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Understanding Embedded - Microprocessors

Embedded microprocessors are specialized computing chips designed to perform specific tasks within an embedded system. Unlike general-purpose microprocessors found in personal computers, embedded microprocessors are tailored for dedicated functions within larger systems, offering optimized performance, efficiency, and reliability. These microprocessors are integral to the operation of countless electronic devices, providing the computational power necessary for controlling processes, handling data, and managing communications.

Applications of Embedded - Microprocessors

Embedded microprocessors are utilized across a broad spectrum of applications, making them indispensable in

Details	
Product Status	Active
Core Processor	ARM® Cortex®-A9
Number of Cores/Bus Width	2 Core, 32-Bit
Speed	852MHz
Co-Processors/DSP	Multimedia; NEON™ SIMD
RAM Controllers	LPDDR2, DDR3L, DDR3
Graphics Acceleration	Yes
Display & Interface Controllers	HDMI, Keypad, LCD, LVDS, MIPI/DSI, Parallel
Ethernet	10/100/1000Mbps (1)
SATA	SATA 3Gbps (1)
USB	USB 2.0 + PHY (3), USB 2.0 OTG + PHY (1)
Voltage - I/O	1.8V, 2.5V, 2.8V, 3.3V
Operating Temperature	-40°C ~ 125°C (TJ)
Security Features	ARM TZ, A-HAB, CAAM, CSU, SJC, SNVS
Package / Case	624-FBGA, FCBGA
Supplier Device Package	624-FCBGA (21x21)
Purchase URL	https://www.e-xfl.com/product-detail/nxp-semiconductors/mcimx6dp6avt8aar

Table 2. i.MX 6DualPlus/6QuadPlus Modules List (continued)

Block Mnemonic	Block Name	Subsystem	Brief Description
CSI	MIPI CSI-2 Interface	Multimedia Peripherals	The CSI IP provides MIPI CSI-2 standard camera interface port. The CSI-2 interface supports up to 1 Gbps for up to 3 data lanes and up to 800 Mbps for 4 data lanes.
CSU	Central Security Unit	Security	The Central Security Unit (CSU) is responsible for setting comprehensive security policy within the i.MX 6DualPlus/6QuadPlus platform. The Security Control Registers (SCR) of the CSU are set during boot time by the HAB and are locked to prevent further writing.
CTI-0 CTI-1 CTI-2 CTI-3 CTI-4	Cross Trigger Interfaces	Debug / Trace	Cross Trigger Interfaces allows cross-triggering based on inputs from masters attached to CTIs. The CTI module is internal to the Cortex-A9 Core Platform.
CTM	Cross Trigger Matrix	Debug / Trace	Cross Trigger Matrix IP is used to route triggering events between CTIs. The CTM module is internal to the Cortex-A9 Core Platform.
DAP	Debug Access Port	System Control Peripherals	The DAP provides real-time access for the debugger without halting the core to: • System memory and peripheral registers • All debug configuration registers The DAP also provides debugger access to JTAG scan chains. The DAP module is internal to the Cortex-A9 Core Platform.
DCIC-0 DCIC-1	Display Content Integrity Checker	Automotive IP	The DCIC provides integrity check on portion(s) of the display. Each i.MX 6DualPlus/6QuadPlus processor has two such modules, one for each IPU.
DSI	MIPI DSI interface	Multimedia Peripherals	The MIPI DSI IP provides DSI standard display port interface. The DSI interface support 80 Mbps to 1 Gbps speed per data lane.
eCSPI1-5	Configurable SPI	Connectivity Peripherals	Full-duplex enhanced Synchronous Serial Interface. It is configurable to support Master/Slave modes, four chip selects to support multiple peripherals.
ENET	Ethernet Controller	Connectivity Peripherals	The Ethernet Media Access Controller (MAC) is designed to support 10/100/1000 Mbps Ethernet/IEEE 802.3 networks. An external transceiver interface and transceiver function are required to complete the interface to the media. The i.MX 6DualPlus/6QuadPlus processors also consist of hardware assist for IEEE 1588 standard. For details, see the ENET chapter of the i.MX 6DualPlus/6QuadPlus reference manual (IMX6DQPRM). **Note:** The theoretical maximum performance of 1 Gbps ENET is limited to 470 Mbps (total for Tx and Rx) due to internal bus throughput limitations. The actual measured performance in optimized environment is up to 400 Mbps. For details, see the ERR004512 erratum in the i.MX 6Dual/6Quad errata document (IMX6DQCE).
EPIT-1 EPIT-2	Enhanced Periodic Interrupt Timer	Timer Peripherals	Each EPIT is a 32-bit “set and forget” timer that starts counting after the EPIT is enabled by software. It is capable of providing precise interrupts at regular intervals with minimal processor intervention. It has a 12-bit prescaler for division of input clock frequency to get the required time setting for the interrupts to occur, and counter value can be programmed on the fly.

Table 2. i.MX 6DualPlus/6QuadPlus Modules List (continued)

Block Mnemonic	Block Name	Subsystem	Brief Description
PRE1 PRE2 PRE3 PRE4	Prefetch/Resolve Engine	Multimedia Peripherals	The PRE includes the Resolve engine, Prefetch engine, and Store engine 3 blocks.The PRE key features are: **The Resolve engine supports:** • GPU 32bpp 4x4 standard tile, 4x4 split tile, 4x4 super tile, 4x4 super split tile format. • GPU 16bpp 8x4 standard tile, 8x4 split tile, 8x4 super tile, 8x4 super split format. • 32/16x4 block mode and scan mode. **The prefetch engine supports:** • Transfer of non-interleaved YUV422(NI422), non-interleaved YUV420(NI420), partial interleaved YUV422(PI422), and partial interleaved YUV420(PI420), inputs to interleaved YUV422. • Vertical flip function both in block mode and scan mode. In block mode, vertical flip function should complete with TPR module enable. • 8bpp, 16bpp, 32bpp and 64bpp data format as generic data. • Transfer of non-interleaved YUV444(NI444), input to interleaved YUV444 output. **The store Engine supports:** 4/8/16 lines handshake modes with PRG.
PRG1 PRG2	Prefetch/Resolve Gasket	Multimedia Peripherals	The PRG is a digital core function which works as a gasket interface between the fabric and the IPU system. The primary function is to re-map the ARADDR from a frame-based address to a band-based address depending on the different ARIDs. The PRG also implements the handshake logic with the Prefetch Resolve Engine (PRE).
PMU	Power-Management Functions	Data Path	Integrated power management unit. Used to provide power to various SoC domains.
PWM-1 PWM-2 PWM-3 PWM-4	Pulse Width Modulation	Connectivity Peripherals	The pulse-width modulator (PWM) has a 16-bit counter and is optimized to generate sound from stored sample audio images and it can also generate tones. It uses 16-bit resolution and a 4x16 data FIFO to generate sound.
RAM 16 KB	Secure/non-secure RAM	Secured Internal Memory	Secure/non-secure Internal RAM, interfaced through the CAAM.
RAM 512 KB	Internal RAM	Internal Memory	Internal RAM, which is accessed through OCRAM memory controllers.
ROM 96 KB	Boot ROM	Internal Memory	Supports secure and regular Boot Modes. Includes read protection on 4K region for content protection
SATA	Serial ATA	Connectivity Peripherals	The SATA controller and PHY is a complete mixed-signal IP solution designed to implement SATA II, 3.0 Gbps HDD connectivity.

Table 2. i.MX 6DualPlus/6QuadPlus Modules List (continued)

Block Mnemonic	Block Name	Subsystem	Brief Description
SDMA	Smart Direct Memory Access	System Control Peripherals	The SDMA is multi-channel flexible DMA engine. It helps in maximizing system performance by off-loading the various cores in dynamic data routing. It has the following features: • Powered by a 16-bit Instruction-Set micro-RISC engine • Multi-channel DMA supporting up to 32 time-division multiplexed DMA channels • 48 events with total flexibility to trigger any combination of channels • Memory accesses including linear, FIFO, and 2D addressing • Shared peripherals between ARM and SDMA • Very fast context-switching with 2-level priority based preemptive multi-tasking • DMA units with auto-flush and prefetch capability • Flexible address management for DMA transfers (increment, decrement, and no address changes on source and destination address) • DMA ports can handle unit-directional and bi-directional flows (copy mode) • Up to 8-word buffer for configurable burst transfers • Support of byte-swapping and CRC calculations • Library of Scripts and API is available
SJC	System JTAG Controller	System Control Peripherals	The SJC provides JTAG interface, which complies with JTAG TAP standards, to internal logic. The i.MX 6DualPlus/6QuadPlus processors use JTAG port for production, testing, and system debugging. In addition, the SJC provides BSR (Boundary Scan Register) standard support, which complies with IEEE1149.1 and IEEE1149.6 standards. The JTAG port must be accessible during platform initial laboratory bring-up, for manufacturing tests and troubleshooting, as well as for software debugging by authorized entities. The i.MX 6DualPlus/6QuadPlus SJC incorporates three security modes for protecting against unauthorized accesses. Modes are selected through eFUSE configuration.
SNVS	Secure Non-Volatile Storage	Security	Secure Non-Volatile Storage, including Secure Real Time Clock, Security State Machine, Master Key Control, and Violation/Tamper Detection and reporting.
SPDIF	Sony Philips Digital Interconnect Format	Multimedia Peripherals	A standard audio file transfer format, developed jointly by the Sony and Phillips corporations. It supports Transmitter and Receiver functionality.
SSI-1 SSI-2 SSI-3	I2S/SSI/AC97 Interface	Connectivity Peripherals	The SSI is a full-duplex synchronous interface, which is used on the processor to provide connectivity with off-chip audio peripherals. The SSI supports a wide variety of protocols (SSI normal, SSI network, I2S, and AC-97), bit depths (up to 24 bits per word), and clock / frame sync options. The SSI has two pairs of 8x24 FIFOs and hardware support for an external DMA controller to minimize its impact on system performance. The second pair of FIFOs provides hardware interleaving of a second audio stream that reduces CPU overhead in use cases where two time slots are being used simultaneously.

4.1.2 Thermal Resistance

NOTE

Per JEDEC JESD51-2, the intent of thermal resistance measurements is solely for a thermal performance comparison of one package to another in a standardized environment. This methodology is not meant to and will not predict the performance of a package in an application-specific environment.

4.1.2.1 FCPBGA Package Thermal Resistance

Table 5 provides the FCPBGA package thermal resistance data for the *lidded* package type.

Table 5. FCPBGA Package Thermal Resistance Data (Lidded)

Thermal Parameter	Test Conditions	Symbol	Value	Unit
Junction to Ambient[1]	Single-layer board (1s); natural convection[2]	$R_{\theta JA}$	24	°C/W
	Four-layer board (2s2p); natural convection[2]	$R_{\theta JA}$	15	°C/W
Junction to Ambient[1]	Single-layer board (1s); air flow 200 ft/min[3]	$R_{\theta JMA}$	17	°C/W
	Four-layer board (2s2p); air flow 200 ft/min[4]	$R_{\theta JMA}$	12	°C/W
Junction to Board[1,4]	—	$R_{\theta JB}$	5	°C/W
Junction to Case (top)[1,5]	—	$R_{\theta JCtop}$	1	°C/W

1 Junction temperature is a function of die size, on-chip power dissipation, package thermal resistance, mounting site (board) temperature, ambient temperature, air flow, power dissipation of other components on the board, and board thermal resistance.

2 Per JEDEC JESD51-3 with the single layer board horizontal. Thermal test board meets JEDEC specification for the specified package.

3 Per JEDEC JESD51-6 with the board horizontal.

4 Thermal resistance between the die and the printed circuit board per JEDEC JESD51-8. Board temperature is measured on the top surface of the board near the package.

5 Thermal resistance between the die and the case top surface as measured by the cold plate method (MIL SPEC-883 Method 1012.1). The cold plate temperature is used for the case temperature. Reported value includes the thermal resistance of the interface layer.

Table 8. Maximum Supply Currents (continued)

Power Supply	Conditions	Maximum Current		Unit
		Power Virus	CoreMark	
NVCC_LVDS2P5	—	NVCC_LVDS2P5 is connected to VDD_HIGH_CAP at the board level. VDD_HIGH_CAP is capable of handing the current required by NVCC_LVDS2P5.		
MISC				
DRAM_VREF	—	1		mA

1 i.MX 6DualPlus numbers assume VDD_ARM23_IN and VDD_ARM23_CAP are connected to ground.

2 The actual maximum current drawn from VDD_HIGH_IN will be as shown plus any additional current drawn from the VDD_HIGH_CAP outputs, depending upon actual application configuration (for example, NVCC_LVDS_2P5, NVCC_MIPI, or HDMI, PCIe, and SATA VPH supplies).

3 Under normal operating conditions, the maximum current on VDD_SNVS_IN is shown Table 8. The maximum VDD_SNVS_IN current may be higher depending on specific operating configurations, such as BOOT_MODE[1:0] not equal to 00, or use of the Tamper feature. During initial power on, VDD_SNVS_IN can draw up to 1 mA if the supply is capable of sourcing that current. If less than 1 mA is available, the VDD_SNVS_CAP charge time will increase.

4 This is the maximum current per active USB physical interface.

5 The DRAM power consumption is dependent on several factors such as external signal termination. DRAM power calculators are typically available from memory vendors which take into account factors such as signal termination.
See the *i.MX 6Dual/6Quad Power Consumption Measurement Application Note* (AN4509) for examples of DRAM power consumption during specific use case scenarios.

6 General equation for estimated, maximum power consumption of an IO power supply:
Imax = N x C x V x (0.5 x F)
Where:
N—Number of IO pins supplied by the power line
C—Equivalent external capacitive load
V—IO voltage
(0.5 xF)—Data change rate. Up to 0.5 of the clock rate (F)
In this equation, Imax is in Amps, C in Farads, V in Volts, and F in Hertz.

4.1.6 Low Power Mode Supply Currents

Table 9 shows the current core consumption (not including I/O) of the i.MX 6DualPlus/6QuadPlus processors in selected low power modes.

Table 9. Stop Mode Current and Power Consumption

Mode	Test Conditions	Supply	Typical[1]	Unit
WAIT	• ARM, SoC, and PU LDOs are set to 1.225 V • HIGH LDO set to 2.5 V • Clocks are gated • DDR is in self refresh • PLLs are active in bypass (24 MHz) • Supply voltages remain ON	VDD_ARM_IN (1.4 V)	6	mA
		VDD_SOC_IN (1.4 V)	23	mA
		VDD_HIGH_IN (3.0 V)	3.7	mA
		Total	52	mW

Table 11. SATA PHY Current Drain (continued)

Mode	Test Conditions	Supply	Typical Current	Unit
P1: Transmitter idle, Rx powered down, LOS disabled	Single Transceiver	SATA_VP	0.67	mA
		SATA_VPH	0.23	
	Clock Module	SATA_VP	6.9	
		SATA_VPH	6.2	
P2: Powered-down state, only LOS and POR enabled	Single Transceiver	SATA_VP	0.53	mA
		SATA_VPH	0.11	
	Clock Module	SATA_VP	0.036	
		SATA_VPH	0.12	
PDDQ mode[3]	Single Transceiver	SATA_VP	0.13	mA
		SATA_VPH	0.012	
	Clock Module	SATA_VP	0.008	
		SATA_VPH	0.004	

1 Programmed for 1.0 V peak-to-peak Tx level.

2 Programmed for 0.9 V peak-to-peak Tx level with no boost or attenuation.

3 LOW power non-functional.

4.1.9 PCIe 2.0 Maximum Power Consumption

Table 12 provides PCIe PHY currents for certain operating modes.

Table 12. PCIe PHY Current Drain

Mode	Test Conditions	Supply	Max Current	Unit
P0: Normal Operation	5G Operations	PCIE_VP (1.1 V)	40	mA
		PCIE_VPTX (1.1 V)	20	
		PCIE_VPH (2.5 V)	21	
	2.5G Operations	PCIE_VP (1.1 V)	27	
		PCIE_VPTX (1.1 V)	20	
		PCIE_VPH (2.5 V)	20	
P0s: Low Recovery Time Latency, Power Saving State	5G Operations	PCIE_VP (1.1 V)	30	mA
		PCIE_VPTX (1.1 V)	2.4	
		PCIE_VPH (2.5 V)	18	
	2.5G Operations	PCIE_VP (1.1 V)	20	
		PCIE_VPTX (1.1 V)	2.4	
		PCIE_VPH (2.5 V)	18	

system to determine when the load capability of the regulator is being exceeded to take the necessary steps. Current-limiting can be enabled to allow for in-rush current requirements during start-up, if needed. Active-pull-down can also be enabled for systems requiring this feature.

For information on external capacitor requirements for this regulator, see the Hardware Development Guide for i.MX 6Quad, 6Dual, 6DualLite, 6Solo Families of Applications Processors (IMX6DQ6SDLHDG).

For additional information, see the i.MX 6DualPlus/6QuadPlus reference manual (IMX6DQPRM).

4.3.2.2 LDO_2P5 / VDDHIGH_CAP

The LDO_2P5 module implements a programmable linear-regulator function from VDD_HIGH_IN (see Table 6 for min and max input requirements). Typical Programming Operating Range is 2.25 V to 2.75 V with the nominal default setting as 2.5 V. The LDO_2P5 supplies the eFuses, PLLs, and USB PHY. Optionally it can be used to supply the HDMI, LVDS, MIPI, PCIe, and SATA PHY's through external connections. A programmable brown-out detector is included in the regulator that can be used by the system to determine when the load capability of the regulator is being exceeded, to take the necessary steps. Current-limiting can be enabled to allow for in-rush current requirements during start-up, if needed. Active-pull-down can also be enabled for systems requiring this feature. An alternate self-biased low-precision weak-regulator is included that can be enabled for applications needing to keep the output voltage alive during low-power modes where the main regulator driver and its associated global bandgap reference module are disabled. The output of the weak-regulator is not programmable and is a function of the input supply as well as the load current. Typically, with a 3 V input supply the weak-regulator output is 2.525 V and its output impedance is approximately 40 Ω.

For information on external capacitor requirements for this regulator, see the Hardware Development Guide for i.MX 6Quad, 6Dual, 6DualLite, 6Solo Families of Applications Processors (IMX6DQ6SDLHDG).

For additional information, see the i.MX 6DualPlus/6QuadPlus reference manual (IMX6DQPRM).

4.3.2.3 LDO_USB / VDD_VBUS_CAP

The LDO_USB module implements a programmable linear-regulator function from the USB_OTG_VBUS and USB_H1_VBUS voltages (4.4 V–5.25 V) to produce a nominal 3.0 V output voltage. A programmable brown-out detector is included in the regulator that can be used by the system to determine when the load capability of the regulator is being exceeded, to take the necessary steps. This regulator has a built in power-mux that allows the user to select to run the regulator from either VBUS supply, when both are present. If only one of the VBUS voltages is present, then the regulator automatically selects this supply. Current limit is also included to help the system meet in-rush current targets. If no VBUS voltage is present, then the VBUSVALID threshold setting will prevent the regulator from being enabled.

For information on external capacitor requirements for this regulator, see the Hardware Development Guide for i.MX 6Quad, 6Dual, 6DualLite, 6Solo Families of Applications Processors (IMX6DQ6SDLHDG).

For additional information, see the i.MX 6DualPlus/6QuadPlus reference manual (IMX6DQPRM).

Figure 14 to Figure 17 provide few examples of basic EIM accesses to external memory devices with the timing parameters mentioned previously for specific control parameters settings.

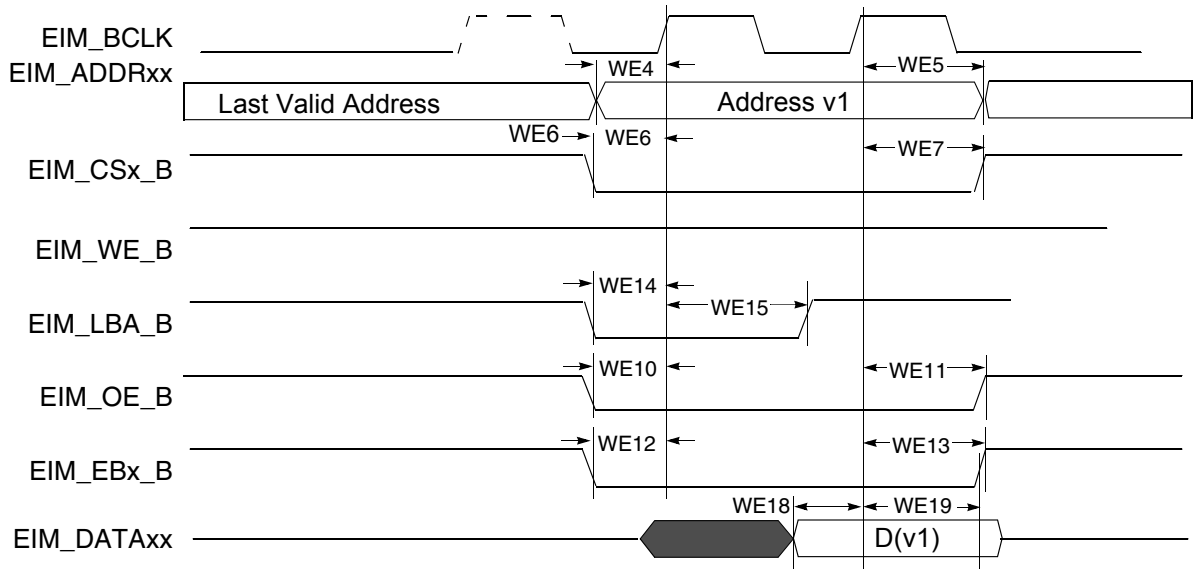

Figure 14. Synchronous Memory Read Access, WSC=1

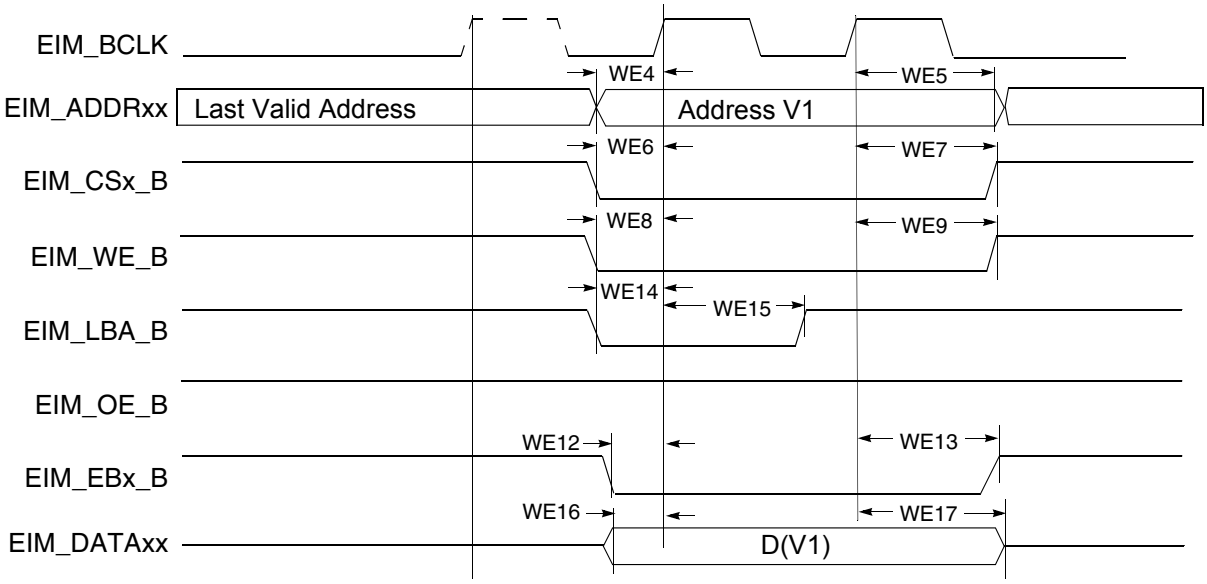

Figure 15. Synchronous Memory, Write Access, WSC=1, WBEA=0 and WADVN=0

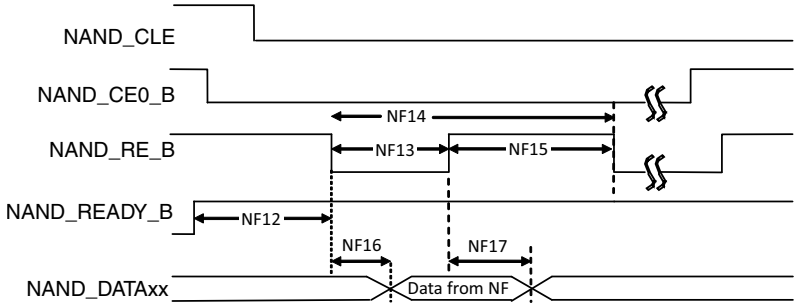

Figure 27. Read Data Latch Cycle Timing Diagram (Non-EDO Mode)

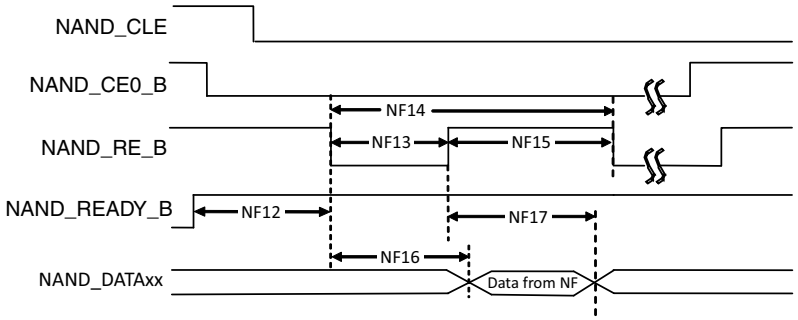

Figure 28. Read Data Latch Cycle Timing Diagram (EDO Mode)

Table 44. Asynchronous Mode Timing Parameters[1]

ID	Parameter	Symbol	Timing T = GPMI Clock Cycle		Unit
			Min	Max	
NF1	NAND_CLE setup time	tCLS	(AS + DS) × T - 0.12 [see [2,3]]		ns
NF2	NAND_CLE hold time	tCLH	DH × T - 0.72 [see [2]]		ns
NF3	NAND_CEx_B setup time	tCS	(AS + DS + 1) × T [see [3,2]]		ns
NF4	NAND_CEx_B hold time	tCH	(DH+1) × T - 1 [see [2]]		ns
NF5	NAND_WE_B pulse width	tWP	DS × T [see [2]]		ns
NF6	NAND_ALE setup time	tALS	(AS + DS) × T - 0.49 [see [3,2]]		ns
NF7	NAND_ALE hold time	tALH	(DH × T - 0.42 [see [2]]		ns
NF8	Data setup time	tDS	DS × T - 0.26 [see [2]]		ns
NF9	Data hold time	tDH	DH × T - 1.37 [see [2]]		ns
NF10	Write cycle time	tWC	(DS + DH) × T [see [2]]		ns
NF11	NAND_WE_B hold time	tWH	DH × T [see [2]]		ns
NF12	Ready to NAND_RE_B low	tRR[4]	(AS + 2) × T [see [3,2]]	—	ns
NF13	NAND_RE_B pulse width	tRP	DS × T [see [2]]		ns
NF14	READ cycle time	tRC	(DS + DH) × T [see [2]]		ns
NF15	NAND_RE_B high hold time	tREH	DH × T [see [2]]		ns

Table 44. Asynchronous Mode Timing Parameters[1] (continued)

ID	Parameter	Symbol	Timing T = GPMI Clock Cycle		Unit
			Min	Max	
NF16	Data setup on read	tDSR	—	(DS × T -0.67)/18.38 [see [5,6]]	ns
NF17	Data hold on read	tDHR	0.82/11.83 [see [5,6]]	—	ns

1 The GPMI asynchronous mode output timing can be controlled by the module's internal registers HW_GPMI_TIMING0_ADDRESS_SETUP, HW_GPMI_TIMING0_DATA_SETUP, and HW_GPMI_TIMING0_DATA_HOLD. This AC timing depends on these registers settings. In the table, AS/DS/DH represents each of these settings.

2 AS minimum value can be 0, while DS/DH minimum value is 1.

3 T = GPMI clock period -0.075ns (half of maximum p-p jitter).

4 NF12 is met automatically by the design.

5 Non-EDO mode.

6 EDO mode, GPMI clock ≈ 100 MHz
(AS=DS=DH=1, GPMI_CTL1 [RDN_DELAY] = 8, GPMI_CTL1 [HALF_PERIOD] = 0).

In EDO mode (Figure 28), NF16/NF17 are different from the definition in non-EDO mode (Figure 27). They are called tREA/tRHOH (NAND_RE_B access time/NAND_RE_B HIGH to output hold). The typical value for them are 16 ns (max for tREA)/15 ns (min for tRHOH) at 50 MB/s EDO mode. In EDO mode, GPMI will sample NAND_DATAxx at rising edge of delayed NAND_RE_B provided by an internal DPLL. The delay value can be controlled by GPMI_CTRL1.RDN_DELAY (see the GPMI chapter of the i.MX 6DualPlus/6QuadPlus reference manual (IMX6DQPRM)). The typical value of this control register is 0x8 at 50 MT/s EDO mode. However, if the board delay is large enough and cannot be ignored, the delay value should be made larger to compensate the board delay.

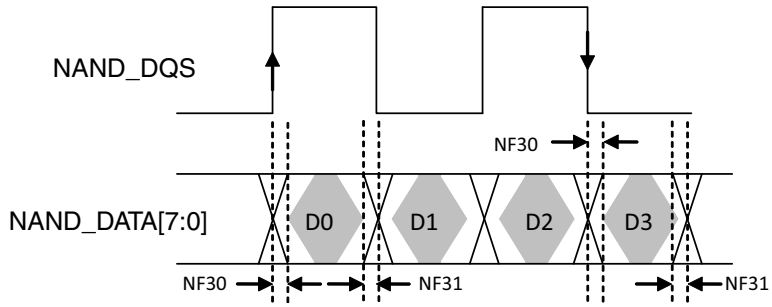

Figure 32. NAND_DQS/NAND_DQ Read Valid Window

Table 45. Source Synchronous Mode Timing Parameters[1]

ID	Parameter	Symbol	Timing T = GPMI Clock Cycle		Unit
			Min	Max	
NF18	NAND_CEx_B access time	tCE	CE_DELAY × T - 0.79 [see 2]		ns
NF19	NAND_CEx_B hold time	tCH	0.5 × tCK - 0.63 [see 2]		ns
NF20	Command/address NAND_DATAxx setup time	tCAS	0.5 × tCK - 0.05		ns
NF21	Command/address NAND_DATAxx hold time	tCAH	0.5 × tCK - 1.23		ns
NF22	clock period	tCK	—		ns
NF23	preamble delay	tPRE	PRE_DELAY × T - 0.29 [see 2]		ns
NF24	postamble delay	tPOST	POST_DELAY × T - 0.78 [see 2]		ns
NF25	NAND_CLE and NAND_ALE setup time	tCALS	0.5 × tCK - 0.86		ns
NF26	NAND_CLE and NAND_ALE hold time	tCALH	0.5 × tCK - 0.37		ns
NF27	NAND_CLK to first NAND_DQS latching transition	tDQSS	T - 0.41 [see 2]		ns
NF28	Data write setup	tDS	0.25 × tCK - 0.35		—
NF29	Data write hold	tDH	0.25 × tCK - 0.85		—
NF30	NAND_DQS/NAND_DQ read setup skew	tDQSQ	—	2.06	—
NF31	NAND_DQS/NAND_DQ read hold skew	tQHS	—	1.95	—

1 The GPMI source synchronous mode output timing can be controlled by the module's internal registers GPMI_TIMING2_CE_DELAY, GPMI_TIMING_PREAMBLE_DELAY, GPMI_TIMING2_POST_DELAY. This AC timing depends on these registers settings. In the table, CE_DELAY/PRE_DELAY/POST_DELAY represents each of these settings.

2 T = tCK (GPMI clock period) -0.075ns (half of maximum p-p jitter).

Figure 32 shows the timing diagram of NAND_DQS/NAND_DATAxx read valid window. For Source Synchronous mode, the typical value of tDQSQ is 0.85 ns (max) and 1 ns (max) for tQHS at 200MB/s. GPMI will sample NAND_DATA[7:0] at both rising and falling edge of a delayed NAND_DQS signal, which can be provided by an internal DPLL. The delay value can be controlled by GPMI register GPMI_READ_DDR_DLL_CTRL.SLV_DLY_TARGET (see the GPMI chapter of the i.MX 6DualPlus/6QuadPlus reference manual (IMX6DQPRM)). Generally, the typical delay value of this register is equal to 0x7 which means 1/4 clock cycle delay expected. However, if the board delay is large enough and cannot be ignored, the delay value should be made larger to compensate the board delay.

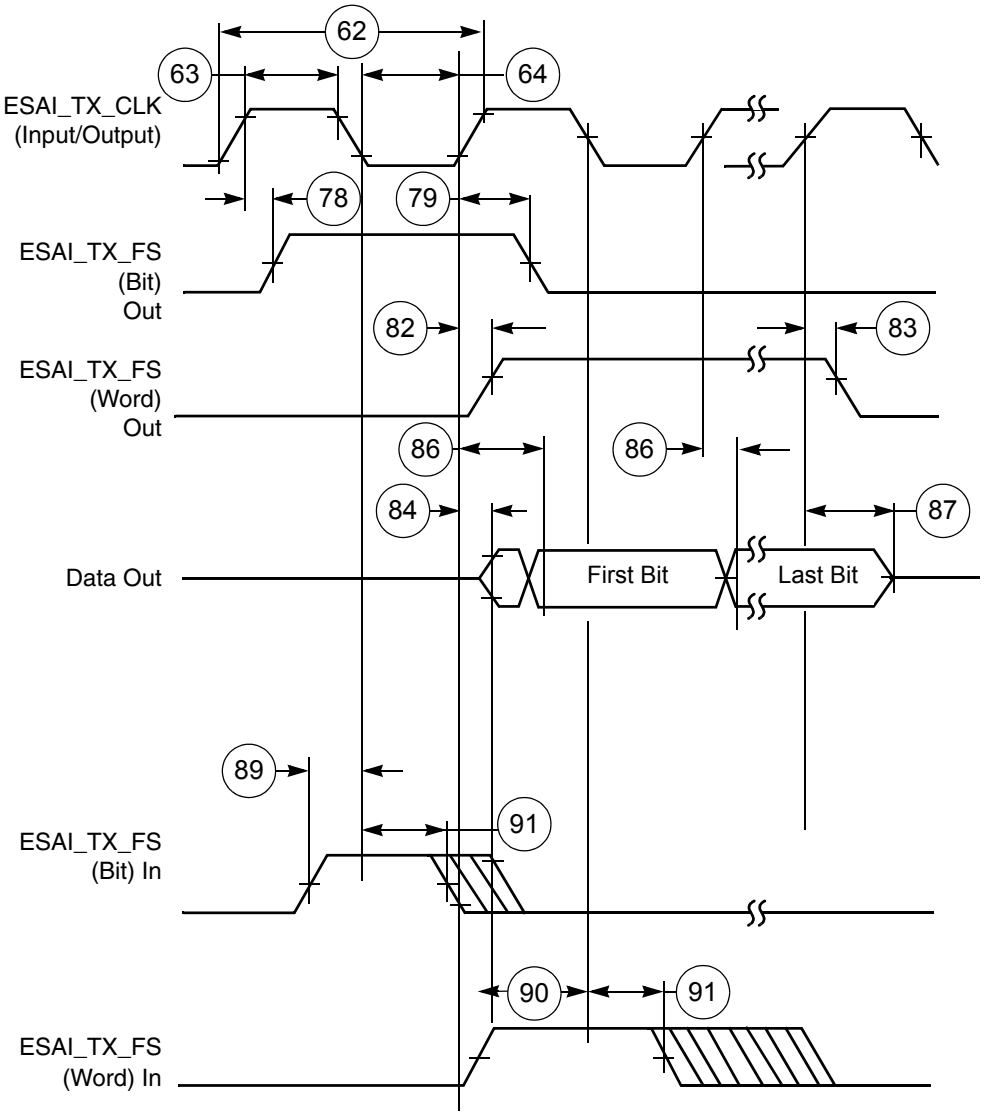

Figure 37. ESAI Transmitter Timing

stops receiving data from the stream. For the next line, the IPU2_CSIx_HSYNC timing repeats. For the next frame, the IPU2_CSIx_VSYNC timing repeats.

4.12.10.2.3 Non-Gated Clock Mode

The timing is the same as the gated-clock mode (described in Section 4.12.10.2.2, "Gated Clock Mode,") except for the IPU2_CSIx_HSYNC signal, which is not used (see Figure 60). All incoming pixel clocks are valid and cause data to be latched into the input FIFO. The IPU2_CSIx_PIX_CLK signal is inactive (states low) until valid data is going to be transmitted over the bus.

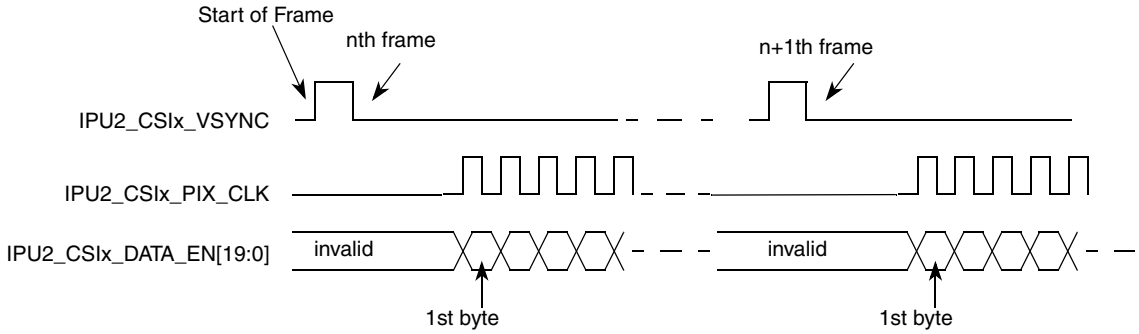

Figure 60. Non-Gated Clock Mode Timing Diagram

The timing described in Figure 60 is that of a typical sensor. Some other sensors may have a slightly different timing. The CSI can be programmed to support rising/falling-edge triggered IPU2_CSIx_VSYNC; active-high/low IPU2_CSIx_HSYNC; and rising/falling-edge triggered IPU2_CSIx_PIX_CLK.

Table 64. Video Signal Cross-Reference (continued)

i.MX 6DualPlus/6QuadPlus	LCD							Comment[1,2]
Port Name (x = 0, 1)	RGB, Signal Name (General)	RGB/TV Signal Allocation (Example)						
		16-bit RGB	18-bit RGB	24 Bit RGB	8-bit YCrCb[3]	16-bit YCrCb	20-bit YCrCb	
IPUx_DISPx_DAT05	DAT[5]	G[0]	B[5]	B[5]	Y/C[5]	C[5]	C[5]	—
IPUx_DISPx_DAT06	DAT[6]	G[1]	G[0]	B[6]	Y/C[6]	C[6]	C[6]	—
IPUx_DISPx_DAT07	DAT[7]	G[2]	G[1]	B[7]	Y/C[7]	C[7]	C[7]	—
IPUx_DISPx_DAT08	DAT[8]	G[3]	G[2]	G[0]	—	Y[0]	C[8]	—
IPUx_DISPx_DAT09	DAT[9]	G[4]	G[3]	G[1]	—	Y[1]	C[9]	—
IPUx_DISPx_DAT10	DAT[10]	G[5]	G[4]	G[2]	—	Y[2]	Y[0]	—
IPUx_DISPx_DAT11	DAT[11]	R[0]	G[5]	G[3]	—	Y[3]	Y[1]	—
IPUx_DISPx_DAT12	DAT[12]	R[1]	R[0]	G[4]	—	Y[4]	Y[2]	—
IPUx_DISPx_DAT13	DAT[13]	R[2]	R[1]	G[5]	—	Y[5]	Y[3]	—
IPUx_DISPx_DAT14	DAT[14]	R[3]	R[2]	G[6]	—	Y[6]	Y[4]	—
IPUx_DISPx_DAT15	DAT[15]	R[4]	R[3]	G[7]	—	Y[7]	Y[5]	—
IPUx_DISPx_DAT16	DAT[16]	—	R[4]	R[0]	—	—	Y[6]	—
IPUx_DISPx_DAT17	DAT[17]	—	R[5]	R[1]	—	—	Y[7]	—
IPUx_DISPx_DAT18	DAT[18]	—	—	R[2]	—	—	Y[8]	—
IPUx_DISPx_DAT19	DAT[19]	—	—	R[3]	—	—	Y[9]	—
IPUx_DISPx_DAT20	DAT[20]	—	—	R[4]	—	—	—	—
IPUx_DISPx_DAT21	DAT[21]	—	—	R[5]	—	—	—	—
IPUx_DISPx_DAT22	DAT[22]	—	—	R[6]	—	—	—	—
IPUx_DISPx_DAT23	DAT[23]	—	—	R[7]	—	—	—	—
IPUx_DIx_DISP_CLK	PixCLK							—
IPUx_DIx_PIN01	—							May be required for anti-tearing
IPUx_DIx_PIN02	HSYNC							—
IPUx_DIx_PIN03	VSYNC							VSYNC out

Table 68. Electrical and Timing Information (continued)

Symbol	Parameters	Test Conditions	Min	Typ	Max	Unit
LP Line Receiver DC Specifications						
V_{IL}	Input low voltage	—	—	—	550	mV
V_{IH}	Input high voltage	—	920	—	—	mV
V_{HYST}	Input hysteresis	—	25	—	—	mV
Contention Line Receiver DC Specifications						
V_{ILF}	Input low fault threshold	—	200	—	450	mV

4.12.12.2 D-PHY Signaling Levels

The signal levels are different for differential HS mode and single-ended LP mode. Figure 66 shows both the HS and LP signal levels on the left and right sides, respectively. The HS signalling levels are below the LP low-level input threshold such that LP receiver always detects low on HS signals.

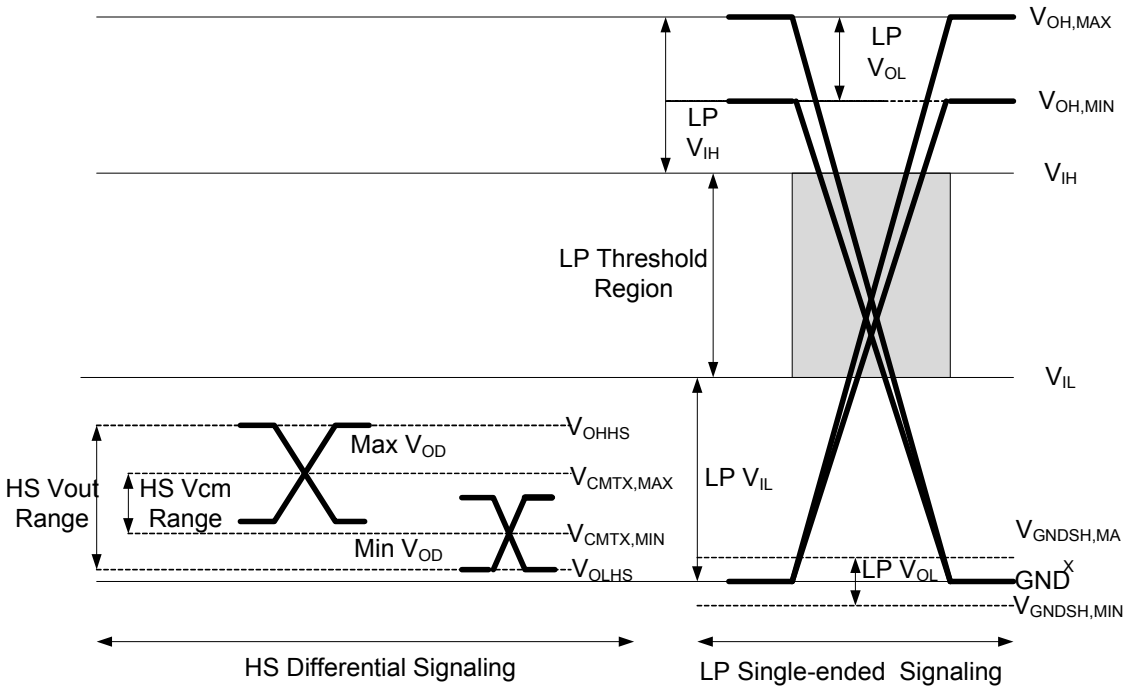

Figure 66. D-PHY Signaling Levels

4.12.12.3 HS Line Driver Characteristics

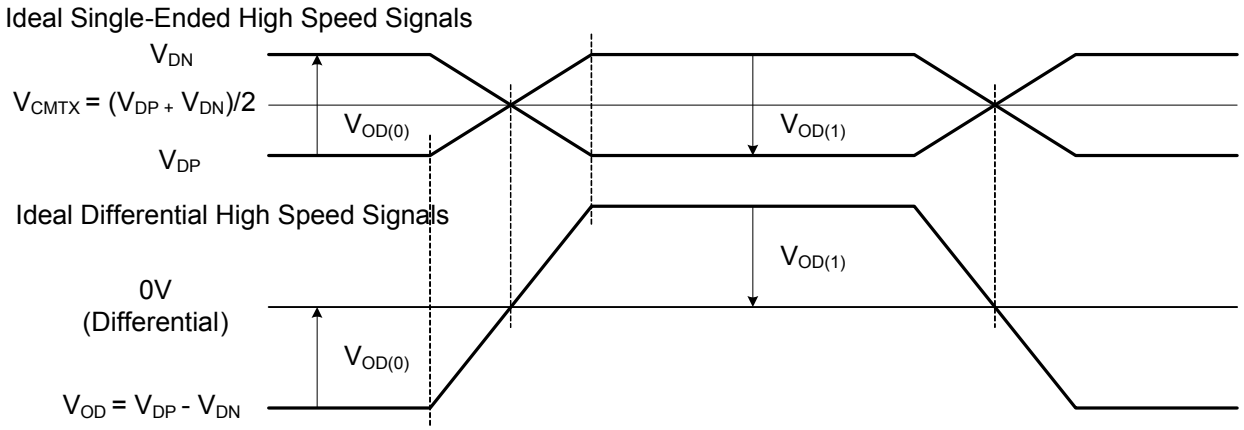

Figure 67. Ideal Single-ended and Resulting Differential HS Signals

4.12.12.4 Possible ΔVCMTX and ΔVOD Distortions of the Single-ended HS Signals

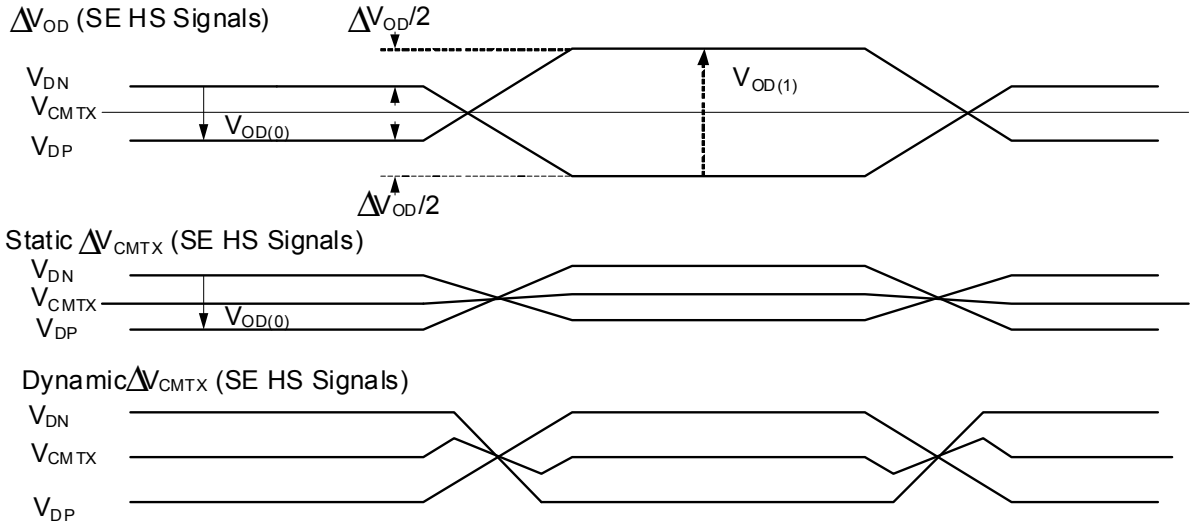

Figure 68. Possible ΔVCMTX and ΔVOD Distortions of the Single-ended HS Signals

4.12.12.5 D-PHY Switching Characteristics

Table 69. Electrical and Timing Information

Symbol	Parameters	Test Conditions	Min	Typ	Max	Unit
HS Line Drivers AC Specifications						
—	Maximum serial data rate (forward direction)	On DATAP/N outputs. 80 Ω <= RL <= 125 Ω	80	—	1000	Mbps
F_{DDRCLK}	DDR CLK frequency	On DATAP/N outputs.	40	—	500	MHz
P_{DDRCLK}	DDR CLK period	80 Ω <= RL< = 125 Ω	2	—	25	ns

4.12.13.9 DATA and FLAG Signal Timing

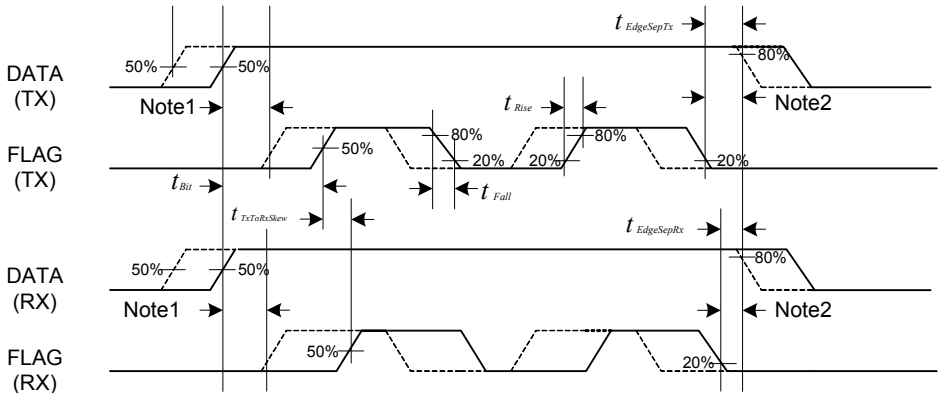

Figure 80. DATA and FLAG Signal Timing

4.12.14 MediaLB (MLB) Characteristics

4.12.14.1 MediaLB (MLB) DC Characteristics

Table 71 lists the MediaLB 3-pin interface electrical characteristics.

Table 71. MediaLB 3-Pin Interface Electrical DC Specifications

Parameter	Symbol	Test Conditions	Min	Max	Unit
Maximum input voltage	—	—	—	3.6	V
Low level input threshold	V_{IL}	—	—	0.7	V
High level input threshold	V_{IH}	See Note[1]	1.8	—	V
Low level output threshold	V_{OL}	I_{OL} = 6 mA	—	0.4	V
High level output threshold	V_{OH}	I_{OH} = –6 mA	2.0	—	V
Input leakage current	I_L	0 < V_{in} < VDD	—	±10	µA

[1] Higher V_{IH} thresholds can be used; however, the risks associated with less noise margin in the system must be evaluated and assumed by the customer.

Table 72 lists the MediaLB 6-pin interface electrical characteristics.

Table 72. MediaLB 6-Pin Interface Electrical DC Specifications

Parameter	Symbol	Test Conditions	Min	Max	Unit
Driver Characteristics					
Differential output voltage (steady-state): I V_{O+} - V_{O-} I	V_{OD}	See Note[1]	300	500	mV
Difference in differential output voltage between (high/low) steady-states: I $V_{OD, high}$ - $V_{OD, low}$ I	ΔV_{OD}	—	-50	50	mV

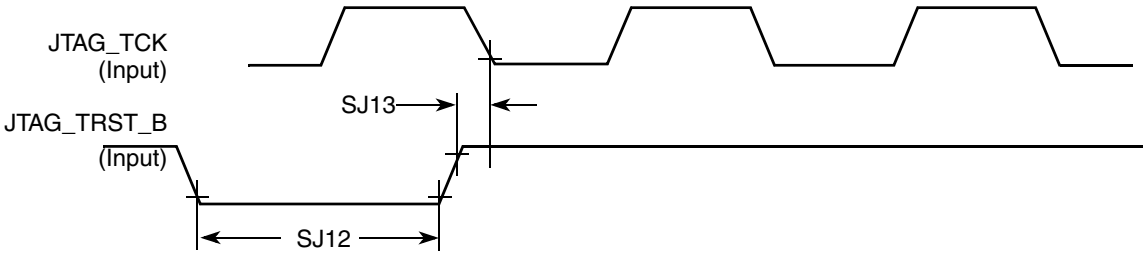

Figure 87. JTAG_TRST_B Timing Diagram

Table 79. JTAG Timing

ID	Parameter[1,2]	All Frequencies		Unit
		Min	Max	
SJ0	JTAG_TCK frequency of operation $1/(3xT_{DC})$[1]	0.001	22	MHz
SJ1	JTAG_TCK cycle time in crystal mode	45	—	ns
SJ2	JTAG_TCK clock pulse width measured at V_M[2]	22.5	—	ns
SJ3	JTAG_TCK rise and fall times	—	3	ns
SJ4	Boundary scan input data set-up time	5	—	ns
SJ5	Boundary scan input data hold time	24	—	ns
SJ6	JTAG_TCK low to output data valid	—	40	ns
SJ7	JTAG_TCK low to output high impedance	—	40	ns
SJ8	JTAG_TMS, JTAG_TDI data set-up time	5	—	ns
SJ9	JTAG_TMS, JTAG_TDI data hold time	25	—	ns
SJ10	JTAG_TCK low to JTAG_TDO data valid	—	44	ns
SJ11	JTAG_TCK low to JTAG_TDO high impedance	—	44	ns
SJ12	JTAG_TRST_B assert time	100	—	ns
SJ13	JTAG_TRST_B set-up time to JTAG_TCK low	40	—	ns

1 T_{DC} = target frequency of SJC
2 V_M = mid-point voltage

4.12.19 SPDIF Timing Parameters

The Sony/Philips Digital Interconnect Format (SPDIF) data is sent using the bi-phase marking code. When encoding, the SPDIF data signal is modulated by a clock that is twice the bit rate of the data signal.

Table 80 and Figure 88 and Figure 89 show SPDIF timing parameters for the Sony/Philips Digital Interconnect Format (SPDIF), including the timing of the modulating Rx clock (SPDIF_SR_CLK) for SPDIF in Rx mode and the timing of the modulating Tx clock (SPDIF_ST_CLK) for SPDIF in Tx mode.

NOTES:

1. ALL DIMENSIONS IN MILLIMETERS.
2. DIMENSIONING AND TOLERANCING PER ASME Y14.5M–1994.
3. MAXIMUM SOLDER BALL DIAMETER MEASURED PARALLEL TO DATUM A.
4. DATUM A, THE SEATING PLANE, IS DETERMINED BY THE SPHERICAL CROWNS OF THE SOLDER BALLS.
5. PARALLELISM MEASUREMENT SHALL EXCLUDE ANY EFFECT OF MARK ON TOP SURFACE OF PACKAGE.
6. 21.2MM MAXIMUM PACKAGE ASSEMBLY (LID + LAMINATE) X AND Y.

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TITLE: 624 I/O FC PBGA, 21 X 21 X 2 PKG, 0.8 MM PITCH, STAMPED LID	DOCUMENT NO: 98ASA00330D	REV: E
	STANDARD: NON–JEDEC	
	SOT1643–1	07 JAN 2016

Figure 101. 21 x 21 mm Lidded Package Top, Bottom, and Side Views (Sheet 2 of 2)

6.2.5 21 x 21 mm, 0.8 mm Pitch Ball Map

Table 98 shows the FCPBGA 21 x 21 mm, 0.8 mm pitch ball map.

Table 98. 21 x 21 mm, 0.8 mm Pitch Ball Map

	A	B	C	D	E	F	G
1		PCIE_RXM	GND	CSI_D1M	CSI_D2M	CSI_D3P	DSI_D0P
2	PCIE_REXT	PCIE_RXP	JTAG_TRSTB	CSI_D1P	CSI_D2P	CSI_D3M	DSI_D0M
3	PCIE_TXM	PCIE_TXP	JTAG_TMS	GND	CSI_D0P	CSI_CLK0P	GND
4	GND	GND	GND	CSI_REXT	CSI_D0M	CSI_CLK0M	DSI_REXT
5	FA_ANA	VDD_FA	CLK2_N	CLK2_P	GND	GND	JTAG_TDI
6	USB_OTG_DP	USB_OTG_DN	GND	GND	GND	GND	JTAG_TDO
7	XTALI	XTALO	CLK1_N	CLK1_P	GND	GND	PCIE_VPH
8	GND	USB_OTG_CHD_B	GPANAIO	GND	NVCC_PLL_OUT	GND	PCIE_VPTX
9	MLB_SN	MLB_SP	RTC_XTALO	RTC_XTALI	USB_OTG_VBUS	VDDUSB_CAP	VDD_SNVS_CAP
10	MLB_DP	MLB_DN	GND	USB_H1_VBUS	USB_H1_DP	USB_H1_DN	GND
11	MLB_CN	MLB_CP	POR_B	PMIC_ON_REQ	TAMPER	PMIC_STBY_REQ	VDD_SNVS_IN
12	SATA_TXP	SATA_TXM	BOOT_MODE0	ONOFF	TEST_MODE	BOOT_MODE1	SATA_VPH
13	GND	SD3_CMD	SD3_DAT5	SD3_DAT4	SD3_DAT6	SD3_DAT7	SATA_VP
14	SATA_RXM	SATA_RXP	SATA_REXT	SD3_CLK	SD3_DAT0	SD3_DAT1	NVCC_SD3
15	SD3_DAT2	SD3_DAT3	NANDF_CLE	SD3_RST	NANDF_WP_B	NANDF_CS0	NVCC_NANDF
16	NANDF_ALE	NANDF_RB0	NANDF_CS1	NANDF_CS3	SD4_CLK	NANDF_D2	NVCC_SD1
17	NANDF_CS2	SD4_CMD	NANDF_D1	NANDF_D3	NANDF_D6	SD4_DAT2	NVCC_SD2
18	NANDF_D0	NANDF_D5	NANDF_D7	SD4_DAT0	SD4_DAT4	SD1_DAT3	NVCC_RGMII
19	NANDF_D4	SD4_DAT1	SD4_DAT5	SD4_DAT7	SD1_DAT2	SD2_CMD	GND
20	SD4_DAT3	SD4_DAT6	SD1_DAT1	SD1_CLK	SD2_DAT1	RGMII_TD1	EIM_D20
21	SD1_DAT0	SD1_CMD	SD2_CLK	RGMII_TXC	RGMII_TD2	EIM_D17	EIM_D19
22	SD2_DAT0	SD2_DAT3	RGMII_TD0	RGMII_RX_CTL	EIM_EB2	EIM_D24	EIM_D25
23	SD2_DAT2	RGMII_RD1	RGMII_TX_CTL	RGMII_RD3	EIM_D22	EIM_EB3	EIM_D28
24	RGMII_TD3	RGMII_RD2	RGMII_RD0	EIM_D18	EIM_D26	EIM_A22	EIM_A17
25	GND	RGMII_RXC	EIM_D16	EIM_D23	EIM_D27	EIM_A24	EIM_A19