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Understanding Embedded - CPLDs (Complex Programmable Logic Devices)

Embedded - CPLDs, or Complex Programmable Logic Devices, are highly versatile digital logic devices used in electronic systems. These programmable components are designed to perform complex logical operations and can be customized for specific applications. Unlike fixed-function ICs, CPLDs offer the flexibility to reprogram their configuration, making them an ideal choice for various embedded systems. They consist of a set of logic gates and programmable interconnects, allowing designers to implement complex logic circuits without needing custom hardware.

Applications of Embedded - CPLDs

Details	
Product Status	Obsolete
Programmable Type	In System Programmable
Delay Time tpd(1) Max	10 ns
Voltage Supply - Internal	3V ~ 3.6V
Number of Logic Elements/Blocks	4
Number of Macrocells	64
Number of Gates	1250
Number of I/O	34
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	44-TQFP
Supplier Device Package	44-TQFP (10x10)
Purchase URL	https://www.e-xfl.com/product-detail/intel/epm3064ati44-10nad

MAX 3000A devices contain 32 to 512 macrocells, combined into groups of 16 macrocells called logic array blocks (LABs). Each macrocell has a programmable–AND / fixed–OR array and a configurable register with independently programmable clock, clock enable, clear, and preset functions. To build complex logic functions, each macrocell can be supplemented with shareable expander and high–speed parallel expander product terms to provide up to 32 product terms per macrocell.

MAX 3000A devices provide programmable speed / power optimization. Speed–critical portions of a design can run at high speed / full power, while the remaining portions run at reduced speed / low power. This speed / power optimization feature enables the designer to configure one or more macrocells to operate at 50% or lower power while adding only a nominal timing delay. MAX 3000A devices also provide an option that reduces the slew rate of the output buffers, minimizing noise transients when non–speed–critical signals are switching. The output drivers of all MAX 3000A devices can be set for 2.5 V or 3.3 V, and all input pins are 2.5–V, 3.3–V, and 5.0-V tolerant, allowing MAX 3000A devices to be used in mixed–voltage systems.

MAX 3000A devices are supported by Altera development systems, which are integrated packages that offer schematic, text—including VHDL, Verilog HDL, and the Altera Hardware Description Language (AHDL)—and waveform design entry, compilation and logic synthesis, simulation and timing analysis, and device programming. The software provides EDIF 2 0 0 and 3 0 0, LPM, VHDL, Verilog HDL, and other interfaces for additional design entry and simulation support from other industry–standard PC– and UNIX–workstation–based EDA tools. The software runs on Windows–based PCs, as well as Sun SPARCstation, and HP 9000 Series 700 / 800 workstations.

For more information on development tools, see the *MAX+PLUS II Programmable Logic Development System & Software Data Sheet* and the *Quartus Programmable Logic Development System & Software Data Sheet*.

Functional Description

The MAX 3000A architecture includes the following elements:

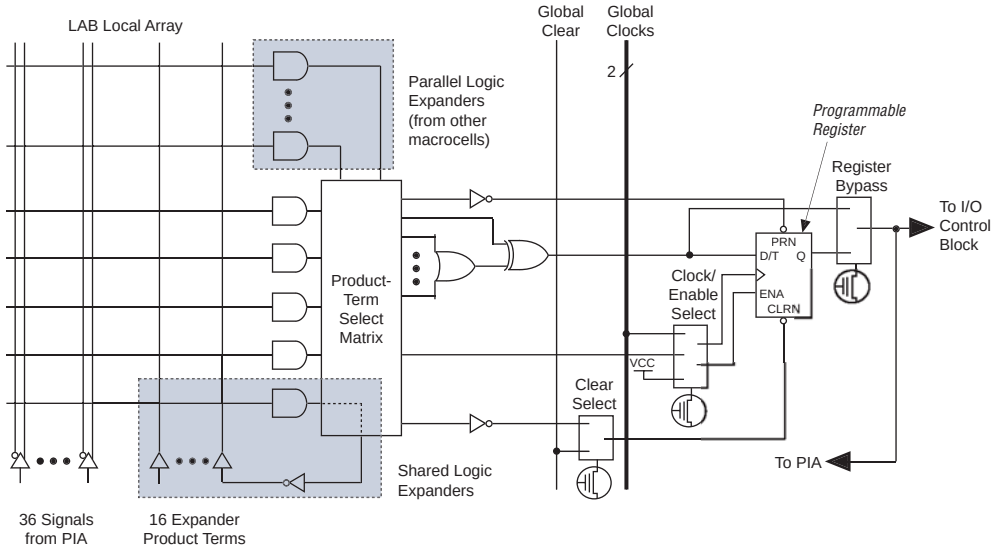
- Logic array blocks (LABs)
- Macrocells
- Expander product terms (shareable and parallel)
- Programmable interconnect array (PIA)
- I / O control blocks

The MAX 3000A architecture includes four dedicated inputs that can be used as general–purpose inputs or as high–speed, global control signals (clock, clear, and two output enable signals) for each macrocell and I / O pin. Figure 1 shows the architecture of MAX 3000A devices.

Macrocells

MAX 3000A macrocells can be individually configured for either sequential or combinatorial logic operation. Macrocells consist of three functional blocks: logic array, product–term select matrix, and programmable register. Figure 2 shows a MAX 3000A macrocell.

Figure 2. MAX 3000A Macrocell



Combinatorial logic is implemented in the logic array, which provides five product terms per macrocell. The product–term select matrix allocates these product terms for use as either primary logic inputs (to the OR and XOR gates) to implement combinatorial functions, or as secondary inputs to the macrocell's register preset, clock, and clock enable control functions.

Two kinds of expander product terms ("expanders") are available to supplement macrocell logic resources:

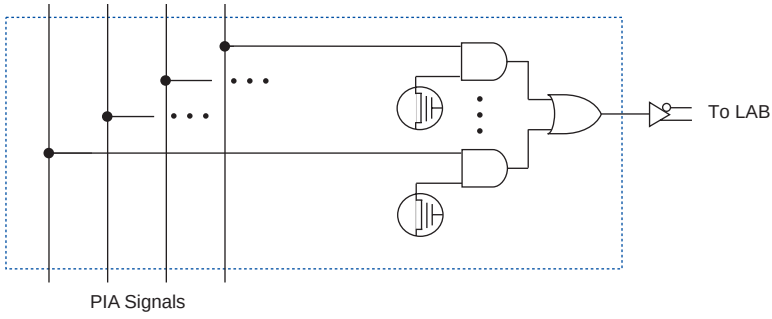
- Shareable expanders, which are inverted product terms that are fed back into the logic array
- Parallel expanders, which are product terms borrowed from adjacent macrocells

The Altera development system automatically optimizes product–term allocation according to the logic requirements of the design.

Programmable Interconnect Array

Logic is routed between LABs on the PIA. This global bus is a programmable path that connects any signal source to any destination on the device. All MAX 3000A dedicated inputs, I/O pins, and macrocell outputs feed the PIA, which makes the signals available throughout the entire device. Only the signals required by each LAB are actually routed from the PIA into the LAB. Figure 5 shows how the PIA signals are routed into the LAB. An EEPROM cell controls one input to a two-input AND gate, which selects a PIA signal to drive into the LAB.

Figure 5. MAX 3000A PIA Routing

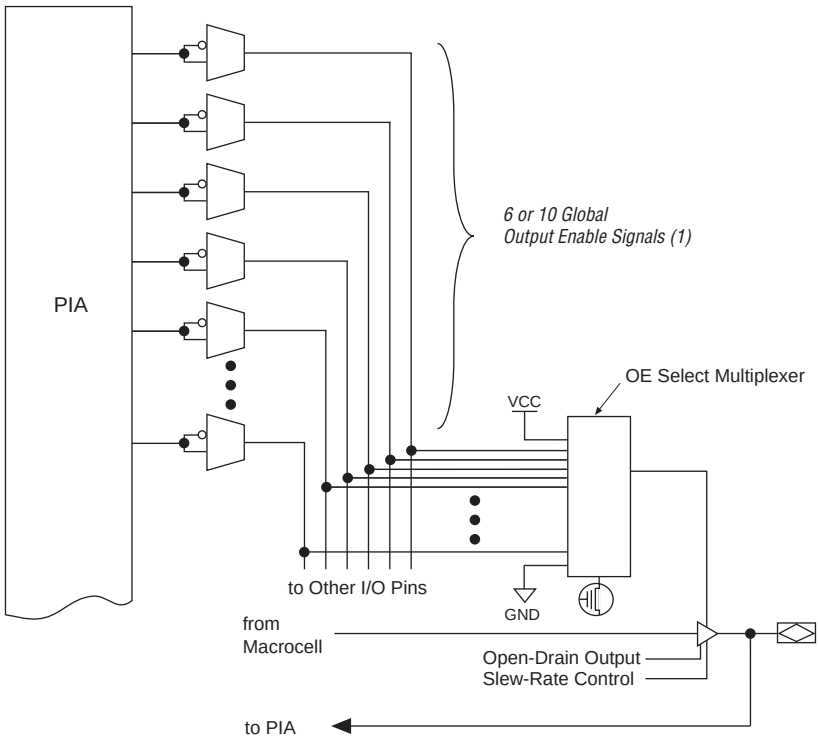


While the routing delays of channel–based routing schemes in masked or FPGAs are cumulative, variable, and path–dependent, the MAX 3000A PIA has a predictable delay. The PIA makes a design's timing performance easy to predict.

I/O Control Blocks

The I/O control block allows each I/O pin to be individually configured for input, output, or bidirectional operation. All I/O pins have a tri–state buffer that is individually controlled by one of the global output enable signals or directly connected to ground or V_{CC}. Figure 6 shows the I/O control block for MAX 3000A devices. The I/O control block has 6 or 10 global output enable signals that are driven by the true or complement of two output enable signals, a subset of the I/O pins, or a subset of the I/O macrocells.

Figure 6. I/O Control Block of MAX 3000A Devices



Note:

(1) EPM3032A, EPM3064A, EPM3128A, and EPM3256A devices have six output enables. EPM3512A devices have 10 output enables.

When the tri–state buffer control is connected to ground, the output is tri-stated (high impedance), and the I/O pin can be used as a dedicated input. When the tri–state buffer control is connected to V_{CC}, the output is enabled.

The MAX 3000A architecture provides dual I/O feedback, in which macrocell and pin feedbacks are independent. When an I/O pin is configured as an input, the associated macrocell can be used for buried logic.

In–System Programmability

MAX 3000A devices can be programmed in–system via an industry–standard four–pin IEEE Std. 1149.1-1990 (JTAG) interface. In-system programmability (ISP) offers quick, efficient iterations during design development and debugging cycles. The MAX 3000A architecture internally generates the high programming voltages required to program its EEPROM cells, allowing in–system programming with only a single 3.3–V power supply. During in–system programming, the I/O pins are tri–stated and weakly pulled–up to eliminate board conflicts. The pull–up value is nominally 50 kΩ.

MAX 3000A devices have an enhanced ISP algorithm for faster programming. These devices also offer an `ISP_Done` bit that ensures safe operation when in–system programming is interrupted. This `ISP_Done` bit, which is the last bit programmed, prevents all I/O pins from driving until the bit is programmed.

ISP simplifies the manufacturing flow by allowing devices to be mounted on a printed circuit board (PCB) with standard pick–and–place equipment before they are programmed. MAX 3000A devices can be programmed by downloading the information via in–circuit testers, embedded processors, the MasterBlaster communications cable, the ByteBlasterMV parallel port download cable, and the BitBlaster serial download cable. Programming the devices after they are placed on the board eliminates lead damage on high–pin–count packages (e.g., QFP packages) due to device handling. MAX 3000A devices can be reprogrammed after a system has already shipped to the field. For example, product upgrades can be performed in the field via software or modem.

The Jam STAPL programming and test language can be used to program MAX 3000A devices with in–circuit testers, PCs, or embedded processors.

For more information on using the Jam STAPL programming and test language, see *Application Note 88 (Using the Jam Language for ISP & ICR via an Embedded Processor)*, *Application Note 122 (Using Jam STAPL for ISP & ICR via an Embedded Processor)* and *AN 111 (Embedded Programming Using the 8051 and Jam Byte-Code)*.

The ISP circuitry in MAX 3000A devices is compliant with the IEEE Std. 1532 specification. The IEEE Std. 1532 is a standard developed to allow concurrent ISP between multiple PLD vendors.

By combining the pulse and shift times for each of the programming stages, the program or verify time can be derived as a function of the TCK frequency, the number of devices, and specific target device(s). Because different ISP-capable devices have a different number of EEPROM cells, both the total fixed and total variable times are unique for a single device.

Programming a Single MAX 3000A Device

The time required to program a single MAX 3000A device in-system can be calculated from the following formula:

$$t_{PROG} = t_{PPULSE} + \frac{Cycle_{PTCK}}{f_{TCK}}$$

where: t_{PROG} = Programming time
t_{PPULSE} = Sum of the fixed times to erase, program, and verify the EEPROM cells
$Cycle_{PTCK}$ = Number of TCK cycles to program a device
f_{TCK} = TCK frequency

The ISP times for a stand-alone verification of a single MAX 3000A device can be calculated from the following formula:

$$t_{VER} = t_{VPULSE} + \frac{Cycle_{VTCK}}{f_{TCK}}$$

where: t_{VER} = Verify time
t_{VPULSE} = Sum of the fixed times to verify the EEPROM cells
$Cycle_{VTCK}$ = Number of TCK cycles to verify a device

The programming times described in Tables 4 through 6 are associated with the worst-case method using the enhanced ISP algorithm.

Table 4. MAX 3000A t_{PULSE} & $Cycle_{TCK}$ Values

Device	Programming		Stand-Alone Verification	
	t_{PPULSE} (s)	$Cycle_{PTCK}$	t_{VPULSE} (s)	$Cycle_{VTCK}$
EPM3032A	2.00	55,000	0.002	18,000
EPM3064A	2.00	105,000	0.002	35,000
EPM3128A	2.00	205,000	0.002	68,000
EPM3256A	2.00	447,000	0.002	149,000
EPM3512A	2.00	890,000	0.002	297,000

Tables 5 and 6 show the in-system programming and stand alone verification times for several common test clock frequencies.

Table 5. MAX 3000A In-System Programming Times for Different Test Clock Frequencies

Device	f_{TCK}								Units
	10 MHz	5 MHz	2 MHz	1 MHz	500 kHz	200 kHz	100 kHz	50 kHz	
EPM3032A	2.01	2.01	2.03	2.06	2.11	2.28	2.55	3.10	s
EPM3064A	2.01	2.02	2.05	2.11	2.21	2.53	3.05	4.10	s
EPM3128A	2.02	2.04	2.10	2.21	2.41	3.03	4.05	6.10	s
EPM3256A	2.05	2.09	2.23	2.45	2.90	4.24	6.47	10.94	s
EPM3512A	2.09	2.18	2.45	2.89	3.78	6.45	10.90	19.80	s

Table 6. MAX 3000A Stand-Alone Verification Times for Different Test Clock Frequencies

Device	f_{TCK}								Units
	10 MHz	5 MHz	2 MHz	1 MHz	500 kHz	200 kHz	100 kHz	50 kHz	
EPM3032A	0.00	0.01	0.01	0.02	0.04	0.09	0.18	0.36	s
EPM3064A	0.01	0.01	0.02	0.04	0.07	0.18	0.35	0.70	s
EPM3128A	0.01	0.02	0.04	0.07	0.14	0.34	0.68	1.36	s
EPM3256A	0.02	0.03	0.08	0.15	0.30	0.75	1.49	2.98	s
EPM3512A	0.03	0.06	0.15	0.30	0.60	1.49	2.97	5.94	s

Programming with External Hardware

MAX 3000A devices can be programmed on Windows–based PCs with an Altera Logic Programmer card, MPU, and the appropriate device adapter. The MPU performs continuity checking to ensure adequate electrical contact between the adapter and the device.

For more information, see the *Altera Programming Hardware Data Sheet*.

The Altera software can use text– or waveform–format test vectors created with the Altera Text Editor or Waveform Editor to test the programmed device. For added design verification, designers can perform functional testing to compare the functional device behavior with the results of simulation.

Data I/O, BP Microsystems, and other programming hardware manufacturers also provide programming support for Altera devices.

For more information, see *Programming Hardware Manufacturers*.

IEEE Std. 1149.1 (JTAG) Boundary–Scan Support

MAX 3000A devices include the JTAG BST circuitry defined by IEEE Std. 1149.1–1990. Table 7 describes the JTAG instructions supported by MAX 3000A devices. The pin-out tables found on the Altera web site (**http://www.altera.com**) or the *Altera Digital Library* show the location of the JTAG control pins for each device. If the JTAG interface is not required, the JTAG pins are available as user I/O pins.

Table 7. MAX 3000A JTAG Instructions

JTAG Instruction	Description
SAMPLE/PRELOAD	Allows a snapshot of signals at the device pins to be captured and examined during normal device operation, and permits an initial data pattern output at the device pins
EXTEST	Allows the external circuitry and board–level interconnections to be tested by forcing a test pattern at the output pins and capturing test results at the input pins
BYPASS	Places the 1–bit bypass register between the `TDI` and `TDO` pins, which allows the BST data to pass synchronously through a selected device to adjacent devices during normal device operation
IDCODE	Selects the IDCODE register and places it between the `TDI` and `TDO` pins, allowing the IDCODE to be serially shifted out of `TDO`
USERCODE	Selects the 32–bit USERCODE register and places it between the `TDI` and `TDO` pins, allowing the USERCODE value to be shifted out of `TDO`
ISP Instructions	These instructions are used when programming MAX 3000A devices via the JTAG ports with the MasterBlaster, ByteBlasterMV, or BitBlaster cable, or when using a Jam STAPL file, JBC file, or SVF file via an embedded processor or test equipment

The instruction register length of MAX 3000A devices is 10 bits. The IDCODE and USERCODE register length is 32 bits. Tables 8 and 9 show the boundary–scan register length and device IDCODE information for MAX 3000A devices.

Table 8. MAX 3000A Boundary–Scan Register Length

Device	Boundary–Scan Register Length
EPM3032A	96
EPM3064A	192
EPM3128A	288
EPM3256A	480
EPM3512A	624

Table 9. 32–Bit MAX 3000A Device IDCODE Value *Note (1)*

Device	IDCODE (32 bits)			
	Version (4 Bits)	Part Number (16 Bits)	Manufacturer's Identity (11 Bits)	1 (1 Bit) *(2)*
EPM3032A	0001	0111 0000 0011 0010	00001101110	1
EPM3064A	0001	0111 0000 0110 0100	00001101110	1
EPM3128A	0001	0111 0001 0010 1000	00001101110	1
EPM3256A	0001	0111 0010 0101 0110	00001101110	1
EPM3512A	0001	0111 0101 0001 0010	00001101110	1

Notes:
(1) The most significant bit (MSB) is on the left.
(2) The least significant bit (LSB) for all JTAG IDCODEs is 1.

See *Application Note 39 (IEEE 1149.1 (JTAG) Boundary–Scan Testing in Altera Devices)* for more information on JTAG BST.

Programmable Speed/Power Control

MAX 3000A devices offer a power–saving mode that supports low-power operation across user–defined signal paths or the entire device. This feature allows total power dissipation to be reduced by 50% or more because most logic applications require only a small fraction of all gates to operate at maximum frequency.

The designer can program each individual macrocell in a MAX 3000A device for either high–speed or low–power operation. As a result, speed-critical paths in the design can run at high speed, while the remaining paths can operate at reduced power. Macrocells that run at low power incur a nominal timing delay adder (t_{LPA}) for the t_{LAD}, t_{LAC}, t_{IC}, t_{ACL}, t_{EN}, t_{CPPW} and t_{SEXP} parameters.

Output Configuration

MAX 3000A device outputs can be programmed to meet a variety of system–level requirements.

MultiVolt I/O Interface

The MAX 3000A device architecture supports the MultiVolt I/O interface feature, which allows MAX 3000A devices to connect to systems with differing supply voltages. MAX 3000A devices in all packages can be set for 2.5–V, 3.3–V, or 5.0–V I/O pin operation. These devices have one set of V_{CC} pins for internal operation and input buffers (VCCINT), and another set for I/O output drivers (VCCIO).

The VCCIO pins can be connected to either a 3.3–V or 2.5–V power supply, depending on the output requirements. When the VCCIO pins are connected to a 2.5–V power supply, the output levels are compatible with 2.5–V systems. When the VCCIO pins are connected to a 3.3–V power supply, the output high is at 3.3 V and is therefore compatible with 3.3-V or 5.0–V systems. Devices operating with V_{CCIO} levels lower than 3.0 V incur a nominally greater timing delay of t_{OD2} instead of t_{OD1}. Inputs can always be driven by 2.5–V, 3.3–V, or 5.0–V signals.

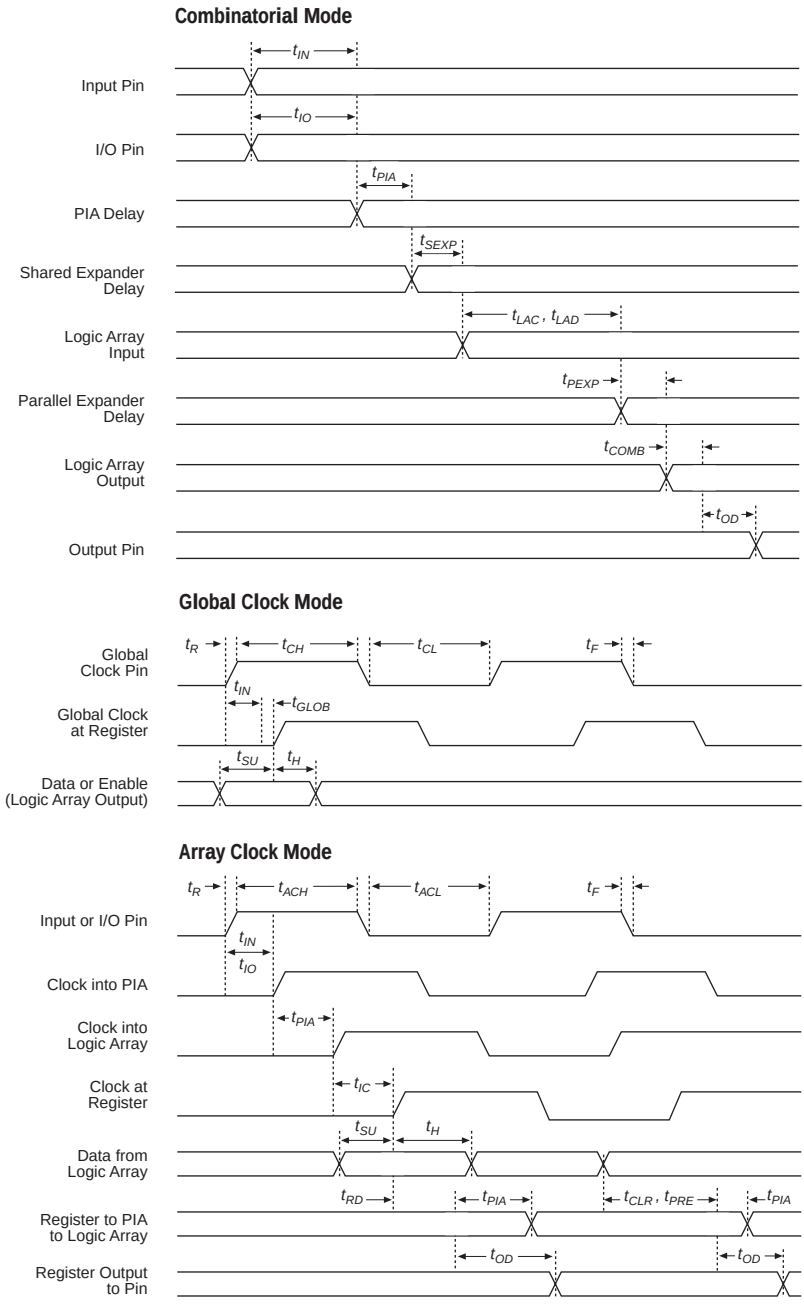
Table 11 summarizes the MAX 3000A MultiVolt I/O support.

Table 11. MAX 3000A MultiVolt I/O Support

V_{CCIO} Voltage	Input Signal (V)			Output Signal (V)		
	2.5	3.3	5.0	2.5	3.3	5.0
2.5	✓	✓	✓	✓		
3.3	✓	✓	✓	✓	✓	✓

Figure 11. MAX 3000A Switching Waveforms

t_R & t_F < 2 ns. Inputs are driven at 3 V for a logic high and 0 V for a logic low. All timing characteristics are measured at 1.5 V.



Tables 16 through 23 show EPM3032A, EPM3064A, EPM3128A, EPM3256A, and EPM3512A timing information.

Table 16. EPM3032A External Timing Parameters *Note (1)*

Symbol	Parameter	Conditions	Speed Grade						Unit
			–4		–7		–10		
			Min	Max	Min	Max	Min	Max	
t_{PD1}	Input to non–registered output	C1 = 35 pF *(2)*		4.5		7.5		10	ns
t_{PD2}	I/O input to non–registered output	C1 = 35 pF *(2)*		4.5		7.5		10	ns
t_{SU}	Global clock setup time	*(2)*	2.9		4.7		6.3		ns
t_{H}	Global clock hold time	*(2)*	0.0		0.0		0.0		ns
t_{CO1}	Global clock to output delay	C1 = 35 pF	1.0	3.0	1.0	5.0	1.0	6.7	ns
t_{CH}	Global clock high time		2.0		3.0		4.0		ns
t_{CL}	Global clock low time		2.0		3.0		4.0		ns
t_{ASU}	Array clock setup time	*(2)*	1.6		2.5		3.6		ns
t_{AH}	Array clock hold time	*(2)*	0.3		0.5		0.5		ns
t_{ACO1}	Array clock to output delay	C1 = 35 pF *(2)*	1.0	4.3	1.0	7.2	1.0	9.4	ns
t_{ACH}	Array clock high time		2.0		3.0		4.0		ns
t_{ACL}	Array clock low time		2.0		3.0		4.0		ns
t_{CPPW}	Minimum pulse width for clear and preset	*(3)*	2.0		3.0		4.0		ns
t_{CNT}	Minimum global clock period	*(2)*		4.4		7.2		9.7	ns
f_{CNT}	Maximum internal global clock frequency	*(2)*, *(4)*	227.3		138.9		103.1		MHz
t_{ACNT}	Minimum array clock period	*(2)*		4.4		7.2		9.7	ns
f_{ACNT}	Maximum internal array clock frequency	*(2)*, *(4)*	227.3		138.9		103.1		MHz

Table 17. EPM3032A Internal Timing Parameters (Part 1 of 2) *Note (1)*

Symbol	Parameter	Conditions	Speed Grade						Unit
			–4		–7		–10		
			Min	Max	Min	Max	Min	Max	
t_{IN}	Input pad and buffer delay			0.7		1.2		1.5	ns
t_{IO}	I/O input pad and buffer delay			0.7		1.2		1.5	ns
t_{SEXP}	Shared expander delay			1.9		3.1		4.0	ns
t_{PEXP}	Parallel expander delay			0.5		0.8		1.0	ns
t_{LAD}	Logic array delay			1.5		2.5		3.3	ns
t_{LAC}	Logic control array delay			0.6		1.0		1.2	ns
t_{IOE}	Internal output enable delay			0.0		0.0		0.0	ns
t_{OD1}	Output buffer and pad delay, slow slew rate = off V_{CCIO} = 3.3 V	C1 = 35 pF		0.8		1.3		1.8	ns
t_{OD2}	Output buffer and pad delay, slow slew rate = off V_{CCIO} = 2.5 V	C1 = 35 pF		1.3		1.8		2.3	ns
t_{OD3}	Output buffer and pad delay, slow slew rate = on V_{CCIO} = 2.5 V or 3.3 V	C1 = 35 pF		5.8		6.3		6.8	ns
t_{ZX1}	Output buffer enable delay, slow slew rate = off V_{CCIO} = 3.3 V	C1 = 35 pF		4.0		4.0		5.0	ns
t_{ZX2}	Output buffer enable delay, slow slew rate = off V_{CCIO} = 2.5 V	C1 = 35 pF		4.5		4.5		5.5	ns
t_{ZX3}	Output buffer enable delay, slow slew rate = on V_{CCIO} = 2.5 V or 3.3 V	C1 = 35 pF		9.0		9.0		10.0	ns
t_{XZ}	Output buffer disable delay	C1 = 5 pF		4.0		4.0		5.0	ns
t_{SU}	Register setup time		1.3		2.0		2.8		ns
t_{H}	Register hold time		0.6		1.0		1.3		ns
t_{RD}	Register delay			0.7		1.2		1.5	ns
t_{COMB}	Combinatorial delay			0.6		1.0		1.3	ns
t_{IC}	Array clock delay			1.2		2.0		2.5	ns
t_{EN}	Register enable time			0.6		1.0		1.2	ns
t_{GLOB}	Global control delay			0.8		1.3		1.9	ns
t_{PRE}	Register preset time			1.2		1.9		2.6	ns
t_{CLR}	Register clear time			1.2		1.9		2.6	ns
t_{PIA}	PIA delay	*(2)*		0.9		1.5		2.1	ns

Table 21. EPM3128A Internal Timing Parameters (Part 2 of 2) *Note (1)*

Symbol	Parameter	Conditions	Speed Grade						Unit
			–5		–7		–10		
			Min	Max	Min	Max	Min	Max	
t_{SU}	Register setup time		1.4		2.1		2.9		ns
t_{H}	Register hold time		0.6		1.0		1.3		ns
t_{RD}	Register delay			0.8		1.2		1.6	ns
t_{COMB}	Combinatorial delay			0.5		0.9		1.3	ns
t_{IC}	Array clock delay			1.2		1.7		2.2	ns
t_{EN}	Register enable time			0.7		1.0		1.3	ns
t_{GLOB}	Global control delay			1.1		1.6		2.0	ns
t_{PRE}	Register preset time			1.4		2.0		2.7	ns
t_{CLR}	Register clear time			1.4		2.0		2.7	ns
t_{PIA}	PIA delay	*(2)*		1.4		2.0		2.6	ns
t_{LPA}	Low–power adder	*(5)*		4.0		4.0		5.0	ns

Table 22. EPM3256A External Timing Parameters *Note (1)*

Symbol	Parameter	Conditions	Speed Grade				Unit
			–7		–10		
			Min	Max	Min	Max	
t_{PD1}	Input to non–registered output	C1 = 35 pF *(2)*		7.5		10	ns
t_{PD2}	I/O input to non–registered output	C1 = 35 pF *(2)*		7.5		10	ns
t_{SU}	Global clock setup time	*(2)*	5.2		6.9		ns
t_{H}	Global clock hold time	*(2)*	0.0		0.0		ns
t_{CO1}	Global clock to output delay	C1 = 35 pF	1.0	4.8	1.0	6.4	ns
t_{CH}	Global clock high time		3.0		4.0		ns
t_{CL}	Global clock low time		3.0		4.0		ns
t_{ASU}	Array clock setup time	*(2)*	2.7		3.6		ns
t_{AH}	Array clock hold time	*(2)*	0.3		0.5		ns
t_{ACO1}	Array clock to output delay	C1 = 35 pF *(2)*	1.0	7.3	1.0	9.7	ns
t_{ACH}	Array clock high time		3.0		4.0		ns
t_{ACL}	Array clock low time		3.0		4.0		ns
t_{CPPW}	Minimum pulse width for clear and preset	*(3)*	3.0		4.0		ns

Table 22. EPM3256A External Timing Parameters *Note (1)*

Symbol	Parameter	Conditions	Speed Grade				Unit
			–7		–10		
			Min	Max	Min	Max	
t_{CNT}	Minimum global clock period	*(2)*		7.9		10.5	ns
f_{CNT}	Maximum internal global clock frequency	*(2)*, *(4)*	126.6		95.2		MHz
t_{ACNT}	Minimum array clock period	*(2)*		7.9		10.5	ns
f_{ACNT}	Maximum internal array clock frequency	*(2)*, *(4)*	126.6		95.2		MHz

Table 23. EPM3256A Internal Timing Parameters (Part 1 of 2) *Note (1)*

Symbol	Parameter	Conditions	Speed Grade				Unit
			–7		–10		
			Min	Max	Min	Max	
t_{IN}	Input pad and buffer delay			0.9		1.2	ns
t_{IO}	I/O input pad and buffer delay			0.9		1.2	ns
t_{SEXP}	Shared expander delay			2.8		3.7	ns
t_{PEXP}	Parallel expander delay			0.5		0.6	ns
t_{LAD}	Logic array delay			2.2		2.8	ns
t_{LAC}	Logic control array delay			1.0		1.3	ns
t_{IOE}	Internal output enable delay			0.0		0.0	ns
t_{OD1}	Output buffer and pad delay, slow slew rate = off V_{CCIO} = 3.3 V	C1 = 35 pF		1.2		1.6	ns
t_{OD2}	Output buffer and pad delay, slow slew rate = off V_{CCIO} = 2.5 V	C1 = 35 pF		1.7		2.1	ns
t_{OD3}	Output buffer and pad delay, slow slew rate = on V_{CCIO} = 2.5 V or 3.3 V	C1 = 35 pF		6.2		6.6	ns
t_{ZX1}	Output buffer enable delay, slow slew rate = off V_{CCIO} = 3.3 V	C1 = 35 pF		4.0		5.0	ns
t_{ZX2}	Output buffer enable delay, slow slew rate = off V_{CCIO} = 2.5 V	C1 = 35 pF		4.5		5.5	ns

Table 24. EPM3512A External Timing Parameters *Note (1)*

Symbol	Parameter	Conditions	Speed Grade				Unit
			-7		-10		
			Min	Max	Min	Max	
t_{ACO1}	Array clock to output delay	C1 = 35 pF *(2)*	1.0	7.8	1.0	10.4	ns
t_{ACH}	Array clock high time		3.0		4.0		ns
t_{ACL}	Array clock low time		3.0		4.0		ns
t_{CPPW}	Minimum pulse width for clear and preset	*(3)*	3.0		4.0		ns
t_{CNT}	Minimum global clock period	*(2)*		8.6		11.5	ns
f_{CNT}	Maximum internal global clock frequency	*(2)*, *(4)*	116.3		87.0		MHz
t_{ACNT}	Minimum array clock period	*(2)*		8.6		11.5	ns
f_{ACNT}	Maximum internal array clock frequency	*(2)*, *(4)*	116.3		87.0		MHz

Table 25. EPM3512A Internal Timing Parameters (Part 1 of 2) *Note (1)*

Symbol	Parameter	Conditions	Speed Grade				Unit
			-7		-10		
			Min	Max	Min	Max	
t_{IN}	Input pad and buffer delay			0.7		0.9	ns
t_{IO}	I/O input pad and buffer delay			0.7		0.9	ns
t_{FIN}	Fast input delay			3.1		3.6	ns
t_{SEXP}	Shared expander delay			2.7		3.5	ns
t_{PEXP}	Parallel expander delay			0.4		0.5	ns
t_{LAD}	Logic array delay			2.2		2.8	ns
t_{LAC}	Logic control array delay			1.0		1.3	ns
t_{IOE}	Internal output enable delay			0.0		0.0	ns
t_{OD1}	Output buffer and pad delay, slow slew rate = off V_{CCIO} = 3.3 V	C1 = 35 pF		1.0		1.5	ns
t_{OD2}	Output buffer and pad delay, slow slew rate = off V_{CCIO} = 2.5 V	C1 = 35 pF		1.5		2.0	ns
t_{OD3}	Output buffer and pad delay, slow slew rate = on V_{CCIO} = 2.5 V or 3.3 V	C1 = 35 pF		6.0		6.5	ns

Table 25. EPM3512A Internal Timing Parameters (Part 2 of 2) *Note (1)*

Symbol	Parameter	Conditions	Speed Grade				Unit
			-7		-10		
			Min	Max	Min	Max	
t_{ZX1}	Output buffer enable delay, slow slew rate = off V_{CCIO} = 3.3 V	C1 = 35 pF		4.0		5.0	ns
t_{ZX2}	Output buffer enable delay, slow slew rate = off V_{CCIO} = 2.5 V	C1 = 35 pF		4.5		5.5	ns
t_{ZX3}	Output buffer enable delay, slow slew rate = on V_{CCIO} = 3.3 V	C1 = 35 pF		9.0		10.0	ns
t_{XZ}	Output buffer disable delay	C1 = 5 pF		4.0		5.0	ns
t_{SU}	Register setup time		2.1		3.0		ns
t_{H}	Register hold time		0.6		0.8		ns
t_{FSU}	Register setup time of fast input		1.6		1.6		ns
t_{FH}	Register hold time of fast input		1.4		1.4		ns
t_{RD}	Register delay			1.3		1.7	ns
t_{COMB}	Combinatorial delay			0.6		0.8	ns
t_{IC}	Array clock delay			1.8		2.3	ns
t_{EN}	Register enable time			1.0		1.3	ns
t_{GLOB}	Global control delay			1.7		2.2	ns
t_{PRE}	Register preset time			1.0		1.4	ns
t_{CLR}	Register clear time			1.0		1.4	ns
t_{PIA}	PIA delay	*(2)*		3.0		4.0	ns
t_{LPA}	Low-power adder	*(5)*		4.5		5.0	ns

Notes to tables:

(1) These values are specified under the recommended operating conditions, as shown in Table 13 on page 22. See Figure 11 on page 26 for more information on switching waveforms.

(2) These values are specified for a PIA fan–out of one LAB (16 macrocells). For each additional LAB fan–out in these devices, add an additional 0.1 ns to the PIA timing value.

(3) This minimum pulse width for preset and clear applies for both global clear and array controls. The t_{LPA} parameter must be added to this minimum width if the clear or reset signal incorporates the t_{LAD} parameter into the signal path.

(4) These parameters are measured with a 16–bit loadable, enabled, up / down counter programmed into each LAB.

(5) The t_{LPA} parameter must be added to the t_{LAD}, t_{LAC}, t_{IC}, t_{EN}, t_{SEXP}, $\mathbf{t_{ACL}}$, and $\mathbf{t_{CPPW}}$ parameters for macrocells running in low–power mode.

Power Consumption

Supply power (P) versus frequency ($\mathbf{f_{MAX}}$, in MHz) for MAX 3000A devices is calculated with the following equation:

$$P = P_{INT} + P_{IO} = I_{CCINT} \times V_{CC} + P_{IO}$$

The P_{IO} value, which depends on the device output load characteristics and switching frequency, can be calculated using the guidelines given in *Application Note 74 (Evaluating Power for Altera Devices)*.

The I_{CCINT} value depends on the switching frequency and the application logic. The I_{CCINT} value is calculated with the following equation:

$$I_{CCINT} =$$

$$(A \times MC_{TON}) + [B \times (MC_{DEV} - MC_{TON})] + (C \times MC_{USED} \times \mathbf{f_{MAX}} \times \mathbf{tog_{LC}})$$

The parameters in the I_{CCINT} equation are:

MC_{TON}	=	Number of macrocells with the Turbo Bit™ option turned on, as reported in the Quartus II or MAX+PLUS II Report File (**.rpt**)
MC_{DEV}	=	Number of macrocells in the device
MC_{USED}	=	Total number of macrocells in the design, as reported in the RPT File
$\mathbf{f_{MAX}}$	=	Highest clock frequency to the device
$\mathbf{tog_{LC}}$	=	Average percentage of logic cells toggling at each clock (typically 12.5%)
A, B, C	=	Constants (shown in Table 26)

Table 26. MAX 3000A I_{CC} Equation Constants

Device	A	B	C
EPM3032A	0.71	0.30	0.014
EPM3064A	0.71	0.30	0.014
EPM3128A	0.71	0.30	0.014
EPM3256A	0.71	0.30	0.014
EPM3512A	0.71	0.30	0.014

The I_{CCINT} calculation provides an I_{CC} estimate based on typical conditions using a pattern of a 16–bit, loadable, enabled, up / down counter in each LAB with no output load. Actual I_{CC} should be verified during operation because this measurement is sensitive to the actual pattern in the device and the environmental operating conditions.

Figures 12 and 13 show the typical supply current versus frequency for MAX 3000A devices.

Figure 15. 100–Pin TQFP Package Pin–Out Diagram

Package outline not drawn to scale.

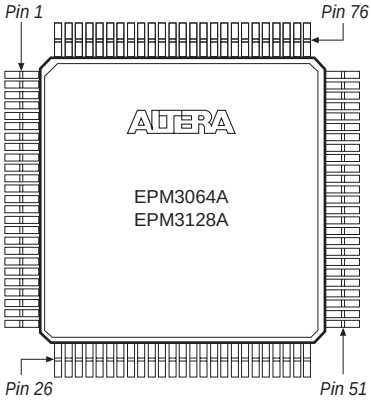
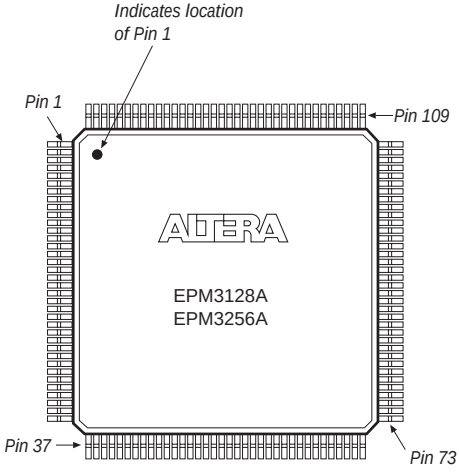


Figure 16. 144–Pin TQFP Package Pin–Out Diagram

Package outline not drawn to scale.



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