quad Flat Pack	208	2.0	22.0	19.8	18.0	°C/W
Ceramic Quad Flat Pack	256	2.0	20.0	16.5	15.0	°C/W

Figure 33 • **42MX SRAM Asynchronous Read Operation—Type 2 (Write Address Controlled)**



3.10.7 Predictable Performance: Tight Delay Distributions

Propagation delay between logic modules depends on the resistive and capacitive loading of the routing tracks, the interconnect elements, and the module inputs being driven. Propagation delay increases as the length of routing tracks, the number of interconnect elements, or the number of inputs increases.

From a design perspective, the propagation delay can be statistically correlated or modeled by the fanout (number of loads) driven by a module. Higher fanout usually requires some paths to have longer routing tracks.

The MX FPGAs deliver a tight fanout delay distribution, which is achieved in two ways: by decreasing the delay of the interconnect elements and by decreasing the number of interconnect elements per path.

Microsemi's patented antifuse offers a very low resistive/capacitive interconnect. The antifuses, fabricated in 0.45 µm lithography, offer nominal levels of 100 Ω resistance and 7.0 fF capacitance per antifuse.

MX fanout distribution is also tight due to the low number of antifuses required for each interconnect path. The proprietary architecture limits the number of antifuses per path to a maximum of four, with 90 percent of interconnects using only two antifuses.

3.11 Timing Characteristics

Device timing characteristics fall into three categories: family-dependent, device-dependent, and design-dependent. The input and output buffer characteristics are common to all MX devices. Internal routing delays are device-dependent; actual delays are not determined until after place-and-route of the user's design is complete. Delay values may then be determined by using the Designer software utility or by performing simulation with post-layout delays.

3.11.1 Critical Nets and Typical Nets

Propagation delays are expressed only for typical nets, which are used for initial design performance evaluation. Critical net delays can then be applied to the most timing critical paths. Critical nets are determined by net property assignment in Microsemi's Designer software prior to placement and routing. Up to 6% of the nets in a design may be designated as critical.

3.11.2 Long Tracks

Some nets in the design use long tracks, which are special routing resources that span multiple rows, columns, or modules. Long tracks employ three and sometimes four antifuse connections, which increase capacitance and resistance, resulting in longer net delays for macros connected to long tracks. Typically, up to 6 percent of nets in a fully utilized device require long tracks. Long tracks add

Table 36 • **A40MX04 Timing Characteristics (Nominal 5.0 V Operation) *(continued)*(Worst-Case Commercial Conditions, VCC = 4.75 V, T_J = 70°C)**

Parameter / Description		–3 Speed		–2 Speed		–1 Speed		Std Speed		–F Speed		Units
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
t_{ENLZ}	Enable Pad LOW to Z		5.9		6.8		7.7		9.0		12.6	ns
d_{TLH}	Delta LOW to HIGH		0.02		0.02		0.03		0.03		0.04	ns/pF
d_{THL}	Delta HIGH to LOW		0.03		0.03		0.03		0.04		0.06	ns/pF

Table 38 • **A42MX09 Timing Characteristics (Nominal 5.0 V Operation) *(continued)*(Worst-Case Commercial Conditions, VCCA = 4.75 V, T_J = 70°C)**

Parameter / Description		–3 Speed		–2 Speed		–1 Speed		Std Speed		–F Speed		Units
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
t_{WASYN}	Flip-Flop (Latch) Asynchronous Pulse Width	4.5		4.9		5.6		6.6		9.2		ns
t_A	Flip-Flop Clock Input Period	3.5		3.8		4.3		5.1		7.1		ns
t_{INH}	Input Buffer Latch Hold	0.0		0.0		0.0		0.0		0.0		ns
t_{INSU}	Input Buffer Latch Set-Up	0.3		0.3		0.4		0.4		0.6		ns
t_{OUTH}	Output Buffer Latch Hold	0.0		0.0		0.0		0.0		0.0		ns
t_{OUTSU}	Output Buffer Latch Set-Up	0.3		0.3		0.4		0.4		0.6		ns
f_{MAX}	Flip-Flop (Latch) Clock Frequency		268		244		224		195		117	MHz

Table 40 • **A42MX16 Timing Characteristics (Nominal 5.0 V Operation) *(continued)*(Worst-Case Commercial Conditions, VCCA = 4.75 V, T_J = 70°C)**

Parameter / Description		–3 Speed		–2 Speed		–1 Speed		Std Speed		–F Speed		Units
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
TTL Output Module Timing[4]												
t_{DLH}	Data-to-Pad HIGH		2.5		2.8		3.2		3.7		5.2	ns
t_{DHL}	Data-to-Pad LOW		3.0		3.3		3.7		4.4		6.1	ns
t_{ENZH}	Enable Pad Z to HIGH		2.7		3.0		3.4		4.0		5.6	ns
t_{ENZL}	Enable Pad Z to LOW		3.0		3.3		3.8		4.4		6.2	ns
t_{ENHZ}	Enable Pad HIGH to Z		5.4		6.0		6.8		8.0		11.2	ns
t_{ENLZ}	Enable Pad LOW to Z		5.0		5.6		6.3		7.4		10.4	ns
t_{GLH}	G-to-Pad HIGH		2.9		3.2		3.6		4.3		6.0	ns
t_{GHL}	G-to-Pad LOW		2.9		3.2		3.6		4.3		6.0	ns
t_{LCO}	I/O Latch Clock-to-Out (Pad-to-Pad), 64 Clock Loading		5.7		6.3		7.1		8.4		11.9	ns
t_{ACO}	Array Clock-to-Out (Pad-to-Pad), 64 Clock Loading		8.0		8.9		10.1		11.9		16.7	ns
d_{TLH}	Capacitive Loading, LOW to HIGH		0.03		0.03		0.03		0.04		0.06	ns/pF
d_{THL}	Capacitive Loading, HIGH to LOW		0.04		0.04		0.04		0.05		0.07	ns/pF

Table 41 • **A42MX16 Timing Characteristics (Nominal 3.3 V Operation) *(continued)*(Worst-Case Commercial Conditions, VCCA = 3.0 V, T_J = 70°C)**

Parameter / Description		–3 Speed		–2 Speed		–1 Speed		Std Speed		–F Speed		Units
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
t_{ACO}	Array Clock-to-Out (Pad-to-Pad),64 Clock Loading		11.3		12.5		14.2		16.7		23.3	ns
d_{TLH}	Capacitive Loading, LOW to HIGH		0.04		0.04		0.05		0.06		0.08	ns/pF
d_{THL}	Capacitive Loading, HIGH to LOW		0.05		0.05		0.06		0.07		0.10	ns/pF

1. For dual-module macros use tPD1 + tRD1 + taped, to + tRD1 + taped, or tPD1 + tRD1 + tusk, whichever is appropriate.
2. Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing ansalysis or simulation is required to determine actual performance.
3. Data applies to macros based on the S-module. Timing parameters for sequential macros constructed from C-modules can be obtained from the Timer utility.
4. Set-up and hold timing parameters for the input buffer latch are defined with respect to the PAD and the D input. External setup/hold timing parameters must account for delay from an external PAD signal to the G inputs. Delay from an external PAD signal to the G input subtracts (adds) to the internal setup (hold) time.
5. Delays based on 35 pF loading.

Table 42 • **A42MX24 Timing Characteristics (Nominal 5.0 V Operation) (Worst-Case Commercial Conditions, VCCA = 4.75 V, T_J = 70°C)**

Parameter / Description		–3 Speed		–2 Speed		–1 Speed		Std Speed		–F Speed		Units
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	
Logic Module Combinatorial Functions[1]												
t_{PD}	Internal Array Module Delay		1.2		1.3		1.5		1.8		2.5	ns
t_{PDD}	Internal Decode Module Delay		1.4		1.6		1.8		2.1		3.0	ns
Logic Module Predicted Routing Delays[2]												
t_{RD1}	FO = 1 Routing Delay		0.8		0.9		1.0		1.2		1.7	ns
t_{RD2}	FO = 2 Routing Delay		1.0		1.2		1.3		1.5		2.1	ns
t_{RD3}	FO = 3 Routing Delay		1.3		1.4		1.6		1.9		2.6	ns
t_{RD4}	FO = 4 Routing Delay		1.5		1.7		1.9		2.2		3.1	ns
t_{RD5}	FO = 8 Routing Delay		2.4		2.7		3.0		3.6		5.0	ns
Logic Module Sequential Timing[3, 4]												
t_{CO}	Flip-Flop Clock-to-Output		1.3		1.4		1.6		1.9		2.7	ns
t_{GO}	Latch Gate-to-Output		1.2		1.3		1.5		1.8		2.5	ns
t_{SUD}	Flip-Flop (Latch) Set-Up Time	0.3		0.4		0.4		0.5		0.7		ns
t_{HD}	Flip-Flop (Latch) Hold Time	0.0		0.0		0.0		0.0		0.0		ns
t_{RO}	Flip-Flop (Latch) Reset-to-Output		1.4		1.6		1.8		2.1		2.9	ns
t_{SUENA}	Flip-Flop (Latch) Enable Set-Up	0.4		0.5		0.5		0.6		0.8		ns
t_{HENA}	Flip-Flop (Latch) Enable Hold	0.0		0.0		0.0		0.0		0.0		ns
t_{WCLKA}	Flip-Flop (Latch) Clock Active Pulse Width	3.3		3.7		4.2		4.9		6.9		ns
t_{WASYN}	Flip-Flop (Latch) Asynchronous Pulse Width	4.4		4.8		5.3		6.5		9.0		ns

Table 45 • **A42MX36 Timing Characteristics (Nominal 3.3 V Operation) *(continued)*(Worst-Case Commercial Conditions, VCCA = 3.0 V, T_J = 70°C)**

Parameter / Description		–3 Speed		–2 Speed		–1 Speed		Std Speed		–F Speed		
		Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Units
t_{ACO}	Array Latch Clock-to-Out (Pad-to-Pad) 32 I/O		10.9		12.1		13.7		16.1		22.5	ns
d_{TLH}	Capacitive Loading, LOW to HIGH		0.10		0.11		0.12		0.14		0.20	ns/pF
d_{THL}	Capacitive Loading, HIGH to LOW		0.10		0.11		0.12		0.14		0.20	ns/pF
CMOS Output Module Timing[5]												
t_{DLH}	Data-to-Pad HIGH		4.9		5.5		6.2		7.3		10.3	ns
t_{DHL}	Data-to-Pad LOW		3.4		3.8		4.3		5.1		7.1	ns
t_{ENZH}	Enable Pad Z to HIGH		3.7		4.1		4.7		5.5		7.7	ns
t_{ENZL}	Enable Pad Z to LOW		4.1		4.6		5.2		6.1		8.5	ns
t_{ENHZ}	Enable Pad HIGH to Z		7.4		8.2		9.3		10.9		15.3	ns
t_{ENLZ}	Enable Pad LOW to Z		6.9		7.6		8.7		10.2		14.3	ns
t_{GLH}	G-to-Pad HIGH		7.0		7.8		8.9		10.4		14.6	ns
t_{GHL}	G-to-Pad LOW		7.0		7.8		8.9		10.4		14.6	ns
t_{LSU}	I/O Latch Set-Up	0.7		0.7		0.8		1.0		1.4		ns
t_{LH}	I/O Latch Hold	0.0		0.0		0.0		0.0		0.0		ns
t_{LCO}	I/O Latch Clock-to-Out (Pad-to-Pad) 32 I/O		7.9		8.8		10.0		11.8		16.5	ns

1. For dual-module macros, use $t_{PD1} + t_{RD1} + t_{PDn}$, $t_{CO} + t_{RD1} + t_{PDn}$, or $t_{PD1} + t_{RD1} + t_{SUD}$, whichever is appropriate.
2. Routing delays are for typical designs across worst-case operating conditions. These parameters should be used for estimating device performance. Post-route timing analysis or simulation is required to determine actual performance.
3. Data applies to macros based on the S-module. Timing parameters for sequential macros constructed from C-modules can be obtained from the Timer utility.
4. *Set-up and hold timing parameters for the Input Buffer Latch are defined with respect to the PAD and the D input. External setup/hold timing parameters must account for delay from an external PAD signal to the G inputs. Delay from an external PAD signal to the G input subtracts (adds) to the internal setup (hold) time.*
5. Delays based on 35 pF loading.

3.12 Pin Descriptions

This section lists the pin descriptions for 40MX and 42MX series FPGAs.

CLK/A/B, I/O Global Clock

Clock inputs for clock distribution networks. CLK is for 40MX while CLKA and CLKB are for 42MX devices. The clock input is buffered prior to clocking the logic modules. This pin can also be used as an I/O.

DCLK, I/ODiagnostic Clock

Clock input for diagnostic probe and device programming. DCLK is active when the MODE pin is HIGH. This pin functions as an I/O when the MODE pin is LOW.

GND, Ground

Input LOW supply voltage.

I/O, Input/Output

Figure 44 • **PQ208**

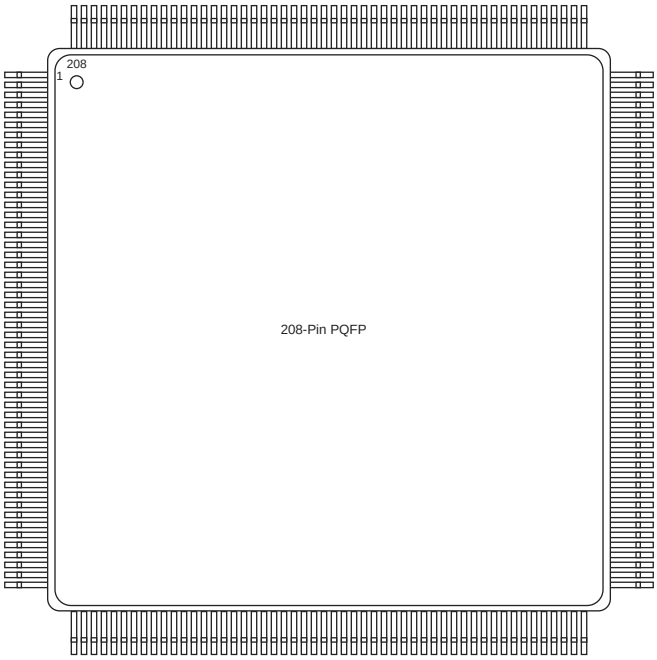


Table 53 • **PQ208**

PQ208			
Pin Number	**A42MX16 Function**	**A42MX24 Function**	**A42MX36 Function**
1	GND	GND	GND
2	NC	VCCA	VCCA
3	MODE	MODE	MODE
4	I/O	I/O	I/O
5	I/O	I/O	I/O
6	I/O	I/O	I/O
7	I/O	I/O	I/O
8	I/O	I/O	I/O
9	NC	I/O	I/O
10	NC	I/O	I/O
11	NC	I/O	I/O
12	I/O	I/O	I/O
13	I/O	I/O	I/O
14	I/O	I/O	I/O
15	I/O	I/O	I/O
16	NC	I/O	I/O
17	VCCA	VCCA	VCCA
18	I/O	I/O	I/O
19	I/O	I/O	I/O
20	I/O	I/O	I/O

Table 54 • **PQ240**

PQ240	
Pin Number	**A42MX36 Function**
89	VCCI
90	VCCA
91	LP
92	TCK, I/O
93	I/O
94	GND
95	I/O
96	I/O
97	I/O
98	I/O
99	I/O
100	I/O
101	I/O
102	I/O
103	I/O
104	I/O
105	I/O
106	I/O
107	I/O
108	VCCI
109	I/O
110	I/O
111	I/O
112	I/O
113	I/O
114	I/O
115	I/O
116	I/O
117	I/O
118	VCCA
119	GND
120	GND
121	GND
122	I/O
123	SDO, TDO, I/O
124	I/O
125	WD, I/O

Table 54 • **PQ240**

PQ240	
Pin Number	**A42MX36 Function**
200	I/O
201	I/O
202	I/O
203	I/O
204	I/O
205	I/O
206	VCCA
207	I/O
208	I/O
209	VCCA
210	VCCI
211	I/O
212	I/O
213	I/O
214	I/O
215	I/O
216	I/O
217	I/O
218	I/O
219	VCCA
220	I/O
221	I/O
222	I/O
223	I/O
224	I/O
225	I/O
226	I/O
227	VCCI
228	I/O
229	I/O
230	I/O
231	I/O
232	I/O
233	I/O
234	I/O
235	I/O
236	I/O

Table 57 • **TQ176**

TQ176			
Pin Number	**A42MX09 Function**	**A42MX16 Function**	**A42MX24 Function**
121	NC	NC	I/O
122	I/O	I/O	I/O
123	I/O	I/O	I/O
124	NC	I/O	I/O
125	NC	I/O	I/O
126	NC	NC	I/O
127	I/O	I/O	I/O
128	I/O	I/O	I/O
129	I/O	I/O	I/O
130	I/O	I/O	I/O
131	I/O	I/O	I/O
132	I/O	I/O	I/O
133	GND	GND	GND
134	I/O	I/O	I/O
135	SDI, I/O	SDI, I/O	SDI, I/O
136	NC	I/O	I/O
137	I/O	I/O	WD, I/O
138	I/O	I/O	WD, I/O
139	I/O	I/O	I/O
140	NC	VCCI	VCCI
141	I/O	I/O	I/O
142	I/O	I/O	I/O
143	NC	I/O	I/O
144	NC	I/O	WD, I/O
145	NC	NC	WD, I/O
146	I/O	I/O	I/O
147	NC	I/O	I/O
148	I/O	I/O	I/O
149	I/O	I/O	I/O
150	I/O	I/O	WD, I/O
151	NC	I/O	WD, I/O
152	PRA, I/O	PRA, I/O	PRA, I/O
153	I/O	I/O	I/O
154	CLKA, I/O	CLKA, I/O	CLKA, I/O
155	VCCA	VCCA	VCCA
156	GND	GND	GND
157	I/O	I/O	I/O

Table 58 • **CQ208**

CQ208	
Pin Number	**A42MX36 Function**
148	I/O
149	I/O
150	GND
151	I/O
152	I/O
153	I/O
154	I/O
155	I/O
156	I/O
157	GND
158	I/O
159	SDI, I/O
160	I/O
161	WD, I/O
162	WD, I/O
163	I/O
164	VCCI
165	I/O
166	I/O
167	I/O
168	WD, I/O
169	WD, I/O
170	I/O
171	QCLKD, I/O
172	I/O
173	I/O
174	I/O
175	I/O
176	WD, I/O
177	WD, I/O
178	PRA, I/O
179	I/O
180	CLKA, I/O
181	I/O
182	VCCI
183	VCCA
184	GND

Table 59 • **CQ256**

CQ256	
Pin Number	**A42MX36 Function**
59	I/O
60	VCCA
61	GND
62	GND
63	NC
64	NC
65	NC
66	I/O
67	SDO, TDO, I/O
68	I/O
69	WD, I/O
70	WD, I/O
71	I/O
72	VCCI
73	I/O
74	I/O
75	I/O
76	WD, I/O
77	GND
78	WD, I/O
79	I/O
80	QCLKB, I/O
81	I/O
82	I/O
83	I/O
84	I/O
85	I/O
86	I/O
87	WD, I/O
88	WD, I/O
89	I/O
90	I/O
91	I/O
92	I/O
93	I/O
94	I/O
95	VCCI

Table 60 • **BG272**

BG272	
Pin Number	**A42MX36 Function**
A6	I/O
A7	WD, I/O
A8	WD, I/O
A9	I/O
A10	I/O
A11	CLKA
A12	I/O
A13	I/O
A14	I/O
A15	I/O
A16	WD, I/O
A17	I/O
A18	I/O
A19	GND
A20	GND
B1	GND
B2	GND
B3	DCLK, I/O
B4	I/O
B5	I/O
B6	I/O
B7	WD, I/O
B8	I/O
B9	PRB, I/O
B10	I/O
B11	I/O
B12	WD, I/O
B13	I/O
B14	I/O
B15	WD, I/O
B16	I/O
B17	WD, I/O
B18	I/O
B19	GND
B20	GND
C1	I/O
C2	MODE

Table 60 • **BG272**

BG272	
Pin Number	**A42MX36 Function**
J9	GND
J10	GND
J11	GND
J12	GND
J17	VCCA
J18	I/O
J19	I/O
J20	I/O
K1	I/O
K2	I/O
K3	I/O
K4	VCCI
K9	GND
K10	GND
K11	GND
K12	GND
K17	I/O
K18	VCCA
K19	VCCA
K20	LP
L1	I/O
L2	I/O
L3	VCCA
L4	VCCA
L9	GND
L10	GND
L11	GND
L12	GND
L17	VCCI
L18	I/O
L19	I/O
L20	TCK, I/O
M1	I/O
M2	I/O
M3	I/O
M4	VCCI
M9	GND

Table 60 • **BG272**

BG272	
Pin Number	**A42MX36 Function**
T19	I/O
T20	I/O
U1	I/O
U2	I/O
U3	I/O
U4	I/O
U5	VCCI
U6	WD, I/O
U7	I/O
U8	I/O
U9	WD, I/O
U10	VCCA
U11	VCCI
U12	I/O
U13	I/O
U14	QCLKB, I/O
U15	I/O
U16	VCCI
U17	I/O
U18	GND
U19	I/O
U20	I/O
V1	I/O
V2	I/O
V3	GND
V4	GND
V5	I/O
V6	I/O
V7	I/O
V8	WD, I/O
V9	I/O
V10	I/O
V11	I/O
V12	I/O
V13	WD, I/O
V14	I/O
V15	WD, I/O

Table 61 • **PG132**

PG132	
Pin Number	**A42MX09 Function**
F2	I/O
F1	I/O
G1	I/O
G4	VSV
H1	I/O
H2	I/O
H3	I/O
H4	I/O
J1	I/O
K1	I/O
L1	I/O
K2	I/O
M1	I/O
K3	I/O
L2	I/O
N1	I/O
L3	BININ
M2	BINOUT
N2	I/O
M3	I/O
L4	I/O
N3	I/O
M4	I/O
N4	I/O
M5	I/O
K6	I/O
N5	I/O
N6	I/O
L6	I/O
M6	I/O
M7	I/O
N7	I/O
N8	I/O
M8	I/O
L8	I/O
K8	I/O
N9	I/O