Intel - EP1C6T144I7 Datasheet

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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications,

Details	
Product Status	Obsolete
Number of LABs/CLBs	598
Number of Logic Elements/Cells	5980
Total RAM Bits	92160
Number of I/O	98
Number of Gates	-
Voltage - Supply	1.425V ~ 1.575V
Mounting Type	Surface Mount
Operating Temperature	-40°C ~ 100°C (TJ)
Package / Case	144-LQFP
Supplier Device Package	144-TQFP (20x20)
Purchase URL	https://www.e-xfl.com/product-detail/intel/ep1c6t144i7

Table 1–1. Cyclone Device Features (Part 2 of 2)

Feature	EP1C3	EP1C4	EP1C6	EP1C12	EP1C20
Total RAM bits	59,904	78,336	92,160	239,616	294,912
PLLs	1	2	2	2	2
Maximum user I/O pins *(1)*	104	301	185	249	301

Note to Table 1–1:

(1) This parameter includes global clock pins.

Cyclone devices are available in quad flat pack (QFP) and space-saving FineLine® BGA packages (see Tables 1–2 through 1–3).

Table 1–2. Cyclone Package Options and I/O Pin Counts

Device	100-Pin TQFP *(1)*	144-Pin TQFP *(1), (2)*	240-Pin PQFP *(1)*	256-Pin FineLine BGA	324-Pin FineLine BGA	400-Pin FineLine BGA
EP1C3	65	104	—	—	—	—
EP1C4	—	—	—	—	249	301
EP1C6	—	98	185	185	—	—
EP1C12	—	—	173	185	249	—
EP1C20	—	—	—	—	233	301

Notes to Table 1–2:

(1) TQFP: thin quad flat pack.
PQFP: plastic quad flat pack.

(2) Cyclone devices support vertical migration within the same package (i.e., designers can migrate between the EP1C3 device in the 144-pin TQFP package and the EP1C6 device in the same package).

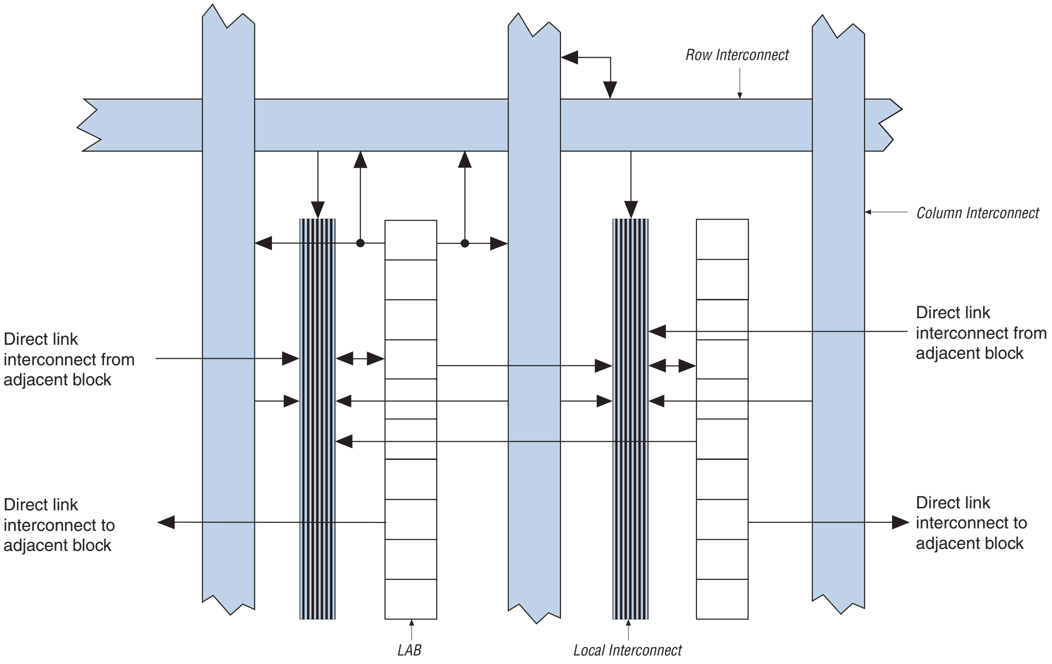
Vertical migration means you can migrate a design from one device to another that has the same dedicated pins, JTAG pins, and power pins, and are subsets or supersets for a given package across device densities. The largest density in any package has the highest number of power pins; you must use the layout for the largest planned density in a package to provide the necessary power pins for migration.

For I/O pin migration across densities, cross-reference the available I/O pins using the device pin-outs for all planned densities of a given package type to identify which I/O pins can be migrated. The Quartus® II software can automatically cross-reference and place all pins for you when given a device migration list. If one device has power or ground pins, but these same pins are user I/O on a different device that is in the migration path,the Quartus II software ensures the pins are not used as user I/O in the Quartus II software. Ensure that these pins are connected

Logic Array Blocks

Each LAB consists of 10 LEs, LE carry chains, LAB control signals, a local interconnect, look-up table (LUT) chain, and register chain connection lines. The local interconnect transfers signals between LEs in the same LAB. LUT chain connections transfer the output of one LE's LUT to the adjacent LE for fast sequential LUT connections within the same LAB. Register chain connections transfer the output of one LE's register to the adjacent LE's register within a LAB. The Quartus® II Compiler places associated logic within a LAB or adjacent LABs, allowing the use of local, LUT chain, and register chain connections for performance and area efficiency. Figure 2–2 details the Cyclone LAB.

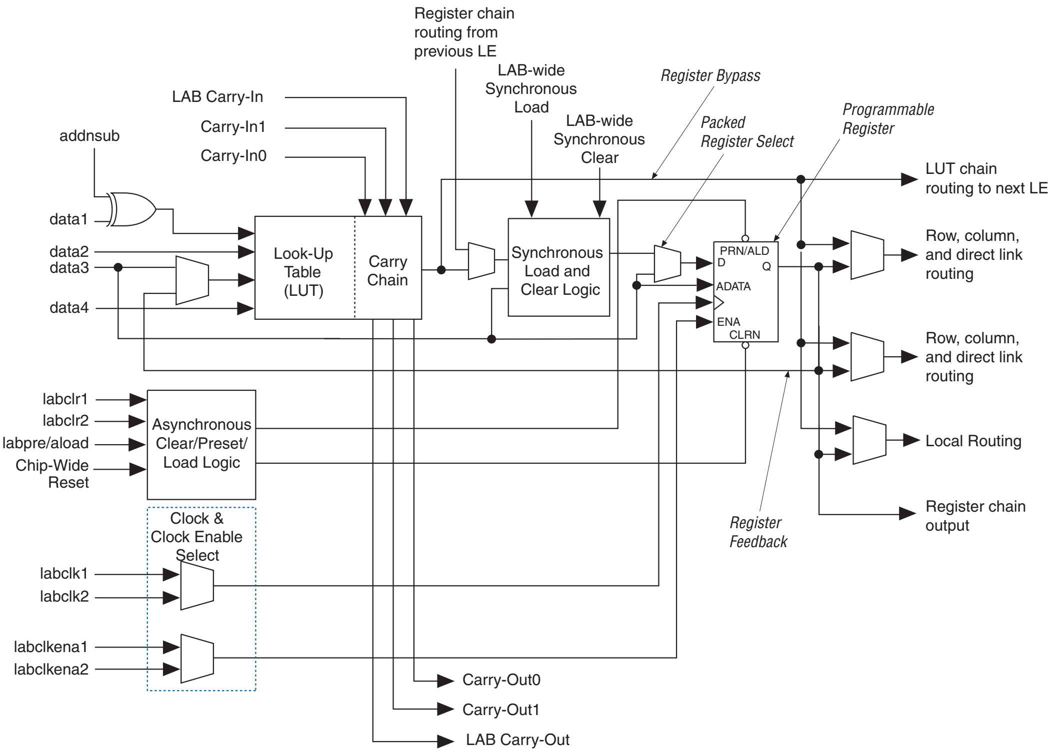
Figure 2–2. Cyclone LAB Structure



LAB Interconnects

The LAB local interconnect can drive LEs within the same LAB. The LAB local interconnect is driven by column and row interconnects and LE outputs within the same LAB. Neighboring LABs, PLLs, and M4K RAM blocks from the left and right can also drive a LAB's local interconnect through the direct link connection. The direct link connection feature minimizes the use of row and column interconnects, providing higher

Figure 2–5. Cyclone LE



Each LE's programmable register can be configured for D, T, JK, or SR operation. Each register has data, true asynchronous load data, clock, clock enable, clear, and asynchronous load/preset inputs. Global signals, general-purpose I/O pins, or any internal logic can drive the register's clock and clear control signals. Either general-purpose I/O pins or internal logic can drive the clock enable, preset, asynchronous load, and asynchronous data. The asynchronous load data input comes from the `data3` input of the LE. For combinatorial functions, the LUT output bypasses the register and drives directly to the LE outputs.

Each LE has three outputs that drive the local, row, and column routing resources. The LUT or register output can drive these three outputs independently. Two LE outputs drive column or row and direct link routing connections and one drives local interconnect resources. This allows the LUT to drive one output while the register drives another output. This feature, called register packing, improves device utilization because the device can use the register and the LUT for unrelated functions. Another special packing mode allows the register output to feed back into the LUT of the same LE so that the register is packed with its own fan-out LUT. This provides another mechanism for improved fitting. The LE can also drive out registered and unregistered versions of the LUT output.

LUT Chain and Register Chain

In addition to the three general routing outputs, the LEs within a LAB have LUT chain and register chain outputs. LUT chain connections allow LUTs within the same LAB to cascade together for wide input functions. Register chain outputs allow registers within the same LAB to cascade together. The register chain output allows a LAB to use LUTs for a single combinatorial function and the registers to be used for an unrelated shift register implementation. These resources speed up connections between LABs while saving local interconnect resources. "MultiTrack Interconnect" on page 2–12 for more information on LUT chain and register chain connections.

addnsub Signal

The LE's dynamic adder/subtractor feature saves logic resources by using one set of LEs to implement both an adder and a subtractor. This feature is controlled by the LAB-wide control signal `addnsub`. The `addnsub` signal sets the LAB to perform either A + B or A –B. The LUT computes addition; subtraction is computed by adding the two's complement of the intended subtractor. The LAB-wide signal converts to two's complement by inverting the B bits within the LAB and setting carry-in = 1 to add one to the least significant bit (LSB). The LSB of an adder/subtractor must be placed in the first LE of the LAB, where the LAB-wide `addnsub` signal automatically sets the carry-in to 1. The Quartus II Compiler automatically places and uses the adder/subtractor feature when using adder/subtractor parameterized functions.

LE Operating Modes

The Cyclone LE can operate in one of the following modes:

- Normal mode
- Dynamic arithmetic mode

Each mode uses LE resources differently. In each mode, eight available inputs to the LE—the four data inputs from the LAB local interconnect, `carry-in0` and `carry-in1` from the previous LE, the LAB carry-in from the previous carry-chain LAB, and the register chain connection—are directed to different destinations to implement the desired logic function. LAB-wide signals provide clock, asynchronous clear, asynchronous

Byte Enables

M4K blocks support byte writes when the write port has a data width of 16, 18, 32, or 36 bits. The byte enables allow the input data to be masked so the device can write to specific bytes. The unwritten bytes retain the previous written value. Table 2–5 summarizes the byte selection.

Table 2–5. Byte Enable for M4K Blocks *Notes (1), (2)*

byteena[3..0]	datain ×18	datain ×36
[0] = 1	[8..0]	[8..0]
[1] = 1	[17..9]	[17..9]
[2] = 1	—	[26..18]
[3] = 1	—	[35..27]

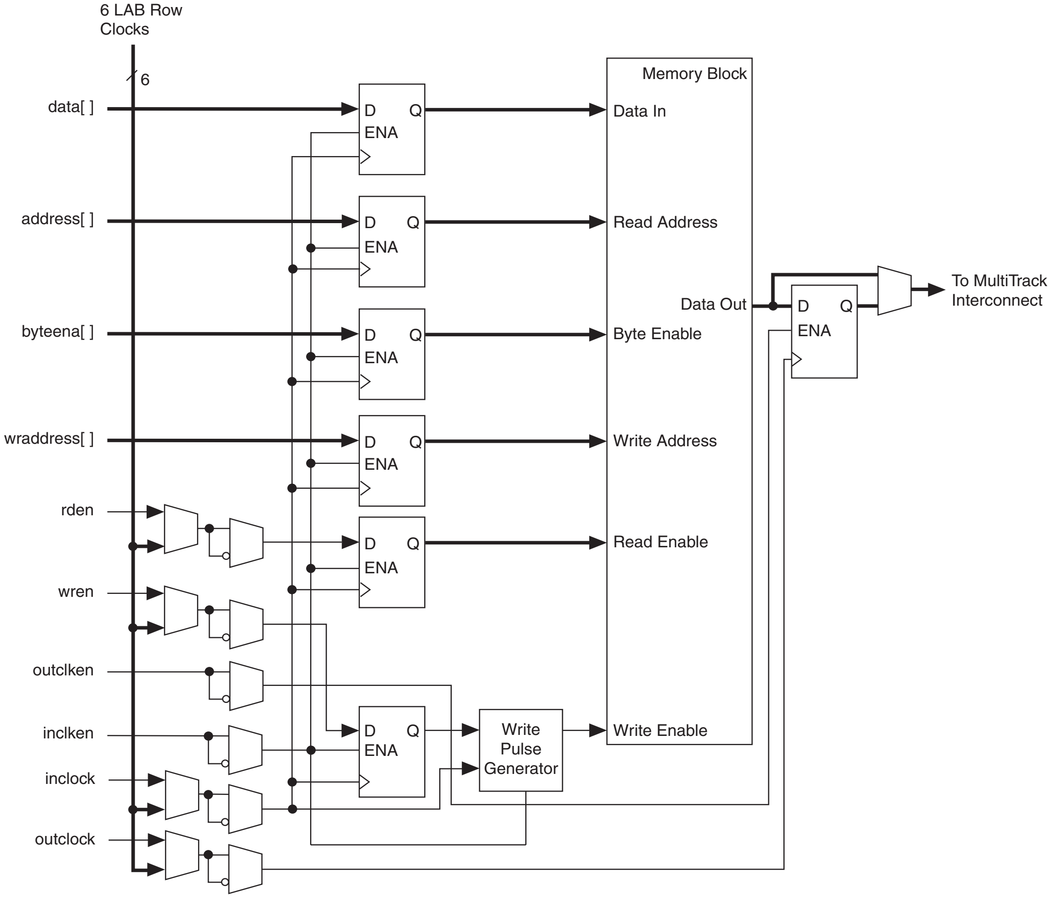
Notes to Table 2–5:
(1) Any combination of byte enables is possible.
(2) Byte enables can be used in the same manner with 8-bit words, i.e., in ×16 and ×32 modes.

Control Signals and M4K Interface

The M4K blocks allow for different clocks on their inputs and outputs. Either of the two clocks feeding the block can clock M4K block registers (`renwe`, address, byte enable, `datain`, and output registers). Only the output register can be bypassed. The six `labclk` signals or local interconnects can drive the control signals for the A and B ports of the M4K block. LEs can also control the `clock_a`, `clock_b`, `renwe_a`, `renwe_b`, `clr_a`, `clr_b`, `clocken_a`, and `clocken_b` signals, as shown in Figure 2–15.

The R4, C4, and direct link interconnects from adjacent LABs drive the M4K block local interconnect. The M4K blocks can communicate with LABs on either the left or right side through these row resources or with LAB columns on either the right or left with the column resources. Up to 10 direct link input connections to the M4K block are possible from the left adjacent LABs and another 10 possible from the right adjacent LAB. M4K block outputs can also connect to left and right LABs through 10 direct link interconnects each. Figure 2–16 shows the M4K block to logic array interface.

Figure 2–19. Input/Output Clock Mode in Simple Dual-Port Mode *Notes (1), (2)*



Notes to Figure 2–19:

(1) All registers shown except the rden register have asynchronous clear ports.

(2) Violating the setup or hold time on the address registers could corrupt the memory contents. This applies to both read and write operations.

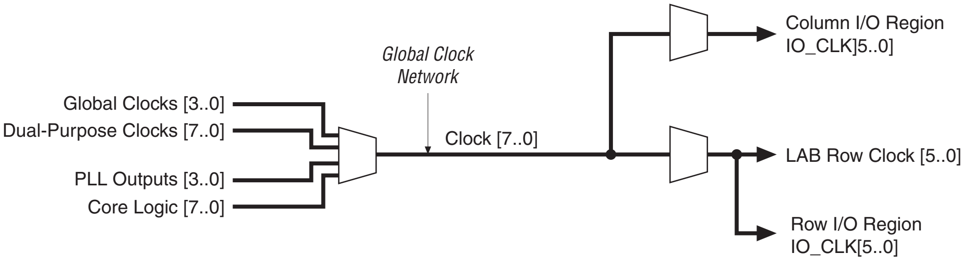
Dual-Purpose Clock Pins

Each Cyclone device except the EP1C3 device has eight dual-purpose clock pins, `DPCLK[7..0]` (two on each I/O bank). EP1C3 devices have five `DPCLK` pins in the 100-pin TQFP package. These dual-purpose pins can connect to the global clock network (see Figure 2–22) for high-fanout control signals such as clocks, asynchronous clears, presets, and clock enables, or protocol control signals such as `TRDY` and `IRDY` for PCI, or DQS signals for external memory interfaces.

Combined Resources

Each Cyclone device contains eight distinct dedicated clocking resources. The device uses multiplexers with these clocks to form six-bit buses to drive LAB row clocks, column IOE clocks, or row IOE clocks. See Figure 2–23. Another multiplexer at the LAB level selects two of the six LAB row clocks to feed the LE registers within the LAB.

Figure 2–23. Global Clock Network Multiplexers



IOE clocks have row and column block regions. Six of the eight global clock resources feed to these row and column regions. Figure 2–24 shows the I/O clock regions.

Table 2–7. Global Clock Network Sources (Part 2 of 2)

Source		GCLK0	GCLK1	GCLK2	GCLK3	GCLK4	GCLK5	GCLK6	GCLK7
Dual-Purpose Clock Pins	DPCLK0 *(3)*	—	—	—	✓	—	—	—	—
	DPCLK1 *(3)*	—	—	✓	—	—	—	—	—
	DPCLK2	✓	—	—	—	—	—	—	—
	DPCLK3	—	—	—	—	✓	—	—	—
	DPCLK4	—	—	—	—	—	—	✓	—
	DPCLK5 *(3)*	—	—	—	—	—	—	—	✓
	DPCLK6	—	—	—	—	—	✓	—	—
	DPCLK7	—	✓	—	—	—	—	—	—

Notes to Table 2–7:

(1) EP1C3 devices only have one PLL (PLL 1).

(2) EP1C3 devices in the 100-pin TQFP package do not have dedicated clock pins `CLK1` and `CLK3`.

(3) EP1C3 devices in the 100-pin TQFP package do not have the `DPCLK0`, `DPCLK1`, or `DPCLK5` pins.

Clock Multiplication and Division

Cyclone PLLs provide clock synthesis for PLL output ports using $m/(n \times$ post scale counter) scaling factors. The input clock is divided by a pre-scale divider, n, and is then multiplied by the m feedback factor. The control loop drives the VCO to match $f_{IN} \times (m/n)$. Each output port has a unique post-scale counter to divide down the high-frequency VCO. For multiple PLL outputs with different frequencies, the VCO is set to the least-common multiple of the output frequencies that meets its frequency specifications. Then, the post-scale dividers scale down the output frequency for each output port. For example, if the output frequencies required from one PLL are 33 and 66 MHz, the VCO is set to 330 MHz (the least-common multiple in the VCO's range).

Each PLL has one pre-scale divider, n, that can range in value from 1 to 32. Each PLL also has one multiply divider, m, that can range in value from 2 to 32. Global clock outputs have two post scale G dividers for global clock outputs, and external clock outputs have an E divider for external clock output, both ranging from 1 to 32. The Quartus II software automatically chooses the appropriate scaling factors according to the input frequency, multiplication, and division values entered.

External Clock Inputs

Each PLL supports single-ended or differential inputs for source-synchronous receivers or for general-purpose use. The dedicated clock pins (`CLK[3..0]`) feed the PLL inputs. These dual-purpose pins can also act as LVDS input pins. See Figure 2–25.

Table 2–8 shows the I/O standards supported by PLL input and output pins.

Table 2–8. PLL I/O Standards

I/O Standard	CLK Input	EXTCLK Output
3.3-V LVTTL/LVCMOS	✓	✓
2.5-V LVTTL/LVCMOS	✓	✓
1.8-V LVTTL/LVCMOS	✓	✓
1.5-V LVCMOS	✓	✓
3.3-V PCI	✓	✓
LVDS	✓	✓
SSTL-2 class I	✓	✓
SSTL-2 class II	✓	✓
SSTL-3 class I	✓	✓
SSTL-3 class II	✓	✓
Differential SSTL-2	—	✓

For more information on LVDS I/O support, refer to "LVDS I/O Pins" on page 2–54.

External Clock Outputs

Each PLL supports one differential or one single-ended output for source-synchronous transmitters or for general-purpose external clocks. If the PLL does not use these `PLL_OUT` pins, the pins are available for use as general-purpose I/O pins. The `PLL_OUT` pins support all I/O standards shown in Table 2–8.

The external clock outputs do not have their own V_{CC} and ground voltage supplies. Therefore, to minimize jitter, do not place switching I/O pins next to these output pins. The EP1C3 device in the 100-pin TQFP package

Programmable Duty Cycle

The programmable duty cycle allows PLLs to generate clock outputs with a variable duty cycle. This feature is supported on each PLL post-scale counter (g0, g1, e). The duty cycle setting is achieved by a low- and high-time count setting for the post-scale dividers. The Quartus II software uses the frequency input and the required multiply or divide rate to determine the duty cycle choices.

Control Signals

There are three control signals for clearing and enabling PLLs and their outputs. You can use these signals to control PLL resynchronization and the ability to gate PLL output clocks for low-power applications.

The `pllenable` signal enables and disables PLLs. When the `pllenable` signal is low, the clock output ports are driven by ground and all the PLLs go out of lock. When the `pllenable` signal goes high again, the PLLs relock and resynchronize to the input clocks. An input pin or LE output can drive the `pllenable` signal.

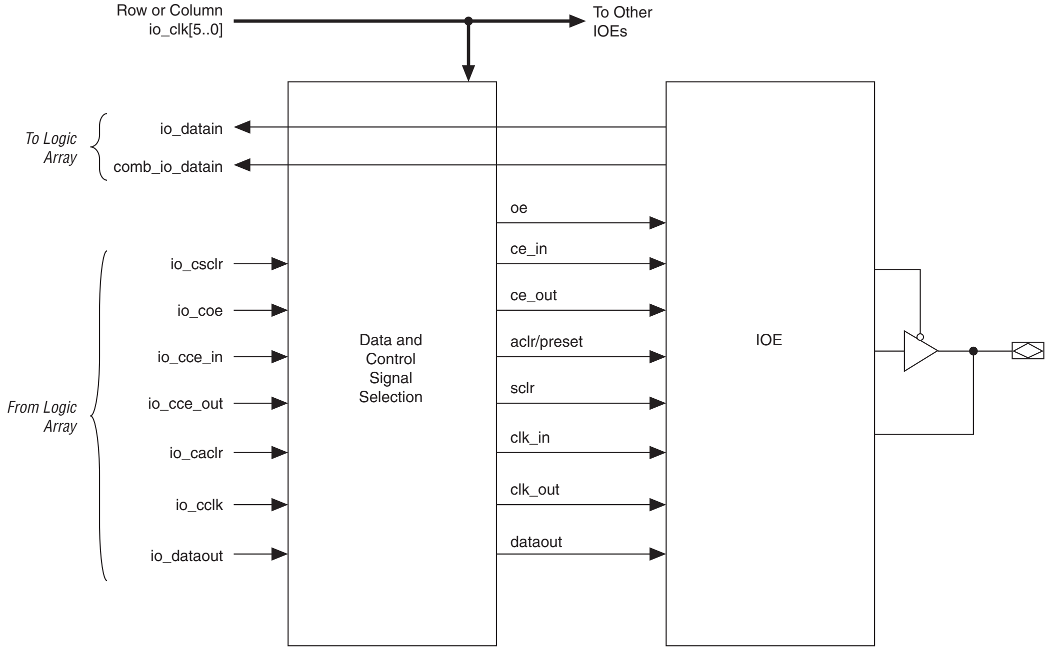
The `areset` signals are reset/resynchronization inputs for each PLL. Cyclone devices can drive these input signals from input pins or from LEs. When `areset` is driven high, the PLL counters will reset, clearing the PLL output and placing the PLL out of lock. When driven low again, the PLL will resynchronize to its input as it relocks.

The `pfdena` signals control the phase frequency detector (PFD) output with a programmable gate. If you disable the PFD, the VCO will operate at its last set value of control voltage and frequency with some drift, and the system will continue running when the PLL goes out of lock or the input clock disables. By maintaining the last locked frequency, the system has time to store its current settings before shutting down. You can either use their own control signal or gated locked status signals to trigger the `pfdena` signal.

For more information about Cyclone PLLs, refer to *Using PLLs in Cyclone Devices* chapter in the *Cyclone Device Handbook*.

The pin's datain signals can drive the logic array. The logic array drives the control and data signals, providing a flexible routing resource. The row or column IOE clocks, `io_clk[5..0]`, provide a dedicated routing resource for low-skew, high-speed clocks. The global clock network generates the IOE clocks that feed the row or column I/O regions (see "Global Clock Network and Phase-Locked Loops" on page 2–29). Figure 2–30 illustrates the signal paths through the I/O block.

Figure 2–30. Signal Path through the I/O Block



Each IOE contains its own control signal selection for the following control signals: `oe`, `ce_in`, `ce_out`, `aclr/preset`, `sclr/preset`, `clk_in`, and `clk_out`. Figure 2–31 illustrates the control signal selection.

Table 2–10. DQ Pin Groups (Part 2 of 2)

Device	Package	Number of × 8 DQ Pin Groups	Total DQ Pin Count
EP1C6	144-pin TQFP	4	32
	240-pin PQFP	4	32
	256-pin FineLine BGA	4	32
EP1C12	240-pin PQFP	4	32
	256-pin FineLine BGA	4	32
	324-pin FineLine BGA	8	64
EP1C20	324-pin FineLine BGA	8	64
	400-pin FineLine BGA	8	64

Note to Table 2–10:

(1) EP1C3 devices in the 100-pin TQFP package do not have any DQ pin groups in I/O bank 1.

A programmable delay chain on each DQS pin allows for either a 90° phase shift (for DDR SDRAM), or a 72° phase shift (for FCRAM) which automatically center-aligns input DQS synchronization signals within the data window of their corresponding DQ data signals. The phase-shifted DQS signals drive the global clock network. This global DQS signal clocks DQ signals on internal LE registers.

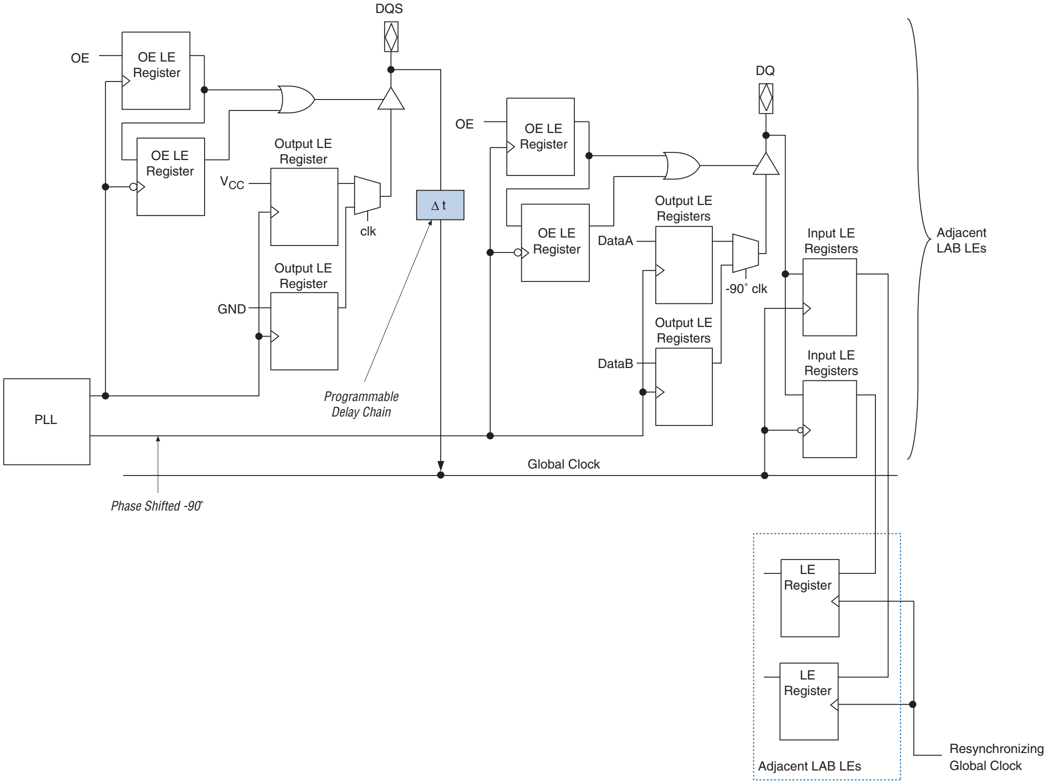
These DQS delay elements combine with the PLL's clocking and phase shift ability to provide a complete hardware solution for interfacing to high-speed memory.

The clock phase shift allows the PLL to clock the DQ output enable and output paths. The designer should use the following guidelines to meet 133 MHz performance for DDR SDRAM and FCRAM interfaces:

- The DQS signal must be in the middle of the DQ group it clocks
- Resynchronize the incoming data to the logic array clock using successive LE registers or FIFO buffers
- LE registers must be placed in the LAB adjacent to the DQ I/O pin column it is fed by

Figure 2–34 illustrates DDR SDRAM and FCRAM interfacing from the I/O through the dedicated circuitry to the logic array.

Figure 2–34. DDR SDRAM and FCRAM Interfacing

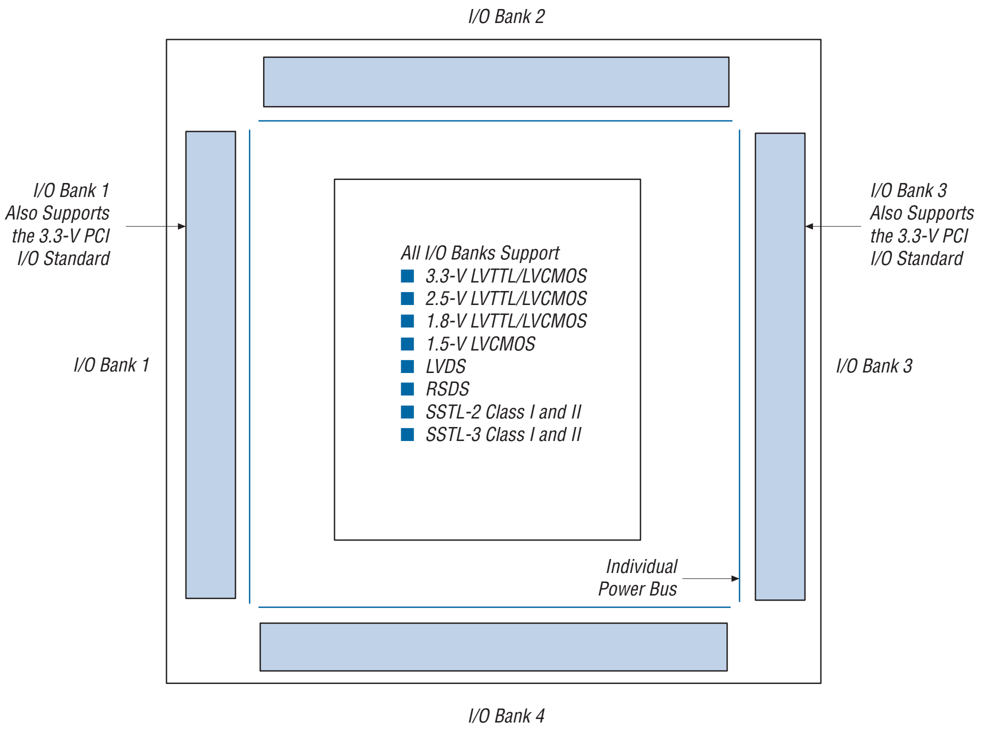


Programmable Drive Strength

The output buffer for each Cyclone device I/O pin has a programmable drive strength control for certain I/O standards. The LVTTL and LVCMOS standards have several levels of drive strength that the designer can control. SSTL-3 class I and II, and SSTL-2 class I and II support a minimum setting, the lowest drive strength that guarantees the I_{OH}/I_{OL}

and DM pins to support a DDR SDRAM or FCRAM interface. I/O bank 1 can also support a DDR SDRAM or FCRAM interface, however, the configuration input pins in I/O bank 1 must operate at 2.5 V. I/O bank 3 can also support a DDR SDRAM or FCRAM interface, however, all the JTAG pins in I/O bank 3 must operate at 2.5 V.

Figure 2–35. Cyclone I/O Banks *Notes (1), (2)*



Notes to Figure 2–35:

(1) Figure 2–35 is a top view of the silicon die.

(2) Figure 2–35 is a graphic representation only. Refer to the pin list and the Quartus II software for exact pin locations.

Each I/O bank has its own `VCCIO` pins. A single device can support 1.5-V, 1.8-V, 2.5-V, and 3.3-V interfaces; each individual bank can support a different standard with different I/O voltages. Each bank also has dual-purpose `VREF` pins to support any one of the voltage-referenced standards (e.g., SSTL-3) independently. If an I/O bank does not use voltage-referenced standards, the V_{REF} pins are available as user I/O pins.

☞ Cyclone devices must be within the first 8 devices in a JTAG chain. All of these devices have the same JTAG controller. If any of the Cyclone devices are in the 9th or after they will fail configuration. This does not affect the SignalTap® II logic analyzer.

For more information on JTAG, refer to the following documents:

- *AN 39: IEEE Std. 1149.1 (JTAG) Boundary-Scan Testing in Altera Devices*
- *Jam Programming & Test Language Specification*

SignalTap II Embedded Logic Analyzer

Cyclone devices feature the SignalTap II embedded logic analyzer, which monitors design operation over a period of time through the IEEE Std. 1149.1 (JTAG) circuitry. A designer can analyze internal logic at speed without bringing internal signals to the I/O pins. This feature is particularly important for advanced packages, such as FineLine BGA packages, because it can be difficult to add a connection to a pin during the debugging process after a board is designed and manufactured.

Configuration

The logic, circuitry, and interconnects in the Cyclone architecture are configured with CMOS SRAM elements. Altera FPGAs are reconfigurable and every device is tested with a high coverage production test program so the designer does not have to perform fault testing and can instead focus on simulation and design verification.

Cyclone devices are configured at system power-up with data stored in an Altera configuration device or provided by a system controller. The Cyclone device's optimized interface allows the device to act as controller in an active serial configuration scheme with the new low-cost serial configuration device. Cyclone devices can be configured in under 120 ms using serial data at 20 MHz. The serial configuration device can be programmed via the ByteBlaster II download cable, the Altera Programming Unit (APU), or third-party programmers.

In addition to the new low-cost serial configuration device, Altera offers in-system programmability (ISP)-capable configuration devices that can configure Cyclone devices via a serial data stream. The interface also enables microprocessors to treat Cyclone devices as memory and configure them by writing to a virtual memory location, making reconfiguration easy. After a Cyclone device has been configured, it can be reconfigured in-circuit by resetting the device and loading new data. Real-time changes can be made during system operation, enabling innovative reconfigurable computing applications.

Table 4–8. 1.5-V I/O Specifications

Symbol	Parameter	Conditions	Minimum	Maximum	Unit
V_{CCIO}	Output supply voltage	—	1.4	1.6	V
V_{IH}	High-level input voltage	—	0.65 × V_{CCIO}	V_{CCIO} + 0.3 *(12)*	V
V_{IL}	Low-level input voltage	—	–0.3	0.35 × V_{CCIO}	V
V_{OH}	High-level output voltage	I_{OH} = –2 mA *(11)*	0.75 × V_{CCIO}	—	V
V_{OL}	Low-level output voltage	I_{OL} = 2 mA *(11)*	—	0.25 × V_{CCIO}	V

Table 4–9. 2.5-V LVDS I/O Specifications *Note (13)*

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Unit
V_{CCIO}	I/O supply voltage	—	2.375	2.5	2.625	V
V_{OD}	Differential output voltage	R_L = 100 Ω	250	—	550	mV
ΔV_{OD}	Change in V_{OD} between high and low	R_L = 100 Ω	—	—	50	mV
V_{OS}	Output offset voltage	R_L = 100 Ω	1.125	1.25	1.375	V
ΔV_{OS}	Change in V_{OS} between high and low	R_L = 100 Ω	—	—	50	mV
V_{TH}	Differential input threshold	V_{CM} = 1.2 V	–100	—	100	mV
V_{IN}	Receiver input voltage range	—	0.0	—	2.4	V
R_L	Receiver differential input resistor	—	90	100	110	Ω

Table 4–10. 3.3-V PCI Specifications (Part 1 of 2)

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Unit
V_{CCIO}	Output supply voltage	—	3.0	3.3	3.6	V
V_{IH}	High-level input voltage	—	0.5 × V_{CCIO}	—	V_{CCIO} + 0.5	V
V_{IL}	Low-level input voltage	—	–0.5	—	0.3 × V_{CCIO}	V

Power Consumption

Designers can use the Altera web Early Power Estimator to estimate the device power.

Cyclone devices require a certain amount of power-up current to successfully power up because of the nature of the leading-edge process on which they are fabricated. Table 4–17 shows the maximum power-up current required to power up a Cyclone device.

Table 4–17. Cyclone Maximum Power-Up Current (I_{CCINT}) Requirements (In-Rush Current)

Device	Commercial Specification	Industrial Specification	Unit
EP1C3	150	180	mA
EP1C4	150	180	mA
EP1C6	175	210	mA
EP1C12	300	360	mA
EP1C20	500	600	mA

Notes to Table 4–17:

(1) The Cyclone devices (except for the EP1C20 device) meet the power up specification for Mini PCI.

(2) The lot codes 9G0082 to 9G2999, or 9G3109 and later comply to the specifications in Table 4–17 and meet the Mini PCI specification. Lot codes appear at the top of the device.

(3) The lot codes 9H0004 to 9H29999, or 9H3014 and later comply to the specifications in this table and meet the Mini PCI specification. Lot codes appear at the top of the device.

Designers should select power supplies and regulators that can supply this amount of current when designing with Cyclone devices. This specification is for commercial operating conditions. Measurements were performed with an isolated Cyclone device on the board. Decoupling capacitors were not used in this measurement. To factor in the current for decoupling capacitors, sum up the current for each capacitor using the following equation:

$$I = C\,(dV/dt)$$

The exact amount of current that is consumed varies according to the process, temperature, and power ramp rate. If the power supply or regulator can supply more current than required, the Cyclone device may consume more current than the maximum current specified in Table 4–17. However, the device does not require any more current to successfully power up than what is listed in Table 4–17.

The duration of the I_{CCINT} power-up requirement depends on the V_{CCINT} voltage supply rise time. The power-up current consumption drops when the V_{CCINT} supply reaches approximately 0.75 V. For example, if the V_{CCINT} rise time has a linear rise of 15 ms, the current consumption spike drops by 7.5 ms.

Table 4–20. Cyclone Device Performance

Resource Used	Design Size and Function	Mode	Resources Used			Performance		
			LEs	M4K Memory Bits	M4K Memory Blocks	-6 Speed Grade (MHz)	-7 Speed Grade (MHz)	-8 Speed Grade (MHz)
M4K memory block	RAM 128 × 36 bit	Single port	—	4,608	1	256.00	222.67	197.01
	RAM 128 × 36 bit	Simple dual-port mode	—	4,608	1	255.95	222.67	196.97
	RAM 256 × 18 bit	True dual-port mode	—	4,608	1	255.95	222.67	196.97
	FIFO 128 × 36 bit	—	40	4,608	1	256.02	222.67	197.01
	Shift register 9 × 4 × 128	Shift register	11	4,536	1	255.95	222.67	196.97

Note to Table 4–20:

(1) The performance numbers for this function are from an EP1C6 device in a 240-pin PQFP package.

Internal Timing Parameters

Internal timing parameters are specified on a speed grade basis independent of device density. Tables 4–21 through 4–24 describe the Cyclone device internal timing microparameters for LEs, IOEs, M4K memory structures, and MultiTrack interconnects.

Table 4–21. LE Internal Timing Microparameter Descriptions

Symbol	Parameter
t_{SU}	LE register setup time before clock
t_H	LE register hold time after clock
t_{CO}	LE register clock-to-output delay
t_{LUT}	LE combinatorial LUT delay for data-in to data-out
t_{CLR}	Minimum clear pulse width
t_{PRE}	Minimum preset pulse width
t_{CLKHL}	Minimum clock high or low time

Tables 4–32 through 4–33 show the external timing parameters on column and row pins for EP1C4 devices.

Table 4–32. EP1C4 Column Pin Global Clock External I/O Timing Parameters Note (1)

Symbol	-6 Speed Grade		-7 Speed Grade		-8 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{INSU}	2.471	—	2.841	—	3.210	—	ns
t_{INH}	0.000	—	0.000	—	0.000	—	ns
t_{OUTCO}	2.000	3.937	2.000	4.526	2.000	5.119	ns
$t_{INSUPLL}$	1.471	—	1.690	—	1.910	—	ns
t_{INHPLL}	0.000	—	0.000	—	0.000	—	ns
$t_{OUTCOPLL}$	0.500	2.080	0.500	2.392	0.500	2.705	ns

Table 4–33. EP1C4 Row Pin Global Clock External I/O Timing Parameters Note (1)

Symbol	-6 Speed Grade		-7 Speed Grade		-8 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{INSU}	2.600	—	2.990	—	3.379	—	ns
t_{INH}	0.000	—	0.000	—	0.000	—	ns
t_{OUTCO}	2.000	3.991	2.000	4.388	2.000	5.189	ns
$t_{INSUPLL}$	1.300	—	1.494	—	1.689	—	ns
t_{INHPLL}	0.000	—	0.000	—	0.000	—	ns
$t_{OUTCOPLL}$	0.500	2.234	0.500	2.569	0.500	2.905	ns

Note to Tables 4–32 and 4–33:

(1) Contact Altera Applications for EP1C4 device timing parameters.

Table 4–49. Cyclone Maximum Input Clock Rate for Row Pins

I/O Standard	-6 Speed Grade	-7 Speed Grade	-8 Speed Grade	Unit
LVTTL	464	428	387	MHz
2.5 V	392	302	207	MHz
1.8 V	387	311	252	MHz
1.5 V	387	320	243	MHz
LVCMOS	405	374	333	MHz
SSTL-3 class I	405	356	293	MHz
SSTL-3 class II	414	365	302	MHz
SSTL-2 class I	464	428	396	MHz
SSTL-2 class II	473	432	396	MHz
3.3-V PCI *(1)*	464	428	387	MHz
LVDS	567	549	531	MHz

Note to Tables 4–48 through 4–49:

(1) EP1C3 devices do not support the PCI I/O standard. These parameters are only available on row I/O pins.

Tables 4–50 and 4–51 show the maximum output clock rate for column and row pins in Cyclone devices.

Table 4–50. Cyclone Maximum Output Clock Rate for Column Pins

I/O Standard	-6 Speed Grade	-7 Speed Grade	-8 Speed Grade	Unit
LVTTL	304	304	304	MHz
2.5 V	220	220	220	MHz
1.8 V	213	213	213	MHz
1.5 V	166	166	166	MHz
LVCMOS	304	304	304	MHz
SSTL-3 class I	100	100	100	MHz
SSTL-3 class II	100	100	100	MHz
SSTL-2 class I	134	134	134	MHz
SSTL-2 class II	134	134	134	MHz
LVDS	320	320	275	MHz

Note to Table 4–50:

(1) EP1C3 devices do not support the PCI I/O standard.