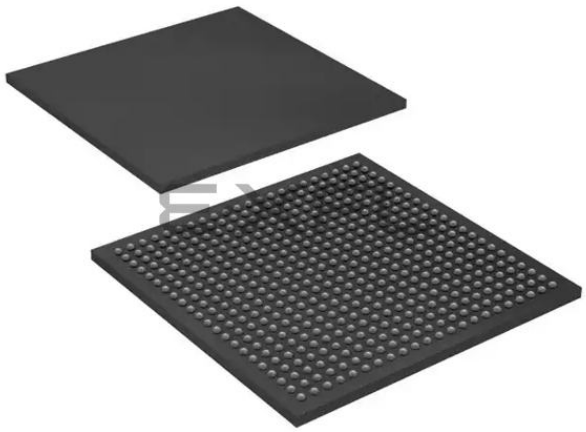

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Understanding Embedded - FPGAs (Field Programmable Gate Array)

Embedded - FPGAs, or Field Programmable Gate Arrays, are advanced integrated circuits that offer unparalleled flexibility and performance for digital systems. Unlike traditional fixed-function logic devices, FPGAs can be programmed and reprogrammed to execute a wide array of logical operations, enabling customized functionality tailored to specific applications. This reprogrammability allows developers to iterate designs quickly and implement complex functions without the need for custom hardware.

Applications of Embedded - FPGAs

The versatility of Embedded - FPGAs makes them indispensable in numerous fields. In telecommunications.

Details	
Product Status	Active
Number of LABs/CLBs	2076
Number of Logic Elements/Cells	33216
Total RAM Bits	483840
Number of I/O	322
Number of Gates	-
Voltage - Supply	1.15V ~ 1.25V
Mounting Type	Surface Mount
Operating Temperature	0°C ~ 85°C (TJ)
Package / Case	484-BGA
Supplier Device Package	484-FBGA (23x23)
Purchase URL	https://www.e-xfl.com/product-detail/intel/ep2c35f484c6

Section I. Cyclone II Device Family Data Sheet

This section provides information for board layout designers to successfully layout their boards for Cyclone® II devices. It contains the required PCB layout guidelines, device pin tables, and package specifications.

This section includes the following chapters:

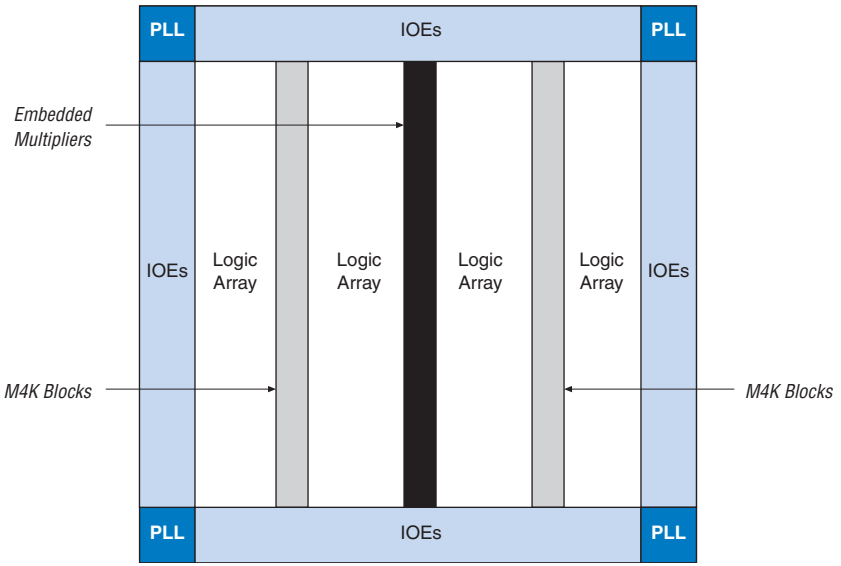
Revision History

Refer to each chapter for its own specific revision history. For information on when each chapter was updated, refer to the Chapter Revision Dates section, which appears in the complete handbook.

phase-align double data rate (DDR) signals) provide interface support for external memory devices such as DDR, DDR2, and single data rate (SDR) SDRAM, and QDRII SRAM devices at up to 167 MHz.

Figure 2–1 shows a diagram of the Cyclone II EP2C20 device.

Figure 2–1. Cyclone II EP2C20 Device Block Diagram



The number of M4K memory blocks, embedded multiplier blocks, PLLs, rows, and columns vary per device.

Logic Elements

The smallest unit of logic in the Cyclone II architecture, the LE, is compact and provides advanced features with efficient logic utilization. Each LE features:

- A four-input look-up table (LUT), which is a function generator that can implement any function of four variables
- A programmable register
- A carry chain connection
- A register chain connection
- The ability to drive all types of interconnects: local, row, column, register chain, and direct link interconnects
- Support for register packing
- Support for register feedback

Embedded multipliers can operate at up to 250 MHz (for the fastest speed grade) for 18 × 18 and 9 × 9 multiplications when using both input and output registers.

Each Cyclone II device has one to three columns of embedded multipliers that efficiently implement multiplication functions. An embedded multiplier spans the height of one LAB row. Table 2–10 shows the number of embedded multipliers in each Cyclone II device and the multipliers that can be implemented.

Table 2–10. Number of Embedded Multipliers in Cyclone II Devices *Note (1)*

Device	Embedded Multiplier Columns	Embedded Multipliers	9 × 9 Multipliers	18 × 18 Multipliers
EP2C5	1	13	26	13
EP2C8	1	18	36	18
EP2C15	1	26	52	26
EP2C20	1	26	52	26
EP2C35	1	35	70	35
EP2C50	2	86	172	86
EP2C70	3	150	300	150

Note to Table 2–10:

(1) Each device has either the number of 9 × 9-, or 18 × 18-bit multipliers shown. The total number of multipliers for each device is not the sum of all the multipliers.

The embedded multiplier consists of the following elements:

- Multiplier block
- Input and output registers
- Input and output interfaces

Figure 2–18 shows the multiplier block architecture.

There is only one signa and one signb signal for each dedicated multiplier. Therefore, all of the data A inputs feeding the same dedicated multiplier must have the same sign representation. Similarly, all of the data B inputs feeding the same dedicated multiplier must have the same sign representation. The signa and signb signals can be changed dynamically to modify the sign representation of the input operands at run time. The multiplier offers full precision regardless of the sign representation and can be registered using dedicated registers located at the input register stage.

Multiplier Modes

Table 2–12 summarizes the different modes that the embedded multipliers can operate in.

Table 2–12. Embedded Multiplier Modes

Multiplier Mode	Description
18-bit Multiplier	An embedded multiplier can be configured to support a single 18 × 18 multiplier for operand widths up to 18 bits. All 18-bit multiplier inputs and results can be registered independently. The multiplier operands can accept signed integers, unsigned integers, or a combination of both.
9-bit Multiplier	An embedded multiplier can be configured to support two 9 × 9 independent multipliers for operand widths up to 9-bits. Both 9-bit multiplier inputs and results can be registered independently. The multiplier operands can accept signed integers, unsigned integers or a combination of both. There is only one signa signal to control the sign representation of both data A inputs and one signb signal to control the sign representation of both data B inputs of the 9-bit multipliers within the same dedicated multiplier.

Table 2–18. Cyclone II Device LVDS Channels (Part 2 of 2)		
Device	**Pin Count**	**Number of LVDS Channels** *(1)*
EP2C70	672	160 (168)
	896	257 (265)

Note to Table 2–18:
(1) The first number represents the number of bidirectional I/O pins which can be used as inputs or outputs. The number in parenthesis includes dedicated clock input pin pairs which can only be used as inputs.

You can use I/O pins and internal logic to implement a high-speed I/O receiver and transmitter in Cyclone II devices. Cyclone II devices do not contain dedicated serialization or deserialization circuitry. Therefore, shift registers, internal PLLs, and IOEs are used to perform serial-to-parallel conversions on incoming data and parallel-to-serial conversion on outgoing data.

The maximum internal clock frequency for a receiver and for a transmitter is 402.5 MHz. The maximum input data rate of 805 Mbps and the maximum output data rate of 640 Mbps is only achieved when DDIO registers are used. The LVDS standard does not require an input reference voltage, but it does require a 100-Ω termination resistor between the two signals at the input buffer. An external resistor network is required on the transmitter side.

For more information on Cyclone II differential I/O interfaces, see the *High-Speed Differential Interfaces in Cyclone II Devices* chapter in Volume 1 of the *Cyclone II Device Handbook*.

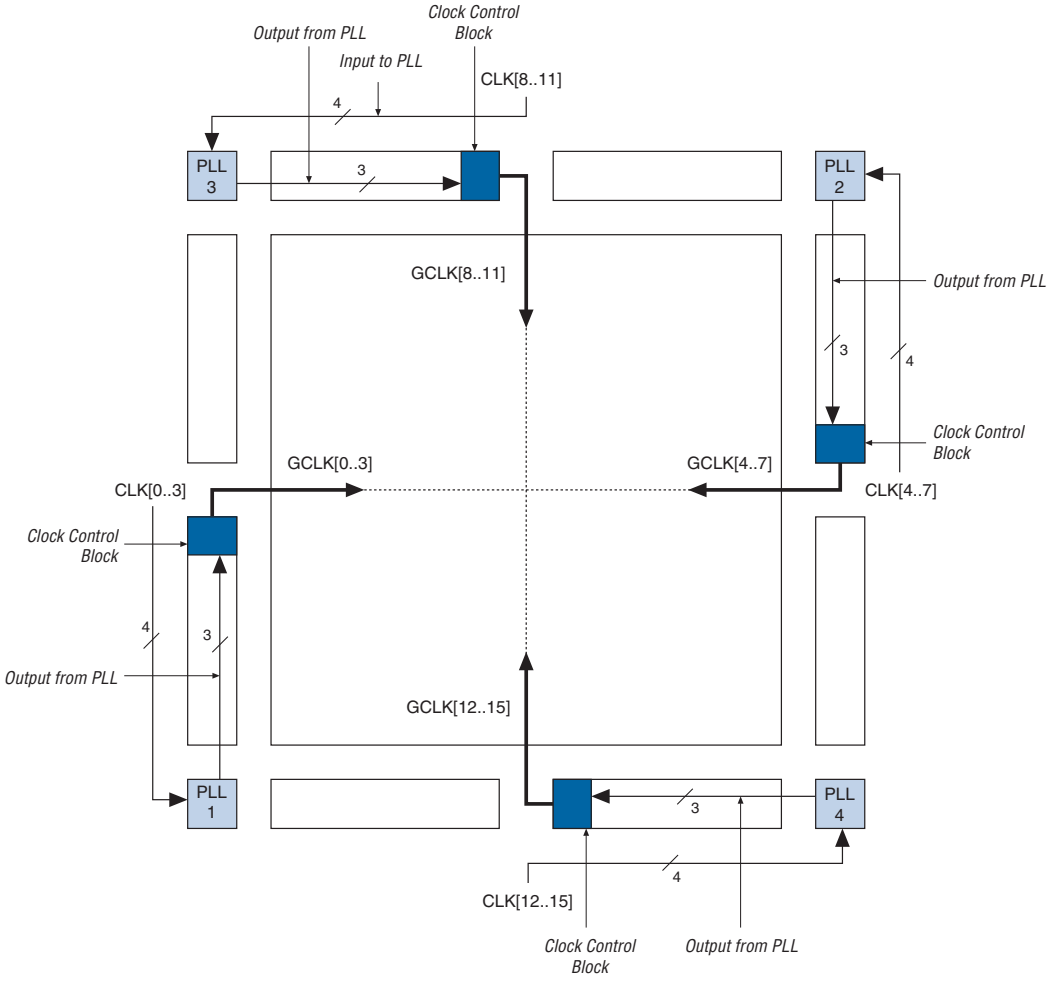
Series On-Chip Termination

On-chip termination helps to prevent reflections and maintain signal integrity. This also minimizes the need for external resistors in high pin count ball grid array (BGA) packages. Cyclone II devices provide I/O driver on-chip impedance matching and on-chip series termination for single-ended outputs and bidirectional pins.

Table 5–42. Cyclone II I/O Output Delay for Column Pins (Part 2 of 6)

I/O Standard	Drive Strength	Parameter	Fast Corner		–6 Speed Grade	–7 Speed Grade (2)	–7 Speed Grade (3)	–8 Speed Grade	Unit
			Industrial/ Automotive	Commer -cial					
2.5V	4 mA	t_{OP}	1208	1267	2478	2614	2743	2750	ps
		t_{DIP}	1340	1406	2648	2808	2969	2969	ps
	8 mA	t_{OP}	1190	1248	2307	2434	2554	2561	ps
		t_{DIP}	1322	1387	2477	2628	2780	2780	ps
	12 mA	t_{OP}	1154	1210	2192	2314	2430	2437	ps
		t_{DIP}	1286	1349	2362	2508	2656	2656	ps
	16 mA (1)	t_{OP}	1140	1195	2152	2263	2375	2382	ps
		t_{DIP}	1272	1334	2322	2457	2601	2601	ps
1.8V	2 mA	t_{OP}	1682	1765	3988	4279	4563	4570	ps
		t_{DIP}	1814	1904	4158	4473	4789	4789	ps
	4 mA	t_{OP}	1567	1644	3301	3538	3768	3775	ps
		t_{DIP}	1699	1783	3471	3732	3994	3994	ps
	6 mA	t_{OP}	1475	1547	2993	3195	3391	3398	ps
		t_{DIP}	1607	1686	3163	3389	3617	3617	ps
	8 mA	t_{OP}	1451	1522	2882	3074	3259	3266	ps
		t_{DIP}	1583	1661	3052	3268	3485	3485	ps
	10 mA	t_{OP}	1438	1508	2853	3041	3223	3230	ps
		t_{DIP}	1570	1647	3023	3235	3449	3449	ps
	12 mA (1)	t_{OP}	1438	1508	2853	3041	3223	3230	ps
		t_{DIP}	1570	1647	3023	3235	3449	3449	ps
1.5V	2 mA	t_{OP}	2083	2186	4477	4870	5256	5263	ps
		t_{DIP}	2215	2325	4647	5064	5482	5482	ps
	4 mA	t_{OP}	1793	1881	3649	3965	4274	4281	ps
		t_{DIP}	1925	2020	3819	4159	4500	4500	ps
	6 mA	t_{OP}	1770	1857	3527	3823	4112	4119	ps
		t_{DIP}	1902	1996	3697	4017	4338	4338	ps
	8 mA (1)	t_{OP}	1703	1787	3537	3827	4111	4118	ps
		t_{DIP}	1835	1926	3707	4021	4337	4337	ps

Figure 7–12. Cyclone II Clock Control Blocks Placement



The inputs to the four clock control blocks on each side are chosen from among the following clock sources:

- Four clock input pins
- Three PLL counter outputs
- Two DPCLK pins and two CDPCLK pins from both the left and right sides and four DPCLK pins and two CDPCLK pins from both the top and bottom
- Four signals from internal logic

QDRII SRAM devices use the following clock signals:

- Input clocks K and K#
- Optional output clocks C and C#
- Echo clocks CQ and CQn

Clocks C#, K#, and CQn are logical complements of clocks C, K, and CQ, respectively. Clocks C, C#, K, and K# are inputs to the QDRII SRAM, and clocks CQ and CQn are outputs from the QDRII SRAM. Cyclone II devices use single-clock mode for QDRII SRAM interfacing. The K and K# clocks are used for both read and write operations, and the C and C# clocks are unused.

You can generate C, C#, K, and K# clocks using any of the I/O registers via the DDR registers. Due to strict skew requirements between K and K# signals, use adjacent pins to generate the clock pair. Surround the pair with buffer pins tied to V_{CC} and pins tied to ground for better noise immunity from other signals.

In Cyclone II devices, another DQS pin implements the CQn pin in the QDRII SRAM memory interface. These pins are denoted by DQS/CQ# in the pin table. Connect CQ and CQn pins to the Cyclone II DQS/CQ and DQS/CQ# pins of the same DQ groups, respectively. You must configure the DQS/CQ and DQS/CQ# as bidirectional pins. However, because CQ and CQn pins are output-only pins from the memory device, the Cyclone II device's QDRII SRAM memory interface requires that you ground the DQS/CQ and DQS/CQ# output enable. To capture data presented by the memory device, connect the shifted CQ signal to register C_I and input register A_I. Connect the shifted CQn to input register B_I. Figure 9–4 shows the CQ and CQn connections for a QDRII SRAM read.

Document Revision History

Table 9–4 shows the revision history for this document.

Table 9–4. Document Revision History

Date & Document Version	Changes Made	Summary of Changes
February 2007 v3.1	• Added document revision history. • Added handpara note in "Data & Data Strobe Pins" section. • Updated "DDR Output Registers" section.	• Elaboration of DDR2 and QDRII interfaces supported by I/O bank included.
November 2005, v2.1	• Introduction • Updated Table 9–2. • Updated Figure 9–7.	
July 2005, v2.0	Updated Table 9–2.	
November 2004, v1.1	• Moved the "External Memory Interface Standards" section to follow the "Introduction" section. • Updated the "Data & Data Strobe Pins" section. • Updated Figures 9–11, 9–12, 9–15, 9–16, and 9–17.	
June 2004, v1.0	Added document to the Cyclone II Device Handbook.	

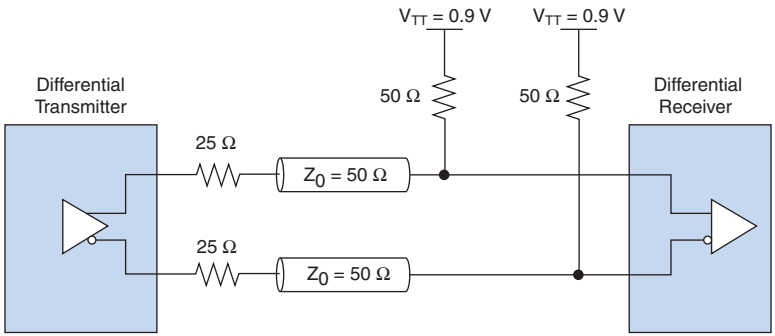
Pseudo-Differential SSTL-18 Class I and Differential SSTL-18 Class II

The 1.8-V differential SSTL-18 standard is formulated under JEDEC Standard, JESD8-15: Stub Series Terminated Logic for 1.8V (SSTL-18).

The differential SSTL-18 I/O standard is a 1.8-V standard used for applications such as high-speed DDR2 SDRAM interfaces. This standard supports differential signals in systems using the SSTL-18 standard and supplements the SSTL-18 standard for differential clocks. Refer to Figures 10–9 and 10–10 for details on differential SSTL-18 termination.

Cyclone II devices do not support true differential SSTL-18 standards. Cyclone II devices support pseudo-differential SSTL-18 outputs for `PLL_OUT` pins and pseudo-differential SSTL-18 inputs for clock pins. Pseudo-differential inputs require an input reference voltage as opposed to the true differential inputs. Refer to Table 10–1 on page 10–2 for information about pseudo-differential SSTL.

Figure 10–9. Differential SSTL-18 Class I Termination



I/O Driver Impedance Matching (R_S) and Series Termination (R_S)

Cyclone II devices support driver impedance matching to the impedance of the transmission line, typically 25 or 50 Ω When used with the output drivers, on-chip termination (OCT) sets the output driver impedance to 25 or 50 Ωby choosing the driver strength. Once matching impedance is selected, driver current can not be changed. Table 10–7 provides a list of output standards that support impedance matching. All I/O banks and I/O pins support impedance matching and series termination. Dedicated configuration pins and JTAG pins do not support impedance matching or series termination.

Table 10–7. Selectable I/O Drivers with Impedance Matching and Series Termination

I/O Standard	Target R_S (Ω)
3.3-V LVTTL/CMOS	25 *(1)*
2.5-V LVTTL/CMOS	50 *(1)*
1.8-V LVTTL/CMOS	50 *(1)*
SSTL-2 class I	50 *(1)*
SSTL-18 class I	50 *(1)*

Note to Table 10–7:

(1) These RS values are nominal values. Actual impedance varies across process, voltage, and temperature conditions. Tolerance is specified in the *DC Characteristics and Timing Specifications* chapter in volume 1 of the *Cyclone II Handbook*.

Pad Placement and DC Guidelines

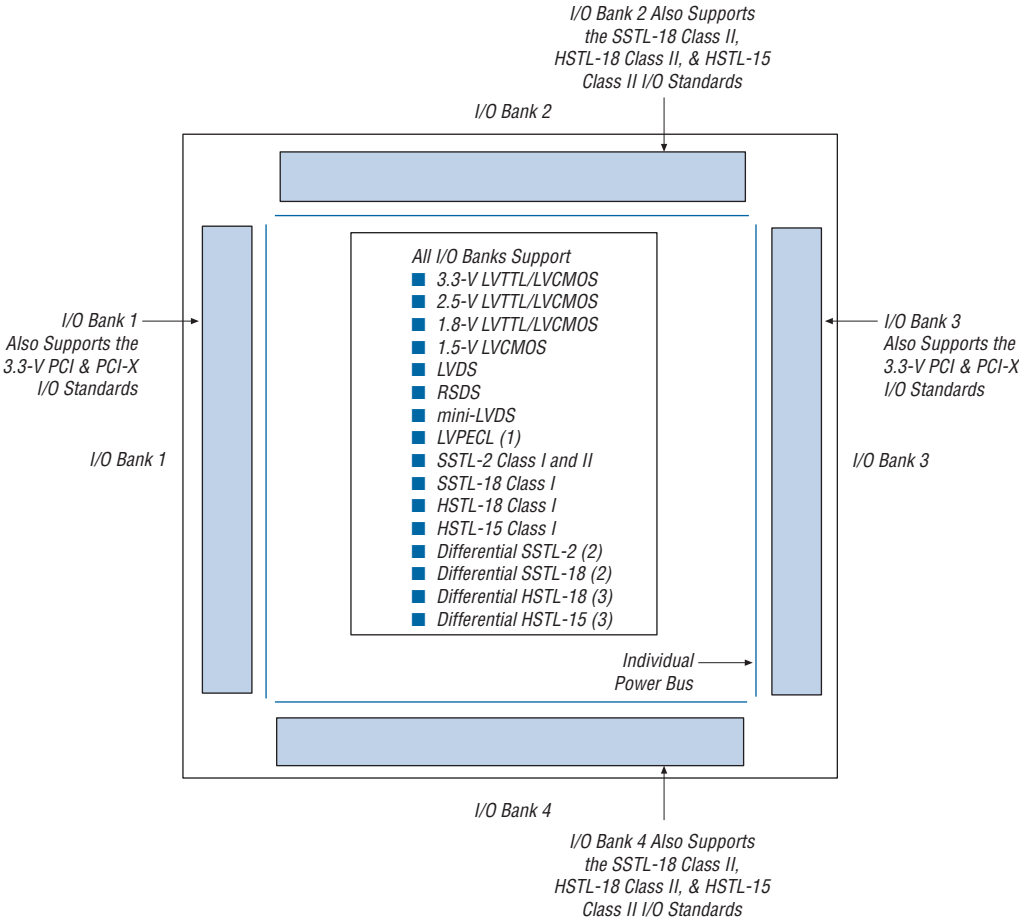
This section provides pad placement guidelines for the programmable I/O standards supported by Cyclone II devices and includes essential information for designing systems using the devices' selectable I/O capabilities. This section also discusses the DC limitations and guidelines.

Quartus II software provides user controlled restriction relaxation options for some placement constraints. When a default restriction is relaxed by a user, the Quartus II fitter generates warnings.

For more information about how Quartus II software checks I/O restrictions, refer to the *I/O Management* chapter in volume 2 of the *Quartus II Handbook*.

pins in each I/O bank (on both rows and columns) support the high-speed I/O interface. Cyclone II pin tables list the pins that support the high-speed I/O interface.

Figure 11–1. I/O Banks in EP2C5 & EP2C8 Devices



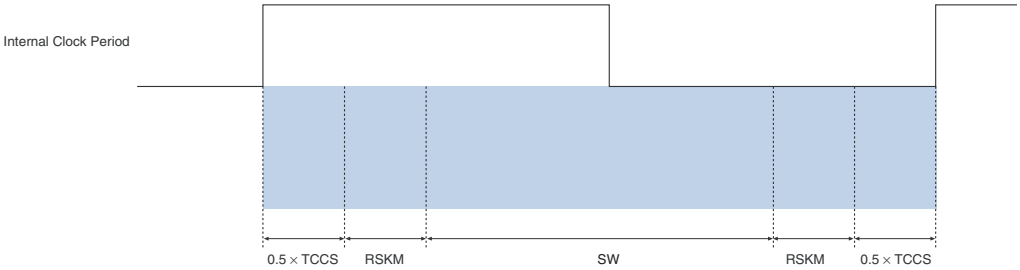
Notes to Figure 11–1:

(1) The LVPECL I/O standard is only supported on clock input pins. This I/O standard is not supported on output pins.

(2) The differential SSTL-18 and SSTL-2 I/O standards are only supported on clock input pins and PLL output clock pins.

(3) The differential 1.8-V and 1.5-V HSTL I/O standards are only supported on clock input pins and PLL output clock pins.

Figure 11–17. Cyclone II High-Speed I/O Timing Budget *Note (1)*



Note to Figure 11–17:
(1) The equation for the high-speed I/O timing budget is: Period = 0.5/TCCS + RSKM + SW + RSKM + 0.5/TCCS.

Design Guidelines

This section provides guidelines for designing with Cyclone II devices.

Differential Pad Placement Guidelines

To maintain an acceptable noise level on the V_{CCIO} supply, there are restrictions on placement of single-ended I/O pins in relation to differential pads.

See the guidelines in the *Selectable I/O Standards in Cyclone II Devices* chapter in Volume 1 of the *Cyclone II Device Handbook* for placing single-ended pads with respect to differential pads in Cyclone II devices.

Board Design Considerations

This section explains how to get the optimal performance from the Cyclone II I/O interface and ensure first-time success in implementing a functional design with optimal signal quality. The critical issues of controlled impedance of traces and connectors, differential routing, and termination techniques must be considered to get the best performance from the IC. The Cyclone II device generates signals that travel over the media at frequencies as high as 805 Mbps. Use the following general guidelines for improved signal quality:

- Base board designs on controlled differential impedance. Calculate and compare all parameters such as trace width, trace thickness, and the distance between two differential traces.

Document Revision History

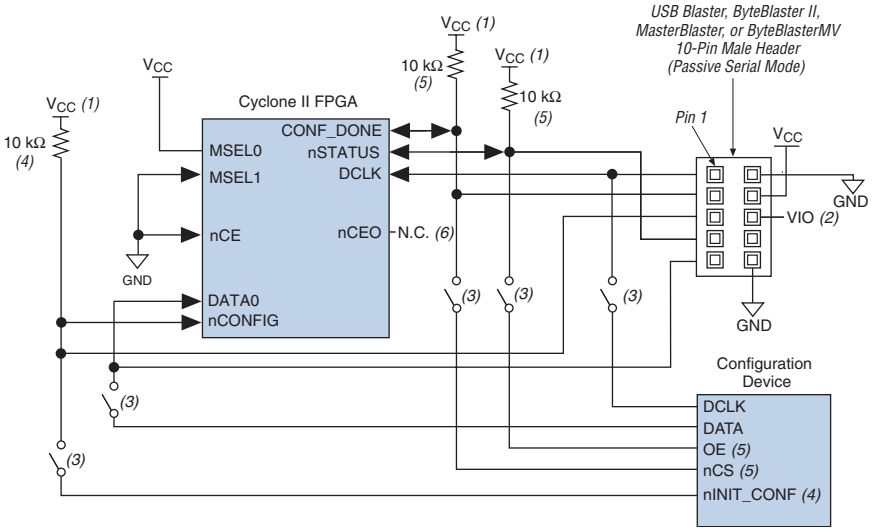
Table 11–6 shows the revision history for this document.

Table 11–6. Document Revision History

Date & Document Version	Changes Made	Summary of Changes
February 2007 v2.2	• Added document revision history. • Added *Note (1)* to Table 11–1. • Updated Figure 11–5 and added *Note (1)* • Added *Note (1)* to Table 11–2. • Updated Figure 11–6 and added *Note (1)* • Added *Note (1)* to Table 11–3. • Added *Note (1)* to Figure 11–9.	• Added information stating LVDS/RSDS/mini-LVDS I/O standards specifications apply at the external resistors network output.
November 2005 v2.1	• Updated Table 11–2. • Updated Figures 11–7 through 11–9. • Added Resistor Network Solution for RSDS. • Updated note for mini-LVDS Resistor Network table.	
July 2005 v2.0	• Updated "I/O Standards Support" section. • Updated Tables 11–1 through 11–3.	
November 2004 v1.1	• Updated Table 11–1. • Updated Figures 11–4, 11–5, 11–7, and 11–9.	
June 2004, v1.0	Added document to the Cyclone II Device Handbook.	

the five common signals (nCONFIG, nSTATUS, DCLK, DATA0, and CONF_DONE) between the cable and the configuration device. You can also remove the configuration device from the board when configuring the FPGA with the cable. Figure 13–21 shows a combination of a configuration device and a download cable to configure an FPGA.

Figure 13–21. PS Configuration with a Download Cable & Configuration Device Circuit



Notes to Figure 13–21:

(1) The pull-up resistor should be connected to the same supply voltage as the configuration device.

(2) Pin 6 of the header is a V_{IO} reference voltage for the MasterBlaster output driver. V_{IO} should match the device's V_{CCIO}. Refer to the *MasterBlaster Serial/USB Communications Cable Data Sheet* for this value. In the ByteBlasterMV, this pin is a no connect. In the USB-Blaster and ByteBlaster II, this pin is connected to nCE when it is used for AS programming, otherwise it is a no connect.

(3) You should not attempt configuration with a download cable while a configuration device is connected to a Cyclone II device. Instead, you should either remove the configuration device from its socket when using the download cable or place a switch on the five common signals between the download cable and the configuration device.

(4) The nINIT_CONF pin (available on enhanced configuration devices and EPC2 devices only) has an internal pull-up resistor that is always active. This means an external pull-up resistor should not be used on the nINIT_CONF to nCONFIG line. The nINIT_CONF pin does not need to be connected if its functionality is not used. If nINIT_CONF is not used or not available (e.g., on EPC1 devices), nCONFIG must be pulled to V_{CC} either directly or through a resistor (if reconfiguration is required, a resistor is necessary).

(5) The enhanced configuration devices' OE and nCS pins have internal programmable pull-up resistors. If internal pull-up resistors are used, external pull-up resistors should not be used on these pins. The internal pull-up resistors are used by default in the Quartus II software. To turn off the internal pull-up resistors, check the **Disable nCS and OE pull-ups on configuration device** option when generating programming files.

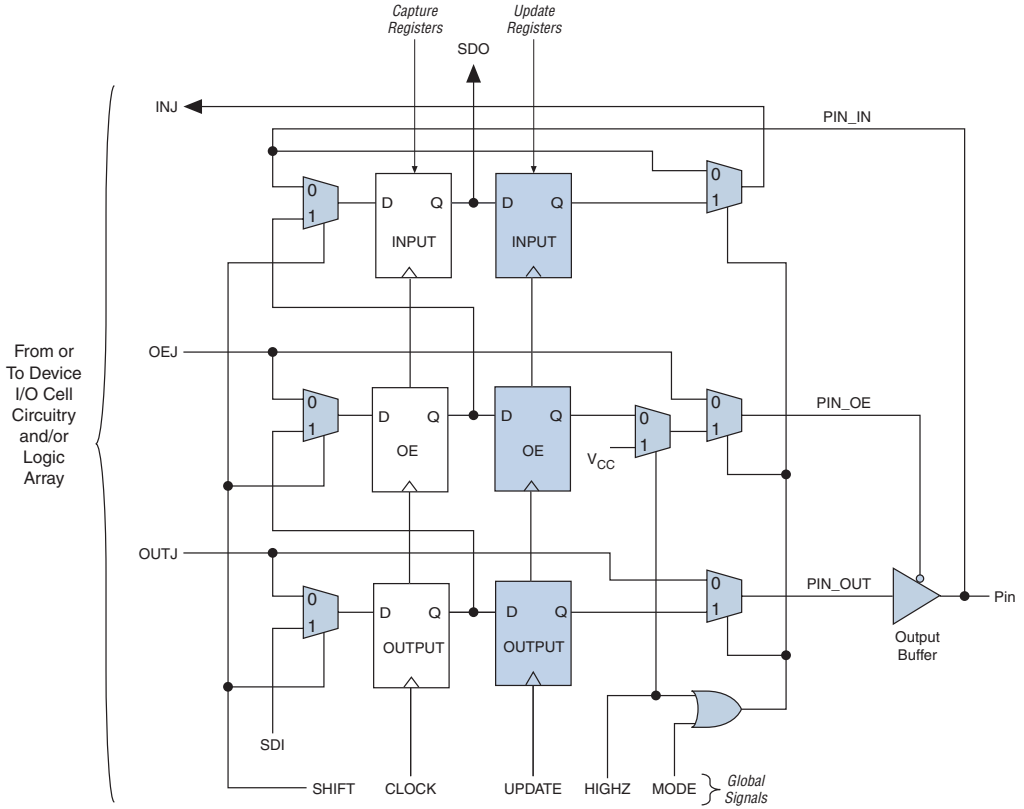
(6) The nCEO pin can be left unconnected or used as a user I/O pin when it does not feed other device's nCE pin.

frequency (up to 40 MHz), which reduces your configuration time. In addition, Cyclone II devices can receive a compressed configuration bitstream and decompress this data on-the-fly in the AS or PS configuration scheme, which further reduces storage requirements and configuration time.

to external device data via the PIN_IN signal, while the update registers connect to external data through the PIN_OUT and PIN_OE signals. The global control signals for the IEEE Std. 1149.1 BST registers (for example, shift, clock, and update) are generated internally by the TAP controller. The MODE signal is generated by a decode of the instruction register. The data signal path for the boundary-scan register runs from the serial data in (SDI) signal to the serial data out (SDO) signal. The scan register begins at the TDI pin and ends at the TDO pin of the device.

Figure 14–4 shows the Cyclone II device's user I/O boundary-scan cell.

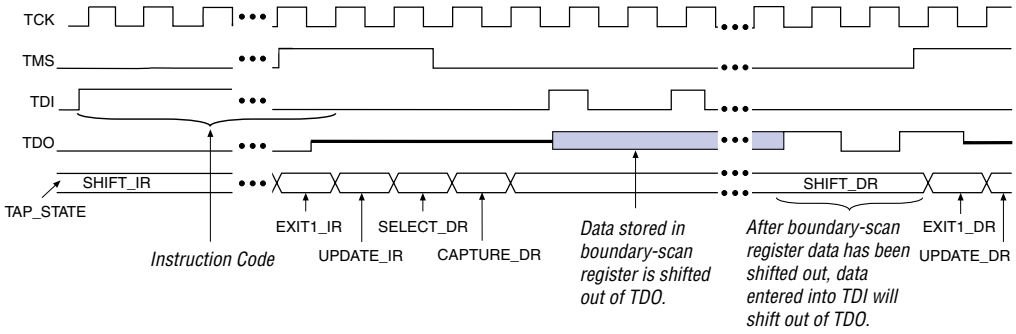
Figure 14–4. Cyclone II Device's User I/O BSC with IEEE Std. 1149.1 BST Circuitry



EXTEST selects data differently than SAMPLE/PRELOAD. EXTEST chooses data from the update registers as the source of the output and output enable signals. Once the EXTEST instruction code is entered, the multiplexers select the update register data. Thus, data stored in these registers from a previous EXTEST or SAMPLE/PRELOAD test cycle can be forced onto the pin signals. In the capture phase, the results of this test data are stored in the capture registers, then shifted out of TDO during the shift phase. New test data can then be stored in the update registers during the update phase.

The EXTEST waveform diagram in Figure 14–11 resembles the SAMPLE/PRELOAD waveform diagram, except for the instruction code. The data shifted out of TDO consists of the data that was present in the capture registers after the capture phase. New test data shifted into the TDI pin appears at the TDO pin after being clocked through the entire boundary-scan register.

Figure 14–11. EXTEST Shift Data Register Waveforms



BYPASS Instruction Mode

The BYPASS mode is activated when an instruction code of all 1's is loaded in the instruction register. The waveforms in Figure 14–12 show how scan data passes through a device once the TAP controller is in the SHIFT_DR state. In this state, data signals are clocked into the bypass register from TDI on the rising edge of TCK and out of TDO on the falling edge of the same clock pulse.

Tables 15–5 and 15–6 show the package information and package outline figure references, respectively, for the 144-pin TQFP package.

Table 15–5. 144-Pin TQFP Package Information

Description	Specification
Ordering code reference	T
Package acronym	TQFP
Lead frame material	Copper
Lead finish (plating)	Regular: 85Sn:15Pb (Typ.) Pb-free: Matte Sn
JEDEC Outline Reference	MS-026 Variation: BFB
Maximum lead coplanarity	0.003 inches (0.08mm)
Weight	1.3 g
Moisture sensitivity level	Printed on moisture barrier bag

Table 15–6. 144-Pin TQFP Package Outline Dimensions

Symbol	Millimeter		
	Min.	Nom.	Max.
A	–	–	1.60
A1	0.05	–	0.15
A2	1.35	1.40	1.45
D	22.00 BSC		
D1	20.00 BSC		
E	22.00 BSC		
E1	20.00 BSC		
L	0.45	0.60	0.75
L1	1.00 REF		
S	0.20	–	–
b	0.17	0.22	0.27
c	0.09	–	0.20
e	0.50 BSC		
θ	0°	3.5°	7°

Figure 15–6 shows a 484-pin Ultra FineLine BGA package outline.

Figure 15–6. 484-Pin Ultra FineLine BGA Package Outline