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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	1206
Number of Logic Elements/Cells	12060
Total RAM Bits	239616
Number of I/O	173
Number of Gates	-
Voltage - Supply	1.425V ~ 1.575V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	240-BFQFP
Supplier Device Package	240-PQFP (32x32)
Purchase URL	https://www.e-xfl.com/product-detail/intel/ep1c12q240c8n

to the appropriate plane on the board. The Quartus II software reserves I/O pins as power pins as necessary for layout with the larger densities in the same package having more power pins.

Table 1–3. Cyclone QFP and FineLine BGA Package Sizes

Dimension	100-Pin TQFP	144-Pin TQFP	240-Pin PQFP	256-Pin FineLine BGA	324-Pin FineLine BGA	400-Pin FineLine BGA
Pitch (mm)	0.5	0.5	0.5	1.0	1.0	1.0
Area (mm^2)	256	484	1,024	289	361	441
Length × width (mm × mm)	16×16	22×22	34.6×34.6	17×17	19×19	21×21

Document Revision History

Table 1–4 shows the revision history for this document.

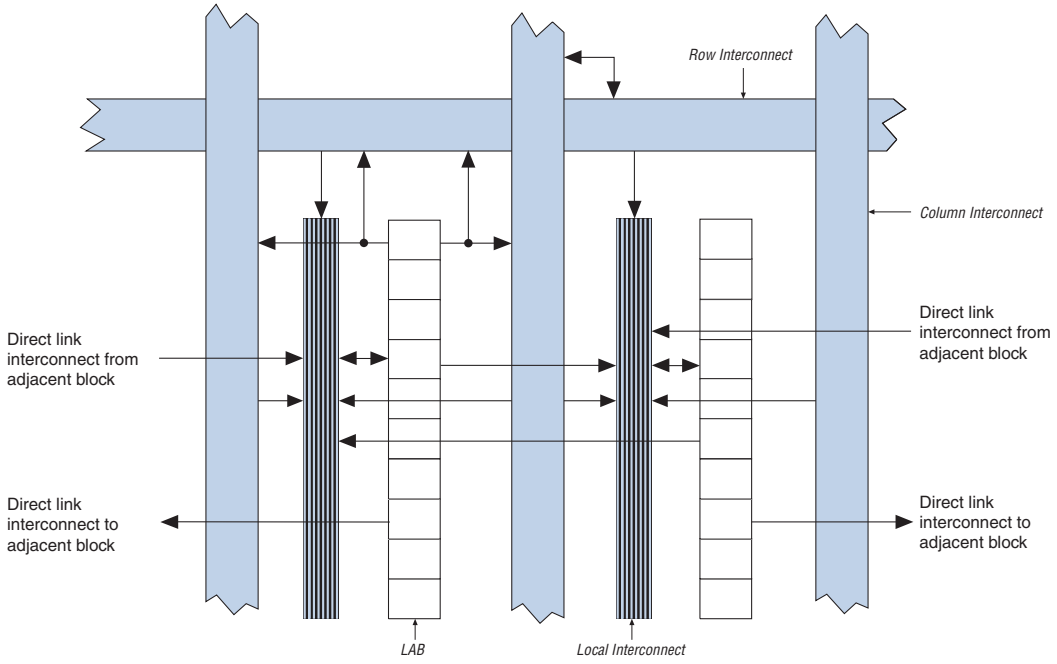
Table 1–4. Document Revision History

Date and Document Version	Changes Made	Summary of Changes
May 2008 v1.5	Minor textual and style changes.	—
January 2007 v1.4	Added document revision history.	—
August 2005 v1.3	Minor updates.	—
October 2003 v1.2	Added 64-bit PCI support information.	—
September 2003 v1.1	• Updated LVDS data rates to 640 Mbps from 311 Mbps. • Updated RSDS feature information.	—
May 2003 v1.0	Added document to Cyclone Device Handbook.	—

Logic Array Blocks

Each LAB consists of 10 LEs, LE carry chains, LAB control signals, a local interconnect, look-up table (LUT) chain, and register chain connection lines. The local interconnect transfers signals between LEs in the same LAB. LUT chain connections transfer the output of one LE's LUT to the adjacent LE for fast sequential LUT connections within the same LAB. Register chain connections transfer the output of one LE's register to the adjacent LE's register within a LAB. The Quartus® II Compiler places associated logic within a LAB or adjacent LABs, allowing the use of local, LUT chain, and register chain connections for performance and area efficiency. Figure 2–2 details the Cyclone LAB.

Figure 2–2. Cyclone LAB Structure



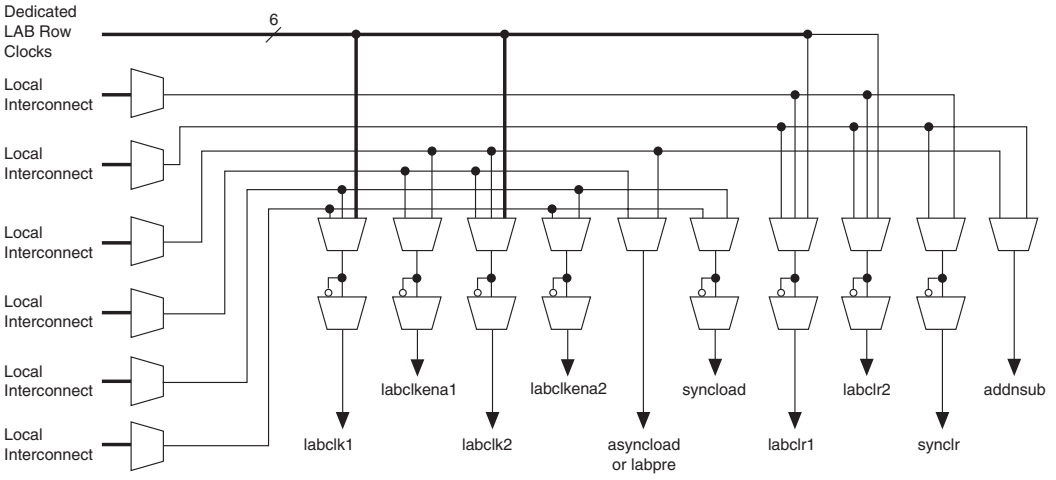
LAB Interconnects

The LAB local interconnect can drive LEs within the same LAB. The LAB local interconnect is driven by column and row interconnects and LE outputs within the same LAB. Neighboring LABs, PLLs, and M4K RAM blocks from the left and right can also drive a LAB's local interconnect through the direct link connection. The direct link connection feature minimizes the use of row and column interconnects, providing higher

With the LAB-wide `addnsub` control signal, a single LE can implement a one-bit adder and subtractor. This saves LE resources and improves performance for logic functions such as DSP correlators and signed multipliers that alternate between addition and subtraction depending on data.

The LAB row clocks `[5..0]` and LAB local interconnect generate the LAB-wide control signals. The MultiTrack™ interconnect's inherent low skew allows clock and control signal distribution in addition to data. Figure 2–4 shows the LAB control signal generation circuit.

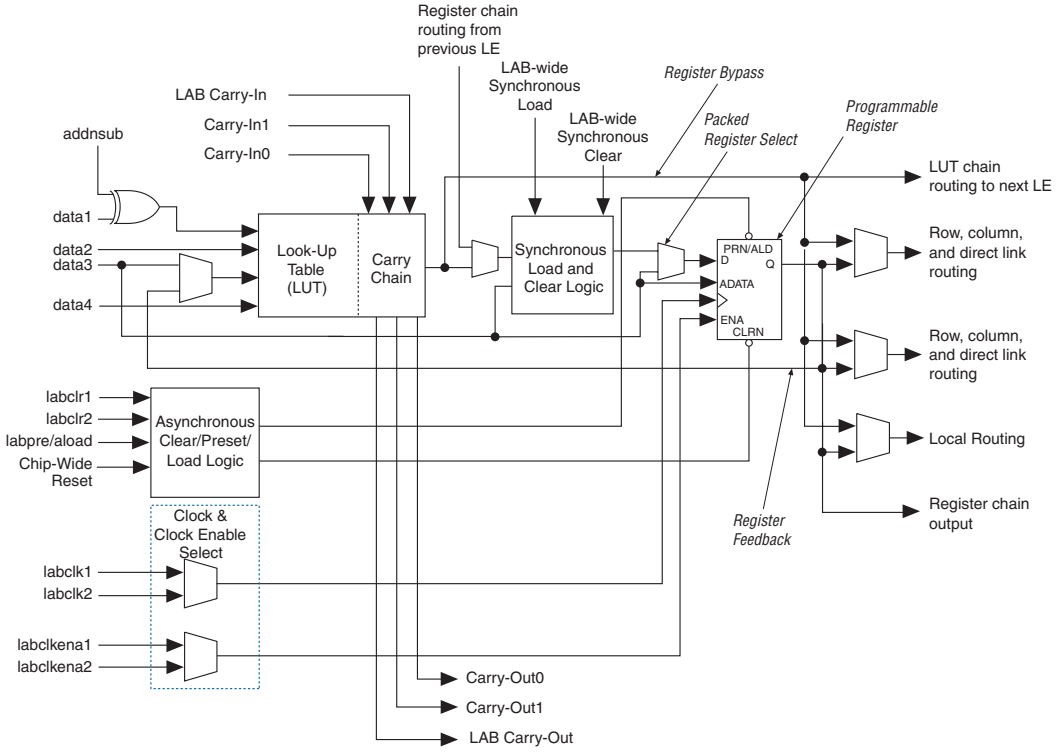
Figure 2–4. LAB-Wide Control Signals



Logic Elements

The smallest unit of logic in the Cyclone architecture, the LE, is compact and provides advanced features with efficient logic utilization. Each LE contains a four-input LUT, which is a function generator that can implement any function of four variables. In addition, each LE contains a programmable register and carry chain with carry select capability. A single LE also supports dynamic single bit addition or subtraction mode selectable by a LAB-wide control signal. Each LE drives all types of interconnects: local, row, column, LUT chain, register chain, and direct link interconnects. See Figure 2–5.

Figure 2–5. Cyclone LE



Each LE's programmable register can be configured for D, T, JK, or SR operation. Each register has data, true asynchronous load data, clock, clock enable, clear, and asynchronous load/preset inputs. Global signals, general-purpose I/O pins, or any internal logic can drive the register's clock and clear control signals. Either general-purpose I/O pins or internal logic can drive the clock enable, preset, asynchronous load, and asynchronous data. The asynchronous load data input comes from the `data3` input of the LE. For combinatorial functions, the LUT output bypasses the register and drives directly to the LE outputs.

Each LE has three outputs that drive the local, row, and column routing resources. The LUT or register output can drive these three outputs independently. Two LE outputs drive column or row and direct link routing connections and one drives local interconnect resources. This allows the LUT to drive one output while the register drives another output. This feature, called register packing, improves device utilization because the device can use the register and the LUT for unrelated

functions. Another special packing mode allows the register output to feed back into the LUT of the same LE so that the register is packed with its own fan-out LUT. This provides another mechanism for improved fitting. The LE can also drive out registered and unregistered versions of the LUT output.

LUT Chain and Register Chain

In addition to the three general routing outputs, the LEs within a LAB have LUT chain and register chain outputs. LUT chain connections allow LUTs within the same LAB to cascade together for wide input functions. Register chain outputs allow registers within the same LAB to cascade together. The register chain output allows a LAB to use LUTs for a single combinatorial function and the registers to be used for an unrelated shift register implementation. These resources speed up connections between LABs while saving local interconnect resources. "MultiTrack Interconnect" on page 2–12 for more information on LUT chain and register chain connections.

addnsub Signal

The LE's dynamic adder/subtractor feature saves logic resources by using one set of LEs to implement both an adder and a subtractor. This feature is controlled by the LAB-wide control signal `addnsub`. The `addnsub` signal sets the LAB to perform either A + B or A –B. The LUT computes addition; subtraction is computed by adding the two's complement of the intended subtractor. The LAB-wide signal converts to two's complement by inverting the B bits within the LAB and setting carry-in = 1 to add one to the least significant bit (LSB). The LSB of an adder/subtractor must be placed in the first LE of the LAB, where the LAB-wide `addnsub` signal automatically sets the carry-in to 1. The Quartus II Compiler automatically places and uses the adder/subtractor feature when using adder/subtractor parameterized functions.

LE Operating Modes

The Cyclone LE can operate in one of the following modes:

- Normal mode
- Dynamic arithmetic mode

Each mode uses LE resources differently. In each mode, eight available inputs to the LE—the four data inputs from the LAB local interconnect, `carry-in0` and `carry-in1` from the previous LE, the LAB carry-in from the previous carry-chain LAB, and the register chain connection—are directed to different destinations to implement the desired logic function. LAB-wide signals provide clock, asynchronous clear, asynchronous

The Quartus II Compiler automatically creates carry chain logic during design processing, or you can create it manually during design entry. Parameterized functions such as LPM functions automatically take advantage of carry chains for the appropriate functions.

The Quartus II Compiler creates carry chains longer than 10 LEs by linking LABs together automatically. For enhanced fitting, a long carry chain runs vertically allowing fast horizontal connections to M4K memory blocks. A carry chain can continue as far as a full column.

Clear and Preset Logic Control

LAB-wide signals control the logic for the register's clear and preset signals. The LE directly supports an asynchronous clear and preset function. The register preset is achieved through the asynchronous load of a logic high. The direct asynchronous preset does not require a NOT-gate push-back technique. Cyclone devices support simultaneous preset/ asynchronous load and clear signals. An asynchronous clear signal takes precedence if both signals are asserted simultaneously. Each LAB supports up to two clears and one preset signal.

In addition to the clear and preset ports, Cyclone devices provide a chip-wide reset pin (`DEV_CLRn`) that resets all registers in the device. An option set before compilation in the Quartus II software controls this pin. This chip-wide reset overrides all other control signals.

MultiTrack Interconnect

In the Cyclone architecture, connections between LEs, M4K memory blocks, and device I/O pins are provided by the MultiTrack interconnect structure with DirectDrive™ technology. The MultiTrack interconnect consists of continuous, performance-optimized routing lines of different speeds used for inter- and intra-design block connectivity. The Quartus II Compiler automatically places critical design paths on faster interconnects to improve design performance.

DirectDrive technology is a deterministic routing technology that ensures identical routing resource usage for any function regardless of placement within the device. The MultiTrack interconnect and DirectDrive technology simplify the integration stage of block-based designing by eliminating the re-optimization cycles that typically follow design changes and additions.

The MultiTrack interconnect consists of row and column interconnects that span fixed distances. A routing structure with fixed length resources for all devices allows predictable and repeatable performance when

Embedded Memory

The Cyclone embedded memory consists of columns of M4K memory blocks. EP1C3 and EP1C6 devices have one column of M4K blocks, while EP1C12 and EP1C20 devices have two columns (refer to Table 1–1 on page 1–1 for total RAM bits per density). Each M4K block can implement various types of memory with or without parity, including true dual-port, simple dual-port, and single-port RAM, ROM, and FIFO buffers. The M4K blocks support the following features:

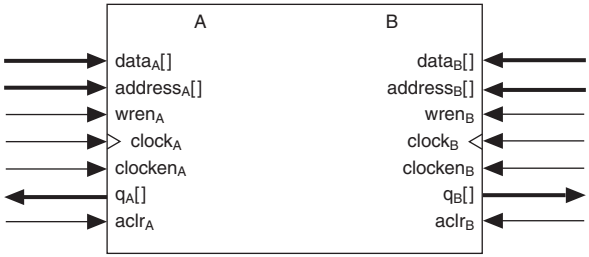
- 4,608 RAM bits
- 250 MHz performance
- True dual-port memory
- Simple dual-port memory
- Single-port memory
- Byte enable
- Parity bits
- Shift register
- FIFO buffer
- ROM
- Mixed clock mode

Violating the setup or hold time on the address registers could corrupt the memory contents. This applies to both read and write operations.

Memory Modes

The M4K memory blocks include input registers that synchronize writes and output registers to pipeline designs and improve system performance. M4K blocks offer a true dual-port mode to support any combination of two-port operations: two reads, two writes, or one read and one write at two different clock frequencies. Figure 2–12 shows true dual-port memory.

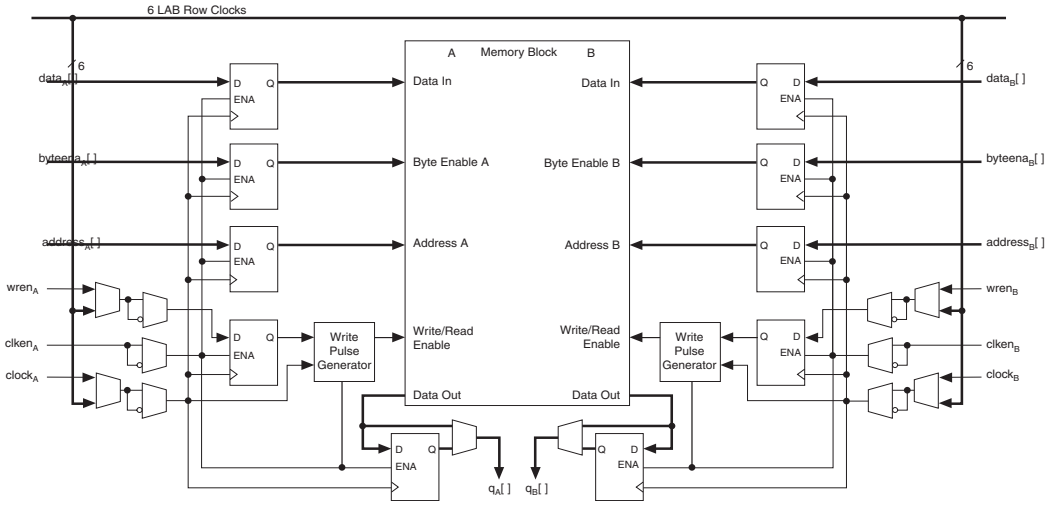
Figure 2–12. True Dual-Port Memory Configuration



Independent Clock Mode

The M4K memory blocks implement independent clock mode for true dual-port memory. In this mode, a separate clock is available for each port (ports A and B). Clock A controls all registers on the port A side, while clock B controls all registers on the port B side. Each port, A and B, also supports independent clock enables and asynchronous clear signals for port A and B registers. Figure 2–17 shows an M4K memory block in independent clock mode.

Figure 2–17. Independent Clock Mode *Notes (1), (2)*



Notes to Figure 2–17:

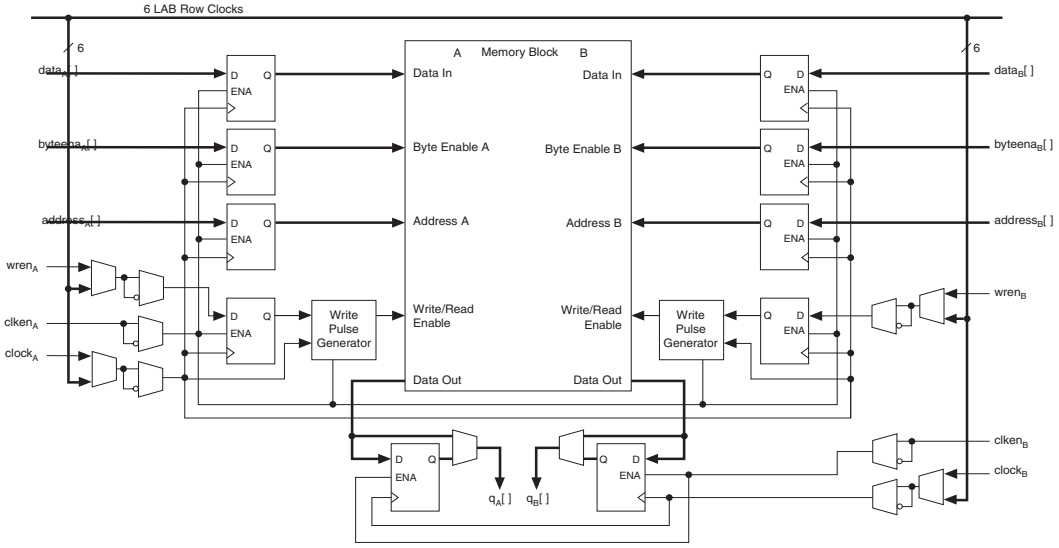
(1) All registers shown have asynchronous clear ports.

(2) Violating the setup or hold time on the address registers could corrupt the memory contents. This applies to both read and write operations.

Input/Output Clock Mode

Input/output clock mode can be implemented for both the true and simple dual-port memory modes. On each of the two ports, A or B, one clock controls all registers for inputs into the memory block: data input, `wren`, and address. The other clock controls the block's data output registers. Each memory block port, A or B, also supports independent clock enables and asynchronous clear signals for input and output registers. Figures 2–18 and 2–19 show the memory block in input/output clock mode.

Figure 2–18. Input/Output Clock Mode in True Dual-Port Mode *Notes (1), (2)*

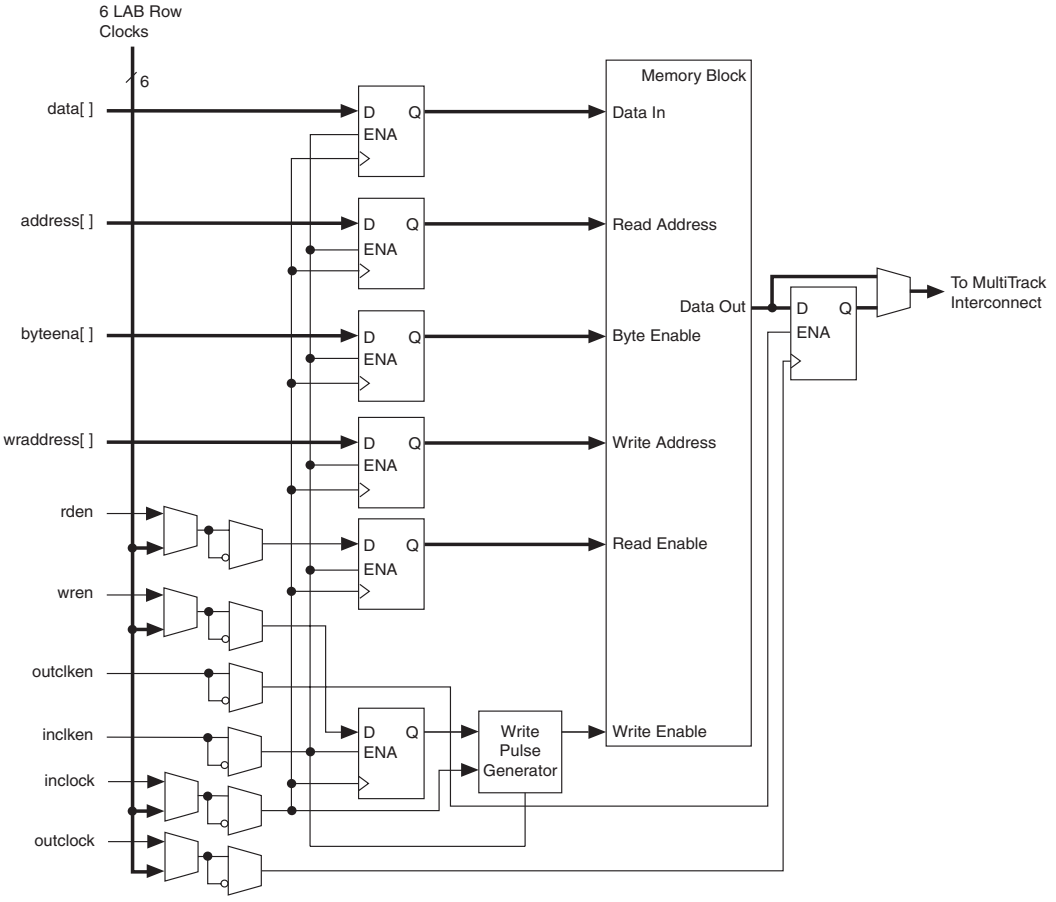


Notes to Figure 2–18:

(1) All registers shown have asynchronous clear ports.

(2) Violating the setup or hold time on the address registers could corrupt the memory contents. This applies to both read and write operations.

Figure 2–19. Input/Output Clock Mode in Simple Dual-Port Mode *Notes (1), (2)*



Notes to Figure 2–19:

(1) All registers shown except the rden register have asynchronous clear ports.

(2) Violating the setup or hold time on the address registers could corrupt the memory contents. This applies to both read and write operations.

External Clock Inputs

Each PLL supports single-ended or differential inputs for source-synchronous receivers or for general-purpose use. The dedicated clock pins (CLK[3..0]) feed the PLL inputs. These dual-purpose pins can also act as LVDS input pins. See Figure 2–25.

Table 2–8 shows the I/O standards supported by PLL input and output pins.

Table 2–8. PLL I/O Standards

I/O Standard	CLK Input	EXTCLK Output
3.3-V LVTTL/LVCMOS	✓	✓
2.5-V LVTTL/LVCMOS	✓	✓
1.8-V LVTTL/LVCMOS	✓	✓
1.5-V LVCMOS	✓	✓
3.3-V PCI	✓	✓
LVDS	✓	✓
SSTL-2 class I	✓	✓
SSTL-2 class II	✓	✓
SSTL-3 class I	✓	✓
SSTL-3 class II	✓	✓
Differential SSTL-2	—	✓

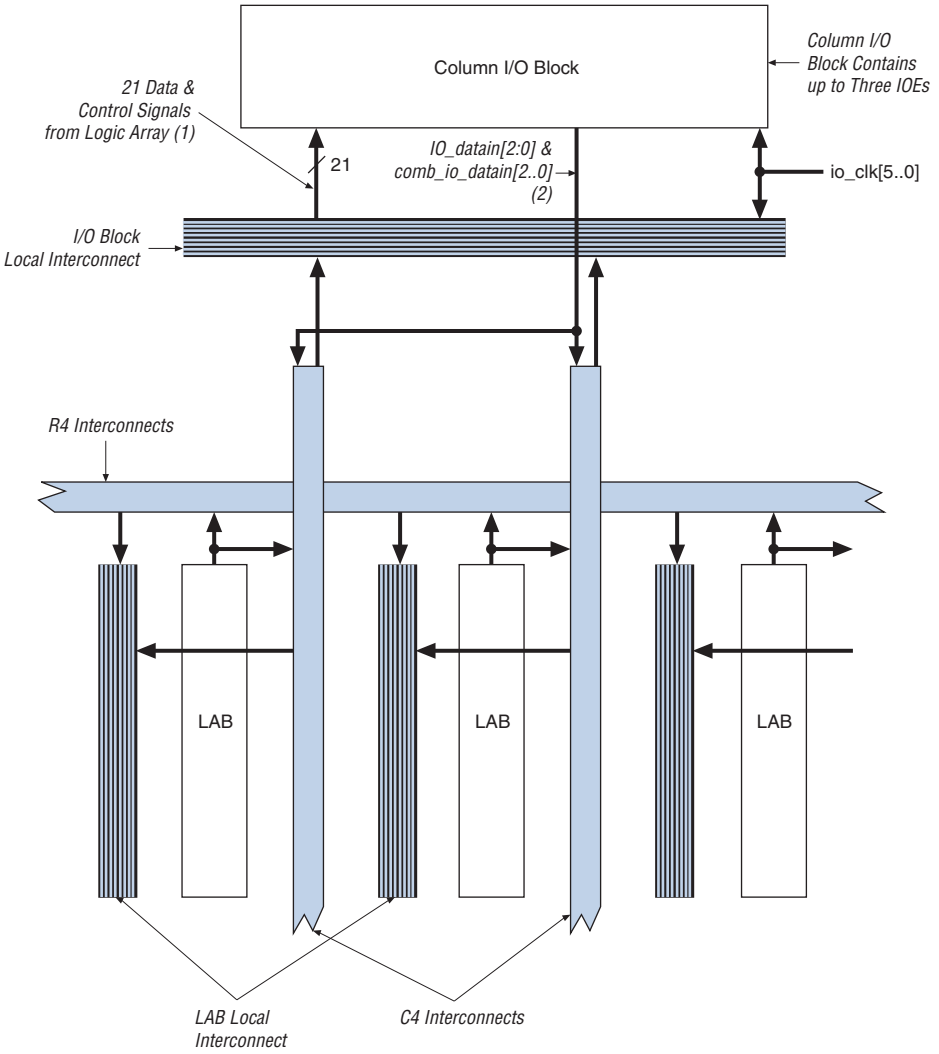
For more information on LVDS I/O support, refer to "LVDS I/O Pins" on page 2–54.

External Clock Outputs

Each PLL supports one differential or one single-ended output for source-synchronous transmitters or for general-purpose external clocks. If the PLL does not use these PLL_OUT pins, the pins are available for use as general-purpose I/O pins. The PLL_OUT pins support all I/O standards shown in Table 2–8.

The external clock outputs do not have their own V_{CC} and ground voltage supplies. Therefore, to minimize jitter, do not place switching I/O pins next to these output pins. The EP1C3 device in the 100-pin TQFP package

Figure 2–29. Column I/O Block Connection to the Interconnect



Notes to Figure 2–29:

(1) The 21 data and control signals consist of three data out lines, `io_dataout[2..0]`, three output enables, `io_coe[2..0]`, three input clock enables, `io_cce_in[2..0]`, three output clock enables, `io_cce_out[2..0]`, three clocks, `io_cclk[2..0]`, three asynchronous clear signals, `io_caclr[2..0]`, and three synchronous clear signals, `io_csclr[2..0]`.

(2) Each of the three IOEs in the column I/O block can have one `io_datain` input (combinatorial or registered) and one `comb_io_datain` (combinatorial) input.

Operating Modes

The Cyclone architecture uses SRAM configuration elements that require configuration data to be loaded each time the circuit powers up. The process of physically loading the SRAM data into the device is called configuration. During initialization, which occurs immediately after configuration, the device resets registers, enables I/O pins, and begins to operate as a logic device. Together, the configuration and initialization processes are called command mode. Normal device operation is called user mode.

SRAM configuration elements allow Cyclone devices to be reconfigured in-circuit by loading new configuration data into the device. With real-time reconfiguration, the device is forced into command mode with a device pin. The configuration process loads different configuration data, reinitializes the device, and resumes user-mode operation. Designers can perform in-field upgrades by distributing new configuration files either within the system or remotely.

A built-in weak pull-up resistor pulls all user I/O pins to V_{CCIO} before and during device configuration.

The configuration pins support 1.5-V/1.8-V or 2.5-V/3.3-V I/O standards. The voltage level of the configuration output pins is determined by the V_{CCIO} of the bank where the pins reside. The bank V_{CCIO} selects whether the configuration inputs are 1.5-V, 1.8-V, 2.5-V, or 3.3-V compatible.

Configuration Schemes

Designers can load the configuration data for a Cyclone device with one of three configuration schemes (see Table 3–5), chosen on the basis of the target application. Designers can use a configuration device, intelligent controller, or the JTAG port to configure a Cyclone device. A low-cost configuration device can automatically configure a Cyclone device at system power-up.

Table 4–8. 1.5-V I/O Specifications

Symbol	Parameter	Conditions	Minimum	Maximum	Unit
V_{CCIO}	Output supply voltage	—	1.4	1.6	V
V_{IH}	High-level input voltage	—	0.65 × V_{CCIO}	V_{CCIO} + 0.3 *(12)*	V
V_{IL}	Low-level input voltage	—	–0.3	0.35 × V_{CCIO}	V
V_{OH}	High-level output voltage	I_{OH} = –2 mA *(11)*	0.75 × V_{CCIO}	—	V
V_{OL}	Low-level output voltage	I_{OL} = 2 mA *(11)*	—	0.25 × V_{CCIO}	V

Table 4–9. 2.5-V LVDS I/O Specifications *Note (13)*

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Unit
V_{CCIO}	I/O supply voltage	—	2.375	2.5	2.625	V
V_{OD}	Differential output voltage	R_L = 100 Ω	250	—	550	mV
Δ V_{OD}	Change in V_{OD} between high and low	R_L = 100 Ω	—	—	50	mV
V_{OS}	Output offset voltage	R_L = 100 Ω	1.125	1.25	1.375	V
Δ V_{OS}	Change in V_{OS} between high and low	R_L = 100 Ω	—	—	50	mV
V_{TH}	Differential input threshold	V_{CM} = 1.2 V	–100	—	100	mV
V_{IN}	Receiver input voltage range	—	0.0	—	2.4	V
R_L	Receiver differential input resistor	—	90	100	110	Ω

Table 4–10. 3.3-V PCI Specifications (Part 1 of 2)

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Unit
V_{CCIO}	Output supply voltage	—	3.0	3.3	3.6	V
V_{IH}	High-level input voltage	—	0.5 × V_{CCIO}	—	V_{CCIO} + 0.5	V
V_{IL}	Low-level input voltage	—	–0.5	—	0.3 × V_{CCIO}	V

Power Consumption

Designers can use the Altera web Early Power Estimator to estimate the device power.

Cyclone devices require a certain amount of power-up current to successfully power up because of the nature of the leading-edge process on which they are fabricated. Table 4–17 shows the maximum power-up current required to power up a Cyclone device.

Table 4–17. Cyclone Maximum Power-Up Current (I_{CCINT}) Requirements (In-Rush Current)

Device	Commercial Specification	Industrial Specification	Unit
EP1C3	150	180	mA
EP1C4	150	180	mA
EP1C6	175	210	mA
EP1C12	300	360	mA
EP1C20	500	600	mA

Notes to Table 4–17:

(1) The Cyclone devices (except for the EP1C20 device) meet the power up specification for Mini PCI.

(2) The lot codes 9G0082 to 9G2999, or 9G3109 and later comply to the specifications in Table 4–17 and meet the Mini PCI specification. Lot codes appear at the top of the device.

(3) The lot codes 9H0004 to 9H29999, or 9H3014 and later comply to the specifications in this table and meet the Mini PCI specification. Lot codes appear at the top of the device.

Designers should select power supplies and regulators that can supply this amount of current when designing with Cyclone devices. This specification is for commercial operating conditions. Measurements were performed with an isolated Cyclone device on the board. Decoupling capacitors were not used in this measurement. To factor in the current for decoupling capacitors, sum up the current for each capacitor using the following equation:

$$I = C\ (dV/dt)$$

The exact amount of current that is consumed varies according to the process, temperature, and power ramp rate. If the power supply or regulator can supply more current than required, the Cyclone device may consume more current than the maximum current specified in Table 4–17. However, the device does not require any more current to successfully power up than what is listed in Table 4–17.

The duration of the I_{CCINT} power-up requirement depends on the V_{CCINT} voltage supply rise time. The power-up current consumption drops when the V_{CCINT} supply reaches approximately 0.75 V. For example, if the V_{CCINT} rise time has a linear rise of 15 ms, the current consumption spike drops by 7.5 ms.

Table 4–22. IOE Internal Timing Microparameter Descriptions	
Symbol	**Parameter**
t_{SU}	IOE input and output register setup time before clock
t_{H}	IOE input and output register hold time after clock
t_{CO}	IOE input and output register clock-to-output delay
$t_{PIN2COMBOUT\_R}$	Row input pin to IOE combinatorial output
$t_{PIN2COMBOUT\_C}$	Column input pin to IOE combinatorial output
$t_{COMBIN2PIN\_R}$	Row IOE data input to combinatorial output pin
$t_{COMBIN2PIN\_C}$	Column IOE data input to combinatorial output pin
t_{CLR}	Minimum clear pulse width
t_{PRE}	Minimum preset pulse width
t_{CLKHL}	Minimum clock high or low time

Table 4–23. M4K Block Internal Timing Microparameter Descriptions	
Symbol	**Parameter**
t_{M4KRC}	Synchronous read cycle time
t_{M4KWC}	Synchronous write cycle time
$t_{M4KWERESU}$	Write or read enable setup time before clock
$t_{M4KWEREH}$	Write or read enable hold time after clock
$t_{M4KBESU}$	Byte enable setup time before clock
t_{M4KBEH}	Byte enable hold time after clock
$t_{M4KDATAASU}$	A port data setup time before clock
$t_{M4KDATAAH}$	A port data hold time after clock
$t_{M4KADDRASU}$	A port address setup time before clock
$t_{M4KADDRAH}$	A port address hold time after clock
$t_{M4KDATABSU}$	B port data setup time before clock
$t_{M4KDATABH}$	B port data hold time after clock
$t_{M4KADDRBSU}$	B port address setup time before clock
$t_{M4KADDRBH}$	B port address hold time after clock
$t_{M4KDATACO1}$	Clock-to-output delay when using output registers
$t_{M4KDATACO2}$	Clock-to-output delay without output registers
$t_{M4KCLKHL}$	Minimum clock high or low time
t_{M4KCLR}	Minimum clear pulse width

Tables 4–34 through 4–35 show the external timing parameters on column and row pins for EP1C6 devices.

Table 4–34. EP1C6 Column Pin Global Clock External I/O Timing Parameters

Symbol	-6 Speed Grade		-7 Speed Grade		-8 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{INSU}	2.691	—	3.094	—	3.496	—	ns
t_{INH}	0.000	—	0.000	—	0.000	—	ns
t_{OUTCO}	2.000	3.917	2.000	4.503	2.000	5.093	ns
$t_{INSUPLL}$	1.513	—	1.739	—	1.964	—	ns
t_{INHPLL}	0.000	—	0.000	—	0.000	—	ns
$t_{OUTCOPLL}$	0.500	2.038	0.500	2.343	0.500	2.651	ns

Table 4–35. EP1C6 Row Pin Global Clock External I/O Timing Parameters

Symbol	-6 Speed Grade		-7 Speed Grade		-8 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{INSU}	2.774	—	3.190	—	3.605	—	ns
t_{INH}	0.000	—	0.000	—	0.000	—	ns
t_{OUTCO}	2.000	3.817	2.000	4.388	2.000	4.963	ns
$t_{INSUPLL}$	1.596	—	1.835	—	2.073	—	ns
t_{INHPLL}	0.000	—	0.000	—	0.000	—	ns
$t_{OUTCOPLL}$	0.500	1.938	0.500	2.228	0.500	2.521	ns

Tables 4–36 through 4–37 show the external timing parameters on column and row pins for EP1C12 devices.

Table 4–36. EP1C12 Column Pin Global Clock External I/O Timing Parameters (Part 1 of 2)

Symbol	-6 Speed Grade		-7 Speed Grade		-8 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{INSU}	2.510	—	2.885	—	3.259	—	ns
t_{INH}	0.000	—	0.000	—	0.000	—	ns
tO_{UTCO}	2.000	3.798	2.000	4.367	2.000	4.940	ns
$t_{INSUPLL}$	1.588	—	1.824	—	2.061	—	ns

Table 4–43. Cyclone I/O Standard Output Delay Adders for Fast Slew Rate on Row Pins (Part 2 of 2)

Standard		-6 Speed Grade		-7 Speed Grade		-8 Speed Grade		Unit
		Min	Max	Min	Max	Min	Max	
1.8-V LVTTL	2 mA	—	1,290	—	1,483	—	1,677	ps
	8 mA	—	4	—	4	—	5	ps
	12 mA	—	–208	—	–240	—	–271	ps
1.5-V LVTTL	2 mA	—	2,288	—	2,631	—	2,974	ps
	4 mA	—	608	—	699	—	790	ps
	8 mA	—	292	—	335	—	379	ps
3.3-V PCI *(1)*		—	–877	—	–1,009	—	–1,141	ps
SSTL-3 class I		—	–410	—	–472	—	–533	ps
SSTL-3 class II		—	–811	—	–933	—	–1,055	ps
SSTL-2 class I		—	–485	—	–558	—	–631	ps
SSTL-2 class II		—	–758	—	–872	—	–986	ps
LVDS		—	–998	—	–1,148	—	–1,298	ps

Table 4–44. Cyclone I/O Standard Output Delay Adders for Slow Slew Rate on Column Pins (Part 1 of 2)

I/O Standard		-6 Speed Grade		-7 Speed Grade		-8 Speed Grade		Unit
		Min	Max	Min	Max	Min	Max	
LVCMOS	2 mA	—	1,800	—	2,070	—	2,340	ps
	4 mA	—	1,311	—	1,507	—	1,704	ps
	8 mA	—	945	—	1,086	—	1,228	ps
	12 mA	—	807	—	928	—	1,049	ps
3.3-V LVTTL	4 mA	—	1,831	—	2,105	—	2,380	ps
	8 mA	—	1,484	—	1,705	—	1,928	ps
	12 mA	—	973	—	1,118	—	1,264	ps
	16 mA	—	1,012	—	1,163	—	1,315	ps
	24 mA	—	838	—	963	—	1,089	ps
2.5-V LVTTL	2 mA	—	2,747	—	3,158	—	3,570	ps
	8 mA	—	1,757	—	2,019	—	2,283	ps
	12 mA	—	1,763	—	2,026	—	2,291	ps
	16 mA	—	1,623	—	1,865	—	2,109	ps
1.8-V LVTTL	2 mA	—	5,506	—	6,331	—	7,157	ps
	8 mA	—	4,220	—	4,852	—	5,485	ps
	12 mA	—	4,008	—	4,608	—	5,209	ps

5. Reference and Ordering Information

C51005-1.4

Software

Cyclone® devices are supported by the Altera® Quartus® II design software, which provides a comprehensive environment for system-on-a-programmable-chip (SOPC) design. The Quartus II software includes HDL and schematic design entry, compilation and logic synthesis, full simulation and advanced timing analysis, SignalTap® II logic analysis, and device configuration.

For more information about the Quartus II software features, refer to the *Quartus II Handbook*.

The Quartus II software supports the Windows 2000/NT/98, Sun Solaris, Linux Red Hat v7.1 and HP-UX operating systems. It also supports seamless integration with industry-leading EDA tools through the NativeLink® interface.

Device Pin-Outs

Device pin-outs for Cyclone devices are available on the Altera website (www.altera.com) and in the *Cyclone Device Handbook*.

Ordering Information

Figure 5–1 describes the ordering codes for Cyclone devices. For more information about a specific package, refer to the *Package Information for Cyclone Devices* chapter in the *Cyclone Device Handbook*.