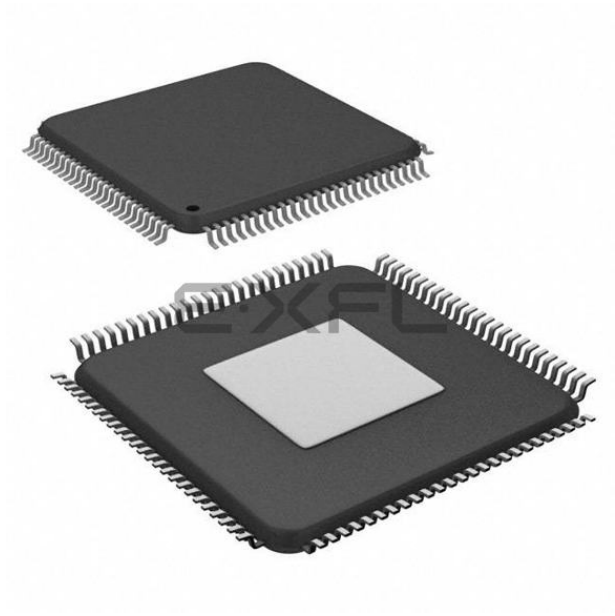
Infineon Technologies - XMC4800F100K1536AAXQMA1 Datasheet



Welcome to E-XFL.COM

What is "Embedded - Microcontrollers"?

"Embedded - Microcontrollers" refer to small, integrated circuits designed to perform specific tasks within larger systems. These microcontrollers are essentially compact computers on a single chip, containing a processor core, memory, and programmable input/output peripherals. They are called "embedded" because they are embedded within electronic devices to control various functions, rather than serving as standalone computers. Microcontrollers are crucial in modern electronics, providing the intelligence and control needed for a wide range of applications.

Applications of "Embedded - Microcontrollers"

Details	
Product Status	Active
Core Processor	ARM® Cortex®-M4
Core Size	32-Bit Single-Core
Speed	144MHz
Connectivity	CANbus, EBI/EMI, Ethernet, I²C, LINbus, MMC/SD, SPI, UART/USART, USB OTG, USIC
Peripherals	DMA, I²S, LED, POR, Touch-Sense, WDT
Number of I/O	75
Program Memory Size	1.5MB (1.5M x 8)
Program Memory Type	FLASH
EEPROM Size	-
RAM Size	276K x 8
Voltage - Supply (Vcc/Vdd)	3.13V ~ 3.63V
Data Converters	A/D 24x12b; D/A 2x12b
Oscillator Type	External
Operating Temperature	-40°C ~ 125°C (TA)
Mounting Type	Surface Mount
Package / Case	100-LQFP Exposed Pad
Supplier Device Package	PG-LQFP-100-25
Purchase URL	https://www.e-xfl.com/product-detail/infineon-technologies/xmc4800f100k1536aaxqma1

Table 10 Package Pin Mapping (cont'd)

Function	LFBGA-196	LQFP-144	LQFP-100	Pad Type	Notes
P5.3	L10	81	-	A2	
P5.4	M10	80	-	A2	
P5.5	L8	79	-	A2	
P5.6	M8	78	-	A2	
P5.7	L7	77	55	A1+	
P5.8	K6	58	-	A2	
P5.9	M6	57	-	A2	
P5.10	K5	56	-	A1+	
P5.11	L5	55	-	A1+	
P6.0	J10	101	-	A2	
P6.1	H9	100	-	A2	
P6.2	K10	99	-	A2	
P6.3	J9	98	-	A1+	
P6.4	H10	97	-	A2	
P6.5	H11	96	-	A2	
P6.6	H12	95	-	A2	
P7.0	L13	-	-	A2	
P7.1	M13	-	-	A2	
P7.2	N13	-	-	A2	
P7.3	M14	-	-	A2	
P7.4	N14	-	-	A1+	
P7.5	L14	-	-	A1+	
P7.6	K14	-	-	A1+	
P7.7	J14	-	-	A1+	
P7.8	H14	-	-	A2	
P7.9	G13	-	-	A1+	
P7.10	G14	-	-	A1+	
P7.11	F14	-	-	A1+	
P8.0	B7	-	-	A2	
P8.1	A7	-	-	A2	
P8.2	B3	-	-	A2	
P8.3	B2	-	-	A2	
P8.4	B6	-	-	A1+	

2.2.2 Port I/O Functions

The following general scheme is used to describe each Port pin:

Table 11 Port I/O Function Description

Function	Outputs			Inputs		
	ALT1	ALTn	HWO0	HWI0	Input	Input
P0.0		MODA.OUT	MODB.OUT	MODB.INA	MODC.INA	
Pn.y	MODA.OUT				MODA.INA	MODC.INB

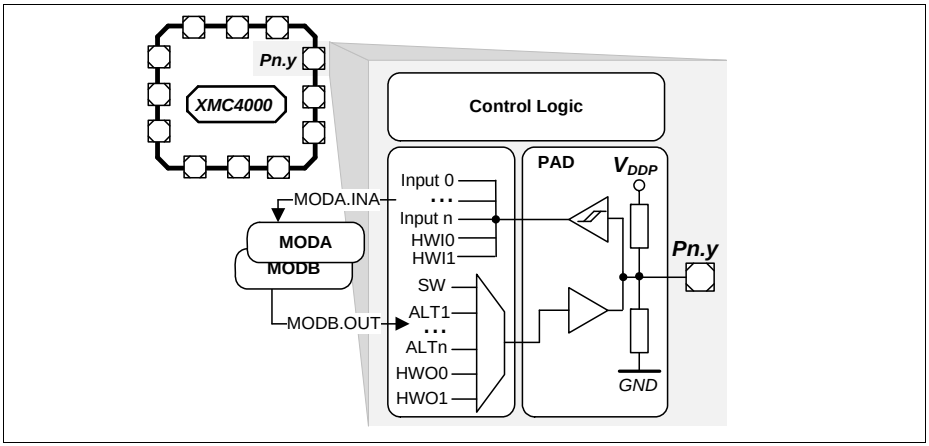


Figure 8 Simplified Port Structure

Pn.y is the port pin name, defining the control and data bits/registers associated with it. As GPIO, the port is under software control. Its input value is read via Pn_IN.y, Pn_OUT defines the output value.

Up to four alternate output functions (ALT1/2/3/4) can be mapped to a single port pin, selected by Pn_IOCR.PC. The output value is directly driven by the respective module, with the pin characteristics controlled by the port registers (within the limits of the connected pad).

The port pin input can be connected to multiple peripherals. Most peripherals have an input multiplexer to select between different possible input sources.
The input path is also active while the pin is configured as output. This allows to feedback an output to on-chip resources without wasting an additional external pin.

By Pn_HWSEL it is possible to select between different hardware “masters” (HWO0/HWI0). The selected peripheral can take control of the pin(s). Hardware control overrules settings in the respective port pin registers.

The XMC4[78]00 has a common ground concept, all V_{SS}, V_{SSA} and V_{SSO} pins share the same ground potential. In packages with an exposed die pad it must be connected to the common ground as well.

V_{AGND} is the low potential to the analog reference V_{AREF}. Depending on the application it can share the common ground or have a different potential. In devices with shared V_{DDA}/V_{AREF} and V_{SSA}/V_{AGND} pins the reference is tied to the supply. Some analog channels can optionally serve as “Alternate Reference”; further details on this operating mode are described in the Reference Manual.

When V_{DDP} is supplied, V_{BAT} must be supplied as well. If no other supply source (e.g. battery) is connected to V_{BAT}, the V_{BAT} pin can also be connected directly to V_{DDP}.

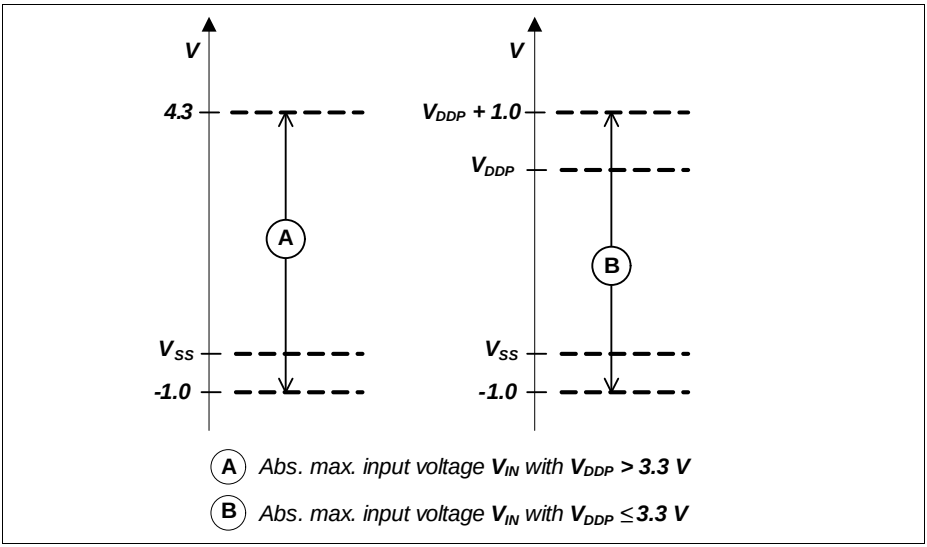


Figure 10 Absolute Maximum Input Voltage Ranges

3.1.3 Pin Reliability in Overload

When receiving signals from higher voltage devices, low-voltage devices experience overload currents and voltages that go beyond their own IO power supplies specification.

Table 14 defines overload conditions that will not cause any negative reliability impact if all the following conditions are met:

- full operation life-time is not exceeded
- **Operating Conditions** are met for
  - pad supply levels (V_{DDP} or V_{DDA})
  - temperature

If a pin current is outside of the **Operating Conditions** but within the overload conditions, then the parameters of this pin as stated in the Operating Conditions can no longer be guaranteed. Operation is still possible in most cases but with relaxed parameters.

Note: An overload condition on one or more pins does not require a reset.

Note: A series resistor at the pin to limit the current to the maximum permitted overload current is sufficient to handle failure situations like short to battery.

Table 14 Overload Parameters

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Input current on any port pin during overload condition	I_{OV} SR	-5	–	5	mA	
Absolute sum of all input circuit currents for one port group during overload condition[1)]	I_{OVG} SR	–	–	20	mA	$\Sigma\|I_{OVx}\|$, for all $I_{OVx} < 0$ mA
		–	–	20	mA	$\Sigma\|I_{OVx}\|$, for all $I_{OVx} > 0$ mA
Absolute sum of all input circuit currents during overload condition	I_{OVS} SR	–	–	80	mA	ΣI_{OVG}

1) The port groups are defined in **Table 17**.

Figure 11 shows the path of the input currents during overload via the ESD protection structures. The diodes against V_{DDP} and ground are a simplified representation of these ESD protection structures.

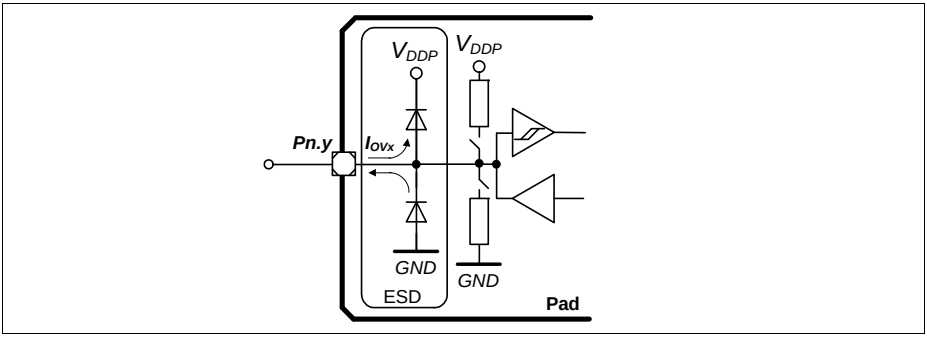


Figure 11 Input Overload Current via ESD structures

Table 15 and **Table 16** list input voltages that can be reached under overload conditions. Note that the absolute maximum input voltages as defined in the **Absolute Maximum Ratings** must not be exceeded during overload.

3.1.4 Pad Driver and Pad Classes Summary

This section gives an overview on the different pad driver classes and their basic characteristics.

Table 18 Pad Driver and Pad Classes Overview

Class	Power Supply	Type	Sub-Class	Speed Grade	Load	Termination
A	3.3 V	LVTTL I/O	**A1** (e.g. GPIO)	6 MHz	100 pF	No
			A1+ (e.g. serial I/Os)	25 MHz	50 pF	Series termination recommended
			A2 (e.g. ext. Bus)	80 MHz	15 pF	Series termination recommended

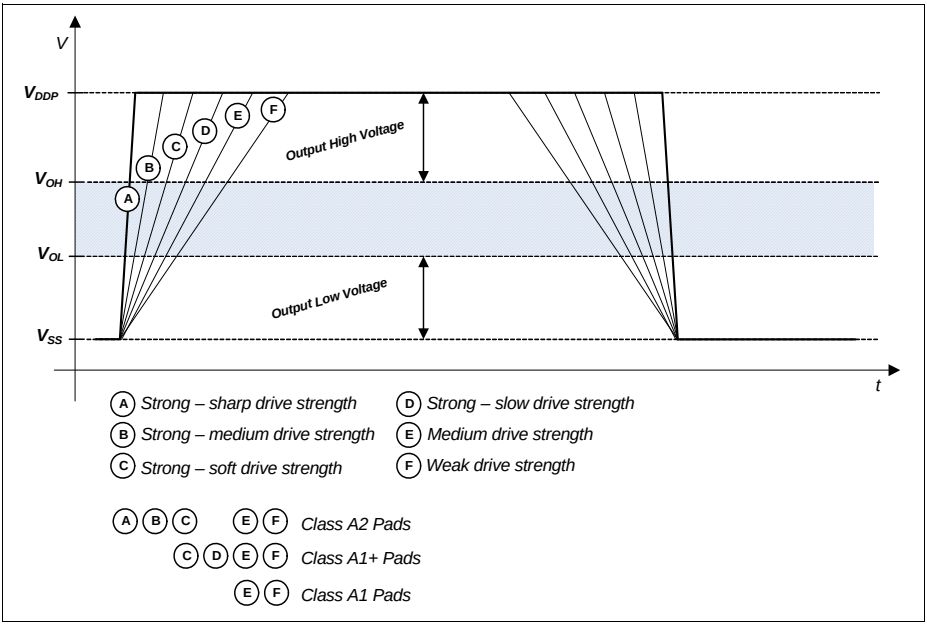


Figure 12 Output Slopes with different Pad Driver Modes

Figure 12 is a qualitative display of the resulting output slope performance with different output driver modes. The detailed input and output characteristics are listed in **Section 3.2.1**.

3.2 DC Parameters

3.2.1 Input/Output Pins

The digital input stage of the shared analog/digital input pins is identical to the input stage of the standard digital input/output pins.
The Pull-up on the $\overline{\text{PORST}}$ pin is identical to the Pull-up on the standard digital input/output pins.

Note: These parameters are not subject to production test, but verified by design and/or characterization.

Table 20 Standard Pad Parameters

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.		Max.		
Pin capacitance (digital inputs/outputs)	C_{IO} CC	–		10	pF	
Pull-down current	$\lvert I_{PDL}\rvert$ SR	150		–	µA	1) $V_{IN} \geq 0.6 \times V_{DDP}$
		–		10	µA	2) $V_{IN} \leq 0.36 \times V_{DDP}$
Pull-Up current	$\lvert I_{PUH}\rvert$ SR	–		10	µA	2) $V_{IN} \geq 0.6 \times V_{DDP}$
		100		–	µA	1) $V_{IN} \leq 0.36 \times V_{DDP}$
Input Hysteresis for pads of all A classes[3]	$HYSA$ CC	$0.1 \times V_{DDP}$		–	V	
$\overline{\text{PORST}}$ spike filter always blocked pulse duration	t_{SF1} CC	–		10	ns	
$\overline{\text{PORST}}$ spike filter pass-through pulse duration	t_{SF2} CC	100		–	ns	
$\overline{\text{PORST}}$ pull-down current	$\lvert I_{PPD}\rvert$ CC	13		–	mA	V_{IN} = 1.0 V

1) Current required to override the pull device with the opposite logic level (“force current”). With active pull device, at load currents between force and keep current the input state is undefined.

2) Load current at which the pull device still maintains the valid logic level (“keep current”). With active pull device, at load currents between force and keep current the input state is undefined.

3) Hysteresis is implemented to avoid metastable states and switching due to internal ground bounce. It can not be guaranteed that it suppresses switching due to external system noise.

Table 22 Standard Pads Class_A1+

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.		Max.		
Output high voltage, POD[1] = weak	V_{OHA1+} CC	V_{DDP} - 0.4		–	V	$I_{OH} \geq$ -400 µA
		2.4		–	V	$I_{OH} \geq$ -500 µA
Output high voltage, POD[1] = medium		V_{DDP} - 0.4		–	V	$I_{OH} \geq$ -1.4 mA
		2.4		–	V	$I_{OH} \geq$ -2 mA
Output high voltage, POD[1] = strong		V_{DDP} - 0.4		–	V	$I_{OH} \geq$ -1.4 mA
		2.4		–	V	$I_{OH} \geq$ -2 mA
Output low voltage	V_{OLA1+} CC	–		0.4	V	$I_{OL} \leq$ 500 µA; POD[1] = weak
		–		0.4	V	$I_{OL} \leq$ 2 mA; POD[1] = medium
		–		0.4	V	$I_{OL} \leq$ 2 mA; POD[1] = strong
Fall time	t_{FA1+} CC	–		150	ns	C_L = 20 pF; POD[1] = weak
		–		50	ns	C_L = 50 pF; POD[1] = medium
		–		28	ns	C_L = 50 pF; POD[1] = strong; edge = slow
		–		16	ns	C_L = 50 pF; POD[1] = strong; edge = soft;
Rise time	t_{RA1+} CC	–		150	ns	C_L = 20 pF; POD[1] = weak
		–		50	ns	C_L = 50 pF; POD[1] = medium
		–		28	ns	C_L = 50 pF; POD[1] = strong; edge = slow
		–		16	ns	C_L = 50 pF; POD[1] = strong; edge = soft

1) POD = Pin Out Driver

Table 25 VADC Parameters (Operating Conditions apply)

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Total Unadjusted Error	TUE CC	-4	–	4	LSB	12-bit resolution; V_{DDA} = 3.3 V; $V_{AREF} = V_{DDA}$ 7)
Differential Non-Linearity Error 8)	EA_{DNL} CC	-3	–	3	LSB	
Gain Error 8)	EA_{GAIN} CC	-4	–	4	LSB	
Integral Non-Linearity 8)	EA_{INL} CC	-3	–	3	LSB	
Offset Error 8)	EA_{OFF} CC	-4	–	4	LSB	
Worst case ADC V_{DDA} power supply current per active converter	I_{DDAA} CC	–	1.5	2	mA	during conversion V_{DDP} = 3.6 V, T_J = 150 °C
Charge consumption on V_{AREF} per conversion 5)	Q_{CONV} CC	–	30	–	pC	0 V ≤ V_{AREF} ≤ V_{DDA} 9)
ON resistance of the analog input path	R_{AIN} CC	–	600	1 200	Ohm	
ON resistance for the ADC test (pull down for AIN7)	R_{AIN7T} CC	180	550	900	Ohm	
Resistance of the reference voltage input path	R_{AREF} CC	–	700	1 700	Ohm	

1) A running conversion may become imprecise in case the normal conditions are violated (voltage overshoot).

2) If the analog reference voltage is below V_{DDA}, then the ADC converter errors increase. If the reference voltage is reduced by the factor k (k<1), TUE, DNL, INL, Gain, and Offset errors increase also by the factor 1/k.

3) The leakage current definition is a continuous function, as shown in figure ADCx Analog Inputs Leakage. The numerical values defined determine the characteristic points of the given continuous linear approximation - they do not define step function (see **Figure 16**).

4) The sampling capacity of the conversion C-network is pre-charged to V_{AREF}/2 before the sampling moment. Because of the parasitic elements, the voltage measured at AINx can deviate from V_{AREF}/2.

5) Applies to AINx, when used as alternate reference input.

6) This represents an equivalent switched capacitance. This capacitance is not switched to the reference voltage at once. Instead, smaller capacitances are successively switched to the reference voltage.

7) For 10-bit conversions, the errors are reduced to 1/4; for 8-bit conversions, the errors are reduced to 1/16. Never less than ±1 LSB.

8) The sum of DNL/INL/GAIN/OFF errors does not exceed the related total unadjusted error TUE.

9) The resulting current for a conversion can be calculated with $I_{AREF} = Q_{CONV} / t_c$.
The fastest 12-bit post-calibrated conversion of t_c = 459 ns results in a typical average current of I_{AREF} = 65.4 µA.

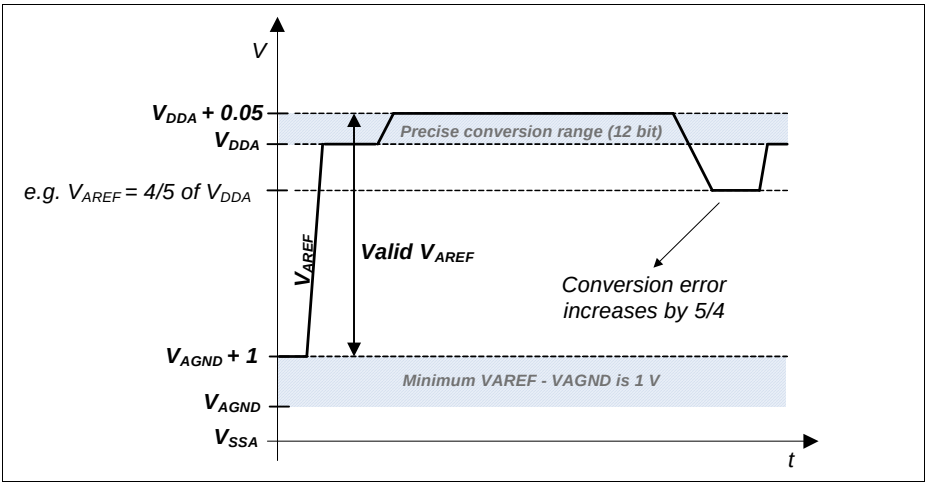


Figure 14 VADC Reference Voltage Range

The power-up calibration of the VADC requires a maximum number of 4 352 f_{ADCI} cycles.

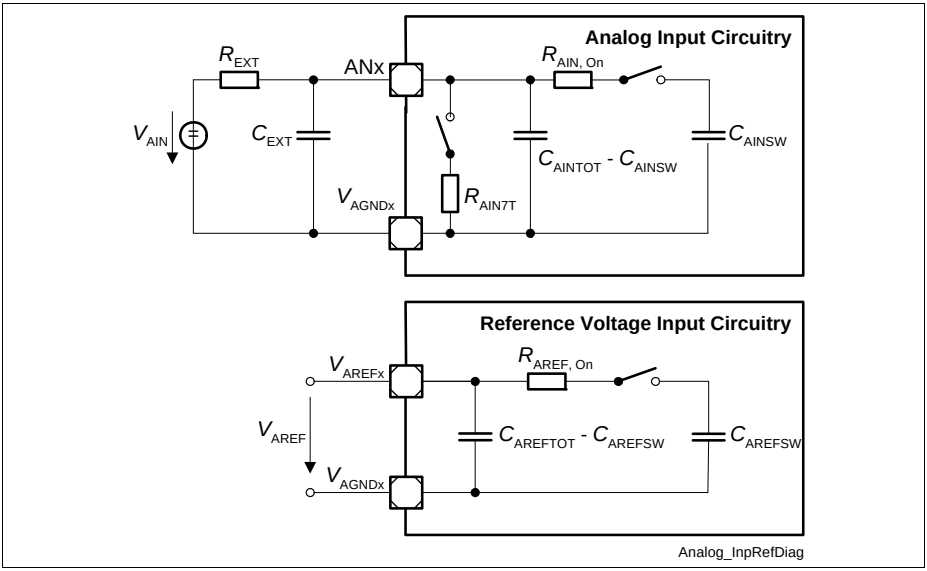


Figure 15 VADC Input Circuits

Table 31 USB OTG Data Line (USB_DP, USB_DM) Parameters (Operating Conditions apply)

Parameter	Symbol		Values			Unit	Note / Test Condition
			Min.	Typ.	Max.		
Input low voltage	V_{IL}	SR	–	–	0.8	V	
Input high voltage (driven)	V_{IH}	SR	2.0	–	–	V	
Input high voltage (floating) [1]	V_{IHZ}	SR	2.7	–	3.6	V	
Differential input sensitivity	V_{DIS}	CC	0.2	–	–	V	
Differential common mode range	V_{CM}	CC	0.8	–	2.5	V	
Output low voltage	V_{OL}	CC	0.0	–	0.3	V	1.5 kOhm pull-up to 3.6 V
Output high voltage	V_{OH}	CC	2.8	–	3.6	V	15 kOhm pull-down to 0 V
DP pull-up resistor (idle bus)	R_{PUI}	CC	900	–	1 575	Ohm	
DP pull-up resistor (upstream port receiving)	R_{PUA}	CC	1 425	–	3 090	Ohm	
DP, DM pull-down resistor	R_{PD}	CC	14.25	–	24.8	kOhm	
Input impedance DP, DM	Z_{INP}	CC	300	–	–	kOhm	$0\ V \le V_{IN} \le V_{DDP}$
Driver output resistance DP, DM	Z_{DRV}	CC	28	–	44	Ohm	

1) Measured at A-connector with 1.5 kOhm ± 5% to 3.3 V ± 0.3 V connected to USB_DP or USB_DM and at B-connector with 15 kOhm ± 5% to ground connected to USB_DP and USB_DM.

3.3.2 Power-Up and Supply Monitoring

$\overline{\text{PORST}}$ is always asserted when V_{DDP} and/or V_{DDC} violate the respective thresholds.

Note: These parameters are not subject to production test, but verified by design and/or characterization.

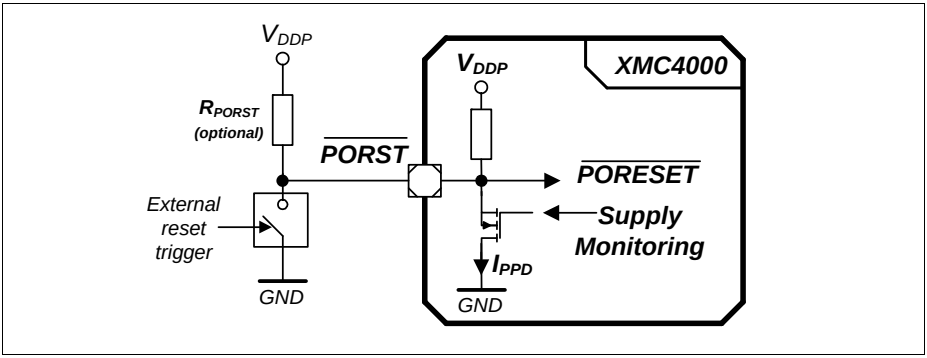


Figure 25 $\overline{\text{PORST}}$ Circuit

Table 37 Supply Monitoring Parameters

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Digital supply voltage reset threshold	V_{POR} CC	2.79[1]	–	3.05[2]	V	[3]
Core supply voltage reset threshold	V_{PV} CC	–	–	1.17	V	
V_{DDP} voltage to ensure defined pad states	V_{DDPPA} CC	–	1.0	–	V	
$\overline{\text{PORST}}$ rise time	t_{PR} SR	–	–	2	µs	[4]
Startup time from power-on reset with code execution from Flash	t_{SSW} CC	–	2.5	3.5	ms	Time to the first user code instruction
V_{DDC} ramp up time	t_{VCR} CC	–	550	–	µs	Ramp up after power-on or after a reset triggered by a violation of V_{POR} or V_{PV}

1) Minimum threshold for reset assertion.

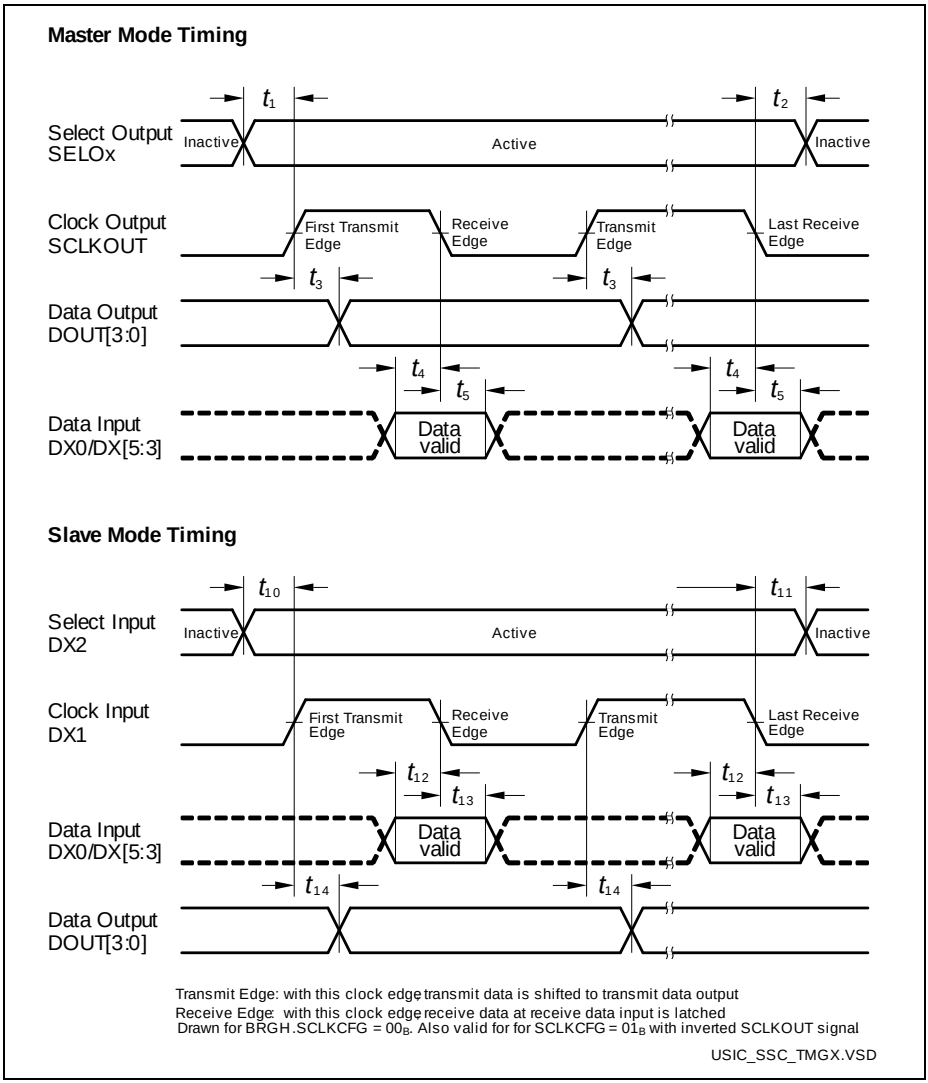


Figure 33 USIC - SSC Master/Slave Mode Timing

Note: This timing diagram shows a standard configuration, for which the slave select signal is low-active, and the serial clock signal is not shifted and not inverted.

3.3.9.3 Inter-IC (IIC) Interface Timing

The following parameters are applicable for a USIC channel operated in IIC mode.

Note: These parameters are not subject to production test, but verified by design and/or characterization.

Table 48 USIC IIC Standard Mode Timing[1)]

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Fall time of both SDA and SCL	t_1 CC/SR	-	-	300	ns	
Rise time of both SDA and SCL	t_2 CC/SR	-	-	1000	ns	
Data hold time	t_3 CC/SR	0	-	-	µs	
Data set-up time	t_4 CC/SR	250	-	-	ns	
LOW period of SCL clock	t_5 CC/SR	4.7	-	-	µs	
HIGH period of SCL clock	t_6 CC/SR	4.0	-	-	µs	
Hold time for (repeated) START condition	t_7 CC/SR	4.0	-	-	µs	
Set-up time for repeated START condition	t_8 CC/SR	4.7	-	-	µs	
Set-up time for STOP condition	t_9 CC/SR	4.0	-	-	µs	
Bus free time between a STOP and START condition	t_{10} CC/SR	4.7	-	-	µs	
Capacitive load for each bus line	C_b SR	-	-	400	pF	

1) Due to the wired-AND configuration of an IIC bus system, the port drivers of the SCL and SDA signal lines need to operate in open-drain mode. The high level on these lines must be held by an external pull-up device, approximalely 10 kOhm for operation at 100 kbit/s, approximately 2 kOhm for operation at 400 kbit/s.

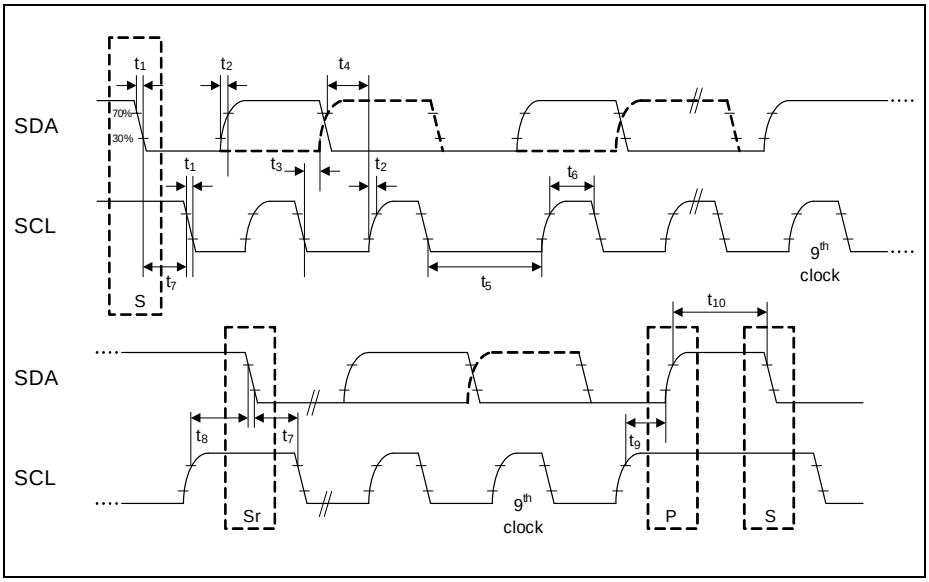


Figure 34 USIC IIC Stand and Fast Mode Timing

3.3.9.4 Inter-IC Sound (IIS) Interface Timing

The following parameters are applicable for a USIC channel operated in IIS mode.

Note: These parameters are not subject to production test, but verified by design and/or characterization.

Table 50 USIC IIS Master Transmitter Timing

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Clock period	t_1 CC	33.3	–	–	ns	
Clock high time	t_2 CC	0.35 x t_{1min}	–	–	ns	
Clock low time	t_3 CC	0.35 x t_{1min}	–	–	ns	
Hold time	t_4 CC	0	–	–	ns	
Clock rise time	t_5 CC	–	–	0.15 x t_{1min}	ns	

3.3.9.5 SDMMC Interface Timing

Note: These parameters are not subject to production test, but verified by design and/or characterization.

Note: Operating Conditions apply, total external capacitive load C_L = 40 pF.

AC Timing Specifications (Full-Speed Mode)

Table 52 SDMMC Timing for Full-Speed Mode

Parameter	Symbol		Values		Unit	Note/ Test Condition
			Min.	Max.		
Clock frequency in full speed transfer mode ($1/t_{pp}$)	f_{pp}	CC	0	24	MHz	
Clock cycle in full speed transfer mode	t_{pp}	CC	40	–	ns	
Clock low time	t_{WL}	CC	10	–	ns	
Clock high time	t_{WH}	CC	10	–	ns	
Clock rise time	t_{TLH}	CC	–	10	ns	
Clock fall time	t_{THL}	CC	–	10	ns	
Inputs setup to clock rising edge	t_{ISU_F}	SR	2	–	ns	
Inputs hold after clock rising edge	t_{IH_F}	SR	2	–	ns	
Outputs valid time in full speed mode	t_{ODLY_F}	CC	–	10	ns	
Outputs hold time in full speed mode	t_{OH_F}	CC	0	–	ns	

Table 53 SD Card Bus Timing for Full-Speed Mode[1)]

Parameter	Symbol	Values		Unit	Note/ Test Condition
		Min.	Max.		
SD card input setup time	t_{ISU}	5	–	ns	
SD card input hold time	t_{IH}	5	–	ns	

Table 55 SD Card Bus Timing for High-Speed Mode[1)]

Parameter	Symbol	Values		Unit	Note/ Test Condition
		Min.	Max.		
SD card input setup time	t_{ISU}	6	–	ns	
SD card input hold time	t_{IH}	2	–	ns	
SD card output valid time	t_{ODLY}	–	14	ns	
SD card output hold time	t_{OH}	2.5	–	ns	

1) Reference card timing values for calculation examples. Not subject to production test and not characterized.

High-Speed Output Path (Write)

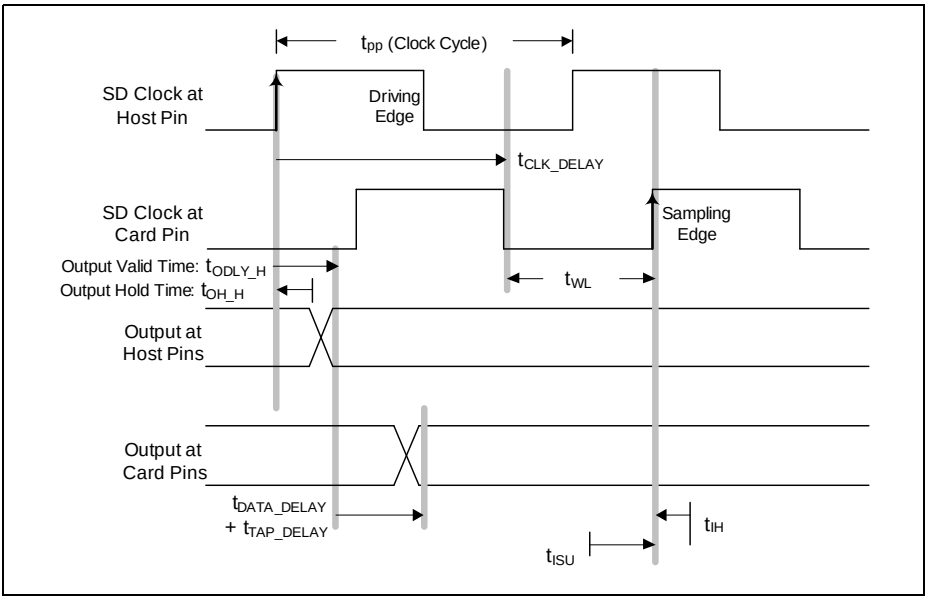


Figure 39 High-Speed Output Path

High-Speed Write Meeting Setup (Maximum Delay)

The following equations show how to calculate the allowed skew range between the SD_CLK and SD_DAT/CMD signals on the PCB.

Table 62 EBU SDRAM Access Signal Timing Parameters

Parameter			Symbol	Limit Values		Unit
				Min.	Max.	
A(15:0) output valid	from SDCLKO low-to-high transition	CC	t_6	–	9	ns
A(15:0) output hold		CC	t_7	3	–	
$\overline{\text{CS}}$(3:0) low		CC	t_8	–	9	
$\overline{\text{CS}}$(3:0) high		CC	t_9	3	–	
$\overline{\text{RAS}}$ low		CC	t_{10}	–	9	
$\overline{\text{RAS}}$ high		SR	t_{11}	3	–	
$\overline{\text{CAS}}$ low		SR	t_{12}	–	9	
$\overline{\text{CAS}}$ high		CC	t_{13}	3	–	
RD/$\overline{\text{WR}}$ low		CC	t_{14}	–	9	
RD/$\overline{\text{WR}}$ high		CC	t_{15}	3	–	
$\overline{\text{BC}}$(3:0) low		CC	t_{16}	–	9	
$\overline{\text{BC}}$(3:0) high		CC	t_{17}	3	–	
D(15:0) output valid		CC	t_{18}	–	9	
D(15:0) output hold		CC	t_{19}	3	–	
CKE output valid[1)]		CC	t_{22}	–	7	
CKE output hold[1)]		CC	t_{23}	2	–	
D(15:0) input hold		SR	t_{21}	3	–	
D(15:0) input setup to SDCLKO low-to-high transition		SR	t_{20}	4	–	

1) Not depicted in the read and write access timing figures below.

3.3.11 USB Interface Characteristics

The Universal Serial Bus (USB) Interface is compliant to the USB Rev. 2.0 Specification and the OTG Specification Rev. 1.3. High-Speed Mode is not supported.

Note: These parameters are not subject to production test, but verified by design and/or characterization.

Table 63 USB Timing Parameters (operating conditions apply)

Parameter	Symbol		Values			Unit	Note / Test Condition
			Min.	Typ.	Max.		
Rise time	t_R	CC	4	–	20	ns	C_L = 50 pF
Fall time	t_F	CC	4	–	20	ns	C_L = 50 pF
Rise/Fall time matching	t_R/t_F	CC	90	–	111.11	%	C_L = 50 pF
Crossover voltage	V_{CRS}	CC	1.3	–	2.0	V	C_L = 50 pF

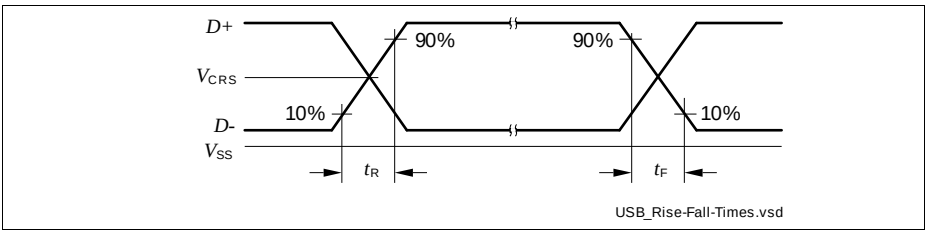


Figure 50 USB Signal Timing

3.3.12.4 ETH RMII Parameters

In the following, the parameters of the RMII (Reduced Media Independent Interface) are described.

Table 66 ETH RMII Signal Timing Parameters

Parameter	Symbol		Values			Unit	Note / Test Condition
			Min.	Typ.	Max.		
ETH_RMII_REF_CL clock period	t_{13}	SR	20	–	–	ns	C_L = 25 pF; 50 ppm
ETH_RMII_REF_CL clock high time	t_{14}	SR	7	–	13	ns	C_L = 25 pF
ETH_RMII_REF_CL clock low time	t_{15}	SR	7	–	13	ns	
ETH_RMII_RXD[1:0], ETH_RMII_CRS setup time	t_{16}	SR	4	–	–	ns	
ETH_RMII_RXD[1:0], ETH_RMII_CRS hold time	t_{17}	SR	2	–	–	ns	
ETH_RMII_TXD[1:0], ETH_RMII_TXEN data valid	t_{18}	CC	4	–	15	ns	

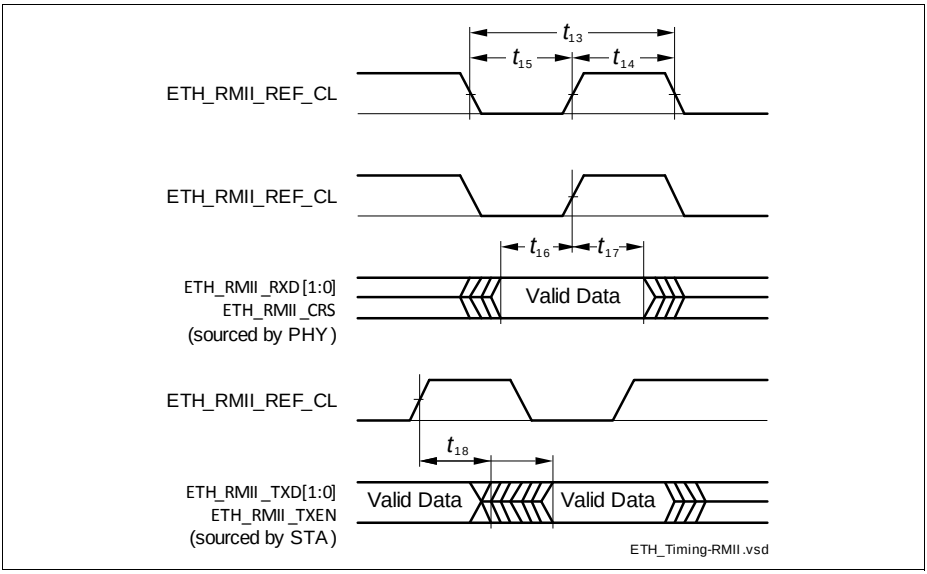


Figure 54 ETH RMII Signal Timing