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Details

Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	64MHz
Connectivity	I ² C, LINbus, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, POR, PWM, WDT
Number of I/O	60
Program Memory Size	32KB (16K x 16)
Program Memory Type	FLASH
EEPROM Size	1K x 8
RAM Size	2K x 8
Voltage - Supply (Vcc/Vdd)	1.8V ~ 3.6V
Data Converters	A/D 45x10b; D/A 1x5b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 85°C (TA)
Mounting Type	Surface Mount
Package / Case	64-VFQFN Exposed Pad
Supplier Device Package	64-VQFN (9x9)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic18lf65k40t-i-mr

TABLE 1: 64-PIN ALLOCATION TABLE (PIC18(L)F65/66K40) (CONTINUED)

I/O ⁽²⁾	64-Pin TQFP, QFN	A/D	DAC	Comparator	Timers	CCP and PWM	CWG	ZCD	SMT	Clock Reference (CLKR)	Interrupt	EUSART	DSM	MSSP	Basic
RC1	29	—	—	—	T6IN ⁽¹⁾	—	—	—	—	—	IOCC1	RX4 ⁽¹⁾ DT4 ⁽¹⁾	—	—	SOSCI
RC2	33	—	—	—	—	—	CWG1IN ⁽¹⁾	—	—	—	IOCC2	—	—	—	—
RC3	34	—	—	—	—	—	—	—	—	—	IOCC3	—	—	SCL1 ^(3,4) SCK1 ⁽¹⁾	—
RC4	35	—	—	—	—	—	—	—	—	—	IOCC4	—	—	SDA1 ^(3,4) SDI1 ⁽¹⁾	—
RC5	36	—	—	—	—	—	—	—	—	—	IOCC5	—	—	—	—
RC6	31	—	—	—	—	—	—	—	—	—	IOCC6	CK1 ⁽³⁾	—	—	—
RC7	32	—	—	—	—	—	—	—	—	—	IOCC7	RX1 ⁽¹⁾ DT1 ⁽³⁾	—	—	—
RD0	58	AND0	—	—	—	—	—	—	—	—	—	—	—	—	—
RD1	55	AND1	—	—	T5CKI ⁽¹⁾ T7G ⁽¹⁾	—	—	—	—	—	—	—	—	—	—
RD2	54	AND2	—	—	—	—	—	—	—	—	—	—	—	—	—
RD3	53	AND3	—	—	—	—	—	—	—	—	—	—	MDCARL ⁽¹⁾	—	—
RD4	52	AND4	—	—	—	—	—	—	—	—	—	—	MDCARH ⁽¹⁾	—	—
RD5	51	AND5	—	—	—	—	—	—	—	—	—	—	MDSRC ⁽¹⁾	SDA2 ^(3,4) SDI2 ⁽¹⁾	—
RD6	50	AND6	—	—	—	—	—	—	—	—	—	—	—	SCL2 ^(3,4) SCK2 ⁽¹⁾	—
RD7	49	AND7	—	—	—	—	—	—	—	—	—	—	—	SS2 ⁽¹⁾	—
RE0	2	ANE0	—	—	—	—	—	—	—	—	IOCE0	CK3 ⁽³⁾	—	—	—
RE1	1	ANE1	—	—	—	—	—	—	—	—	IOCE1	RX3 ⁽¹⁾ DT3 ⁽³⁾	—	—	—
RE2	64	ANE2	—	—	—	—	—	—	—	—	IOCE2	CK5 ⁽³⁾	—	—	—
RE3	63	ANE3	—	—	—	—	—	—	—	—	IOCE3	RX5 ⁽¹⁾ DT5 ⁽³⁾	—	—	—
RE4	62	ANE4	—	—	T4IN ⁽¹⁾	CCP2 ⁽¹⁾	—	—	—	—	IOCE4	—	—	—	—

- Note**
- 1: This is a PPS remappable input signal. The input function may be moved from the default location shown to one of several other PORTx pins. Refer to Table 17-1 for details on which PORT pins may be used for this signal.
 - 2: All output signals shown in this row are PPS remappable. These signals may be mapped to output onto one of several PORTx pin options as described in Register 17-2
 - 3: This is a bidirectional signal. For normal module operation, the firmware should map this signal to the same pin in both the PPS input and PPS output registers.
 - 4: These pins are configured for I²C™ logic levels; The SCLx/SDAx signals may be assigned to any of these pins. PPS assignments to the other pins (e.g., RB1) will operate, but input logic levels will be standard TTL/ST as selected by the INLV register, instead of the I²C specific or SMBus input buffer thresholds.

FIGURE 4-3: QUARTZ CRYSTAL OPERATION (LP, XT OR HS MODE)

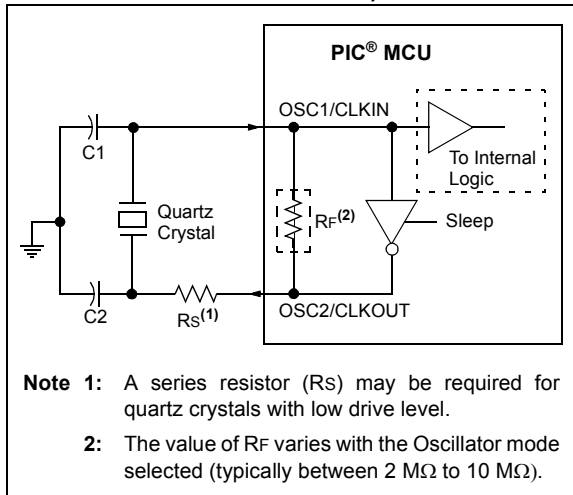
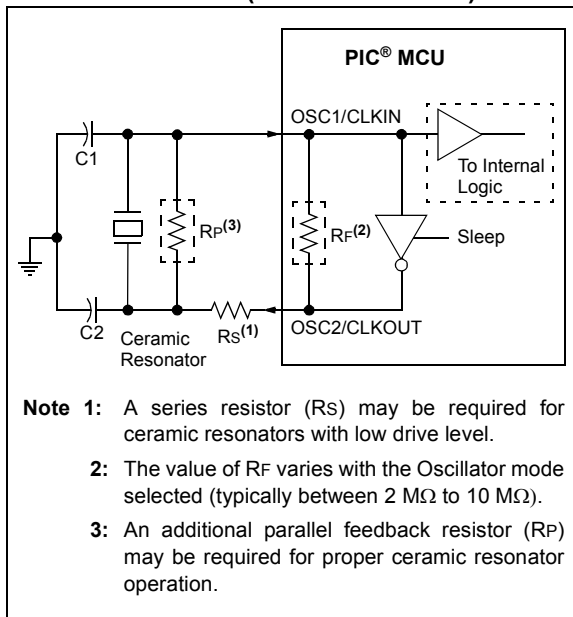


FIGURE 4-4: CERAMIC RESONATOR OPERATION (XT OR HS MODE)



4.3.1.3 Oscillator Start-up Timer (OST)

If the oscillator module is configured for LP, XT or HS modes, the Oscillator Start-up Timer (OST) counts 1024 oscillations from OSC1. This occurs following a Power-on Reset (POR), or a wake-up from Sleep. The OST ensures that the oscillator circuit, using a quartz crystal resonator or ceramic resonator, has started and is providing a stable system clock to the oscillator module.

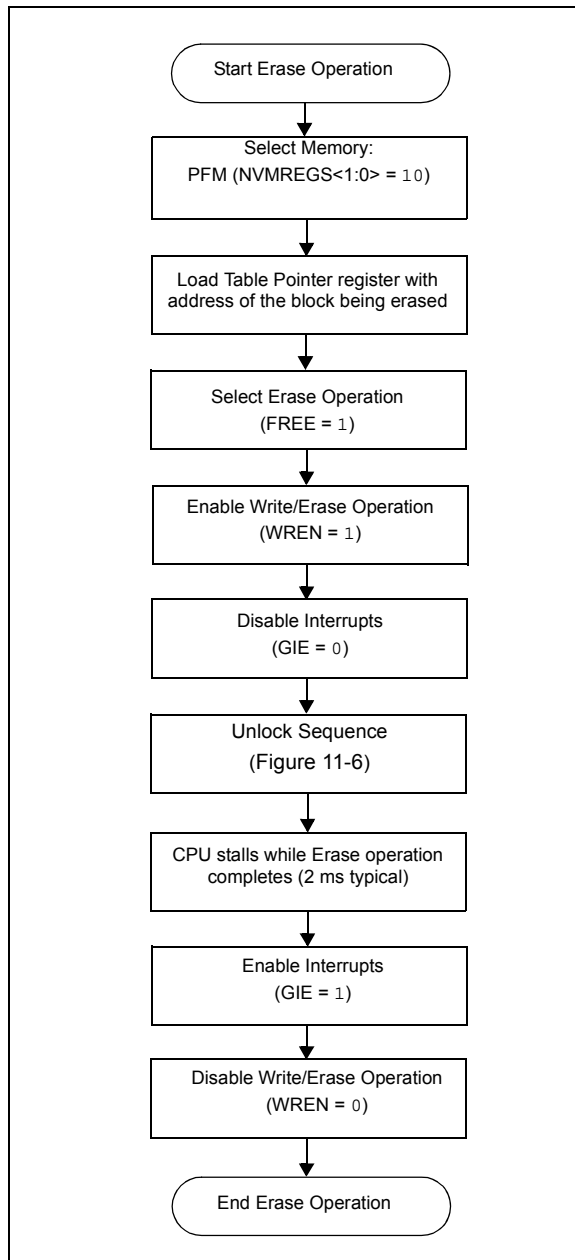
4.3.1.4 4x PLL

The oscillator module contains a 4x PLL that can be used with the external clock sources to provide a system clock source. The input frequency for the PLL must fall within specifications. See the PLL Clock Timing Specifications in Table 37-9.

The PLL can be enabled for use by one of two methods:

1. Program the RSTOSC bits in the Configuration Word 1 to 010 (enable EXTOSC with 4x PLL).
2. Write the NOSC bits in the OSCCON1 register to 010 (enable EXTOSC with 4x PLL).

FIGURE 11-7: PFM ROW ERASE FLOWCHART



11.1.6 WRITING TO PROGRAM FLASH MEMORY

The programming write block size is described in Table 11-3. Word or byte programming is not supported. Table writes are used internally to load the holding registers needed to program the Flash memory. There are only as many holding registers as there are bytes in a write block. Refer to Table 11-3 for write latch size.

Since the table latch (TABLAT) is only a single byte, the TBLWT instruction needs to be executed multiple times for each programming operation. The write protection state is ignored for this operation. All of the table write operations will essentially be short writes because only the holding registers are written. NVMIF is not affected while writing to the holding registers.

After all the holding registers have been written, the programming operation of that block of memory is started by configuring the NVMCON1 register for a program memory write and performing the long write sequence.

If the PFM address in the TBLPTR is write-protected or if TBLPTR points to an invalid location, the WR bit is cleared without any effect and the WREER is signaled.

The long write is necessary for programming the internal Flash. CPU operation is suspended during a long write cycle and resumes when the operation is complete. The long write operation completes in one instruction cycle. When complete, WR is cleared in hardware and NVMIF is set and an interrupt will occur if NVMIE is also set. The latched data is reset to all '1s'. WREN is not changed.

The internal programming timer controls the write time. The write/erase voltages are generated by an on-chip charge pump, rated to operate over the voltage range of the device.

Note: The default value of the holding registers on device Resets and after write operations is FFh. A write of FFh to a holding register does not modify that byte. This means that individual bytes of program memory may be modified, provided that the change does not attempt to change any bit from a '0' to a '1'. When modifying individual bytes, it is not necessary to load all holding registers before executing a long write operation.

REGISTER 13-6: CRCACCL: CRC ACCUMULATOR LOW BYTE REGISTER

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
ACC<7:0>							
bit 7				bit 0			

Legend:

R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 **ACC<7:0>**: CRC Accumulator Register bits
Writing to this register writes to the CRC accumulator register through the CRC write bus. Reading from this register reads the CRC accumulator.

REGISTER 13-7: CRCSHIFTH: CRC SHIFT HIGH BYTE REGISTER

R-0	R-0	R-0	R-0	R-0	R-0	R-0	R-0
SHIFT<15:8>							
bit 7				bit 0			

Legend:

R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 **SHIFT<15:8>**: CRC Shifter Register bits
Reading from this register reads the CRC Shifter.

REGISTER 13-8: CRCSHIFTL: CRC SHIFT LOW BYTE REGISTER

R-0	R-0	R-0	R-0	R-0	R-0	R-0	R-0
SHIFT<7:0>							
bit 7				bit 0			

Legend:

R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 **SHIFT<7:0>**: CRC Shifter Register bits
Reading from this register reads the CRC Shifter.

13.3 CRC Functional Overview

The CRC module can be used to detect bit errors in the Flash memory using the built-in memory scanner or through user input RAM memory. The CRC module can accept up to a 16-bit polynomial with up to a 16-bit seed value. A CRC calculated check value (or checksum) will then be generated into the CRCACC<15:0> registers for user storage. The CRC module uses an XOR shift register implementation to perform the polynomial division required for the CRC calculation.

EXAMPLE 13-1: CRC EXAMPLE

Rev. 10-000206A
1/8/2014

CRC-16-ANSI

$$x^{16} + x^{15} + x^2 + 1 \text{ (17 bits)}$$

Standard 16-bit representation = 0x8005

CRCXORH = 0b10000000
CRCXORL = 0b0000010- ⁽¹⁾

Data Sequence:
0x55, 0x66, 0x77, 0x88

DLEN = 0b0111
PLEN = 0b1111

Data entered into the CRC:
SHIFTM = 0:
01010101 01100110 01110111 10001000

SHIFTM = 1:
10101010 01100110 11101110 00010001

Check Value (ACCM = 1):

SHIFTM = 0: 0x32D6
CRCACCH = 0b00110010
CRCACCL = 0b11010110

SHIFTM = 1: 0x6BA2
CRCACCH = 0b01101011
CRCACCL = 0b10100010

Note 1: Bit 0 is unimplemented. The LSb of any CRC polynomial is always '1' and will always be treated as a '1' by the CRC for calculating the CRC check value. This bit will be read in software as a '0'.

REGISTER 14-19: PIE7: PERIPHERAL INTERRUPT ENABLE REGISTER 7

U-0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
—	—	—	CCP5IE	CCP4IE	CCP3IE	CCP2IE	CCP1IE
bit 7							bit 0

Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

-n = Value at POR

'1' = Bit is set

'0' = Bit is cleared

x = Bit is unknown

bit 7-5	Unimplemented: Read as '0'
bit 4	CCP5IE: ECCP5 Interrupt Enable bit 1 = Enabled 0 = Disabled
bit 3	CCP4IE: ECCP4 Interrupt Enable bit 1 = Enabled 0 = Disabled
bit 2	CCP3IE: ECCP3 Interrupt Enable bit 1 = Enabled 0 = Disabled
bit 1	CCP2IE: ECCP2 Interrupt Enable bit 1 = Enabled 0 = Disabled
bit 0	CCP1IE: ECCP1 Interrupt Enable bit 1 = Enabled 0 = Disabled

REGISTER 14-28: IPR6: PERIPHERAL INTERRUPT PRIORITY REGISTER 6

U-0	U-0	U-0	U-0	R/W-1/1	R/W-1/1	R/W-1/1	R/W-1/1
—	—	—	—	TMR7GIP	TMR5GIP	TMR3GIP	TMR1GIP
bit 7				bit 0			

Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

-n = Value at POR

'1' = Bit is set

'0' = Bit is cleared

x = Bit is unknown

bit 7-4 **Unimplemented:** Read as '0'

bit 3 **TMR7GIP:** TMR7 Gate Interrupt Priority bit
1 = High priority
0 = Low priority

bit 2 **TMR5GIP:** TMR5 Gate Interrupt Priority bit
1 = High priority
0 = Low priority

bit 1 **TMR3GIP:** TMR3 Gate Interrupt Priority bit
1 = High priority
0 = Low priority

bit 0 **TMR1GIP:** TMR1 Gate Interrupt Priority bit
1 = High priority
0 = Low priority

REGISTER 18-2: T0CON1: TIMER0 CONTROL REGISTER 1

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
T0CS<2:0>			T0ASYNC	T0CKPS<3:0>			
bit 7							bit 0

Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

u = Bit is unchanged

x = Bit is unknown

-n/n = Value at POR and BOR/Value at all other Resets

'1' = Bit is set

'0' = Bit is cleared

bit 7-5 **T0CS<2:0>**: Timer0 Clock Source Select bits

111 = Reserved

110 = Reserved

101 = SOSC

100 = LFINTOSC

011 = HFINTOSC

010 = Fosc/4

001 = Pin selected by T0CKIPPS (Inverted)

000 = Pin selected by T0CKIPPS (Non-inverted)

bit 4 **T0ASYNC**: TMR0 Input Asynchronization Enable bit

1 = The input to the TMR0 counter is not synchronized to system clocks

0 = The input to the TMR0 counter is synchronized to Fosc/4

bit 3-0 **T0CKPS<3:0>**: Prescaler Rate Select bit

1111 = 1:32768

1110 = 1:16384

1101 = 1:8192

1100 = 1:4096

1011 = 1:2048

1010 = 1:1024

1001 = 1:512

1000 = 1:256

0111 = 1:128

0110 = 1:64

0101 = 1:32

0100 = 1:16

0011 = 1:8

0010 = 1:4

0001 = 1:2

0000 = 1:1

TABLE 21-3: EXAMPLE PWM FREQUENCIES AND RESOLUTIONS (Fosc = 20 MHz)

PWM Frequency	1.22 kHz	4.88 kHz	19.53 kHz	78.12 kHz	156.3 kHz	208.3 kHz
Timer Prescale	16	4	1	1	1	1
PR2 Value	0xFF	0xFF	0xFF	0x3F	0x1F	0x17
Maximum Resolution (bits)	10	10	10	8	7	6.6

TABLE 21-4: EXAMPLE PWM FREQUENCIES AND RESOLUTIONS (Fosc = 8 MHz)

PWM Frequency	1.22 kHz	4.90 kHz	19.61 kHz	76.92 kHz	153.85 kHz	200.0 kHz
Timer Prescale	16	4	1	1	1	1
PR2 Value	0x65	0x65	0x65	0x19	0x0C	0x09
Maximum Resolution (bits)	8	8	8	6	5	5

21.5.7 OPERATION IN SLEEP MODE

In Sleep mode, the TMR2 register will not increment and the state of the module will not change. If the CCPx pin is driving a value, it will continue to drive that value. When the device wakes up, TMR2 will continue from its previous state.

21.5.8 CHANGES IN SYSTEM CLOCK FREQUENCY

The PWM frequency is derived from the system clock frequency. Any changes in the system clock frequency will result in changes to the PWM frequency. See **Section 4.0 “Oscillator Module (with Fail-Safe Clock Monitor)”** for additional details.

21.5.9 EFFECTS OF RESET

Any Reset will force all ports to Input mode and the CCP registers to their Reset states.

24.0 COMPLEMENTARY WAVEFORM GENERATOR (CWG) MODULE

The Complementary Waveform Generator (CWG) produces half-bridge, full-bridge, and steering of PWM waveforms. It is backwards compatible with previous CCP functions. The PIC18(L)F6xK40 family has one instance of the CWG module.

The CWG has the following features:

- Six operating modes:
 - Synchronous Steering mode
 - Asynchronous Steering mode
 - Full-Bridge mode, Forward
 - Full-Bridge mode, Reverse
 - Half-Bridge mode
 - Push-Pull mode
- Output polarity control
- Output steering
- Independent 6-bit rising and falling event dead-band timers
 - Clocked dead band
 - Independent rising and falling dead-band enables
- Auto-shutdown control with:
 - Selectable shutdown sources
 - Auto-restart option
 - Auto-shutdown pin override control

24.1 Fundamental Operation

The CWG generates two output waveforms from the selected input source.

The off-to-on transition of each output can be delayed from the on-to-off transition of the other output, thereby, creating a time delay immediately where neither output is driven. This is referred to as dead time and is covered in **Section 24.6 “Dead-Band Control”**.

It may be necessary to guard against the possibility of circuit faults or a feedback event arriving too late or not at all. In this case, the active drive must be terminated before the Fault condition causes damage. This is referred to as auto-shutdown and is covered in **Section 24.10 “Auto-Shutdown”**.

24.2 Operating Modes

The CWG module can operate in six different modes, as specified by the MODE<2:0> bits of the CWG1CON0 register:

- Half-Bridge mode
- Push-Pull mode
- Asynchronous Steering mode
- Synchronous Steering mode
- Full-Bridge mode, Forward
- Full-Bridge mode, Reverse

All modes accept a single pulse data input, and provide up to four outputs as described in the following sections.

All modes include auto-shutdown control as described in **Section 24.10 “Auto-Shutdown”**

Note: Except as noted for Full-bridge mode (Section 24.2.3 “Full-Bridge Modes”), mode changes should only be performed while EN = 0 (Register 24-1).
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24.2.1 HALF-BRIDGE MODE

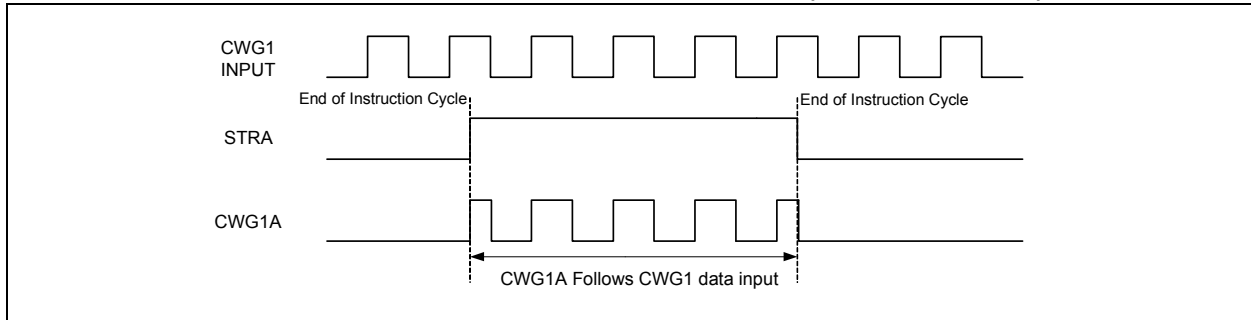
In Half-Bridge mode, two output signals are generated as true and inverted versions of the input as illustrated in Figure 24-2. A non-overlap (dead-band) time is inserted between the two outputs to prevent shoot through current in various power supply applications. Dead-band control is described in **Section 24.6 “Dead-Band Control”**. The output steering feature cannot be used in this mode. A basic block diagram of this mode is shown in Figure 24-1.

The unused outputs CWG1C and CWG1D drive similar signals, with polarity independently controlled by the POLC and POLD bits of the CWG1CON1 register, respectively.

24.2.4.2 Asynchronous Steering Mode

In Asynchronous mode (MODE<2:0> bits = 000, Register 24-1), steering takes effect at the end of the instruction cycle that writes to STR. In Asynchronous Steering mode, the output signal may be an incomplete waveform (Figure 24-10). This operation may be useful when the user firmware needs to immediately remove a signal from the output pin.

FIGURE 24-10: EXAMPLE OF ASYNCHRONOUS STEERING (MODE<2:0>= 000)



24.2.4.3 Start-up Considerations

The application hardware must use the proper external pull-up and/or pull-down resistors on the CWG output pins. This is required because all I/O pins are forced to high-impedance at Reset.

The POLy bits (Register 24-2) allow the user to choose whether the output signals are active-high or active-low.

27.4 Register Definitions: MSSP Control

REGISTER 27-1: SSPxSTAT: MSSPx STATUS REGISTER (SPI MODE)

R/W-0	R/W-0	R-0	R-0	R-0	R-0	R-0	R-0
SMP	CKE ⁽¹⁾	D/A	P	S	R/W	UA	BF
bit 7							bit 0

Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

-n = Value at POR

'1' = Bit is set

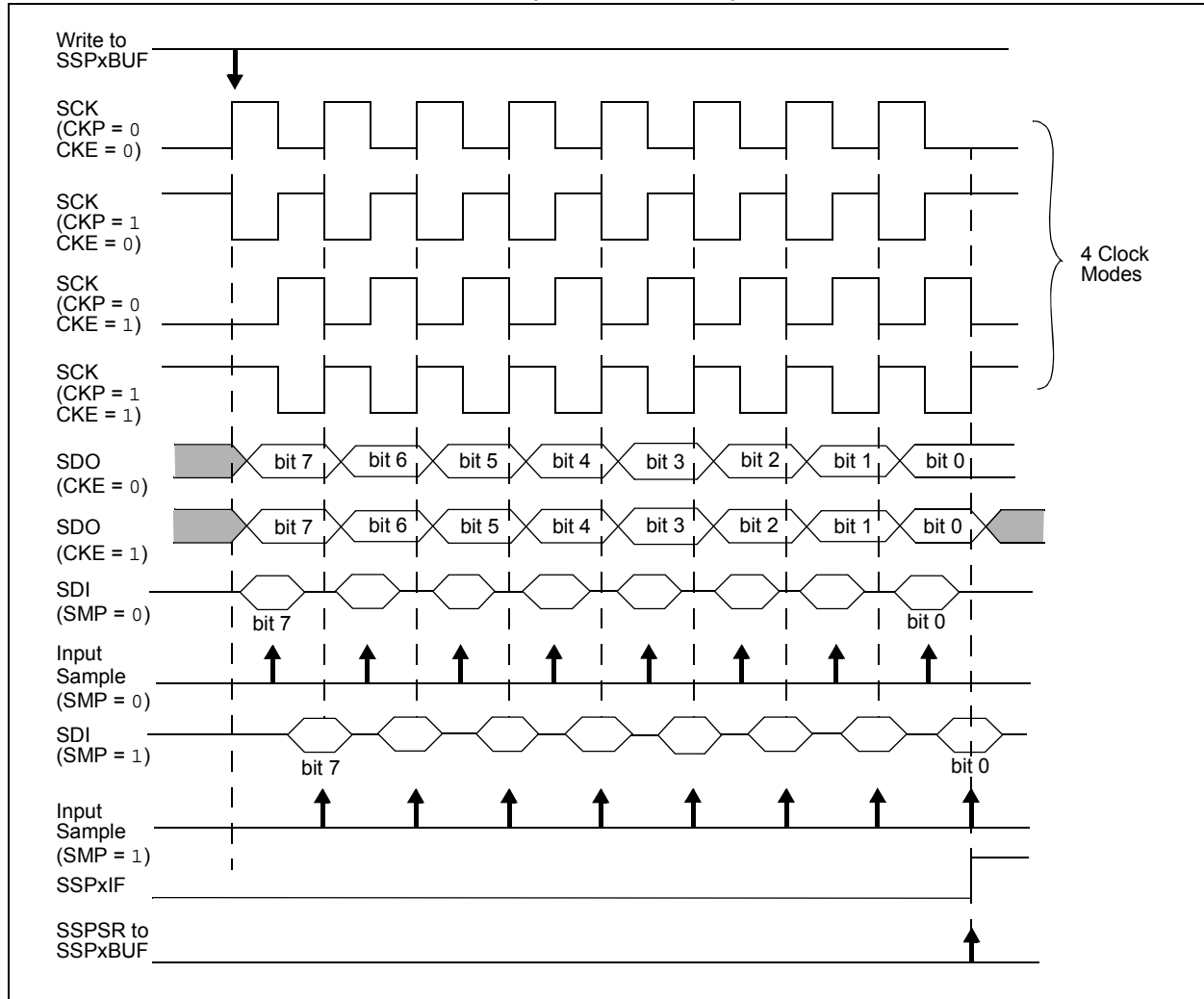
'0' = Bit is cleared

x = Bit is unknown

- bit 7 **SMP:** Sample bit
SPI Master mode:
 1 = Input data is sampled at the end of data output time
 0 = Input data is sampled at the middle of data output time
SPI Slave mode:
 SMP must be cleared when SPI is used in Slave mode.
- bit 6 **CKE:** SPI Clock Select bit⁽¹⁾
 1 = Transmit occurs on the transition from active to Idle clock state
 0 = Transmit occurs on the transition from Idle to active clock state
- bit 5 **D/A:** Data/Address bit
 Used in I²C mode only.
- bit 4 **P:** Stop bit
 Used in I²C mode only. This bit is cleared when the MSSPx module is disabled; SSPEN is cleared.
- bit 3 **S:** Start bit
 Used in I²C mode only.
- bit 2 **R/W:** Read/Write Information bit
 Used in I²C mode only.
- bit 1 **UA:** Update Address bit
 Used in I²C mode only.
- bit 0 **BF:** Buffer Full Status bit (Receive mode only)
 1 = Receive is complete, SSPxBUF is full
 0 = Receive is not complete, SSPxBUF is empty

Note 1: Polarity of clock state is set by the CKP bit (SSPxCON1<4>).

FIGURE 27-4: SPI MODE WAVEFORM (MASTER MODE)



27.5.2 SPI SLAVE MODE

In Slave mode, the data is transmitted and received as external clock pulses appear on SCK. When the last bit is latched, the SSPxIF interrupt flag bit is set.

Before enabling the module in SPI Slave mode, the clock line must match the proper Idle state. The clock line can be observed by reading the SCK pin. The Idle state is determined by the CKP bit of the SSPxCON1 register.

While in Slave mode, the external clock is supplied by the external clock source on the SCK pin. This external clock must meet the minimum high and low times as specified in the electrical specifications.

While in Sleep mode, the slave can transmit/receive data. The shift register is clocked from the SCK pin input and when a byte is received, the device will generate an interrupt. If enabled, the device will wake-up from Sleep.

27.5.3 DAISY-CHAIN CONFIGURATION

The SPI bus can sometimes be connected in a daisy-chain configuration. The first slave output is connected to the second slave input, the second slave output is connected to the third slave input, and so on. The final slave output is connected to the master input. Each slave sends out, during a second group of clock pulses, an exact copy of what was received during the first group of clock pulses. The whole chain acts as one large communication shift register. The daisy-chain feature only requires a single Slave Select line from the master device.

Figure 27-5 shows the block diagram of a typical daisy-chain connection when operating in SPI mode.

In a daisy-chain configuration, only the most recent byte on the bus is required by the slave. Setting the BOEN bit of the SSPxCON3 register will enable writes to the SSPxBUF register, even if the previous byte has not been read. This allows the software to ignore data that may not apply to it.

30.4 ADC Acquisition Time

To ensure accurate temperature measurements, the user must wait at least 200 μ s after the ADC input multiplexer is connected to the temperature indicator output before the conversion is performed. In addition, the user must wait 200 μ s between consecutive conversions of the temperature indicator output.

TABLE 30-2: SUMMARY OF REGISTERS ASSOCIATED WITH THE TEMPERATURE INDICATOR

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Register on page
FVRCON	FVREN	FVRRDY	TSEN	TSRNG	CDFVR<1:0>		ADFVR<1:0>		479

Legend: — = Unimplemented location, read as '0'. Shaded cells are unused by the temperature indicator module.

REGISTER 32-32: ADACT: ADC AUTO CONVERSION TRIGGER CONTROL REGISTER

U-0	U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
—	—	—	ADACT<4:0>				
bit 7			bit 0				

33.0 COMPARATOR MODULE (C1/2/3)

Comparators are used to interface analog circuits to a digital circuit by comparing two analog voltages and providing a digital indication of their relative magnitudes. Comparators are very useful mixed signal building blocks because they provide analog functionality independent of program execution. The PIC18(L)F6xK40 family has three comparators (C1/2/3).

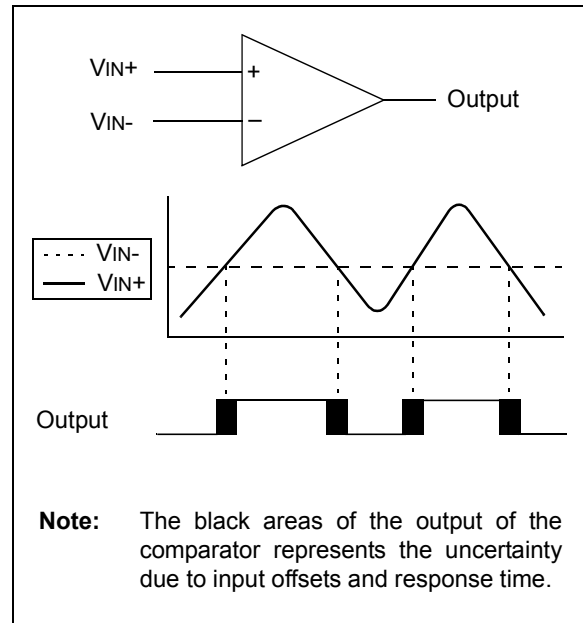
The analog comparator module includes the following features:


- Programmable input selection
- Programmable output polarity
- Rising/falling output edge interrupts
- Wake-up from Sleep
- CWG Auto-shutdown source
- Selectable voltage reference
- ADC Auto-trigger
- TMR1/3/5/7 Gate
- TMR2/4/6/8 Reset
- CCP Capture Mode Input
- DSM Modulator Source
- Input and Window signal to Signal Measurement Timer

33.1 Comparator Overview

A single comparator is shown in Figure 33-1 along with the relationship between the analog input levels and the digital output. When the analog voltage at V_{IN+} is less than the analog voltage at V_{IN-} , the output of the comparator is a digital low level. When the analog voltage at V_{IN+} is greater than the analog voltage at V_{IN-} , the output of the comparator is a digital high level.

FIGURE 33-1: SINGLE COMPARATOR



RRNCF		Rotate Right f (No Carry)											
Syntax:	RRNCF f {,d {,a}}												
Operands:	$0 \leq f \leq 255$ $d \in [0,1]$ $a \in [0,1]$												
Operation:	$(f < n) \rightarrow \text{dest} < n - 1 >$, $(f < 0) \rightarrow \text{dest} < 7 >$												
Status Affected:	N, Z												
Encoding:	<table><tr><td>0100</td><td>00da</td><td>ffff</td><td>ffff</td></tr></table>					0100	00da	ffff	ffff				
0100	00da	ffff	ffff										
Description:	<p>The contents of register 'f' are rotated one bit to the right. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed back in register 'f' (default). If 'a' is '0', the Access Bank will be selected (default), overriding the BSR value. If 'a' is '1', then the bank will be selected as per the BSR value. If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $f \leq 95$ (5Fh). See Section 36.2.3 “Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode” for details.</p> <div></div>												
Words:	1												
Cycles:	1												
Q Cycle Activity:	<table><tr><th>Q1</th><th>Q2</th><th>Q3</th><th>Q4</th></tr><tr><td>Decode</td><td>Read register 'f'</td><td>Process Data</td><td>Write to destination</td></tr></table>					Q1	Q2	Q3	Q4	Decode	Read register 'f'	Process Data	Write to destination
Q1	Q2	Q3	Q4										
Decode	Read register 'f'	Process Data	Write to destination										

Example 1: RRNCF REG, 1, 0

Before Instruction
REG = 1101 0111
After Instruction
REG = 1110 1011

Example 2: RRNCF REG, 0, 0

Before Instruction
W = ?
REG = 1101 0111
After Instruction
W = 1110 1011
REG = 1101 0111

SETF		Set f							
Syntax:	SETF f {,a}								
Operands:	$0 \leq f \leq 255$ $a \in [0,1]$								
Operation:	$\text{FFh} \rightarrow f$								
Status Affected:	None								
Encoding:	<table border="1"><tr><td>0110</td><td>100a</td><td>ffff</td><td>ffff</td></tr></table>					0110	100a	ffff	ffff
0110	100a	ffff	ffff						
Description:	<p>The contents of the specified register are set to FFh.</p> <p>If 'a' is '0', the Access Bank is selected.</p> <p>If 'a' is '1', the BSR is used to select the GPR bank.</p> <p>If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever $f \leq 95$ (5Fh). See Section 36.2.3 "Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode" for details.</p>								
Words:	1								
Cycles:	1								
Q Cycle Activity:									
	Q1	Q2	Q3	Q4					
	Decode	Read register 'f'	Process Data	Write register 'f'					

Example: SETF REG, 1

Before Instruction
REG = 5Ah
After Instruction
REG = FFh

36.2.3 BYTE-ORIENTED AND BIT-ORIENTED INSTRUCTIONS IN INDEXED LITERAL OFFSET MODE

Note: Enabling the PIC18 instruction set extension may cause legacy applications to behave erratically or fail entirely.

In addition to eight new commands in the extended set, enabling the extended instruction set also enables Indexed Literal Offset Addressing mode (**Section 10.7.1 “Indexed Addressing with Literal Offset”**). This has a significant impact on the way that many commands of the standard PIC18 instruction set are interpreted.

When the extended set is disabled, addresses embedded in opcodes are treated as literal memory locations: either as a location in the Access Bank ('a' = 0), or in a GPR bank designated by the BSR ('a' = 1). When the extended instruction set is enabled and 'a' = 0, however, a file register argument of 5Fh or less is interpreted as an offset from the pointer value in FSR2 and not as a literal address. For practical purposes, this means that all instructions that use the Access RAM bit as an argument – that is, all byte-oriented and bit-oriented instructions, or almost half of the core PIC18 instructions – may behave differently when the extended instruction set is enabled.

When the content of FSR2 is 00h, the boundaries of the Access RAM are essentially remapped to their original values. This may be useful in creating backward compatible code. If this technique is used, it may be necessary to save the value of FSR2 and restore it when moving back and forth between C and assembly routines in order to preserve the Stack Pointer. Users must also keep in mind the syntax requirements of the extended instruction set (see **Section 36.2.3.1 “Extended Instruction Syntax with Standard PIC18 Commands”**).

Although the Indexed Literal Offset Addressing mode can be very useful for dynamic stack and pointer manipulation, it can also be very annoying if a simple arithmetic operation is carried out on the wrong register. Users who are accustomed to the PIC18 programming must keep in mind that, when the extended instruction set is enabled, register addresses of 5Fh or less are used for Indexed Literal Offset Addressing.

Representative examples of typical byte-oriented and bit-oriented instructions in the Indexed Literal Offset Addressing mode are provided on the following page to show how execution is affected. The operand conditions shown in the examples are applicable to all instructions of these types.

36.2.3.1 Extended Instruction Syntax with Standard PIC18 Commands

When the extended instruction set is enabled, the file register argument, 'f', in the standard byte-oriented and bit-oriented commands is replaced with the literal offset value, 'k'. As already noted, this occurs only when 'f' is less than or equal to 5Fh. When an offset value is used, it must be indicated by square brackets ("[]"). As with the extended instructions, the use of brackets indicates to the compiler that the value is to be interpreted as an index or an offset. Omitting the brackets, or using a value greater than 5Fh within brackets, will generate an error in the MPASM assembler.

If the index argument is properly bracketed for Indexed Literal Offset Addressing, the Access RAM argument is never specified; it will automatically be assumed to be '0'. This is in contrast to standard operation (extended instruction set disabled) when 'a' is set on the basis of the target address. Declaring the Access RAM bit in this mode will also generate an error in the MPASM assembler.

The destination argument, 'd', functions as before.

In the latest versions of the MPASM™ assembler, language support for the extended instruction set must be explicitly invoked. This is done with either the command line option, /y, or the PE directive in the source listing.

36.2.4 CONSIDERATIONS WHEN ENABLING THE EXTENDED INSTRUCTION SET

It is important to note that the extensions to the instruction set may not be beneficial to all users. In particular, users who are not writing code that uses a software stack may not benefit from using the extensions to the instruction set.

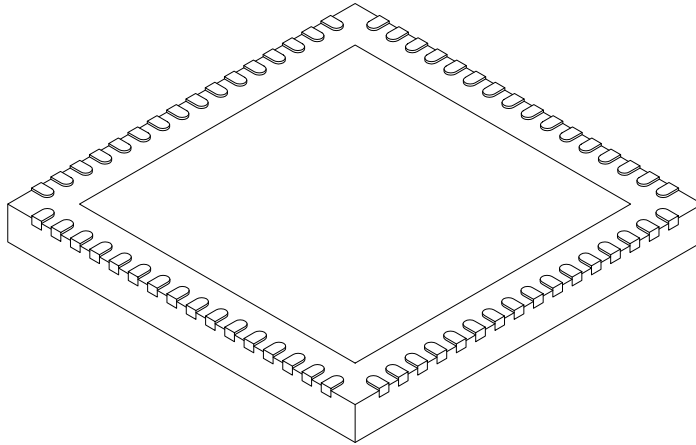
Additionally, the Indexed Literal Offset Addressing mode may create issues with legacy applications written to the PIC18 assembler. This is because instructions in the legacy code may attempt to address registers in the Access Bank below 5Fh. Since these addresses are interpreted as literal offsets to FSR2 when the instruction set extension is enabled, the application may read or write to the wrong data addresses.

When porting an application to the PIC18(L)F6xK40, it is very important to consider the type of code. A large, re-entrant application that is written in 'C' and would benefit from efficient compilation will do well when using the instruction set extensions. Legacy applications that heavily use the Access Bank will most likely not benefit from using the extended instruction set.

PIC18(L)F65/66K40

64-Lead Plastic Quad Flat, No Lead Package (MR) – 9x9x0.9 mm Body [QFN] With 7.15 x 7.15 Exposed Pad [QFN]

Note: For the most current package drawings, please see the Microchip Packaging Specification located at <http://www.microchip.com/packaging>



Units		MILLIMETERS		
Dimension Limits		MIN	NOM	MAX
Number of Pins	N	64		
Pitch	e	0.50 BSC		
Overall Height	A	0.80	0.90	1.00
Standoff	A1	0.00	0.02	0.05
Contact Thickness	A3	0.20 REF		
Overall Width	E	9.00 BSC		
Exposed Pad Width	E2	7.05	7.15	7.50
Overall Length	D	9.00 BSC		
Exposed Pad Length	D2	7.05	7.15	7.50
Contact Width	b	0.18	0.25	0.30
Contact Length	L	0.30	0.40	0.50
Contact-to-Exposed Pad	K	0.20	-	-

Notes:

- Pin 1 visual index feature may vary, but must be located within the hatched area.
- Package is saw singulated.
- Dimensioning and tolerancing per ASME Y14.5M.
 - BSC: Basic Dimension. Theoretically exact value shown without tolerances.
 - REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-149C Sheet 2 of 2