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"[Embedded - Microcontrollers](#)" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "[Embedded - Microcontrollers](#)"

Details

Product Status	Obsolete
Core Processor	C166SV2
Core Size	16/32-Bit
Speed	128MHz
Connectivity	CANbus, EBI/EMI, FlexRay, I ² C, LINbus, SPI, UART/USART
Peripherals	I ² S, POR, PWM, WDT
Number of I/O	118
Program Memory Size	1.06MB (1.06M x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	90K x 8
Voltage - Supply (Vcc/Vdd)	3V ~ 5.5V
Data Converters	A/D 24x12b
Oscillator Type	Internal
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	-
Package / Case	-
Supplier Device Package	-
Purchase URL	https://www.e-xfl.com/product-detail/infineon-technologies/xc2788x136f128laakxuma1

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Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
6	$\overline{\text{TRST}}$	I	In/B	Test-System Reset Input For normal system operation, pin $\overline{\text{TRST}}$ should be held low. A high level at this pin at the rising edge of $\overline{\text{PORST}}$ activates the XC2788X's debug system. In this case, pin $\overline{\text{TRST}}$ must be driven low once to reset the debug system. An internal pull-down device will hold this pin low when nothing is driving it.
7	P8.3	O0 / I	St/B	Bit 3 of Port 8, General Purpose Input/Output
	CCU60_COUT60	O1	St/B	CCU60 Channel 0 Output
	CCU62_CC60	O2	St/B	CCU62 Channel 0 Output
	TDI_D	IH	St/B	JTAG Test Data Input If JTAG pos. D is selected during start-up, an internal pull-up device will hold this pin low when nothing is driving it.
	CCU62_CC60INB	I	St/B	CCU62 Channel 0 Input
8	P7.0	O0 / I	St/B	Bit 0 of Port 7, General Purpose Input/Output
	T3OUT	O1	St/B	GPT12E Timer T3 Toggle Latch Output
	T6OUT	O2	St/B	GPT12E Timer T6 Toggle Latch Output
	TDO_A	OH / IH	St/B	JTAG Test Data Output / DAP1 Input/Output If DAP pos. 0 or 2 is selected during start-up, an internal pull-down device will hold this pin low when nothing is driving it.
	ESR2_1	I	St/B	ESR2 Trigger Input 1

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
63	P4.1	O0 / I	St/B	Bit 1 of Port 4, General Purpose Input/Output
	CC2_CC25	O3 / I	St/B	CAPCOM2 CC25IO Capture Inp./ Compare Out.
	CS1	OH	St/B	External Bus Interface Chip Select 1 Output
	CCU62_CCP OS0B	I	St/B	CCU62 Position Input 0
	T4EUDB	I	St/B	GPT12E Timer T4 External Up/Down Control Input
	ESR1_8	I	St/B	ESR1 Trigger Input 8
64	P2.4	O0 / I	St/B	Bit 4 of Port 2, General Purpose Input/Output
	U0C1_DOUT	O1	St/B	USIC0 Channel 1 Shift Data Output
	TxDC0	O2	St/B	CAN Node 0 Transmit Data Output
	CC2_CC17	O3 / I	St/B	CAPCOM2 CC17IO Capture Inp./ Compare Out.
	A17	OH	St/B	External Bus Interface Address Line 17
	ESR1_0	I	St/B	ESR1 Trigger Input 0
	U0C0_DX0F	I	St/B	USIC0 Channel 0 Shift Data Input
	RxDC1A	I	St/B	CAN Node 1 Receive Data Input
65	P11.1	O0 / I	St/B	Bit 1 of Port 11, General Purpose Input/Output
	CCU61_COU T61	O1	St/B	CCU61 Channel 1 Output
	TxDC0	O2	St/B	CAN Node 0 Transmit Data Output
	CCU63_CCP OS1A	I	St/B	CCU63 Position Input 1
	CCU61_CTR APD	I	St/B	CCU61 Emergency Trap Input
66	P11.0	O0 / I	St/B	Bit 0 of Port 11, General Purpose Input/Output
	CCU61_COU T60	O1	St/B	CCU61 Channel 0 Output
	CCU63_CCP OS0A	I	St/B	CCU63 Position Input 0
	RxDC0F	I	St/B	CAN Node 0 Receive Data Input
	ESR1_7	I	St/B	ESR1 Trigger Input 7

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
87	P0.3	O0 / I	St/B	Bit 3 of Port 0, General Purpose Input/Output
	U1C0_SELO0	O1	St/B	USIC1 Channel 0 Select/Control 0 Output
	U1C1_SELO1	O2	St/B	USIC1 Channel 1 Select/Control 1 Output
	CCU61_COUT60	O3	St/B	CCU61 Channel 0 Output
	A3	OH	St/B	External Bus Interface Address Line 3
	U1C0_DX2A	I	St/B	USIC1 Channel 0 Shift Control Input
	RxDC0B	I	St/B	CAN Node 0 Receive Data Input
88	P3.1	O0 / I	St/B	Bit 1 of Port 3, General Purpose Input/Output
	U2C0_DOUT	O1	St/B	USIC2 Channel 0 Shift Data Output
	U2C0_DX0B	I	St/B	USIC2 Channel 0 Shift Data Input
89	P10.2	O0 / I	St/B	Bit 2 of Port 10, General Purpose Input/Output
	U0C0_SCLKOUT	O1	St/B	USIC0 Channel 0 Shift Clock Output
	CCU60_CC62	O2	St/B	CCU60 Channel 2 Output
	AD2	OH / IH	St/B	External Bus Interface Address/Data Line 2
	CCU60_CC62INA	I	St/B	CCU60 Channel 2 Input
	U0C0_DX1B	I	St/B	USIC0 Channel 0 Shift Clock Input

Table 6 Pin Definitions and Functions (cont'd)

Pin	Symbol	Ctrl.	Type	Function
105	P10.7	O0 / I	St/B	Bit 7 of Port 10, General Purpose Input/Output
	U0C1_DOUT	O1	St/B	USIC0 Channel 1 Shift Data Output
	CCU60_COU T63	O2	St/B	CCU60 Channel 3 Output
	AD7	OH / IH	St/B	External Bus Interface Address/Data Line 7
	U0C1_DX0B	I	St/B	USIC0 Channel 1 Shift Data Input
	CCU60_CCP OS0A	I	St/B	CCU60 Position Input 0
	T4INB	I	St/B	GPT12E Timer T4 Count/Gate Input
106	P0.7	O0 / I	St/B	Bit 7 of Port 0, General Purpose Input/Output
	U1C1_DOUT	O1	St/B	USIC1 Channel 1 Shift Data Output
	U1C0_SELO 3	O2	St/B	USIC1 Channel 0 Select/Control 3 Output
	A7	OH	St/B	External Bus Interface Address Line 7
	U1C1_DX0B	I	St/B	USIC1 Channel 1 Shift Data Input
	CCU61_CTR APB	I	St/B	CCU61 Emergency Trap Input
107	P3.7	O0 / I	St/B	Bit 7 of Port 3, General Purpose Input/Output
	U2C1_DOUT	O1	St/B	USIC2 Channel 1 Shift Data Output
	U2C0_SELO 3	O2	St/B	USIC2 Channel 0 Select/Control 3 Output
	U0C0_SELO 7	O3	St/B	USIC0 Channel 0 Select/Control 7 Output
	U2C1_DX0B	I	St/B	USIC2 Channel 1 Shift Data Input

Memory Content Protection

The contents of on-chip memories can be protected against soft errors (induced e.g. by radiation) by activating the parity mechanism or the Error Correction Code (ECC).

The parity mechanism can detect a single-bit error and prevent the software from using incorrect data or executing incorrect instructions.

The ECC mechanism can detect and automatically correct single-bit errors. This supports the stable operation of the system.

It is strongly recommended to activate the ECC mechanism wherever possible because this dramatically increases the robustness of an application against such soft errors.

Functional Description

to a dedicated vector table location). The occurrence of a hardware trap is also indicated by a single bit in the trap flag register (TFR). Unless another higher-priority trap service is in progress, a hardware trap will interrupt any ongoing program execution. In turn, hardware trap services can normally not be interrupted by standard or PEC interrupts.

Depending on the package option up to 3 External Service Request (ESR) pins are provided. The ESR unit processes their input values and allows to implement user controlled trap functions (System Requests SR0 and SR1). In this way reset, wakeup and power control can be efficiently realized.

Software interrupts are supported by the 'TRAP' instruction in combination with an individual trap (interrupt) number. Alternatively to emulate an interrupt by software a program can trigger interrupt requests by writing the Interrupt Request (IR) bit of an interrupt control register.

3.7 On-Chip Debug Support (OCDS)

The On-Chip Debug Support system built into the XC2788X provides a broad range of debug and emulation features. User software running on the XC2788X can be debugged within the target system environment.

The OCDS is controlled by an external debugging device via the debug interface. This either consists of the 2-pin Device Access Port (DAP) or of the JTAG port conforming to IEEE-1149. The debug interface can be completed with an optional break interface.

The debugger controls the OCDS with a set of dedicated registers accessible via the debug interface (DAP or JTAG). In addition the OCDS system can be controlled by the CPU, e.g. by a monitor program. An injection interface allows the execution of OCDS-generated instructions by the CPU.

Multiple breakpoints can be triggered by on-chip hardware, by software, or by an external trigger input. Single stepping is supported, as is the injection of arbitrary instructions and read/write access to the complete internal address space. A breakpoint trigger can be answered with a CPU halt, a monitor call, a data transfer, or/and the activation of an external signal.

Tracing of data can be obtained via the debug interface, or via the external bus interface for increased performance.

Tracing of program execution is supported by the XC2000 Family emulation device. With this device the DAP can operate on clock rates of up to 20 MHz.

The DAP interface uses two interface signals, the JTAG interface uses four interface signals, to communicate with external circuitry. The debug interface can be amended with two optional break lines.

MultiCAN Features

- CAN functionality conforming to CAN specification V2.0 B active for each CAN node (compliant to ISO 11898)
- Independent CAN nodes
- Set of independent message objects (shared by the CAN nodes)
- Dedicated control registers for each CAN node
- Data transfer rate up to 1 Mbit/s, individually programmable for each node
- Flexible and powerful message transfer control and error handling capabilities
- Full-CAN functionality for message objects:
 - Can be assigned to one of the CAN nodes
 - Configurable as transmit or receive objects, or as message buffer FIFO
 - Handle 11-bit or 29-bit identifiers with programmable acceptance mask for filtering
 - Remote Monitoring Mode, and frame counter for monitoring
- Automatic Gateway Mode support
- 16 individually programmable interrupt nodes
- Analyzer mode for CAN bus monitoring

3.15 System Timer

The System Timer consists of a programmable prescaler and two concatenated timers (10 bits and 6 bits). Both timers can generate interrupt requests. The clock source can be selected and the timers can also run during power reduction modes.

Therefore, the System Timer enables the software to maintain the current time for scheduling functions or for the implementation of a clock.

3.16 Watchdog Timer

The Watchdog Timer is one of the fail-safe mechanisms which have been implemented to prevent the controller from malfunctioning for longer periods of time.

The Watchdog Timer is always enabled after an application reset of the chip. It can be disabled and enabled at any time by executing the instructions DISWDT and ENWDT respectively. The software has to service the Watchdog Timer before it overflows. If this is not the case because of a hardware or software failure, the Watchdog Timer overflows, generating a prewarning interrupt and then a reset request.

The Watchdog Timer is a 16-bit timer clocked with the system clock divided by 16,384 or 256. The Watchdog Timer register is set to a prespecified reload value (stored in WDTRRL) in order to allow further variation of the monitored time interval. Each time it is serviced by the application software, the Watchdog Timer is reloaded and the prescaler is cleared.

Time intervals between 3.2 μ s and 13.4 s can be monitored (@ 80 MHz).

Time intervals between 2.0 μ s and 8.39 s can be monitored (@ 128 MHz).

4.2.1 DC Parameters for Upper Voltage Area

Keeping signal levels within the limits specified in this table ensures operation without overload conditions. For signal levels outside these specifications, also refer to the specification of the overload current I_{OV} .

Note: Operating Conditions apply.

Table 16 is valid under the following conditions:

$V_{DDP} \geq 4.5 \text{ V}$; $V_{DDP\text{typ}} = 5 \text{ V}$; $V_{DDP} \leq 5.5 \text{ V}$

Table 16 DC Characteristics for Upper Voltage Range

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Pin capacitance (digital inputs/outputs). To be doubled for double bond pins. ¹⁾	$C_{IO} \text{ CC}$	–	–	10	pF	not subject to production test
Input Hysteresis ²⁾	HYS CC	$0.11 \times V_{DDP}$	–	–	V	$R_S = 0 \Omega$
Absolute input leakage current on pins of analog ports ³⁾	$ I_{OZ1} \text{ CC}$	–	10	200	nA	$V_{IN} > 0 \text{ V}$; $V_{IN} < V_{DDP}$
Absolute input leakage current for all other pins. To be doubled for double bond pins. ³⁾¹⁾⁴⁾	$ I_{OZ2} \text{ CC}$	–	0.2	5	μA	$T_J \leq 110 \text{ }^\circ\text{C}$; $V_{IN} < V_{DDP}$; $V_{IN} > V_{SS}$
		–	0.2	15	μA	$T_J \leq 150 \text{ }^\circ\text{C}$; $V_{IN} < V_{DDP}$; $V_{IN} > V_{SS}$
Pull Level Force Current ⁵⁾	$ I_{PLF} \text{ SR}$	250	–	–	μA	⁶⁾
Pull Level Keep Current ⁷⁾	$ I_{PLK} \text{ SR}$	–	–	30	μA	⁶⁾
Input high voltage (all except XTAL1)	$V_{IH} \text{ SR}$	$0.7 \times V_{DDP}$	–	$V_{DDP} + 0.3$	V	
Input low voltage (all except XTAL1)	$V_{IL} \text{ SR}$	-0.3	–	$0.3 \times V_{DDP}$	V	
Output High voltage ⁸⁾	$V_{OH} \text{ CC}$	$V_{DDP} - 1.0$	–	–	V	$I_{OH} \geq I_{OH\text{max}}$
		$V_{DDP} - 0.4$	–	–	V	$I_{OH} \geq I_{OH\text{nom}}$ ⁹⁾

4.2.2 DC Parameters for Lower Voltage Area

Keeping signal levels within the limits specified in this table ensures operation without overload conditions. For signal levels outside these specifications, also refer to the specification of the overload current I_{OV} .

Note: Operating Conditions apply.

Table 17 is valid under the following conditions:

$V_{DDP} \geq 3.0 \text{ V}$; $V_{DDPtyp} = 3.3 \text{ V}$; $V_{DDP} \leq 4.5 \text{ V}$

Table 17 DC Characteristics for Lower Voltage Range

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Pin capacitance (digital inputs/outputs). To be doubled for double bond pins. ¹⁾	$C_{IO} \text{ CC}$	–	–	10	pF	not subject to production test
Input Hysteresis ²⁾	HYS CC	$0.07 \times V_{DDP}$	–	–	V	$R_S = 0 \Omega$
Absolute input leakage current on pins of analog ports ³⁾	$ I_{OZ1} \text{ CC}$	–	10	200	nA	$V_{IN} > V_{SS}$; $V_{IN} < V_{DDP}$
Absolute input leakage current for all other pins. To be doubled for double bond pins. ³⁾¹⁾⁴⁾	$ I_{OZ2} \text{ CC}$	–	0.2	2.5	μA	$T_J \leq 110^\circ\text{C}$; $V_{IN} < V_{DDP}$; $V_{IN} > V_{SS}$
		–	0.2	8	μA	$T_J \leq 150^\circ\text{C}$; $V_{IN} < V_{DDP}$; $V_{IN} > V_{SS}$
Pull Level Force Current ⁵⁾	$ I_{PLF} \text{ SR}$	150	–	–		⁶⁾
Pull Level Keep Current ⁷⁾	$ I_{PLK} \text{ SR}$	–	–	10	μA	⁶⁾
Input high voltage (all except XTAL1)	$V_{IH} \text{ SR}$	$0.7 \times V_{DDP}$	–	$V_{DDP} + 0.3$	V	
Input low voltage (all except XTAL1)	$V_{IL} \text{ SR}$	-0.3	–	$0.3 \times V_{DDP}$	V	
Output High voltage ⁸⁾	$V_{OH} \text{ CC}$	$V_{DDP} - 1.0$	–	–	V	$I_{OH} \geq I_{OHmax}$
		$V_{DDP} - 0.4$	–	–	V	$I_{OH} \geq I_{OHnom}$ ⁹⁾

Electrical Parameters

Table 17 DC Characteristics for Lower Voltage Range (cont'd)

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Output Low Voltage ⁸⁾	V_{OL} CC	—	—	1.0	V	$I_{OL} \leq I_{OLmax}$
		—	—	0.4	V	$I_{OL} \leq I_{OLnom}$ ¹⁰⁾

- 1) Because each double bond pin is connected to two pads (standard pad and high-speed pad), it has twice the normal value. For a list of affected pins refer to the pin definitions table in chapter 2.
- 2) Not subject to production test - verified by design/characterization. Hysteresis is implemented to avoid metastable states and switching due to internal ground bounce. It cannot suppress switching due to external system noise under all conditions.
- 3) If the input voltage exceeds the respective supply voltage due to ground bouncing ($V_{IN} < V_{SS}$) or supply ripple ($V_{IN} > V_{DDP}$), a certain amount of current may flow through the protection diodes. This current adds to the leakage current. An additional error current (I_{INJ}) will flow if an overload current flows through an adjacent pin. Please refer to the definition of the overload coupling factor K_{OV} .
- 4) The given values are worst-case values. In production test, this leakage current is only tested at 125 °C; other values are ensured by correlation. For derating, please refer to the following descriptions: Leakage derating depending on temperature (T_J = junction temperature [°C]): $I_{OZ} = 0.05 \times e^{(1.5 + 0.028 \times T_J)}$ [μA]. For example, at a temperature of 95 °C the resulting leakage current is 3.2 μA. Leakage derating depending on voltage level ($DV = V_{DDP} - V_{PIN}$ [V]): $I_{OZ} = I_{OZtempmax} - (1.6 \times DV)$ (μA). This voltage derating formula is an approximation which applies for maximum temperature.
- 5) Drive the indicated minimum current through this pin to change the default pin level driven by the enabled pull device: $V_{PIN} \leq V_{IL}$ for a pullup; $V_{PIN} \geq V_{IH}$ for a pulldown.
- 6) These values apply to the fixed pull-devices in dedicated pins and to the user-selectable pull-devices in general purpose IO pins.
- 7) Limit the current through this pin to the indicated value so that the enabled pull device can keep the default pin level: $V_{PIN} \geq V_{IH}$ for a pullup; $V_{PIN} \leq V_{IL}$ for a pulldown.
- 8) The maximum deliverable output current of a port driver depends on the selected output driver mode. This specification is not valid for outputs which are switched to open drain mode. In this case the respective output will float and the voltage is determined by the external circuit.
- 9) As a rule, with decreasing output current the output levels approach the respective supply level (VOL->VSS, VOH->VDDP). However, only the levels for nominal output currents are verified.
- 10) As a rule, with decreasing output current the output levels approach the respective supply level (V_{OL} -> V_{SS} , V_{OH} -> V_{DDP}). However, only the levels for nominal output currents are verified.

Electrical Parameters

- 2) This parameter includes the sample time (also the additional sample time specified by STC), the time to determine the digital result and the time to load the result register with the conversion result. Values for the basic clock t_{ADCl} depend on programming.
- 3) The broken wire detection delay against V_{AGND} is measured in numbers of consecutive precharge cycles at a conversion rate of not more than 500 μs . Result below 10% (66_H).
- 4) The broken wire detection delay against V_{AREF} is measured in numbers of consecutive precharge cycles at a conversion rate of not more than 10 μs . This function is influenced by leakage current, in particular at high temperature. Result above 80% (332_H).
- 5) V_{AIN} may exceed V_{AGND} or V_{AREF} up to the absolute maximum ratings. However, the conversion result in these cases will be X000_H or X3FF_H, respectively.

Table 21 ADC Parameters for Upper Voltage Range

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Input resistance of the selected analog channel	R_{AIN} CC	—	0.9	1.5	k Ω	not subject to production test ¹⁾
Input resistance of the reference input	R_{AREF} CC	—	0.5	1	k Ω	not subject to production test ¹⁾
Differential Non-Linearity Error ²⁾³⁾⁴⁾⁵⁾	$ EA_{\text{DNL}} $ CC	—	1.5	3.0	LSB	not subject to production test
Gain Error ²⁾³⁾⁴⁾⁵⁾	$ EA_{\text{GAIN}} $ CC	—	0.5	3.5	LSB	not subject to production test
Integral Non-Linearity ²⁾³⁾⁴⁾⁵⁾	$ EA_{\text{INL}} $ CC	—	1.5	3.0	LSB	not subject to production test
Offset Error ²⁾³⁾⁴⁾⁵⁾	$ EA_{\text{OFF}} $ CC	—	1.0	4.0	LSB	not subject to production test
Total Unadjusted Error ³⁾⁴⁾	$ TUE $ CC	—	2.5	4	LSB	⁶⁾
Analog clock frequency	f_{ADCl} SR	2	—	20	MHz	Std. reference input (V_{AREF})
		2	—	17.5	MHz	Alt. reference input (CH0)
Wakeup time from analog powerdown, fast mode	t_{WAF} CC	—	—	7	μs	
Wakeup time from analog powerdown, slow mode	t_{WAS} CC	—	—	11.5	μs	

Electrical Parameters

- 1) The short-term frequency deviation refers to a timeframe of a few hours and is measured relative to the current frequency at the beginning of the respective timeframe. This parameter is useful to determine a time span for re-triggering a LIN synchronization.
- 2) This parameter is tested for the fastest and the slowest selection. The medium selections are not subject to production test - verified by design/characterization
- 3) f_{WU} in MHz
- 4) This value includes a hysteresis of approximately 50 mV for rising voltage.
- 5) V_{LV} = selected PVC/SWD voltage level
- 6) The limit $V_{LV} - 0.10$ V is valid for the OK1 level. The limit for the OK2 level is $V_{LV} - 0.15$ V.

Conditions for t_{SPO} Timing Measurement

The time required for the transition from **Power-On** to **Base** mode is called t_{SPO} . It is measured under the following conditions:

Precondition: The pad supply is valid, i.e. V_{DDPB} is above 3.0 V and remains above 3.0 V even though the XC2788X is starting up. See also V_{DDPB} requirements in [Table 13](#).

Start condition: Power-on reset is removed ($\overline{PORST} = 1$).

End condition: External pin toggle caused by first user instruction executed from FLASH after startup.

Conditions for t_{SSB} Timing Measurement

The time required for the transition from **Standby** to **Base** mode is called t_{SSB} . It is measured under the following conditions:

Precondition: The **Standby** mode has been entered using the procedure defined in the Programmer's Guide.

Start condition: Pin toggle on \overline{ESR} pin triggering the startup sequence.

End condition: External pin toggle caused by first user instruction executed from FLASH after startup.

Conditions for t_{SSO} Timing Measurement

The time required for the transition from **Stopover** to **Stopover Waked-Up** mode is called t_{SSO} . It is measured under the following conditions:

Precondition: The **Stopover** mode has been entered using the procedure defined in the Programmer's Guide.

Start condition: Pin toggle on \overline{ESR} pin triggering the startup sequence.

End condition: External pin toggle caused by first user instruction executed from PSRAM after startup.

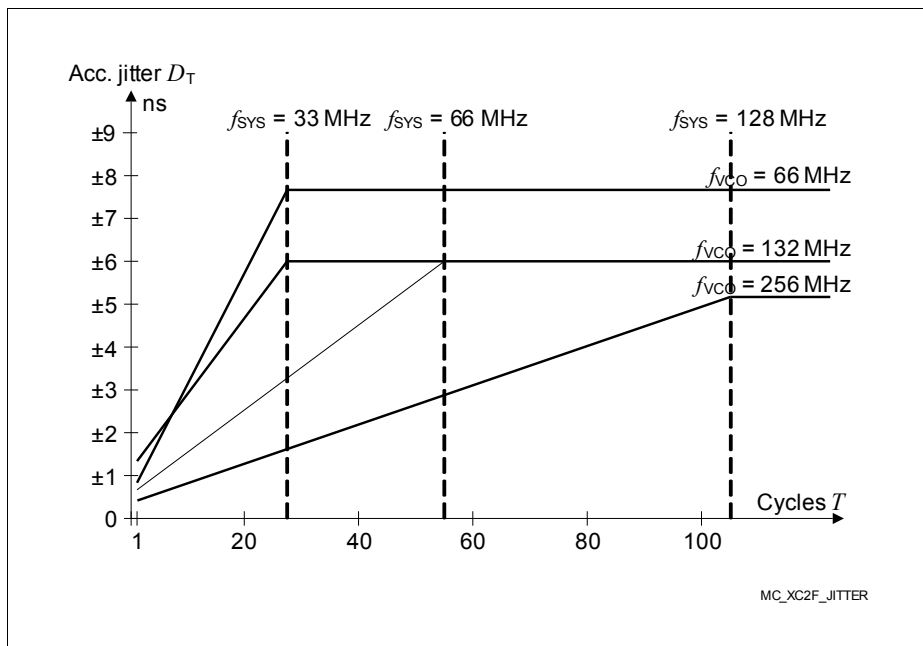


Figure 17 Approximated Accumulated PLL Jitter

Note: The specified PLL jitter values are valid if the capacitive load per pin does not exceed $C_L = 20 \text{ pF}$.

The maximum peak-to-peak noise on the pad supply voltage (measured between V_{DDPB} pin 144 and V_{SS} pin 1) is limited to a peak-to-peak voltage of $V_{PP} = 50 \text{ mV}$. This can be achieved by appropriate blocking of the supply voltage as close as possible to the supply pins and using PCB supply and ground planes.

4.6.2.2 Wakeup Clock

When wakeup operation is selected ($\text{SYSCON0.CLKSEL} = 00_{\text{B}}$), the system clock is derived from the low-frequency wakeup clock source:

$$f_{\text{SYS}} = f_{\text{WU}}$$

In this mode, a basic functionality can be maintained without requiring an external clock source and while minimizing the power consumption.

4.6.2.3 Selecting and Changing the Operating Frequency

When selecting a clock source and the clock generation method, the required parameters must be carefully written to the respective bitfields, to avoid unintended intermediate states.

Many applications change the frequency of the system clock (f_{SYS}) during operation in order to optimize system performance and power consumption. Changing the operating frequency also changes the switching currents, which influences the power supply.

To ensure proper operation of the on-chip EVRs while they generate the core voltage, the operating frequency shall only be changed in certain steps. This prevents overshoots and undershoots of the supply voltage.

To avoid the indicated problems, recommended sequences are provided which ensure the intended operation of the clock system interacting with the power system.

Please refer to the Programmer's Guide.

4.6.5.1 Bus Cycle Control with the READY Input

The duration of an external bus cycle can be controlled by the external circuit using the READY input signal. The polarity of this input signal can be selected.

Synchronous READY permits the shortest possible bus cycle but requires the input signal to be synchronous to the reference signal CLKOUT.

An asynchronous READY signal puts no timing constraints on the input signal but incurs a minimum of one waitstate due to the additional synchronization stage. The minimum duration of an asynchronous READY signal for safe synchronization is one CLKOUT period plus the input setup time.

An active READY signal can be deactivated in response to the trailing (rising) edge of the corresponding command (\overline{RD} or \overline{WR}).

If the next bus cycle is controlled by READY, an active READY signal must be disabled before the first valid sample point in the next bus cycle. This sample point depends on the programmed phases of the next cycle.

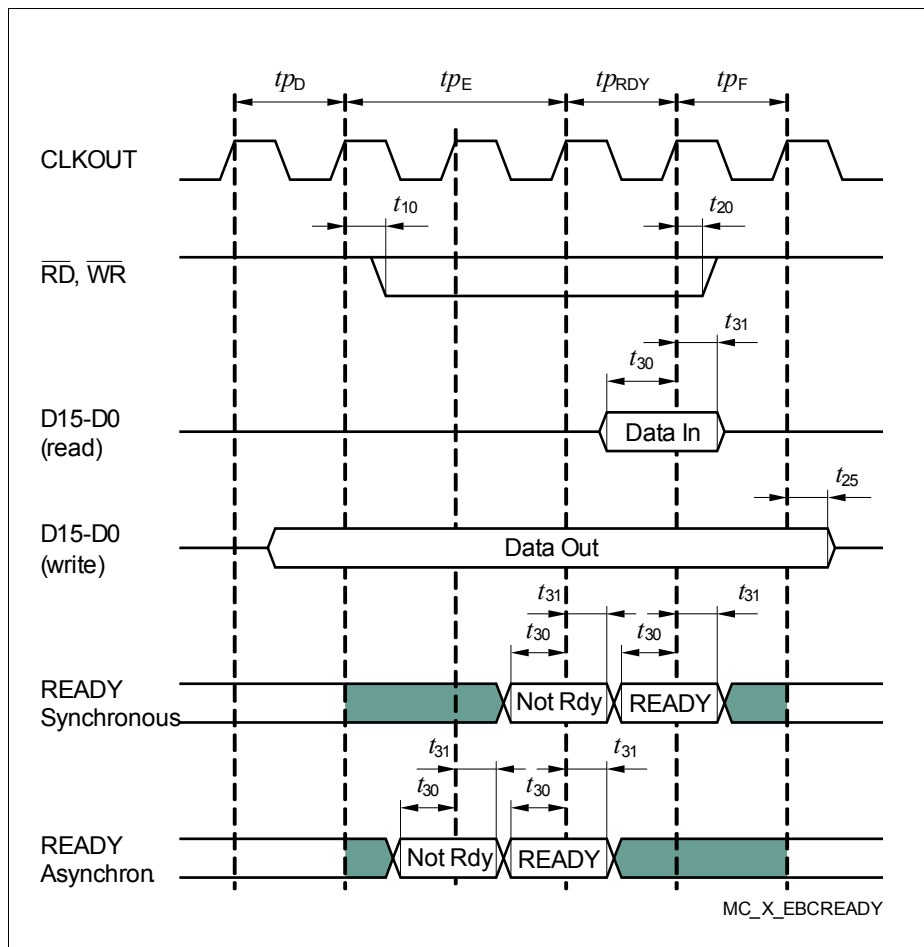


Figure 22 **READY Timing**

Note: If the READY input is sampled inactive at the indicated sampling point ("Not Rdy") a READY-controlled waitstate is inserted (tp_{RDY}), sampling the READY input active at the indicated sampling point ("Ready") terminates the currently running bus cycle.

Note the different sampling points for synchronous and asynchronous READY. This example uses one mandatory waitstate (see tp_E) before the READY input value is used.

4.6.6 Synchronous Serial Interface Timing

The following parameters are applicable for a USIC channel operated in SSC mode.

Note: These parameters are not subject to production test but verified by design and/or characterization.

Note: Operating Conditions apply; $C_L = 20$ pF.

Table 36 USIC SSC Master Mode Timing for Upper Voltage Range

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Slave select output SELO active to first SCLKOUT transmit edge	t_1 CC	$t_{SYS} - 8^{1)}$	—	—	ns	
Slave select output SELO inactive after last SCLKOUT receive edge	t_2 CC	$t_{SYS} - 6^{1)}$	—	—	ns	
Data output DOUT valid time	t_3 CC	-6	—	9	ns	
Receive data input setup time to SCLKOUT receive edge	t_4 SR	31	—	—	ns	
Data input DX0 hold time from SCLKOUT receive edge	t_5 SR	-4	—	—	ns	

1) $t_{SYS} = 1 / f_{SYS}$

Table 37 USIC SSC Master Mode Timing for Lower Voltage Range

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Slave select output SELO active to first SCLKOUT transmit edge	t_1 CC	$t_{SYS} - 10^{1)}$	—	—	ns	
Slave select output SELO inactive after last SCLKOUT receive edge	t_2 CC	$t_{SYS} - 9^{1)}$	—	—	ns	
Data output DOUT valid time	t_3 CC	-7	—	11	ns	