

Welcome to **E-XFL.COM**

Understanding <u>Embedded - FPGAs (Field Programmable Gate Array)</u>

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Obsolete
Number of LABs/CLBs	832
Number of Logic Elements/Cells	8320
Total RAM Bits	106496
Number of I/O	382
Number of Gates	526000
Voltage - Supply	2.375V ~ 2.625V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	484-BBGA
Supplier Device Package	484-FBGA (23x23)
Purchase URL	https://www.e-xfl.com/product-detail/intel/ep20k200bc484-2x

Email: info@E-XFL.COM

Address: Room A, 16/F, Full Win Commercial Centre, 573 Nathan Road, Mongkok, Hong Kong

General Description

APEXTM 20K devices are the first PLDs designed with the MultiCore architecture, which combines the strengths of LUT-based and product-term-based devices with an enhanced memory structure. LUT-based logic provides optimized performance and efficiency for data-path, register-intensive, mathematical, or digital signal processing (DSP) designs. Product-term-based logic is optimized for complex combinatorial paths, such as complex state machines. LUT- and product-term-based logic combined with memory functions and a wide variety of MegaCore and AMPP functions make the APEX 20K device architecture uniquely suited for system-on-a-programmable-chip designs. Applications historically requiring a combination of LUT-, product-term-, and memory-based devices can now be integrated into one APEX 20K device.

APEX 20KE devices are a superset of APEX 20K devices and include additional features such as advanced I/O standard support, CAM, additional global clocks, and enhanced ClockLock clock circuitry. In addition, APEX 20KE devices extend the APEX 20K family to 1.5 million gates. APEX 20KE devices are denoted with an "E" suffix in the device name (e.g., the EP20K1000E device is an APEX 20KE device). Table 8 compares the features included in APEX 20K and APEX 20KE devices.

All APEX 20K devices are reconfigurable and are 100% tested prior to shipment. As a result, test vectors do not have to be generated for fault coverage purposes. Instead, the designer can focus on simulation and design verification. In addition, the designer does not need to manage inventories of different application-specific integrated circuit (ASIC) designs; APEX 20K devices can be configured on the board for the specific functionality required.

APEX 20K devices are configured at system power-up with data stored in an Altera serial configuration device or provided by a system controller. Altera offers in-system programmability (ISP)-capable EPC1, EPC2, and EPC16 configuration devices, which configure APEX 20K devices via a serial data stream. Moreover, APEX 20K devices contain an optimized interface that permits microprocessors to configure APEX 20K devices serially or in parallel, and synchronously or asynchronously. The interface also enables microprocessors to treat APEX 20K devices as memory and configure the device by writing to a virtual memory location, making reconfiguration easy.

After an APEX 20K device has been configured, it can be reconfigured in-circuit by resetting the device and loading new data. Real-time changes can be made during system operation, enabling innovative reconfigurable computing applications.

APEX 20K devices are supported by the Altera Quartus II development system, a single, integrated package that offers HDL and schematic design entry, compilation and logic synthesis, full simulation and worst-case timing analysis, SignalTap logic analysis, and device configuration. The Quartus II software runs on Windows-based PCs, Sun SPARCstations, and HP 9000 Series 700/800 workstations.

The Quartus II software provides NativeLink interfaces to other industry-standard PC- and UNIX workstation-based EDA tools. For example, designers can invoke the Quartus II software from within third-party design tools. Further, the Quartus II software contains built-in optimized synthesis libraries; synthesis tools can use these libraries to optimize designs for APEX 20K devices. For example, the Synopsys Design Compiler library, supplied with the Quartus II development system, includes DesignWare functions optimized for the APEX 20K architecture.

Each LAB contains dedicated logic for driving control signals to its LEs and ESBs. The control signals include clock, clock enable, asynchronous clear, asynchronous preset, asynchronous load, synchronous clear, and synchronous load signals. A maximum of six control signals can be used at a time. Although synchronous load and clear signals are generally used when implementing counters, they can also be used with other functions.

Each LAB can use two clocks and two clock enable signals. Each LAB's clock and clock enable signals are linked (e.g., any LE in a particular LAB using CLK1 will also use CLKENA1). LEs with the same clock but different clock enable signals either use both clock signals in one LAB or are placed into separate LABs.

If both the rising and falling edges of a clock are used in a LAB, both LAB-wide clock signals are used.

The LAB-wide control signals can be generated from the LAB local interconnect, global signals, and dedicated clock pins. The inherent low skew of the FastTrack Interconnect enables it to be used for clock distribution. Figure 4 shows the LAB control signal generation circuit.

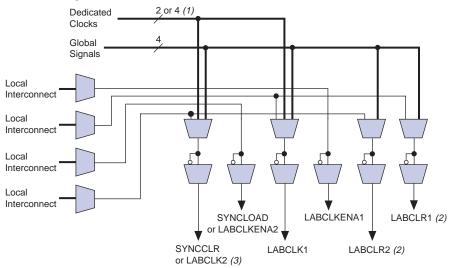


Figure 4. LAB Control Signal Generation

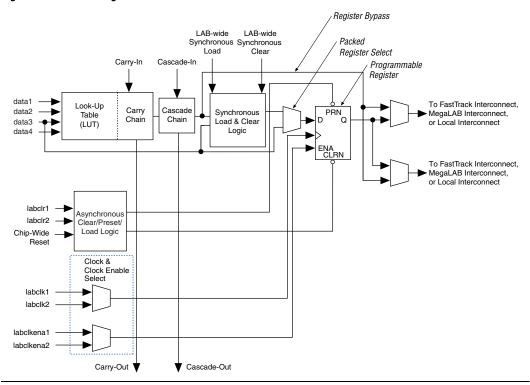
Notes to Figure 4:

- (1) APEX 20KE devices have four dedicated clocks.
- (2) The LABCLR1 and LABCLR2 signals also control asynchronous load and asynchronous preset for LEs within the LAB.
- (3) The SYNCCLR signal can be generated by the local interconnect or global signals.

Logic Element

The LE, the smallest unit of logic in the APEX 20K architecture, is compact and provides efficient logic usage. Each LE contains a four-input LUT, which is a function generator that can quickly implement any function of four variables. In addition, each LE contains a programmable register and carry and cascade chains. Each LE drives the local interconnect, MegaLAB interconnect, and FastTrack Interconnect routing structures. See Figure 5.

Figure 5. APEX 20K Logic Element



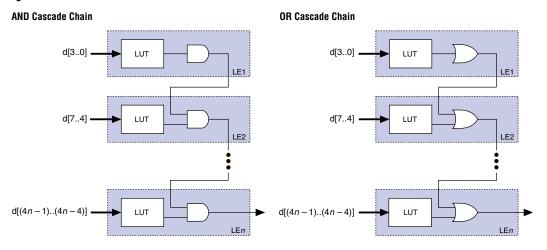
Each LE's programmable register can be configured for D, T, JK, or SR operation. The register's clock and clear control signals can be driven by global signals, general-purpose I/O pins, or any internal logic. For combinatorial functions, the register is bypassed and the output of the LUT drives the outputs of the LE.

Cascade Chain

With the cascade chain, the APEX 20K architecture can implement functions with a very wide fan-in. Adjacent LUTs can compute portions of a function in parallel; the cascade chain serially connects the intermediate values. The cascade chain can use a logical AND or logical OR (via De Morgan's inversion) to connect the outputs of adjacent LEs. Each additional LE provides four more inputs to the effective width of a function, with a short cascade delay. Cascade chain logic can be created automatically by the Quartus II software Compiler during design processing, or manually by the designer during design entry.

Cascade chains longer than ten LEs are implemented automatically by linking LABs together. For enhanced fitting, a long cascade chain skips alternate LABs in a MegaLAB structure. A cascade chain longer than one LAB skips either from an even-numbered LAB to the next even-numbered LAB, or from an odd-numbered LAB to the next odd-numbered LAB. For example, the last LE of the first LAB in the upper-left MegaLAB structure carries to the first LE of the third LAB in the MegaLAB structure. Figure 7 shows how the cascade function can connect adjacent LEs to form functions with a wide fan-in.

Figure 7. APEX 20K Cascade Chain



The counter mode uses two three-input LUTs: one generates the counter data, and the other generates the fast carry bit. A 2-to-1 multiplexer provides synchronous loading, and another AND gate provides synchronous clearing. If the cascade function is used by an LE in counter mode, the synchronous clear or load overrides any signal carried on the cascade chain. The synchronous clear overrides the synchronous load. LEs in arithmetic mode can drive out registered and unregistered versions of the LUT output.

Clear & Preset Logic Control

Logic for the register's clear and preset signals is controlled by LAB-wide signals. The LE directly supports an asynchronous clear function. The Quartus II software Compiler can use a NoT-gate push-back technique to emulate an asynchronous preset. Moreover, the Quartus II software Compiler can use a programmable NoT-gate push-back technique to emulate simultaneous preset and clear or asynchronous load. However, this technique uses three additional LEs per register. All emulation is performed automatically when the design is compiled. Registers that emulate simultaneous preset and load will enter an unknown state upon power-up or when the chip-wide reset is asserted.

In addition to the two clear and preset modes, APEX 20K devices provide a chip-wide reset pin (DEV_CLRn) that resets all registers in the device. Use of this pin is controlled through an option in the Quartus II software that is set before compilation. The chip-wide reset overrides all other control signals. Registers using an asynchronous preset are preset when the chip-wide reset is asserted; this effect results from the inversion technique used to implement the asynchronous preset.

FastTrack Interconnect

In the APEX 20K architecture, connections between LEs, ESBs, and I/O pins are provided by the FastTrack Interconnect. The FastTrack Interconnect is a series of continuous horizontal and vertical routing channels that traverse the device. This global routing structure provides predictable performance, even in complex designs. In contrast, the segmented routing in FPGAs requires switch matrices to connect a variable number of routing paths, increasing the delays between logic resources and reducing performance.

The FastTrack Interconnect consists of row and column interconnect channels that span the entire device. The row interconnect routes signals throughout a row of MegaLAB structures; the column interconnect routes signals throughout a column of MegaLAB structures. When using the row and column interconnect, an LE, IOE, or ESB can drive any other LE, IOE, or ESB in a device. See Figure 9.

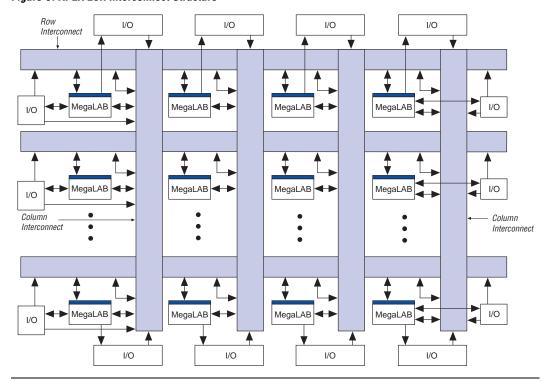


Figure 9. APEX 20K Interconnect Structure

A row line can be driven directly by LEs, IOEs, or ESBs in that row. Further, a column line can drive a row line, allowing an LE, IOE, or ESB to drive elements in a different row via the column and row interconnect. The row interconnect drives the MegaLAB interconnect to drive LEs, IOEs, or ESBs in a particular MegaLAB structure.

A column line can be directly driven by LEs, IOEs, or ESBs in that column. A column line on a device's left or right edge can also be driven by row IOEs. The column line is used to route signals from one row to another. A column line can drive a row line; it can also drive the MegaLAB interconnect directly, allowing faster connections between rows.

Figure 10 shows how the FastTrack Interconnect uses the local interconnect to drive LEs within MegaLAB structures.

ESBs can implement synchronous RAM, which is easier to use than asynchronous RAM. A circuit using asynchronous RAM must generate the RAM write enable (WE) signal, while ensuring that its data and address signals meet setup and hold time specifications relative to the WE signal. In contrast, the ESB's synchronous RAM generates its own WE signal and is self-timed with respect to the global clock. Circuits using the ESB's self-timed RAM must only meet the setup and hold time specifications of the global clock.

ESB inputs are driven by the adjacent local interconnect, which in turn can be driven by the MegaLAB or FastTrack Interconnect. Because the ESB can be driven by the local interconnect, an adjacent LE can drive it directly for fast memory access. ESB outputs drive the MegaLAB and FastTrack Interconnect. In addition, ten ESB outputs, nine of which are unique output lines, drive the local interconnect for fast connection to adjacent LEs or for fast feedback product-term logic.

When implementing memory, each ESB can be configured in any of the following sizes: 128×16 , 256×8 , 512×4 , $1,024 \times 2$, or $2,048 \times 1$. By combining multiple ESBs, the Quartus II software implements larger memory blocks automatically. For example, two 128×16 RAM blocks can be combined to form a 128×32 RAM block, and two 512×4 RAM blocks can be combined to form a 512×8 RAM block. Memory performance does not degrade for memory blocks up to 2,048 words deep. Each ESB can implement a 2,048-word-deep memory; the ESBs are used in parallel, eliminating the need for any external control logic and its associated delays.

To create a high-speed memory block that is more than 2,048 words deep, ESBs drive tri-state lines. Each tri-state line connects all ESBs in a column of MegaLAB structures, and drives the MegaLAB interconnect and row and column FastTrack Interconnect throughout the column. Each ESB incorporates a programmable decoder to activate the tri-state driver appropriately. For instance, to implement 8,192-word-deep memory, four ESBs are used. Eleven address lines drive the ESB memory, and two more drive the tri-state decoder. Depending on which 2,048-word memory page is selected, the appropriate ESB driver is turned on, driving the output to the tri-state line. The Quartus II software automatically combines ESBs with tri-state lines to form deeper memory blocks. The internal tri-state control logic is designed to avoid internal contention and floating lines. See Figure 18.

Clock Phase & Delay Adjustment

The APEX 20KE ClockShift feature allows the clock phase and delay to be adjusted. The clock phase can be adjusted by 90° steps. The clock delay can be adjusted to increase or decrease the clock delay by an arbitrary amount, up to one clock period.

LVDS Support

Two PLLs are designed to support the LVDS interface. When using LVDS, the I/O clock runs at a slower rate than the data transfer rate. Thus, PLLs are used to multiply the I/O clock internally to capture the LVDS data. For example, an I/O clock may run at 105 MHz to support 840 megabits per second (Mbps) LVDS data transfer. In this example, the PLL multiplies the incoming clock by eight to support the high-speed data transfer. You can use PLLs in EP20K400E and larger devices for high-speed LVDS interfacing.

Lock Signals

The APEX 20KE ClockLock circuitry supports individual LOCK signals. The LOCK signal drives high when the ClockLock circuit has locked onto the input clock. The LOCK signals are optional for each ClockLock circuit; when not used, they are I/O pins.

ClockLock & ClockBoost Timing Parameters

For the ClockLock and ClockBoost circuitry to function properly, the incoming clock must meet certain requirements. If these specifications are not met, the circuitry may not lock onto the incoming clock, which generates an erroneous clock within the device. The clock generated by the ClockLock and ClockBoost circuitry must also meet certain specifications. If the incoming clock meets these requirements during configuration, the APEX 20K ClockLock and ClockBoost circuitry will lock onto the clock during configuration. The circuit will be ready for use immediately after configuration. In APEX 20KE devices, the clock input standard is programmable, so the PLL cannot respond to the clock until the device is configured. The PLL locks onto the input clock as soon as configuration is complete. Figure 30 shows the incoming and generated clock specifications.



For more information on ClockLock and ClockBoost circuitry, see *Application Note 115: Using the ClockLock and ClockBoost PLL Features in APEX Devices*.

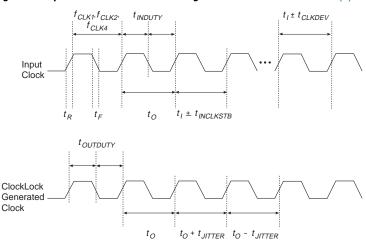


Figure 30. Specifications for the Incoming & Generated Clocks Note (1)

Note to Figure 30:

(1) The tl parameter refers to the nominal input clock period; the tO parameter refers to the nominal output clock period.

Table 15 summarizes the APEX 20K ClockLock and ClockBoost parameters for -1 speed-grade devices.

Symbol	Parameter	Min	Max	Unit	
f _{OUT}	Output frequency	25	180	MHz	
f _{CLK1} (1)	Input clock frequency (ClockBoost clock multiplication factor equals 1)	25	180 (1)	MHz	
f _{CLK2}	Input clock frequency (ClockBoost clock multiplication factor equals 2)	16	90	MHz	
f _{CLK4}	Input clock frequency (ClockBoost clock multiplication factor equals 4)	10	48	MHz	
^t OUTDUTY	Duty cycle for ClockLock/ClockBoost-generated clock	40	60	%	
f _{CLKDEV}	Input deviation from user specification in the Quartus II software (ClockBoost clock multiplication factor equals 1) (2)		25,000 (3)	PPM	
t _R	Input rise time		5	ns	
t _F	Input fall time		5	ns	
t _{LOCK}	Time required for ClockLock/ClockBoost to acquire lock (4)		10	μs	



For DC Operating Specifications on APEX 20KE I/O standards, please refer to *Application Note 117 (Using Selectable I/O Standards in Altera Devices).*

Table 30. APEX 20KE Device Capacitance Note (15)								
Symbol	Parameter	Conditions	Min	Max	Unit			
C _{IN}	Input capacitance	V _{IN} = 0 V, f = 1.0 MHz		8	pF			
C _{INCLK}	Input capacitance on dedicated clock pin	V _{IN} = 0 V, f = 1.0 MHz		12	pF			
C _{OUT}	Output capacitance	V _{OUT} = 0 V, f = 1.0 MHz		8	pF			

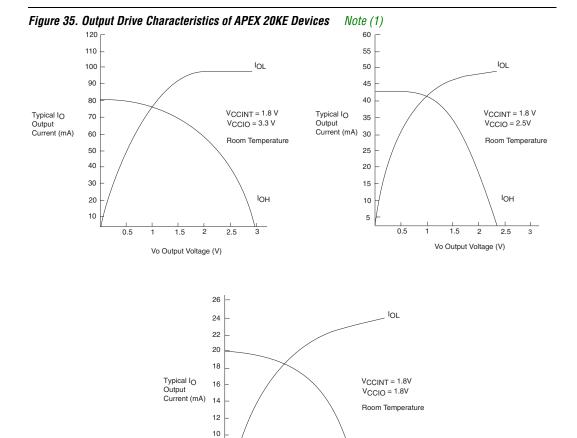
Notes to Tables 27 through 30:

- (1) See the Operating Requirements for Altera Devices Data Sheet.
- (2) Minimum DC input is -0.5 V. During transitions, the inputs may undershoot to -2.0 V or overshoot to 5.75 V for input currents less than 100 mA and periods shorter than 20 ns.
- (3) Numbers in parentheses are for industrial-temperature-range devices.
- (4) Maximum V_{CC} rise time is 100 ms, and V_{CC} must rise monotonically.
- (5) Minimum DC input is -0.5 V. During transitions, the inputs may undershoot to -2.0 V or overshoot to the voltage shown in the following table based on input duty cycle for input currents less than 100 mA. The overshoot is dependent upon duty cycle of the signal. The DC case is equivalent to 100% duty cycle.

Vin Max. Duty Cycle 4.0V 100% (DC) 4.1 90% 4.2 50% 4.3 30% 4.4 17% 4.5 10%

- (6) All pins, including dedicated inputs, clock, I/O, and JTAG pins, may be driven before V_{CCINT} and V_{CCIO} are powered.
- Typical values are for $T_A = 25^{\circ}$ C, $V_{CCINT} = 1.8$ V, and $V_{CCIO} = 1.8$ V, 2.5 V or 3.3 V.
- (8) These values are specified under the APEX 20KE device recommended operating conditions, shown in Table 24 on page 60.
- (9) Refer to Application Note 117 (Using Selectable I/O Standards in Altera Devices) for the V_{IH}, V_{IL}, V_{OH}, V_{OL}, and I_I parameters when VCCIO = 1.8 V.
- (10) The APEX 20KE input buffers are compatible with 1.8-V, 2.5-V and 3.3-V (LVTTL and LVCMOS) signals. Additionally, the input buffers are 3.3-V PCI compliant. Input buffers also meet specifications for GTL+, CTT, AGP, SSTL-2, SSTL-3, and HSTL.
- (11) The I_{OH} parameter refers to high-level TTL, PCI, or CMOS output current.
- (12) The I_{OL} parameter refers to low-level TTL, PCI, or CMOS output current. This parameter applies to open-drain pins as well as output pins.
- (13) This value is specified for normal device operation. The value may vary during power-up.
- (14) Pin pull-up resistance values will be lower if an external source drives the pin higher than V_{CCIO}.
- (15) Capacitance is sample-tested only.

Figure 33 shows the relationship between V_{CCIO} and V_{CCINT} for 3.3-V PCI compliance on APEX 20K devices.



8

4 2

0.5

Figure 35 shows the output drive characteristics of APEX 20KE devices.

Note to Figure 35:

(1) These are transient (AC) currents.

Timing Model

The high-performance FastTrack and MegaLAB interconnect routing resources ensure predictable performance, accurate simulation, and accurate timing analysis. This predictable performance contrasts with that of FPGAs, which use a segmented connection scheme and therefore have unpredictable performance.

Vo Output Voltage (V)

IOH

2.0

Table 31. APEX 20K f _{MAX} Timing Parameters (Part 2 of 2)							
Symbol	Parameter						
t _{ESBDATACO2}	ESB clock-to-output delay without output registers						
t _{ESBDD}	ESB data-in to data-out delay for RAM mode						
t _{PD}	ESB macrocell input to non-registered output						
t _{PTERMSU}	ESB macrocell register setup time before clock						
t _{PTERMCO}	ESB macrocell register clock-to-output delay	-					
t _{F1-4}	Fanout delay using local interconnect						
t _{F5-20}	Fanout delay using MegaLab Interconnect						
t _{F20+}	Fanout delay using FastTrack Interconnect						
t _{CH}	Minimum clock high time from clock pin						
t _{CL}	Minimum clock low time from clock pin						
t _{CLRP}	LE clear pulse width						
t _{PREP}	LE preset pulse width						
t _{ESBCH}	Clock high time						
t _{ESBCL}	Clock low time						
t _{ESBWP}	Write pulse width						
t _{ESBRP}	Read pulse width						

Tables 32 and 33 describe APEX 20K external timing parameters.

Table 32. APEX 20K External Timing Parameters Note (1)						
Symbol	Clock Parameter					
t _{INSU}	Setup time with global clock at IOE register					
t _{INH}	Hold time with global clock at IOE register					
tоитсо	Clock-to-output delay with global clock at IOE register					

Table 33. APEX 20K External Bidirectional Timing Parameters Note (1)							
Symbol	Parameter	Conditions					
t _{INSUBIDIR}	Setup time for bidirectional pins with global clock at same-row or same-column LE register						
t _{INHBIDIR}	Hold time for bidirectional pins with global clock at same-row or same-column LE register						
^t OUTCOBIDIR	Clock-to-output delay for bidirectional pins with global clock at IOE register	C1 = 10 pF					
t _{XZBIDIR}	Synchronous IOE output buffer disable delay	C1 = 10 pF					
t _{ZXBIDIR}	Synchronous IOE output buffer enable delay, slow slew rate = off	C1 = 10 pF					

Table 43. EP20K100 External Timing Parameters									
Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit		
	Min	Max	Min	Max	Min	Max			
t _{INSU} (1)	2.3		2.8		3.2		ns		
t _{INH} (1)	0.0		0.0		0.0		ns		
t _{OUTCO} (1)	2.0	4.5	2.0	4.9	2.0	6.6	ns		
t _{INSU} (2)	1.1		1.2		-		ns		
t _{INH} (2)	0.0		0.0		-		ns		
t _{OUTCO} (2)	0.5	2.7	0.5	3.1	_	4.8	ns		

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	1
t _{INSUBIDIR} (1)	2.3		2.8		3.2		ns
t _{INHBIDIR} (1)	0.0		0.0		0.0		ns
toutcobidir (1)	2.0	4.5	2.0	4.9	2.0	6.6	ns
t _{XZBIDIR} (1)		5.0		5.9		6.9	ns
t _{ZXBIDIR} (1)		5.0		5.9		6.9	ns
t _{INSUBIDIR} (2)	1.0		1.2		-		ns
t _{INHBIDIR} (2)	0.0		0.0		-		ns
toutcobidir (2)	0.5	2.7	0.5	3.1	-	-	ns
t _{XZBIDIR} (2)		4.3		5.0		_	ns
t _{ZXBIDIR} (2)		4.3		5.0		_	ns

Table 45. EP20K200 External Timing Parameters										
Symbol	-1 Speed Grade		-2 Spe	-2 Speed Grade		-3 Speed Grade				
	Min	Max	Min	Max	Min	Max				
t _{INSU} (1)	1.9		2.3		2.6		ns			
t _{INH} (1)	0.0		0.0		0.0		ns			
t _{OUTCO} (1)	2.0	4.6	2.0	5.6	2.0	6.8	ns			
t _{INSU} (2)	1.1		1.2		-		ns			
t _{INH} (2)	0.0		0.0		-		ns			
t _{оитсо} <i>(2)</i>	0.5	2.7	0.5	3.1	-	_	ns			

Symbol	-1		-2		-3		Unit
	Min	Max	Min	Max	Min	Max	1
t _{CH}	0.55		0.78		1.15		ns
t _{CL}	0.55		0.78		1.15		ns
t _{CLRP}	0.22		0.31		0.46		ns
t _{PREP}	0.22		0.31		0.46		ns
t _{ESBCH}	0.55		0.78		1.15		ns
t _{ESBCL}	0.55		0.78		1.15		ns
t _{ESBWP}	1.43		2.01		2.97		ns
t _{ESBRP}	1.15		1.62		2.39		ns

Symbol	-1		-2		-3		Unit
	Min	Max	Min	Max	Min	Max	7
t _{INSU}	2.02		2.13		2.24		ns
t _{INH}	0.00		0.00		0.00		ns
t _{outco}	2.00	4.88	2.00	5.36	2.00	5.88	ns
t _{INSUPLL}	2.11		2.23		=		ns
t _{INHPLL}	0.00		0.00		=		ns
t _{OUTCOPLL}	0.50	2.60	0.50	2.88	-	-	ns

Symbol	-1		-2		-3		Unit
	Min	Max	Min	Max	Min	Max	
t _{INSUBIDIR}	1.85		1.77		1.54		ns
t _{INHBIDIR}	0.00		0.00		0.00		ns
toutcobidir	2.00	4.88	2.00	5.36	2.00	5.88	ns
t _{XZBIDIR}		7.48		8.46		9.83	ns
t _{ZXBIDIR}		7.48		8.46		9.83	ns
t _{INSUBIDIRPLL}	4.12		4.24		-		ns
t _{INHBIDIRPLL}	0.00		0.00		-		ns
t _{OUTCOBIDIRPLL}	0.50	2.60	0.50	2.88	-	-	ns
t _{XZBIDIRPLL}		5.21		5.99		-	ns
tzxbidirpll		5.21		5.99		-	ns

Tables 67 through 72 describe f_{MAX} LE Timing Microparameters, f_{MAX} ESB Timing Microparameters, f_{MAX} Routing Delays, Minimum Pulse Width Timing Parameters, External Timing Parameters, and External Bidirectional Timing Parameters for EP20K160E APEX 20KE devices.

Table 67. EP20K160E f _{MAX} LE Timing Microparameters										
Symbol	_	1		-2	-3		Unit			
	Min	Max	Min	Max	Min	Max				
t _{SU}	0.22		0.24		0.26		ns			
t _H	0.22		0.24		0.26		ns			
t _{CO}		0.25		0.31		0.35	ns			
t _{LUT}		0.69		0.88		1.12	ns			

Symbol	-1		-2		-3		Unit
	Min	Max	Min	Max	Min	Max	
t _{ESBARC}		1.68		2.06		2.24	ns
t _{ESBSRC}		2.27		2.77		3.18	ns
t _{ESBAWC}		3.10		3.86		4.50	ns
t _{ESBSWC}		2.90		3.67		4.21	ns
t _{ESBWASU}	0.55		0.67		0.74		ns
t _{ESBWAH}	0.36		0.46		0.48		ns
t _{ESBWDSU}	0.69		0.83		0.95		ns
t _{ESBWDH}	0.36		0.46		0.48		ns
t _{ESBRASU}	1.61		1.90		2.09		ns
t _{ESBRAH}	0.00		0.00		0.01		ns
t _{ESBWESU}	1.42		1.71		2.01		ns
t _{ESBWEH}	0.00		0.00		0.00		ns
t _{ESBDATASU}	-0.06		-0.07		0.05		ns
t _{ESBDATAH}	0.13		0.13		0.13		ns
t _{ESBWADDRSU}	0.11		0.13		0.31		ns
t _{ESBRADDRSU}	0.18		0.23		0.39		ns
t _{ESBDATACO1}		1.09		1.35		1.51	ns
t _{ESBDATACO2}		2.19		2.75		3.22	ns
t _{ESBDD}		2.75		3.41		4.03	ns
t _{PD}		1.58		1.97		2.33	ns
t _{PTERMSU}	1.00		1.22		1.51		ns
t _{PTERMCO}		1.10		1.37		1.09	ns

Table 75. EP20K200E f _{MAX} Routing Delays										
Symbol	Symbol -1 -2 -3									
	Min	Max	Min	Max	Min	Max				
t _{F1-4}		0.25		0.27		0.29	ns			
t _{F5-20}		1.02		1.20		1.41	ns			
t _{F20+}		1.99		2.23		2.53	ns			

Tables 85 through 90 describe f_{MAX} LE Timing Microparameters, f_{MAX} ESB Timing Microparameters, f_{MAX} Routing Delays, Minimum Pulse Width Timing Parameters, External Timing Parameters, and External Bidirectional Timing Parameters for EP20K400E APEX 20KE devices.

Table 85. EP2	OK400E f _{MAX}	LE Timing Mic	roparameter	s			
Symbol	-1 Spee	d Grade	-2 Spec	ed Grade	-3 Spee	Unit	
	Min	Max	Min	Max	Min	Max	
t _{SU}	0.23		0.23		0.23		ns
t _H	0.23		0.23		0.23		ns
t _{CO}		0.25		0.29		0.32	ns
t _{LUT}		0.70		0.83		1.01	ns

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t _{ESBARC}		1.67		1.91		1.99	ns
t _{ESBSRC}		2.30		2.66		2.93	ns
t _{ESBAWC}		3.09		3.58		3.99	ns
t _{ESBSWC}		3.01		3.65		4.05	ns
t _{ESBWASU}	0.54		0.63		0.65		ns
t _{ESBWAH}	0.36		0.43		0.42		ns
t _{ESBWDSU}	0.69		0.77		0.84		ns
t _{ESBWDH}	0.36		0.43		0.42		ns
t _{ESBRASU}	1.61		1.77		1.86		ns
t _{ESBRAH}	0.00		0.00		0.01		ns
t _{ESBWESU}	1.35		1.47		1.61		ns
t _{ESBWEH}	0.00		0.00		0.00		ns
t _{ESBDATASU}	-0.18		-0.30		-0.27		ns
t _{ESBDATAH}	0.13		0.13		0.13		ns
t _{ESBWADDRSU}	-0.02		-0.11		-0.03		ns
t _{ESBRADDRSU}	0.06		-0.01		-0.05		ns
t _{ESBDATACO1}		1.16		1.40		1.54	ns
t _{ESBDATACO2}		2.18		2.55		2.85	ns
t _{ESBDD}		2.73		3.17		3.58	ns
t _{PD}		1.57		1.83		2.07	ns
t _{PTERMSU}	0.92		0.99		1.18		ns
t _{PTERMCO}		1.18		1.43		1.17	ns

Table 110. Selectable I/O Standard Output Delays								
Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit	
	Min	Max	Min	Max	Min	Max	Min	
LVCMOS		0.00		0.00		0.00	ns	
LVTTL		0.00		0.00		0.00	ns	
2.5 V		0.00		0.09		0.10	ns	
1.8 V		2.49		2.98		3.03	ns	
PCI		-0.03		0.17		0.16	ns	
GTL+		0.75		0.75		0.76	ns	
SSTL-3 Class I		1.39		1.51		1.50	ns	
SSTL-3 Class II		1.11		1.23		1.23	ns	
SSTL-2 Class I		1.35		1.48		1.47	ns	
SSTL-2 Class II		1.00		1.12		1.12	ns	
LVDS		-0.48		-0.48		-0.48	ns	
CTT		0.00		0.00		0.00	ns	
AGP		0.00		0.00		0.00	ns	

Power Consumption

To estimate device power consumption, use the interactive power calculator on the Altera web site at http://www.altera.com.

Configuration & Operation

The APEX 20K architecture supports several configuration schemes. This section summarizes the device operating modes and available device configuration schemes.

Operating Modes

The APEX architecture uses SRAM configuration elements that require configuration data to be loaded each time the circuit powers up. The process of physically loading the SRAM data into the device is called configuration. During initialization, which occurs immediately after configuration, the device resets registers, enables I/O pins, and begins to operate as a logic device. The I/O pins are tri-stated during power-up, and before and during configuration. Together, the configuration and initialization processes are called *command mode*; normal device operation is called *user mode*.

Before and during device configuration, all I/O pins are pulled to $V_{\mbox{\scriptsize CCIO}}$ by a built-in weak pull-up resistor.