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"[Embedded - Microcontrollers](#)" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "[Embedded - Microcontrollers](#)"

Details

Product Status	Active
Core Processor	PIC
Core Size	8-Bit
Speed	64MHz
Connectivity	I ² C, LINbus, SPI, UART/USART
Peripherals	Brown-out Detect/Reset, DMA, HLVD, POR, PWM, WDT
Number of I/O	25
Program Memory Size	16KB (8K x 16)
Program Memory Type	FLASH
EEPROM Size	256 x 8
RAM Size	1K x 8
Voltage - Supply (Vcc/Vdd)	2.3V ~ 5.5V
Data Converters	A/D 24x12b; D/A 1x5b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	28-UFQFN Exposed Pad
Supplier Device Package	28-UQFN (4x4)
Purchase URL	https://www.e-xfl.com/product-detail/microchip-technology/pic18f24k42-e-mv

8.13 Power Control (PCON0/PCON1) Register

The Power Control (PCON0/PCON1) register contains flag bits to differentiate between a:

- Brown-out Reset ($\overline{\text{BOR}}$)
- Power-on Reset ($\overline{\text{POR}}$)
- Reset Instruction Reset ($\overline{\text{RI}}$)
- MCLR Reset ($\overline{\text{RMCLR}}$)
- Watchdog Timer Reset ($\overline{\text{RWDI}}$)
- Watchdog Window Violation ($\overline{\text{WDTWV}}$)
- Stack Underflow Reset (STKUNF)
- Stack Overflow Reset (STKOVF)
- Memory Violation Reset ($\overline{\text{MEMV}}$)

The PCON0/1 register bits are shown in [Register 8-2](#) and [Register 8-3](#). Hardware will change the corresponding register bit during the Reset process; if the Reset was not caused by the condition, the bit remains unchanged ([Table 8-3](#)).

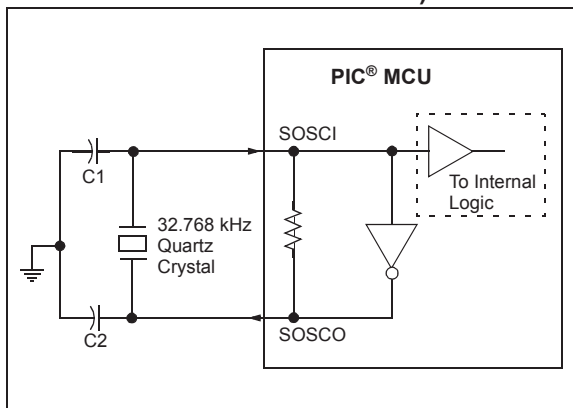
Software should reset the bit to the inactive state after restart (hardware will not reset the bit). Software may also set any PCON0 bit to the active state, so that user code may be tested, but no Reset action will be generated.

9.2.1.5 Secondary Oscillator

The secondary oscillator is a separate oscillator block that can be used as an alternate system clock source. The secondary oscillator is optimized for 32.768 kHz, and can be used with an external crystal oscillator connected to the SOSC1 and SOSCO device pins, or an external clock source connected to the SOSCIN pin. The secondary oscillator can be selected during run-time using clock switching. Refer to [Section 9.3 “Clock Switching”](#) for more information.

Two power modes are available for the secondary oscillator. These modes are selected with the SOSCPWR (OSCCON3<6>). Clearing this bit selects the lower Crystal Gain mode which provides lowest microcontroller power consumption. Setting this bit enables a higher Gain mode to support faster crystal start-up or crystals with higher ESR.

FIGURE 9-5: QUARTZ CRYSTAL OPERATION (SECONDARY OSCILLATOR)



Note 1: Quartz crystal characteristics vary according to type, package and manufacturer. The user should consult the manufacturer data sheets for specifications and recommended application.

2: Always verify oscillator performance over the VDD and temperature range that is expected for the application.

3: For oscillator design assistance, reference the following Microchip Application Notes:

- AN826, “Crystal Oscillator Basics and Crystal Selection for PIC® and PIC® Devices” (DS00826)
- AN849, “Basic PIC® Oscillator Design” (DS00849)
- AN943, “Practical PIC® Oscillator Analysis and Design” (DS00943)
- AN949, “Making Your Oscillator Work” (DS00949)
- TB097, “Interfacing a Micro Crystal MS1V-T1K 32.768 kHz Tuning Fork Crystal to a PIC16F690/SS” (DS91097)
- AN1288, “Design Practices for Low-Power External Oscillators” (DS01288)

10.0 REFERENCE CLOCK OUTPUT MODULE

The reference clock output module provides the ability to send a clock signal to the clock reference output pin (CLKR). The reference clock output can also be used as a signal for other peripherals, such as the Data Signal Modulator (DSM), Memory Scanner and Timer module.

The reference clock output module has the following features:

- Selectable clock source using the CLKRCLK register
- Programmable clock divider
- Selectable duty cycle

FIGURE 10-1: CLOCK REFERENCE BLOCK DIAGRAM

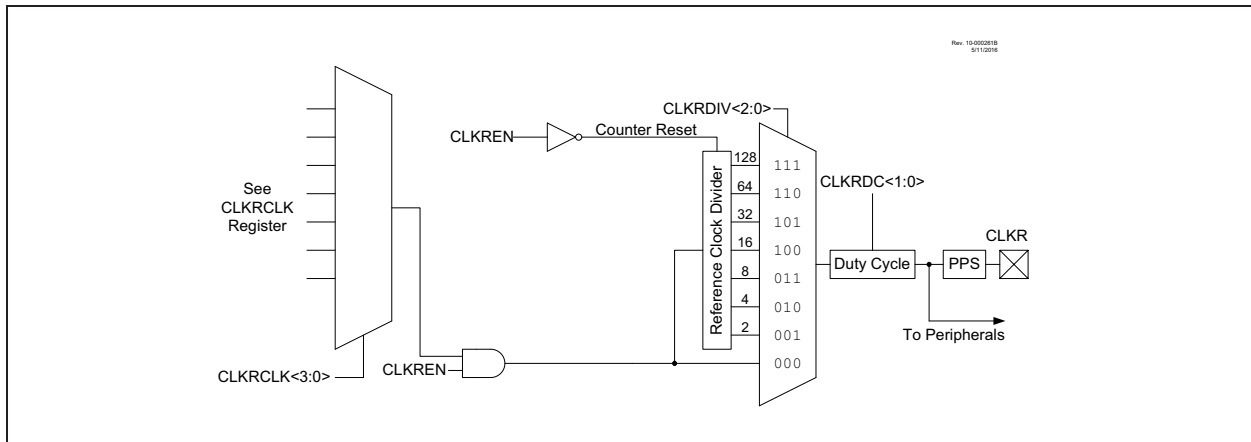
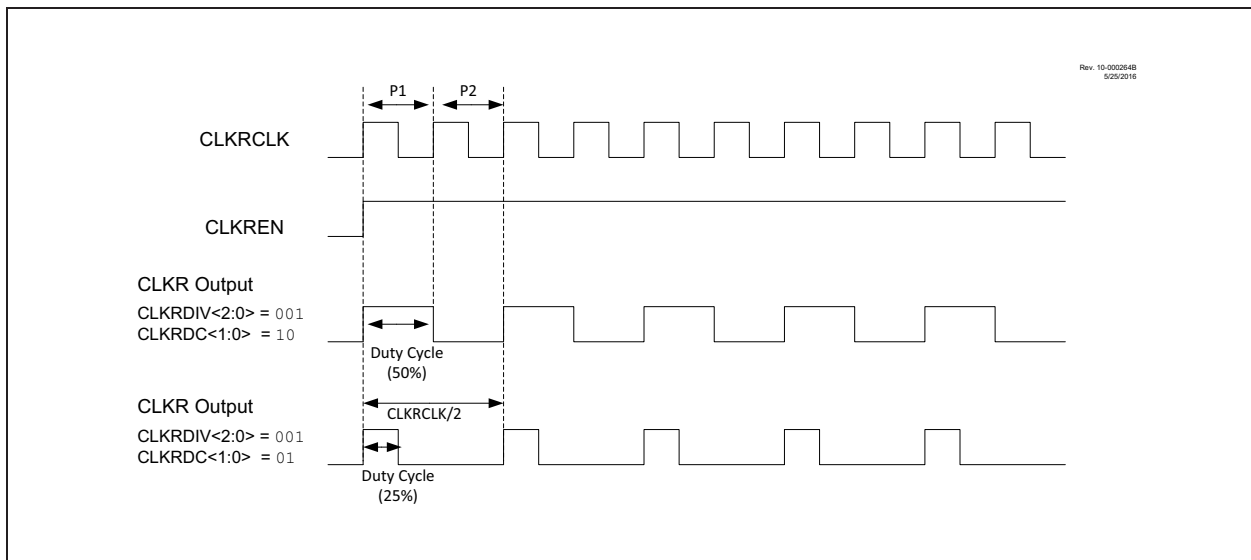


FIGURE 10-2: CLOCK REFERENCE TIMING



11.5 Context Saving

The Interrupt controller supports a two-level deep context saving (Main routine context and Low ISR context). Refer to state machine shown in [Figure 11-6](#) for details.

The Program Counter (PC) is saved on the dedicated device PC stack. CPU registers saved include STATUS, WREG, BSR, FSR0/1/2, PRODL/H and PCLATH/U.

After WREG has been saved to the context registers, the resolved vector number of the interrupt source to be serviced is copied into WREG. Context save and restore operation is completed by the interrupt controller based on current state of the interrupts and the order in which they were sent to the CPU.

Context save/restore works the same way in both states of MVECEN. When IPEN = 0, there is only one level interrupt active. Hence, only the main context is saved when an interrupt is received.

11.5.1 ACCESSING SHADOW REGISTERS

The Interrupt controller automatically saves the context information in the shadow registers available in Bank 56. Both the saved context values (i.e., main routine and low ISR) can be accessed using the same set of shadow registers. By clearing the SHADLO bit in the SHADCON register ([Register 11-43](#)), the CPU register values saved for main routine context can accessed, and by setting the SHADLO bit of the CPU register, values saved for low ISR context can accessed. Low ISR context is automatically restored to the CPU registers upon exiting the high ISR. Similarly, the main context is automatically restored to the CPU registers upon exiting the low ISR.

The Shadow registers in Bank 56 are readable and writable, so if the user desires to modify the context then the corresponding shadow register should be modified and the value will be restored when exiting the ISR. Depending on the user's application, other registers may also need to be saved.

REGISTER 11-10: PIR7: PERIPHERAL INTERRUPT REGISTER 7

U-0	U-0	R/W/HS-0/0	R/W/HS-0/0	R/W/HS-0/0	U-0	R/W/HS-0/0	R/W/HS-0/0
—	—	INT2IF	CLC2IF	CWG2IF	—	CCP2IF	TMR4IF
bit 7						bit 0	

Legend:

R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	HS = Bit is set in hardware

bit 7-6	Unimplemented: Read as '0'
bit 5	INT2IF: External Interrupt 2 Interrupt Flag bit 1 = Interrupt has occurred (must be cleared by software) 0 = Interrupt event has not occurred
bit 4	CLC2IF: CLC2 Interrupt Flag bit 1 = Interrupt has occurred (must be cleared by software) 0 = Interrupt event has not occurred
bit 3	CWG2IF: CWG2 Interrupt Flag bit 1 = Interrupt has occurred (must be cleared by software) 0 = Interrupt event has not occurred
bit 2	Unimplemented: Read as '0'
bit 1	CCP2IF: CCP2 Interrupt Flag bit 1 = Interrupt has occurred (must be cleared by software) 0 = Interrupt event has not occurred
bit 0	TMR4IF: TMR4 Interrupt Flag bit 1 = Interrupt has occurred (must be cleared by software) 0 = Interrupt event has not occurred

Note: Interrupt flag bits get set when an interrupt condition occurs, regardless of the state of its corresponding enable bit, or the global enable bit. User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt.

15.1.2 CONTROL REGISTERS

Several control registers are used in conjunction with the `TBLRD` and `TBLWT` instructions. These include the following registers:

- NVMCON1 register
- NVMCON2 register
- TABLAT register
- TBLPTR registers

15.1.2.1 NVMCON1 and NVMCON2 Registers

The NVMCON1 register ([Register 15-1](#)) is the control register for memory accesses. The NVMCON2 register is not a physical register; it is used exclusively in the memory write and erase sequences. Reading NVMCON2 will read all '0's.

The `REG<1:0>` control bits determine if the access will be to Data EEPROM Memory locations, PFM locations or User IDs, Configuration bits, Rev ID and Device ID. When `REG<1:0> = 00`, any subsequent operations will operate on the Data EEPROM Memory. When `REG<1:0> = 10`, any subsequent operations will operate on the program memory. When `REG<1:0> = x1`, any subsequent operations will operate on the Configuration bits, User IDs, Rev ID and Device ID.

The `FREE` bit allows the program memory erase operation. When the `FREE` bit is set, an erase operation is initiated on the next `WR` command. When `FREE` is clear, only writes are enabled. This bit is applicable only to the PFM and not to data EEPROM.

When set, the `WREN` bit will allow a program/erase operation. The `WREN` bit is cleared on power-up.

The `WRERR` bit is set by hardware when the `WR` bit is set and is cleared when the internal programming timer expires and the write operation is successfully complete.

The `WR` control bit initiates erase/write cycle operation when the `REG<1:0>` bits point to the Data EEPROM Memory location, and it initiates a write operation when the `REG<1:0>` bits point to the PFM location. The `WR` bit cannot be cleared by firmware; it can only be set by firmware. Then the `WR` bit is cleared by hardware at the completion of the write operation.

The `NVMIF` Interrupt Flag bit is set when the write is complete. The `NVMIF` flag stays set until cleared by firmware.

15.1.2.2 TABLAT – Table Latch Register

The Table Latch (`TABLAT`) is an 8-bit register mapped into the SFR space. The Table Latch register is used to hold 8-bit data during data transfers between program memory and data RAM.

15.1.2.3 TBLPTR – Table Pointer Register

The Table Pointer (`TBLPTR`) register addresses a byte within the program memory. The `TBLPTR` is comprised of three SFR registers: Table Pointer Upper Byte, Table Pointer High Byte and Table Pointer Low Byte (`TBLPTRU:TBLPTRH:TBLPTRL`). These three registers join to form a 22-bit wide pointer. The low-order 21 bits allow the device to address up to 2 Mbytes of program memory space. The 22nd bit allows access to the Device ID, the User ID and the Configuration bits.

The Table Pointer register, `TBLPTR`, is used by the `TBLRD` and `TBLWT` instructions. These instructions can update the `TBLPTR` in one of four ways based on the table operation. These operations on the `TBLPTR` affect only the low-order 21 bits.

15.1.2.4 Table Pointer Boundaries

`TBLPTR` is used in reads, writes and erases of the Program Flash Memory.

When a `TBLRD` is executed, all 22 bits of the `TBLPTR` determine which byte is read from program memory directly into the `TABLAT` register.

When a `TBLWT` is executed the byte in the `TABLAT` register is written, not to memory but, to a holding register in preparation for a program memory write. The holding registers constitute a write block which varies depending on the device (see [Table 15-3](#)). The 3, 4, or 5 LSBs of the `TBLPTRL` register determine which specific address within the holding register block is written to. The MSBs of the Table Pointer have no effect during `TBLWT` operations.

When a program memory write is executed the entire holding register block is written to the memory at the address determined by the MSBs of the `TBLPTR`. The 3, 4, or 5 LSBs are ignored during memory writes. For more detail, see [Section 15.1.6 “Writing to Program Flash Memory”](#).

[Figure 15-3](#) describes the relevant boundaries of `TBLPTR` based on Program Flash Memory operations.

TABLE 16-2: SCANNER OPERATING MODES⁽¹⁾

TRIGEN	BURSTMD	Scanner Operation
0	0	Memory access is requested when the CRC module is ready to accept data; the request is granted if no other higher priority source request is pending.
1	0	Memory access is requested when the CRC module is ready to accept data and trigger selection is true; the request is granted if no other higher priority source request is pending.
x	1	Memory access is always requested, the request is granted if no other higher priority source request is pending.

Note 1: See [Section 3.1 “System Arbitration”](#) for Priority selection and [Section 3.2 “Memory Access Scheme”](#) for Memory Access Scheme.

REGISTER 16-12: SCANLADRU: SCAN LOW ADDRESS UPPER BYTE REGISTER

U-0	U-0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
—	—	LADR<21:16> ^(1,2)					
bit 7							bit 0

Legend:

R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set	‘0’ = Bit is cleared	

bit 7-6 **Unimplemented:** Read as ‘0’

bit 5-0 **LADR<21:16>:** Scan Start/Current Address bits^(1,2)

Upper bits of the current address to be fetched from, value increments on each fetch of memory.

Note 1: Registers SCANLADRU/H/L form a 22-bit value, but are not guarded for atomic or asynchronous access; registers should only be read or written while SGO = 0 (SCANCON0 register).

2: While SGO = 1 (SCANCON0 register), writing to this register is ignored.

REGISTER 16-13: SCANLADRH: SCAN LOW ADDRESS HIGH BYTE REGISTER

R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0	R/W-0/0
LADR<15:8> ^(1, 2)							
bit 7							bit 0

Legend:

R = Readable bit	W = Writable bit	U = Unimplemented bit, read as ‘0’
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
‘1’ = Bit is set	‘0’ = Bit is cleared	

bit 7-0 **LADR<15:8>:** Scan Start/Current Address bits^(1, 2)

Most Significant bits of the current address to be fetched from, value increments on each fetch of memory.

Note 1: Registers SCANLADRU/H/L form a 22-bit value, but are not guarded for atomic or asynchronous access; registers should only be read or written while SGO = 0 (SCANCON0 register).

2: While SGO = 1 (SCANCON0 register), writing to this register is ignored.

REGISTER 19-3: PPSLOCK: PPS LOCK REGISTER

U-0	U-0	U-0	U-0	U-0	U-0	U-0	R/W-0/0
—	—	—	—	—	—	—	PPSLOCKED
bit 7							bit 0

Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

u = Bit is unchanged

x = Bit is unknown

-n/n = Value at POR and BOR/Value at all other Resets

'1' = Bit is set

'0' = Bit is cleared

bit 7-1

Unimplemented: Read as '0'

bit 0

PPSLOCKED: PPS Locked bit

1 = PPS is locked. PPS selections can not be changed.

0 = PPS is not locked. PPS selections can be changed.

FIGURE 27-1: SMT BLOCK DIAGRAM

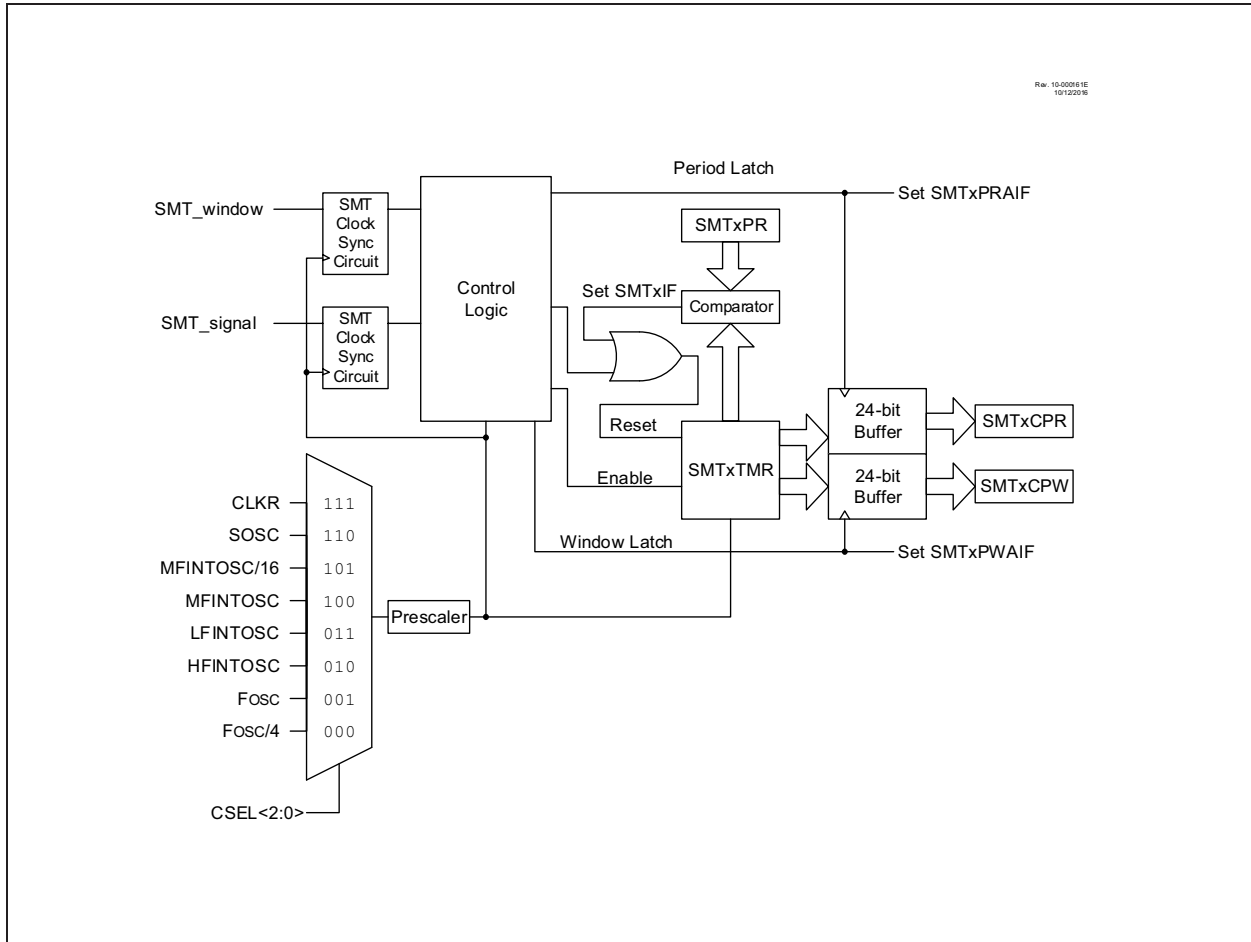
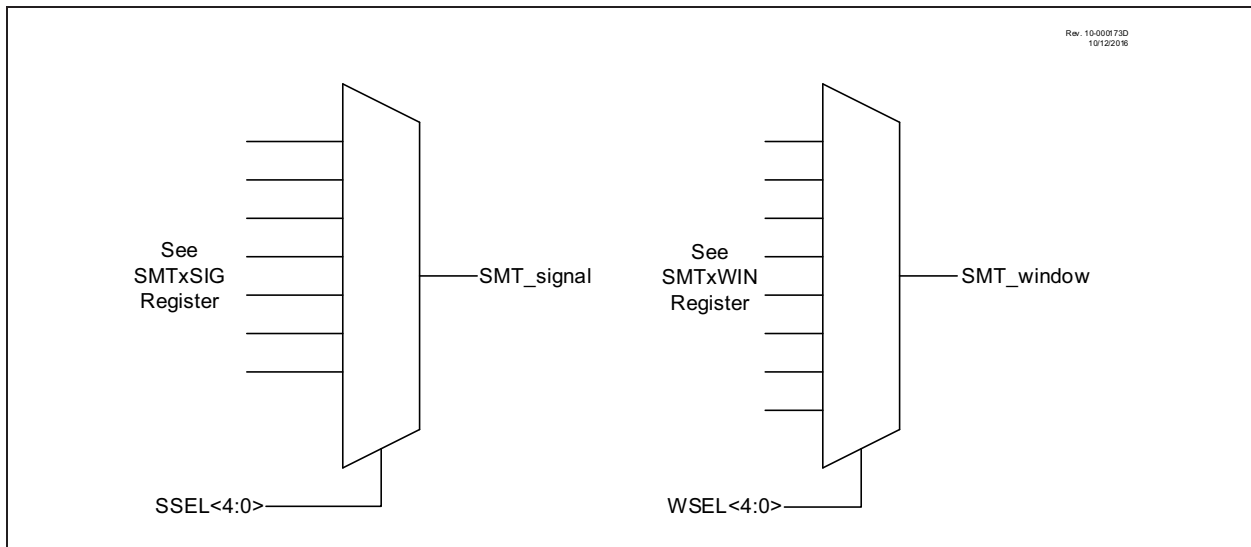


FIGURE 27-2: SMT SIGNAL AND WINDOW BLOCK DIAGRAM



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12/19/2013

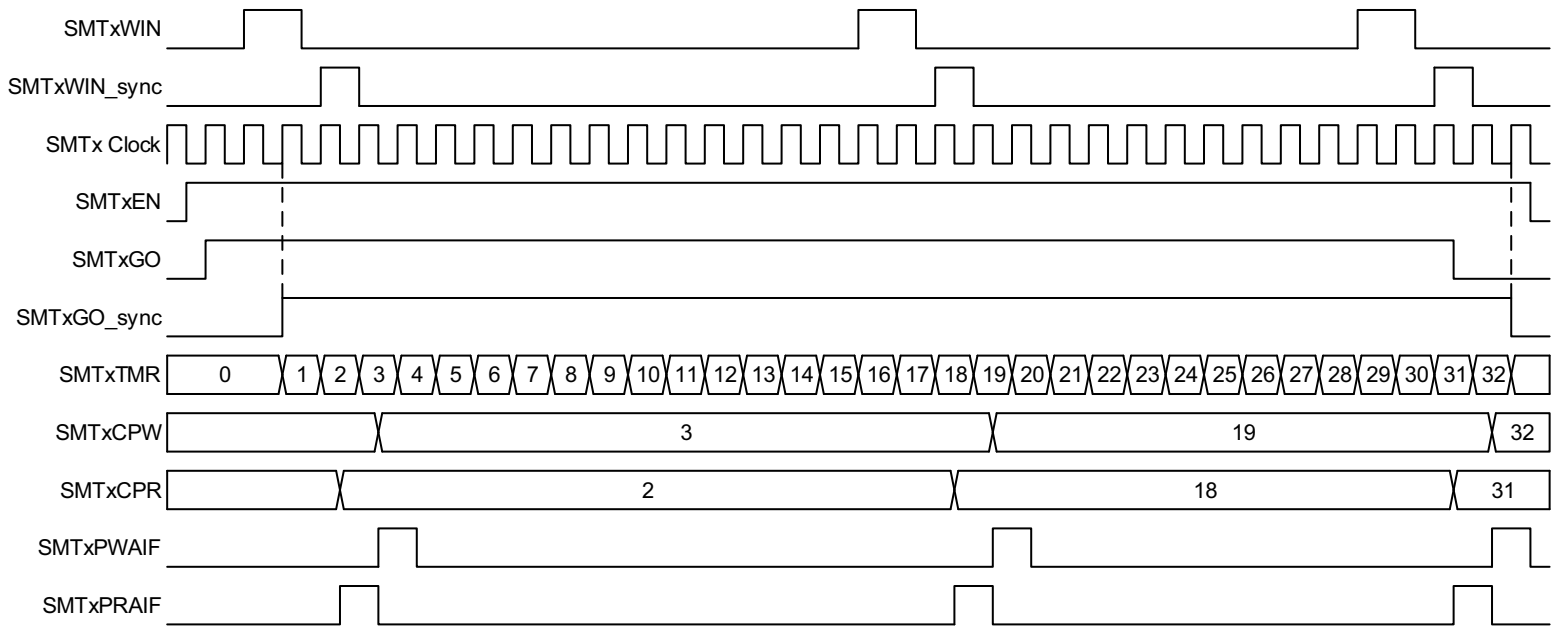


FIGURE 27-16: CAPTURE MODE REPEAT ACQUISITION TIMING DIAGRAM

FIGURE 28-12: DEAD-BAND OPERATION, CWGxDBR = 0x01, CWGxDBF = 0x02

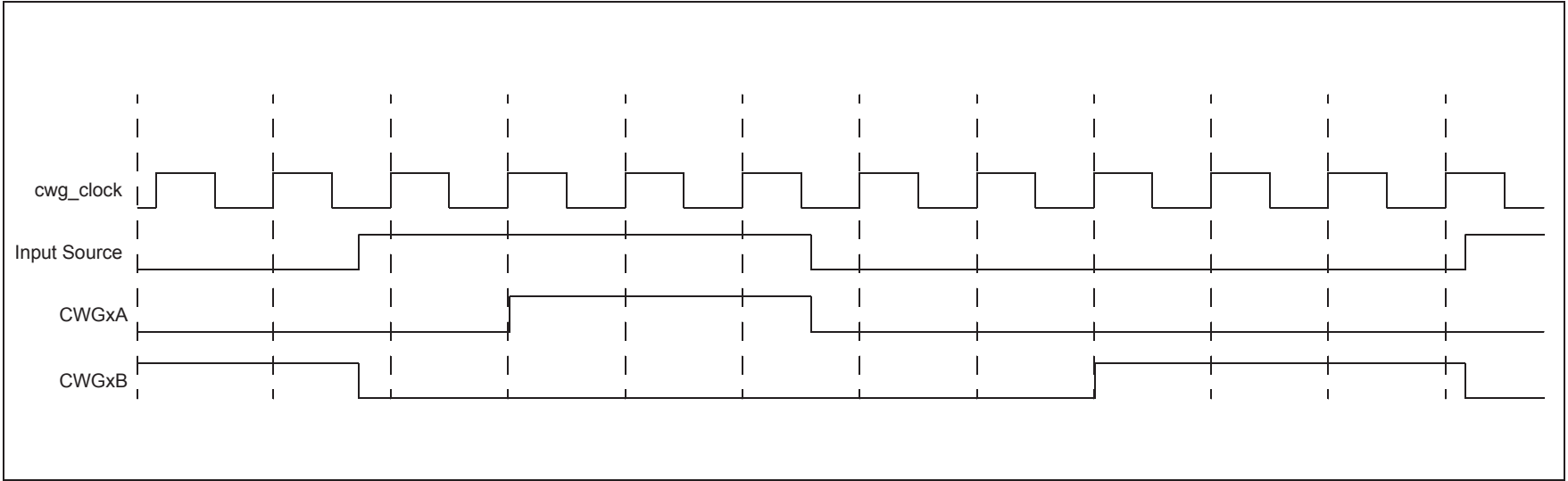


FIGURE 28-13: DEAD-BAND OPERATION, CWGxDBR = 0x03, CWGxDBF = 0x06, SOURCE SHORTER THAN DEAD BAND

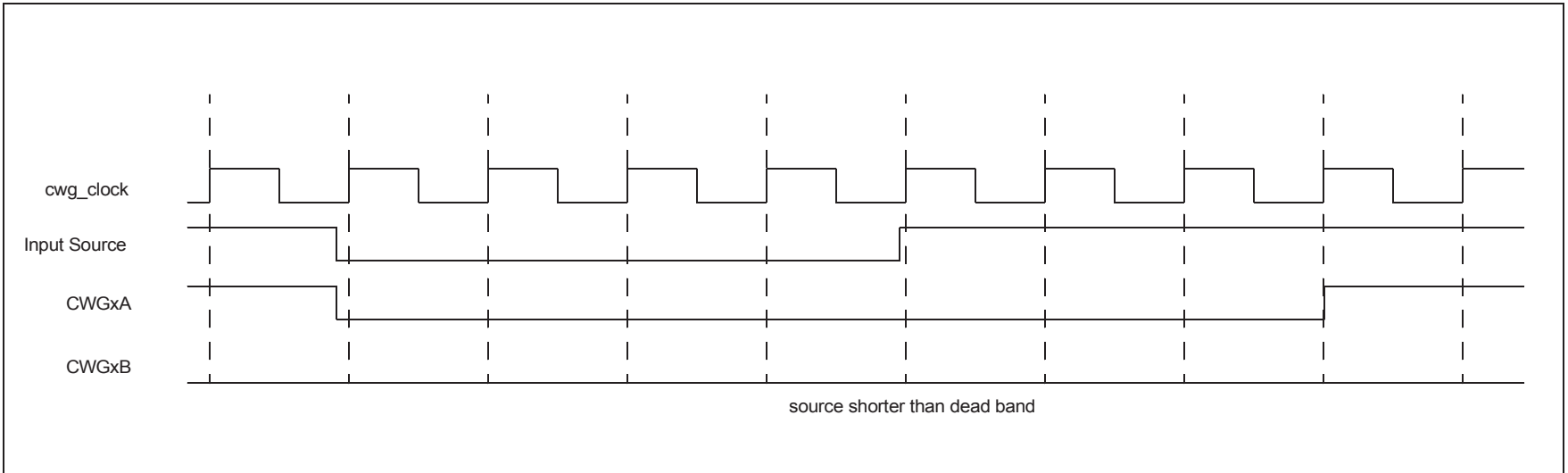
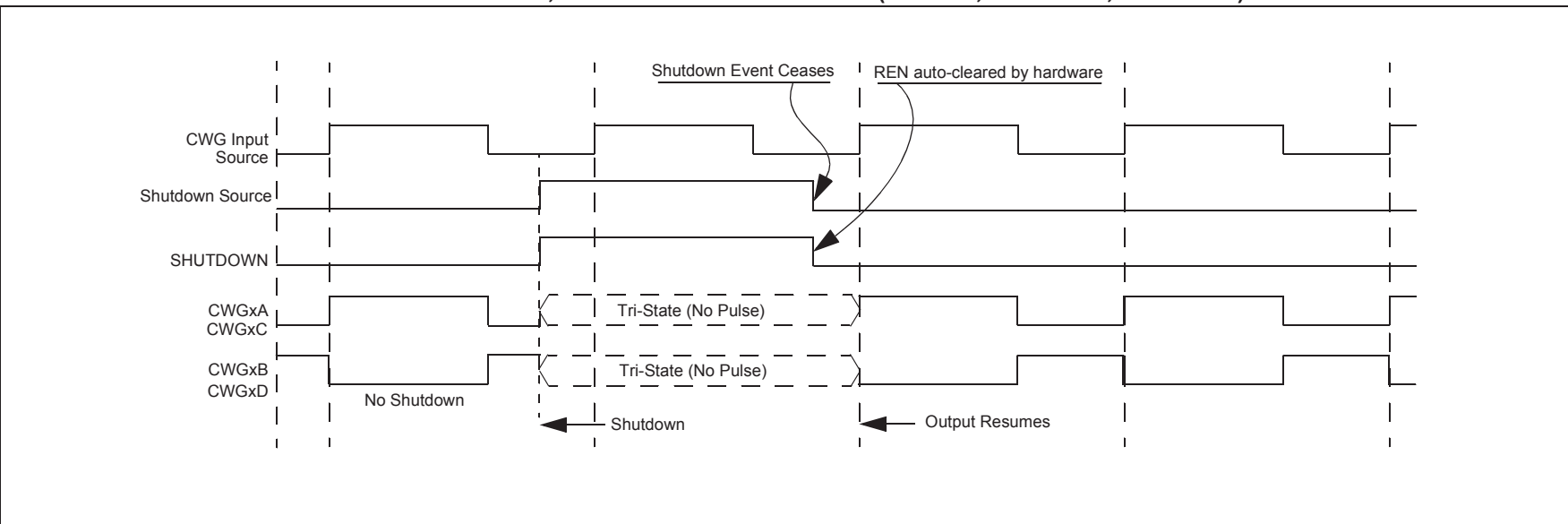


FIGURE 28-16: SHUTDOWN FUNCTIONALITY, AUTO-RESTART ENABLED (REN = 1, LSAC = 01, LSB0 = 01)



33.0 UNIVERSAL ASYNCHRONOUS RECEIVER TRANSMITTER (UART) WITH PROTOCOL SUPPORT

The Universal Asynchronous Receiver Transmitter (UART) module is a serial I/O communications peripheral. It contains all the clock generators, shift registers and data buffers necessary to perform an input or output serial data transfer, independent of device program execution. The UART, also known as a Serial Communications Interface (SCI), can be configured as a full-duplex asynchronous system or one of several automated protocols. Full-Duplex mode is useful for communications with peripheral systems, such as CRT terminals and personal computers.

Supported protocols include:

- LIN Master and Slave
- DMX mode
- DALI control gear and control device

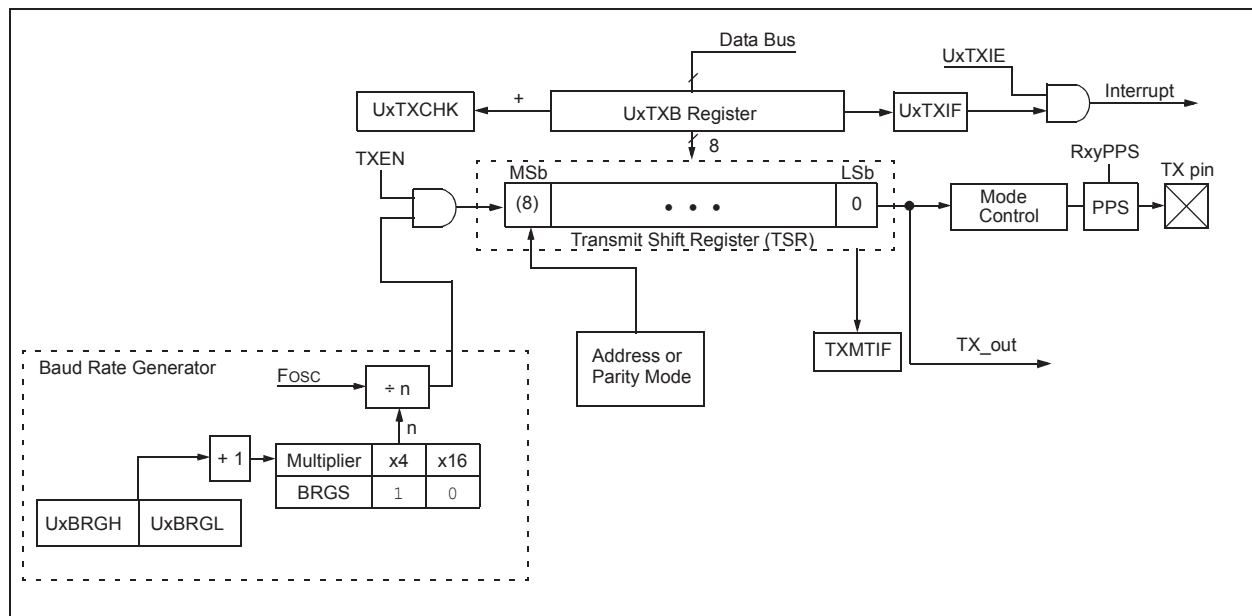
The UART module includes the following capabilities:

- Full-duplex asynchronous transmit and receive
- Two-character input buffer
- One-character output buffer
- Programmable 7-bit or 8-bit character length
- 9th bit Address detection
- 9th bit even or odd parity
- Input buffer overrun error detection
- Received character framing error detection
- Hardware and software flow control
- Automatic checksums
- Programmable 1, 1.5, and 2 Stop bits
- Programmable data polarity
- Manchester encoder/decoder
- Operation in Sleep
- Automatic detection and calibration of the baud rate
- Wake-up on Break reception
- Automatic and user timed Break period generation
- RX and TX inactivity timeouts (with Timer2)

Block diagrams of the UART transmitter and receiver are shown in [Figure 33-1](#) and [Figure 33-2](#).

The UART transmit output (TX_out) is available to the TX pin and internally to various peripherals.

FIGURE 33-1: UART TRANSMIT BLOCK DIAGRAM



PIC18(L)F24/25K42

35.1 I²C Features

- Inter-Integrated Circuit (I²C) interface supports the following modes in hardware:
 - Master mode
 - Slave mode with byte NACKing
 - Multi-Master mode
- Dedicated Address, Receive and Transmit buffers
- Up to four Slave addresses matching
- General Call address matching
- 7-bit and 10-bit addressing with masking
- Start, Restart, Stop, Address, Write, and ACK Interrupts
- Clock Stretching hardware for:
 - RX Buffer Full
 - TX Buffer Empty
 - After Address, Write, and ACK
- Bus Collision Detection with arbitration
- Bus Timeout Detection
- SDA hold time selection
- I²C, SMBus 2.0, and SMBus 3.0 input level selections

35.2 I²C Module Overview

The I²C module provides a synchronous interface between the microcontroller and other I²C-compatible devices using the two-wire I²C serial bus. Devices communicate in a master/slave environment. The I²C bus specifies two signal connections:

- Serial Clock (SCL)
- Serial Data (SDA)

Both the SCL and SDA connections are bidirectional open-drain lines, each requiring pull-up resistors to the supply voltage. Pulling the line to ground is considered a logical zero and letting the line float is considered a logical one. Every transaction on the I²C bus has to be initiated by the Master.

Figure 35-25 shows a typical connection between a master and more than one slave.

FIGURE 35-2: I²C MASTER/SLAVE CONNECTIONS

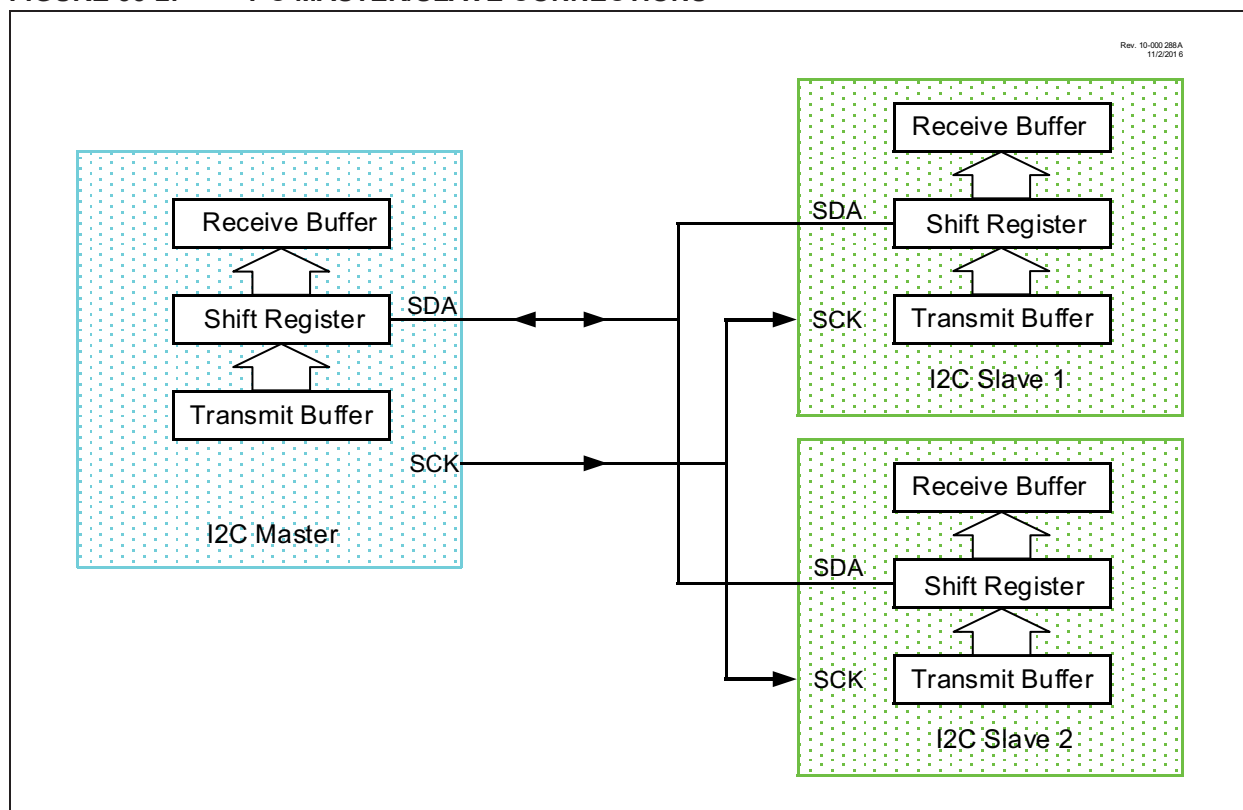


TABLE 38-1: ADC CLOCK PERIOD (T_{AD}) Vs. DEVICE OPERATING FREQUENCIES^(1,4)

ADC Clock Period (T _{AD})		Device Frequency (F _{osc})						
ADC Clock Source	CS<5:0>	64 MHz	32 MHz	20 MHz	16 MHz	8 MHz	4 MHz	1 MHz
FOSC/2	000000	31.25 ns ⁽²⁾	62.5 ns ⁽²⁾	100 ns ⁽²⁾	125 ns ⁽²⁾	250 ns ⁽²⁾	500 ns ⁽²⁾	2.0 μs
FOSC/4	000001	62.5 ns ⁽²⁾	125 ns ⁽²⁾	200 ns ⁽²⁾	250 ns ⁽²⁾	500 ns ⁽²⁾	1.0 μs	4.0 μs
FOSC/6	000010	125 ns ⁽²⁾	187.5 ns ⁽²⁾	300 ns ⁽²⁾	375 ns ⁽²⁾	750 ns ⁽²⁾	1.5 μs	6.0 μs
FOSC/8	000011	187.5 ns ⁽²⁾	250 ns ⁽²⁾	400 ns ⁽²⁾	500 ns ⁽²⁾	1.0 μs	2.0 μs	9.0 μs ⁽³⁾
...
FOSC/16	000111	250 ns ⁽²⁾	500 ns ⁽²⁾	800 ns ⁽²⁾	1.0 μs	2.0 μs	4.0 μs	16.0 μs ⁽³⁾
...
FOSC/128	111111	2.0 μs	4.0 μs	6.4 μs	8.0 μs	16.0 μs ⁽³⁾	32.0 μs ⁽²⁾	128.0 μs ⁽²⁾
FRC	CS(ADCON0<4>) = 1	1.0-6.0 μs	1.0-6.0 μs	1.0-6.0 μs	1.0-6.0 μs	1.0-6.0 μs	1.0-6.0 μs	1.0-6.0 μs

Legend: Shaded cells are outside of recommended range.

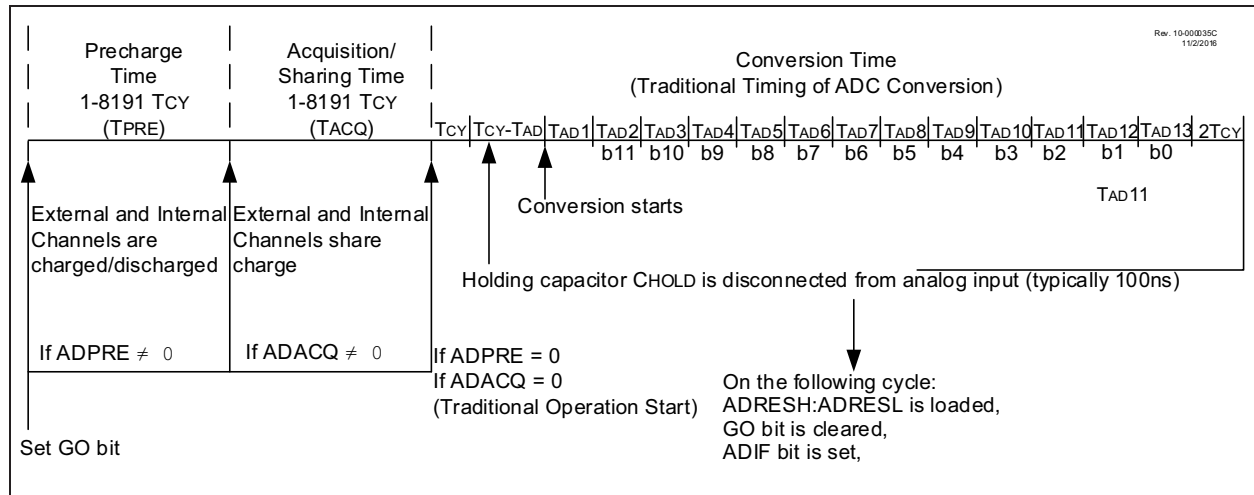
Note 1: See T_{AD} parameter for FRC source typical T_{AD} value.

2: These values violate the required T_{AD} time.

3: Outside the recommended T_{AD} time.

4: The ADC clock period (T_{AD}) and total ADC conversion time can be minimized when the ADC clock is derived from the system clock FOSC. However, the FRC oscillator source must be used when conversions are to be performed with the device in Sleep mode.

FIGURE 38-2: ANALOG-TO-DIGITAL CONVERSION T_{AD} CYCLES



38.3 ADC Acquisition Requirements

For the ADC to meet its specified accuracy, the charge holding capacitor (CHOLD) must be allowed to fully charge to the input channel voltage level. The Analog Input model is shown in [Figure 38-4](#). The source impedance (RS) and the internal sampling switch (RSS) impedance directly affect the time required to charge the capacitor CHOLD. The sampling switch (RSS) impedance varies over the device voltage (VDD), refer to [Figure 38-4](#). **The maximum recommended impedance for analog sources is 10 kΩ.** If the source

impedance is decreased, the acquisition time may be decreased. After the analog input channel is selected (or changed), an ADC acquisition must be completed before the conversion can be started. To calculate the minimum acquisition time, [Equation 38-1](#) may be used. This equation assumes that 1/2 LSB error is used (4,096 steps for the ADC). The 1/2 LSB error is the maximum error allowed for the ADC to meet its specified resolution.

EQUATION 38-1: ACQUISITION TIME EXAMPLE

Assumptions: Temperature = 50°C and external impedance of 10kΩ 5.0V VDD

$$\begin{aligned} T_{ACQ} &= \text{Amplifier Settling Time} + \text{Hold Capacitor Charging Time} + \text{Temperature Coefficient} \\ &= T_{AMP} + T_C + T_{COFF} \\ &= 2\mu s + T_C + [(Temperature - 25^\circ C)(0.05\mu s/^\circ C)] \end{aligned}$$

The value for TC can be approximated with the following equations:

$$V_{APPLIED} \left(1 - \frac{1}{(2^{n+1}) - 1} \right) = V_{CHOLD} \quad ;[1] \text{ } V_{CHOLD} \text{ charged to within } 1/2 \text{ lsb}$$

$$V_{APPLIED} \left(1 - e^{\frac{-T_C}{RC}} \right) = V_{CHOLD} \quad ;[2] \text{ } V_{CHOLD} \text{ charge response to } V_{APPLIED}$$

$$V_{APPLIED} \left(1 - e^{\frac{-T_C}{RC}} \right) = V_{APPLIED} \left(1 - \frac{1}{(2^{n+1}) - 1} \right) \quad ;\text{combining [1] and [2]}$$

Note: Where n = number of bits of the ADC.

Solving for TC:

$$\begin{aligned} T_C &= -CHOLD(RIC + RSS + RS) \ln(1/8191) \\ &= -28pF(1k\Omega + 7k\Omega + 10k\Omega) \ln(0.0001221) \\ &= 4.54\mu s \end{aligned}$$

Therefore:

$$\begin{aligned} T_{ACQ} &= 2\mu s + 4.54\mu s + [(50^\circ C - 25^\circ C)(0.05\mu s/^\circ C)] \\ &= 7.79\mu s \end{aligned}$$

Note 1: The reference voltage (VREF) has no effect on the equation, since it cancels itself out.

2: The charge holding capacitor (CHOLD) is not discharged after each conversion.

3: The maximum recommended impedance for analog sources is 10 kΩ. This is required to meet the pin leakage specification.

REGISTER 38-22: ADPREVH: ADC PREVIOUS RESULT REGISTER

R-x	R-x	R-x	R-x	R-x	R-x	R-x	R-x
PREV<15:8>							
bit 7 bit 0							

Legend:

R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 **PREV<15:8>**: Previous ADC Results bits
If ADPSIS = 1:
Upper byte of FLTR at the start of current ADC conversion
If ADPSIS = 0:
Upper bits of ADRES at the start of current ADC conversion⁽¹⁾

Note 1: If ADPSIS = 0, ADPREVH and ADPREVL are formatted the same way as ADRES is, depending on the FM bit.

REGISTER 38-23: ADPREVL: ADC PREVIOUS RESULT REGISTER

R-x	R-x	R-x	R-x	R-x	R-x	R-x	R-x
PREV<7:0>							
bit 7 bit 0							

Legend:

R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'
u = Bit is unchanged	x = Bit is unknown	-n/n = Value at POR and BOR/Value at all other Resets
'1' = Bit is set	'0' = Bit is cleared	

bit 7-0 **PREV<7:0>**: Previous ADC Results bits
If ADPSIS = 1:
Lower byte of FLTR at the start of current ADC conversion
If ADPSIS = 0:
Lower bits of ADRES at the start of current ADC conversion⁽¹⁾

Note 1: If ADPSIS = 0, ADPREVH and ADPREVL are formatted the same way as ADRES is, depending on the FM bit.

INFSNZ		Increment f, skip if not 0							
Syntax:	INFSNZ f {,d {,a}}								
Operands:	0 ≤ f ≤ 255								
	d ∈ [0,1]								
	a ∈ [0,1]								
Operation:	(f) + 1 → dest, skip if result ≠ 0								
Status Affected:	None								
Encoding:	<table border="1"><tr><td>0100</td><td>10da</td><td>ffff</td><td>ffff</td></tr></table>					0100	10da	ffff	ffff
0100	10da	ffff	ffff						
Description:	<p>The contents of register 'f' are incremented. If 'd' is '0', the result is placed in W. If 'd' is '1', the result is placed back in register 'f' (default). If the result is not '0', the next instruction, which is already fetched, is discarded and a <code>NOP</code> is executed instead, making it a 2-cycle instruction.</p> <p>If 'a' is '0', the Access Bank is selected. If 'a' is '1', the BSR is used to select the GPR bank.</p> <p>If 'a' is '0' and the extended instruction set is enabled, this instruction operates in Indexed Literal Offset Addressing mode whenever f ≤ 95 (5Fh). See Section 43.2.3 “Byte-Oriented and Bit-Oriented Instructions in Indexed Literal Offset Mode” for details.</p>								
Words:	1								
Cycles:	1(2)								
	Note: 3 cycles if skip and followed by a 2-word instruction.								

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write to destination

If skip:

Q1	Q2	Q3	Q4
No operation	No operation	No operation	No operation

If skip and followed by 2-word instruction:

Q1	Q2	Q3	Q4
No operation	No operation	No operation	No operation
No operation	No operation	No operation	No operation

Example:

```

HERE    INFSNZ  REG, 1, 0
ZERO
NZERO

```

Before Instruction

PC = Address (HERE)

After Instruction

REG = REG + 1

If REG \neq 0;

PC = Address (NZERO)

If REG = 0;

PC = Address (ZERO)

IORLW	Inclusive OR literal with W				
Syntax:	IORLW k				
Operands:	$0 \leq k \leq 255$				
Operation:	(W) .OR. k \rightarrow W				
Status Affected:	N, Z				
Encoding:	<table><tr><td>0000</td><td>1001</td><td>kkkk</td><td>kkkk</td></tr></table>	0000	1001	kkkk	kkkk
0000	1001	kkkk	kkkk		
Description:	The contents of W are ORed with the 8-bit literal 'k'. The result is placed in W.				
Words:	1				
Cycles:	1				

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read literal 'k'	Process Data	Write to W

Example: IORLW 35h

Before Instruction

W = 9Ah

After Instruction

W = BFh

ADDWF ADD W to Indexed (Indexed Literal Offset mode)

Syntax: ADDWF [k] {,d}

Operands: $0 \leq k \leq 95$
 $d \in [0,1]$

Operation: $(W) + ((FSR2) + k) \rightarrow \text{dest}$

Status Affected: N, OV, C, DC, Z

Encoding:

0010	01d0	kkkk	kkkk
------	------	------	------

Description: The contents of W are added to the contents of the register indicated by FSR2, offset by the value 'k'. If 'd' is '0', the result is stored in W. If 'd' is '1', the result is stored back in register 'f' (default).

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read 'k'	Process Data	Write to destination

Example: ADDWF [OFST], 0

Before Instruction

W = 17h
 OFST = 2Ch
 FSR2 = 0A00h
 Contents of 0A2Ch = 20h

After Instruction

W = 37h
 Contents of 0A2Ch = 20h

BSF Bit Set Indexed (Indexed Literal Offset mode)

Syntax: BSF [k], b

Operands: $0 \leq f \leq 95$
 $0 \leq b \leq 7$

Operation: $1 \rightarrow ((FSR2) + k) \langle b \rangle$

Status Affected: None

Encoding:

1000	bbb0	kkkk	kkkk
------	------	------	------

Description: Bit 'b' of the register indicated by FSR2, offset by the value 'k', is set.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read register 'f'	Process Data	Write to destination

Example: BSF [FLAG_OFST], 7

Before Instruction

FLAG_OFST = 0Ah
 FSR2 = 0A00h
 Contents of 0A0Ah = 55h

After Instruction

Contents of 0A0Ah = D5h

SETF Set Indexed (Indexed Literal Offset mode)

Syntax: SETF [k]

Operands: $0 \leq k \leq 95$

Operation: $FFh \rightarrow ((FSR2) + k)$

Status Affected: None

Encoding:

0110	1000	kkkk	kkkk
------	------	------	------

Description: The contents of the register indicated by FSR2, offset by 'k', are set to FFh.

Words: 1

Cycles: 1

Q Cycle Activity:

Q1	Q2	Q3	Q4
Decode	Read 'k'	Process Data	Write register

Example: SETF [OFST]

Before Instruction

OFST = 2Ch
 FSR2 = 0A00h
 Contents of 0A2Ch = 00h

After Instruction

Contents of 0A2Ch = FFh

46.4 AC Characteristics

FIGURE 46-4: LOAD CONDITIONS

