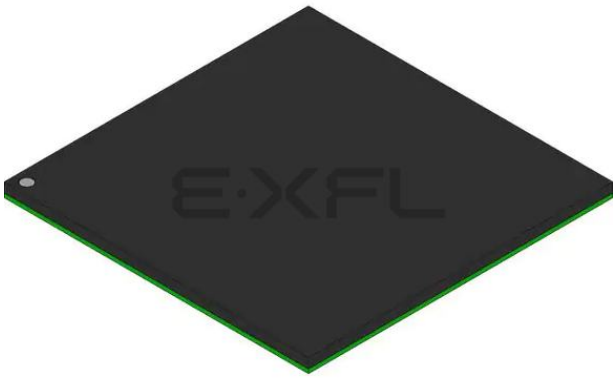

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Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Active
Core Processor	MPC8xx
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	66MHz
Co-Processors/DSP	Communications; CPM
RAM Controllers	DRAM
Graphics Acceleration	No
Display & Interface Controllers	-
Ethernet	10Mbps (1)
SATA	-
USB	USB 1.x (1)
Voltage - I/O	3.3V
Operating Temperature	0°C ~ 95°C (TA)
Security Features	-
Package / Case	256-BGA
Supplier Device Package	256-PBGA (23x23)
Purchase URL	https://www.e-xfl.com/pro/item?MUrl=&PartUrl=xpc850zt66bu

2 Features

Figure 1 is a block diagram of the MPC850, showing its major components and the relationships among those components:

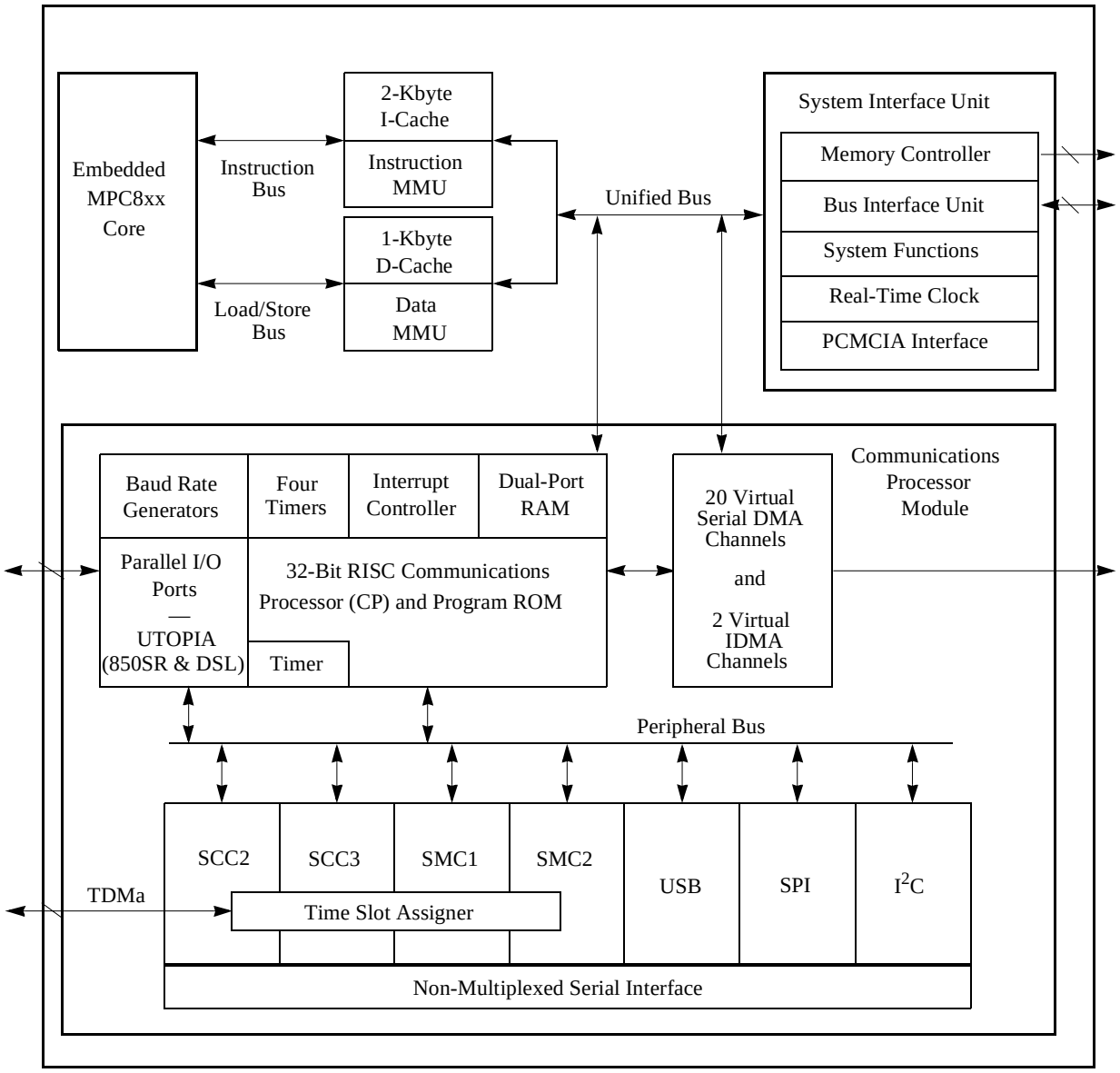


Figure 1. MPC850 Microprocessor Block Diagram

The following list summarizes the main features of the MPC850:

- Embedded single-issue, 32-bit MPC8xx core (implementing the PowerPC architecture) with thirty-two 32-bit general-purpose registers (GPRs)
  - Performs branch folding and branch prediction with conditional prefetch, but without conditional execution

4 Thermal Characteristics

Table 3 shows the thermal characteristics for the MPC850.

Table 3. Thermal Characteristics

Characteristic	Symbol	Value	Unit
Thermal resistance for BGA [1]	θ_{JA}	40 [2]	°C/W
	θ_{JA}	31 [3]	°C/W
	θ_{JA}	24 [4]	°C/W
Thermal Resistance for BGA (junction-to-case)	θ_{JC}	8	°C/W

[1] For more information on the design of thermal vias on multilayer boards and BGA layout considerations in general, refer to AN-1231/D, Plastic Ball Grid Array Application Note available from your local Freescale sales office.

[2] Assumes natural convection and a single layer board (no thermal vias).

[3] Assumes natural convection, a multilayer board with thermal vias[4], 1 watt MPC850 dissipation, and a board temperature rise of 20°C above ambient.

[4] Assumes natural convection, a multilayer board with thermal vias[4], 1 watt MPC850 dissipation, and a board temperature rise of 13°C above ambient.

$T_J = T_A + (P_D \bullet \theta_{JA})$

$P_D = (V_{DD} \bullet I_{DD}) + P_{I/O}$

where:

$P_{I/O}$ is the power dissipation on pins

Table 4 provides power dissipation information.

Table 4. Power Dissipation (P_D)

Characteristic	Frequency (MHz)	Typical [1]	Maximum [2]	Unit
Power Dissipation All Revisions (1:1) Mode	33	TBD	515	mW
	40	TBD	590	mW
	50	TBD	725	mW

[1] Typical power dissipation is measured at 3.3V

[2] Maximum power dissipation is measured at 3.65 V

Table 5 provides the DC electrical characteristics for the MPC850.

Table 5. DC Electrical Specifications

Characteristic	Symbol	Min	Max	Unit
Operating voltage at 40 MHz or less	VDDH, VDDL, KAPWR, VDDSYN	3.0	3.6	V
Operating voltage at 40 MHz or higher	VDDH, VDDL, KAPWR, VDDSYN	3.135	3.465	V
Input high voltage (address bus, data bus, EXTAL, EXTCLK, and all bus control/status signals)	VIH	2.0	3.6	V
Input high voltage (all general purpose I/O and peripheral pins)	VIH	2.0	5.5	V

Table 5. DC Electrical Specifications (continued)

Characteristic	Symbol	Min	Max	Unit
Input low voltage	VIL	GND	0.8	V
EXTAL, EXTCLK input high voltage	VIHC	0.7*(VCC)	VCC+0.3	V
Input leakage current, Vin = 5.5 V (Except TMS, TRST, DSCK and DSDI pins)	I_{in}	—	100	µA
Input leakage current, Vin = 3.6V (Except TMS, TRST, DSCK and DSDI pins)	I_{In}	—	10	µA
Input leakage current, Vin = 0V (Except TMS, TRST, DSCK and DSDI pins)	I_{In}	—	10	µA
Input capacitance	C_{in}	—	20	pF
Output high voltage, IOH = -2.0 mA, VDDH = 3.0V except XTAL, XFC, and open-drain pins	VOH	2.4	—	V
Output low voltage CLKOUT [3] IOL = 3.2 mA [1] IOL = 5.3 mA [2] IOL = 7.0 mA PA[14]/USBOE, PA[12]/TXD2 IOL = 8.9 mA TS, TA, TEA, BI, BB, HRESET, SRESET	VOL	—	0.5	V

1 A[6:31], TSIZ0/REG, TSIZ1, D[0:31], DP[0:3]/IRQ[3:6], RD/WR, BURST, RSV/IRQ2, IP_B[0:1]/IWP[0:1]/VFLS[0:1], IP_B2/IOIS16_B/AT2, IP_B3/IWP2/VF2, IP_B4/LWP0/VF0, IP_B5/LWP1/VF1, IP_B6/DSDI/AT0, IP_B7/PTR/AT3, PA[15]/USBRXD, PA[13]/RXD2, PA[9]/L1TXDA/SMRXD2, PA[8]/L1RXDA/SMTXD2, PA[7]/CLK1/TIN1/L1RCLKA/BRGO1, PA[6]/CLK2/TOUT1/TIN3, PA[5]/CLK3/TIN2/L1TCLKA/BRGO2, PA[4]/CLK4/TOUT2/TIN4, PB[31]/SPISEL, PB[30]/SPICLK/TXD3, PB[29]/SPIMOSI /RXD3, PB[28]/SPIMISO/BRGO3, PB[27]/I2CSDA/BRGO1, PB[26]/I2CSCL/BRGO2, PB[25]/SMTXD1/TXD3, PB[24]/SMRXD1/RXD3, PB[23]/SMSYN1/SDACK1, PB[22]/SMSYN2/SDACK2, PB[19]/L1ST1, PB[18]/RTS2/L1ST2, PB[17]/L1ST3, PB[16]/L1RQa/L1ST4, PC[15]/DREQ0/L1ST5, PC[14]/DREQ1/RTS2/L1ST6, PC[13]/L1ST7/RTS3, PC[12]/L1RQa/L1ST8, PC[11]/USBRXP, PC[10]/TGATE1/USBRXN, PC[9]/CTS2, PC[8]/CD2/TGATE1, PC[7]/USBTXP, PC[6]/USBTXN, PC[5]/CTS3/L1TSYNCA/SDACK1, PC[4]/CD3/L1RSYNCA, PD[15], PD[14], PD[13], PD[12], PD[11], PD[10], PD[9], PD[8], PD[7], PD[6], PD[5], PD[4], PD[3]

2 BDIP/GPL_B5, BR, BG, FRZ/IRQ6, CS[0:5], CS6/CE1_B, CS7/CE2_B, WE0/BS_AB0/IORD, WE1/BS_AB1/IOWR, WE2/BS_AB2/PCOE, WE3/BS_AB3/PCWE, GPL_A0/GPL_B0, OE/GPL_A1/GPL_B1, GPL_A[2:3]/GPL_B[2:3]/CS[2:3], UPWAITA/GPL_A4/AS, UPWAITB/GPL_B4, GPL_A5, ALE_B/DSCK/AT1, OP2/MODCK1/STS, OP3/MODCK2/DSDO

3 The MPC850 IBIS model must be used to accurately model the behavior of the Clkout output driver for the full and half drive setting. Due to the nature of the Clkout output buffer, IOH and IOL for Clkout should be extracted from the IBIS model at any output voltage level.

5 Power Considerations

The average chip-junction temperature, T_J, in °C can be obtained from the equation:

$T_J = T_A + (P_D \bullet \theta_{JA})(1)$

where

T_A = Ambient temperature, °C

θ_{JA} = Package thermal resistance, junction to ambient, °C/W

$P_D = P_{INT} + P_{I/O}$

$P_{INT} = I_{DD} \times V_{DD}$, watts—chip internal power

$P_{I/O}$ = Power dissipation on input and output pins—user determined

For most applications $P_{I/O} < 0.3 \bullet P_{INT}$ and can be neglected. If $P_{I/O}$ is neglected, an approximate relationship between P_D and T_J is:

$P_D = K \div (T_J + 273°C)$(2)

Solving equations (1) and (2) for K gives:

$K = P_D \bullet (T_A + 273°C) + \theta_{JA} \bullet P_D^2$(3)

where K is a constant pertaining to the particular part. K can be determined from equation (3) by measuring P_D (at equilibrium) for a known T_A. Using this value of K, the values of P_D and T_J can be obtained by solving equations (1) and (2) iteratively for any value of T_A.

5.1 Layout Practices

Each V_{CC} pin on the MPC850 should be provided with a low-impedance path to the board's supply. Each GND pin should likewise be provided with a low-impedance path to ground. The power supply pins drive distinct groups of logic on chip. The V_{CC} power supply should be bypassed to ground using at least four 0.1 µF by-pass capacitors located as close as possible to the four sides of the package. The capacitor leads and associated printed circuit traces connecting to chip V_{CC} and GND should be kept to less than half an inch per capacitor lead. A four-layer board is recommended, employing two inner layers as V_{CC} and GND planes.

All output pins on the MPC850 have fast rise and fall times. Printed circuit (PC) trace interconnection length should be minimized in order to minimize undershoot and reflections caused by these fast output switching times. This recommendation particularly applies to the address and data busses. Maximum PC trace lengths of six inches are recommended. Capacitance calculations should consider all device loads as well as parasitic capacitances due to the PC traces. Attention to proper PCB layout and bypassing becomes especially critical in systems with higher capacitive loads because these loads create higher transient currents in the V_{CC} and GND circuits. Pull up all unused inputs or signals that will be inputs during reset. Special care should be taken to minimize the noise levels on the PLL supply pins.

6 Bus Signal Timing

Table 6 provides the bus operation timing for the MPC850 at 50 MHz, 66 MHz, and 80 MHz. Timing information for other bus speeds can be interpolated by equation using the MPC850 Electrical Specifications Spreadsheet found at http://www.mot.com/netcomm.

The maximum bus speed supported by the MPC850 is 50 MHz. Higher-speed parts must be operated in half-speed bus mode (for example, an MPC850 used at 66 MHz must be configured for a 33 MHz bus).

The timing for the MPC850 bus shown assumes a 50-pF load. This timing can be derated by 1 ns per 10 pF. Derating calculations can also be performed using the MPC850 Electrical Specifications Spreadsheet.

Table 6. Bus Operation Timing [1] (continued)

Num	Characteristic	50 MHz		66 MHz		80 MHz		FFACT	Cap Load (default 50 pF)	Unit
		Min	Max	Min	Max	Min	Max			
B33a	CLKOUT rising edge to GPL valid - as requested by control bit GxT3 in the corresponding word in the UPM	5.00	12.00	8.00	14.00	6.00	13.00	0.250	50.00	ns
B34	A[6–31] and D[0–31] to $\overline{CS}$ valid - as requested by control bit CST4 in the corresponding word in the UPM	3.00	—	6.00	—	4.00	—	0.250	50.00	ns
B34a	A[6–31] and D[0–31] to $\overline{CS}$ valid - as requested by control bit CST1 in the corresponding word in the UPM	8.00	—	13.00	—	11.00	—	0.500	50.00	ns
B34b	A[6–31] and D[0–31] to $\overline{CS}$ valid - as requested by CST2 in the corresponding word in UPM	13.00	—	21.00	—	17.00	—	0.750	50.00	ns
B35	A[6–31] to $\overline{CS}$ valid - as requested by control bit BST4 in the corresponding word in UPM	3.00	—	6.00	—	4.00	—	0.250	50.00	ns
B35a	A[6–31] and D[0–31] to $\overline{BS}$ valid - as requested by BST1 in the corresponding word in the UPM	8.00	—	13.00	—	11.00	—	0.500	50.00	ns
B35b	A[6–31] and D[0–31] to $\overline{BS}$ valid - as requested by control bit BST2 in the corresponding word in the UPM	13.00	—	21.00	—	17.00	—	0.750	50.00	ns
B36	A[6–31] and D[0–31] to GPL valid - as requested by control bit GxT4 in the corresponding word in the UPM	3.00	—	6.00	—	4.00	—	0.250	50.00	ns
B37	UPWAIT valid to CLKOUT falling edge [10]	6.00	—	6.00	—	6.00	—	—	50.00	ns
B38	CLKOUT falling edge to UPWAIT valid [10]	1.00	—	1.00	—	1.00	—	—	50.00	ns
B39	$\overline{AS}$ valid to CLKOUT rising edge [11]	7.00	—	7.00	—	7.00	—	—	50.00	ns
B40	A[6–31], TSIZ[0–1], RD/$\overline{WR}$, $\overline{BURST}$, valid to CLKOUT rising edge.	7.00	—	7.00	—	7.00	—	—	50.00	ns
B41	$\overline{TS}$ valid to CLKOUT rising edge (setup time)	7.00	—	7.00	—	7.00	—	—	50.00	ns

Table 6. Bus Operation Timing [1] (continued)

Num	Characteristic	50 MHz		66 MHz		80 MHz		FFACT	Cap Load (default 50 pF)	Unit
		Min	Max	Min	Max	Min	Max			
B42	CLKOUT rising edge to $\overline{TS}$ valid (hold time)	2.00	—	2.00	—	2.00	—	—	50.00	ns
B43	$\overline{AS}$ negation to memory controller signals negation	—	TBD	—	TBD	TBD	—	—	50.00	ns

1 The minima provided assume a 0 pF load, whereas maxima assume a 50pF load. For frequencies not marked on the part, new bus timing must be calculated for all frequency-dependent AC parameters. Frequency-dependent AC parameters are those with an entry in the FFactor column. AC parameters without an FFactor entry do not need to be calculated and can be taken directly from the frequency column corresponding to the frequency marked on the part. The following equations should be used in these calculations.

For a frequency F, the following equations should be applied to each one of the above parameters:

For minima:

$$D = \frac{FFACTOR \times 1000}{F} + (D_{50} - 20 \times FFACTOR)$$

For maxima:

$$D = \frac{FFACTOR \times 1000}{F} + (D_{50} - 20 \times FFACTOR) + 1ns(CAP\ LOAD - 50) / 10$$

where:

D is the parameter value to the frequency required in ns

F is the operation frequency in MHz

D_{50} is the parameter value defined for 50 MHz

CAP LOAD is the capacitance load on the signal in question.

FFACTOR is the one defined for each of the parameters in the table.

2 Phase and frequency jitter performance results are valid only if the input jitter is less than the prescribed value.

3 If the rate of change of the frequency of EXTAL is slow (i.e. it does not jump between the minimum and maximum values in one cycle) or the frequency of the jitter is fast (i.e., it does not stay at an extreme value for a long time) then the maximum allowed jitter on EXTAL can be up to 2%.

4 The timing for $\overline{BR}$ output is relevant when the MPC850 is selected to work with external bus arbiter. The timing for $\overline{BG}$ output is relevant when the MPC850 is selected to work with internal bus arbiter.

5 The setup times required for $\overline{TA}$, $\overline{TEA}$, and $\overline{BI}$ are relevant only when they are supplied by an external device (and not when the memory controller or the PCMCIA interface drives them).

6 The timing required for $\overline{BR}$ input is relevant when the MPC850 is selected to work with the internal bus arbiter. The timing for $\overline{BG}$ input is relevant when the MPC850 is selected to work with the external bus arbiter.

7 The D[0–31] and DP[0–3] input timings B20 and B21 refer to the rising edge of the CLKOUT in which the $\overline{TA}$ input signal is asserted.

8 The D[0:31] and DP[0:3] input timings B20 and B21 refer to the falling edge of CLKOUT. This timing is valid only for read accesses controlled by chip-selects controlled by the UPM in the memory controller, for data beats where DLT3 = 1 in the UPM RAM words. (This is only the case where data is latched on the falling edge of CLKOUT.

9 The timing B30 refers to $\overline{CS}$ when ACS = '00' and to $\overline{WE[0:3]}$ when CSNT = '0'.

10 The signal UPWAIT is considered asynchronous to CLKOUT and synchronized internally. The timings specified in B37 and B38 are specified to enable the freeze of the UPM output signals.

11 The $\overline{AS}$ signal is considered asynchronous to CLKOUT.

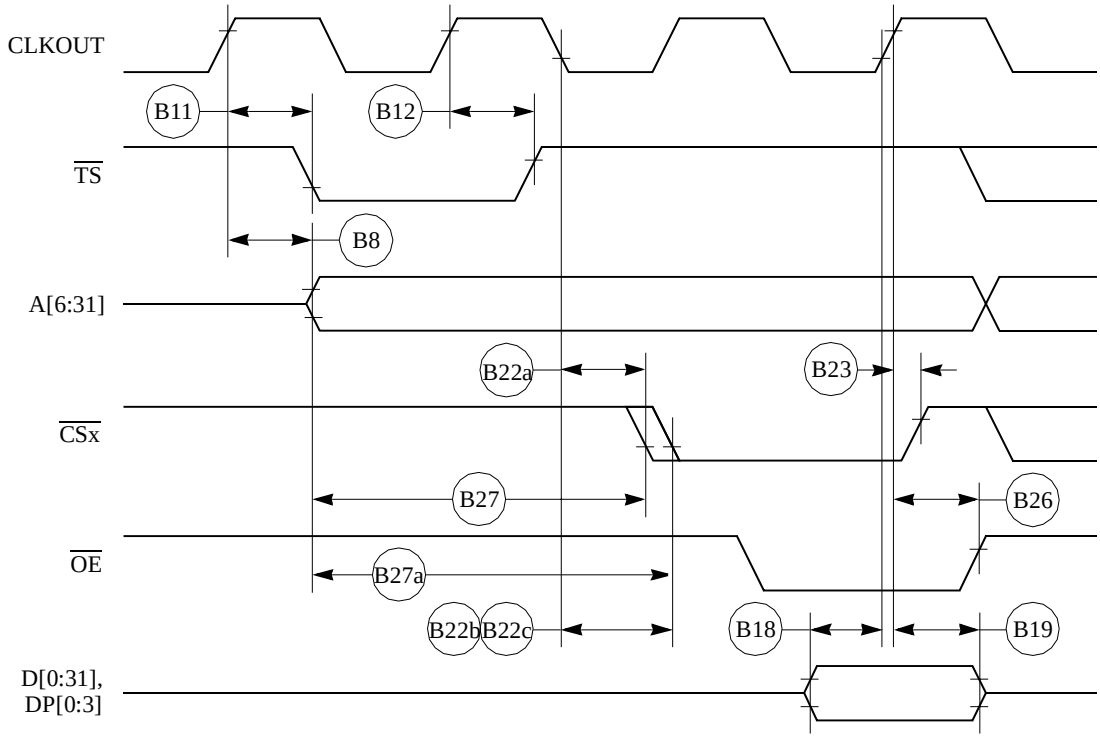


Figure 12. External Bus Read Timing (GPCM Controlled—TRLX = 1, ACS = 10, ACS = 11)

Figure 13 through Figure 15 provide the timing for the external bus write controlled by various GPCM factors.

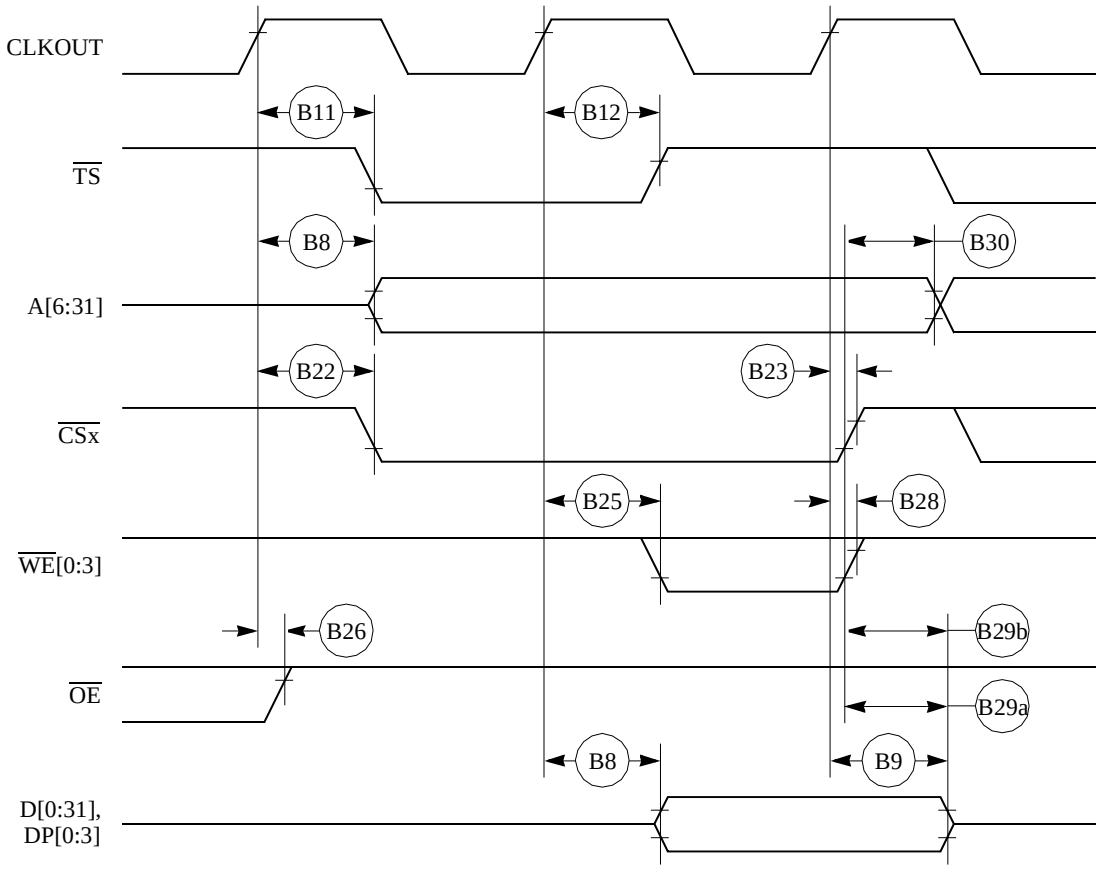


Figure 13. External Bus Write Timing (GPCM Controlled—TRLX = 0, CSNT = 0)

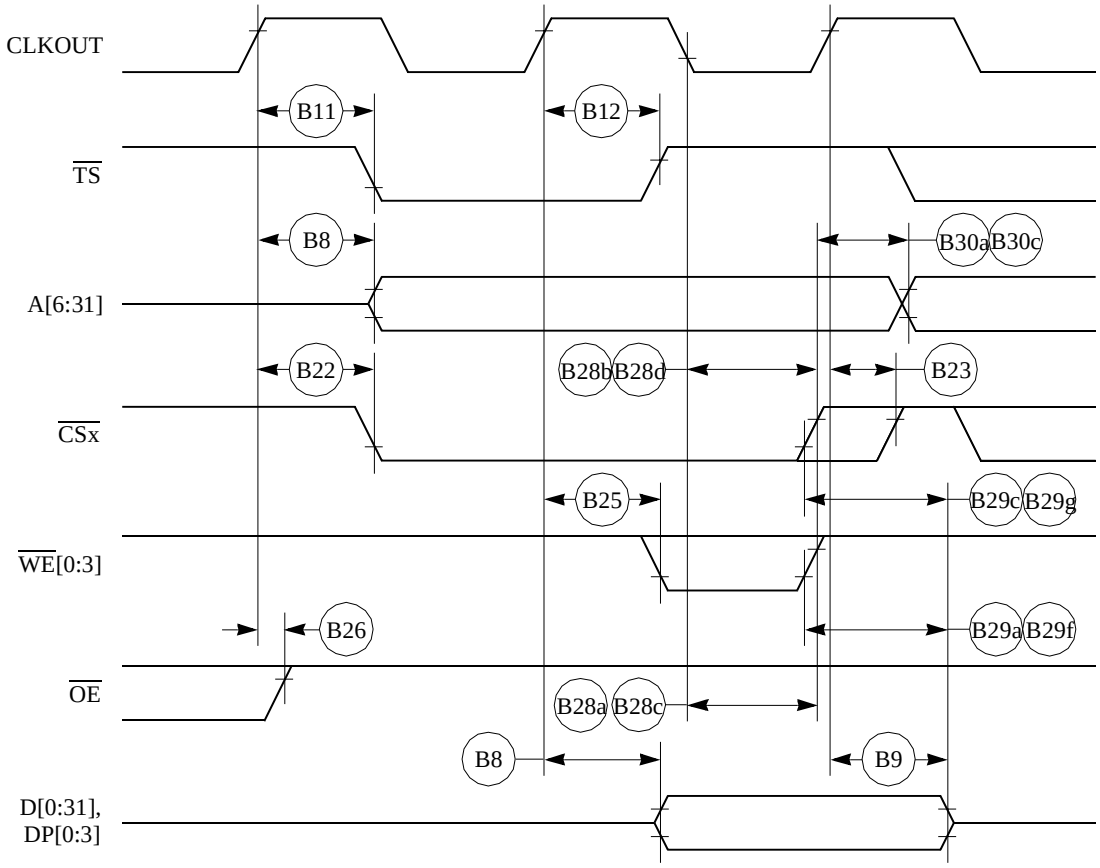


Figure 14. External Bus Write Timing (GPCM Controlled—TRLX = 0, CSNT = 1)

Table 10 shows the debug port timing for the MPC850.

Table 10. Debug Port Timing

Num	Characteristic	50 MHz		66 MHz		80 MHz		Unit
		Min	Max	Min	Max	Min	Max	
D61	DSCK cycle time	60.00	—	91.00	—	75.00	—	ns
D62	DSCK clock pulse width	25.00	—	38.00	—	31.00	—	ns
D63	DSCK rise and fall times	0.00	3.00	0.00	3.00	0.00	3.00	ns
D64	DSDI input data setup time	8.00	—	8.00	—	8.00	—	ns
D65	DSDI data hold time	5.00	—	5.00	—	5.00	—	ns
D66	DSCK low to DSDO data valid	0.00	15.00	0.00	15.00	0.00	15.00	ns
D67	DSCK low to DSDO invalid	0.00	2.00	0.00	2.00	0.00	2.00	ns

Figure 29 provides the input timing for the debug port clock.

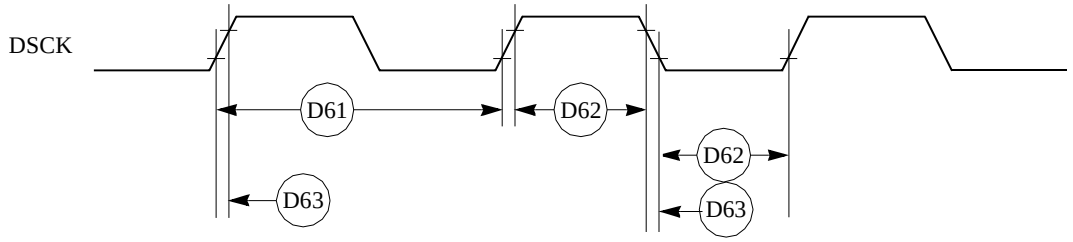


Figure 29. Debug Port Clock Input Timing

Figure 30 provides the timing for the debug port.

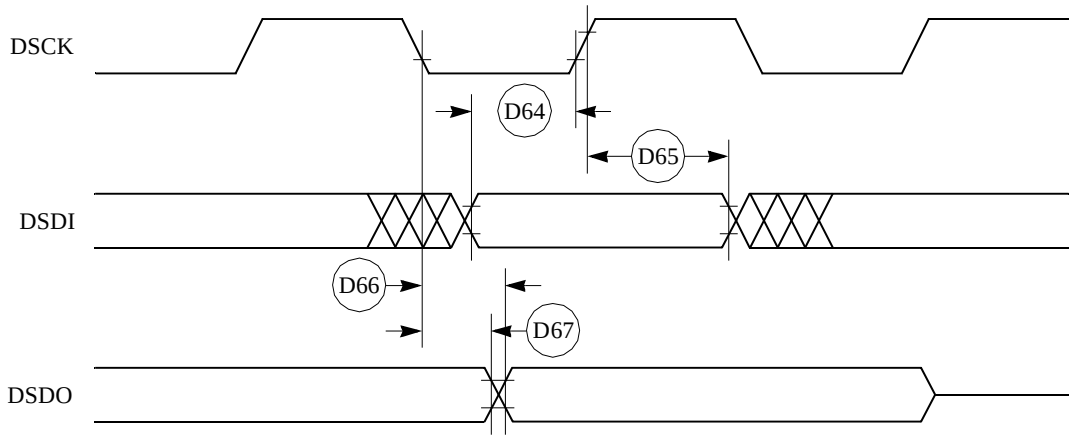


Figure 30. Debug Port Timings

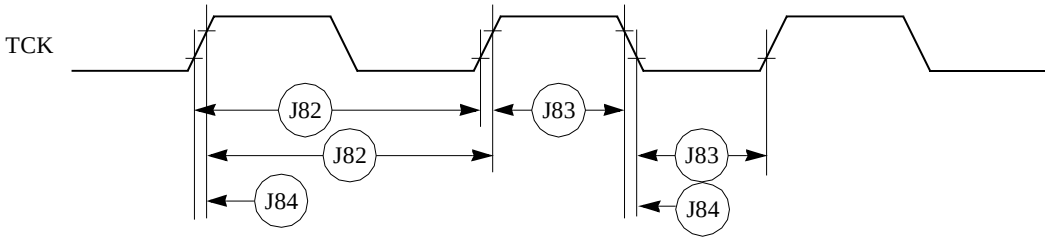


Figure 34. JTAG Test Clock Input Timing

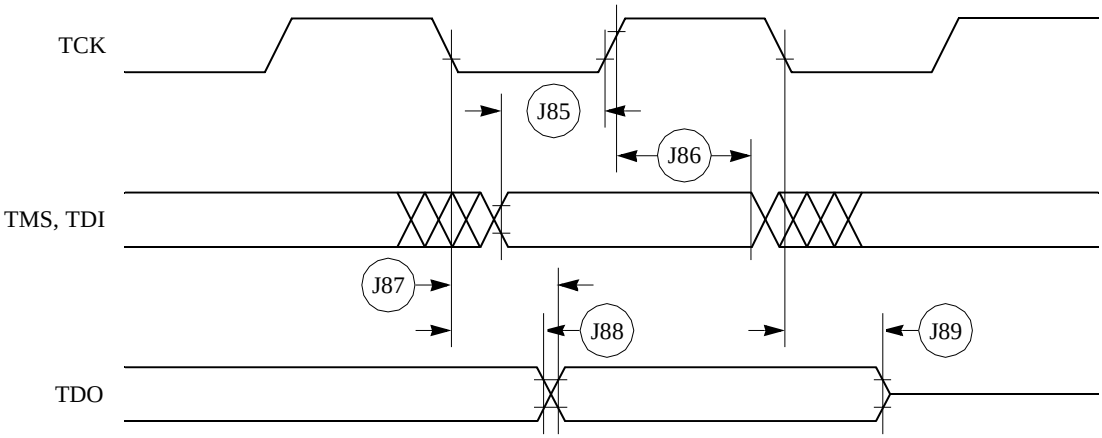


Figure 35. JTAG Test Access Port Timing Diagram

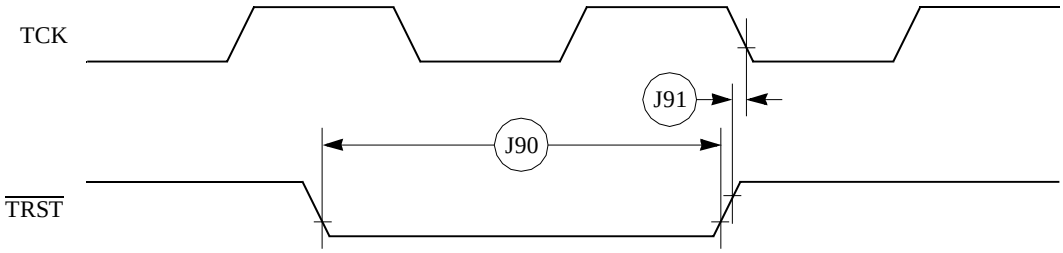


Figure 36. JTAG $\overline{\text{TRST}}$ Timing Diagram

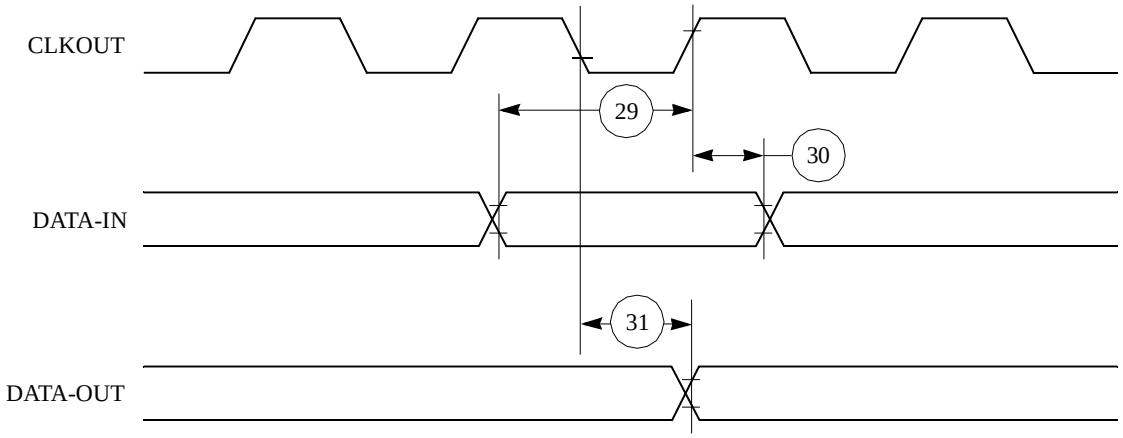


Figure 38. Parallel I/O Data-In/Data-Out Timing Diagram

8.2 IDMA Controller AC Electrical Specifications

Table 14 provides the IDMA controller timings as shown in Figure 39 to Figure 42.

Table 14. IDMA Controller Timing

Num	Characteristic	All Frequencies		Unit
		Min	Max	
40	$\overline{\text{DREQ}}$ setup time to clock high	7.00	—	ns
41	$\overline{\text{DREQ}}$ hold time from clock high	3.00	—	ns
42	$\overline{\text{SDACK}}$ assertion delay from clock high	—	12.00	ns
43	$\overline{\text{SDACK}}$ negation delay from clock low	—	12.00	ns
44	$\overline{\text{SDACK}}$ negation delay from $\overline{\text{TA}}$ low	—	20.00	ns
45	$\overline{\text{SDACK}}$ negation delay from clock high	—	15.00	ns
46	$\overline{\text{TA}}$ assertion to falling edge of the clock setup time (applies to external $\overline{\text{TA}}$)	7.00	—	ns

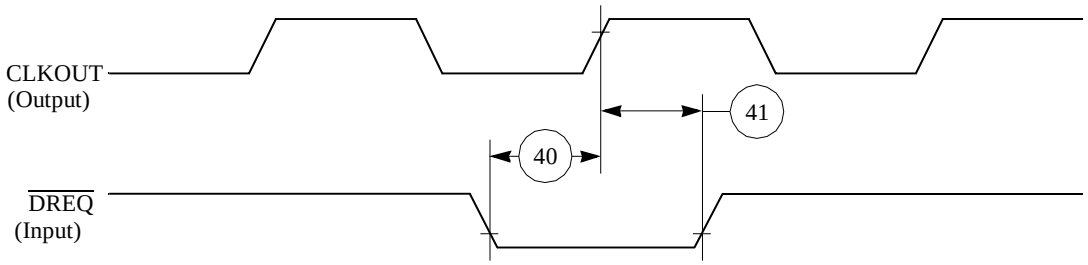


Figure 39. IDMA External Requests Timing Diagram

Table 17. SI Timing (continued)

Num	Characteristic	All Frequencies		Unit
		Min	Max	
82	L1RCLK, L1TCLK frequency (DSC =1)	—	16.00 or SYNCCLK/2	MHz
83	L1RCLK, L1TCLK width low (DSC =1)	P + 10	—	ns
83A	L1RCLK, L1TCLK width high (DSC = 1)[3]	P + 10	—	ns
84	L1CLK edge to L1CLKO valid (DSC = 1)	—	30.00	ns
85	$\overline{\text{L1RQ}}$ valid before falling edge of L1TSYNC[4]	1.00	—	L1TCLK
86	L1GR setup time[2]	42.00	—	ns
87	L1GR hold time	42.00	—	ns
88	L1xCLK edge to L1SYNC valid (FSD = 00) CNT = 0000, BYT = 0, DSC = 0)	—	0.00	ns

1 The ratio SyncCLK/L1RCLK must be greater than 2.5/1.

2 These specs are valid for IDL mode only.

3 Where P = 1/CLKOUT. Thus for a 25-MHz CLKO1 rate, P = 40 ns.

4 These strobes and TxD on the first bit of the frame become valid after L1CLK edge or L1SYNC, whichever is later.

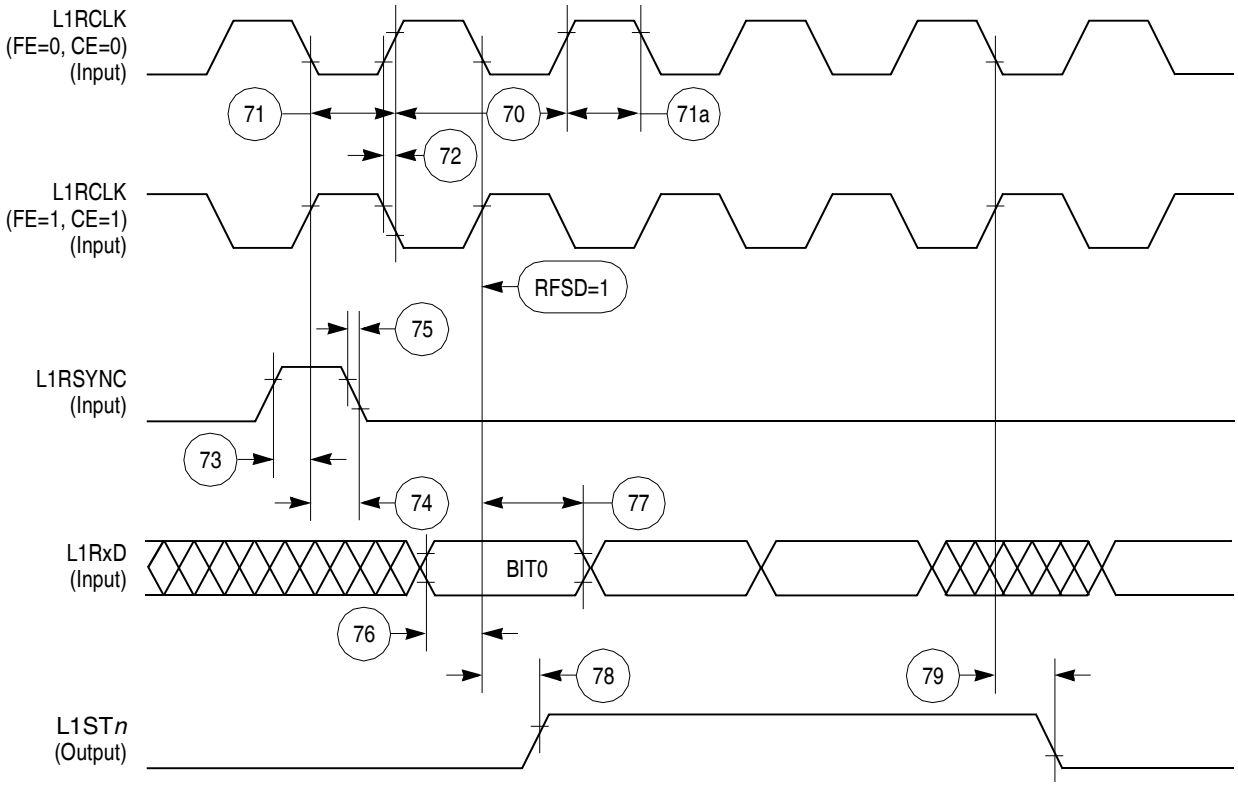


Figure 45. SI Receive Timing Diagram with Normal Clocking (DSC = 0)

Table 20. Ethernet Timing (continued)

Num	Characteristic	All Frequencies		Unit
		Min	Max	
134	TENA inactive delay (from TCLKx rising edge)	10.00	50.00	ns
138	CLKOUT low to $\overline{\text{SDACK}}$ asserted [2]	—	20.00	ns
139	CLKOUT low to $\overline{\text{SDACK}}$ negated [2]	—	20.00	ns

1 The ratios SyncCLK/RCLKx and SyncCLK/TCLKx must be greater or equal to 2/1.

2 $\overline{\text{SDACK}}$ is asserted whenever the SDMA writes the incoming frame destination address into memory.

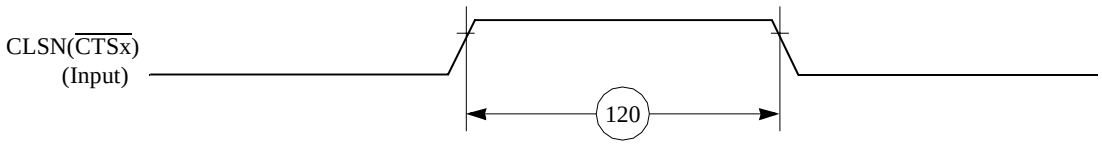


Figure 53. Ethernet Collision Timing Diagram

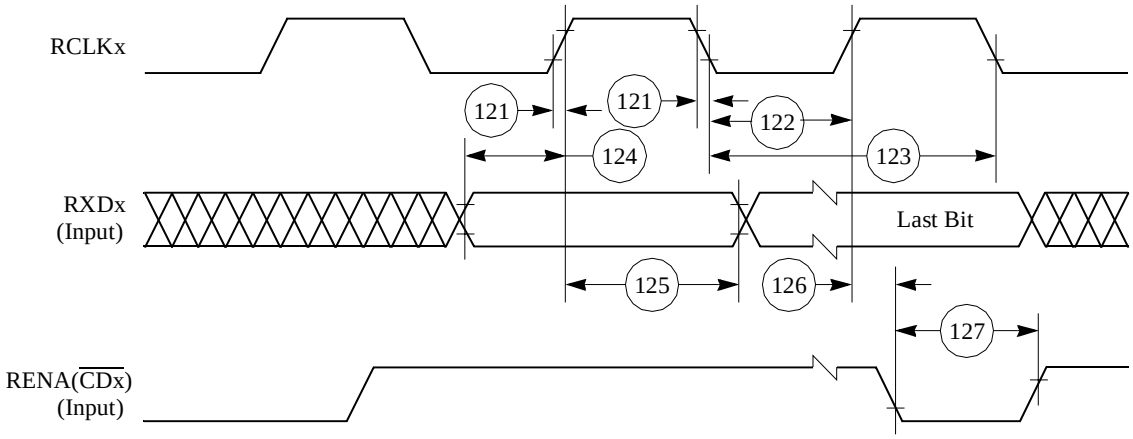
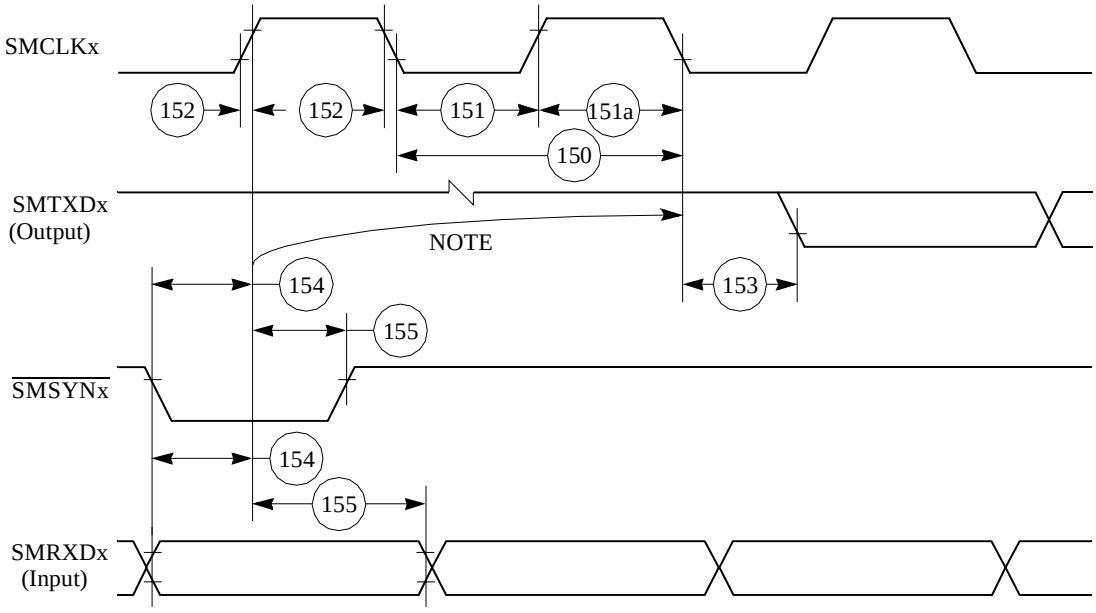


Figure 54. Ethernet Receive Timing Diagram



NOTE:
1. This delay is equal to an integer number of character-length clocks.

Figure 56. SMC Transparent Timing Diagram

8.9 SPI Master AC Electrical Specifications

Table 22 provides the SPI master timings as shown in Figure 57 and Figure 58.

Table 22. SPI Master Timing

Num	Characteristic	All Frequencies		Unit
		Min	Max	
160	MASTER cycle time	4	1024	t_{cyc}
161	MASTER clock (SCK) high or low time	2	512	t_{cyc}
162	MASTER data setup time (inputs)	50.00	—	ns
163	Master data hold time (inputs)	0.00	—	ns
164	Master data valid (after SCK edge)	—	20.00	ns
165	Master data hold time (outputs)	0.00	—	ns
166	Rise time output	—	15.00	ns
167	Fall time output	—	15.00	ns

8.10 SPI Slave AC Electrical Specifications

Table 23 provides the SPI slave timings as shown in Figure 59 and Figure 60.

Table 23. SPI Slave Timing

Num	Characteristic	All Frequencies		Unit
		Min	Max	
170	Slave cycle time	2	—	t_{cyc}
171	Slave enable lead time	15.00	—	ns
172	Slave enable lag time	15.00	—	ns
173	Slave clock (SPICLK) high or low time	1	—	t_{cyc}
174	Slave sequential transfer delay (does not require deselect)	1	—	t_{cyc}
175	Slave data setup time (inputs)	20.00	—	ns
176	Slave data hold time (inputs)	20.00	—	ns
177	Slave access time	—	50.00	ns
178	Slave SPI MISO disable time	—	50.00	ns
179	Slave data valid (after SPICLK edge)	—	50.00	ns
180	Slave data hold time (outputs)	0.00	—	ns
181	Rise time (input)	—	15.00	ns
182	Fall time (input)	—	15.00	ns

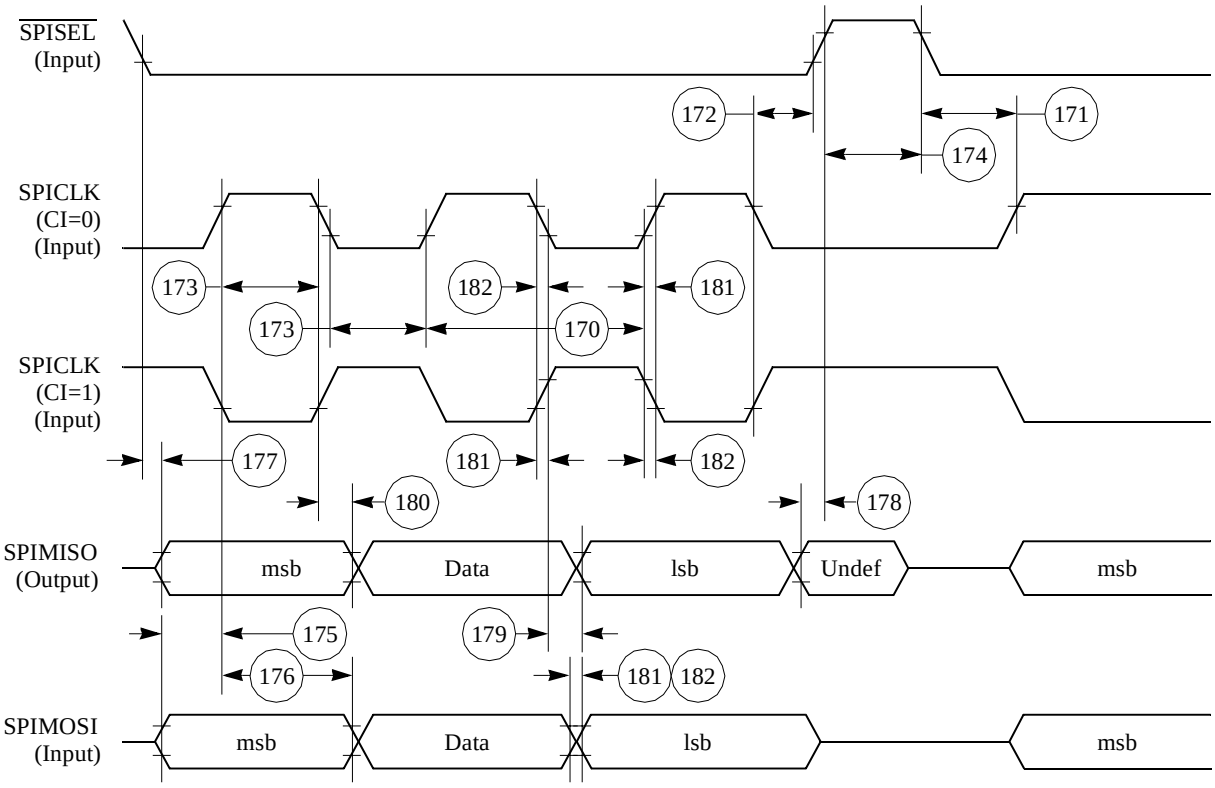


Figure 59. SPI Slave (CP = 0) Timing Diagram

Table 24. I²C Timing (SCL < 100 KHz) (CONTINUED)

Num	Characteristic	All Frequencies		Unit
		Min	Max	
210	SDL/SCL fall time	—	300.00	ns
211	Stop condition setup time	4.70	—	µs

1 SCL frequency is given by SCL = BRGCLK_frequency / ((BRG register + 3) * pre_scaler * 2). The ratio SyncClk/(BRGCLK/pre_scaler) must be greater or equal to 4/1.

Table 25 provides the I²C (SCL > 100 KHz) timings.

Table 25. I²C Timing (SCL > 100 KHz)

Num	Characteristic	Expression	All Frequencies		Unit
			Min	Max	
200	SCL clock frequency (slave)	fSCL	0	BRGCLK/48	Hz
200	SCL clock frequency (master) [1]	fSCL	BRGCLK/16512	BRGCLK/48	Hz
202	Bus free time between transmissions		1/(2.2 * fSCL)	—	s
203	Low period of SCL		1/(2.2 * fSCL)	—	s
204	High period of SCL		1/(2.2 * fSCL)	—	s
205	Start condition setup time		1/(2.2 * fSCL)	—	s
206	Start condition hold time		1/(2.2 * fSCL)	—	s
207	Data hold time		0	—	s
208	Data setup time		1/(40 * fSCL)	—	s
209	SDL/SCL rise time		—	1/(10 * fSCL)	s
210	SDL/SCL fall time		—	1/(33 * fSCL)	s
211	Stop condition setup time		1/2(2.2 * fSCL)	—	s

1 SCL frequency is given by SCL = BrgClk_frequency / ((BRG register + 3) * pre_scaler * 2). The ratio SyncClk/(Brg_Clk/pre_scaler) must be greater or equal to 4/1.

Figure 61 shows the I²C bus timing.

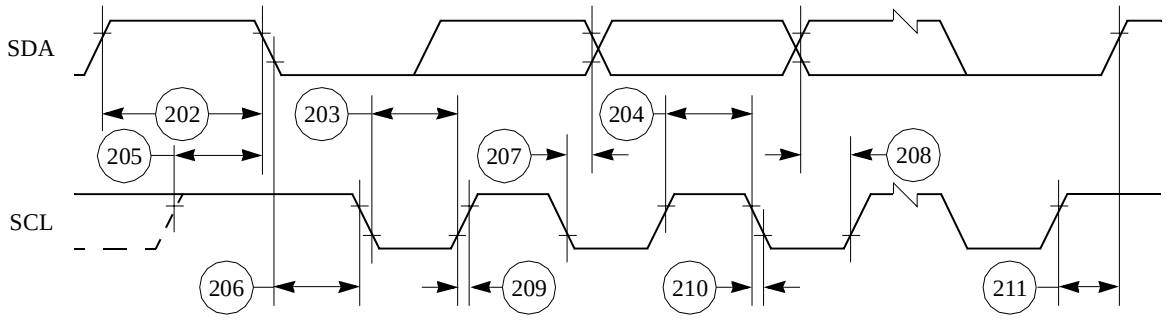


Figure 61. I²C Bus Timing Diagram

customers that are currently using the non-JEDEC pin numbering scheme, two sets of pinouts, JEDEC and non-JEDEC, are presented in this document.

Figure 62 shows the non-JEDEC pinout of the PBGA package as viewed from the top surface.

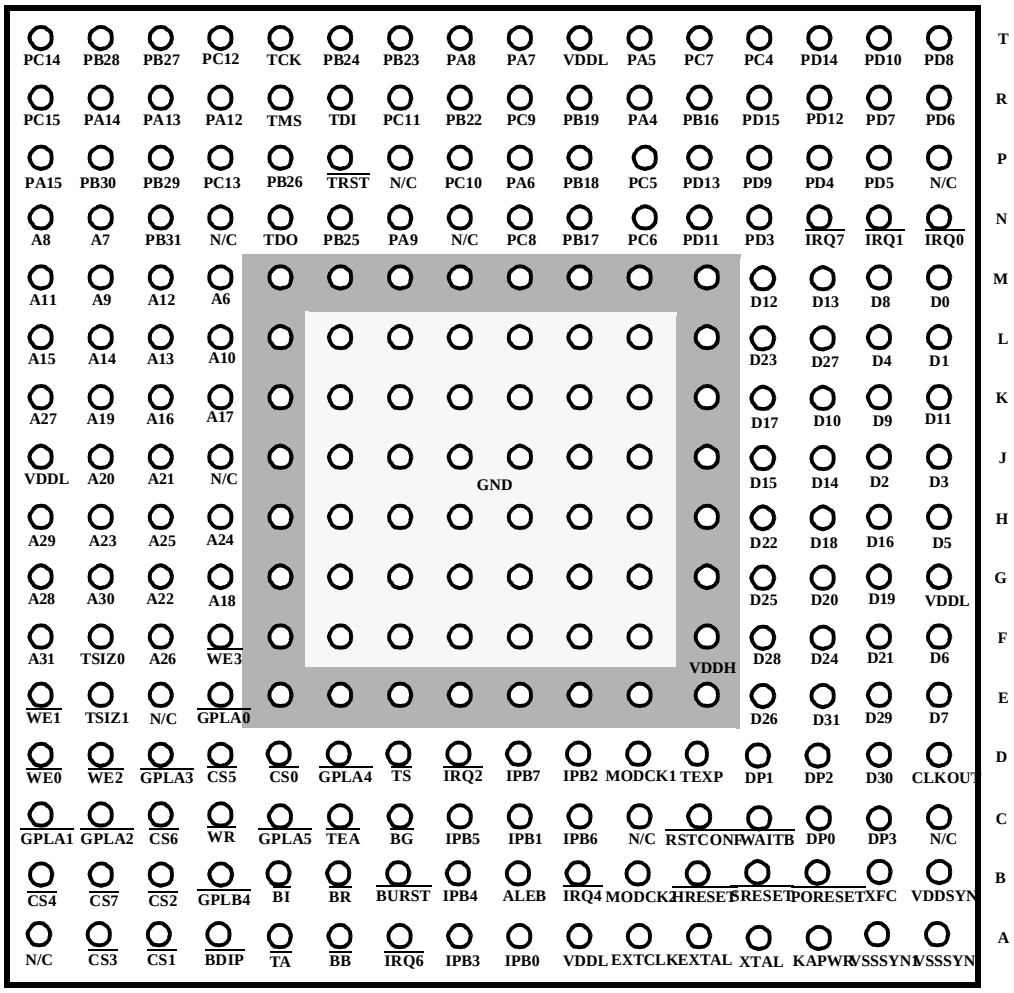
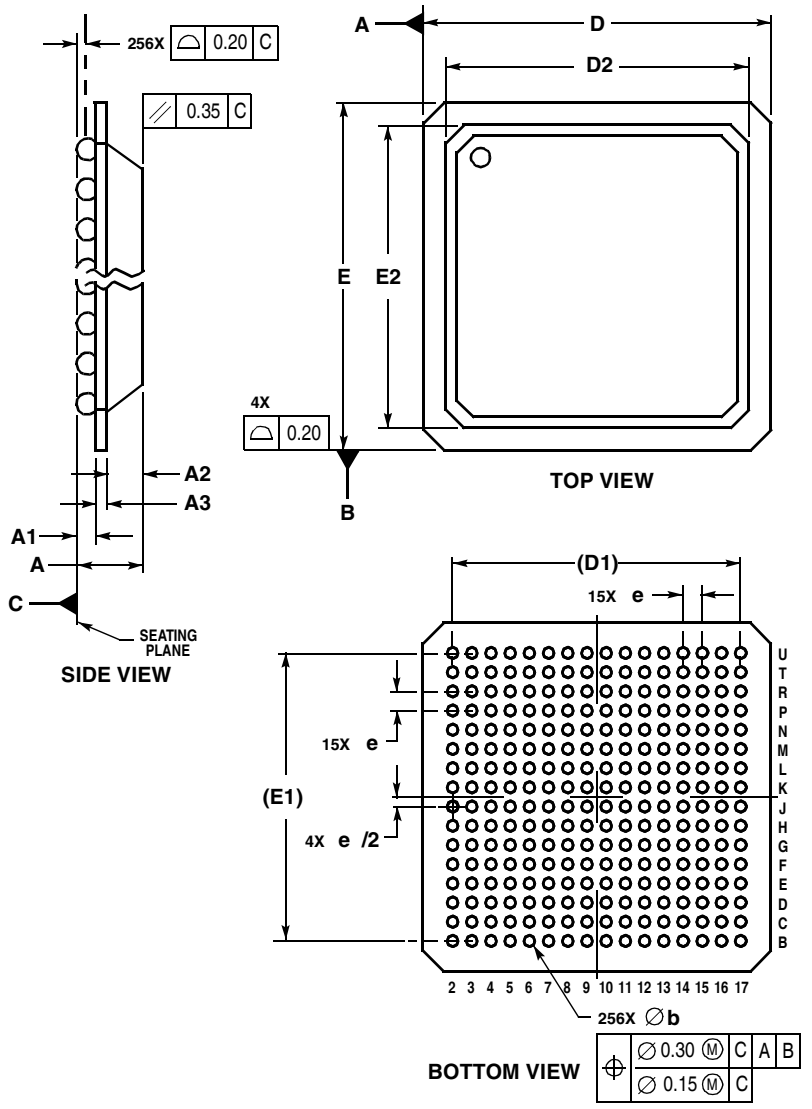


Figure 62. Pin Assignments for the PBGA (Top View)—non-JEDEC Standard

Figure 65 shows the JEDEC package dimensions of the PBGA.



NOTES:
1. DIMENSIONING AND TOLERANCING PER ASME Y14.5M, 1994.
2. DIMENSIONS IN MILLIMETERS.
3. DIMENSION b IS MEASURED AT THE MAXIMUM SOLDER BALL DIAMETER, PARALLEL TO PRIMARY DATUM C.
4. PRIMARY DATUM C AND THE SEATING PLANE ARE

	MILLIMETERS	
DIM	MIN	MAX
A	1.91	2.35
A1	0.50	0.70
A2	1.12	1.22
A3	0.29	0.43
b	0.60	0.90
D	23.00 BSC	
D1	19.05 REF	
D2	19.00	20.00
E	23.00 BSC	
E1	19.05 REF	
E2	19.00	20.00
e	1.27 BSC	

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Figure 65. Package Dimensions for the Plastic Ball Grid Array (PBGA)—JEDEC Standard