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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	-
Number of Logic Elements/Cells	128
Total RAM Bits	-
Number of I/O	56
Number of Gates	3000
Voltage - Supply	2.3V ~ 2.7V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 70°C (TA)
Package / Case	100-LQFP
Supplier Device Package	100-TQFP (14x14)
Purchase URL	https://www.e-xfl.com/product-detail/microsemi/ex64-ptqg100

Table of Contents

1 – eX FPGA Architecture and Characteristics

General Description

The eX family of FPGAs is a low-cost solution for low-power, high-performance designs. The inherent low power attributes of the antifuse technology, coupled with an additional low static power mode, make these devices ideal for power-sensitive applications. Fabricated with an advanced 0.22 mm CMOS antifuse technology, these devices achieve high performance with no power penalty.

eX Family Architecture

Microsemi eX family is implemented on a high-voltage twin-well CMOS process using 0.22 µm design rules. The eX family architecture uses a “sea-of-modules” structure where the entire floor of the device is covered with a grid of logic modules with virtually no chip area lost to interconnect elements or routing. Interconnection among these logic modules is achieved using Microsemi patented metal-to-metal programmable antifuse interconnect elements. The antifuse interconnect is made up of a combination of amorphous silicon and dielectric material with barrier metals and has an "on" state resistance of 25Ω with a capacitance of 1.0fF for low-signal impedance. The antifuses are normally open circuit and, when programmed, form a permanent low-impedance connection. The eX family provides two types of logic modules, the register cell (R-cell) and the combinatorial cell (C-cell).

The R-cell contains a flip-flop featuring asynchronous clear, asynchronous preset, and clock enable (using the S0 and S1 lines) control signals (Figure 1-1). The R-cell registers feature programmable clock polarity selectable on a register-by-register basis. This provides additional flexibility while allowing mapping of synthesized functions into the eX FPGA. The clock source for the R-cell can be chosen from either the hard-wired clock or the routed clock.

The C-cell implements a range of combinatorial functions up to five inputs (Figure 1-2 on page 1-2). Inclusion of the DB input and its associated inverter function enables the implementation of more than 4,000 combinatorial functions in the eX architecture in a single module.

Two C-cells can be combined together to create a flip-flop to imitate an R-cell via the use of the CC macro. This is particularly useful when implementing non-timing-critical paths and when the design engineer is running out of R-cells. More information about the CC macro can be found in the *Maximizing Logic Utilization in eX, SX and SX-A FPGA Devices Using CC Macros* application note.

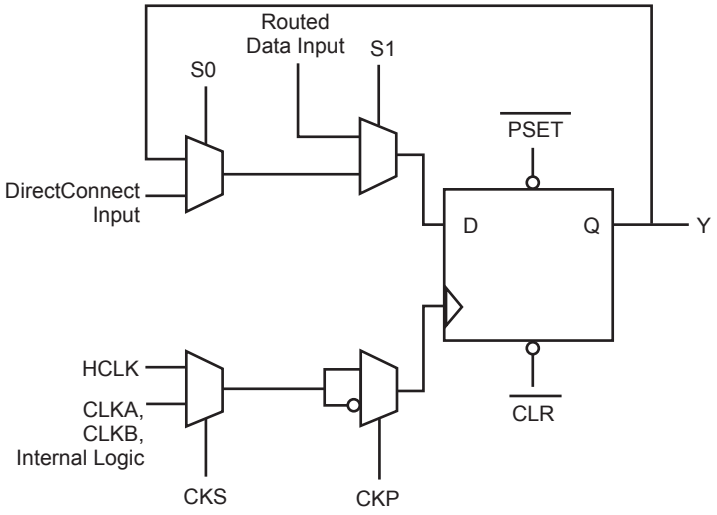


Figure 1-1 • **R-Cell**

Routing Resources

Clusters and SuperClusters can be connected through the use of two innovative local routing resources called FastConnect and DirectConnect, which enable extremely fast and predictable interconnection of modules within Clusters and SuperClusters (Figure 1-4). This routing architecture also dramatically reduces the number of antifuses required to complete a circuit, ensuring the highest possible performance.

DirectConnect is a horizontal routing resource that provides connections from a C-cell to its neighboring R-cell in a given SuperCluster. DirectConnect uses a hard-wired signal path requiring no programmable interconnection to achieve its fast signal propagation time of less than 0.1 ns (–P speed grade).

FastConnect enables horizontal routing between any two logic modules within a given SuperCluster and vertical routing with the SuperCluster immediately below it. Only one programmable connection is used in a FastConnect path, delivering maximum pin-to-pin propagation of 0.3 ns (–P speed grade).

In addition to DirectConnect and FastConnect, the architecture makes use of two globally oriented routing resources known as segmented routing and high-drive routing. The segmented routing structure of Microsemi provides a variety of track lengths for extremely fast routing between SuperClusters. The exact combination of track lengths and antifuses within each path is chosen by the fully automatic place-and-route software to minimize signal propagation delays.

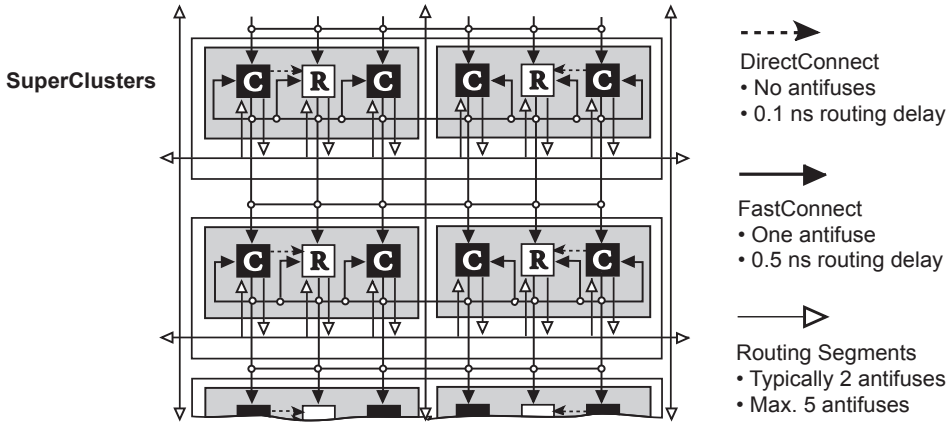


Figure 1-4 • **DirectConnect and FastConnect for SuperClusters**

Clock Resources

eX's high-drive routing structure provides three clock networks. The first clock, called HCLK, is hardwired from the HCLK buffer to the clock select MUX in each R-Cell. HCLK cannot be connected to combinational logic. This provides a fast propagation path for the clock signal, enabling the 3.9 ns clock-to-out (pad-to-pad) performance of the eX devices. The hard-wired clock is tuned to provide a clock skew of less than 0.1 ns worst case. If not used, the HCLK pin must be tied LOW or HIGH and must not be left floating. Figure 1-5 describes the clock circuit used for the constant load HCLK.

HCLK does not function until the fourth clock cycle each time the device is powered up to prevent false output levels due to any possible slow power-on-reset signal and fast start-up clock circuit. To activate HCLK from the first cycle, the TRST pin must be reserved in the Design software and the pin must be tied to GND on the board. (See the "TRST, I/O Boundary Scan Reset Pin" on page 1-32).

The remaining two clocks (CLKA, CLKB) are global routed clock networks that can be sourced from external pins or from internal logic signals (via the CLKINT routed clock buffer) within the eX device. CLKA and CLKB may be connected to sequential cells or to combinational logic. If CLKA or CLKB is sourced from internal logic signals, the external clock pin cannot be used for any other input and must be tied LOW or HIGH and must not float. Figure 1-6 describes the CLKA and CLKB circuit used in eX devices.

To exit the LP mode, the LP pin must be driven LOW for over 200 µs to allow for the charge pumps to power-up and device initialization can begin.

Table 1-3 illustrates the standby current of eX devices in LP mode.

Table 1-3 • **Standby Power of eX Devices in LP Mode Typical Conditions, V_{CCA}, V_{CCI} = 2.5 V, T_J = 25° C**

Product	Low Power Standby Current	Units
eX64	100	µA
eX128	111	µA
eX256	134	µA

Designer software is a place-and-route tool and provides a comprehensive suite of backend support tools for FPGA development. The Designer software includes timing-driven place-and-route, and a world-class integrated static timing analyzer and constraints editor. With the Designer software, a user can lock his/her design pins before layout while minimally impacting the results of place-and-route. Additionally, the back-annotation flow is compatible with all the major simulators and the simulation results can be cross-probed with Silicon Explorer II, Microsemi integrated verification and logic analysis tool. Another tool included in the Designer software is the SmartGen core generator, which easily creates popular and commonly used logic functions for implementation into your schematic or HDL design. Microsemi's Designer software is compatible with the most popular FPGA design entry and verification tools from companies such as Mentor Graphics, Synplicity, Synopsys, and Cadence Design Systems. The Designer software is available for both the Windows and UNIX operating systems.

Related Documents

Datasheet

eX Automotive Family FPGAs
www.microsemi.com/soc/documents/eX_Auto_DS.pdf

Application Notes

Maximizing Logic Utilization in eX, SX and SX-A FPGA Devices Using CC Macros
www.microsemi.com/soc/documents/CC_Macro_AN.pdf
Implementation of Security in Microsemi Antifuse FPGAs
www.microsemi.com/soc/documents/Antifuse_Security_AN.pdf
Microsemi eX, SX-A, and RT54SX-S I/Os
www.microsemi.com/soc/documents/antifuseIO_AN.pdf
Microsemi SX-A and RT54SX-S Devices in Hot-Swap and Cold-Sparing Applications
www.microsemi.com/soc/documents/HotSwapColdSparing_AN.pdf
Design For Low Power in Microsemi Antifuse FPGAs
www.microsemi.com/soc/documents/Low_Power_AN.pdf
Programming Antifuse Devices
www.microsemi.com/soc/documents/AntifuseProgram_AN.pdf

User Guides

Silicon Sculptor II User's Guide
www.microsemi.com/soc/documents/SiliSculptII_Sculpt3_ug.pdf

Miscellaneous

Libero IDE flow
www.microsemi.com/soc/products/tools/libero/flow.html

2.5 V LVCMOS2 Electrical Specifications

Symbol	Parameter		Commercial		Industrial		Units
			Min.	Max.	Min.	Max.	
VOH	VCCI = MIN, VI = VIH or VIL	(IOH = –100 mA)	2.1		2.1		V
	VCCI = MIN, VI = VIH or VIL	(IOH = –1 mA)	2.0		2.0		V
	VCCI = MIN, VI = VIH or VIL	(IOH = –2 mA)	1.7		1.7		V
VOL	VCCI = MIN, VI = VIH or VIL	(IOL = 100 mA)		0.2		0.2	V
	VCCI = MIN, VI = VIH or VIL	(IOL = 1mA)		0.4		0.4	V
	VCCI = MIN,VI = VIH or VIL	(IOL = 2 mA)		0.7		0.7	V
VIL	Input Low Voltage, VOUT ≤ VOL (max.)		–0.3	0.7	-0.3	0.7	V
VIH	Input High Voltage, VOUT ≥ VOH (min.)		1.7	VCCI + 0.3	1.7	VCCI + 0.3	V
IIL/ IIH	Input Leakage Current, VIN = VCCI or GND		–10	10	–10	10	µA
IOZ	3-State Output Leakage Current, VOUT = VCCI or GND		–10	10	–10	10	µA
t_R, $t_{F1,2}$	Input Transition Time			10		10	ns
C_{IO}	I/O Capacitance			10		10	pF
ICC[3,4]	Standby Current			1.0		3.0	mA
IV Curve	Can be derived from the IBIS model at www.microsemi.com/soc/custsup/models/ibis.html.						

Notes:

1. *t_R is the transition time from 0.7 V to 1.7 V.*
2. *t_F is the transition time from 1.7 V to 0.7 V.*
3. *I_{CC} max Commercial –F = 5.0 mA*
4. *$I_{CC} = I_{CCI} + I_{CCA}$*

Power Dissipation

Power consumption for eX devices can be divided into two components: static and dynamic.

Static Power Component

The power due to standby current is typically a small component of the overall power. Typical standby current for eX devices is listed in the Table 1-11 on page 1-16. For example, the typical static power for eX128 at 3.3 V V_{CCI} is:

ICC * VCCA = 795 µA x 2.5 V = 1.99 mW

Dynamic Power Component

Power dissipation in CMOS devices is usually dominated by the dynamic power dissipation. This component is frequency-dependent and a function of the logic and the external I/O. Dynamic power dissipation results from charging internal chip capacitance due to PC board traces and load device inputs. An additional component of the dynamic power dissipation is the totem pole current in the CMOS transistor pairs. The net effect can be associated with an equivalent capacitance that can be combined with frequency and voltage to represent dynamic power dissipation.

Dynamic power dissipation = CEQ * $VCCA^2$ x F

where:

CEQ = Equivalent capacitance

F = switching frequency

Equivalent capacitance is calculated by measuring ICCA at a specified frequency and voltage for each circuit component of interest. Measurements have been made over a range of frequencies at a fixed value of VCC. Equivalent capacitance is frequency-independent, so the results can be used over a wide range of operating conditions. Equivalent capacitance values are shown below.

CEQ Values for eX Devices

Combinatorial modules (Ceqcm)	1.70 pF
Sequential modules (Ceqsm)	1.70 pF
Input buffers (Ceqi)	1.30 pF
Output buffers (Ceqo)	7.40 pF
Routed array clocks (Ceqcr)	1.05 pF

The variable and fixed capacitance of other device components must also be taken into account when estimating the dynamic power dissipation.

Table 1-12 shows the capacitance of the clock components of eX devices.

Table 1-12 • **Capacitance of Clock Components of eX Devices**

	eX64	eX128	eX256
Dedicated array clock – variable (Ceqhv)	0.85 pF	0.85 pF	0.85 pF
Dedicated array clock – fixed (Ceqhf)	18.00 pF	20.00 pF	25.00 pF
Routed array clock A (r1)	23.00 pF	28.00 pF	35.00 pF
Routed array clock B (r2)	23.00 pF	28.00 pF	35.00 pF

The estimation of the dynamic power dissipation is a piece-wise linear summation of the power dissipation of each component.

$$\text{Dynamic power dissipation} = VCCA^2 * [(m_c * C_{eqcm} * fm_C)_{\text{Comb Modules}} + (m_s * C_{eqsm} * fm_S)_{\text{Seq Modules}} + (n * C_{eqi} * fn)_{\text{Input Buffers}} + (0.5 * (q1 * C_{eqcr} * fq1) + (r1 * fq1))_{RCLKA} + (0.5 * (q2 * C_{eqcr} * fq2) + (r2 * fq2))_{RCLKB} + (0.5 * (s1 * C_{eqhv} * fs1)+(C_{eqhf} * fs1))_{HCLK}] + V_{CCI}^2 * [(p * (C_{eqo} + C_L) * fp)_{\text{Output Buffers}}]$$

where:

m_c = Number of combinatorial cells switching at frequency fm, typically 20% of C-cells
m_s = Number of sequential cells switching at frequency fm, typically 20% of R-cells
n = Number of input buffers switching at frequency fn, typically number of inputs / 4
p = Number of output buffers switching at frequency fp, typically number of outputs / 4
q1 = Number of R-cells driven by routed array clock A
q2 = Number of R-cells driven by routed array clock B
r1 = Fixed capacitance due to routed array clock A
r2 = Fixed capacitance due to routed array clock B
s1 = Number of R-cells driven by dedicated array clock
C_{eqcm} = Equivalent capacitance of combinatorial modules
C_{eqsm} = Equivalent capacitance of sequential modules
C_{eqi} = Equivalent capacitance of input buffers
C_{eqcr} = Equivalent capacitance of routed array clocks
C_{eqhv} = Variable capacitance of dedicated array clock
C_{eqhf} = Fixed capacitance of dedicated array clock
C_{eqo} = Equivalent capacitance of output buffers
C_L = Average output loading capacitance, typically 10 pF
fm_c = Average C-cell switching frequency, typically F/10
fm_s = Average R-cell switching frequency, typically F/10
fn = Average input buffer switching frequency, typically F/5
fp = Average output buffer switching frequency, typically F/5
fq1 = Frequency of routed clock A
fq2 = Frequency of routed clock B
fs1 = Frequency of dedicated array clock

The eX, SX-A and RTSX-S Power Calculator can be used to estimate the total power dissipation (static and dynamic) of eX devices: www.microsemi.com/soc/techdocs/calculators.aspx.

Timing Characteristics

Timing characteristics for eX devices fall into three categories: family-dependent, device-dependent, and design-dependent. The input and output buffer characteristics are common to all eX family members. Internal routing delays are device-dependent. Design dependency means actual delays are not determined until after placement and routing of the user's design are complete. Delay values may then be determined by using the Timer utility or performing simulation with post-layout delays.

Critical Nets and Typical Nets

Propagation delays are expressed only for typical nets, which are used for initial design performance evaluation. Critical net delays can then be applied to the most timing critical paths. Critical nets are determined by net property assignment prior to placement and routing. Up to six percent of the nets in a design may be designated as critical.

Long Tracks

Some nets in the design use long tracks. Long tracks are special routing resources that span multiple rows, columns, or modules. Long tracks employ three to five antifuse connections. This increases capacitance and resistance, resulting in longer net delays for macros connected to long tracks. Typically, no more than six percent of nets in a fully utilized device require long tracks. Long tracks contribute approximately 4 ns to 8.4 ns delay. This additional delay is represented statistically in higher fanout routing delays.

Timing Derating

eX devices are manufactured with a CMOS process. Therefore, device performance varies according to temperature, voltage, and process changes. Minimum timing parameters reflect maximum operating voltage, minimum operating temperature, and best-case processing. Maximum timing parameters reflect minimum operating voltage, maximum operating temperature, and worst-case processing.

Temperature and Voltage Derating Factors

***Table 1-16* • Temperature and Voltage Derating Factors**
(Normalized to Worst-Case Commercial, T_J = 70°C, VCCA = 2.3V)

VCCA	Junction Temperature (T_J)						
	–55	**–40**	**0**	**25**	**70**	**85**	**125**
2.3	0.79	0.80	0.87	0.88	1.00	1.04	1.13
2.5	0.74	0.74	0.81	0.83	0.93	0.97	1.06
2.7	0.69	0.70	0.76	0.78	0.88	0.91	1.00

Pin Description

CLKA/B **Routed Clock A and B**

These pins are clock inputs for clock distribution networks. Input levels are compatible with standard TTL or LVTTL specifications. The clock input is buffered prior to clocking the R-cells. If not used, this pin must be set LOW or HIGH on the board. It must not be left floating.

GND **Ground**

LOW supply voltage.

HCLK **Dedicated (Hardwired) Array Clock**

This pin is the clock input for sequential modules. Input levels are compatible with standard TTL or LVTTL specifications. This input is directly wired to each R-cell and offers clock speeds independent of the number of R-cells being driven. If not used, this pin must be set LOW or HIGH on the board. It must not be left floating.

I/O **Input/Output**

The I/O pin functions as an input, output, tristate, or bidirectional buffer. Based on certain configurations, input and output levels are compatible with standard TTL or LVTTL specifications. Unused I/O pins are automatically tristated by the Designer software.

LP **Low Power Pin**

Controls the low power mode of the eX devices. The device is placed in the low power mode by connecting the LP pin to logic HIGH. In low power mode, all I/Os are tristated, all input buffers are turned OFF, and the core of the device is turned OFF. To exit the low power mode, the LP pin must be set LOW. The device enters the low power mode 800 ns after the LP pin is driven to a logic HIGH. It will resume normal operation 200 µs after the LP pin is driven to a logic LOW. LP pin should not be left floating. Under normal operating condition it should be tied to GND via 10 kΩ resistor.

NC **No Connection**

This pin is not connected to circuitry within the device. These pins can be driven to any voltage or can be left floating with no effect on the operation of the device.

PRA/PRB, I/O **Probe A/B**

The Probe pin is used to output data from any user-defined design node within the device. This diagnostic pin can be used independently or in conjunction with the other probe pin to allow real-time diagnostic output of any signal path within the device. The Probe pin can be used as a user-defined I/O when verification has been completed. The pin's probe capabilities can be permanently disabled to protect programmed design confidentiality.

TCK, I/O **Test Clock**

Test clock input for diagnostic probe and device programming. In flexible mode, TCK becomes active when the TMS pin is set LOW (refer to Table 1-4 on page 1-10). This pin functions as an I/O when the boundary scan state machine reaches the "logic reset" state.

TDI, I/O **Test Data Input**

Serial input for boundary scan testing and diagnostic probe. In flexible mode, TDI is active when the TMS pin is set LOW (refer to Table 1-4 on page 1-10). This pin functions as an I/O when the boundary scan state machine reaches the "logic reset" state.

TDO, I/O **Test Data Output**

Serial output for boundary scan testing. In flexible mode, TDO is active when the TMS pin is set LOW (refer to Table 1-4 on page 1-10). This pin functions as an I/O when the boundary scan state machine reaches the "logic reset" state. When Silicon Explorer is being used, TDO will act as an output when the "checksum" command is run. It will return to user I/O when "checksum" is complete.

TMS **Test Mode Select**

The TMS pin controls the use of the IEEE 1149.1 Boundary scan pins (TCK, TDI, TDO, TRST). In flexible mode when the TMS pin is set LOW, the TCK, TDI, and TDO pins are boundary scan pins (refer to Table 1-4 on page 1-10). Once the boundary scan pins are in test mode, they will remain in that mode until the internal boundary scan state machine reaches the “logic reset” state. At this point, the boundary scan pins will be released and will function as regular I/O pins. The “logic reset” state is reached five TCK cycles after the TMS pin is set HIGH. In dedicated test mode, TMS functions as specified in the IEEE 1149.1 specifications.

TRST, I/O **Boundary Scan Reset Pin**

Once it is configured as the JTAG Reset pin, the TRST pin functions as an active-low input to asynchronously initialize or reset the boundary scan circuit. The TRST pin is equipped with an internal pull-up resistor. This pin functions as an I/O when the **Reserve JTAG Reset** Pin is not selected in the Designer software.

VCCI **Supply Voltage**

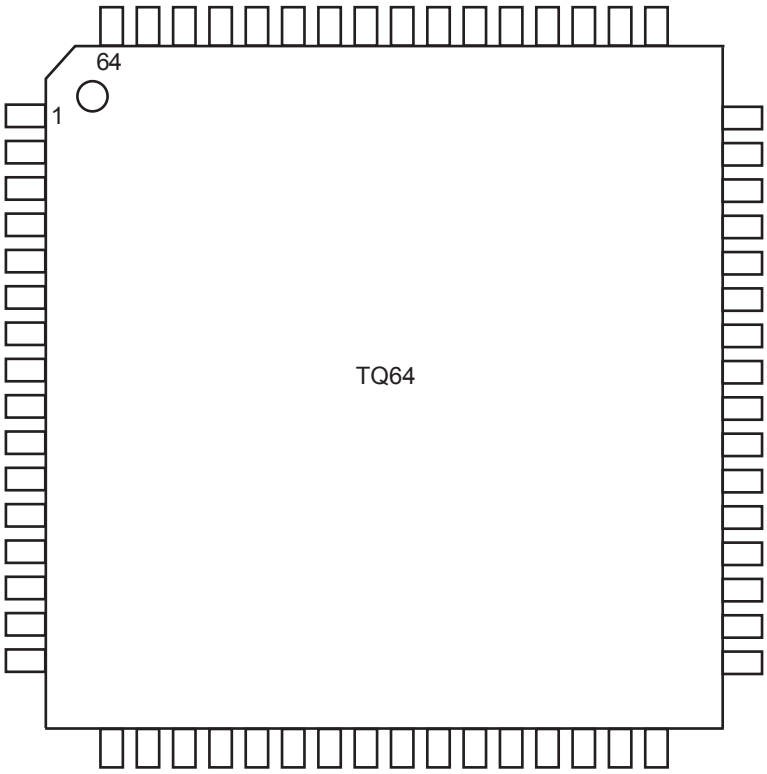
Supply voltage for I/Os.

VCCA **Supply Voltage**

Supply voltage for Array.

2 – Package Pin Assignments

TQ64



Note: For Package Manufacturing and Environmental information, visit Resource center at www.microsemi.com/soc/products/rescenter/package/index.html.

TQ100			
Pin Number	eX64 Function	eX128 Function	eX256 Function
1	GND	GND	GND
2	TDI, I/O	TDI, I/O	TDI, I/O
3	NC	NC	I/O
4	NC	NC	I/O
5	NC	NC	I/O
6	I/O	I/O	I/O
7	TMS	TMS	TMS
8	VCCI	VCCI	VCCI
9	GND	GND	GND
10	NC	I/O	I/O
11	NC	I/O	I/O
12	I/O	I/O	I/O
13	NC	I/O	I/O
14	I/O	I/O	I/O
15	NC	I/O	I/O
16	TRST, I/O	TRST, I/O	TRST, I/O
17	NC	I/O	I/O
18	I/O	I/O	I/O
19	NC	I/O	I/O
20	VCCI	VCCI	VCCI
21	I/O	I/O	I/O
22	NC	I/O	I/O
23	NC	NC	I/O
24	NC	NC	I/O
25	I/O	I/O	I/O
26	I/O	I/O	I/O
27	I/O	I/O	I/O
28	I/O	I/O	I/O
29	I/O	I/O	I/O
30	I/O	I/O	I/O
31	I/O	I/O	I/O
32	I/O	I/O	I/O
33	I/O	I/O	I/O
34	PRB, I/O	PRB, I/O	PRB, I/O
35	VCCA	VCCA	VCCA

TQ100			
Pin Number	eX64 Function	eX128 Function	eX256 Function
36	GND	GND	GND
37	NC	NC	NC
38	I/O	I/O	I/O
39	HCLK	HCLK	HCLK
40	I/O	I/O	I/O
41	I/O	I/O	I/O
42	I/O	I/O	I/O
43	I/O	I/O	I/O
44	VCCI	VCCI	VCCI
45	I/O	I/O	I/O
46	I/O	I/O	I/O
47	I/O	I/O	I/O
48	I/O	I/O	I/O
49	TDO, I/O	TDO, I/O	TDO, I/O
50	NC	I/O	I/O
51	GND	GND	GND
52	NC	NC	I/O
53	NC	NC	I/O
54	NC	NC	I/O
55	I/O	I/O	I/O
56	I/O	I/O	I/O
57	VCCA	VCCA	VCCA
58	VCCI	VCCI	VCCI
59	NC	I/O	I/O
60	I/O	I/O	I/O
61	NC	I/O	I/O
62	I/O	I/O	I/O
63	NC	I/O	I/O
64	I/O	I/O	I/O
65	NC	I/O	I/O
66	I/O	I/O	I/O
67	VCCA	VCCA	VCCA
68	GND/LP	GND/LP	GND/LP
69	GND	GND	GND
70	I/O	I/O	I/O

*Note: *Please read the LP pin descriptions for restrictions on their use.*

TQ100			
Pin Number	**eX64 Function**	**eX128 Function**	**eX256 Function**
71	I/O	I/O	I/O
72	NC	I/O	I/O
73	NC	NC	I/O
74	NC	NC	I/O
75	NC	NC	I/O
76	NC	I/O	I/O
77	I/O	I/O	I/O
78	I/O	I/O	I/O
79	I/O	I/O	I/O
80	I/O	I/O	I/O
81	I/O	I/O	I/O
82	VCCI	VCCI	VCCI
83	I/O	I/O	I/O
84	I/O	I/O	I/O
85	I/O	I/O	I/O
86	I/O	I/O	I/O
87	CLKA	CLKA	CLKA
88	CLKB	CLKB	CLKB
89	NC	NC	NC
90	VCCA	VCCA	VCCA
91	GND	GND	GND
92	PRA, I/O	PRA, I/O	PRA, I/O
93	I/O	I/O	I/O
94	I/O	I/O	I/O
95	I/O	I/O	I/O
96	I/O	I/O	I/O
97	I/O	I/O	I/O
98	I/O	I/O	I/O
99	I/O	I/O	I/O
100	TCK, I/O	TCK, I/O	TCK, I/O

*Note: *Please read the LP pin descriptions for restrictions on their use.*

Revision	Changes	Page
v4.0 (continued)	The "Flexible Mode" section was updated.	1-10
	Table 1-5 •Boundary-Scan Pin Configurations and Functions is new.	1-11
	The "TRST Pin" section was updated.	1-11
	The "Probing Capabilities" section is new.	1-12
	The "Programming" section was updated.	1-12
	The "Probing Capabilities" section was updated.	1-12
	The "Silicon Explorer II Probe" section was updated.	1-12
	The "Design Considerations" section was updated.	1-13
	The "Development Tool Support" section was updated.	1-13
	The "Absolute Maximum Ratings*" section was updated.	1-16
	The "Temperature and Voltage Derating Factors" section was updated.	1-26
	The "TDI, I/O Test Data Input" section was updated.	1-31
	The "TDO, I/O Test Data Output" section was updated.	1-31
	The "TMS Test Mode Select" section was updated.	1-32
	The "TRST, I/O Boundary Scan Reset Pin" section was updated.	1-32
	All VSV pins were changed to VCCA. The change affected the following pins: 64-Pin TQFP – Pin 36 100-Pin TQFP – Pin 57 49-Pin CSP – Pin D5 128-Pin CSP– Pin H11 and Pin J1 for eX256 180-Pin CSP – Pins J12 and K2	
v3.0	The "Recommended Operating Conditions" section has been changed.	1-16
	The "3.3 V LVTTL Electrical Specifications" section has been updated.	1-18
	The "5.0 V TTL Electrical Specifications" section has been updated.	1-18
	The "Total Dynamic Power (mW)" section is new.	1-9
	The "System Power at 5%, 10%, and 15% Duty Cycle" section is new.	1-9
	The "eX Timing Model" section has been updated.	1-22
v2.0.1	The I/O Features table, Table 1-2 on page 1-6, was updated.	1-6
	The table, "Standby Power of eX Devices in LP Mode Typical Conditions, VCCA, VCCI = 2.5 V, TJ = 25° C" section, was updated.	1-7
	"Typical eX Standby Current at 25°C" section is a new table.	1-16
	The table in the section, "Package Thermal Characteristics" section has been updated for the 49-Pin CSP.	1-21
	The "eX Timing Model" section has been updated.	1-22
	The timing numbers found in, "eX Family Timing Characteristics" section have been updated.	1-27
	The V_{SV} pin has been added to the "Pin Description" section.	1-31
	Please see the following pin tables for the V_{SV} pin and an important footnote including the pin: "TQ64", "TQ100", "128-Pin CSP", and "180-Pin CSP".	2-1, 2-3, 2-6, 2-11
	The figure, "TQ64" section has been updated.	2-1

Revision	Changes	Page
Advance v0.4	In the Product Profile, the Maximum User I/Os for eX64 was changed to 84.	1-I
	In the Product Profile table, the Maximum User I/Os for eX128 was changed to 100.	1-I
Advance v0.3	The Mechanical Drawings section has been removed from the data sheet. The mechanical drawings are now contained in a separate document, “Package Characteristics and Mechanical Drawings,” available on the Actel web site.	
	A new section describing "Clock Resources"has been added.	1-3
	A new table describing "I/O Features"has been added.	1-6
	The "Pin Description"section has been updated and clarified.	1-31
	The original Electrical Specifications table was separated into two tables (2.5V and 3.3/5.0V). In both tables, several different currents are specified for V_{OH} and V_{OL}.	Page 8 and 9
	A new table listing 2.5V low power specifications and associated power graphs were added.	page 9
	Pin functions for eX256 TQ100 have been added to the "TQ100"table.	2-3
	A CS49 pin drawing and pin assignment table including eX64 and eX128 pin functions have been added.	page 26
	A CS128 pin drawing and pin assignment table including eX64, eX128, and eX256 pin functions have been added.	pages 26-27
	A CS180 pin drawing and pin assignment table for eX256 pin functions have been added.	pages 27, 31
Advance v0.2	The following table note was added to the eX Timing Characteristics table for clarification: Clock skew improves as the clock network becomes more heavily loaded.	pages 14-15

Datasheet Categories

Categories

In order to provide the latest information to designers, some datasheet parameters are published before data has been fully characterized from silicon devices. The data provided for a given device, as highlighted in the "eX Device Status" table on page II, is designated as either "Product Brief," "Advance," "Preliminary," or "Production." The definitions of these categories are as follows:

Product Brief

The product brief is a summarized version of a datasheet (advance or production) and contains general product information. This document gives an overview of specific device and family information.

Advance

This version contains initial estimated information based on simulation, other products, devices, or speed grades. This information can be used as estimates, but not for production. This label only applies to the DC and Switching Characteristics chapter of the datasheet and will only be used when the data has not been fully characterized.

Preliminary

The datasheet contains information based on simulation and/or initial characterization. The information is believed to be correct, but changes are possible.

Production

This version contains information that is considered to be final.

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