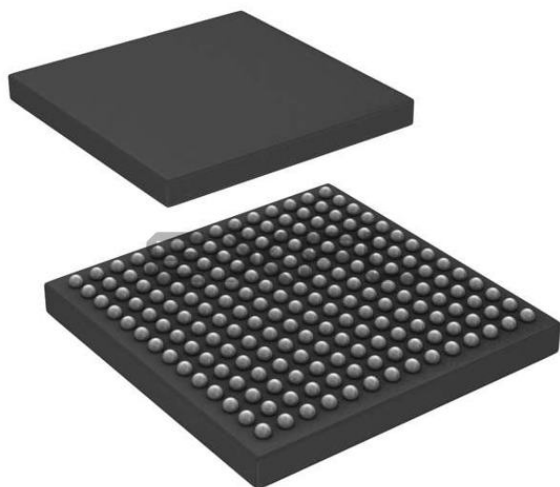




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Details

Product Status	Active
Core Processor	ARM® Cortex®-M4
Core Size	32-Bit Single-Core
Speed	144MHz
Connectivity	CANbus, EBI/EMI, Ethernet, I ² C, LINbus, MMC/SD, SPI, UART/USART, USB OTG, USIC
Peripherals	DMA, I ² S, LED, POR, Touch-Sense, WDT
Number of I/O	155
Program Memory Size	1.5MB (1.5M x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	276K x 8
Voltage - Supply (Vcc/Vdd)	3.13V ~ 3.63V
Data Converters	A/D 32x12b; D/A 2x12b
Oscillator Type	External
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	196-LFBGA
Supplier Device Package	PG-LFBGA-196-2
Purchase URL	https://www.e-xfl.com/product-detail/infineon-technologies/xmc4700e196k1536aaxqma1

XMC4[78]00 Data Sheet

Revision History: V1.0 2016-01

Previous Versions:

V0.7 2015-10 (preliminary)

Page	Subjects
8	Corrected EtherCAT features to 8 Fieldbus Memory Management Units (FMMU) and 8 Sync Manager.
46	Added footnote explaining minimum V_{BAT} requirements to start the hibernate domain and/or oscillation of a crystal on RTC_XTAL.
53	Added HIBIO characteristics.
58	Corrected DAC INL and gain error.
70	Changed frequency dependency of the current consumption.
73	Added peripheral idle current overview.
127ff	Updated package parameters and drawings.
132	Higher HBM and CDM ESD limits.

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About this Document

This Data Sheet is addressed to embedded hardware and software developers. It provides the reader with detailed descriptions about the ordering designations, available features, electrical and physical characteristics of the XMC4[78]00 series devices.

The document describes the characteristics of a superset of the XMC4[78]00 series devices. For simplicity, the various device types are referred to by the collective term XMC4[78]00 throughout this manual.

XMC4000 Family User Documentation

The set of user documentation includes:

- **Reference Manual**
 - describes the functionality of the superset of devices.
- **Data Sheets**
 - list the complete ordering designations, available features and electrical characteristics of derivative devices.
- **Errata Sheets**
 - list deviations from the specifications given in the related Reference Manual or Data Sheets. Errata Sheets are provided for the superset of devices.

Attention: Please consult all parts of the documentation set to attain consolidated knowledge about your device.

Application related guidance is provided by **Users Guides** and **Application Notes**.

Please refer to <http://www.infineon.com/xmc4000> to get access to the latest versions of those documents.

Table 5 SRAM Memory Ranges

Total SRAM Size	Program SRAM	System Data SRAM	Communication Data SRAM
200 Kbytes	1FFE E000 _H – 1FFF FFFF _H	2000 0000 _H – 2001 FFFF _H	–
276 Kbytes	1FFE 8000 _H – 1FFF FFFF _H	2000 0000 _H – 2001 FFFF _H	2002 0000 _H – 2002 CFFF _H
352 Kbytes	1FFE 8000 _H – 1FFF FFFF _H	2000 0000 _H – 2001 FFFF _H	2002 0000 _H – 2003 FFFF _H

Table 6 ADC Channels¹⁾

Package	VADC G0	VADC G1	VADC G2	VADC G3
PG-LQFP-144 PG-LFBGA-196	CH0..CH7	CH0..CH7	CH0..CH7	CH0..CH7
PG-LQFP-100	CH0..CH7	CH0..CH7	CH0..CH3	CH0..CH3

1) Some pins in a package may be connected to more than one channel. For the detailed mapping see the Port I/O Function table.

1.5 Identification Registers

The identification registers allow software to identify the marking.

Table 7 XMC4700 Identification Registers

Register Name	Value	Marking
SCU_IDCHIP	0004 7001 _H	EES-AA, ES-AA, AA
JTAG IDCODE	101D F083 _H	EES-AA, ES-AA, AA

Table 8 XMC4800 Identification Registers

Register Name	Value	Marking
SCU_IDCHIP	0004 8001 _H	EES-AA, ES-AA, AA
JTAG IDCODE	101D F083 _H	EES-AA, ES-AA, AA

2 General Device Information

This section summarizes the logic symbols and package pin configurations with a detailed list of the functional I/O mapping.

2.1 Logic Symbols

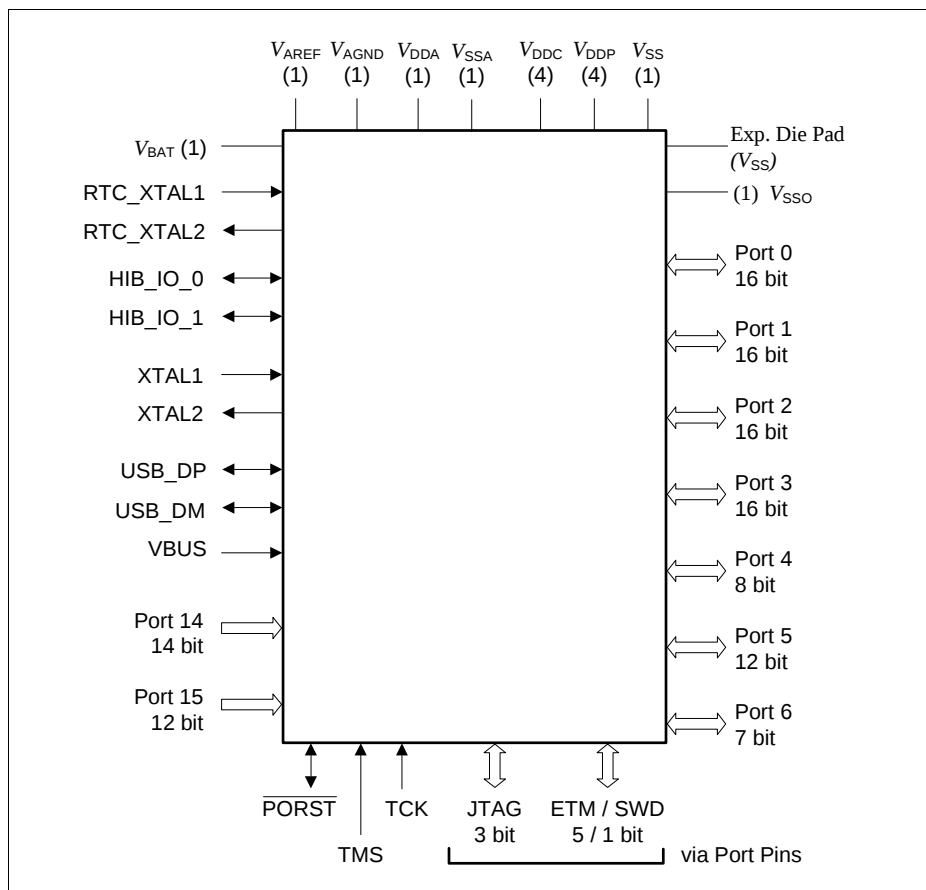


Figure 2 XMC4[78]00 Logic Symbol PG-LQFP-144

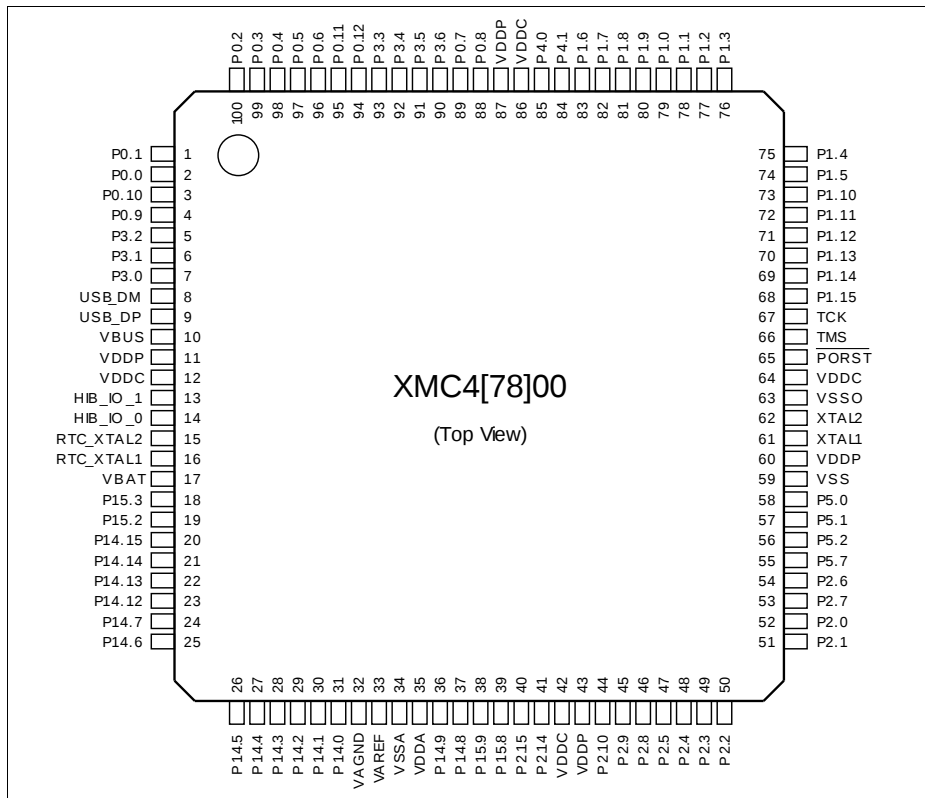


Figure 7 XMC4[78]00 PG-LQFP-100 Pin Configuration (top view)

General Device Information
Table 10 Package Pin Mapping (cont'd)

Function	LFBGA-196	LQFP-144	LQFP-100	Pad Type	Notes
P2.10	N8	66	44	A2	
P2.11	P8	65	-	A2	
P2.12	N7	64	-	A2	
P2.13	P7	63	-	A2	
P2.14	M7	60	41	A2	
P2.15	L6	59	40	A2	
P3.0	E1	7	7	A2	
P3.1	D2	6	6	A2	
P3.2	D3	5	5	A2	
P3.3	H7	132	93	A1+	
P3.4	G7	131	92	A1+	
P3.5	D6	130	91	A2	
P3.6	C7	129	90	A2	
P3.7	G4	14	-	A1+	
P3.8	G3	13	-	A1+	
P3.9	H5	12	-	A1+	
P3.10	H6	11	-	A1+	
P3.11	F3	10	-	A1+	
P3.12	F2	9	-	A2	
P3.13	E2	8	-	A2	
P3.14	F6	134	-	A1+	
P3.15	F7	133	-	A1+	
P4.0	D8	124	85	A2	
P4.1	C9	123	84	A2	
P4.2	G8	122	-	A1+	
P4.3	H8	121	-	A1+	
P4.4	E7	120	-	A1+	
P4.5	F8	119	-	A1+	
P4.6	E8	118	-	A1+	
P4.7	E9	117	-	A1+	
P5.0	K9	84	58	A1+	
P5.1	K8	83	57	A1+	
P5.2	K7	82	56	A1+	

Table 12 Port I/O Functions (cont'd)

Function	Outputs						Inputs									
	ALT1	ALT2	ALT3	ALT4	HWO0	HWO1	HWI0	HWI1	Input	Input	Input	Input	Input	Input	Input	Input
P2.12	ETH0_TXD2	ECAT0_P1_TXD1	CCU80_OUT33	ETH0_TXD0	DB_ETM_TRACEDA_TAI	EBU_AD30		EBU_D30					CCU43_IN3C			
P2.13	ETH0_TXD3	ECAT0_P1_TXD2		ETH0_TXD1	DB_ETM_TRACEDA_TAO	EBU_AD31		EBU_D31					CCU43_IN2C			
P2.14	VADC_EMUX11	UIC0_DOUT0	CCU80_OUT21	CAN_N4_TXD	DB_ETM_TRACECLK	EBU_BC0				UIC0_DX0D			CCU43_IN0B	CCU43_IN1B	CCU43_IN2B	CCU43_IN3B
P2.15	VADC_EMUX12	ECAT0_P1_TXD3	CCU80_OUT11	LEDTS0_LINE6	LEDTS0_EXTENDED6	EBU_BCI	LEDTS0_TSINGA		ETH0_COLA	UIC0_DX0C		CAN_N4_RXDA	CCU42_IN0B	CCU42_IN1B	CCU42_IN2B	CCU42_IN3B
P3.0	U2C1_SEL00	U0C1_SCLKOUT	CCU42_OUT0	ECAT0_P1_TX_ENA		EBU_RD			U0C1_DX1B				CCU80_IN2C	CCU81_IN0C		
P3.1		U0C1_SEL00	ECAT0_P1_TXD0			EBU_RD_WR			U0C1_DX2B		ERU0_0B1		CCU80_IN1C			
P3.2	USB_DRIVEVBUS	CAN_N0_TXD	ECAT0_P1_TXD1	LEDTS0_COLA		EBU_CS0					ERU0_0A1		CCU80_IN0C			
P3.3		UIC1_SEL01	CCU42_MCLK	ECAT0_OUT3	SDMMC_LED			EBU_WAIT		DS0_DIN3B			CCU42_IN3A	CCU80_IN3B		
P3.4	U2C1_MCLKOUT	UIC1_SEL02	CCU42_OUT2	DS0_MCLK3	SDMMC_BUS_POWER			EBU_HOLD	U2C1_DX0B	DS0_MCLK3B			CCU42_IN2A	CCU80_IN0B	ECAT0_P1_LINKA	
P3.5	U2C1_DOUT0	UIC1_SEL03	CCU42_OUT1	U0C1_DOUT0	SDMMC_CMD_OUT	EBU_AD4	SDMMC_CMD_IN	EBU_D4	U2C1_DX0A		ERU0_3B1		CCU42_IN1A		ECAT0_P1_RX_ERRA	
P3.6	U2C1_SCLKOUT	UIC1_SEL04	CCU42_OUT0	U0C1_SCLKOUT	SDMMC_CLK_OUT	EBU_AD5	SDMMC_CLK_IN	EBU_D5	U2C1_DX1B		ERU0_3A1		CCU42_IN0A			
P3.7	ECAT0_SYNC0	CAN_N2_TXD	CCU41_OUT3	LEDTS0_LINE0						U2C0_DX0C						
P3.8	U2C0_DOUT0	U0C1_SEL03	CCU41_OUT2	LEDTS0_LINE1							CAN_N2_RXDB		POSIF1_IN2B			
P3.9	U2C0_SCLKOUT	CAN_N1_TXD	CCU41_OUT1	LEDTS0_LINE2									POSIF1_IN1B			
P3.10	U2C0_SEL00	CAN_N0_TXD	CCU41_OUT0	LEDTS0_LINE3	U0C1_DOUT3		U0C1_HWI3						POSIF1_IN0B			
P3.11	U2C1_DOUT0	U0C1_SEL02	CCU42_OUT3	LEDTS0_LINE4	U0C1_DOUT2		U0C1_HWI2		CAN_N1_RXDB					CCU81_IN0C		
P3.12	ECAT0_P1_LINK_ACT	U0C1_SEL01	CCU42_OUT2	LEDTS0_LINE5	U0C1_DOUT1		U0C1_HWI1		CAN_N0_RXDC	U2C1_DX0D				CCU81_IN2C		
P3.13	U2C1_SCLKOUT	U0C1_DOUT0	CCU42_OUT1	LEDTS0_LINE6	U0C1_DOUT0		U0C1_HWI0		U0C1_DX0D				CCU80_IN3C	CCU81_IN1C		
P3.14		UIC0_SEL03			UIC1_DOUT1		UIC1_HWI1			UIC1_DX0B			CCU42_IN1C			
P3.15		UIC1_DOUT0			UIC1_DOUT0		UIC1_HWI0			UIC1_DX0A			CCU42_IN0C			
P4.0	CAN_N3_TXD	ECAT0_PHY_CLK25	DS0_MCLK1	UIC0_SCLKOUT	SDMMC_DATA0_OUT	EBU_AD6	SDMMC_DATA0_IN	EBU_D6	UIC1_DX1C	DS0_MCLK1B	U0C1_DX0E	U2C1_DX0C				ECAT0_P0_RX_ERRA
P4.1	U2C1_SEL00	UIC1_MCLKOUT	DS0_MCLK0	U0C1_SEL00	SDMMC_DATA3_OUT	EBU_AD9	SDMMC_DATA3_IN	EBU_D9	U2C1_DX2B	DS0_MCLK0B		U2C1_DX2A	DS0_MCLK1D			ECAT0_P0_LINKA
P4.2	U2C1_SEL01	UIC1_DOUT0		UIC1_SCLKOUT	ECAT0_MDO		ECAT0_MDI0		UIC1_DX0C			U2C1_DX1A	CCU43_IN1C			

2.3 Power Connection Scheme

Figure 9. shows a reference power connection scheme for the XMC4[78]00.

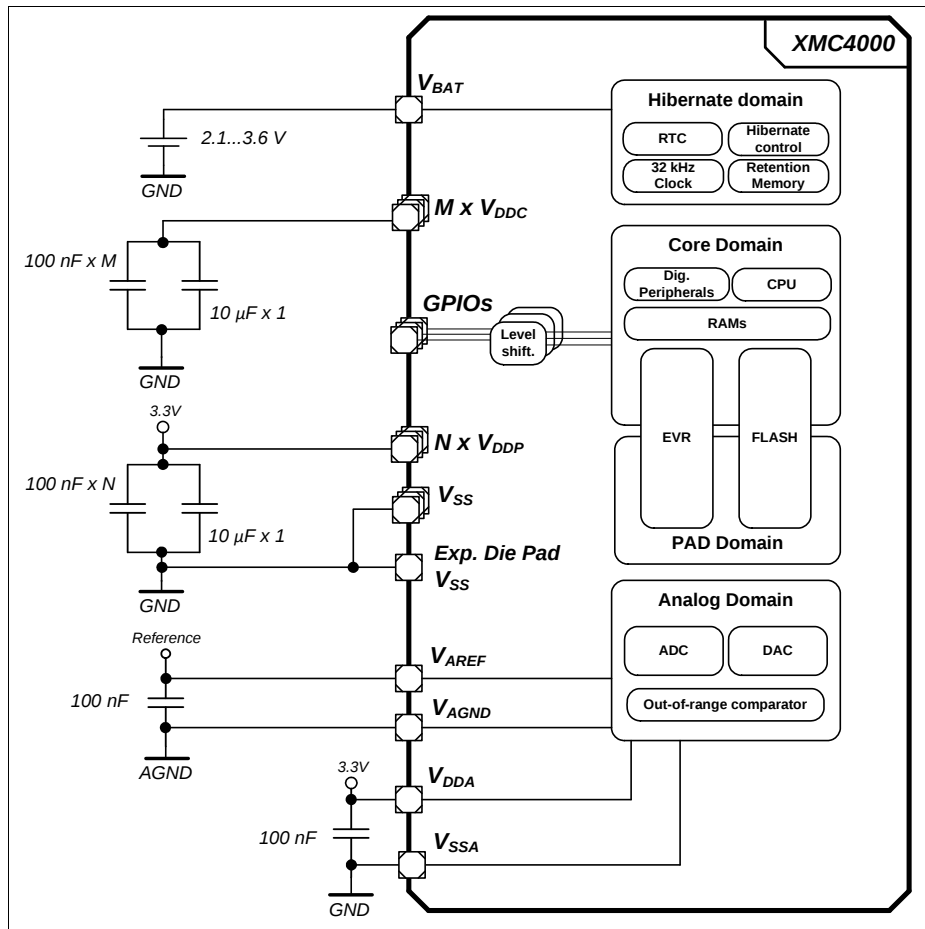


Figure 9 Power Connection Scheme

Every power supply pin needs to be connected. Different pins of the same supply need also to be externally connected. As example, all V_{DDP} pins must be connected externally to one V_{DDP} net. In this reference scheme one 100 nF capacitor is connected at each supply pin against V_{SS} . An additional 10 μF capacitor is connected to the V_{DDP} nets and an additional 10 μF capacitor to the V_{DDC} nets.

Electrical Parameters
Table 25 VADC Parameters (Operating Conditions apply)

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Total Unadjusted Error	TUE_{CC}	-4	—	4	LSB	12-bit resolution; $V_{DDA} = 3.3 V$; $V_{AREF} = V_{DDA}$ ⁷⁾
Differential Non-Linearity Error ⁸⁾	$EA_{DNL_{CC}}$	-3	—	3	LSB	
Gain Error ⁸⁾	$EA_{GAIN_{CC}}$	-4	—	4	LSB	
Integral Non-Linearity ⁸⁾	$EA_{INL_{CC}}$	-3	—	3	LSB	
Offset Error ⁸⁾	$EA_{OFF_{CC}}$	-4	—	4	LSB	
Worst case ADC V_{DDA} power supply current per active converter	$I_{DDAA_{CC}}$	—	1.5	2	mA	during conversion $V_{DDP} = 3.6 V$, $T_j = 150 ^\circ C$
Charge consumption on V_{AREF} per conversion ⁵⁾	$Q_{CONV_{CC}}$	—	30	—	pC	$0 V \leq V_{AREF} \leq V_{DDA}$ ⁹⁾
ON resistance of the analog input path	$R_{AIN_{CC}}$	—	600	1 200	Ohm	
ON resistance for the ADC test (pull down for AIN7)	$R_{AIN7T_{CC}}$	180	550	900	Ohm	
Resistance of the reference voltage input path	$R_{AREF_{CC}}$	—	700	1 700	Ohm	

- 1) A running conversion may become imprecise in case the normal conditions are violated (voltage overshoot).
- 2) If the analog reference voltage is below V_{DDA} , then the ADC converter errors increase. If the reference voltage is reduced by the factor k ($k < 1$), TUE, DNL, INL, Gain, and Offset errors increase also by the factor $1/k$.
- 3) The leakage current definition is a continuous function, as shown in figure ADCx Analog Inputs Leakage. The numerical values defined determine the characteristic points of the given continuous linear approximation - they do not define step function (see [Figure 16](#)).
- 4) The sampling capacity of the conversion C-network is pre-charged to $V_{AREF}/2$ before the sampling moment. Because of the parasitic elements, the voltage measured at AINx can deviate from $V_{AREF}/2$.
- 5) Applies to AINx, when used as alternate reference input.
- 6) This represents an equivalent switched capacitance. This capacitance is not switched to the reference voltage at once. Instead, smaller capacitances are successively switched to the reference voltage.
- 7) For 10-bit conversions, the errors are reduced to 1/4; for 8-bit conversions, the errors are reduced to 1/16. Never less than ± 1 LSB.
- 8) The sum of DNL/INL/GAIN/OFF errors does not exceed the related total unadjusted error TUE.
- 9) The resulting current for a conversion can be calculated with $I_{AREF} = Q_{CONV} / t_c$.
The fastest 12-bit post-calibrated conversion of $t_c = 459$ ns results in a typical average current of $I_{AREF} = 65.4 \mu A$.

Table 38 Power Sequencing Parameters

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Positive Load Step Current	ΔI_{PLS} SR	-	–	50	mA	Load increase on V_{DDP} $\Delta t \leq 10$ ns
Negative Load Step Current	ΔI_{NLS} SR	-	–	150	mA	Load decrease on V_{DDP} $\Delta t \leq 10$ ns
V_{DDC} Voltage Over- / Undershoot from Load Step	ΔV_{LS} CC	-	–	± 100	mV	For maximum positive or negative load step
Positive Load Step Settling Time	t_{PLSS} SR	50	–	-	μ s	
Negative Load Step Settling Time	t_{NLSS} SR	100	–	-	μ s	
External Buffer Capacitor on V_{DDC}	C_{EXT} SR	-	10	-	μ F	In addition $C = 100$ nF capacitor on each V_{DDC} pin

Positive Load Step Examples

System assumptions:

$f_{CPU} = f_{SYS}$, target frequency $f_{CPU} = 144$ MHz, main PLL $f_{VCO} = 288$ MHz, stepping done by K2 divider, t_{PLSS} between individual steps:

24 MHz - 48 MHz - 72 MHz - 96 MHz - 144 MHz (K2 steps 12 - 6 - 4 - 3 - 2)

24 MHz - 48 MHz - 96 MHz - 144 MHz (K2 steps 12 - 6 - 3 - 2)

24 MHz - 72 MHz - 144 MHz (K2 steps 12 - 4 - 2)

3.3.5 Internal Clock Source Characteristics

Note: These parameters are not subject to production test, but verified by design and/or characterization.

Fast Internal Clock Source

Table 40 Fast Internal Clock Parameters

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Nominal frequency	f_{OFINC} CC	–	36.5	–	MHz	not calibrated
		–	24	–	MHz	calibrated
Accuracy	Δf_{OFI} CC	-0.5	–	0.5	%	automatic calibration ¹⁾²⁾
		-15	–	15	%	factory calibration, $V_{\text{DDP}} = 3.3 \text{ V}$
		-25	–	25	%	no calibration, $V_{\text{DDP}} = 3.3 \text{ V}$
		-7	–	7	%	Variation over voltage range ³⁾ $3.13 \text{ V} \leq V_{\text{DDP}} \leq 3.63 \text{ V}$
Start-up time	t_{OFIS} CC	–	50	–	μs	

1) Error in addition to the accuracy of the reference clock.

2) Automatic calibration compensates variations of the temperature and in the V_{DDP} supply voltage.

3) Deviations from the nominal V_{DDP} voltage induce an additional error to the uncalibrated and/or factory calibrated oscillator frequency.

Slow Internal Clock Source
Table 41 Slow Internal Clock Parameters

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Nominal frequency	f_{OSI} CC	–	32.768	–	kHz	
Accuracy	Δf_{OSI} CC	-4	–	4	%	$V_{BAT} = \text{const.}$ $0\text{ }^{\circ}\text{C} \leq T_A \leq 85\text{ }^{\circ}\text{C}$
		-5	–	5	%	$V_{BAT} = \text{const.}$ $T_A < 0\text{ }^{\circ}\text{C}$ or $T_A > 85\text{ }^{\circ}\text{C}$
		-5	–	5	%	$2.4\text{ V} \leq V_{BAT}$, $T_A = 25\text{ }^{\circ}\text{C}$
		-10	–	10	%	$1.95\text{ V} \leq V_{BAT} < 2.4\text{ V}$, $T_A = 25\text{ }^{\circ}\text{C}$
Start-up time	t_{OSIS} CC	–	50	–	μs	

3.3.9 Peripheral Timing

3.3.9.1 Delta-Sigma Demodulator Digital Interface Timing

The following parameters are applicable for the digital interface of the Delta-Sigma Demodulator (DSD).

The data timing is relative to the active clock edge. Depending on the operation mode of the connected modulator that can be the rising and falling clock edge.

Note: These parameters are not subject to production test, but verified by design and/or characterization.

Table 45 DSD Interface Timing Parameters

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
MCLK period in master mode	t_1 CC	33.3	—	—	ns	$t_1 \geq 4 \times t_{\text{PERIPH}}^{1)}$
MCLK high time in master mode	t_2 CC	9	—	—	ns	$t_2 > t_{\text{PERIPH}}^{1)}$
MCLK low time in master mode	t_3 CC	9	—	—	ns	$t_3 > t_{\text{PERIPH}}^{1)}$
MCLK period in slave mode	t_1 SR	33.3	—	—	ns	$t_1 \geq 4 \times t_{\text{PERIPH}}^{1)}$
MCLK high time in slave mode	t_2 SR	t_{PERIPH}	—	—	ns	$^{1)}$
MCLK low time in slave mode	t_3 SR	t_{PERIPH}	—	—	ns	$^{1)}$
DIN input setup time to the active clock edge	t_4 SR	$t_{\text{PERIPH}} + 4$	—	—	ns	$^{1)}$
DIN input hold time from the active clock edge	t_5 SR	$t_{\text{PERIPH}} + 3$	—	—	ns	$^{1)}$

$^{1)} t_{\text{PERIPH}} = 1 / f_{\text{PERIPH}}$

High-Speed Input Path (Read)

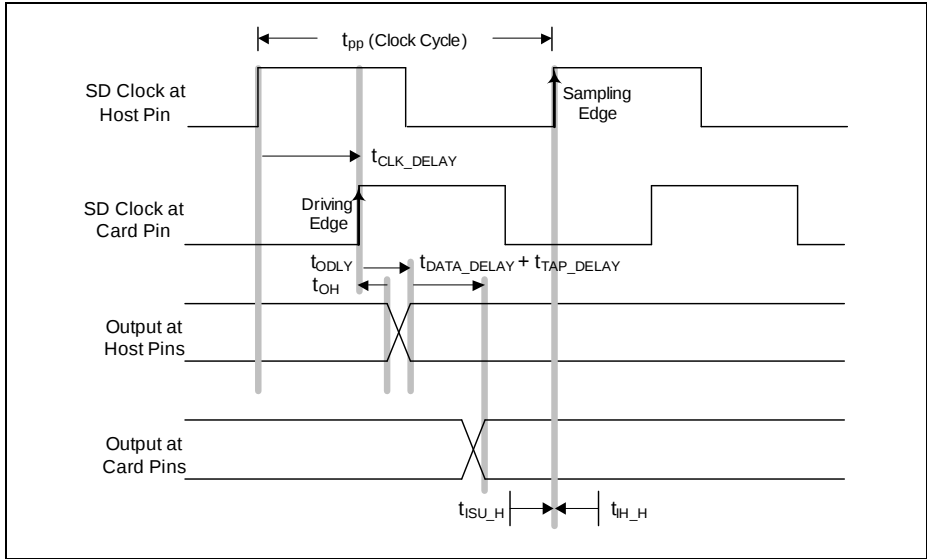


Figure 40 High-Speed Input Path

High-Speed Read Meeting Setup (Maximum Delay)

The following equations show how to calculate the allowed combined propagation delay range of the SD_CLK and SD_DAT/CMD signals on the PCB.

(11)

$$t_{CLK_DELAY} + t_{DATA_DELAY} + t_{TAP_DELAY} + t_{ODLY} + t_{ISU_H} < t_{pp}$$

$$t_{CLK_DELAY} + t_{DATA_DELAY} < t_{pp} - t_{ODLY} - t_{ISU_H} - t_{TAP_DELAY}$$

$$t_{CLK_DELAY} + t_{DATA_DELAY} < 20 - 14 - 2 - t_{TAP_DELAY}$$

$$t_{CLK_DELAY} + t_{DATA_DELAY} < 4 - t_{TAP_DELAY}$$

The data + clock delay can be up to 4 ns for a 20 ns clock cycle.

Write Timing

Table 58 Asynchronous Write Timing, Multiplexed and Demultiplexed

Parameter			Symbol	Limit Values		Unit
				Min.	Max.	
A(24:0) output delay	to RD/ $\overline{\text{WR}}$ rising edge, deviation from the ideal programmed value.	CC	t_{30}	-2.5	2.5	ns
A(24:0) output delay		CC	t_{31}	-2.5	2.5	
$\overline{\text{CS}}$ rising edge		CC	t_{32}	-2	2	
$\overline{\text{ADV}}$ rising edge		CC	t_{33}	-2	4.5	
$\overline{\text{BC}}$ rising edge		CC	t_{34}	-2.5	2	
$\overline{\text{WAIT}}$ input setup		SR	t_{35}	12	–	
$\overline{\text{WAIT}}$ input hold		SR	t_{36}	0	–	
Data output delay		CC	t_{37}	-5.5	2	
Data output delay		CC	t_{38}	-5.5	2	
RD / $\overline{\text{WR}}$ output delay		CC	t_{39}	-2.5	1.5	

Demultiplexed Write Timing

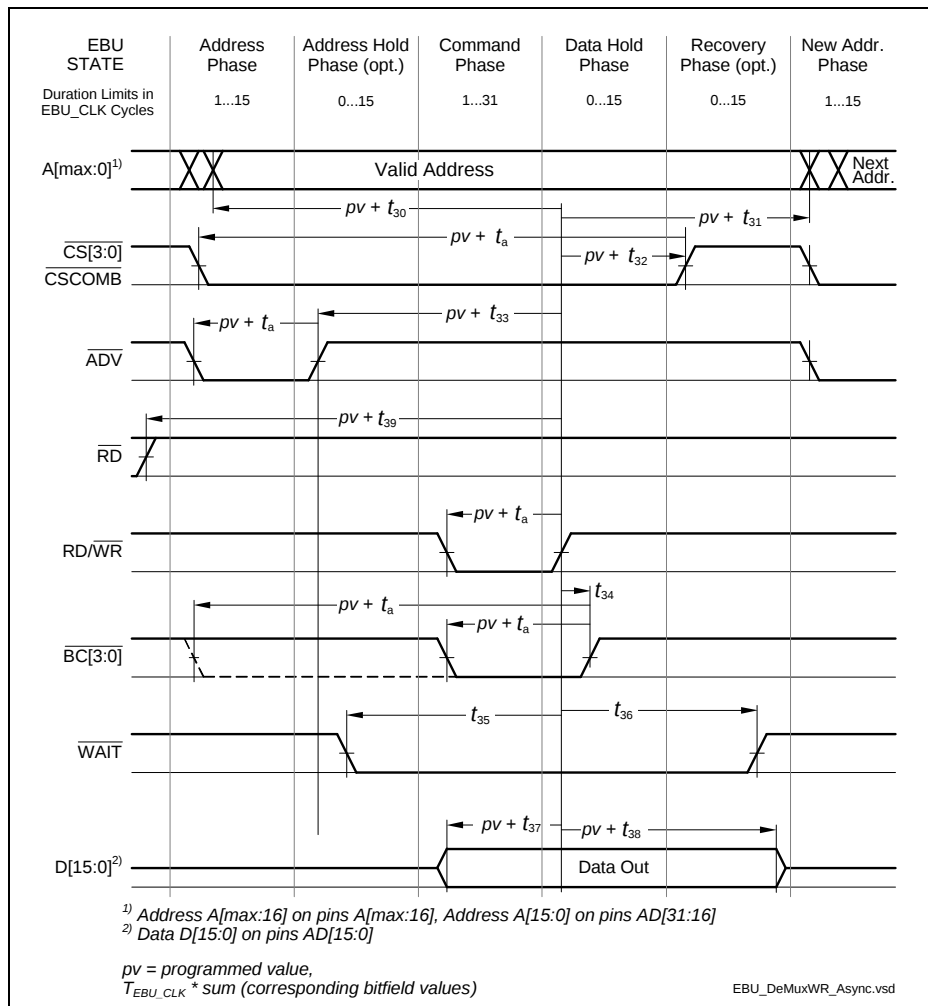


Figure 44 Demultiplexed Write Access

3.3.10.2 EBU Burst Mode Access Timing

Note: These parameters are not subject to production test, but verified by design and/or characterization.

Note: Operating Conditions apply, with Class A2 pins and $C_L = 16$ pF.

Table 59 EBU Burst Mode Read / Write Access Timing Parameters

Parameter	Symbol		Values			Unit	Note / Test Condition
			Min.	Typ.	Max.		
Output delay from BFCLKO rising edge	t_{10}	CC	-2	—	2	ns	—
$\overline{RD}$ and $\overline{RD}/\overline{WR}$ active/inactive after BFCLKO active edge ¹⁾	t_{12}	CC	-2	—	2	ns	—
$\overline{CSx}$ output delay from BFCLKO active edge ¹⁾	t_{21}	CC	-2.5	—	1.5	ns	—
$\overline{ADV}$ active/inactive after BFCLKO active edge ²⁾	t_{22}	CC	-2	—	2	ns	—
$\overline{BAA}$ active/inactive after BFCLKO active edge ²⁾	t_{22a}	CC	-2.5	—	1.5	ns	—
Data setup to BFCLKI rising edge ³⁾	t_{23}	SR	3	—	—	ns	—
Data hold from BFCLKI rising edge ³⁾	t_{24}	SR	0	—	—	ns	—
$\overline{WAIT}$ setup (low or high) to BFCLKI rising edge ³⁾	t_{25}	SR	3	—	—	ns	—
$\overline{WAIT}$ hold (low or high) from BFCLKI rising edge ³⁾	t_{26}	SR	0	—	—	ns	—

1) An active edge can be a rising or falling edge, depending on the settings of bits BFCN.EBSE / ECSE and the clock divider ratio.

Negative minimum values for these parameters mean that the last data read during a burst may be corrupted. However, with clock feedback enabled, this value is an oversampling not required for the internal bus transaction, and will be discarded.

2) This parameter is valid for BUSCONx.EBSE = 1 and BUSAPx.EXTCLK = 00_B.

For BUSCONx.EBSE = 1 and other values of BUSAPx.EXTCLK, ADV and BAA will be delayed by 1/2 of the internal bus clock period $T_{CPU} = 1 / f_{CPU}$.

For BUSCONx. EBSE = 0 and BUSAPx.EXTCLK = 11_B, add 2 internal bus clock periods.

For BUSCONx. EBSE = 0 and other values of BUSAPx.EXTCLK, add 1 internal bus clock period.

Table 62 EBU SDRAM Access Signal Timing Parameters

Parameter			Symbol	Limit Values		Unit
				Min.	Max.	
A(15:0) output valid	from SDCLKO low-to-high transition	CC	t_6	–	9	ns
A(15:0) output hold		CC	t_7	3	–	
$\overline{\text{CS}}(3:0)$ low		CC	t_8	–	9	
$\overline{\text{CS}}(3:0)$ high		CC	t_9	3	–	
$\overline{\text{RAS}}$ low		CC	t_{10}	–	9	
$\overline{\text{RAS}}$ high		SR	t_{11}	3	–	
$\overline{\text{CAS}}$ low		SR	t_{12}	–	9	
$\overline{\text{CAS}}$ high		CC	t_{13}	3	–	
RD/WR low		CC	t_{14}	–	9	
RD/WR high		CC	t_{15}	3	–	
$\overline{\text{BC}}(3:0)$ low		CC	t_{16}	–	9	
$\overline{\text{BC}}(3:0)$ high		CC	t_{17}	3	–	
D(15:0) output valid		CC	t_{18}	–	9	
D(15:0) output hold		CC	t_{19}	3	–	
CKE output valid ¹⁾		CC	t_{22}	–	7	
CKE output hold ¹⁾		CC	t_{23}	2	–	
D(15:0) input hold	D(15:0) input setup to SDCLKO low-to-high transition	SR	t_{21}	3	–	
D(15:0) input setup to SDCLKO low-to-high transition		SR	t_{20}	4	–	

1) Not depicted in the read and write access timing figures below.

3.3.11 USB Interface Characteristics

The Universal Serial Bus (USB) Interface is compliant to the USB Rev. 2.0 Specification and the OTG Specification Rev. 1.3. High-Speed Mode is not supported.

Note: These parameters are not subject to production test, but verified by design and/or characterization.

Table 63 USB Timing Parameters (operating conditions apply)

Parameter	Symbol		Values			Unit	Note / Test Condition
			Min.	Typ.	Max.		
Rise time	t_R	CC	4	–	20	ns	$C_L = 50 \text{ pF}$
Fall time	t_F	CC	4	–	20	ns	$C_L = 50 \text{ pF}$
Rise/Fall time matching	t_R/t_F	CC	90	–	111.11	%	$C_L = 50 \text{ pF}$
Crossover voltage	V_{CRS}	CC	1.3	–	2.0	V	$C_L = 50 \text{ pF}$

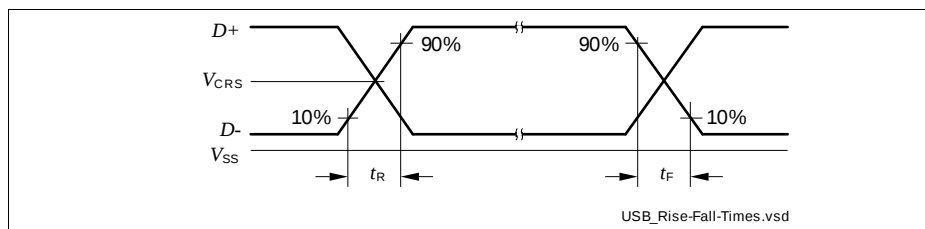


Figure 50 USB Signal Timing

Table 68 **ETH MII TX Signal Timing Parameters (cont'd)**

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
PHY setup requirement: TXEN/TXD[3:0] with respect to TX_CLK	t_{TX_setup} SR	15	—	0	ns	PHY dependent IEEE802.3 limit is 15 ns
PHY hold requirement: TXEN/TXD[3:0] with respect to TX_CLK	t_{TX_hold} CC	0	—	25	ns	PHY dependent IEEE802.3 limit is 0 ns

Note: ECAT0_CONPx.TX_SHIFT can be adjusted by displaying TX_CLK of a PHY and TXEN/TXD[3:0] on an oscilloscope. TXEN/TXD[3:0] is allowed to change between 0 ns and 25 ns after a rising edge of TX_CLK (according to IEEE802.3 – check your PHY's documentation). Configure TX_SHIFT so that TXEN/TXD[3:0] change near the middle of this range. It is sufficient to check just one of the TXEN/TXD[3:0] signals, because they are nearly generated at the same time.

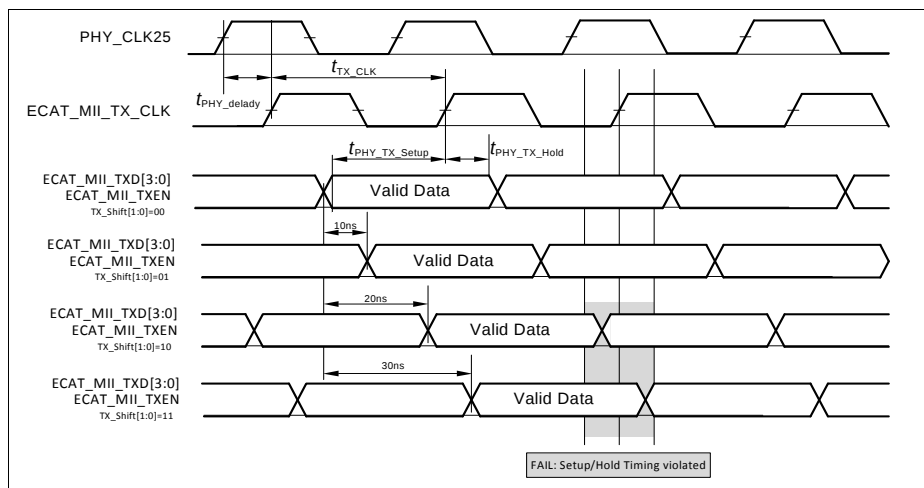


Figure 57 **MII TX Characteristics**