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# Understanding <u>Embedded - FPGAs (Field Programmable Gate Array)</u>

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

## **Applications of Embedded - FPGAs**

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	291
Number of Logic Elements/Cells	2910
Total RAM Bits	59904
Number of I/O	65
Number of Gates	-
Voltage - Supply	1.425V ~ 1.575V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	100-TQFP
Supplier Device Package	100-TQFP (14x14)
Purchase URL	https://www.e-xfl.com/product-detail/intel/ep1c3t100c8n

Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

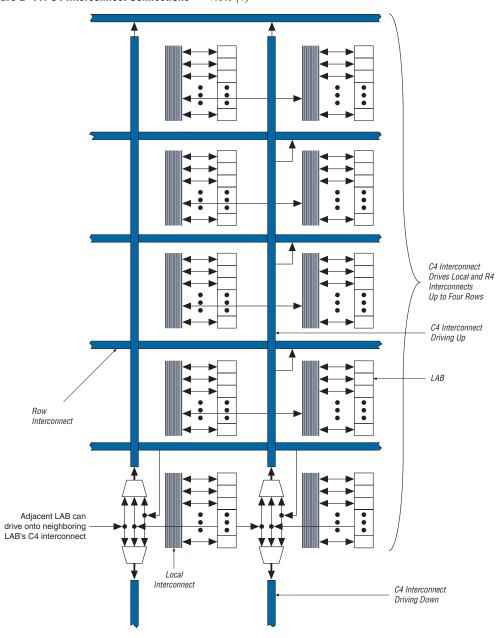


Figure 2–11. C4 Interconnect Connections Note (1)

Note to Figure 2–11:

(1) Each C4 interconnect can drive either up or down four rows.

All embedded blocks communicate with the logic array similar to LAB-to-LAB interfaces. Each block (i.e., M4K memory or PLL) connects to row and column interconnects and has local interconnect regions driven by row and column interconnects. These blocks also have direct link interconnects for fast connections to and from a neighboring LAB.

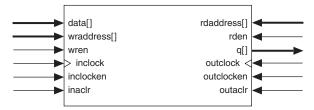
Table 2–2 shows the Cyclone device's routing scheme.

Table 2–2. Cyclone Device Routing Scheme											
	Destination										
Source	LUT Chain	Register Chain	Local Interconnect	Direct Link Interconnect	R4 Interconnect	C4 Interconnect	31	M4K RAM Block	PLL	Column 10E	Row 10E
LUT Chain	_	_	_	_	_	_	<b>✓</b>	_	_	_	_
Register Chain	_	_	_	_	_	_	<b>✓</b>	_	_	_	_
Local Interconnect	_	_	_	_	_	_	<b>✓</b>	~	<b>✓</b>	~	<b>✓</b>
Direct Link Interconnect	_	_	<b>✓</b>	_	_	_	_	_	_	_	_
R4 Interconnect	_	_	<b>✓</b>	_	<b>✓</b>	<b>✓</b>	_	_	_	_	_
C4 Interconnect	_	_	<b>✓</b>	_	<b>✓</b>	<b>✓</b>	_	_	_	_	_
LE	<b>✓</b>	<b>✓</b>	~	~	<b>✓</b>	<b>✓</b>	_	_	_	_	_
M4K RAM Block	_	_	<b>✓</b>	<b>✓</b>	<b>✓</b>	<b>✓</b>	_	_	_	_	_
PLL	_	_	_	<b>✓</b>	<b>✓</b>	<b>✓</b>	_	_	_	_	_
Column IOE	_	_	_	_	_	<b>✓</b>	_	_	_	_	_
Row IOE	_	_	_	<b>✓</b>	<b>✓</b>	<b>✓</b>	_	_	_	_	_

In addition to true dual-port memory, the M4K memory blocks support simple dual-port and single-port RAM. Simple dual-port memory supports a simultaneous read and write. Single-port memory supports non-simultaneous reads and writes. Figure 2–13 shows these different M4K RAM memory port configurations.

Figure 2–13. Simple Dual-Port and Single-Port Memory Configurations

#### Simple Dual-Port Memory



#### Single-Port Memory (1)



#### *Note to Figure 2–13:*

 Two single-port memory blocks can be implemented in a single M4K block as long as each of the two independent block sizes is equal to or less than half of the M4K block size.

The memory blocks also enable mixed-width data ports for reading and writing to the RAM ports in dual-port RAM configuration. For example, the memory block can be written in  $\times 1$  mode at port A and read out in  $\times 16$  mode from port B.

The Cyclone memory architecture can implement fully synchronous RAM by registering both the input and output signals to the M4K RAM block. All M4K memory block inputs are registered, providing synchronous write cycles. In synchronous operation, the memory block generates its own self-timed strobe write enable (wren) signal derived from a global clock. In contrast, a circuit using asynchronous RAM must generate the RAM wren signal while ensuring its data and address signals meet setup and hold time specifications relative to the wren

### **Byte Enables**

M4K blocks support byte writes when the write port has a data width of 16, 18, 32, or 36 bits. The byte enables allow the input data to be masked so the device can write to specific bytes. The unwritten bytes retain the previous written value. Table 2–5 summarizes the byte selection.

Table 2–5. Byte Enable for M4K BlocksNotes (1), (2)						
byteena[30]	datain ×18	datain ×36				
[0] = 1	[80]	[80]				
[1] = 1	[179]	[179]				
[2] = 1	_	[2618]				
[3] = 1	_	[3527]				

Notes to Table 2-5:

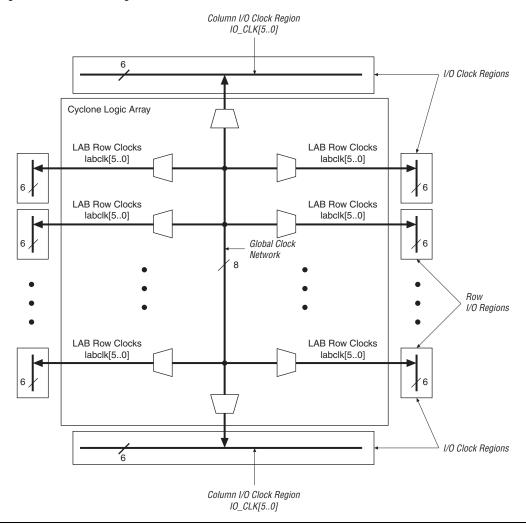
- (1) Any combination of byte enables is possible.
- (2) Byte enables can be used in the same manner with 8-bit words, i.e., in ×16 and ×32 modes.

### **Control Signals and M4K Interface**

The M4K blocks allow for different clocks on their inputs and outputs. Either of the two clocks feeding the block can clock M4K block registers (renwe, address, byte enable, datain, and output registers). Only the output register can be bypassed. The six labclk signals or local interconnects can drive the control signals for the A and B ports of the M4K block. LEs can also control the clock\_a, clock\_b, renwe\_a, renwe\_b, clr\_a, clr\_b, clocken\_a, and clocken\_b signals, as shown in Figure 2–15.

The R4, C4, and direct link interconnects from adjacent LABs drive the M4K block local interconnect. The M4K blocks can communicate with LABs on either the left or right side through these row resources or with LAB columns on either the right or left with the column resources. Up to 10 direct link input connections to the M4K block are possible from the left adjacent LABs and another 10 possible from the right adjacent LAB. M4K block outputs can also connect to left and right LABs through 10 direct link interconnects each. Figure 2–16 shows the M4K block to logic array interface.

Figure 2-24. I/O Clock Regions



#### **PLLs**

Cyclone PLLs provide general-purpose clocking with clock multiplication and phase shifting as well as outputs for differential I/O support. Cyclone devices contain two PLLs, except for the EP1C3 device, which contains one PLL.

## **External Clock Inputs**

Each PLL supports single-ended or differential inputs for source-synchronous receivers or for general-purpose use. The dedicated clock pins (CLK[3..0]) feed the PLL inputs. These dual-purpose pins can also act as LVDS input pins. See Figure 2–25.

Table 2–8 shows the I/O standards supported by PLL input and output pins.

Table 2–8. PLL I/O Standards		
I/O Standard	CLK Input	EXTCLK Output
3.3-V LVTTL/LVCMOS	✓	✓
2.5-V LVTTL/LVCMOS	✓	✓
1.8-V LVTTL/LVCMOS	✓	✓
1.5-V LVCMOS	✓	✓
3.3-V PCI	✓	✓
LVDS	✓	✓
SSTL-2 class I	✓	✓
SSTL-2 class II	✓	✓
SSTL-3 class I	✓	✓
SSTL-3 class II	✓	✓
Differential SSTL-2	_	✓

For more information on LVDS I/O support, refer to "LVDS I/O Pins" on page 2–54.

# **External Clock Outputs**

Each PLL supports one differential or one single-ended output for source-synchronous transmitters or for general-purpose external clocks. If the PLL does not use these PLL\_OUT pins, the pins are available for use as general-purpose I/O pins. The PLL\_OUT pins support all I/O standards shown in Table 2–8.

The external clock outputs do not have their own  $V_{CC}$  and ground voltage supplies. Therefore, to minimize jitter, do not place switching I/O pins next to these output pins. The EP1C3 device in the 100-pin TQFP package

to automatically minimize setup time while providing a zero hold time. Programmable delays can increase the register-to-pin delays for output registers. Table 2–9 shows the programmable delays for Cyclone devices.

Table 2–9. Cyclone Programmable Delay Chain						
Programmable Delays Quartus II Logic Option						
Input pin to logic array delay	Decrease input delay to internal cells					
Input pin to input register delay	Decrease input delay to input registers					
Output pin delay	Increase delay to output pin					

There are two paths in the IOE for a combinatorial input to reach the logic array. Each of the two paths can have a different delay. This allows you adjust delays from the pin to internal LE registers that reside in two different areas of the device. The designer sets the two combinatorial input delays by selecting different delays for two different paths under the **Decrease input delay to internal cells** logic option in the Quartus II software. When the input signal requires two different delays for the combinatorial input, the input register in the IOE is no longer available.

The IOE registers in Cyclone devices share the same source for clear or preset. The designer can program preset or clear for each individual IOE. The designer can also program the registers to power up high or low after configuration is complete. If programmed to power up low, an asynchronous clear can control the registers. If programmed to power up high, an asynchronous preset can control the registers. This feature prevents the inadvertent activation of another device's active-low input upon power up. If one register in an IOE uses a preset or clear signal then all registers in the IOE must use that same signal if they require preset or clear. Additionally a synchronous reset signal is available to the designer for the IOE registers.

# **External RAM Interfacing**

Cyclone devices support DDR SDRAM and FCRAM interfaces at up to 133 MHz through dedicated circuitry.

#### DDR SDRAM and FCRAM

Cyclone devices have dedicated circuitry for interfacing with DDR SDRAM. All I/O banks support DDR SDRAM and FCRAM I/O pins. However, the configuration input pins in bank 1 must operate at 2.5 V because the SSTL-2  $\rm V_{CCIO}$  level is 2.5 V. Additionally, the configuration

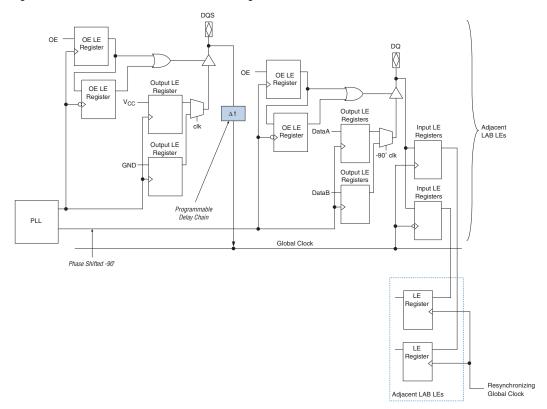


Figure 2-34. DDR SDRAM and FCRAM Interfacing

# **Programmable Drive Strength**

The output buffer for each Cyclone device I/O pin has a programmable drive strength control for certain I/O standards. The LVTTL and LVCMOS standards have several levels of drive strength that the designer can control. SSTL-3 class I and II, and SSTL-2 class I and II support a minimum setting, the lowest drive strength that guarantees the  $\rm I_{OH}/I_{OL}$ 

of the standard. Using minimum settings provides signal slew rate control to reduce system noise and signal overshoot. Table 2–11 shows the possible settings for the I/O standards with drive strength control.

Table 2–11. Programmable Drive Strength Note (1)				
I/O Standard	I <sub>OH</sub> /I <sub>OL</sub> Current Strength Setting (mA)			
LVTTL (3.3 V)	4			
	8			
	12			
	16			
	24(2)			
LVCMOS (3.3 V)	2			
	4			
	8			
	12(2)			
VTTL (2.5 V)	2			
	8			
	12			
	16(2)			
LVTTL (1.8 V)	2			
	8			
	12(2)			
LVCMOS (1.5 V)	2			
	4			
	8(2)			

#### *Notes to Table 2–11:*

- SSTL-3 class I and II, SSTL-2 class I and II, and 3.3-V PCI I/O Standards do not support programmable drive strength.
- (2) This is the default current strength setting in the Quartus II software.

# **Open-Drain Output**

Cyclone devices provide an optional open-drain (equivalent to an open-collector) output for each I/O pin. This open-drain output enables the device to provide system-level control signals (e.g., interrupt and write-enable signals) that can be asserted by any of several devices.

#### Slew-Rate Control

The output buffer for each Cyclone device I/O pin has a programmable output slew-rate control that can be configured for low noise or high-speed performance. A faster slew rate provides high-speed transitions for high-performance systems. However, these fast transitions may introduce noise transients into the system. A slow slew rate reduces system noise, but adds a nominal delay to rising and falling edges. Each I/O pin has an individual slew-rate control, allowing the designer to specify the slew rate on a pin-by-pin basis. The slew-rate control affects both the rising and falling edges.

#### **Bus Hold**

Each Cyclone device I/O pin provides an optional bus-hold feature. The bus-hold circuitry can hold the signal on an I/O pin at its last-driven state. Since the bus-hold feature holds the last-driven state of the pin until the next input signal is present, an external pull-up or pull-down resistor is not necessary to hold a signal level when the bus is tri-stated.

The bus-hold circuitry also pulls undriven pins away from the input threshold voltage where noise can cause unintended high-frequency switching. The designer can select this feature individually for each I/O pin. The bus-hold output will drive no higher than  $V_{\rm CCIO}$  to prevent overdriving signals. If the bus-hold feature is enabled, the device cannot use the programmable pull-up option. Disable the bus-hold feature when the I/O pin is configured for differential signals.

The bus-hold circuitry uses a resistor with a nominal resistance (RBH) of approximately 7 k $\Omega$ to pull the signal level to the last-driven state. Table 4–15 on page 4–6 gives the specific sustaining current for each  $V_{\text{CCIO}}$  voltage level driven through this resistor and overdrive current used to identify the next-driven input level.

The bus-hold circuitry is only active after configuration. When going into user mode, the bus-hold circuit captures the value on the pin present at the end of configuration.

# Programmable Pull-Up Resistor

Each Cyclone device I/O pin provides an optional programmable pull-up resistor during user mode. If the designer enables this feature for an I/O pin, the pull-up resistor (typically 25 k $\Omega$ ) holds the output to the V<sub>CCIO</sub> level of the output pin's bank. Dedicated clock pins do not have the optional programmable pull-up resistor.



Cyclone devices must be within the first 8 devices in a JTAG chain. All of these devices have the same JTAG controller. If any of the Cyclone devices are in the 9th or after they will fail configuration. This does not affect the SignalTap® II logic analyzer.



For more information on JTAG, refer to the following documents:

- AN 39: IEEE Std. 1149.1 (JTAG) Boundary-Scan Testing in Altera Devices
- Jam Programming & Test Language Specification

# SignalTap II Embedded Logic Analyzer

Cyclone devices feature the SignalTap II embedded logic analyzer, which monitors design operation over a period of time through the IEEE Std. 1149.1 (JTAG) circuitry. A designer can analyze internal logic at speed without bringing internal signals to the I/O pins. This feature is particularly important for advanced packages, such as FineLine BGA packages, because it can be difficult to add a connection to a pin during the debugging process after a board is designed and manufactured.

# Configuration

The logic, circuitry, and interconnects in the Cyclone architecture are configured with CMOS SRAM elements. Altera FPGAs are reconfigurable and every device is tested with a high coverage production test program so the designer does not have to perform fault testing and can instead focus on simulation and design verification.

Cyclone devices are configured at system power-up with data stored in an Altera configuration device or provided by a system controller. The Cyclone device's optimized interface allows the device to act as controller in an active serial configuration scheme with the new low-cost serial configuration device. Cyclone devices can be configured in under 120 ms using serial data at 20 MHz. The serial configuration device can be programmed via the ByteBlaster II download cable, the Altera Programming Unit (APU), or third-party programmers.

In addition to the new low-cost serial configuration device, Altera offers in-system programmability (ISP)-capable configuration devices that can configure Cyclone devices via a serial data stream. The interface also enables microprocessors to treat Cyclone devices as memory and configure them by writing to a virtual memory location, making reconfiguration easy. After a Cyclone device has been configured, it can be reconfigured in-circuit by resetting the device and loading new data. Real-time changes can be made during system operation, enabling innovative reconfigurable computing applications.

# Power Consumption

Designers can use the Altera web Early Power Estimator to estimate the device power.

Cyclone devices require a certain amount of power-up current to successfully power up because of the nature of the leading-edge process on which they are fabricated. Table 4–17 shows the maximum power-up current required to power up a Cyclone device.

Table 4–17. Cyclone Maximum Power-Up Current (I <sub>CCINT</sub> ) Requirements (In-Rush Current)							
Device	Commercial Specification	Industrial Specification	Unit				
EP1C3	150	180	mA				
EP1C4	150	180	mA				
EP1C6	175	210	mA				
EP1C12	300	360	mA				
EP1C20	500	600	mA				

#### *Notes to Table 4–17:*

- The Cyclone devices (except for the EP1C20 device) meet the power up specification for Mini PCI.
- (2) The lot codes 9G0082 to 9G2999, or 9G3109 and later comply to the specifications in Table 4–17 and meet the Mini PCI specification. Lot codes appear at the top of the device.
- (3) The lot codes 9H0004 to 9H29999, or 9H3014 and later comply to the specifications in this table and meet the Mini PCI specification. Lot codes appear at the top of the device.

Designers should select power supplies and regulators that can supply this amount of current when designing with Cyclone devices. This specification is for commercial operating conditions. Measurements were performed with an isolated Cyclone device on the board. Decoupling capacitors were not used in this measurement. To factor in the current for decoupling capacitors, sum up the current for each capacitor using the following equation:

$$I = C (dV/dt)$$

The exact amount of current that is consumed varies according to the process, temperature, and power ramp rate. If the power supply or regulator can supply more current than required, the Cyclone device may consume more current than the maximum current specified in Table 4–17. However, the device does not require any more current to successfully power up than what is listed in Table 4–17.

The duration of the  $I_{CCINT}$  power-up requirement depends on the  $V_{CCINT}$  voltage supply rise time. The power-up current consumption drops when the  $V_{CCINT}$  supply reaches approximately 0.75 V. For example, if the  $V_{CCINT}$  rise time has a linear rise of 15 ms, the current consumption spike drops by 7.5 ms.

Table 4–22. IOE Internal Timing Microparameter Descriptions						
Symbol	Parameter					
$t_{SU}$	IOE input and output register setup time before clock					
t <sub>H</sub>	IOE input and output register hold time after clock					
t <sub>CO</sub>	IOE input and output register clock-to-output delay					
t <sub>PIN2COMBOUT_R</sub>	Row input pin to IOE combinatorial output					
t <sub>PIN2COMBOUT_C</sub>	Column input pin to IOE combinatorial output					
t <sub>COMBIN2PIN_R</sub>	Row IOE data input to combinatorial output pin					
t <sub>COMBIN2PIN_C</sub>	Column IOE data input to combinatorial output pin					
t <sub>CLR</sub>	Minimum clear pulse width					
t <sub>PRE</sub>	Minimum preset pulse width					
t <sub>CLKHL</sub>	Minimum clock high or low time					

Table 4–23. M4	K Block Internal Timing Microparameter Descriptions
Symbol	Parameter
t <sub>M4KRC</sub>	Synchronous read cycle time
t <sub>M4KWC</sub>	Synchronous write cycle time
t <sub>M4KWERESU</sub>	Write or read enable setup time before clock
t <sub>M4KWEREH</sub>	Write or read enable hold time after clock
t <sub>M4KBESU</sub>	Byte enable setup time before clock
t <sub>M4KBEH</sub>	Byte enable hold time after clock
t <sub>M4KDATAASU</sub>	A port data setup time before clock
t <sub>M4KDATAAH</sub>	A port data hold time after clock
t <sub>M4KADDRASU</sub>	A port address setup time before clock
t <sub>M4KADDRAH</sub>	A port address hold time after clock
t <sub>M4KDATABSU</sub>	B port data setup time before clock
t <sub>M4KDATABH</sub>	B port data hold time after clock
t <sub>M4KADDRBSU</sub>	B port address setup time before clock
t <sub>M4KADDRBH</sub>	B port address hold time after clock
t <sub>M4KDATACO1</sub>	Clock-to-output delay when using output registers
t <sub>M4KDATACO2</sub>	Clock-to-output delay without output registers
t <sub>M4KCLKHL</sub>	Minimum clock high or low time
t <sub>M4KCLR</sub>	Minimum clear pulse width

Table 4–40. Cyclone I/O S	Standard Col	umn Pin Inp	ut Delay Ad	ders (Part 2	of 2)		
I/O Standard	-6 Spee	d Grade	-7 Spee	d Grade	-8 Speed Grade		Hnit
I/O Standard	Min	Max	Min	Max	Min	Max	Unit
SSTL-2 class II		-278	_	-320	_	-362	ps
LVDS		-261	_	-301	_	-340	ps

Table 4–41. Cyclone I/O Standard Row Pin Input Delay Adders								
I/O Standard	-6 Speed Grade		-7 Spee	ed Grade	-8 Spee	II:A		
	Min	Max	Min	Max	Min	Max	Unit	
LVCMOS	_	0	_	0	_	0	ps	
3.3-V LVTTL	_	0	_	0	_	0	ps	
2.5-V LVTTL	_	27	_	31	_	35	ps	
1.8-V LVTTL	_	182	_	209	_	236	ps	
1.5-V LVTTL	_	278	_	319	_	361	ps	
3.3-V PCI (1)	_	0	_	0	_	0	ps	
SSTL-3 class I	_	-250	_	-288	_	-325	ps	
SSTL-3 class II	_	-250	_	-288	_	-325	ps	
SSTL-2 class I	_	-278	_	-320	_	-362	ps	
SSTL-2 class II	_	-278	_	-320	_	-362	ps	
LVDS	_	-261	_	-301	_	-340	ps	

Standard		-6 Speed Grade		-7 Spe	ed Grade	-8 Spe	1124	
		Min	Max	Min Max		Min	Max	Unit
LVCMOS	2 mA	_	0	_	0	_	0	ps
	4 mA	_	-489	_	-563	_	-636	ps
	8 mA	_	-855	_	-984	_	-1,112	ps
	12 mA	_	-993	_	-1,142	_	-1,291	ps
3.3-V LVTTL	4 mA	_	0	_	0	_	0	ps
	8 mA	_	-347	_	-400	_	-452	ps
	12 mA	_	-858	_	-987	_	-1,116	ps
	16 mA	_	-819	_	-942	_	-1,065	ps
	24 mA	_	-993	_	-1,142	_	-1,291	ps

Table 4–43. Cyclone I/O Standard Output Delay Adders for Fast Slew Rate on Row Pins (Part 2 of 2)								
Standard		-6 Speed Grade		-7 Speed Grade		-8 Speed Grade		1124
		Min	Max	Min	Max	Min	Max	Unit
1.8-V LVTTL	2 mA	_	1,290	_	1,483	_	1,677	ps
	8 mA	_	4	_	4	_	5	ps
	12 mA	_	-208	_	-240	_	-271	ps
1.5-V LVTTL	2 mA	_	2,288	_	2,631	_	2,974	ps
	4 mA	_	608	_	699	_	790	ps
	8 mA	_	292	_	335	_	379	ps
3.3-V PCI (1)		_	-877	_	-1,009	_	-1,141	ps
SSTL-3 class I		_	-410	_	-472	_	-533	ps
SSTL-3 class I	I	_	-811	_	-933	_	-1,055	ps
SSTL-2 class I		_	-485	_	-558	_	-631	ps
SSTL-2 class II		_	-758	_	-872	_	-986	ps
LVDS		_	-998	_	-1,148	_	-1,298	ps

Table 4–44. Cyclone I/O Standard Output Delay Adders for Slow Slew Rate on Column Pins (Part 1 of 2)								
I/O Standard		-6 Speed Grade		-7 Speed Grade		-8 Speed Grade		11
1/U Sta	nuaru	Min	Max	Min	Max	Min	Max	Unit
LVCMOS	2 mA	_	1,800	_	2,070	_	2,340	ps
	4 mA	_	1,311	_	1,507	_	1,704	ps
	8 mA	_	945	_	1,086	_	1,228	ps
	12 mA	_	807	_	928	_	1,049	ps
3.3-V LVTTL	4 mA	_	1,831	_	2,105	_	2,380	ps
	8 mA	_	1,484	_	1,705	_	1,928	ps
	12 mA	_	973	_	1,118	_	1,264	ps
	16 mA	_	1,012	_	1,163	_	1,315	ps
	24 mA	_	838	_	963	_	1,089	ps
2.5-V LVTTL	2 mA	_	2,747	_	3,158	_	3,570	ps
	8 mA	_	1,757	_	2,019	_	2,283	ps
	12 mA	_	1,763	_	2,026	_	2,291	ps
	16 mA	_	1,623	_	1,865	_	2,109	ps
1.8-V LVTTL	2 mA	_	5,506	_	6,331	_	7,157	ps
	8 mA	_	4,220	_	4,852	_	5,485	ps
	12 mA	_	4,008	_	4,608	_	5,209	ps

Table 4–47. Cyclone IOE Programmable Delays on Row Pins								
Davamatav	Setting	-6 Speed Grade		-7 Speed Grade		-8 Speed Grade		IImit.
Parameter		Min	Max	Min	Max	Min	Max	Unit
Decrease input delay to	Off	_	154	_	177	_	200	ps
internal cells	Small	_	2,212	_	2,543	_	2,875	ps
	Medium	_	2,639	_	3,034	_	3,430	ps
	Large	_	3,057	_	3,515	_	3,974	ps
	On	_	154	_	177	_	200	ps
Decrease input delay to input	Off	_	0	_	0	_	0	ps
register	On	_	3,057	_	3,515	_	3,974	ps
Increase delay to output pin	Off	_	0	_	0	_	0	ps
	On	_	556	_	639	_	722	ps

Note to Table 4-47:

# **Maximum Input and Output Clock Rates**

Tables 4--48 and 4--49 show the maximum input clock rate for column and row pins in Cyclone devices.

Table 4–48. Cyclone Maximum Input Clock Rate for Column Pins						
I/O Standard	-6 Speed Grade	-7 Speed Grade	-8 Speed Grade	Unit		
LVTTL	464	428	387	MHz		
2.5 V	392	302	207	MHz		
1.8 V	387	311	252	MHz		
1.5 V	387	320	243	MHz		
LVCMOS	405	374	333	MHz		
SSTL-3 class I	405	356	293	MHz		
SSTL-3 class II	414	365	302	MHz		
SSTL-2 class I	464	428	396	MHz		
SSTL-2 class II	473	432	396	MHz		
LVDS	567	549	531	MHz		

<sup>(1)</sup> EPC1C3 devices do not support the PCI I/O standard.

Table 4–49. Cyclone Maximum Input Clock Rate for Row Pins						
I/O Standard	-6 Speed Grade	-7 Speed Grade	-8 Speed Grade	Unit		
LVTTL	464	428	387	MHz		
2.5 V	392	302	207	MHz		
1.8 V	387	311	252	MHz		
1.5 V	387	320	243	MHz		
LVCMOS	405	374	333	MHz		
SSTL-3 class I	405	356	293	MHz		
SSTL-3 class II	414	365	302	MHz		
SSTL-2 class I	464	428	396	MHz		
SSTL-2 class II	473	432	396	MHz		
3.3-V PCI (1)	464	428	387	MHz		
LVDS	567	549	531	MHz		

Note to Tables 4–48 through 4–49:

Tables 4–50 and 4–51 show the maximum output clock rate for column and row pins in Cyclone devices.

Table 4–50. Cyclone Maximum Output Clock Rate for Column Pins						
I/O Standard	-6 Speed Grade	-7 Speed Grade	-8 Speed Grade	Unit		
LVTTL	304	304	304	MHz		
2.5 V	220	220	220	MHz		
1.8 V	213	213	213	MHz		
1.5 V	166	166	166	MHz		
LVCMOS	304	304	304	MHz		
SSTL-3 class I	100	100	100	MHz		
SSTL-3 class II	100	100	100	MHz		
SSTL-2 class I	134	134	134	MHz		
SSTL-2 class II	134	134	134	MHz		
LVDS	320	320	275	MHz		

Note to Table 4-50:

(1) EP1C3 devices do not support the PCI I/O standard.

<sup>(1)</sup> EP1C3 devices do not support the PCI I/O standard. These parameters are only available on row I/O pins.

Table 4–52. Cyclone PLL Specifications (Part 2 of 2)							
Symbol	Parameter	Min	Max	Unit			
f <sub>OUT</sub> (to global clock)	PLL output frequency (-6 speed grade)	15.625	405	MHz			
	PLL output frequency (-7 speed grade)	15.625	320	MHz			
	PLL output frequency (-8 speed grade)	15.625	275	MHz			
t <sub>OUT</sub> DUTY	Duty cycle for external clock output (when set to 50%)	45.00	55	%			
t <sub>JITTER</sub> (1)	Period jitter for external clock output	_	±300 (2)	ps			
t <sub>LOCK</sub> (3)	Time required to lock from end of device configuration	10.00	100	μs			
f <sub>vco</sub>	PLL internal VCO operating range	500.00	1,000	MHz			
-	Minimum areset time	10	_	ns			
N, G0, G1, E	Counter values	1	32	integer			

#### Notes to Table 4-52:

- (1) The t<sub>JITTER</sub> specification for the PLL[2..1]\_OUT pins are dependent on the I/O pins in its V<sub>CCIO</sub> bank, how many of them are switching outputs, how much they toggle, and whether or not they use programmable current strength or slow slew rate.
- (2)  $f_{OUT} \ge 100$  MHz. When the PLL external clock output frequency ( $f_{OUT}$ ) is smaller than 100 MHz, the jitter specification is 60 mUI.
- (3)  $f_{IN/N}$  must be greater than 200 MHz to ensure correct lock detect circuit operation below -20 C. Otherwise, the PLL operates with the specified parameters under the specified conditions.

July 2003 v1.1	Updated timing information. Timing finalized for EP1C6 and EP1C20 devices. Updated performance information. Added PLL Timing section.	_
May 2003 v1.0	Added document to Cyclone Device Handbook.	_



# 5. Reference and Ordering Information

C51005-1.4

## Software

Cyclone® devices are supported by the Altera® Quartus® II design software, which provides a comprehensive environment for system-on-a-programmable-chip (SOPC) design. The Quartus II software includes HDL and schematic design entry, compilation and logic synthesis, full simulation and advanced timing analysis, SignalTap® II logic analysis, and device configuration.



For more information about the Quartus II software features, refer to the *Quartus II Handbook*.

The Quartus II software supports the Windows 2000/NT/98, Sun Solaris, Linux Red Hat v7.1 and HP-UX operating systems. It also supports seamless integration with industry-leading EDA tools through the NativeLink® interface.

## **Device Pin-Outs**

Device pin-outs for Cyclone devices are available on the Altera website (www.altera.com) and in the *Cyclone Device Handbook*.

# Ordering Information

Figure 5–1 describes the ordering codes for Cyclone devices. For more information about a specific package, refer to the *Package Information for Cyclone Devices* chapter in the *Cyclone Device Handbook*.