Welcome to E-XFL.COM

Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Active
Core Processor	ARM1136JF-S
Number of Cores/Bus Width	1 Core, 32-Bit
Speed	400MHz
Co-Processors/DSP	Multimedia; GPU, IPU, MPEG-4, VFP
RAM Controllers	DDR
Graphics Acceleration	Yes
Display & Interface Controllers	Keyboard, Keypad, LCD
Ethernet	-
SATA	-
USB	USB 2.0 (3)
Voltage - I/O	1.8V, 2.0V, 2.5V, 2.7V, 3.0V
Operating Temperature	-40°C ~ 85°C (TA)
Security Features	Random Number Generator, RTIC, Secure Fusebox, Secure JTAG, Secure Memory
Package / Case	473-LFBGA
Supplier Device Package	473-LFBGA (19x19)
Purchase URL	https://www.e-xfl.com/pro/item?MUrl=&PartUrl=mcimx31cvmn4cr2

MCIMX31C provides the optimal performance versus leakage current balance.

The performance of the MCIMX31C is boosted by a multi-level cache system, and features peripheral devices such as an MPEG-4 Hardware Encoder (VGA, 30 fps), an Autonomous Image Processing Unit, a Vector Floating Point (VFP11) co-processor, and a RISC-based SDMA controller.

The MCIMX31C supports connections to various types of external memories, such as DDR, NAND Flash, NOR Flash, SDRAM, and SRAM. The MCIMX31C can be connected to a variety of external devices using technology, such as high-speed USB2.0 OTG, ATA, MMC/SDIO, and compact flash.

1.1 Features

The MCIMX31C is designed for automotive and industrial markets where extended operating temperature is required. They provide low-power solutions for high-performance demanding multimedia and graphics applications.

The MCIMX31C is built around the ARM11 MCU core and implemented in the 90 nm technology.

The systems include the following features:

- Multimedia and floating-point hardware acceleration supporting:
  - MPEG-4 real-time encode of up to VGA at 30 fps
  - MPEG-4 real-time video post-processing of up to VGA at 30 fps
  - Video conference call of up to QCIF-30 fps (decoder in software), 128 kbps
  - Video streaming (playback) of up to VGA-30 fps, 384 kbps
  - 3D graphics and other applications acceleration with the ARM® tightly-coupled Vector Floating Point co-processor
  - On-the-fly video processing that reduces system memory load (for example, the power-efficient viewfinder application with no involvement of either the memory system or the ARM CPU)
- Advanced power management
  - Dynamic voltage and frequency scaling
  - Multiple clock and power domains
  - Independent gating of power domains
- Multiple communication and expansion ports including a fast parallel interface to an external graphic accelerator (supporting major graphic accelerator vendors)
- Security

1.3 Block Diagram

Figure 1 shows the MCIMX31C simplified interface block diagram.

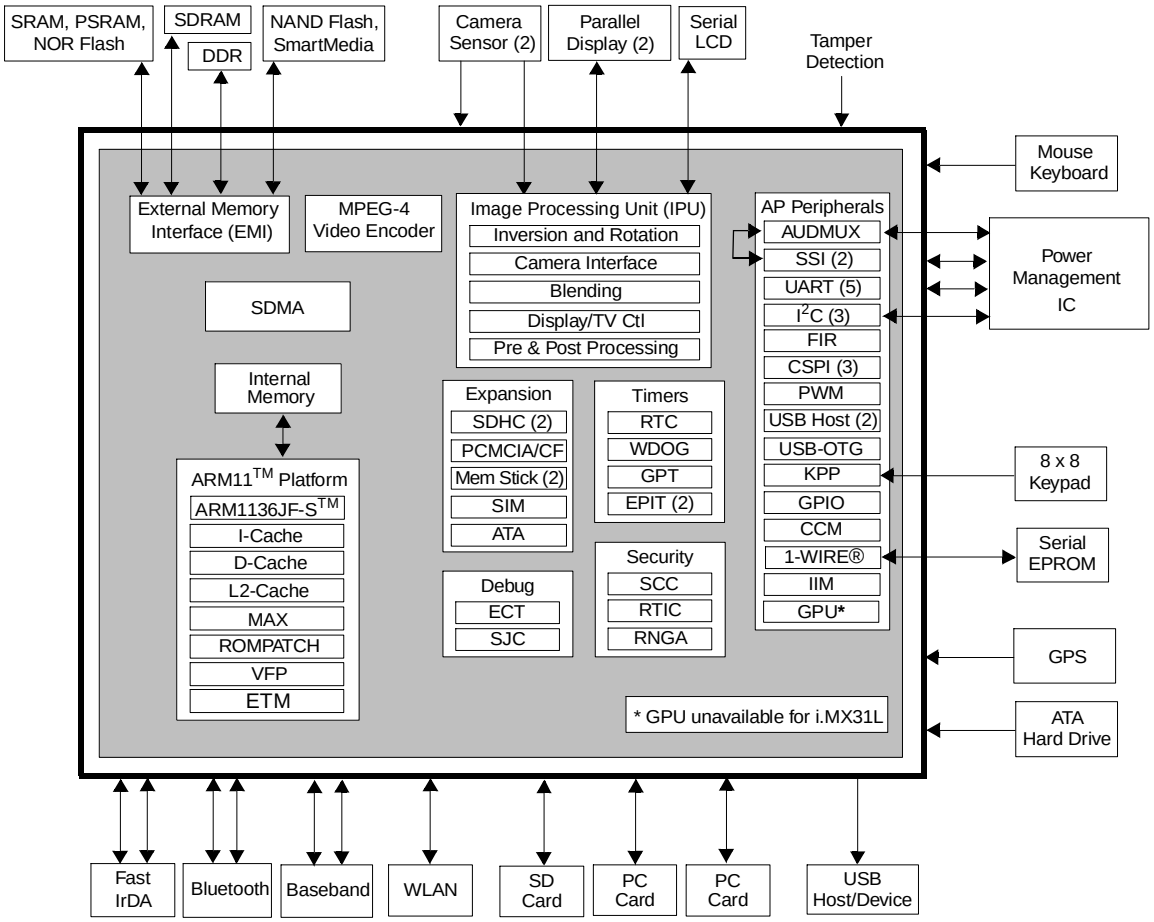


Figure 1. MCIMX31C Simplified Interface Block Diagram

2 Functional Description and Application Information

2.1 ARM11 Microprocessor Core

The CPU of the MCIMX31C is the ARM1136JF-S core based on the ARM v6 architecture. It supports the ARM Thumb® instruction sets, features Jazelle® technology (which enables direct execution of Java byte codes), and a range of SIMD DSP instructions that operate on 16-bit or 8-bit data values in 32-bit registers.

The ARM1136JF-S processor core features:

- Integer unit with integral EmbeddedICE™ logic
- Eight-stage pipeline
- Branch prediction with return stack
- Low-interrupt latency

3 Signal Descriptions

Signal descriptions are in the reference manual. Special signal considerations are listed following this paragraph. The BGA ball assignment is in Section 5, “Package Information and Pinout,” on page 99.

Special Signal Considerations:

- **Tamper detect (GPIO1_6)**

  Tamper detect logic is used to issue a security violation. This logic is activated if the tamper detect input is asserted.

  The tamper detect logic is disabled after reset. After enabling the logic, it is impossible to disable it until the next reset. The GPR[16] bit functions as the tamper detect enable bit.

  GPIO1_6 functions similarly to other I/O with GPIO capabilities regardless of the status of the tamper detect enable bit. (For example, the GPIO1_6 can function as an input with GPIO capabilities, such as sampling through PSR or generating interrupts.)
- **Power ready (GPIO1_5)**

  The power ready input, GPIO1_5, should be connected to an external power management IC power ready output signal. If not used, GPIO1_5 must either be (a) externally pulled-up to NVCC1 or (b) a no connect, internally pulled-up by enabling the on-chip pull-up resistor. GPIO1_5 is a dedicated input and cannot be used as a general-purpose input/output.
- **SJC_MOD**

  SJC_MOD must be externally connected to GND for normal operation. Termination to GND through an external pull-down resistor (such as 1 kΩ) is allowed, but the value should be much smaller than the on-chip 100 kΩ pull-up.
- **CE_CONTROL**

  CE_CONTROL is a reserved input and must be externally tied to GND through a 1 kΩ resistor.
- **M_REQUEST and M_GRANT**

  These two signals are not utilized internally. The user should make no connection to these signals.
- **Clock Source Select (CLKSS)**

  The CLKSS is the input that selects the default reference clock source providing input to the DPLL. To select CKIH, tie CLKSS to NVCC1. To select CKIL, tie CLKSS to ground. After initialization, the reference clock source can be changed (initial setting is overwritten) by programming the PRCS bits in the CCMR.

4 Electrical Characteristics

This section provides the device-level and module-level electrical characteristics for the MCIMX31C.

4.1 Chip-Level Conditions

This section provides the device-level electrical characteristics for the IC. See Table 4 for a quick reference to the individual tables and sections.

NOTE

The term NVCC in this section refers to the associated supply rail of an input or output. The association is shown in the Signal Multiplexing chapter of the reference manual. NVCC for Table 12 refers to NVCC1 and NVCC3–10; QVCC refers to QVCC, QVCC1, and QVCC4.

Table 12. GPIO DC Electrical Parameters

Parameter	Symbol	Test Conditions	Min	Typ	Max	Units
High-level output voltage	V_{OH}	I_{OH} = –1 mA	NVCC –0.15	—	—	V
		I_{OH} = specified Drive	0.8*NVCC	—	—	V
Low-level output voltage	V_{OL}	I_{OL} = 1 mA	—	—	0.15	V
		I_{OL} = specified Drive	—	—	0.2*NVCC	V
High-level output current, slow slew rate	I_{OH_S}	V_{OH}=0.8*NVCC Std Drive High Drive Max Drive	 –2 –4 –8	—	—	mA
High-level output current, fast slew rate	I_{OH_F}	V_{OH}=0.8*NVCC Std Drive High Drive Max Drive	 –4 –6 –8	—	—	mA
Low-level output current, slow slew rate	I_{OL_S}	V_{OL}=0.2*NVCC Std Drive High Drive Max Drive	 2 4 8	—	—	mA
Low-level output current, fast slew rate	I_{OL_F}	V_{OL}=0.2*NVCC Std Drive High Drive Max Drive	 4 6 8	—	—	mA
High-Level DC input voltage	V_{IH}	—	0.7*NVCC	—	NVCC	V
Low-Level DC input voltage	V_{IL}	—	0	—	0.3*QVCC	V
Input Hysteresis	V_{HYS}	Hysteresis enabled	0.25	—	—	V
Schmitt trigger VT+	V_T +	Hysteresis enabled	0.5*QVCC	—	—	V
Schmitt trigger VT–	V_T –	Hysteresis enabled	—	—	0.5*QVCC	V
Pull-up resistor (100 kΩ PU)	R_{PU}[1]	—	—	100	—	kΩ
Pull-down resistor (100 kΩ PD)	R_{PD}[1]	—	—	100	—	
Input current (no PU/PD)	I_{IN}	V_I = NVCC or GND	—	—	±1	µA
Input current (100 kΩ PU)	I_{IN}	V_I = 0	—	—	25	µA µA
Input current (100 kΩ PD)	I_{IN}	V_I = NVCC	—	—	28	µA
Tri-state leakage current	I_{OZ}	V_I = NVCC or GND I/O = High Z	—	—	±2	µA

1 Not a precise value. Measurements made on small sample size have shown variations of ±50% or more.

Table 19. WR0 Sequence Timing Parameters

ID	Parameter	Symbol	Min	Typ	Max	Units
OW5	Write 0 Low Time	t_{WR0_low}	60	100	120	µs
OW6	Transmission Time Slot	t_{SLOT}	OW5	117	120	µs

Figure 8 depicts Write 1 Sequence timing, Figure 9 depicts the Read Sequence timing, and Table 20 lists the timing parameters.

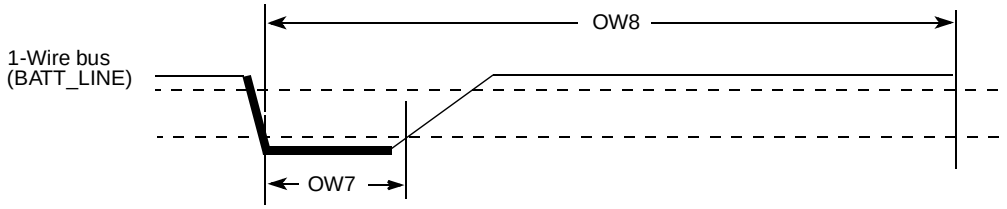


Figure 8. Write 1 Sequence Timing Diagram

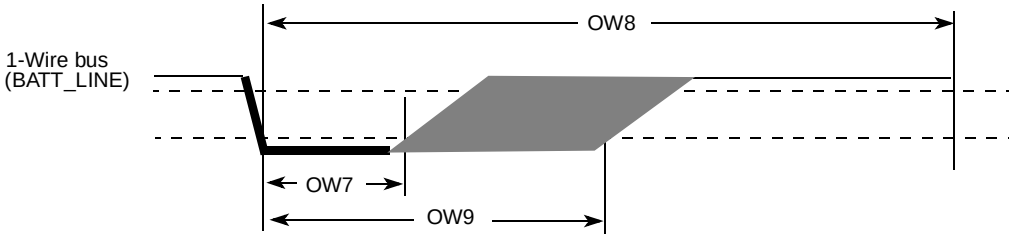


Figure 9. Read Sequence Timing Diagram

Table 20. WR1/RD Timing Parameters

ID	Parameter	Symbol	Min	Typ	Max	Units
OW7	Write 1 / Read Low Time	t_{LOW1}	1	5	15	µs
OW8	Transmission Time Slot	t_{SLOT}	60	117	120	µs
OW9	Release Time	$t_{RELEASE}$	15	—	45	µs

4.3.5 ATA Electrical Specifications (ATA Bus, Bus Buffers)

This section discusses ATA parameters. For a detailed description, refer to the ATA specification.

The user needs to use level shifters for 3.3 Volt or 5.0 Volt compatibility on the ATA interface.

The use of bus buffers introduces delay on the bus and introduces skew between signal lines. These factors make it difficult to operate the bus at the highest speed (UDMA-5) when bus buffers are used. If fast UDMA mode operation is needed, this may not be compatible with bus buffers.

Another area of attention is the slew rate limit imposed by the ATA specification on the ATA bus. According to this limit, any signal driven on the bus should have a slew rate between 0.4 and 1.2 V/ns with a 40 pF load. Not many vendors of bus buffers specify slew rate of the outgoing signals.

Table 21. ATA Timing Parameters (continued)

Name	Description	Value/ Contributing Factor[1]
tskew5	Max difference in cable propagation delay between (ata_dior, ata_diow, ata_dmack) and ata_cs0, ata_cs1, ata_da2, ata_da1, ata_da0, ata_data(write)	cable
tskew6	Max difference in cable propagation delay without accounting for ground bounce	cable

1 Values provided where applicable.

4.3.5.2 PIO Mode Timing

Figure 10 shows timing for PIO read, and Table 22 lists the timing parameters for PIO read.

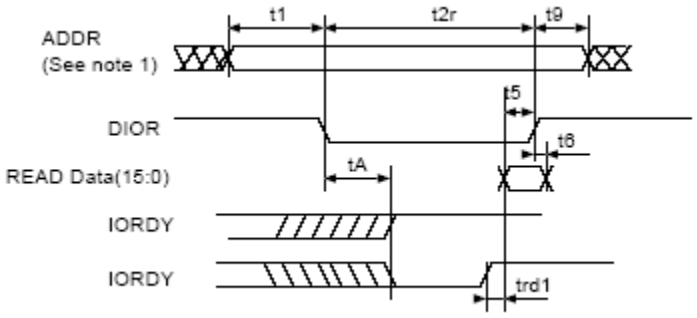


Figure 10. PIO Read Timing Diagram

Table 22. PIO Read Timing Parameters

ATA Parameter	Parameter from Figure 10	Value	Controlling Variable
t1	t1	t1 (min) = time_1 * T – (tskew1 + tskew2 + tskew5)	time_1
t2	t2r	t2 min) = time_2r * T – (tskew1 + tskew2 + tskew5)	time_2r
t9	t9	t9 (min) = time_9 * T – (tskew1 + tskew2 + tskew6)	time_3
t5	t5	t5 (min) = tco + tsu + tbuf + tbuf + tcable1 + tcable2	If not met, increase time_2
t6	t6	0	—
tA	tA	tA (min) = (1.5 + time_ax) * T – (tco + tsui + tcable2 + tcable2 + 2*tbuf)	time_ax
trd	trd1	trd1 (max) = (–trd) + (tskew3 + tskew4) trd1 (min) = (time_pio_rdx – 0.5)*T – (tsu + thi) (time_pio_rdx – 0.5) * T > tsu + thi + tskew3 + tskew4	time_pio_rdx
t0	—	t0 (min) = (time_1 + time_2 + time_9) * T	time_1, time_2r, time_9

Figure 11 shows timing for PIO write, and Table 23 lists the timing parameters for PIO write.

Table 28. DPLL Specifications (continued)

Parameter	Min	Typ	Max	Unit	Comments
Phase lock time	—	—	100	µs	In addition to the frequency
Maximum allowed PLL supply voltage ripple	—	—	25	mV	$F_{modulation}$ < 50 kHz
Maximum allowed PLL supply voltage ripple	—	—	20	mV	50 kHz < $F_{modulation}$ < 300 kHz
Maximum allowed PLL supply voltage ripple	—	—	25	mV	$F_{modulation}$ > 300 kHz
PLL output clock phase jitter	—	—	5.2	ns	Measured on CLKO pin
PLL output clock period jitter	—	—	420	ps	Measured on CLKO pin

1 The user or board designer must take into account that the use of a frequency other than 26 MHz would require adjustment to the DPTC–DVFS table, which is incorporated into operating system code.

2 The PLL reference frequency must be ≤ 35 MHz. Therefore, for frequencies between 35 MHz and 70 MHz, program the predivider to divide by 2 or more. If the CKIH frequency is above 70 MHz, program the predivider to 3 or more. For PD bit description, see the reference manual.

4.3.9 EMI Electrical Specifications

This section provides electrical parametrics and timings for EMI module.

4.3.9.1 NAND Flash Controller Interface (NFC)

The NFC supports normal timing mode, using two flash clock cycles for one access of $\overline{RE}$ and $\overline{WE}$. AC timings are provided as multiplications of the clock cycle and fixed delay. Figure 22, Figure 23, Figure 24, and Figure 25 depict the relative timing requirements among different signals of the NFC at module level, for normal mode, and Table 29 lists the timing parameters.

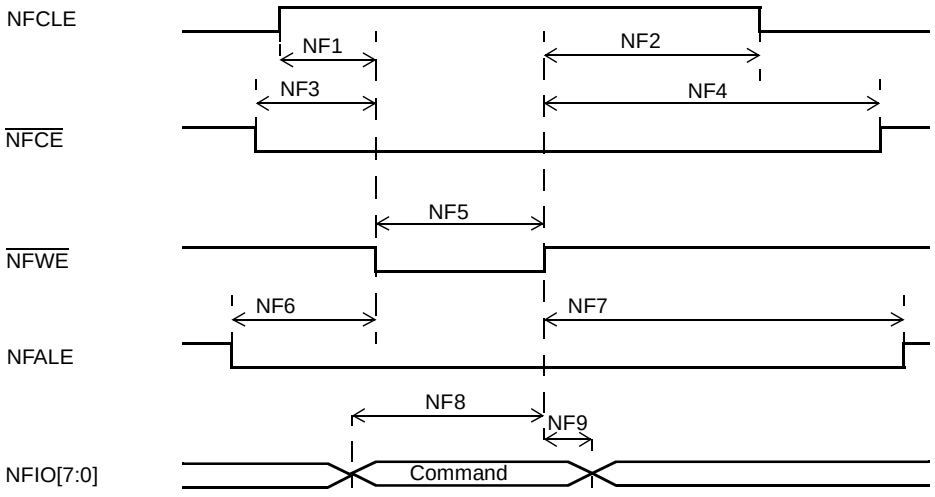
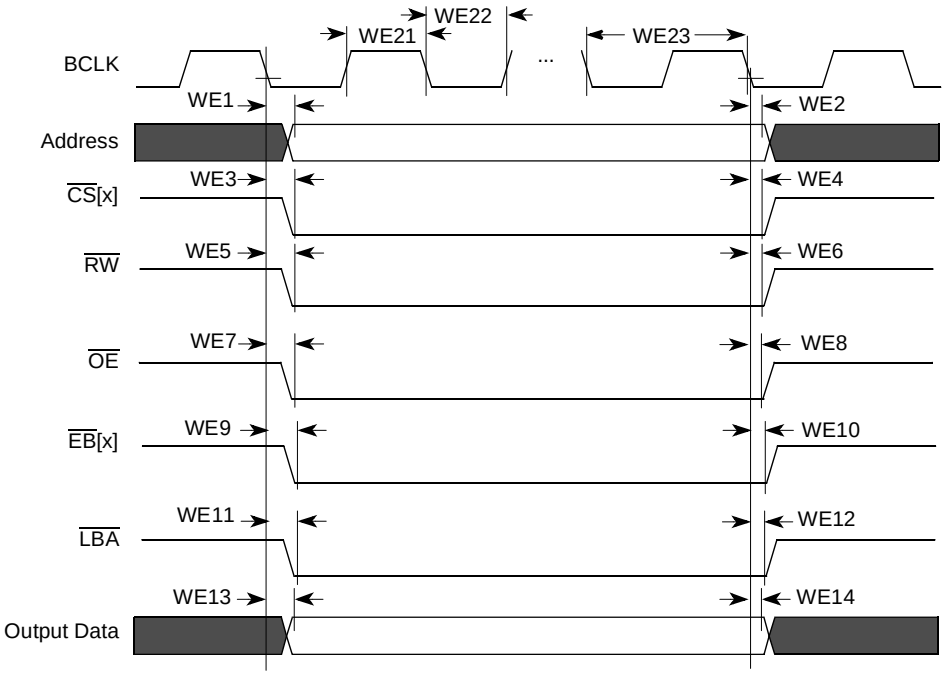


Figure 22. Command Latch Cycle Timing Diagram

WEIM Outputs Timing



WEIM Inputs Timing

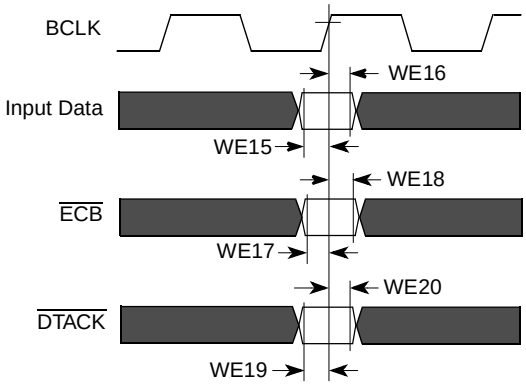


Figure 26. WEIM Bus Timing Diagram

Table 30. WEIM Bus Timing Parameters

ID	Parameter	Min	Max	Unit
WE1	Clock fall to Address Valid	–0.5	2.5	ns
WE2	Clock rise/fall to Address Invalid	–0.5	5	ns
WE3	Clock rise/fall to $\overline{CS}$[x] Valid	–3	3	ns
WE4	Clock rise/fall to $\overline{CS}$[x] Invalid	–3	3	ns
WE5	Clock rise/fall to $\overline{RW}$ Valid	–3	3	ns
WE6	Clock rise/fall to $\overline{RW}$ Invalid	–3	3	ns
WE7	Clock rise/fall to $\overline{OE}$ Valid	–3	3	ns

Table 30. WEIM Bus Timing Parameters (continued)

ID	Parameter	Min	Max	Unit
WE8	Clock rise/fall to $\overline{OE}$ Invalid	–3	3	ns
WE9	Clock rise/fall to $\overline{EB}$[x] Valid	–3	3	ns
WE10	Clock rise/fall to $\overline{EB}$[x] Invalid	–3	3	ns
WE11	Clock rise/fall to $\overline{LBA}$ Valid	–3	3	ns
WE12	Clock rise/fall to $\overline{LBA}$ Invalid	–3	3	ns
WE13	Clock rise/fall to Output Data Valid	–2.5	4	ns
WE14	Clock rise to Output Data Invalid	–2.5	4	ns
WE15	Input Data Valid to Clock rise, FCE=0 FCE=1	8 2.5	—	ns
WE16	Clock rise to Input Data Invalid, FCE=0 FCE=1	–2 –2	—	ns
WE17	$\overline{ECB}$ setup time, FCE=0 FCE=1	6.5 3.5	—	ns
WE18	$\overline{ECB}$ hold time, FCE=0 FCE=1	–2 2	—	ns
WE19	$\overline{DTACK}$ setup time[1]	0	—	ns
WE20	$\overline{DTACK}$ hold time[1]	4.5	—	ns
WE21	BCLK High Level Width[2, 3]	—	T/2 – 3	ns
WE22	BCLK Low Level Width[2, 3]	—	T/2 – 3	ns
WE23	BCLK Cycle time[2]	15	—	ns

1 Applies to rising edge timing

2 BCLK parameters are being measured from the 50% VDD.

3 The actual cycle time is derived from the AHB bus clock frequency.

NOTE

High is defined as 80% of signal value and low is defined as 20% of signal value.

Test conditions: load capacitance, 25 pF. Recommended drive strength for all controls, address, and BCLK is Max drive.

Figure 27, Figure 28, Figure 29, Figure 30, Figure 31, and Figure 32 depict some examples of basic WEIM accesses to external memory devices with the timing parameters mentioned in Table 30 for specific control parameter settings.

Table 33. SDRAM Refresh Timing Parameters

ID	Parameter	Symbol	Min	Max	Unit
SD1	SDRAM clock high-level width	tCH	3.4	4.1	ns
SD2	SDRAM clock low-level width	tCL	3.4	4.1	ns
SD3	SDRAM clock cycle time	tCK	7.5	—	ns
SD6	Address setup time	tAS	1.8	—	ns
SD7	Address hold time	tAH	1.8	—	ns
SD10	Precharge cycle period[1]	tRP	1	4	clock
SD11	Auto precharge command period[1]	tRC	2	20	clock

1 SD10 and SD11 are determined by SDRAM controller register settings.

NOTE

SDR SDRAM CLK parameters are being measured from the 50% point—that is, high is defined as 50% of signal value and low is defined as 50% of signal value.

NOTE

The timing parameters are similar to the ones used in SDRAM data sheets—that is, Table 33 indicates SDRAM requirements. All output signals are driven by the ESDCTL at the negative edge of SDCLK and the parameters are measured at maximum memory frequency.

Table 37. ETM TRACECLK Timing Parameters

ID	Parameter	Min	Max	Unit
T_{cyc}	Clock period	Frequency dependent	—	ns
T_{wl}	Low pulse width	2	—	ns
T_{wh}	High pulse width	2	—	ns
T_r	Clock and data rise time	—	3	ns
T_f	Clock and data fall time	—	3	ns

Figure 40 depicts the setup and hold requirements of the trace data pins with respect to TRACECLK, and Table 38 lists the timing parameters.

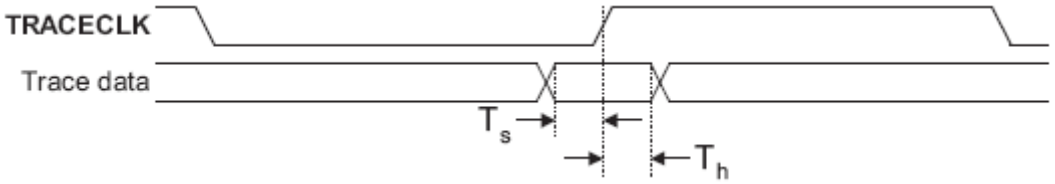


Figure 40. Trace Data Timing Diagram

Table 38. ETM Trace Data Timing Parameters

ID	Parameter	Min	Max	Unit
T_s	Data setup	2	—	ns
T_h	Data hold	1	—	ns

4.3.10.1 Half-Rate Clocking Mode

When half-rate clocking is used, the trace data signals are sampled by the TPA on both the rising and falling edges of TRACECLK, where TRACECLK is half the frequency of the clock shown in Figure 40.

4.3.11 FIR Electrical Specifications

FIR implements asynchronous infrared protocols (FIR, MIR) that are defined by IrDA® (Infrared Data Association). Refer to http://www.IrDA.org for details on FIR and MIR protocols.

4.3.12 Fusebox Electrical Specifications

Table 39. Fusebox Timing Characteristics

Ref. Num	Description	Symbol	Minimum	Typical	Maximum	Units
1	Program time for eFuse[1]	$t_{program}$	125	—	—	µs

1 The program length is defined by the value defined in the epm_pgm_length[2:0] bits of the IIM module. The value to program is based on a 32 kHz clock source (4 * 1/32 kHz = 125 µs)

4.3.13 I²C Electrical Specifications

This section describes the electrical information of the I²C Module.

4.3.13.1 I²C Module Timing

Figure 41 depicts the timing of I²C module. Table 40 lists the I²C module timing parameters where the I/O supply is 2.7 V. 1

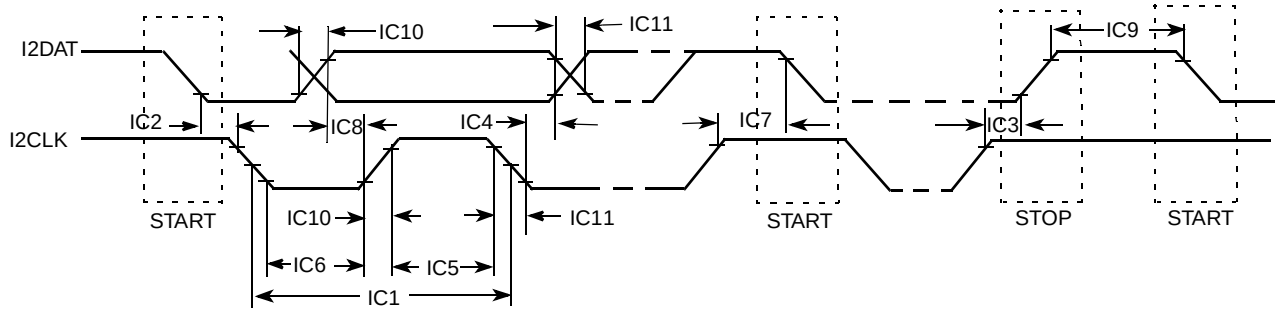


Figure 41. I²C Bus Timing Diagram

Table 40. I²C Module Timing Parameters—I²C Pin I/O Supply=2.7 V

ID	Parameter	Standard Mode		Fast Mode		Unit
		Min	Max	Min	Max	
IC1	I2CLK cycle time	10	—	2.5	—	µs
IC2	Hold time (repeated) START condition	4.0	—	0.6	—	µs
IC3	Set-up time for STOP condition	4.0	—	0.6	—	µs
IC4	Data hold time	0[1]	3.45[2]	0[1]	0.9[2]	µs
IC5	HIGH Period of I2CLK Clock	4.0	—	0.6	—	µs
IC6	LOW Period of the I2CLK Clock	4.7	—	1.3	—	µs
IC7	Set-up time for a repeated START condition	4.7	—	0.6	—	µs
IC8	Data set-up time	250	—	100[3]	—	ns
IC9	Bus free time between a STOP and START condition	4.7	—	1.3	—	µs
IC10	Rise time of both I2DAT and I2CLK signals	—	1000	$20+0.1C_b$[4]	300	ns
IC11	Fall time of both I2DAT and I2CLK signals	—	300	$20+0.1C_b$[4]	300	ns
IC12	Capacitive load for each bus line (C_b)	—	400	—	400	pF

1 A device must internally provide a hold time of at least 300 ns for I2DAT signal in order to bridge the undefined region of the falling edge of I2CLK.

2 The maximum hold time has to be met only if the device does not stretch the LOW period (ID IC6) of the I2CLK signal.

3 A Fast-mode I²C-bus device can be used in a standard-mode I²C-bus system, but the requirement of set-up time (ID IC7) of 250 ns must then be met. This will automatically be the case if the device does not stretch the LOW period of the I2CLK signal. If such a device does stretch the LOW period of the I2CLK signal, it must output the next data bit to the I2DAT line max_rise_time (ID No IC10) + data_setup_time (ID No IC8) = 1000 + 250 = 1250 ns (according to the Standard-mode I²C-bus specification) before the I2CLK line is released.

4 C_b = total capacitance of one bus line in pF.

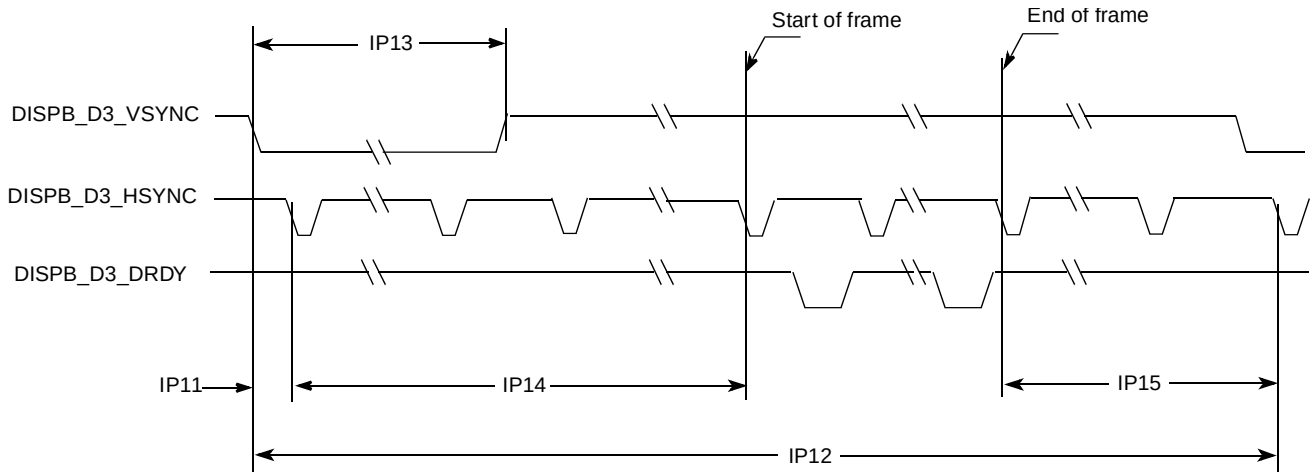


Figure 47. TFT Panels Timing Diagram—Vertical Sync Pulse

Table 44 shows timing parameters of signals presented in Figure 46 and Figure 47.

Table 44. Synchronous Display Interface Timing Parameters—Pixel Level

ID	Parameter	Symbol	Value	Units
IP5	Display interface clock period	Tdicp	Tdicp[1]	ns
IP6	Display pixel clock period	Tdpcp	(DISP3_IF_CLK_CNT_D+1) * Tdicp	ns
IP7	Screen width	Tsw	(SCREEN_WIDTH+1) * Tdpcp	ns
IP8	HSYNC width	Thsw	(H_SYNC_WIDTH+1) * Tdpcp	ns
IP9	Horizontal blank interval 1	Thbi1	BGXP * Tdpcp	ns
IP10	Horizontal blank interval 2	Thbi2	(SCREEN_WIDTH – BGXP – FW) * Tdpcp	ns
IP11	HSYNC delay	Thsd	H_SYNC_DELAY * Tdpcp	ns
IP12	Screen height	Tsh	(SCREEN_HEIGHT+1) * Tsw	ns
IP13	VSYNC width	Tvsw	if V_SYNC_WIDTH_L = 0 than (V_SYNC_WIDTH+1) * Tdpcp else (V_SYNC_WIDTH+1) * Tsw	ns
IP14	Vertical blank interval 1	Tvbi1	BGYP * Tsw	ns
IP15	Vertical blank interval 2	Tvbi2	(SCREEN_HEIGHT – BGYP – FH) * Tsw	ns

1 Display interface clock period immediate value.

$$Tdicp = \begin{cases} T_{HSP\_CLK} \cdot \frac{DISP3\_IF\_CLK\_PER\_WR}{HSP\_CLK\_PERIOD}, & \text{for integer } \frac{DISP3\_IF\_CLK\_PER\_WR}{HSP\_CLK\_PERIOD} \\ T_{HSP\_CLK} \cdot \left(floor\left[\frac{DISP3\_IF\_CLK\_PER\_WR}{HSP\_CLK\_PERIOD} \right] + 0.5 \pm 0.5 \right), & \text{for fractional } \frac{DISP3\_IF\_CLK\_PER\_WR}{HSP\_CLK\_PERIOD} \end{cases}$$

Display interface clock period average value.

$$\overline{T}dicp = T_{HSP\_CLK} \cdot \frac{DISP3\_IF\_CLK\_PER\_WR}{HSP\_CLK\_PERIOD}$$

Table 46. Sharp Synchronous Display Interface Timing Parameters—Pixel Level

ID	Parameter	Symbol	Value	Units
IP21	SPL rise time	Tsplr	(BGXP – 1) * Tdpcp	ns
IP22	CLS rise time	Tclsr	CLS_RISE_DELAY * Tdpcp	ns
IP23	CLS fall time	Tclsf	CLS_FALL_DELAY * Tdpcp	ns
IP24	CLS rise and PS fall time	Tpsf	PS_FALL_DELAY * Tdpcp	ns
IP25	PS rise time	Tpsr	PS_RISE_DELAY * Tdpcp	ns
IP26	REV toggle time	Trev	REV_TOGGLE_DELAY * Tdpcp	ns

4.3.15.4 Synchronous Interface to Dual-Port Smart Displays

Functionality and electrical characteristics of the synchronous interface to dual-port smart displays are identical to parameters of the synchronous interface. See Section 4.3.15.2.2, “Interface to Active Matrix TFT LCD Panels, Electrical Characteristics” on page 55.

4.3.15.4.1 Interface to a TV Encoder, Functional Description

The interface has an 8-bit data bus, transferring a single 8-bit value (Y/U/V) in each cycle. The bits D7–D0 of the value are mapped to bits LD17–LD10 of the data bus, respectively. Figure 50 depicts the interface timing,

- The frequency of the clock DISPB_D3_CLK is 27 MHz (within 10%).
- The DISPB_D3_HSYNC, DISPB_D3_VSYNC and DISPB_D3_DRDY signals are active low.
- The transition to the next row is marked by the negative edge of the DISPB_D3_HSYNC signal. It remains low for a single clock cycle.
- The transition to the next field/frame is marked by the negative edge of the DISPB_D3_VSYNC signal. It remains low for at least one clock cycle.
  - At a transition to an odd field (of the next frame), the negative edges of DISPB_D3_VSYNC and DISPB_D3_HSYNC coincide.
  - At a transition to an even field (of the same frame), they do not coincide.
- The active intervals—during which data is transferred—are marked by the DISPB_D3_HSYNC signal being high.

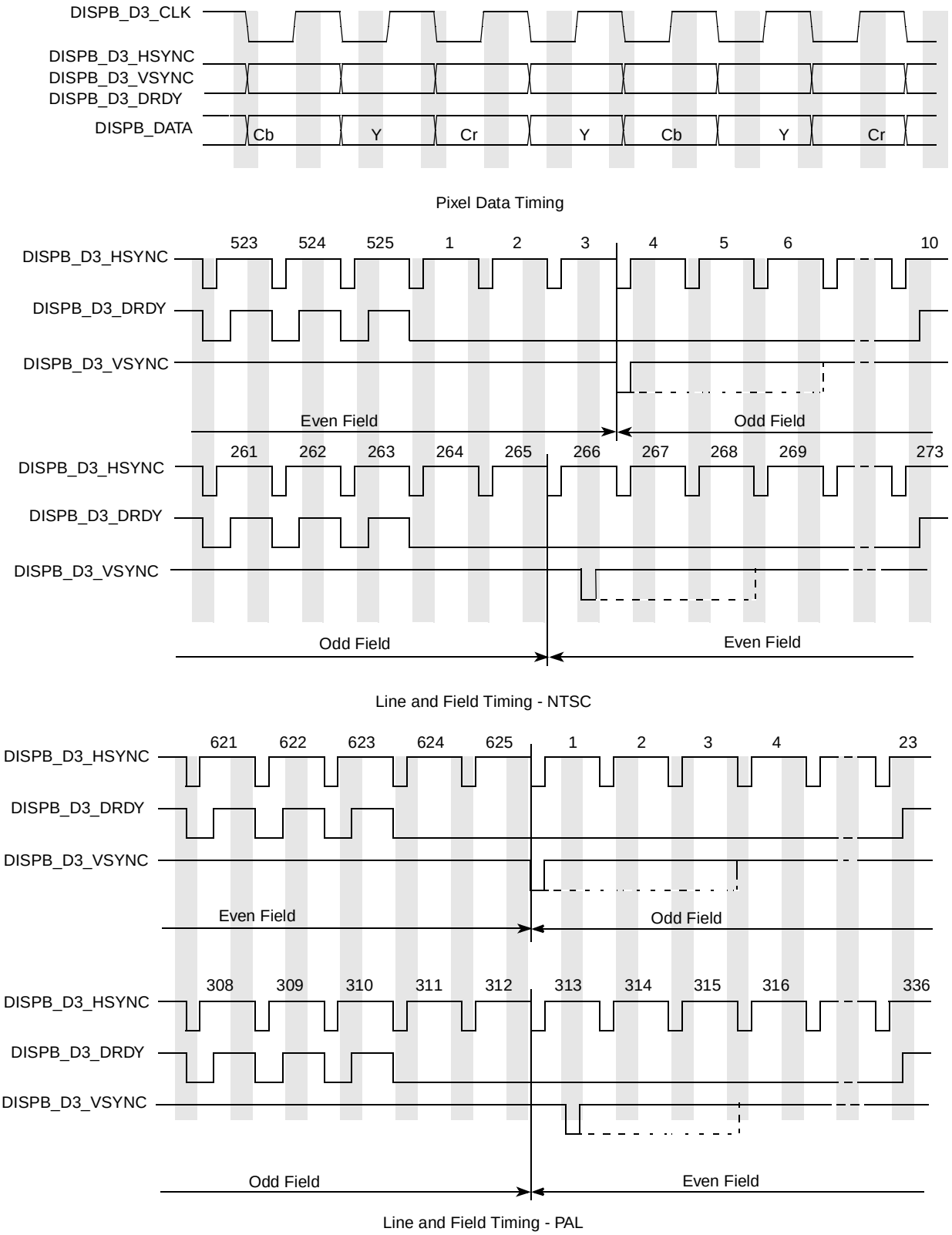


Figure 50. TV Encoder Interface Timing Diagram

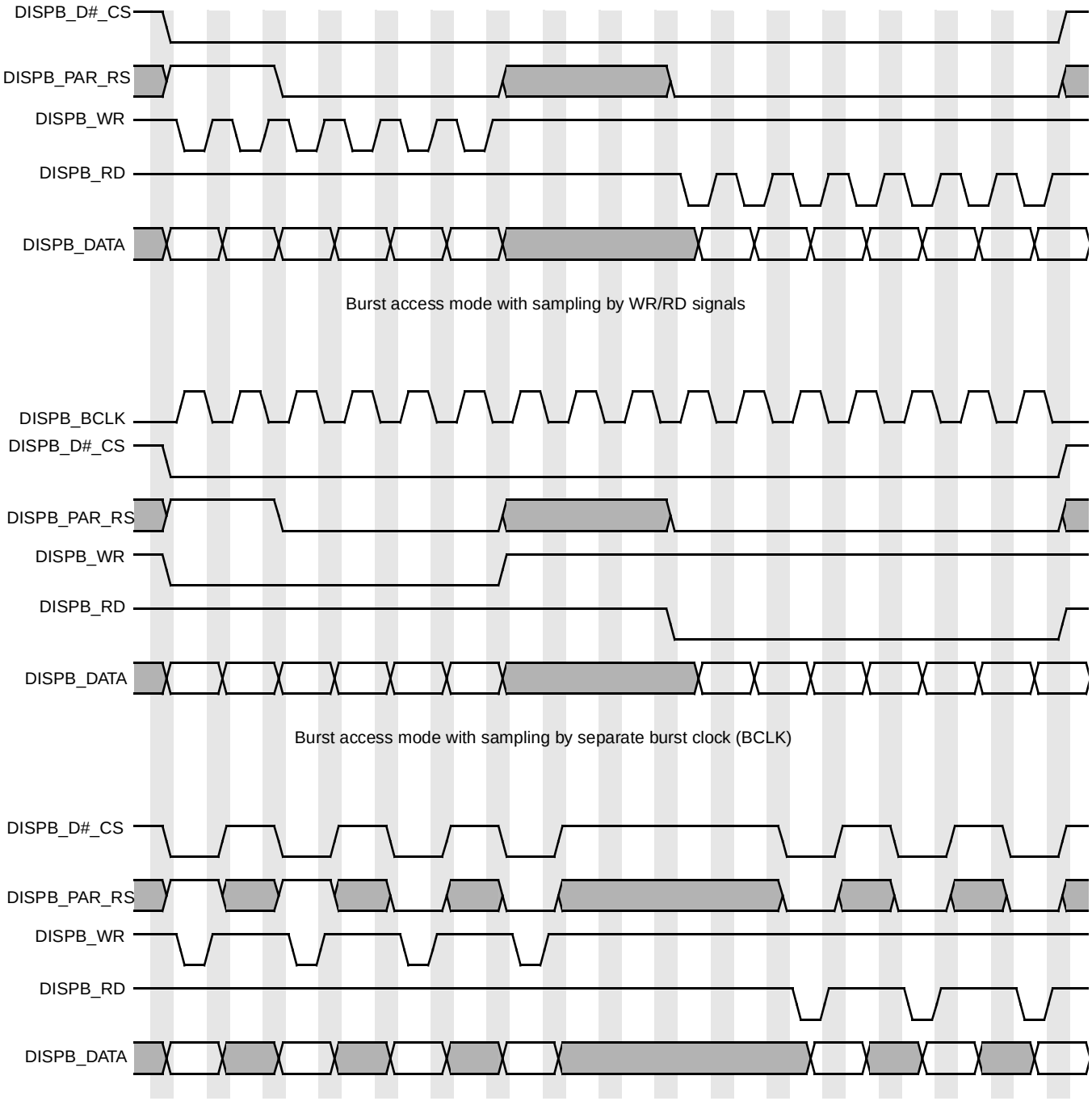


Figure 52. Asynchronous Parallel System 80 Interface (Type 2) Burst Mode Timing Diagram

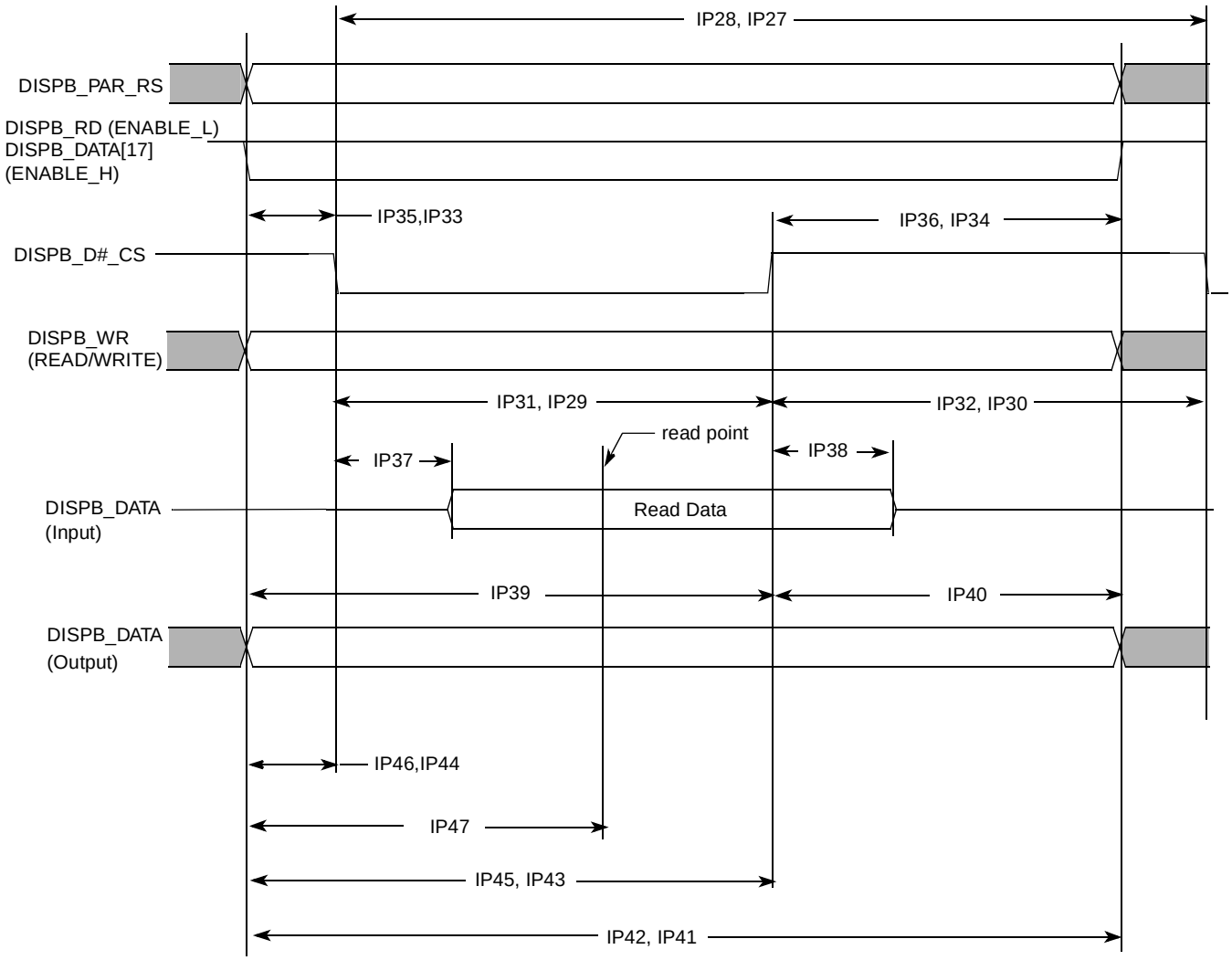


Figure 58. Asynchronous Parallel System 68k Interface (Type 1) Timing Diagram

4.3.22.2 SSI Receiver Timing with Internal Clock

Figure 81 depicts the SSI receiver timing with internal clock, and Table 58 lists the timing parameters.

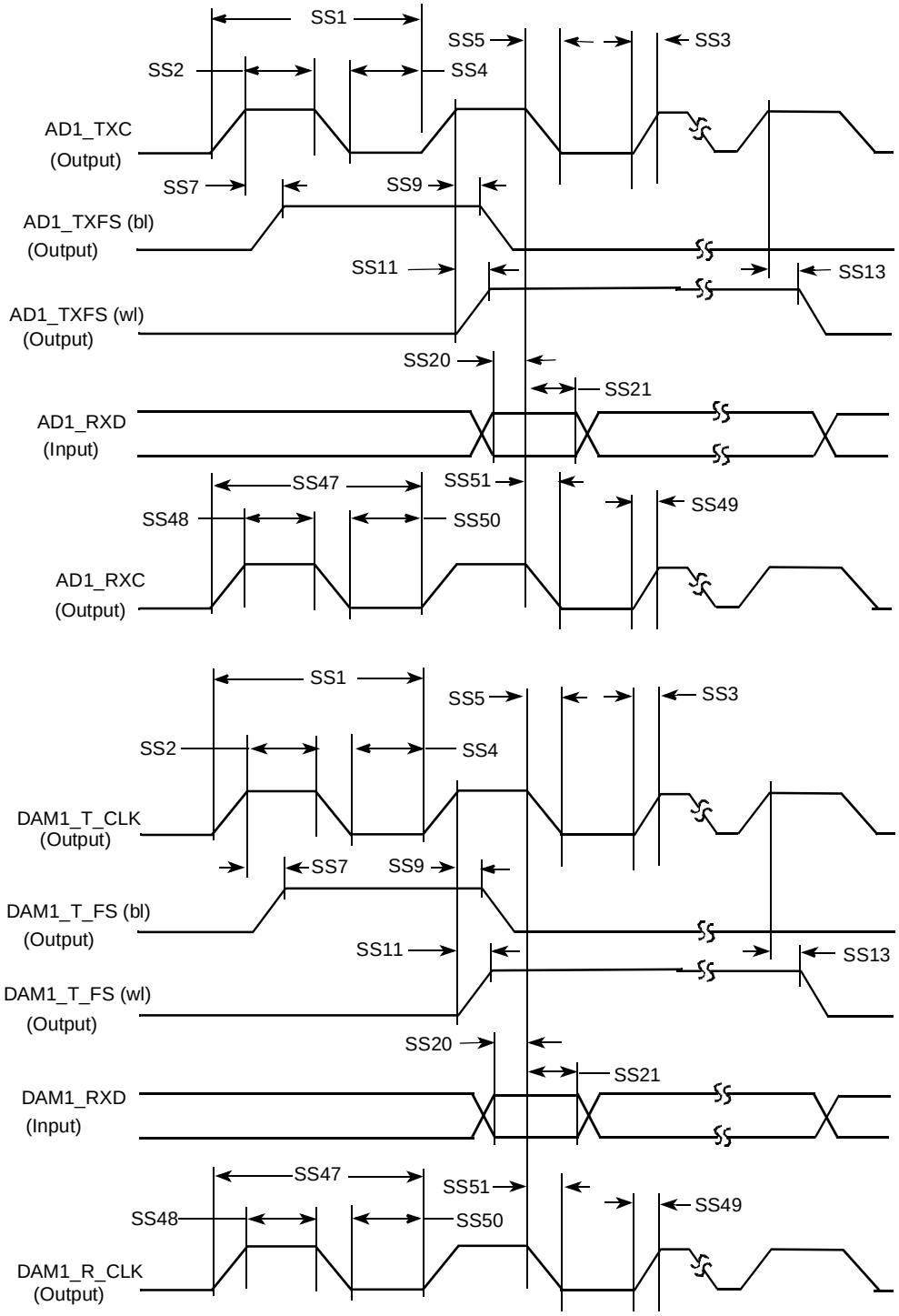


Figure 81. SSI Receiver with Internal Clock Timing Diagram

4.3.22.4 SSI Receiver Timing with External Clock

Figure 83 depicts the SSI receiver timing with external clock, and Table 60 lists the timing parameters.

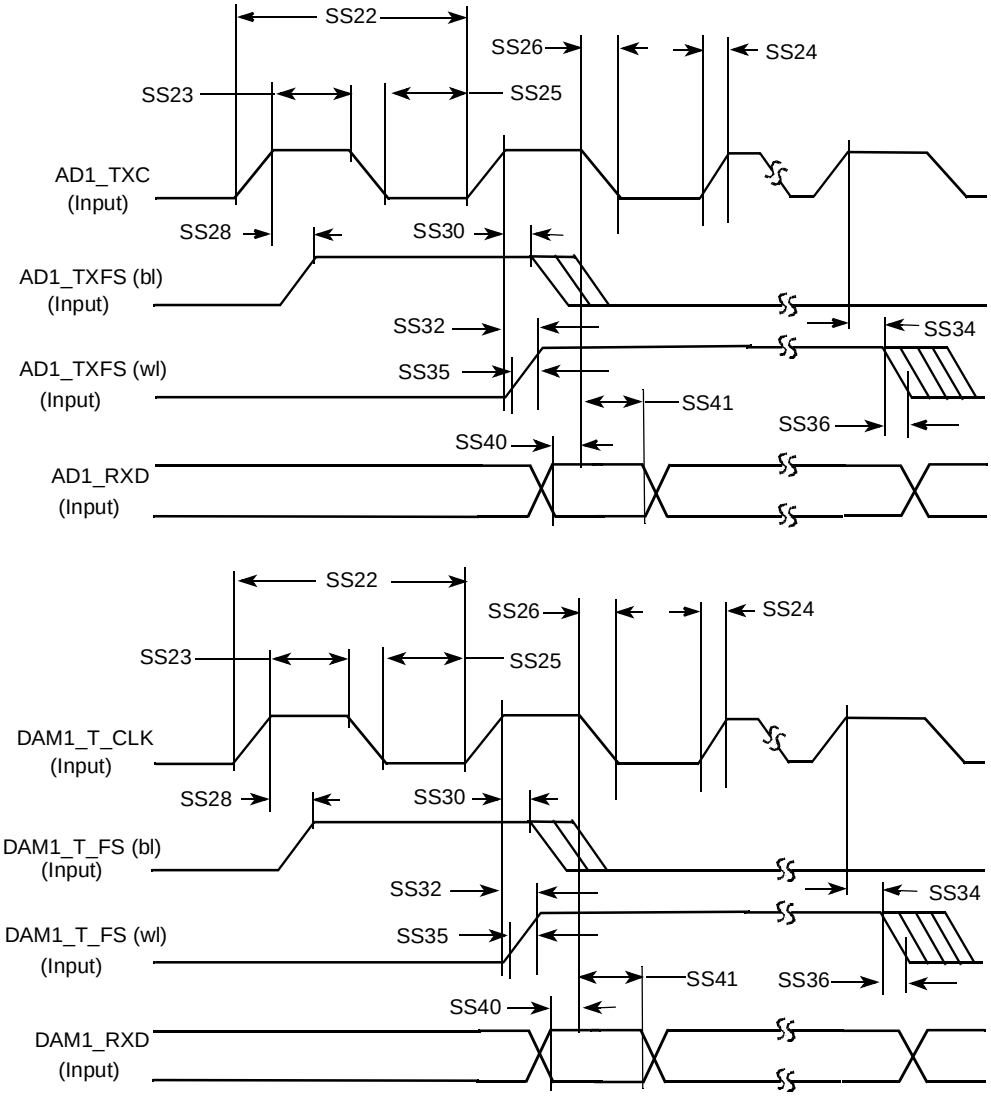


Figure 83. SSI Receiver with External Clock Timing Diagram

Table 60. SSI Receiver with External Clock Timing Parameters

ID	Parameter	Min	Max	Unit
External Clock Operation				
SS22	(Tx/Rx) CK clock period	81.4	—	ns
SS23	(Tx/Rx) CK clock high period	36.0	—	ns
SS24	(Tx/Rx) CK clock rise time	—	6.0	ns
SS25	(Tx/Rx) CK clock low period	36.0	—	ns
SS26	(Tx/Rx) CK clock fall time	—	6.0	ns

How to Reach Us:

Home Page:
www.freescale.com

E-mail:
support@freescale.com

USA/Europe or Locations Not Listed:
Freescale Semiconductor
Technical Information Center, CH370
1300 N. Alma School Road
Chandler, Arizona 85224
+1-800-521-6274 or +1-480-768-2130
support@freescale.com

Europe, Middle East, and Africa:
Freescale Halbleiter Deutschland GmbH
Technical Information Center
Schatzbogen 7
81829 Muenchen, Germany
+44 1296 380 456 (English)
+46 8 52200080 (English)
+49 89 92103 559 (German)
+33 1 69 35 48 48 (French)
support@freescale.com

Japan:
Freescale Semiconductor Japan Ltd.
Headquarters
ARCO Tower 15F
1-8-1, Shimo-Meguro, Meguro-ku,
Tokyo 153-0064, Japan
0120 191014 or +81 3 5437 9125
support.japan@freescale.com

Asia/Pacific:
Freescale Semiconductor Hong Kong Ltd.
Technical Information Center
2 Dai King Street
Tai Po Industrial Estate
Tai Po, N.T., Hong Kong
+800 2666 8080
support.asia@freescale.com

***For Literature Requests Only*:**
Freescale Semiconductor Literature Distribution Center
P.O. Box 5405
Denver, Colorado 80217
1-800-521-6274 or 303-675-2140
Fax: 303-675-2150
LDCForFreescaleSemiconductor@hibbertgroup.com

Document Number: MCIMX31C
Rev. 4.3
2/2010

freescale™
semiconductor